

An abstract digital graphic on the left side of the slide. It features several 3D cubes in various shades of blue. The faces of the cubes are covered in a pattern of binary code (0s and 1s). Bright blue and red light beams emanate from some of the cube's edges and corners, creating a sense of depth and digital activity.

# Lecture 8-2: Nyquist Criterion, Phase Margin, and Gain Margin

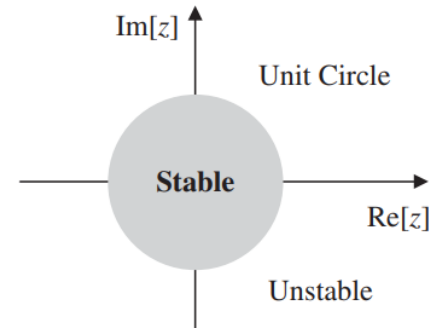
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**ELEN 472: Introduction to Digital Control**

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Louisiana Tech University

# Review



- **Stability of Discrete-time Systems**

- **Definition:** Bounded input Bounded output

- **Stability Check**

- **Manual Check:** If all poles lies inside the closed unit circle ( $|\text{poles}| < 1$ )
- **For High-order Systems:** Jury Test

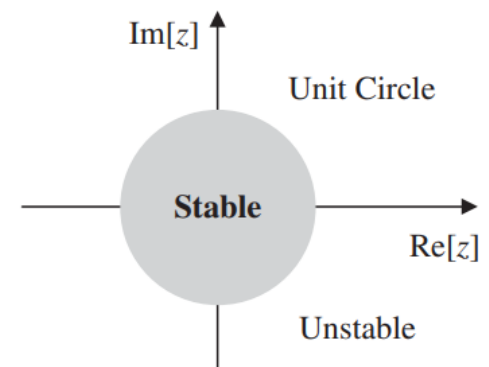
- **Internal Stability**

- **Definition:** For a closed-loop system, internal stability requires all signals in the system are bounded.
- **Two Conditions** for Internal Stability

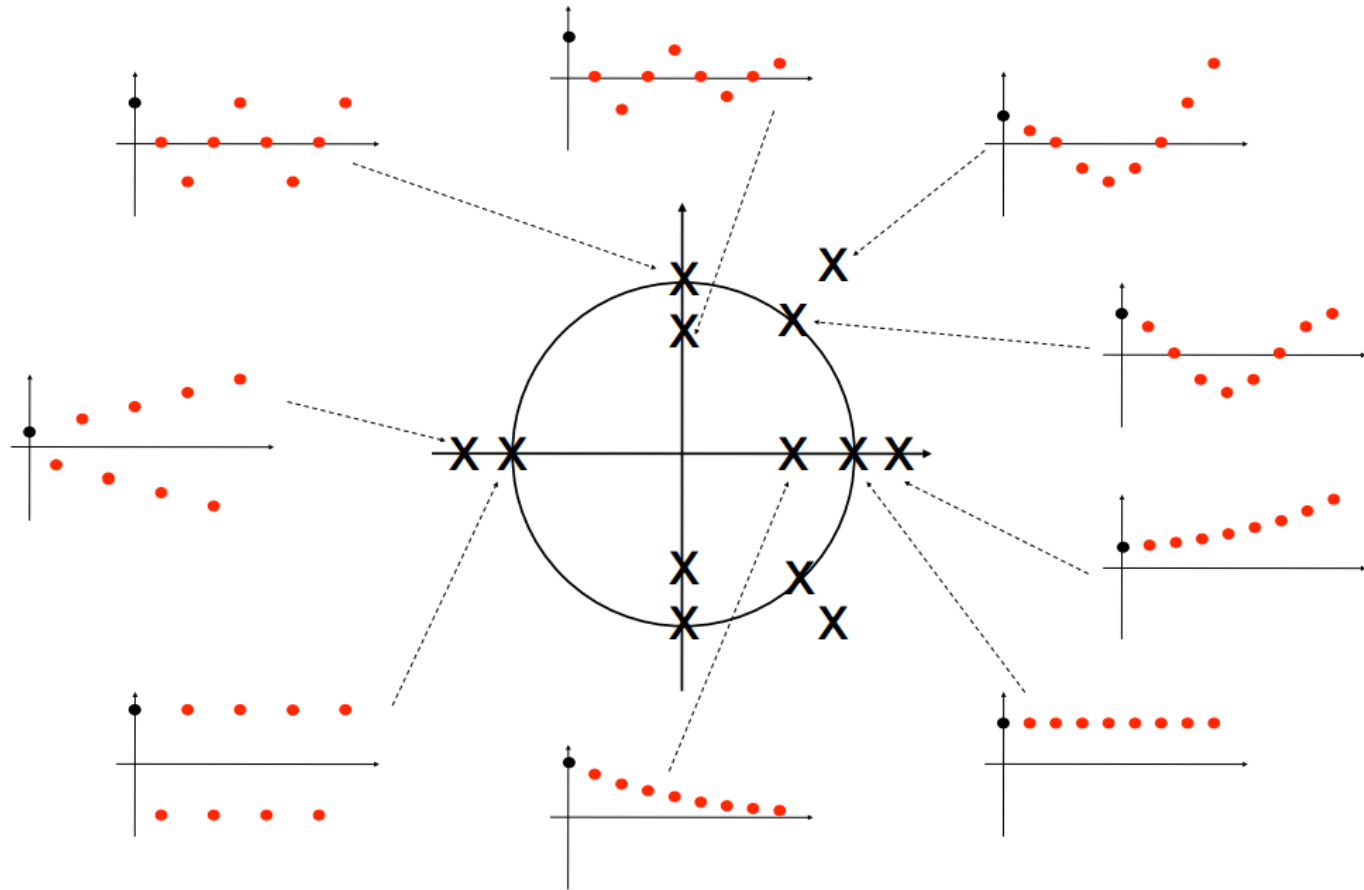
- A closed-loop system is **internal stable** if the following two conditions hold:
  - The characteristic Polynomial  $1 + C(z)G_{ZAS}(z)$  has no zeros **on** or **outside** the unit circle.
  - The loop gain  $C(z)G_{ZAS}(z)$  has no pole-zero cancellation **on** or **outside** the unit circle.

