Digital Image Processing

Lecture #9 Ming-Sui (Amy) Lee

Course Information

Following Schedule

03/28	Lecture 4	05/16	Proposal
04/04	溫書假	05/23	Lecture 9
04/11	Lecture 5	05/30	Lecture 10
04/18	Lecture 6	06/06	Lecture 11
04/25	Midterm	06/13	Demo
05/02	Lecture 7	06/20	Demo
05/09	Lecture 8	06/27	Final Package Due

Announcement

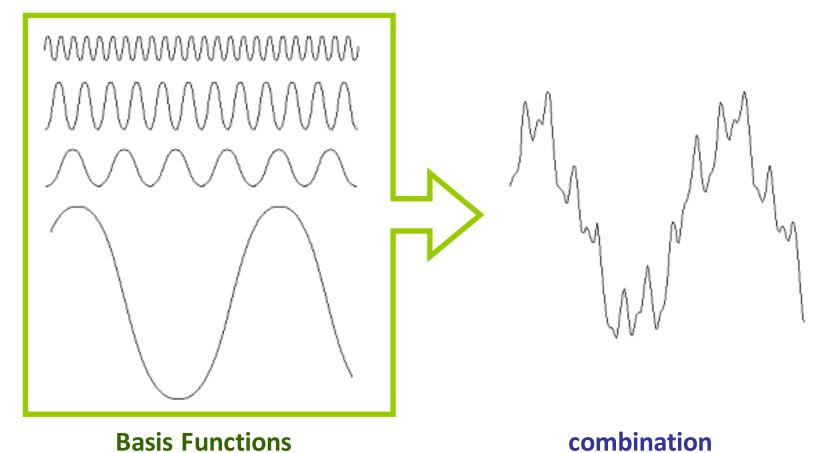
- Proposal Due: 11:59pm on May 15, 2018
 - 1) Demo video
 - 6-minute video in mpg, avi, mp4 or wmv
 - Provide a valid link on CEIBA for us to download
 - Teamxx_ProposalDemo.mpg/avi/mp4/wmv
 - 2) Written report
 - Teamxx_ProposalReport.PDF
 - Upload to CEIBA
 - Remember to include
 - Paper title / Motivation / Problem definition /
 - Algorithm / Expected results
 - Reference

Image Enhancement in Frequency Domain

Background

- Jean Baptiste Joseph Fourier (1807)
 - Any periodic function
 - Can be expressed as the sum of sines and/or cosines of different frequencies, each multiplied by a different coefficient → Fourier series
 - Functions that are not periodic
 - Can be expressed as the integral of sines and/or cosines multiplied by a weighting function → Fourier transform
 - Important characteristic
 - A function can be reconstructed completely without losing any information
 - Fourier/frequency domain processing

Example



- Continuous cases
 - (Continuous) Fourier Transform (CFT)
 - One-dimensional

$$F(u) = \int_{-\infty}^{\infty} f(x)e^{-j2\pi ux}dx$$
$$f(x) = \int_{\infty}^{\infty} F(u)e^{j2\pi ux}du$$

Euler's formula:

$$e^{j\theta} = \cos\theta + j\sin\theta$$

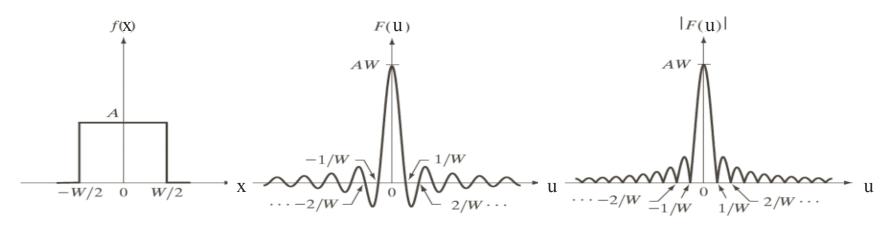
Two-dimensional

$$F(u,v) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x,y)e^{-j2\pi(ux+vy)}dxdy$$
$$f(x,y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u,v)e^{j2\pi(ux+vy)}dudv$$

