



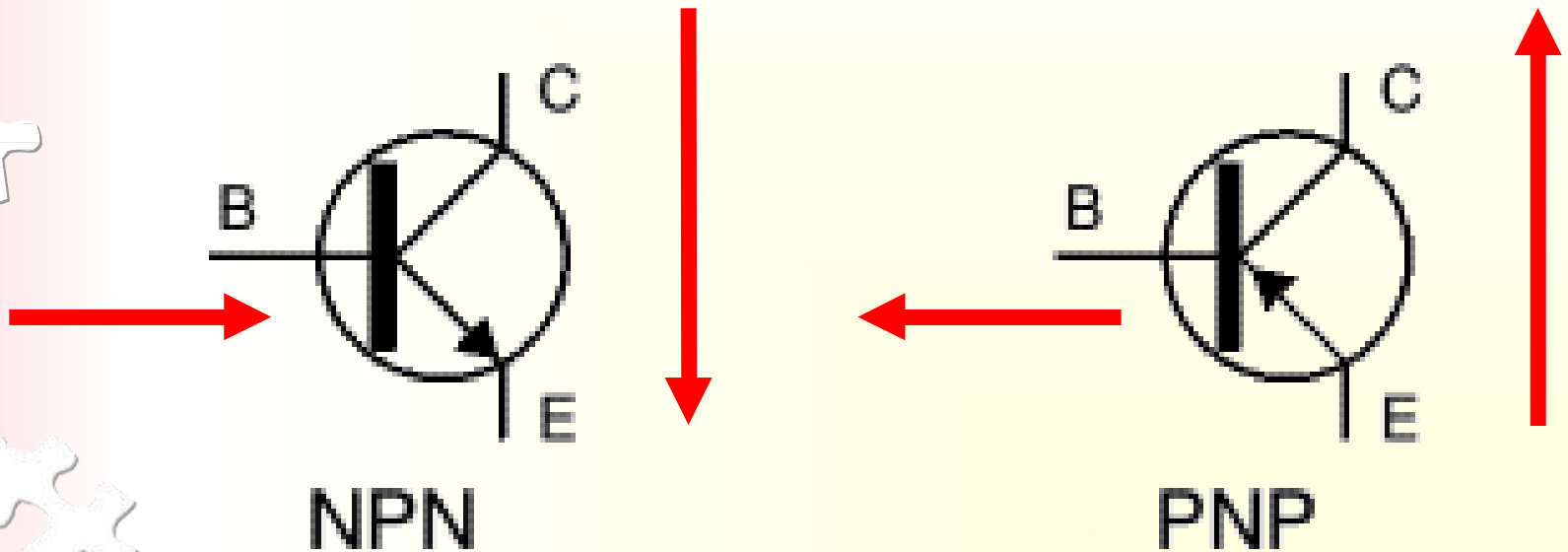
# *Bipolar Junction Transistor*

# Introduction

- The basic of electronic system nowadays is semiconductor device.
- The famous and commonly use of this device is BJTs (Bipolar Junction Transistors).
- It can be use as amplifier and logic switches.
- Transistors consists of three terminal:
  - ◇ collector : C
  - ◇ base : B
  - ◇ emitter : E
- Two types of Transistors : pnp and npn

# Types of Transistors

- Circuit Symbols

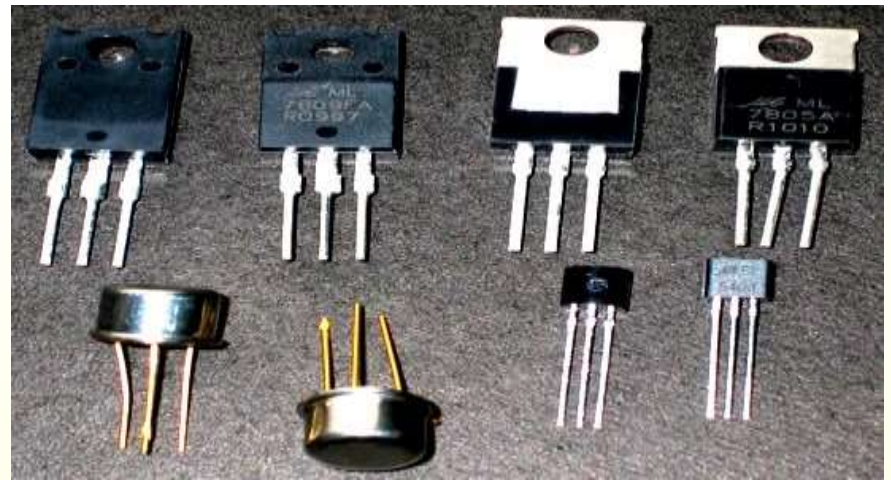


There are two types of standard transistors, **NPN** and **PNP**, with different circuit symbols. The letters refer to the layers of semiconductor material used to make the transistor. Most transistors used today are NPN because this is the easiest type to make from silicon. If you are new to electronics it is best to start by learning how to use NPN transistors. The leads are labeled **base** (B), **collector** (C) and **emitter** (E).



# Transistors

- A Transistor is a small electronic component that can be said to be a switch. Transistors are used in electronic regulation of energy and power.
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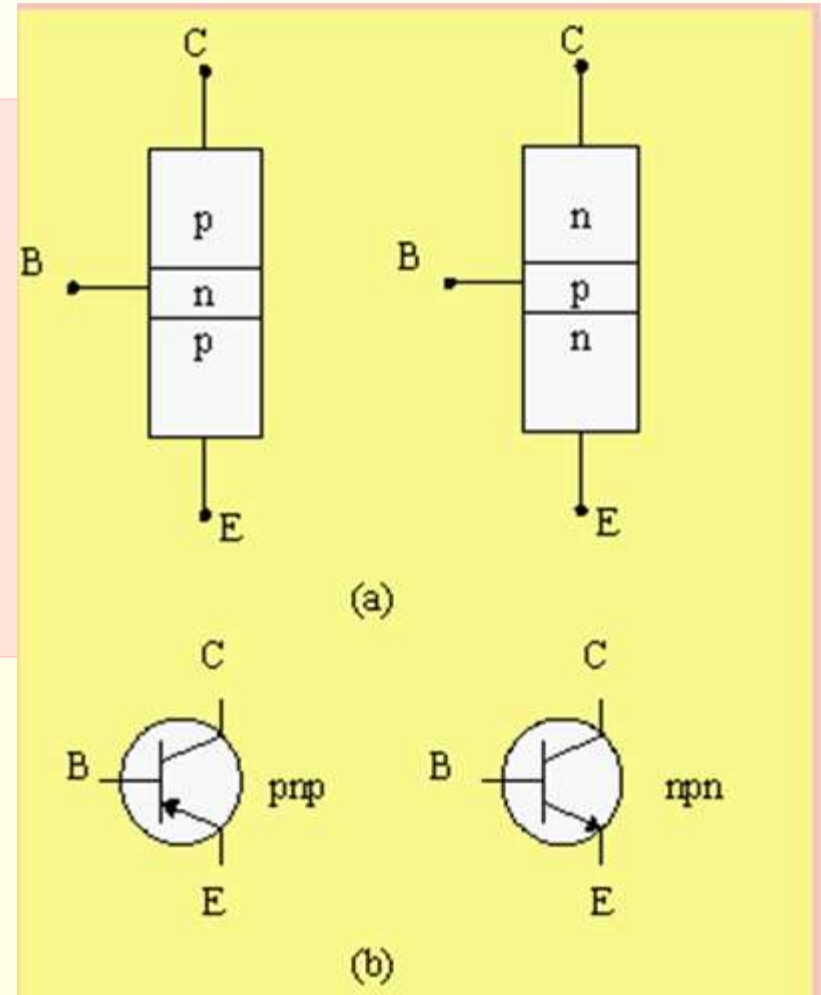
- A Transistor is an semiconductor which is a fundamental component in almost all electronic devices. Transistors are often said to be the most significant invention of the 20th Century. Transistors have many uses including switching, voltage/current regulation, and amplification - all of which are useful in renewable energy applications.
- The transistor is one of the most important component used in electronics, even integrated circuits have many thousands of them within it. Many electronic products and their circuits rely on sensors in order to function, most sensors provide only a very small output current or voltage, it is necessary to amplify this current or voltage

# Transistor Construction

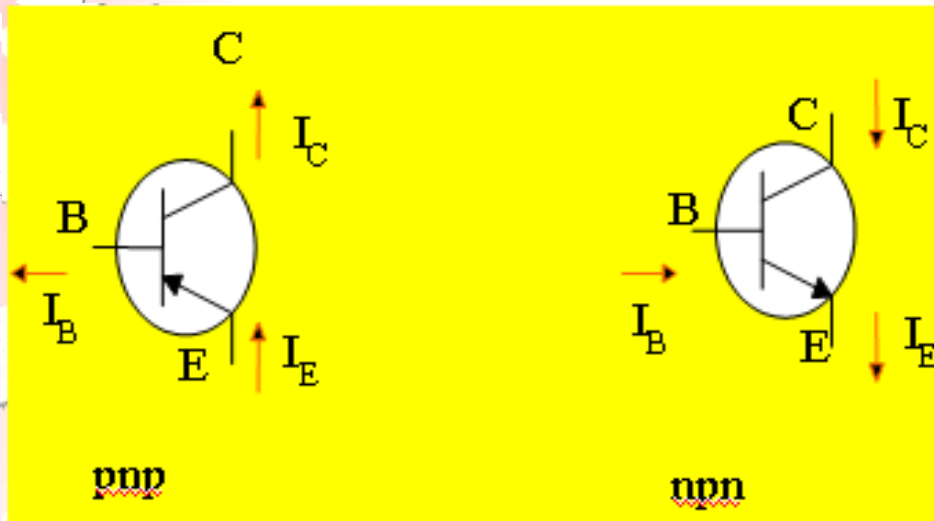
- 3 layer semiconductor device consisting:
  - 2 n- and 1 p-type layers of material  $\diamond$  npn transistor
  - 2 p- and 1 n-type layers of material  $\diamond$  pnp transistor
- The term bipolar reflects the fact that holes and electrons participate in the injection process into the oppositely polarized material
- A single pn junction has two different types of bias:
  - forward bias
  - reverse bias
- Thus, a two-pn-junction device has four types of bias.