# More Details in Manual Check

- BIBO Stability Definition:
  - For a **stable** discrete-time system, if a **bounded** input signal is provided, the output will also be **bounded**.
  - Bounded Input Bounded Output – BIBO definition.
- How to determine whether a discrete-time system is stable?
  - Depending on system's poles.
    - If all poles are **within** the unit circle -> System is **stable**
    - If one or more (not repeated) poles are **on** the unit circle -> **Marginally Stable**
    - If **repeated** poles **on** the unit circle -> **Unstable**
    - If **one** pole is outside of the unit circle -> **Unstable**



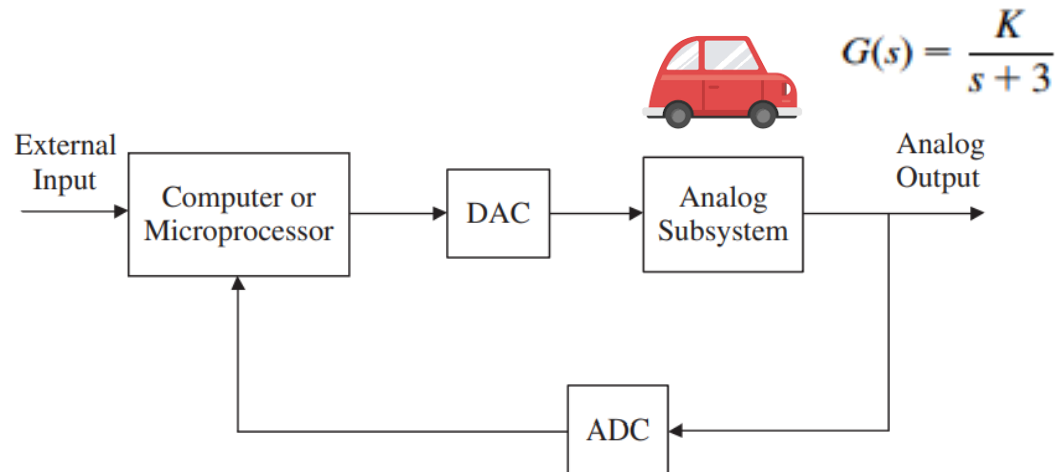
# Review of Stability for Discrete-time Systems



*Time responses as a function of the poles location*

## Example

- Find the **stable** range of the gain  $K$  for the unity feedback digital cruise control system with the analog transfer function (the sampling time is 0.02 s):



- Solution:**

- First, we obtain the overall z-domain transfer function DAC, Analog Subsystem, and ADC (i.e.,  $G_{ZAS}(z)$ )

$$\begin{aligned} G_{ZAS}(z) &= (1 - z^{-1}) \mathcal{Z} \left\{ \mathcal{Z}^{-1} \left[ \frac{G(s)}{s} \right] \right\} \\ &= (1 - z^{-1}) \mathcal{Z} \left\{ \mathcal{Z}^{-1} \left[ \frac{K}{s(s + 3)} \right] \right\} \end{aligned}$$

## Example (Continued)

- **Solution (Continued):**

$$\begin{aligned} G_{ZAS}(z) &= (1 - z^{-1}) \mathcal{Z} \left\{ \mathcal{Z}^{-1} \left[ \frac{G(s)}{s} \right] \right\} \\ &= (1 - z^{-1}) \mathcal{Z} \left\{ \mathcal{Z}^{-1} \left[ \frac{K}{s(s+3)} \right] \right\} \end{aligned}$$

- Using the partial fraction expansion:

$$\frac{K}{s(s+3)} = \frac{K}{3} \left[ \frac{1}{s} - \frac{1}{s+3} \right]$$

- We obtain the transfer function:

$$G_{ZAS}(z) = \frac{1.9412 \times 10^{-2} K}{z - 0.9418}$$

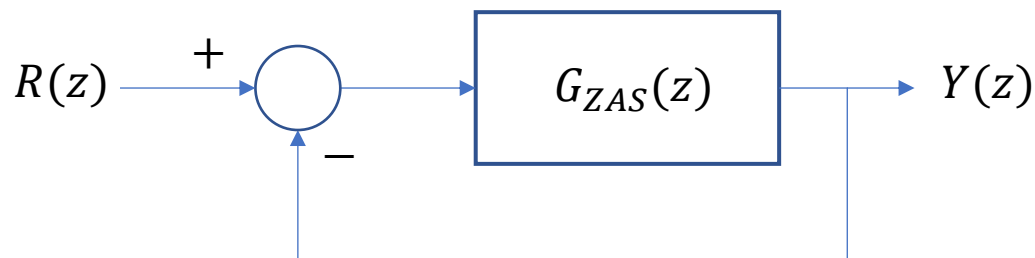
Note: Z-transform pairs

$$\frac{1}{s} \rightarrow \frac{1}{1 - z^{-1}}$$

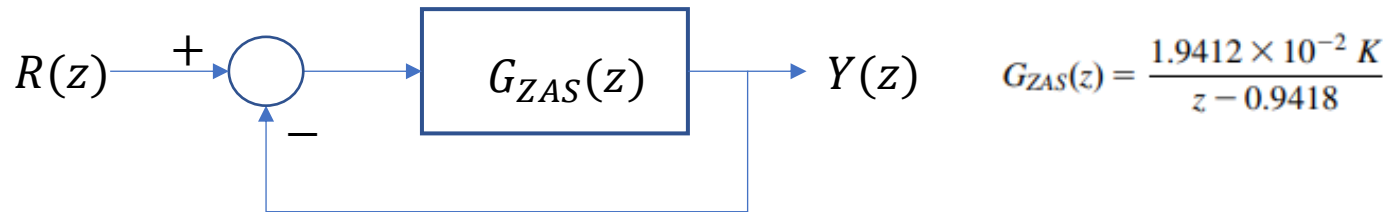
$$\frac{1}{s+a} \rightarrow \frac{1}{1 - e^{-aT} z^{-1}}$$

( $T = 0.02$  for this question)

- Thus, the original system diagram can be simplified as:



## Example (Continued)



### • Solution (Continued)

- For the above unity feedback systems, the closed-loop characteristic equation is

$$1 + G_{ZAS}(z) = 0$$

- Which can be simplified to:

$$z - 0.9418 + 1.9412 \times 10^{-2} K = 0$$

- From the above equation, we have a pole at

$$z = 0.9418 - 1.9412 \times 10^{-2} K$$

- To make the system stable, we need to make sure that this pole is **within** the **unit circle**, i.e.,  $|z| < 1$  or  $-1 < z < 1$ .