Continuous Cases

• CFT of a square wave between $\left[-\frac{W}{2}, \frac{W}{2}\right]$ with amplitude A:

$$F(u) = \int_{-\infty}^{\infty} f(x) e^{-j2\pi ux} dx = \int_{-W/2}^{W/2} A e^{-j2\pi ux} dx = \frac{-A}{j2\pi u} \left[e^{-j2\pi ux} \right]_{-W/2}^{W/2}$$
$$= \frac{-A}{j2\pi u} \left[e^{-j\pi uW} - e^{j\pi uW} \right] = \frac{A}{j2\pi u} \left[e^{j\pi uW} - e^{-j\pi uW} \right] = AW \frac{\sin(\pi uW)}{\pi uW}$$



- Discrete cases
 - Discrete Fourier Transform (DFT)
 - One-dimensional

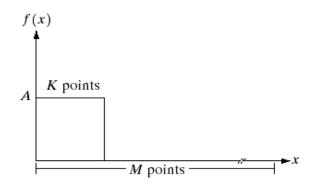
Let
$$F(u) = F(u\Delta u)$$
 $\Delta u = \frac{1}{M\Delta j}$

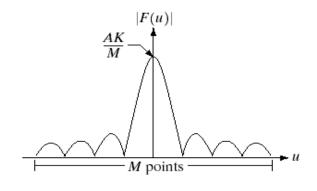
$$f(k) = f(j_0 + k\Delta j)$$

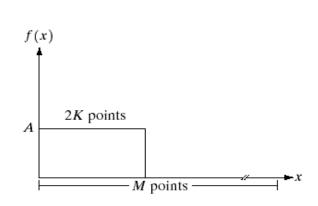
$$F(u) = \frac{1}{M} \sum_{k=0}^{M-1} f(k) e^{-j2\pi uk/M}, \quad u = 0,1,...,M-1$$

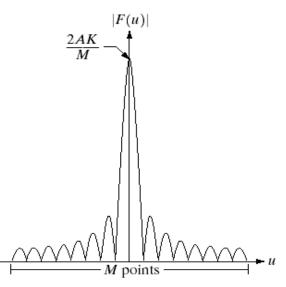
$$f(k) = \sum_{u=0}^{M-1} F(u) e^{j2\pi uk/M}, \quad k = 0,1,...,M-1$$

Example











- Discrete cases
 - One-dimensional
 - Mathematical prism
 - Separate a function into various components frequency

$$F(u) = \frac{1}{M} \sum_{k=0}^{M-1} f(k) e^{-j2\pi u k/M}, \quad u = 0,1,...,M-1$$

frequency component

$$= \frac{1}{M} \sum_{k=0}^{M-1} f(k) \left[\cos(2\pi u k / M) - j \sin(2\pi u k / M) \right]$$

Discrete cases

Two-dimensional

$$F(u,v) = \frac{1}{MN} \sum_{j=0}^{M-1} \sum_{k=0}^{N-1} f(j,k) e^{-j2\pi (\frac{uj}{M} + \frac{vk}{N})} \to \text{Complex}$$

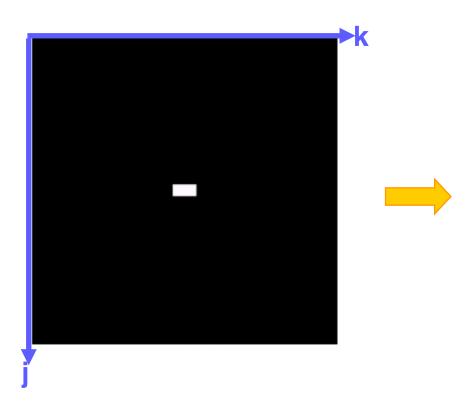
$$f(j,k) = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} F(u,v) e^{j2\pi \left(\frac{uj}{M} + \frac{vk}{N}\right)}$$