# Position of the terminals and symbol of transistor.

- **Base** is located at the middle and more thin from the level of **collector** and **emitter**
- The emitter and collector terminals are made of the same type of semiconductor material, while the base of the other type of material



# Transistor currents



$I_C$  = the collector current  
 $I_B$  = the base current  
 $I_E$  = the emitter current

- The arrow is always drawn on the emitter

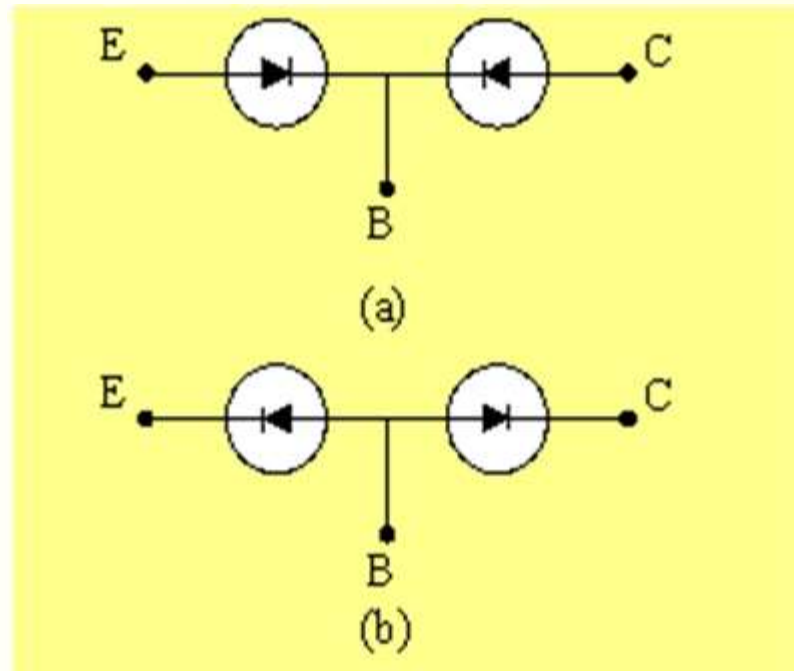
- The arrow always point toward the n-type

- The arrow indicates the direction of the emitter current:

pnp: E  $\diamond$  B

npn: B  $\diamond$  E

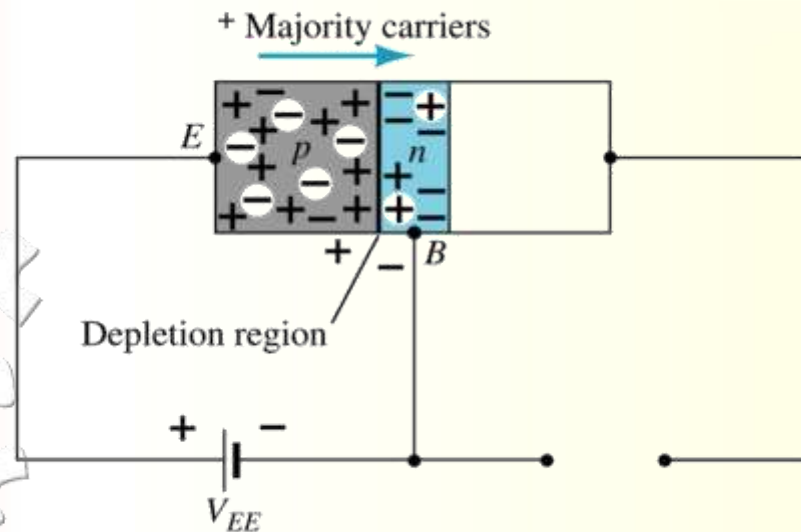
- By imaging the analogy of diode, transistor can be construct like two diodes that connetecd together.
- It can be conclude that the work of transistor is base on work of diode.



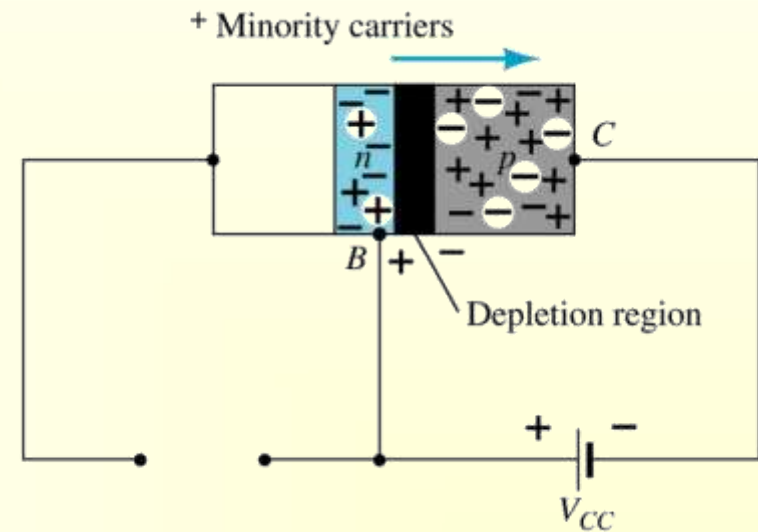


# Transistor Operation

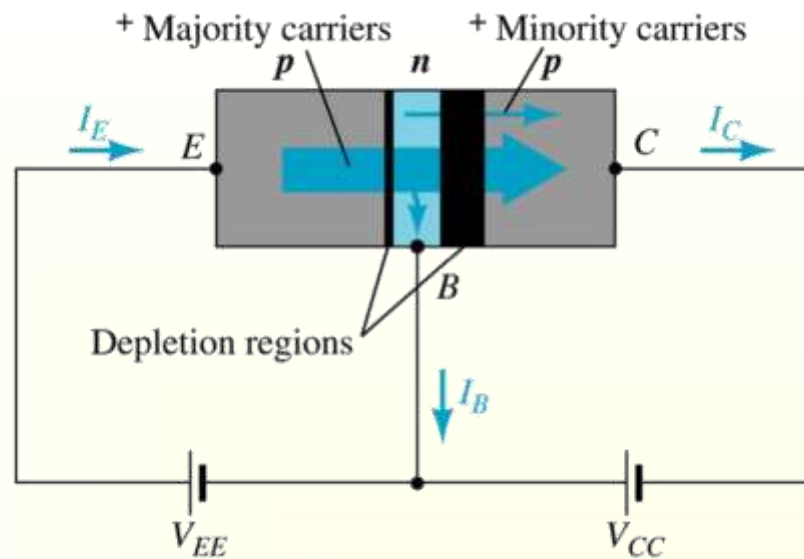
- The basic operation will be described using the pnp transistor. The operation of the pnp transistor is exactly the same if the roles played by the electron and hole are interchanged.
- One p-n junction of a transistor is reverse-biased, whereas the other is forward-biased.



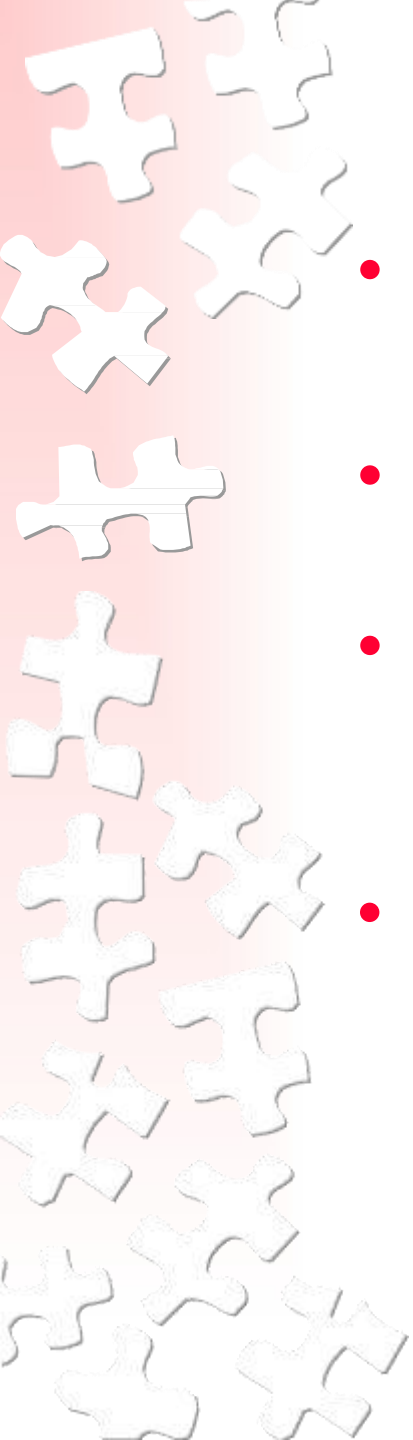
Forward-biased junction  
of a pnp transistor



Reverse-biased junction  
of a pnp transistor



- Both biasing potentials have been applied to a pnp transistor and resulting majority and minority carrier flows indicated.
- Majority carriers (+) will diffuse across the forward-biased p-n junction into the n-type material.
- A very small number of carriers (+) will through n-type material to the base terminal. Resulting  $I_B$  is typically in order of microamperes.
- The large number of majority carriers will diffuse across the reverse-biased junction into the p-type material connected to the collector terminal.

- 
- Majority carriers can cross the reverse-biased junction because the injected majority carriers will appear as minority carriers in the n-type material.
  - Applying KCL to the transistor :

$$I_E = I_C + I_B$$

- The  $I_C$  comprises of two components – the majority and minority carriers

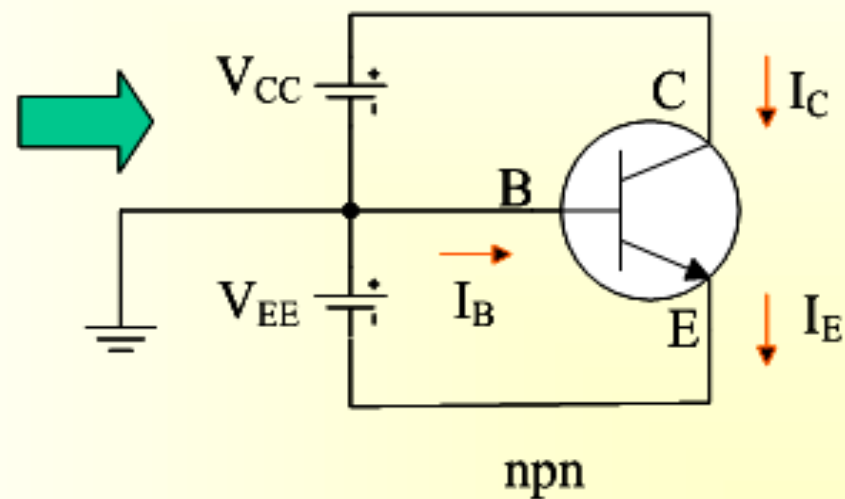
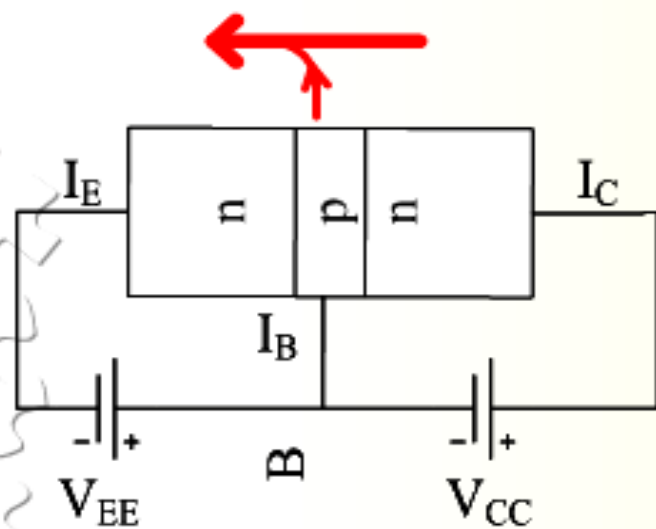
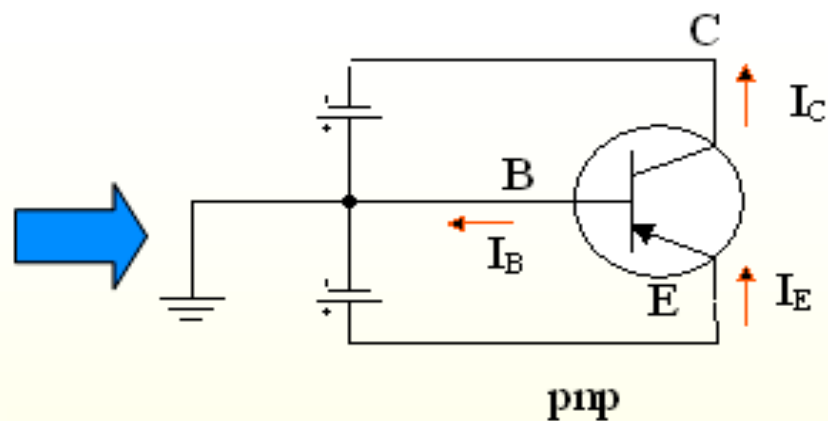
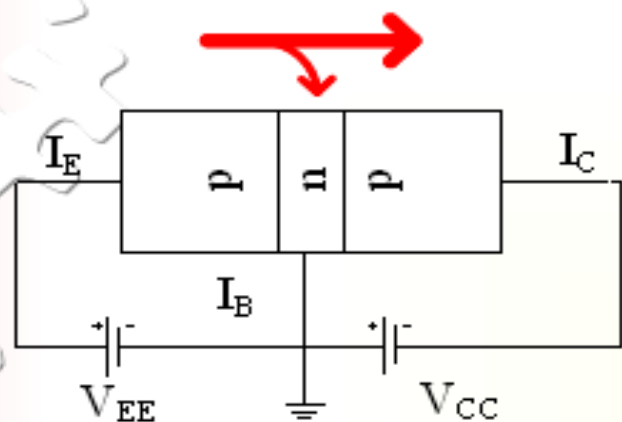
$$I_C = I_{Cmajority} + I_{Cminority}$$

- $I_{\infty}$  –  $I_C$  current with emitter terminal open and is called **leakage current**.