- Thus, the stability conditions are

$$\begin{aligned} 0.9418 - 1.9412 \times 10^{-2} K &< 1 & K < 100.03 \\ -0.9418 + 1.9412 \times 10^{-2} K &< 1 & K < 100.03 \end{aligned}$$

- Thus, the stable range of K is

$$-3 < K < 100.03$$

# Nyquist Criterion Introduction

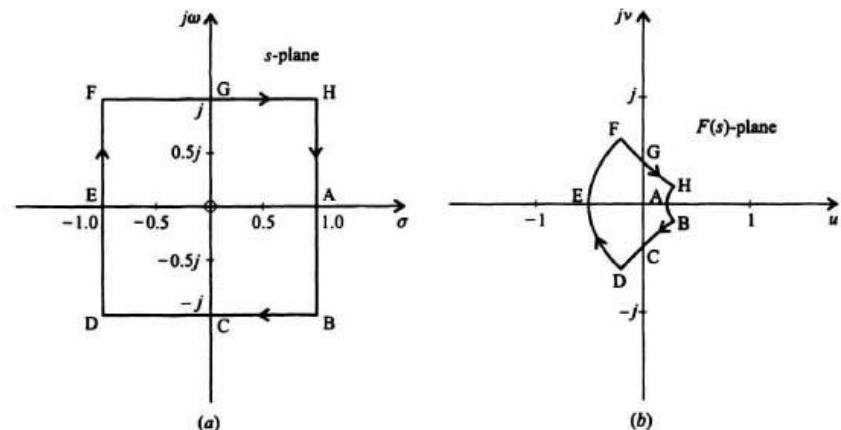
- The Nyquist Criterion is a tool that solves two questions:
  - Does the system have closed-loop **poles outside the unit circle**?
  - If yes, **how many poles** are outside the unit circle?
- The answer of the first question can be used to determine the system's stability.
  - If the system have closed-loop poles outside the unit circle (no matter how many) -> The system is unstable.
- Before we introduce the Nyquist Criterion, we need to understand the concept of **Contour**.

## Definition of Contour:

A Contour is a **closed directed** simple curve.

## A contour must be

- Closed
- Have a Direction (either clockwise or counter-clockwise)





# Nyquist Curve

- **Nyquist Curve:**

- It is a **contour plot** of the **frequency response** of a system.
- For a continuous-time system  $L(s)$ , the **frequency response** is  $L(j\omega)$ ,  $\omega \in [0, \infty)$ .

$|L(j\omega)|$  and  $\angle L(j\omega)$   **Continuous-time System frequency response**

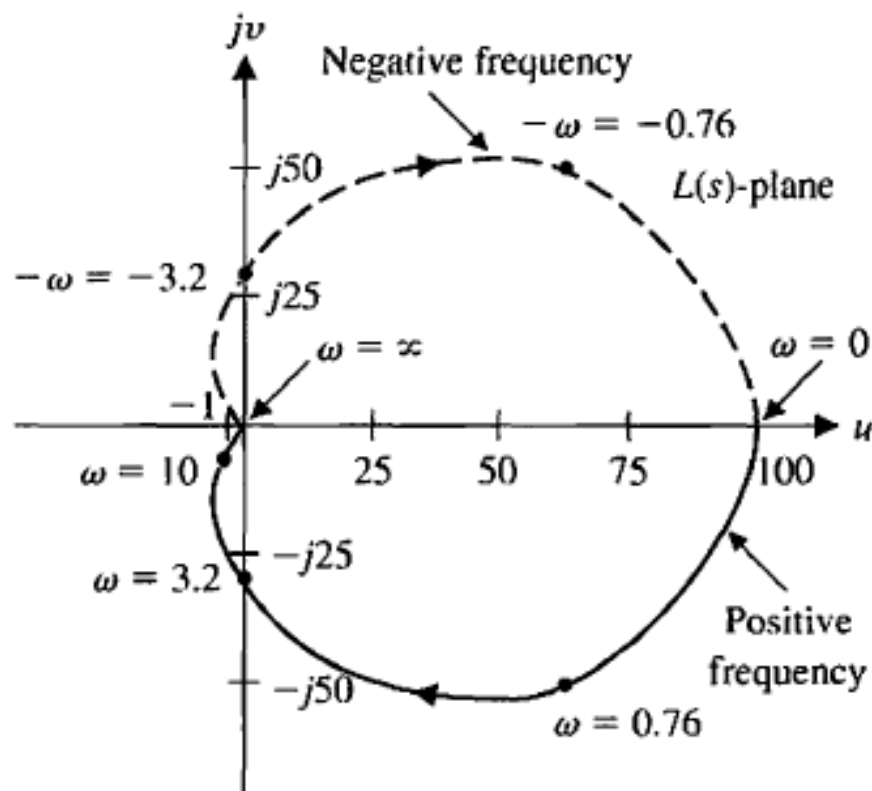
- The **graphical contour** of frequency response of  $L(s)$  -> **Nyquist Curve**.
- For instance, let  $L(s) = \frac{100}{(s+1)(\frac{1}{10}s+1)}$ ,
- We can calculate  $|L(j\omega)|$  and  $\angle L(j\omega)$  with different  $\omega$  values.

$\omega$	0	0.1	0.76	1	2	10	20	100	$\infty$
$ L(j\omega) $	100	96	79.6	70.7	50.2	6.8	2.24	0.10	0
$\angle L(j\omega)$ (degrees)	0	-5.7	-41.5	-50.7	-74.7	-129.3	-150.5	-173.7	-180

- Then, make a diagram for  $|L(j\omega)|$  and  $\angle L(j\omega)$ , we will have

# Nyquist Curve

$\omega$	0	0.1	0.76	1	2	10	20	100	$\infty$
$ L(j\omega) $	100	96	79.6	70.7	50.2	6.8	2.24	0.10	0
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Nyquist Curve

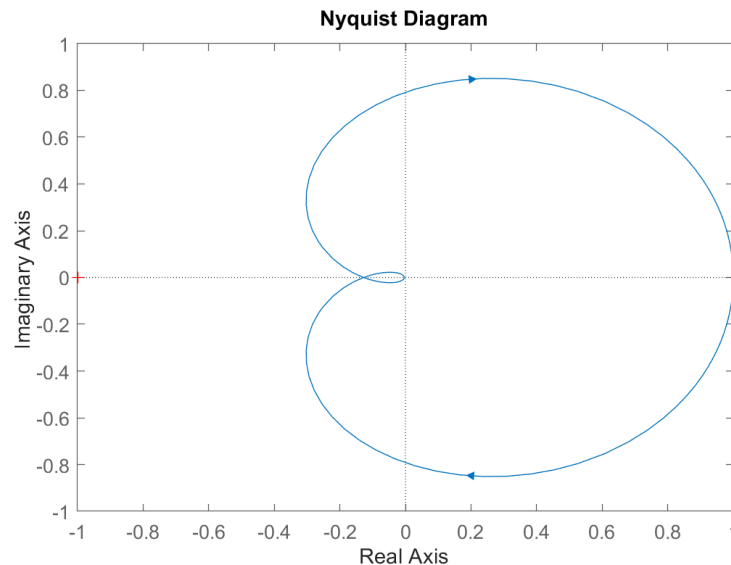
# Nyquist Curve

- For a **discrete-time** system  $G(z)$ , its **frequency response** include:  
 $|G(e^{j\omega T})| \angle G(e^{j\omega T})$ 
  - With different values of  $\omega$ , we can also generate a **Nyquist Curve**.