Fourier Spectrum:
$$|F(u,v)| = [R^2(u,v) + I^2(u,v)]^{/2}$$

Phase Angle:
$$\phi(u,v) = \tan^{-1} \left| \frac{I(u,v)}{R(u,v)} \right|$$

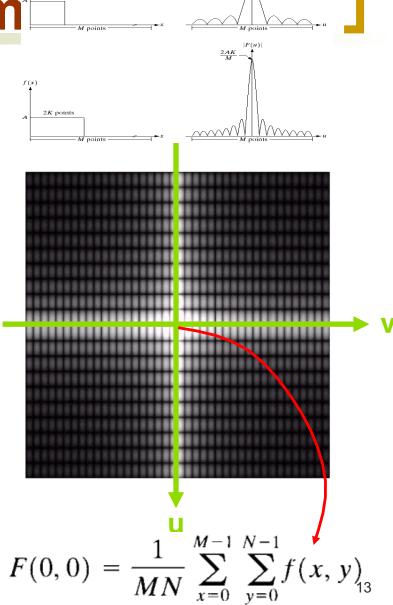
Power Spectrum:
$$P(u,v) = |F(u,v)|^2 = R^2(u,v) + I^2(u,v)|_{12}$$





$$\mathcal{F}[f(j,k)(-1)^{j+k}] = F(u - M/2, v - N/2)$$

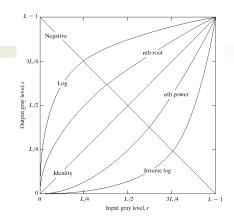
Centering & log transformation



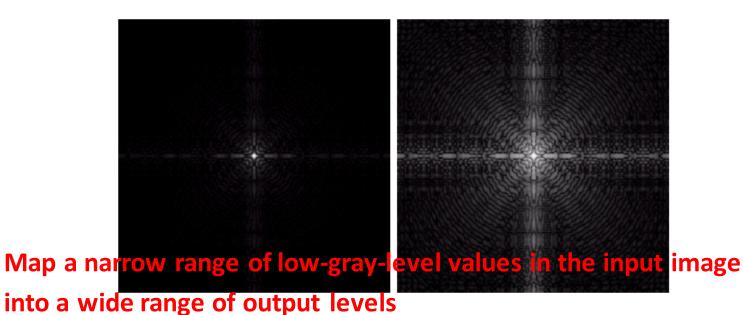
Average/dc component (zero frequency)

Log Transformation

$$s = c \log(1+r), \ r \ge 0$$



- Expand the dynamic range of low gray-level values
- Compress the dynamic range of images with large variations in pixel values



• Example Oxide protrusions

- Strong edges \rightarrow ± 45°directions
- Oxide protrusions → vertical component slightly slant to the left
- Zeros in the vertical frequency component → narrow vertical span of the oxide protrusions

- Properties
 - Centering

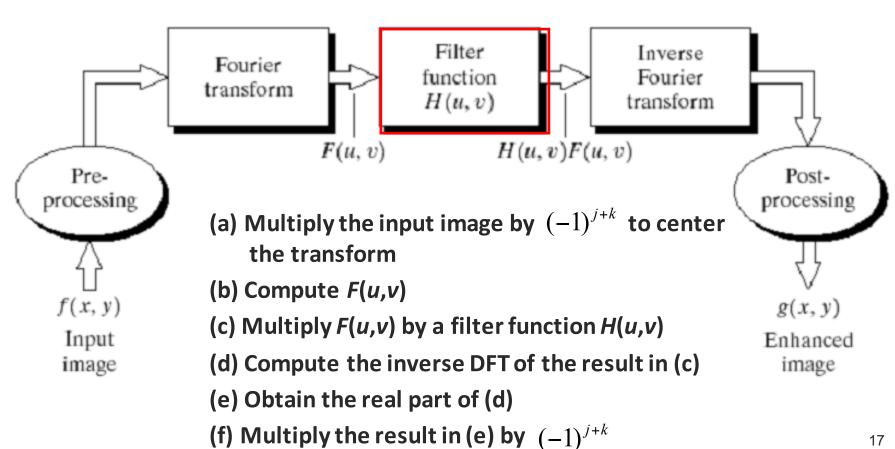
$$\mathcal{F}[f(j,k)(-1)^{j+k}] = F(u - M/2, v - N/2)$$