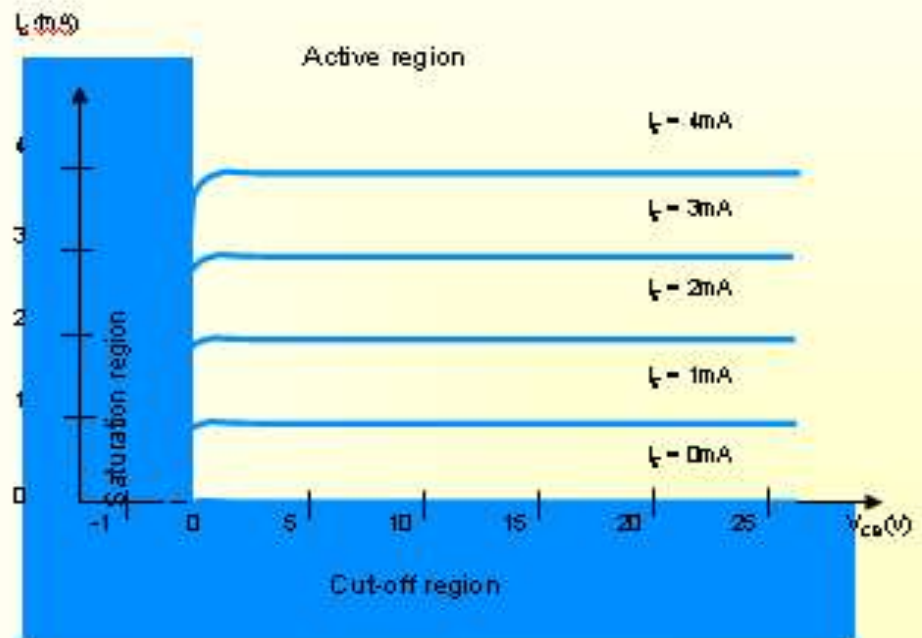
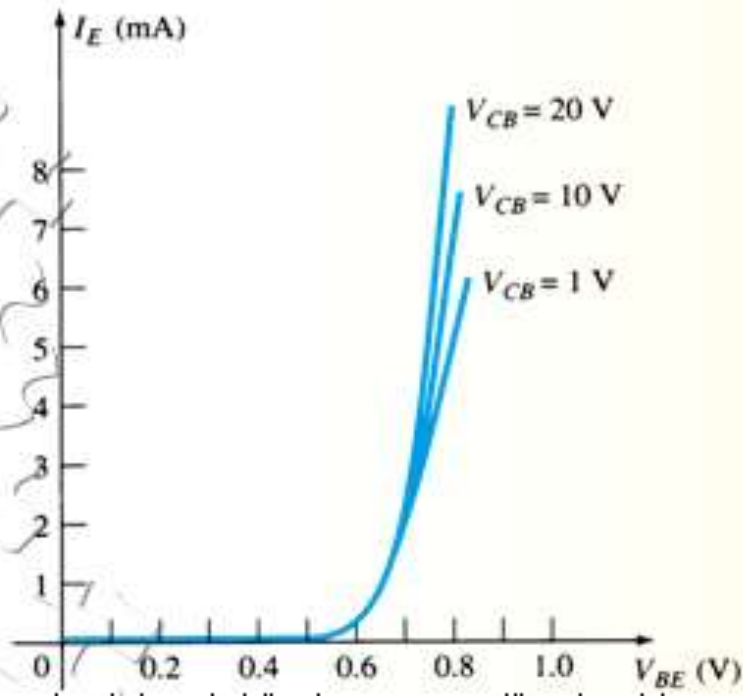


# Common-Base Configuration

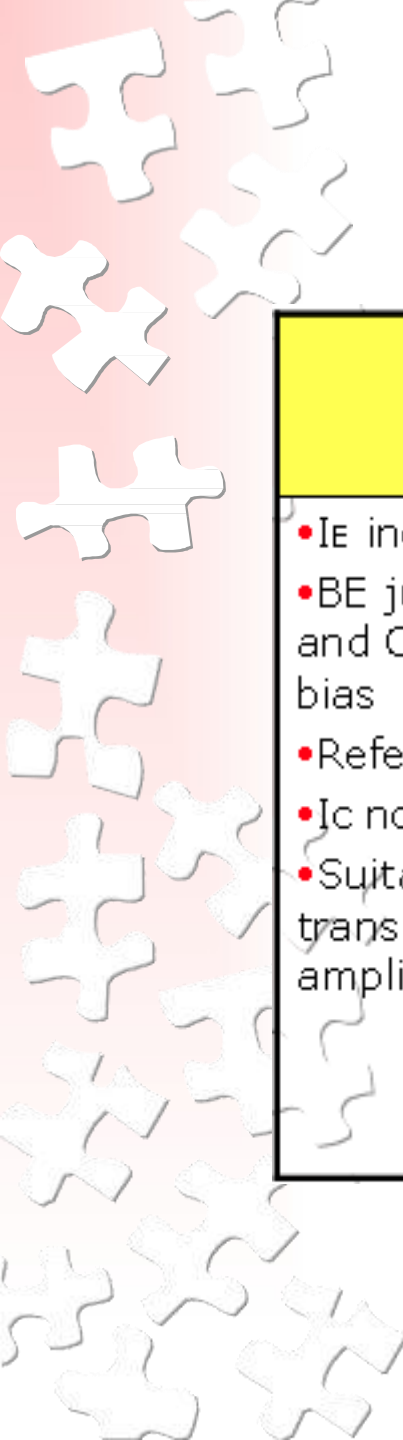
- Common-base terminology is derived from the fact that the :
  - base is common to both input and output of the configuration.
  - base is usually the terminal closest to or at ground potential.
- All current directions will refer to **conventional** (hole) flow and the arrows in all electronic symbols have a direction defined by this convention.
- Note that the applied **biasing** (voltage sources) are such as to establish current in the direction indicated for each branch.



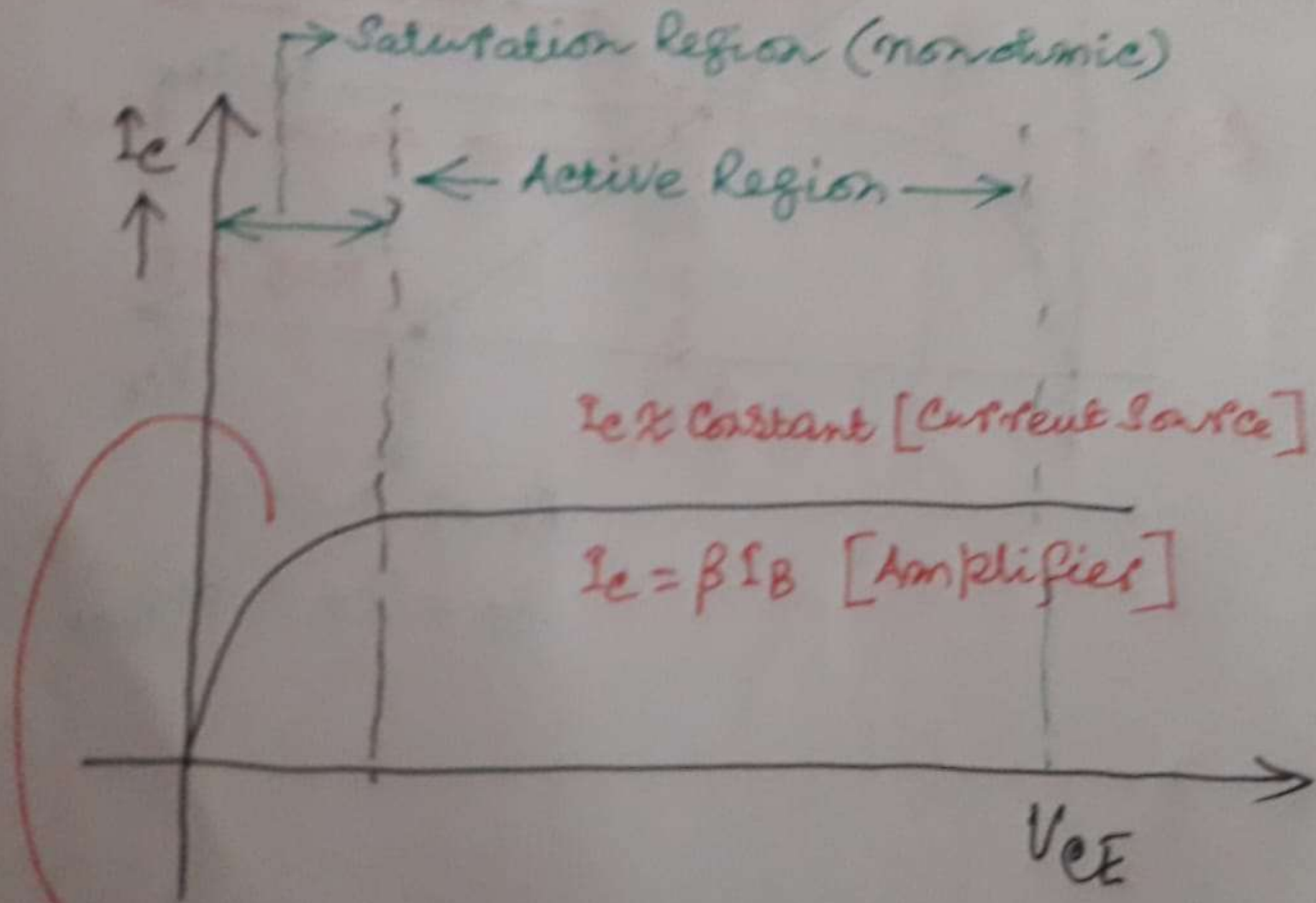
- To describe the behavior of common-base amplifiers requires two set of characteristics:
  - Input or driving point characteristics.
  - Output or collector characteristics
- The output characteristics has 3 basic regions:
  - Active region –defined by the biasing arrangements
  - Cutoff region – region where the collector current is 0A
  - Saturation region- region of the characteristics to the left of  $V_{CB} = 0V$



Output of collector characteristics for common-base transistor



<b>Active region</b>	<b>Saturation region</b>	<b>Cut-off region</b>
<ul style="list-style-type: none"><li>• <math>I_E</math> increased, <math>I_C</math> increased</li><li>• BE junction forward bias and CB junction reverse bias</li><li>• Refer to the <u>graf</u>, <math>I_C \approx I_E</math></li><li>• <math>I_C</math> not depends on <math>V_{CB}</math></li><li>• Suitable region for the transistor working as amplifier</li></ul>	<ul style="list-style-type: none"><li>• BE and CB junction is forward bias</li><li>• Small changes in <math>V_{CB}</math> will cause big different to <math>I_C</math></li><li>• The allocation for this region is to the left of <math>V_{CB} = 0\text{ V}</math>.</li></ul>	<ul style="list-style-type: none"><li>• Region below the line of <math>I_E = 0\text{ A}</math></li><li>• BE and CB is reverse bias</li><li>• no current flow at collector, only leakage current</li></ul>



→ By imposing cond<sup>n</sup> of Hard Saturation BJT may act as SWITCH.