- For instance, if

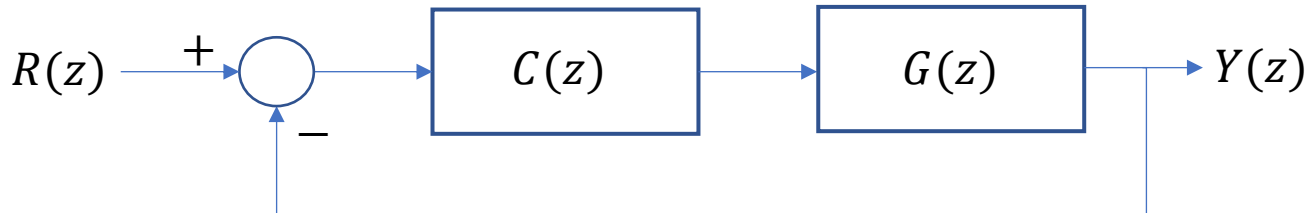
$$G(z) = \frac{0.066z + 0.055}{z^2 - 1.45z + 0.571}$$

- Assume sample period  $T = 0.4$ , the **Nyquist Curve** can be plotted as:



## An Important Observation

- Consider a closed-loop discrete-time control system as shown below:



- The TF of the above system is

$$H(z) = \frac{C(z)G(z)}{1 + C(z)G(z)}$$

- Thus, the characteristic equation is

$$1 + C(z)G(z) = 0$$

- Important Observation:

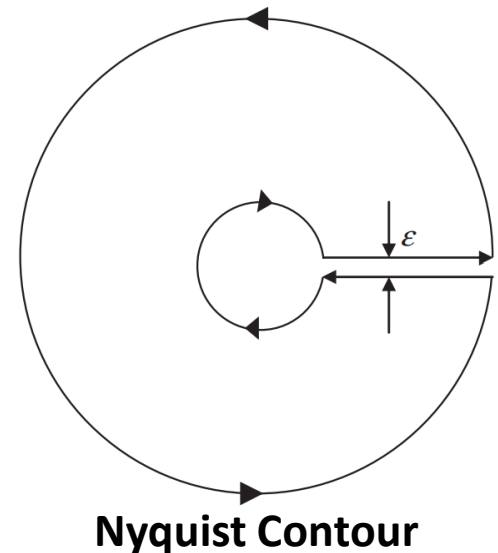
- The **Zeros** of  $1 + C(z)G(z) = 0$  are equal to the **Poles** of  $H(z)$
- We can quickly verify it by assuming  $C(z) = 1$  and  $G(z) = \frac{z}{z-1}$

## An Important Observation (Continued)

The **Zeros** of  $1 + C(z)G(z) = 0$  are equal to the **Poles** of  $H(z)$

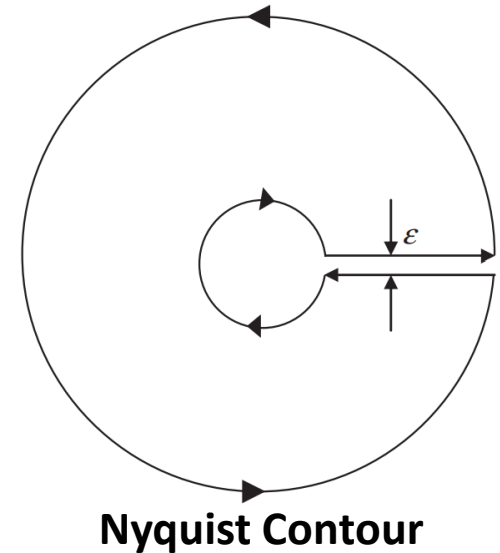
- Thus, to check the stability of a closed-loop system (i.e.,  $H(z)$ ), we can find **Zeros** of  $1 + C(z)G(z) = 0$ 
  - If all **Zeros** of  $1 + C(z)G(z) = 0$  are **within** the unit circle -> Closed-loop System is **Stable**
  - If any **Zeros** of  $1 + C(z)G(z) = 0$  is **outside** the unit circle -> Closed-loop System is **Unstable**
- Unstable Region: Outside of the unit circle.
  - We can present the Unstable Region as a contour:
    - The inner circle is the unit circle.
    - The outer circle includes everything outside.

If any **Zeros** of  $1 + C(z)G(z) = 0$  lie within **Nyquist Contour** -> The Closed-loop System is Unstable



## An Important Observation (Continued)

If any **Zeros** of  $1 + C(z)G(z) = 0$  lie within **Nyquist Contour** -> The Closed-loop System is Unstable



- Therefore, for a stable closed-loop system, we don't want any Zeros of  $1 + C(z)G(z) = 0$  lie within **Nyquist Contour**.
  - Denote the **number** of Zeros of  $1+C(z)G(z)$  inside this Contour as  $Z$
  - For a Stable System,  $Z = 0$
  - For an Unstable System,  $Z \neq 0$
- How to determine the value of  $Z$ ?
  - **Nyquist Criterion**

# Nyquist Criterion

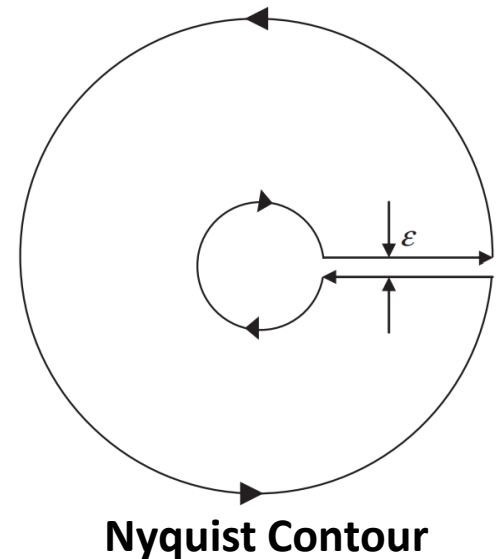
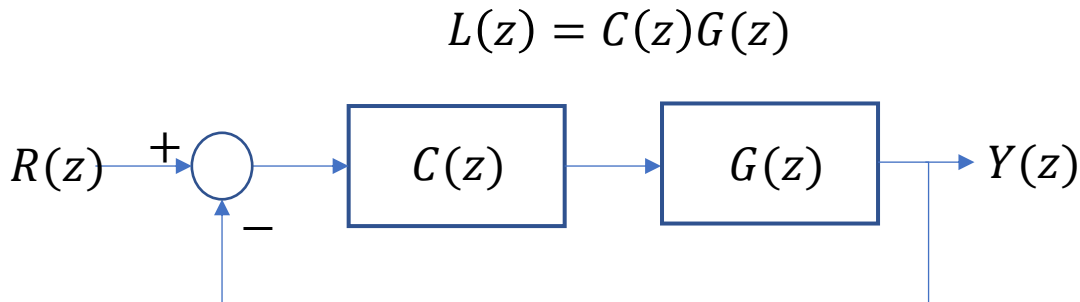
## Nyquist Criterion

A Closed-Loop system has  $Z$  poles **inside** the Nyquist Contour with

$$Z = P - N$$

$P$  is the number of Poles of  $L(z)$  **inside** the Nyquist Contour;

$N$  is the number of **anticlockwise contour encirclement** of  $(-1, 0)$  given the **Nyquist Curve** of  $L(z)$ .