Conjugate symmetry

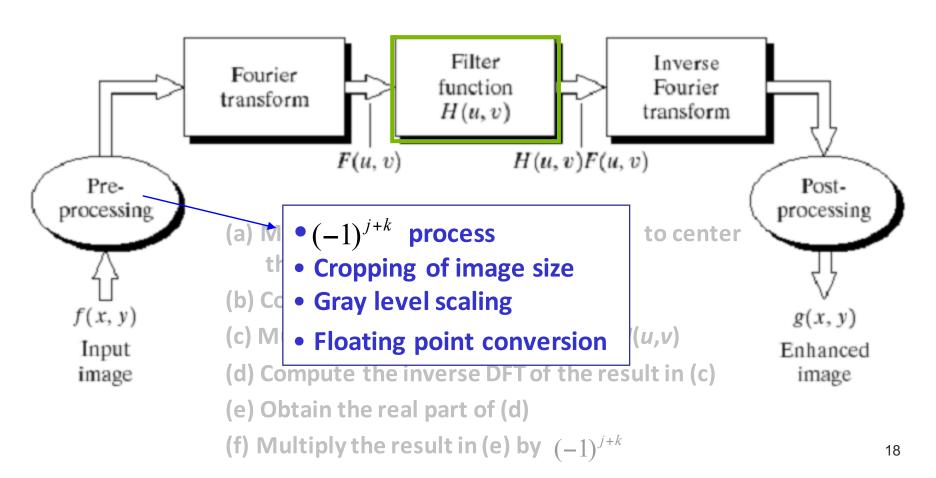
If
$$f(j,k)$$
 is real,
$$F(u,v) = F^*(-u,-v)$$

$$|F(u,v)| = |F(-u,-v)|$$

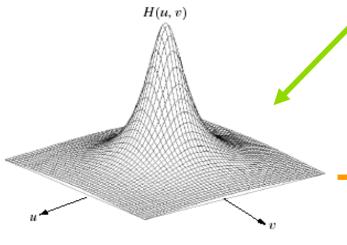
Filtering in the frequency domain

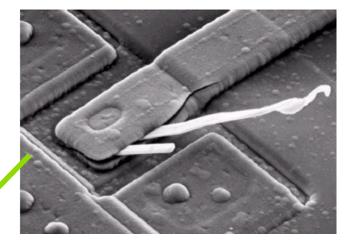


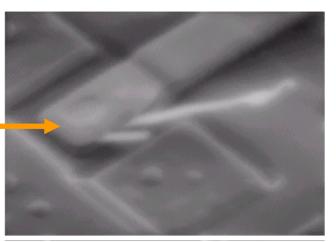
Filtering in the frequency domain



- Basic filters and their properties
 - Low-pass filter
 - High-pass filter
 - Notch filter

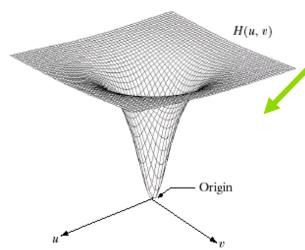




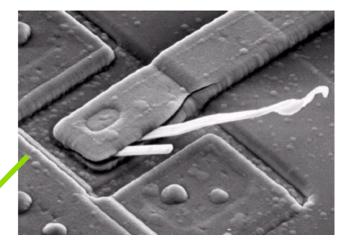


- Attenuates high frequencies
- Passes low frequencies

- Basic filters and their properties
 - Low-pass filter
 - High-pass filter
 - Notch filter

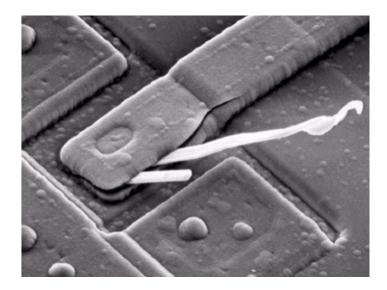


- Attenuates low frequencies
- Passes high frequencies

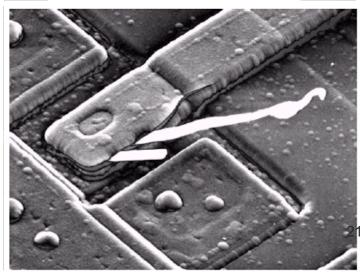




- Basic filters and their properties
 - Low-pass filter
 - Modified high-pass filter
 - Adding a constant of one-half the filter height to the filter function to avoid complete elimination
 - Notch filter