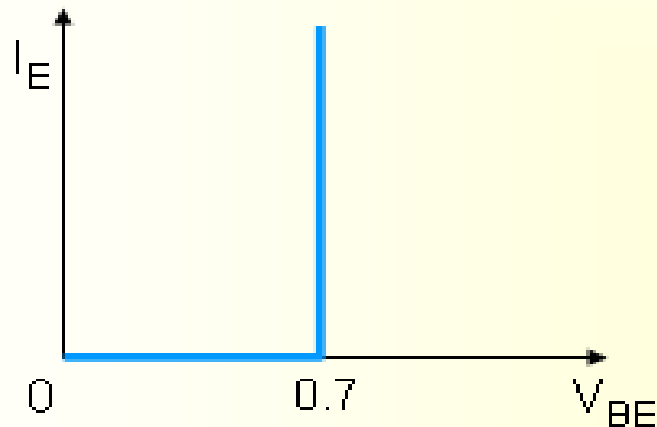


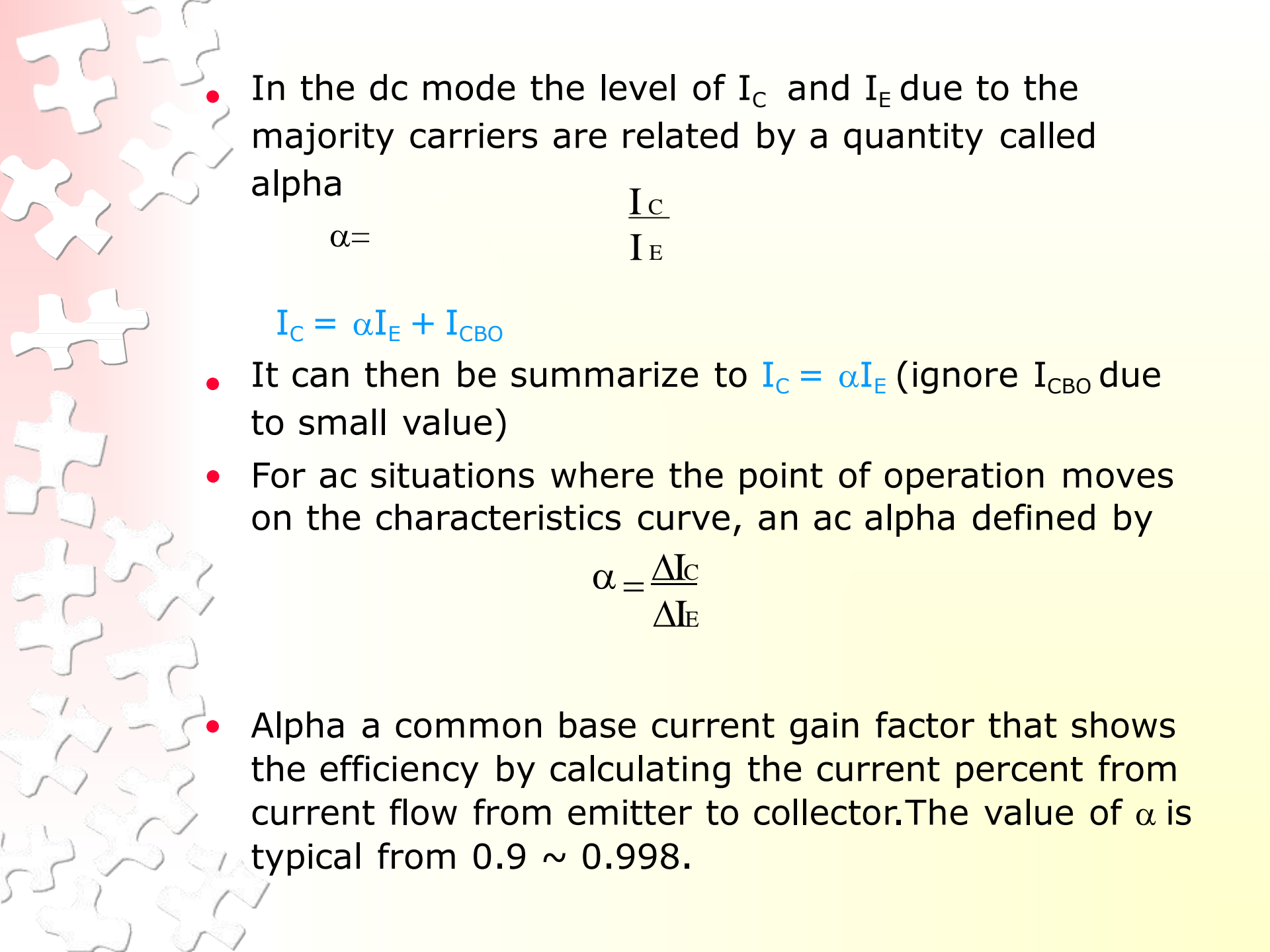
- The curves (output characteristics) clearly indicate that a first approximation to the relationship between  $I_E$  and  $I_C$  in the active region is given by

$$I_C \approx I_E$$

- Once a transistor is in the 'on' state, the base-emitter voltage will be assumed to be

$$V_{BE} = 0.7V$$



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- In the dc mode the level of  $I_C$  and  $I_E$  due to the majority carriers are related by a quantity called alpha

$$\alpha = \frac{I_C}{I_E}$$

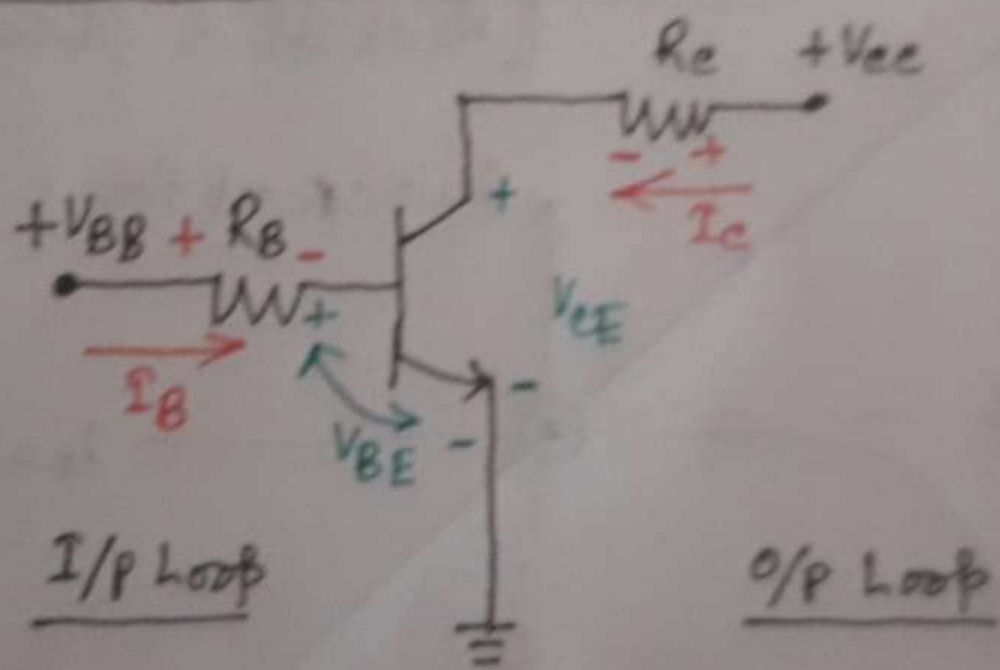
$$I_C = \alpha I_E + I_{CBO}$$

- It can then be summarize to  $I_C = \alpha I_E$  (ignore  $I_{CBO}$  due to small value)
- For ac situations where the point of operation moves on the characteristics curve, an ac alpha defined by

$$\alpha = \frac{\Delta I_C}{\Delta I_E}$$

- Alpha a common base current gain factor that shows the efficiency by calculating the current percent from current flow from emitter to collector. The value of  $\alpha$  is typical from 0.9 ~ 0.998.