# Nyquist Criterion Explanation

## Nyquist Criterion

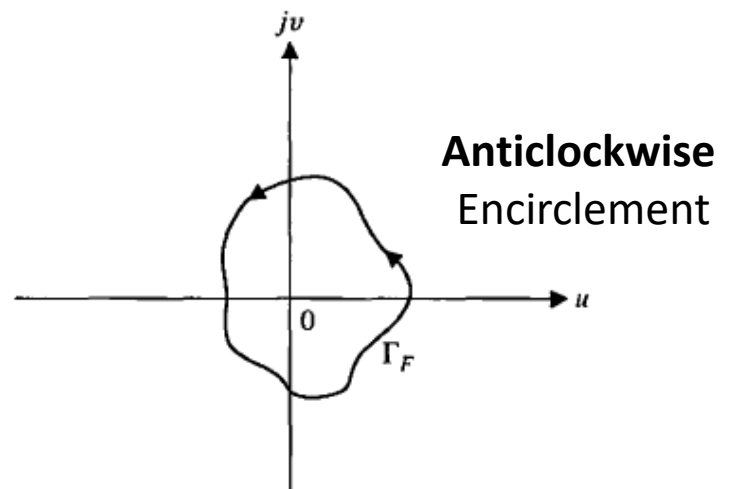
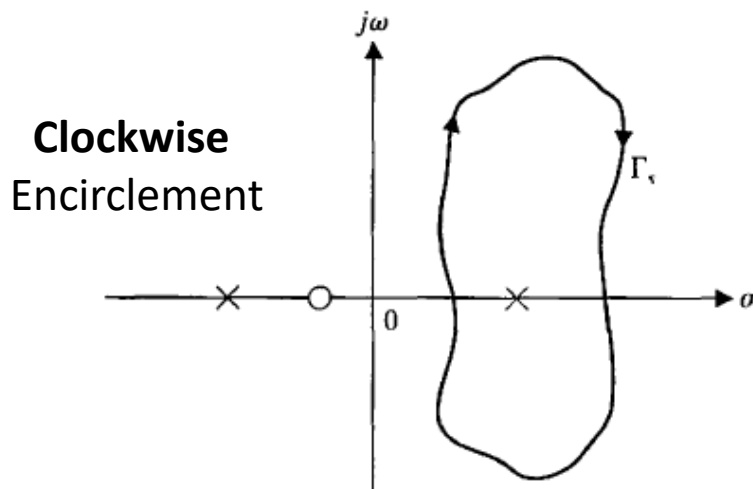
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- Examples of encirclement in **different directions**:





# Nyquist Criterion Explanation

## Nyquist Criterion

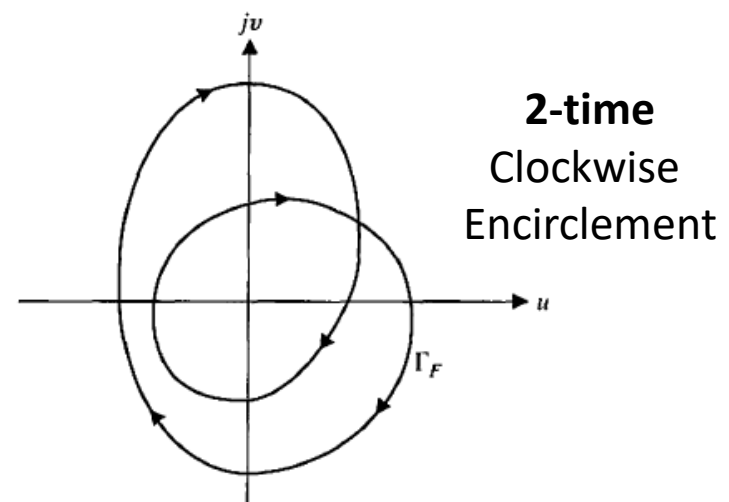
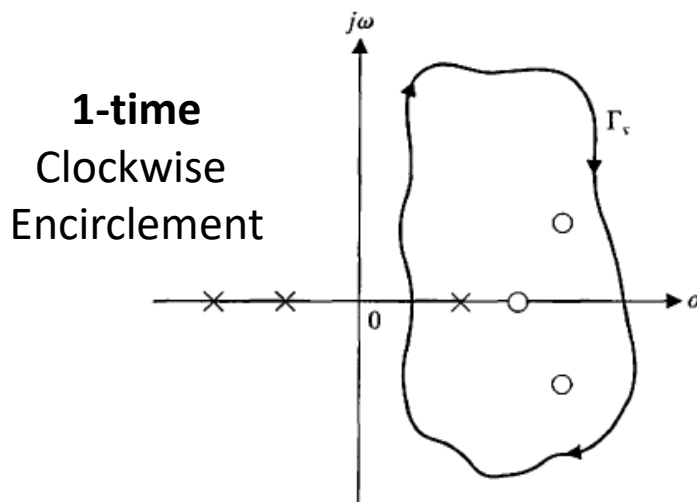
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- Examples of encirclement in **different times**:



# Nyquist Criterion Explanation

## Nyquist Criterion

A Closed-Loop system has  $Z$  poles **inside** the Nyquist Contour with

$$Z = P - N$$

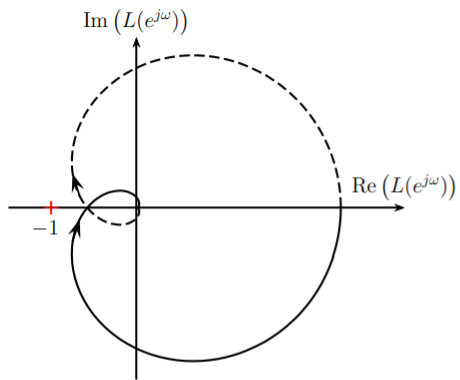
$P$  is the number of Poles of  $L(z)$  **inside** the Nyquist Contour;

$N$  is the number of **anticlockwise contour encirclement** of  $(-1, 0)$  given the **Nyquist Curve** of  $L(z)$ .

- The Nyquist Criterion is simple to use:
  - If  $L(z)$  has no poles inside the Nyquist Contour, i.e.,  $P = 0$
  - Then, for a closed-loop system to be **stable** (i.e.,  $Z = 0$ ),  **$N$  must be 0**
    - Since  $N = P - Z = 0$
    - In other words, the Nyquist Curve of  $L(z)$  must not encircle  $(-1, 0)$

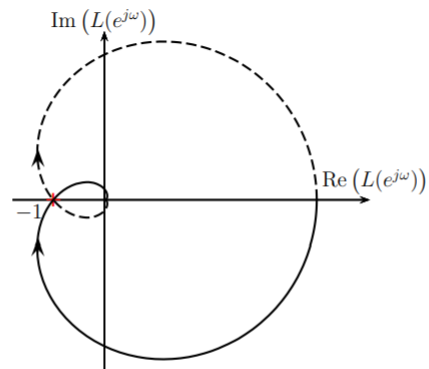
# Examples

- Given the following Nyquist Curves, determine the closed-loop system's stability.
  - Assume  $P = 0$  for all cases.