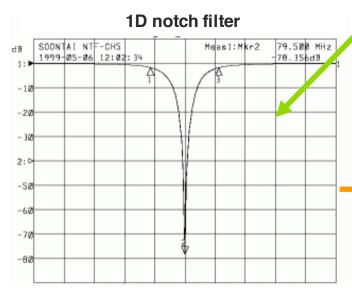
Modified high-pass filter



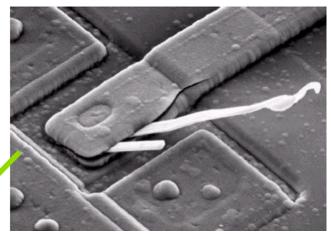
- Basic filters and their properties
 - Notch filter

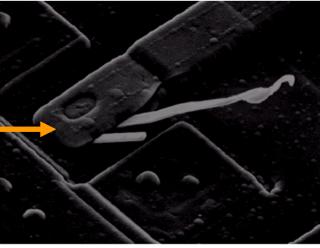
$$H(u,v) = \begin{cases} 0 & if \quad (u,v) = (\alpha,\beta) \\ 1 & otherwise \end{cases}$$

→ A constant function with a hole at the origin



http://www.soontai.com/da_filterprofile.html





- Convolution Theorem
 - Discrete convolution

$$f(j,k) * h(j,k) = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f(m,n)h(j-m,k-n)$$

- Flip → shift → sum of products
- Fourier transform pairs

$$f(j,k)*h(j,k) \Leftrightarrow F(u,v)H(u,v)$$

$$f(j,k)\underline{h(j,k)} \Leftrightarrow F(u,v)*\underline{H(u,v)}$$
Impulse Impulse response

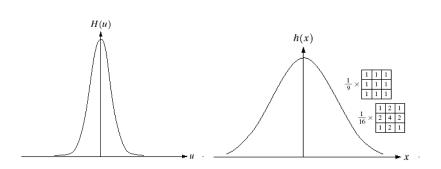
Convolution Theorem

Gaussian filter

One-dimensional

$$H(u) = Ae^{-u^2/2\sigma^2}$$

$$h(k) = \sqrt{2\pi}\sigma Ae^{-2\pi^2\sigma^2k^2}$$



Important characteristics

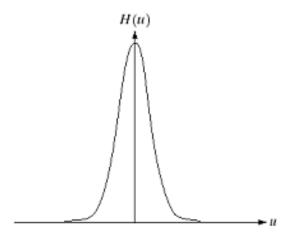
- Easy to specify and manipulate
- Fourier transform of a Gaussian function is a 'real'
 Gaussian function
- Two functions behave reciprocally w.r.t one another

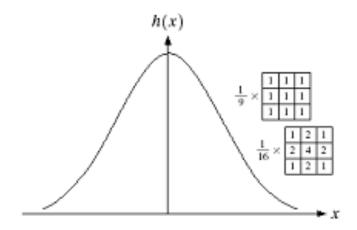
24

- Gaussian filter
 - Low-pass filter

$$H(u) = Ae^{-u^2/2\sigma^2}$$

$$h(k) = \sqrt{2\pi} \sigma A e^{-2\pi^2 \sigma^2 k^2}$$



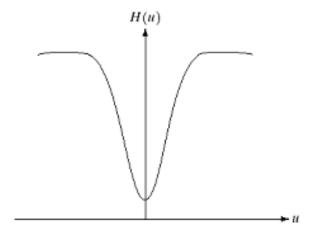


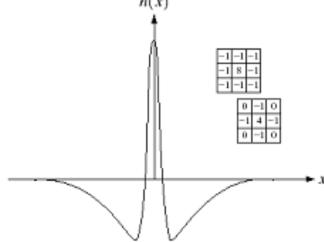
- → All the values are positive in both domains
- → The narrower the frequency domain filter, the wider the spatial domain filter, i.e. more severe blurring effect