# Q Point & Load line



$$V_{BB} - I_B R_B - V_{BE} = 0$$

$$\therefore I_B = \frac{V_{BB} - V_{BE}}{R_B}$$

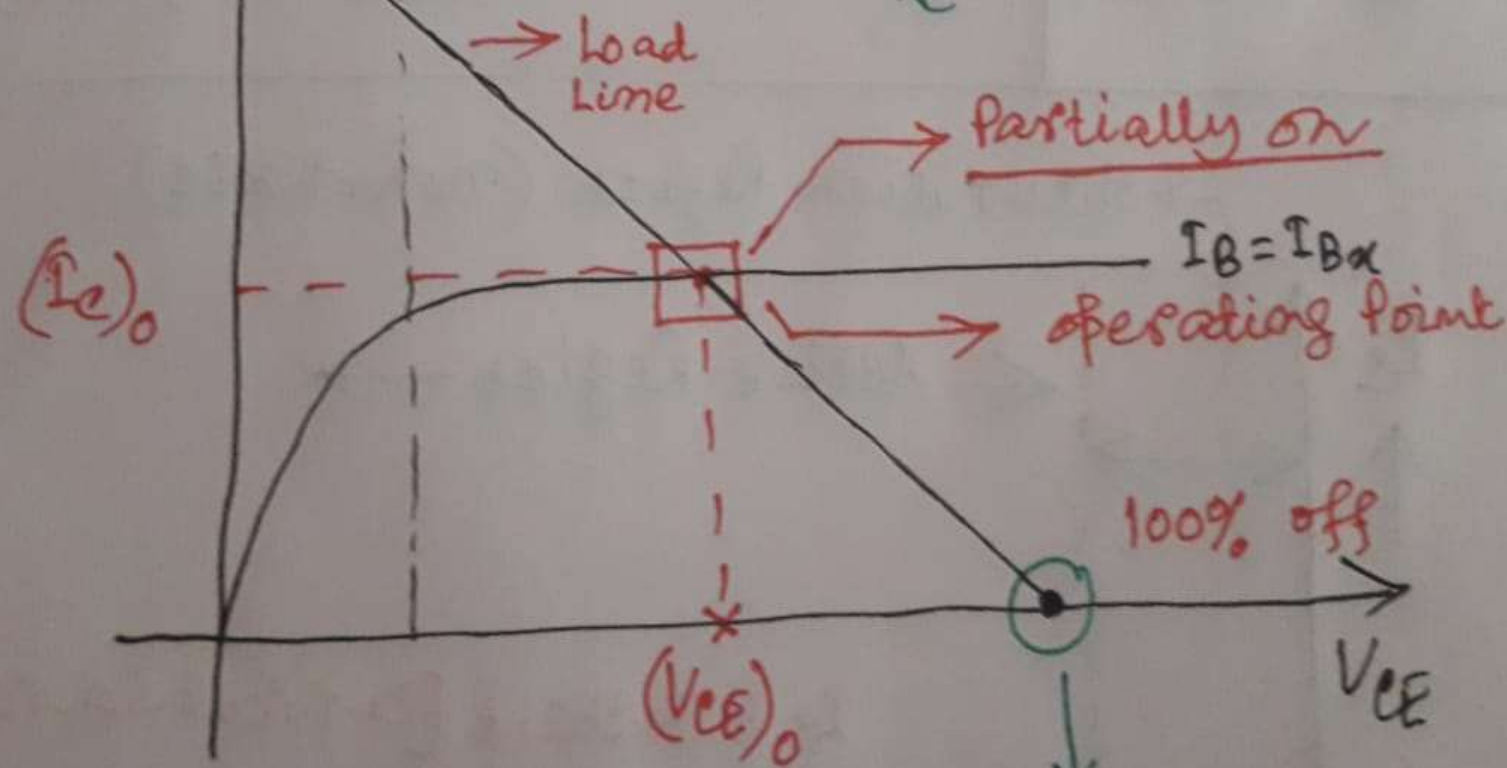
$$V_{CC} - I_C R_C - V_{CE} = 0$$

$$I_C = \frac{V_{CC} - V_{CE}}{R_C}$$

$I_c$  ↑

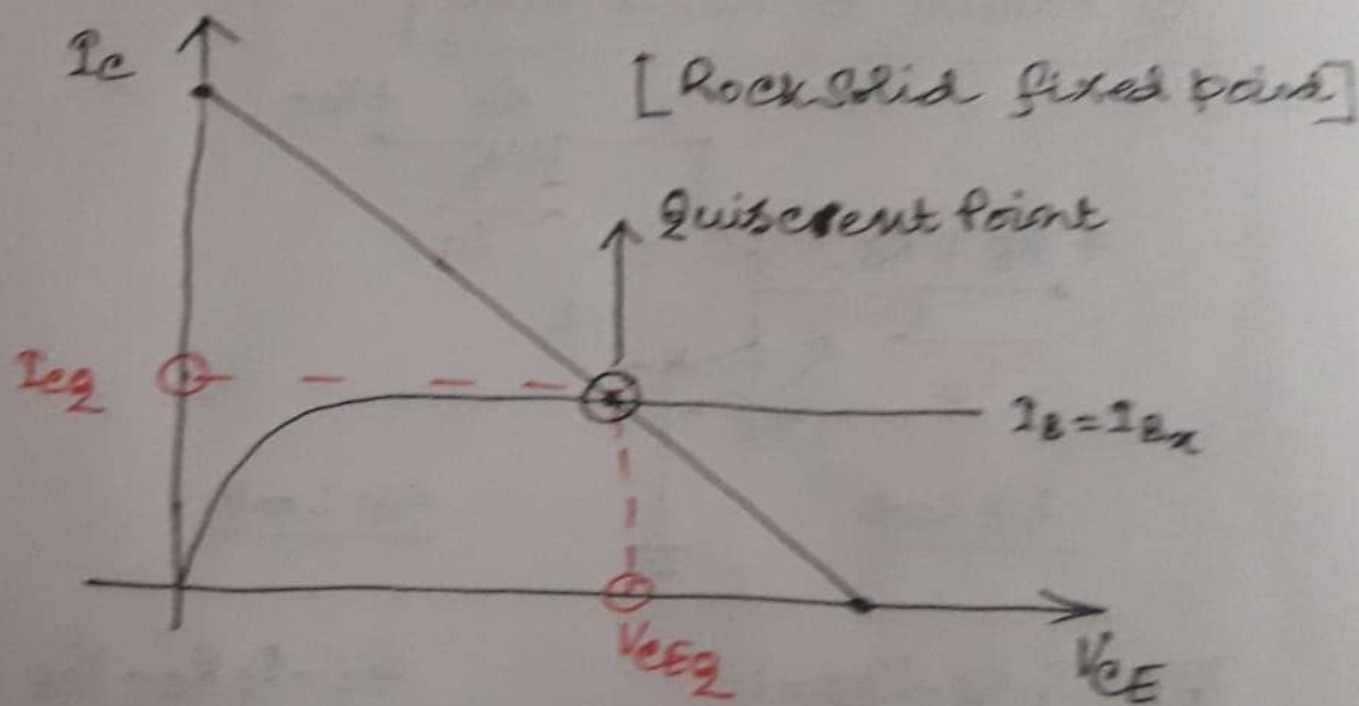
100% on

Pt. of Saturation

$$I_c = \frac{V_{cc}}{R_c} = I_{c,max} = (I_c)_{sat}$$


Pt of Cut off

$$V_{cc} = V_{ce} = (V_{ce})_{max} = (V_{ce})_{sat}$$



For current source or Amplifier

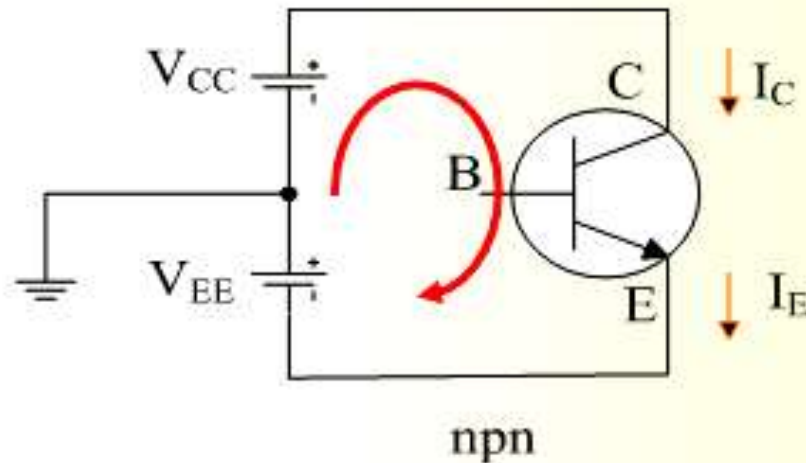
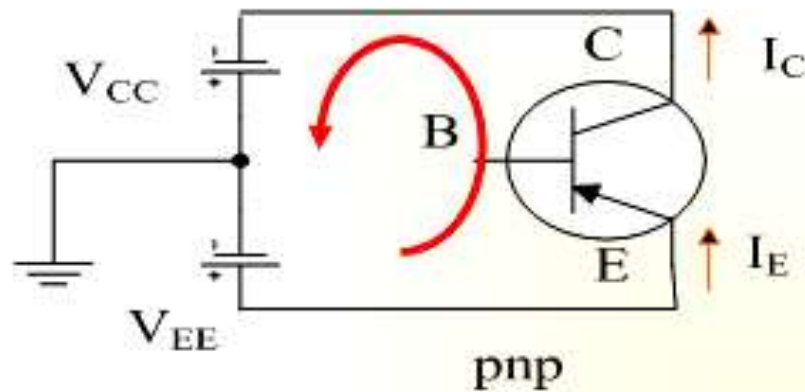
Operating point is the

$$I_{CQ} = f(\beta, V_{BE}, I_{EBO})$$

Q point [ $I_C$  &  $V_{CE}$  value must be fixed]

# Biasing

- Proper biasing CB configuration in active region by approximation  $I_C \approx I_E$  ( $I_B \approx 0 \text{ } \mu\text{A}$ )



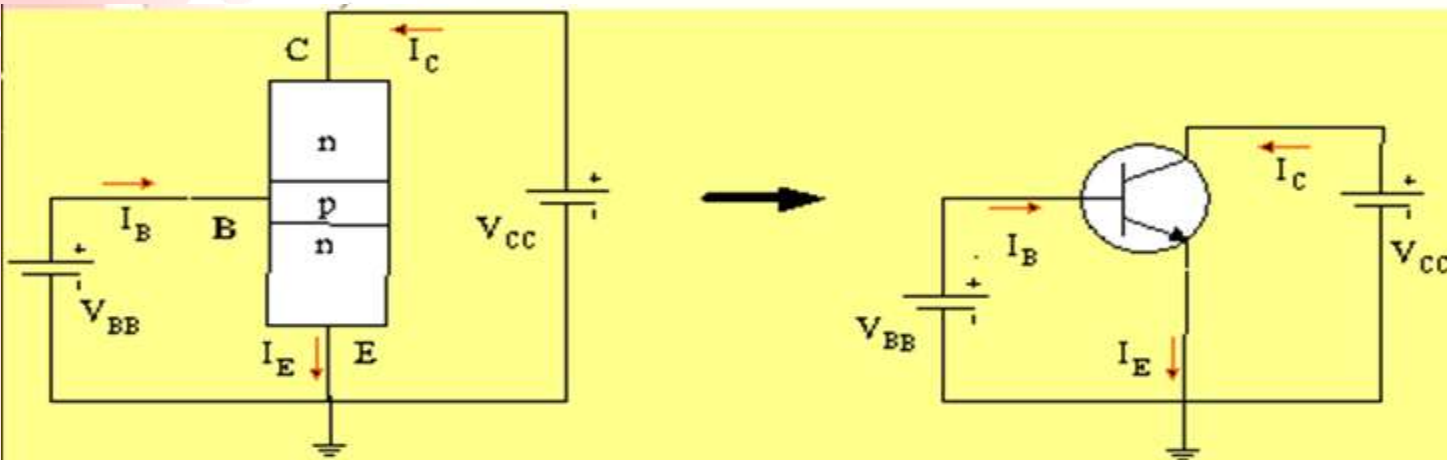


# Common-Emitter Configuration

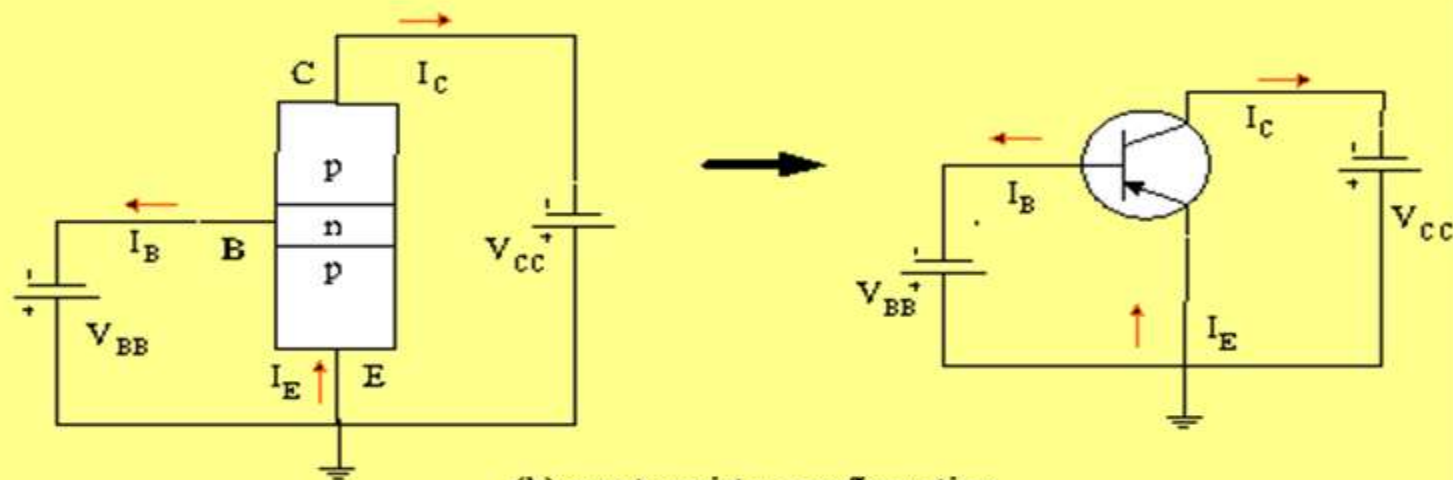
- It is called common-emitter configuration since :
  - emitter is common or reference to both input and output terminals.
  - emitter is usually the terminal closest to or at ground potential.
- Almost amplifier design is using connection of CE due to the high gain for current and voltage.
- Two set of characteristics are necessary to describe the behavior for CE ;input (base terminal) and output (collector terminal) parameters.