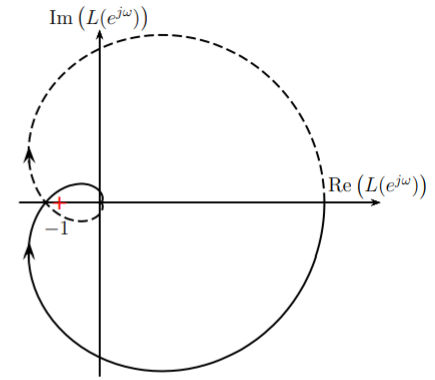
$$\Rightarrow \frac{L(z)}{1 + L(z)}$$

Stable



$$\Rightarrow \frac{L(z)}{1 + L(z)}$$

Marginally Stable

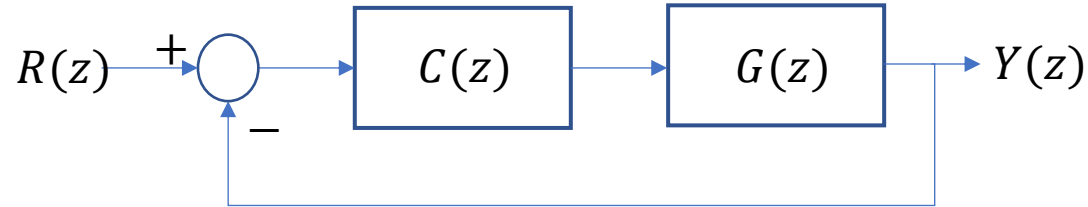


$$\Rightarrow \frac{L(z)}{1 + L(z)}$$

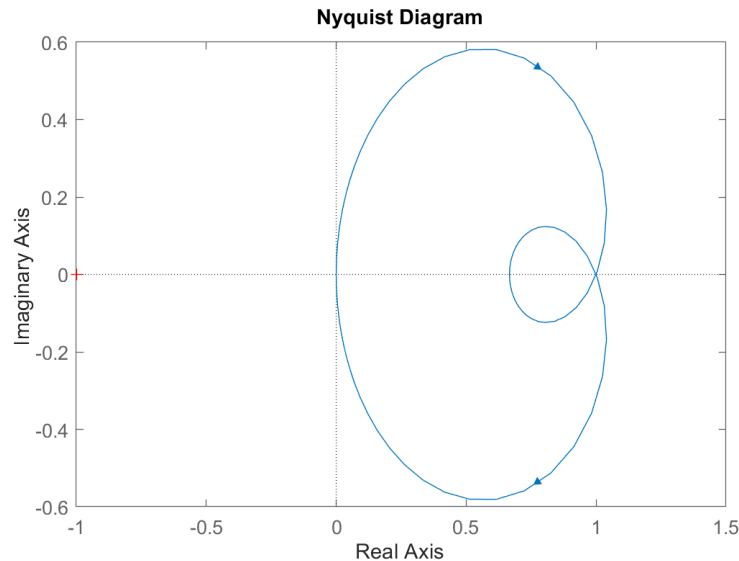
Unstable

## Example 2

- Consider the following closed-loop discrete-time



- Where  $C(z) = 1$  and  $G(z) = \frac{5z+4}{(z-2)(z-3)}$ .
- Given the Nyquist Curve of  $G(z)$ , determine the stability of this closed-loop system.



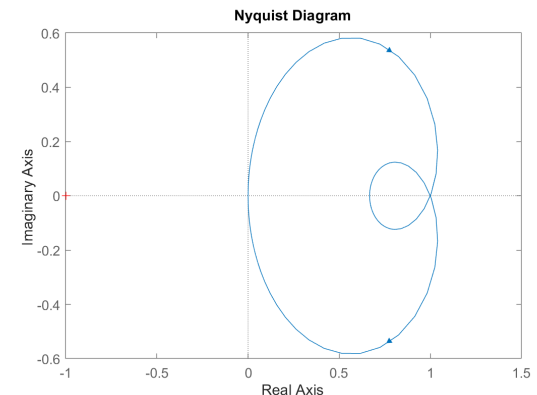
## Example 2 Solution

- **Solution:**

- We first need to determine the value of  $P$ .
  - $P$  is the number of poles of **loop gain** outside the unit circle.
  - In this work, **loop gain** is

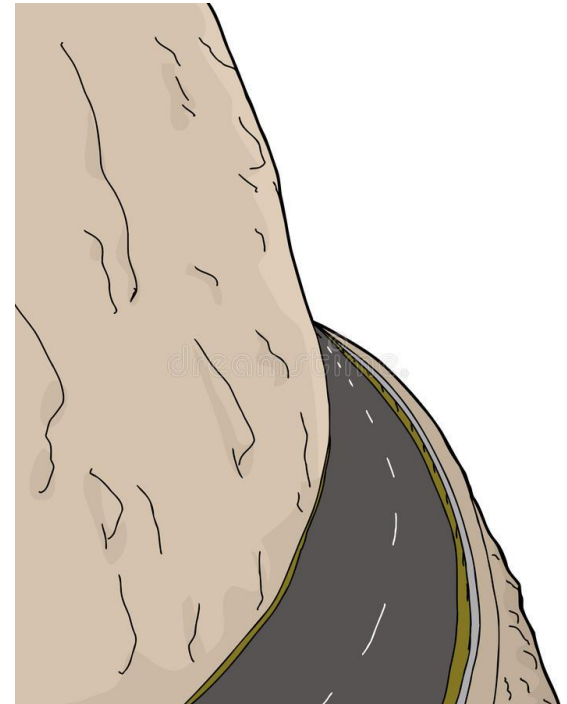
$$G(z) = \frac{5z + 4}{(z - 2)(z - 3)}$$

- Two poles are at 2 and 3, which are both outside the unit circle.
  - Thus,  $P = 2$
- For a stable system,  $Z = P - N$  should be 0
  - Here  $P = 2$ , thus,  $Z = 2 - N = 0$
  - $N$  is the encirclement of  $(-1, 0)$  for the Nyquist curve. However, in this diagram, the Nyquist curve does not encircle  $(-1, 0)$ .
  - Thus, the system is **not stable**.