- Gaussian filter
 - High-pass filter
 - Construct a high-pass filter as a difference of Gaussians

$$H(u) = Ae^{-u^2/2\sigma_1^2} - Be^{-u^2/2\sigma_2^2}$$

$$h(k) = \sqrt{2\pi} \left(\sigma_1 A e^{-2\pi^2 \sigma_1^2 k^2} - \sigma_2 B e^{-2\pi^2 \sigma_2^2 k^2} \right)$$





Comparison

- Comparison of spatial-domain and frequency-domain filtering
 - Filtering in spatial domain
 - Specific masks are needed
 - Filtering in frequency domain
 - $f(j,k)*h(j,k) \Leftrightarrow F(u,v)H(u,v)$ $f(j,k)h(j,k) \Leftrightarrow F(u,v)*H(u,v)$
 - Easy to implement
 - Fast Fourier Transform (FFT)
 - Save computation complexity for larger signal size

Smoothing frequency-domain filters

$$G(u, v) = H(u, v)F(u, v)$$

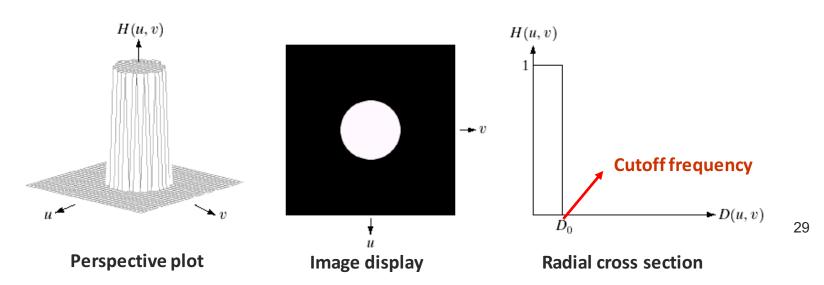
- Ideal low-pass filters (ILPF)
- Butterworth low-pass filters (BLPF)
- Gaussian low-pass filters (GLPF)

Ideal low-pass filters

$$H(u,v) = \begin{cases} 1 & \text{if } D(u,v) \leq D_0 \end{cases} \text{ Non-negative } \\ 0 & \text{if } D(u,v) > D_0 \end{cases}$$

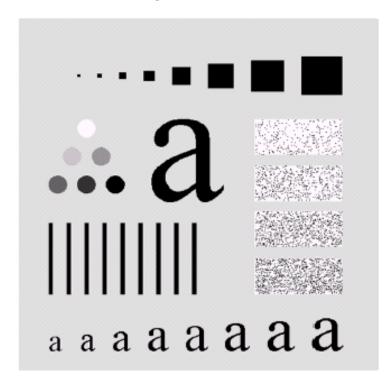
$$D(u,v) = [(u - M/2)^2 + (v - N/2)^2]^{1/2}$$

< Distance from point (u,v) to the center of the frequency rectangle >

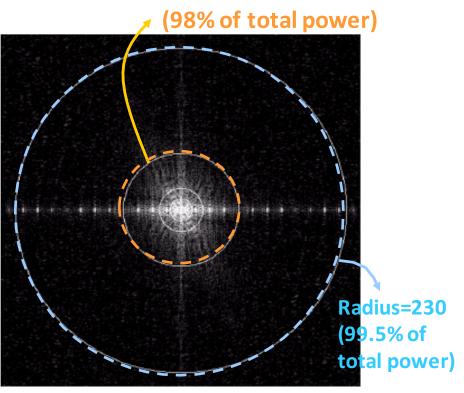


Ideal low-pass filters

Example



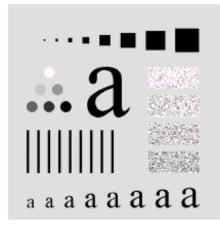
Original image (in spatial domain)