# Proper Biasing common-emitter configuration in active region



(a) npn transistor configuration

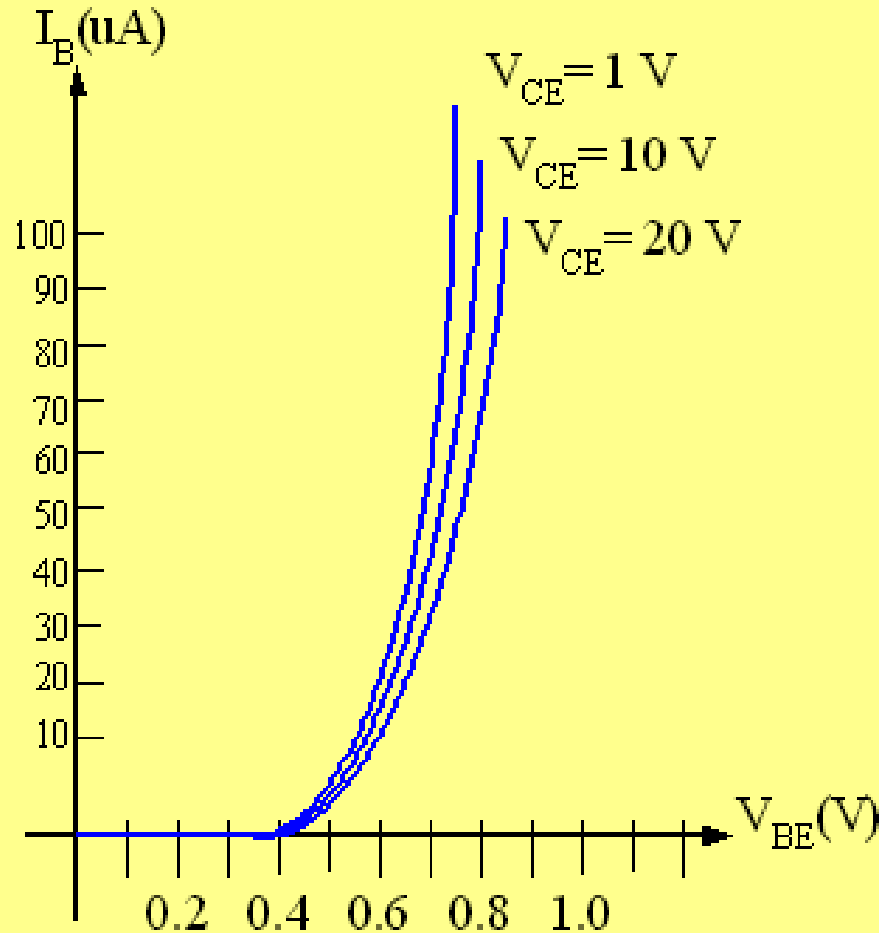


(b) pnp transistor configuration

Common-emitter configuration

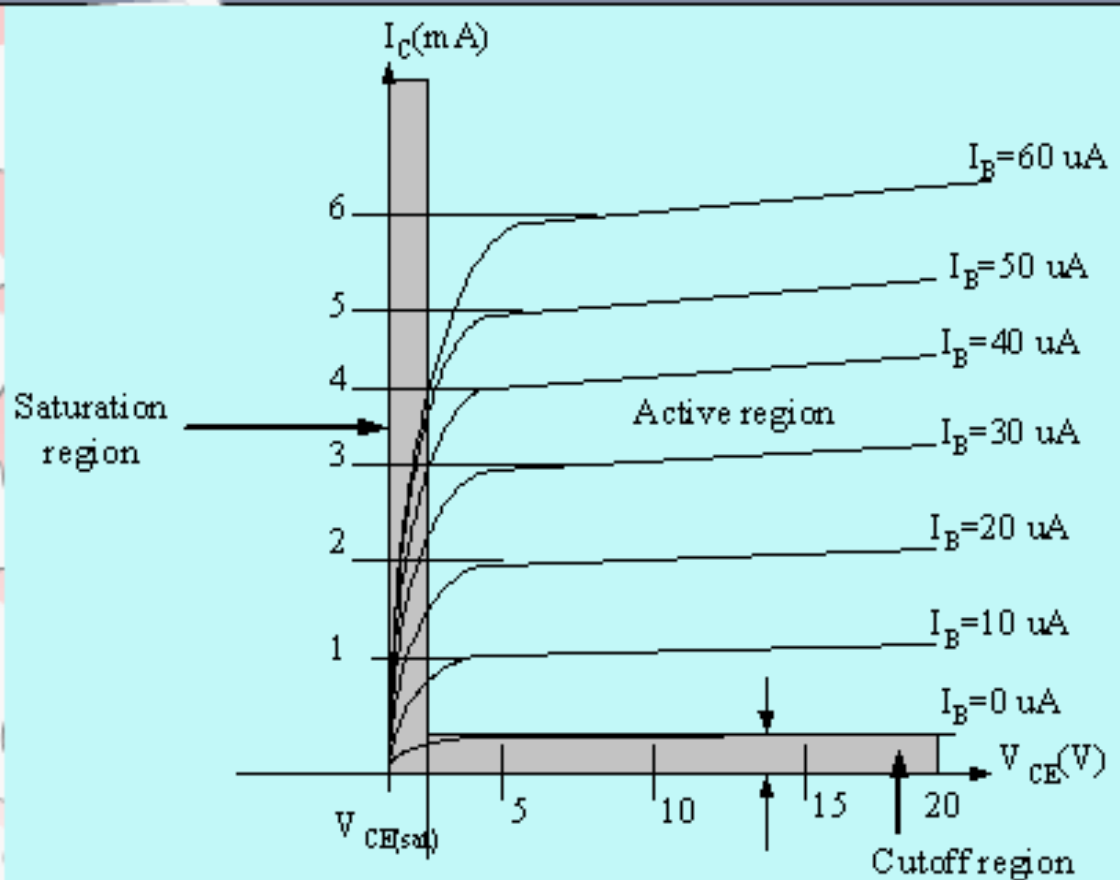
$$I_E = I_C + I_B$$





Input characteristics for a common-emitter NPN transistor

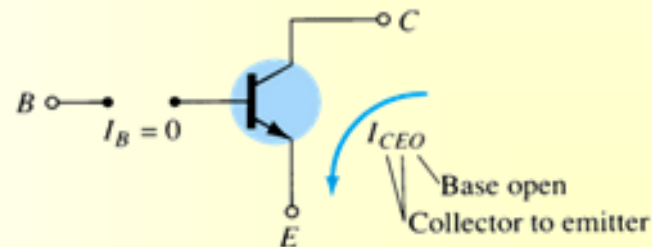
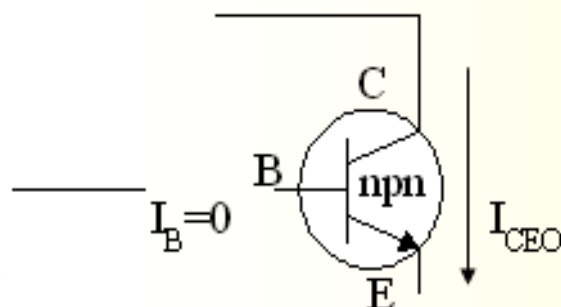
- $I_B$  is microamperes compared to milliamperes of  $I_C$ .
- $I_B$  will flow when  $V_{BE} > 0.7$  V for silicon and 0.3 V for germanium
- Before this value  $I_B$  is very small and no  $I_B$ .
- Base-emitter junction is forward bias
- Increasing  $V_{CE}$  will reduce  $I_B$  for different values.



Output characteristics for a common-emitter npn transistor

- For small  $V_{CE}$  ( $V_{CE} < V_{CE(sat)}$ ),  $I_C$  increase linearly with increasing of  $V_{CE}$
- $V_{CE} > V_{CE(sat)}$ ,  $I_C$  not totally depends on  $V_{CE}$   $\diamond$  constant  $I_C$
- $I_B$  ( $\mu A$ ) is very small compare to  $I_C$  (mA). Small increase in  $I_B$  cause big increase in  $I_C$
- $I_B = 0$  A  $\diamond$   $I_{CEO}$  occur.
- Noticing the value when  $I_C = 0$  A. There is still some value of current flows.

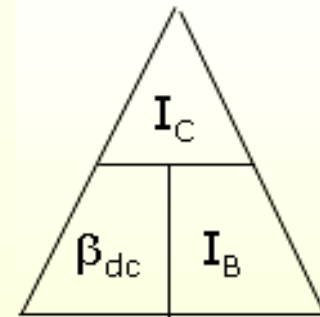
Active region	Saturation region	Cut-off region
<ul style="list-style-type: none"> <li>• B-E junction is forward bias</li> <li>• C-B junction is reverse bias</li> <li>• can be employed for voltage, current and power amplification</li> </ul>	<ul style="list-style-type: none"> <li>• B-E and C-B junction is forward bias, thus the values of <math>I_B</math> and <math>I_C</math> is too big.</li> <li>• The value of <math>V_{CE}</math> is so small.</li> <li>• Suitable region when the transistor as a logic switch.</li> <li>• NOT and avoid this region when the transistor as an amplifier.</li> </ul>	<ul style="list-style-type: none"> <li>• region below <math>I_B=0\mu A</math> is to be avoided if an undistorted o/p signal is required</li> <li>• B-E junction and C-B junction is reverse bias</li> <li>• <math>I_B=0</math>, <math>I_C</math> not zero, during this condition <math>I_C=I_{CEO}</math> where is this current flow when B-E is reverse bias.</li> </ul>



# Beta ( $\beta$ ) or amplification factor

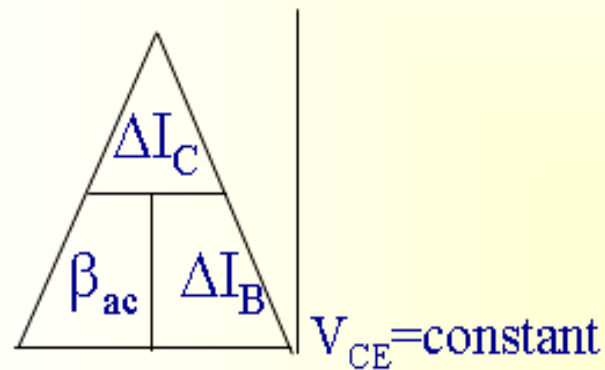
- The ratio of dc collector current ( $I_C$ ) to the dc base current ( $I_B$ ) is dc beta ( $\beta_{dc}$ ) which is dc current gain where  $I_C$  and  $I_B$  are determined at a particular operating point, Q-point (quiescent point).
- It's define by the following equation:

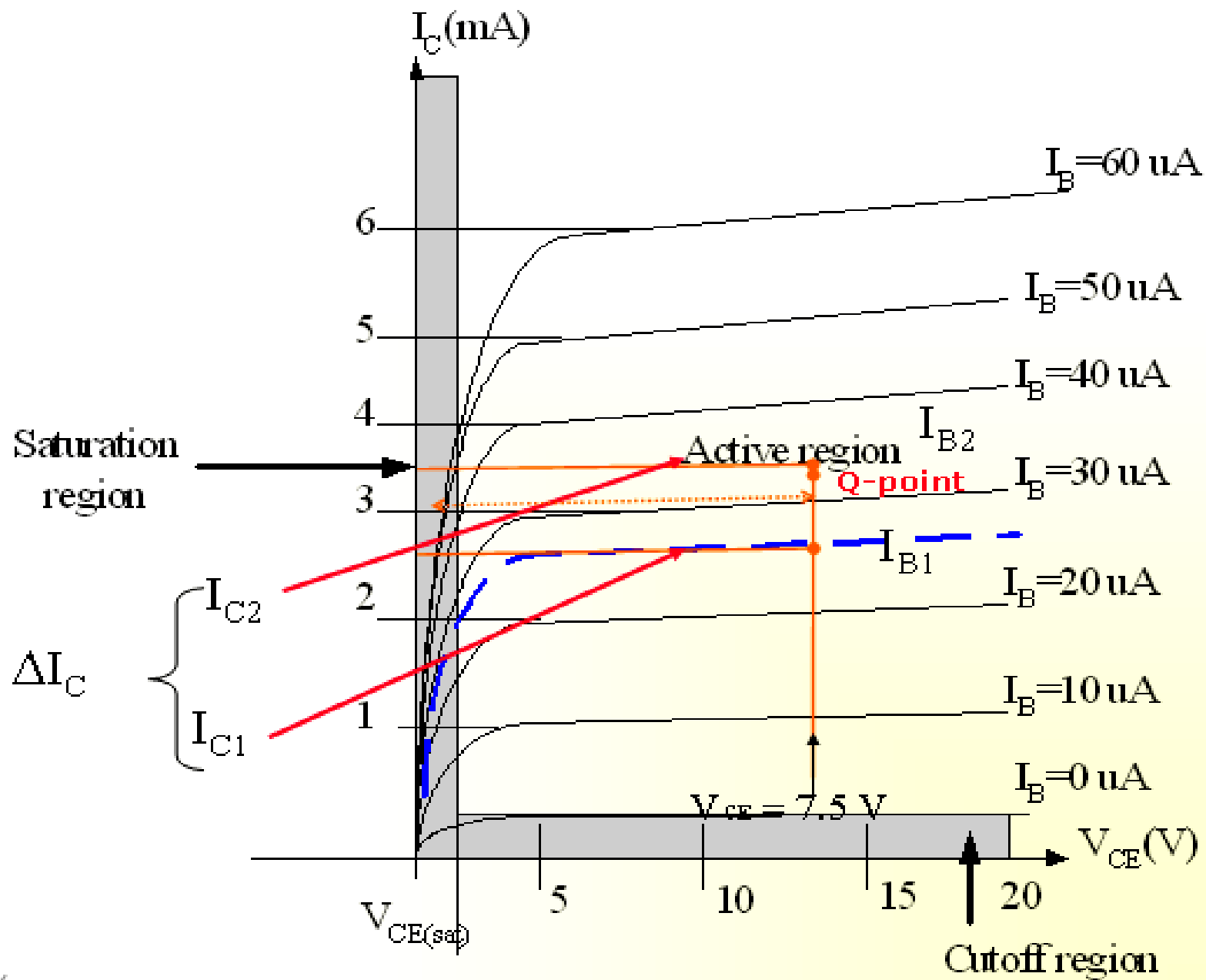
$$30 < \beta_{dc} < 300 \diamond 2N3904$$



- On data sheet,  $\beta_{dc} = h_{FE}$  with  $h$  is derived from ac hybrid equivalent cct. FE are derived from forward-current amplification and common-emitter configuration respectively.

- For ac conditions an ac beta has been defined as the changes of collector current ( $I_C$ ) compared to the changes of base current ( $I_B$ ) where  $I_C$  and  $I_B$  are determined at operating point.
- On data sheet,  $\beta_{ac} = h_{fe}$
- It can be defined by the following equation:





# Relationship analysis between $\alpha$ and $\beta$

CASE 1

$$I_E = I_C + I_B \quad (1)$$

substitute equ.  $I_C = \beta I_B$  into (1) we get

$$\underline{\underline{I_E = (\beta + 1)I_B}}$$

CASE 2

$$\text{known : } \alpha = \frac{I_C}{I_E} \Rightarrow I_E = \frac{I_C}{\alpha} \quad (2)$$

$$\text{known : } \beta = \frac{I_C}{I_B} \Rightarrow I_B = \frac{I_C}{\beta} \quad (3)$$

substitute (2) and (3) into (1) we get,

$$\underline{\underline{\alpha = \frac{\beta}{\beta + 1}}}$$

and

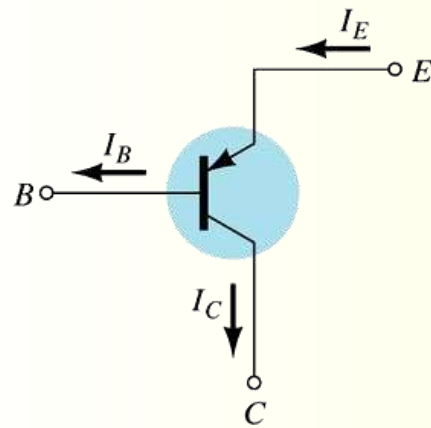
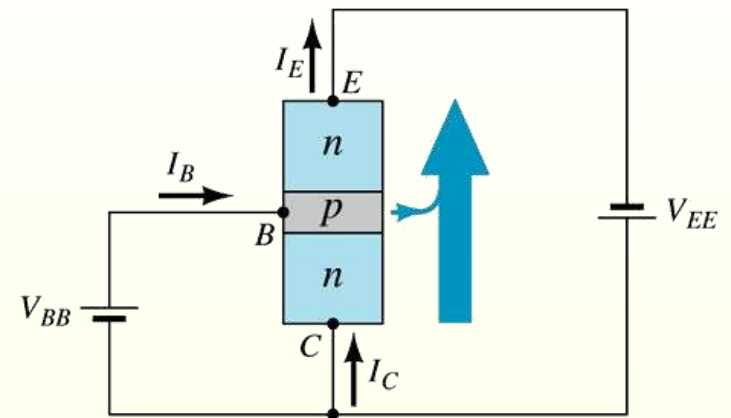
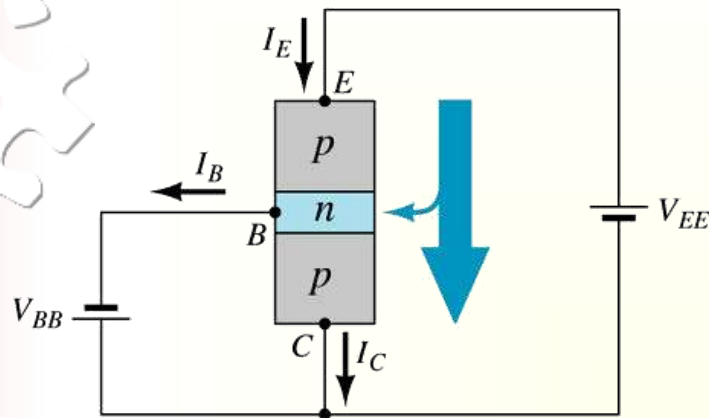
$$\underline{\underline{\beta = \frac{\alpha}{1 - \alpha}}}$$



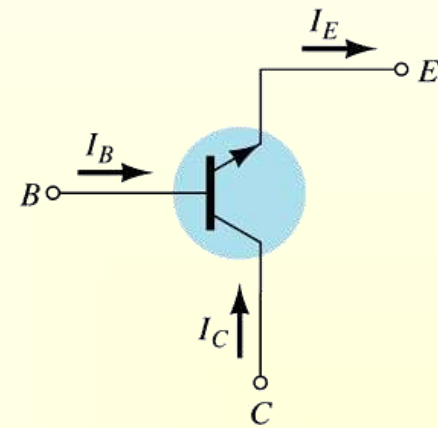
# Common – Collector Configuration

- Also called emitter-follower (EF).
- It is called common-emitter configuration since both the signal source and the load share the collector terminal as a common connection point.
- The output voltage is obtained at emitter terminal.
- The input characteristic of common-collector configuration is similar with common-emitter configuration.
- Common-collector circuit configuration is provided with the load resistor connected from emitter to ground.
- It is used primarily for impedance-matching purpose since it has high input impedance and low output impedance.





(a)



(b)

Notation and symbols used with the common-collector configuration:  
 (a) pnp transistor ; (b) npn transistor.