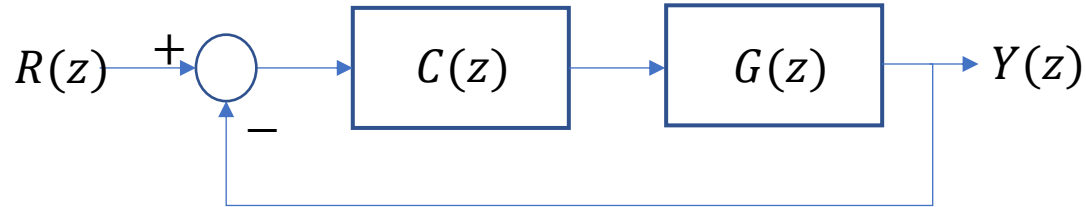
# Gain Margin and Phase Margin

- Stability Criteria – Gain Margin and Phase Margin
  - Think of both of these as **safety margins** for an open-loop system which you would like to make closed-loop.
- Take a real-world example
  - Think if you are walking next to a cliff, you want a **positive space** or “**margin**” of safety between you and a big disaster.
  - So that **positive margins** indicate there is still a safety margin before instability.
  - Conversely, **negative margins** in an open-loop system indicate instability issues if you try to close this loop.



# Closing the Loop

- Consider a Closed-loop Control System:



- Assume  $C(z) = 1$
- The closed-loop TF is:  $G_{cl}(z) = \frac{G(z)}{1+G(z)}$ 
  - Replacing  $z$  with  $e^{j\omega T}$ , we have the frequency response (assume  $T = 1$ ):
$$G_{cl}(e^{j\omega}) = \frac{G(e^{j\omega})}{1 + G(e^{j\omega})}$$
- For this closed-loop system, the **instability** occurs when:
$$1 + G(e^{j\omega}) = 0$$
- For this to happen, we need  $G(e^{j\omega}) = -1$ , which means:
  - $|G(e^{j\omega})| = 1$  and  $\angle G(e^{j\omega}) = -180^\circ$ .

# Gain Margin and Phase Margin

- Stability Margins, including **Gain Margin** and **Phase Margin**, measure how far we are from the point  $|G|=1$  and  $\angle G=-180^\circ$
- We can find Gain Margin and Phase Margin using **Bode Plot**.
  - Bode Plot is a diagram that demonstrate a system's frequency response.
  - Bode Plot includes two sub-diagrams:

## Magnitude Plot

Shows  $\omega$  vs.  $|G(e^{j\omega T})|$

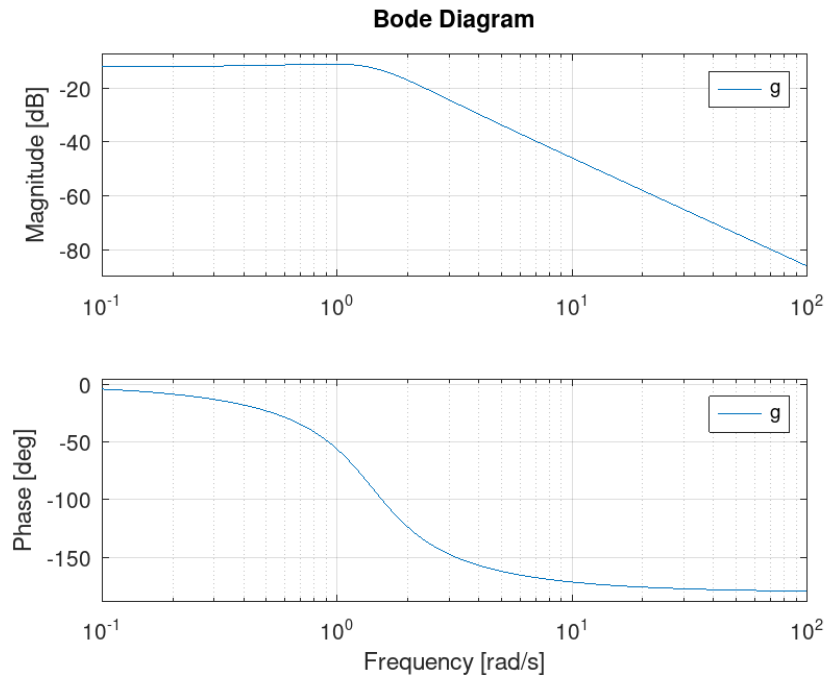
Usually, we convert magnitude in to decibel via

$$db = 20 \log_{10} |G(e^{j\omega T})|$$

## Phase Plot

Shows  $\omega$  vs.  $\angle G(e^{j\omega})$

Usually, the unit for phase angle is degree.

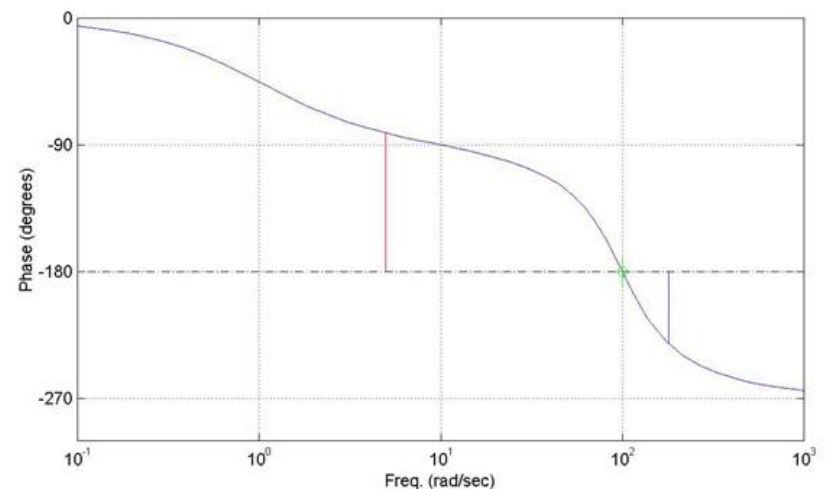
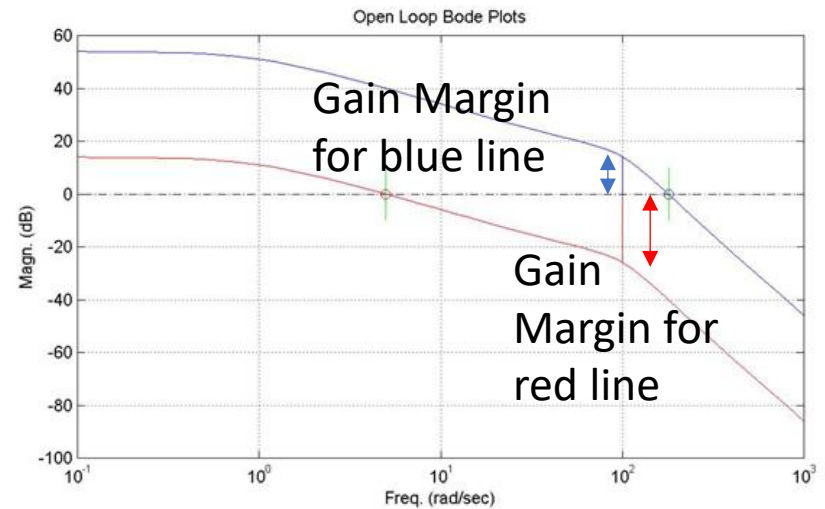