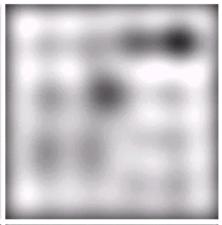


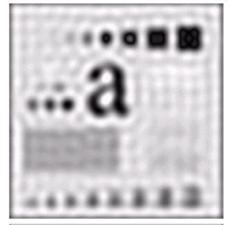
Radius=80

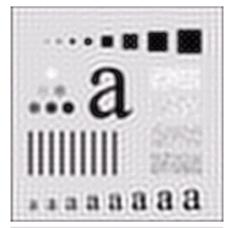
Radius = 5, 15, 30, 80 & 230 (in frequency domain)

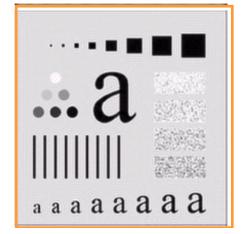
- Ideal low-pass filters
 - Example (with radius = 5, 15, 30, 80 & 230)

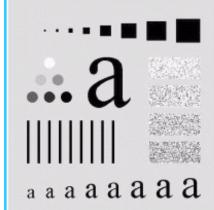




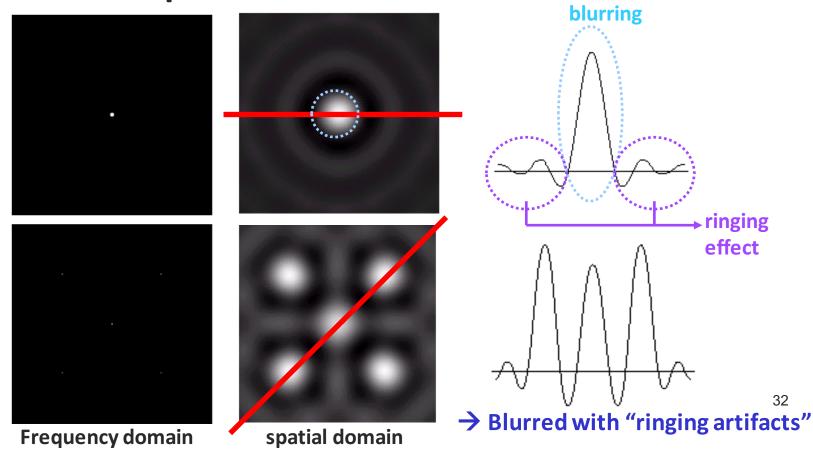




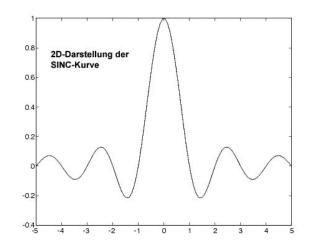


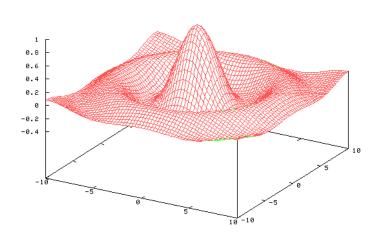


- Ideal low-pass filters
 - Example



- Ideal low-pass filters
 - Ideal LPF presents a Sinc function in the spatial domain
 - Radius of the main lobe is inversely proportional to the cutoff frequency