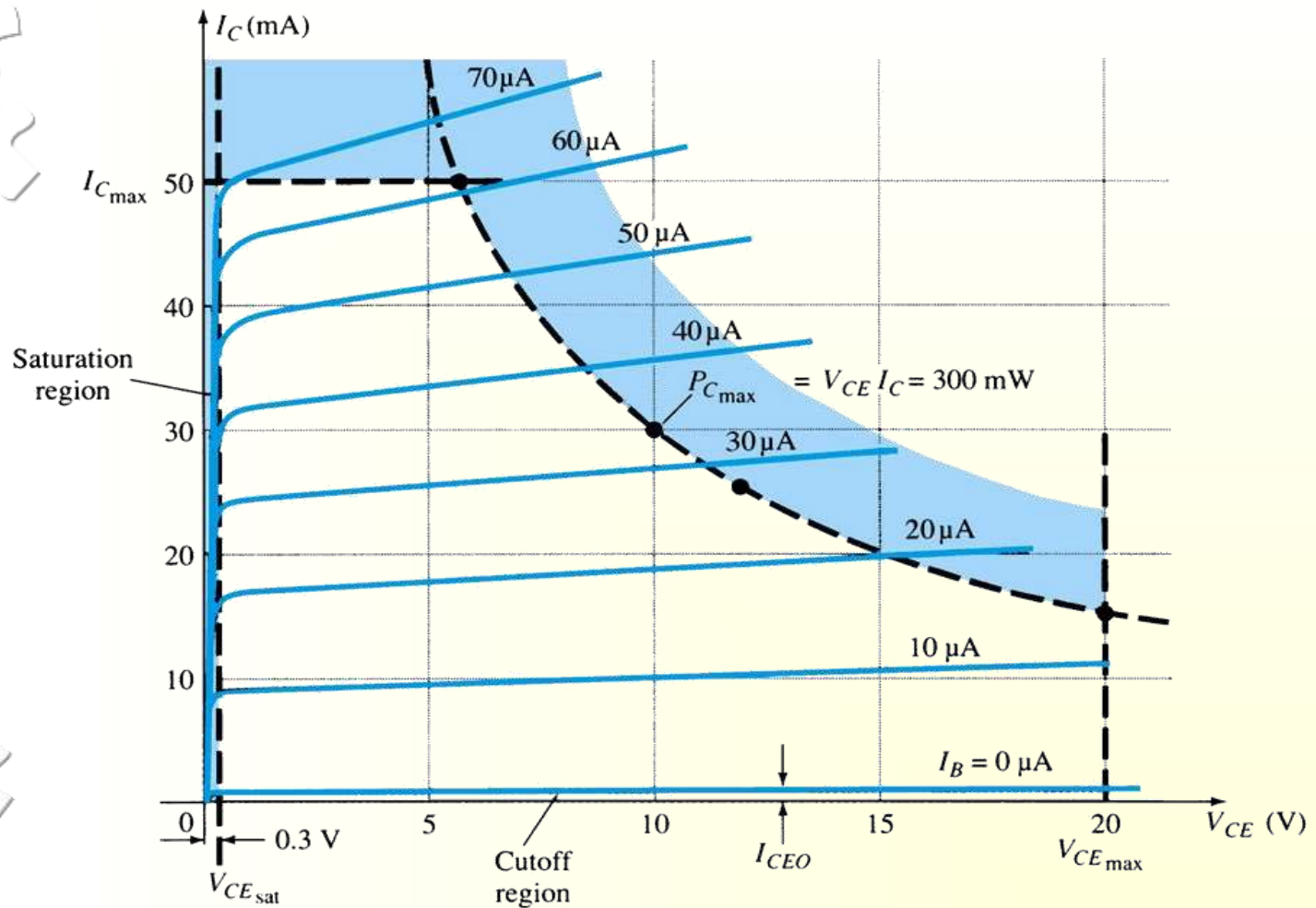
- 
- Saturation region**



**Saturation region**

# Limits of Operation

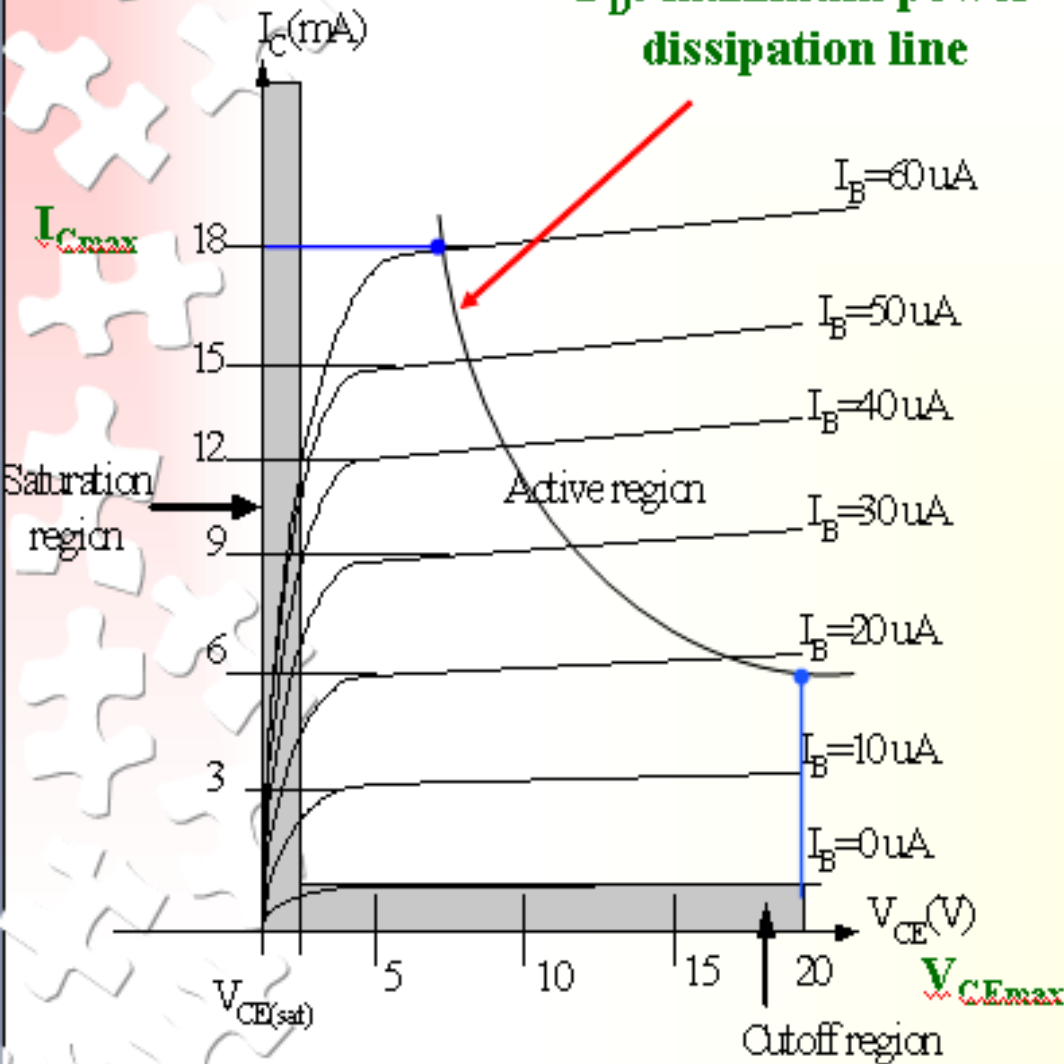
- Many BJT transistor used as an amplifier. Thus it is important to notice the limits of operations.
- At least 3 maximum values is mentioned in data sheet.
- There are:
  - a) Maximum power dissipation at collector:  $P_{G_{max}}$  or  $P_D$
  - b) Maximum collector-emitter voltage:  $V_{CE_{max}}$  sometimes named as  $V_{BR(CEO)}$  or  $V_{CEO}$ .
  - c) Maximum collector current:  $I_{C_{max}}$
- There are few rules that need to be followed for BJT transistor used as an amplifier. The rules are:
  - i) transistor need to be operate in active region!
  - ii)  $I_C < I_{G_{max}}$
  - ii)  $P_C < P_{G_{max}}$



Note:  $V_{CE}$  is at maximum and  $I_C$  is at minimum ( $I_{C\max} = I_{CEO}$ ) in the cutoff region.  $I_C$  is at maximum and  $V_{CE}$  is at minimum ( $V_{CE\max} = V_{CE\text{sat}} = V_{CEO}$ ) in the saturation region. The transistor operates in the active region between saturation and cutoff.

# Example 1:

$P_D$ : maximum power dissipation line



Refer to the fig.

Step 1:

The maximum collector power dissipation,

$$P_D = I_{Cmax} \times V_{CEmax} \quad (1)$$

$$= 18 \text{ m} \times 20 = 360 \text{ mW}$$

Step 2:

At any point on the characteristics the product of and must be equal to 360 mW.

Ex. 1. If choose  $I_{Cmax} = 5 \text{ mA}$ , substitute into the (1), we get

$$V_{CEmax} I_{Cmax} = 360 \text{ mW}$$

$$V_{CEmax} (5 \text{ m}) = 360 / 5 = \underline{7.2 \text{ V}}$$

Ex.2. If choose  $V_{CEmax} = 18 \text{ V}$ , substitute into (1), we get

$$V_{CEmax} I_{Cmax} = 360 \text{ mW}$$

$$(10) I_{Cmax} = 360 \text{ m} / 18 = \underline{20 \text{ mA}}$$



# Derating $P_{Dmax}$

- $P_{Dmax}$  is usually specified at 25°C.
- The higher temperature goes, the less is  $P_{Dmax}$
- Example;
  - A derating factor of 2mW/°C indicates the power dissipation is reduced 2mW each degree centigrade increase of temperature.



# Transistor Specification Sheet

## MAXIMUM RATINGS

Rating	Symbol	2N4123	Unit
Collector-Emitter Voltage	$V_{CE0}$	30	Vdc
Collector-Base Voltage	$V_{CB0}$	40	Vdc
Emitter-Base Voltage	$V_{EB0}$	5.0	Vdc
Collector Current - Continuous	$I_C$	200	mA
Total Device Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	625 5.0	mW mW/°C
Operating and Storage Junction Temperature Range	$T_j, T_{stg}$	-55 to +150	°C

## THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	83.3	°C/W
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	200	°C/W

## ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
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## OFF CHARACTERISTICS

Collector-Emitter Breakdown Voltage (1) ( $I_C = 1.0\text{ mA}$ , $I_E = 0$ )	$V_{(BR)CE0}$	30		Vdc
Collector-Base Breakdown Voltage ( $I_C = 10\text{ }\mu\text{A}$ , $I_E = 0$ )	$V_{(BR)CB0}$	40		Vdc
Emitter-Base Breakdown Voltage ( $I_E = 10\text{ }\mu\text{A}$ , $I_C = 0$ )	$V_{(BR)EB0}$	5.0	—	Vdc
Collector Cutoff Current ( $V_{CE} = 20\text{ Vdc}$ , $I_E = 0$ )	$I_{CBO}$	—	50	nA
Emitter Cutoff Current ( $V_{BE} = 3.0\text{ Vdc}$ , $I_C = 0$ )	$I_{EBO}$	—	50	nA

## ON CHARACTERISTICS

DC Current Gain (1) ( $I_C = 2.0\text{ mA}$ , $V_{CE} = 1.0\text{ Vdc}$ ) ( $I_C = 50\text{ mA}$ , $V_{CE} = 1.0\text{ Vdc}$ )	$h_{FE}$	50 25	150 —	—
Collector-Emitter Saturation Voltage (1) ( $I_C = 50\text{ mA}$ , $I_B = 5.0\text{ mA}$ )	$V_{CE(sat)}$	—	0.3	Vdc
Base-Emitter Saturation Voltage (1) ( $I_C = 50\text{ mA}$ , $I_B = 5.0\text{ mA}$ )	$V_{BE(sat)}$	—	0.95	Vdc

## SMALL-SIGNAL CHARACTERISTICS

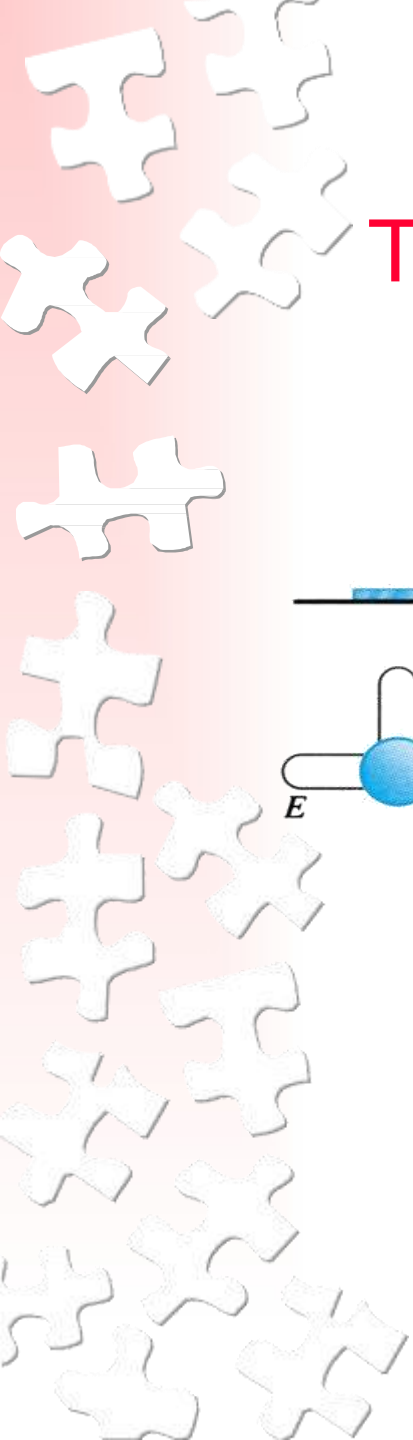
Current-Gain - Bandwidth Product ( $I_C = 10\text{ mA}$ , $V_{CE} = 20\text{ Vdc}$ , $f = 100\text{ MHz}$ )	$f_T$	250		MHz
Output Capacitance ( $V_{CE} = 5.0\text{ Vdc}$ , $I_E = 0$ , $f = 100\text{ MHz}$ )	$C_{obo}$	—	4.0	pF
Input Capacitance ( $V_{BE} = 0.5\text{ Vdc}$ , $I_C = 0$ , $f = 100\text{ kHz}$ )	$C_{ibo}$	—	8.0	pF
Collector-Base Capacitance ( $I_B = 0$ , $V_{CB} = 5.0\text{ V}$ , $f = 100\text{ kHz}$ )	$C_{cb}$	—	4.0	pF
Small-Signal Current Gain ( $I_C = 2.0\text{ mA}$ , $V_{CE} = 10\text{ Vdc}$ , $f = 1.0\text{ kHz}$ )	$h_{ie}$	50	200	—
Current Gain - High Frequency ( $I_C = 10\text{ mA}$ , $V_{CE} = 20\text{ Vdc}$ , $f = 100\text{ MHz}$ ) ( $I_C = 2.0\text{ mA}$ , $V_{CE} = 10\text{ V}$ , $f = 1.0\text{ kHz}$ )	$h_{fe}$	2.5 50	— 200	—
Noise Figure ( $I_C = 100\text{ }\mu\text{A}$ , $V_{CE} = 5.0\text{ Vdc}$ , $R_S = 1.0\text{ k ohm}$ , $f = 1.0\text{ kHz}$ )	NF	—	6.0	dB

(1) Pulse Test: Pulse Width = 300  $\mu\text{s}$ , Duty Cycle = 2.0%

**2N4123**  
CASE 29-04, STYLE 1  
TO-92 (TO-226AA)

3 Collector  
2 Base  
1 Emitter

**GENERAL PURPOSE  
TRANSISTOR**  
NPN SILICON





# Transistor Testing

1. Curve Tracer  
Provides a graph of the characteristic curves.
2. DMM  
Some DMM's will measure  $\beta_{DC}$  or  $HFE$ .
3. Ohmmeter