**Frequency  $\omega$**



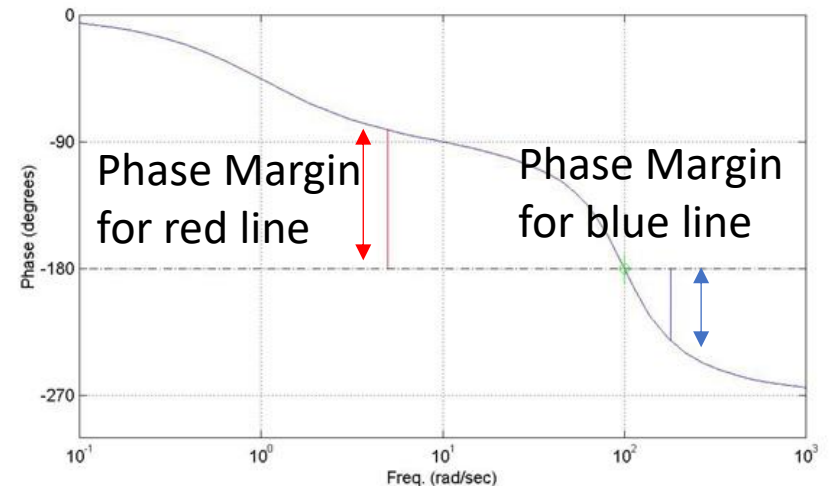
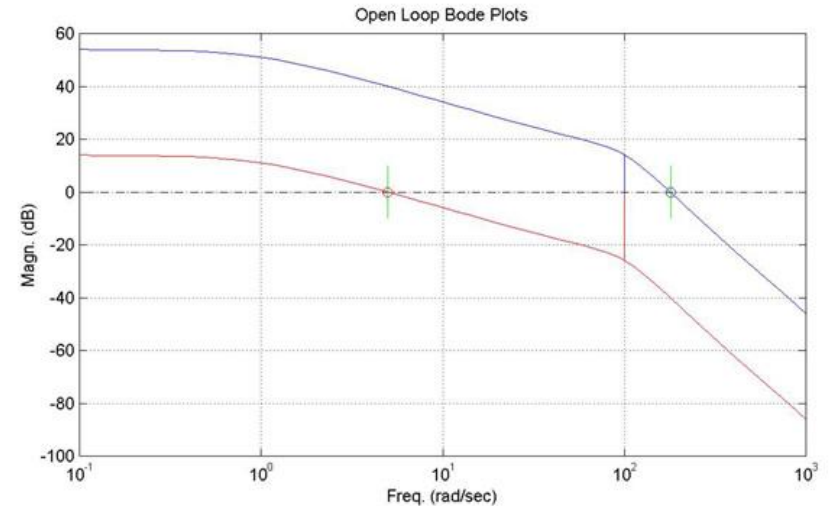
# Gain Margin

- **Gain margin (GM)** is defined as the reciprocal of the gain at the frequency at which the phase angle reaches  $-180^\circ$ .
- Here is an example of finding Gain Margin:
  1. Find the frequency where the PHASE becomes  $-180$  degrees.
    - In the pic, this frequency is 100 rad/sec.
  2. Find the GAIN,  $G$  (in dB), at the same frequency (from the upper plot).
    - For blue line,  $G = 14.1$  dB
    - For red line,  $G = -25.9$  dB
  3. **Gain Margin** =  $0 - G$  dB
    - For blue line, GM = -14.1 dB
    - For red line, GM = 25.9 dB



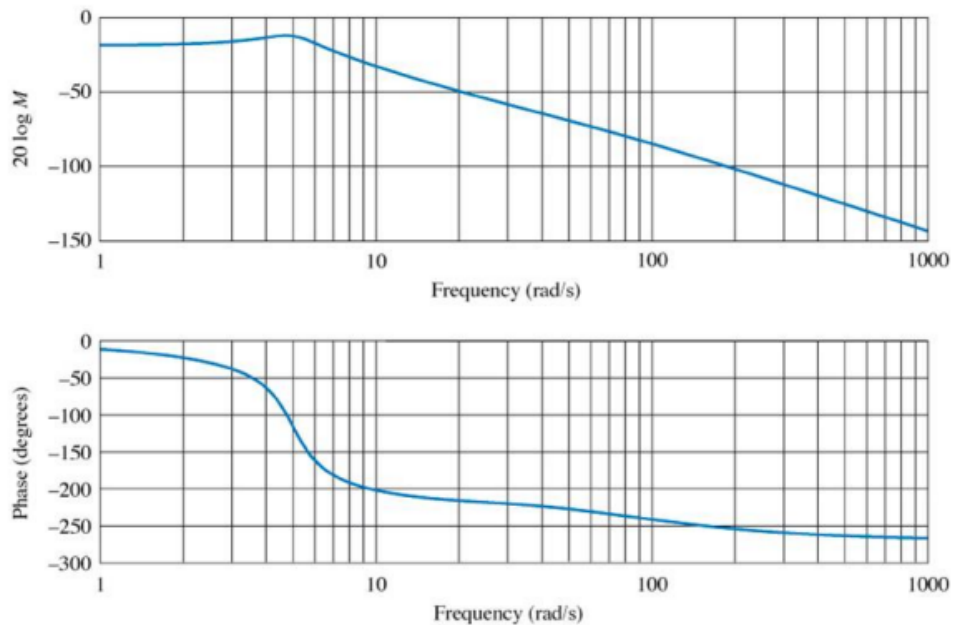
# Phase Margin

- **Phase Margin** refers to the amount of phase, which can be increased or decreased without making the system unstable.
- To find Phase Margin
  1. Find the frequency where the GAIN is 0 dB.
    - For the blue Bode plot, the 0 dB crossover occurs at 181 rad/sec; For the red Bode plot, this happens at 5 rad/sec.
  2. Find the PHASE,  $P$  (in degrees), at the same frequency (look at the lower plot)
  3. Phase Margin =  $P + 180$  degrees
    - For blue line,  $P = -231$  degrees. Thus, PM = -50.0 degrees
    - For red line,  $P = -81.3$  degrees. Thus, PM = 98.7 degrees.



# Undefined Stability Margins

- Note that sometimes the margins are undefined
  - When there is no crossover at 0 dB
  - When there is no crossover at  $180^\circ$



# Example

- Given the following Bode Plot of a closed-loop discrete-time system, determine the Gain Margin and Phase Margin.
- Solution:
  - Gain Margin:  $0 - 7.75 = -7.75$  dB
  - Phase Margin:  $-133 + 180 = 47$  degrees