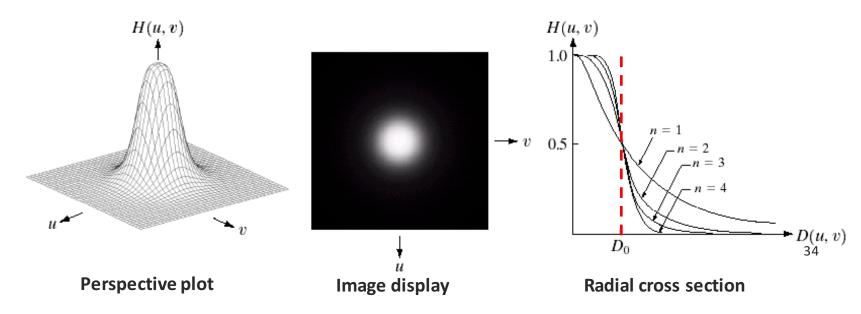


Butterworth low-pass filters

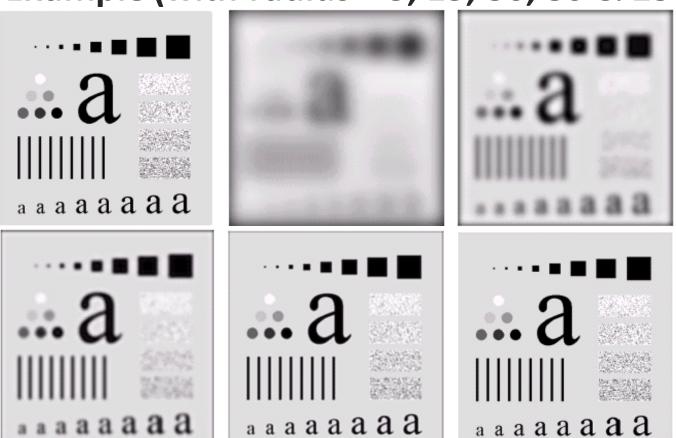
$$H(u,v) = \frac{1}{1 + [D(u,v)/D_0]^{2n}}$$
 n: order (must be an integer)

$$D(u,v) = [(u - M/2)^2 + (v - N/2)^2]^{1/2}$$

< Distance from point (u,v) to the center of the frequency rectangle >



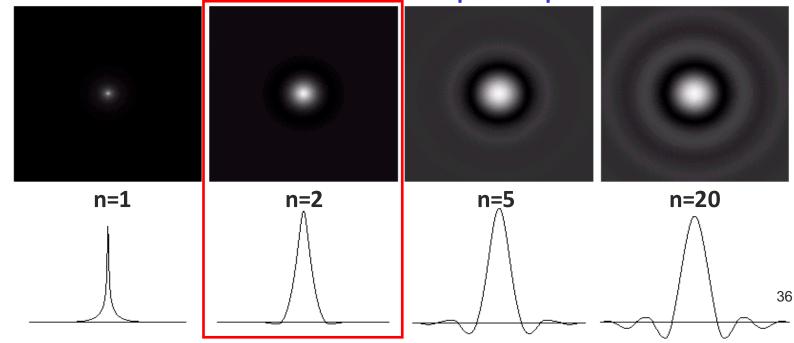
- Butterworth low-pass filters (n=2)
 - Example (with radius = 5, 15, 30, 80 & 230)



- Butterworth low-pass filters
 - Order n = 1, 2, 5, and 20 with cutoff=5 pixels

$$H(u,v) = \frac{1}{1 + [D(u,v)/D_0]^{2n}}$$

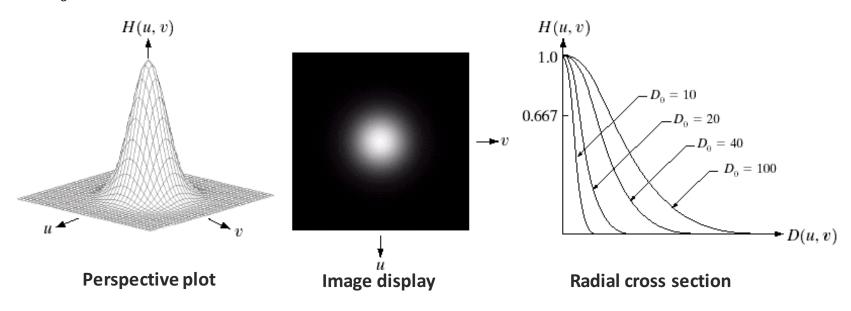
Spatial representation of BLPF



Gaussian low-pass filters

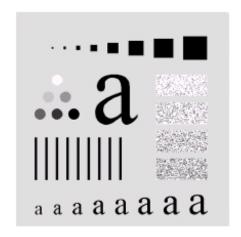
$$H(u,v) = e^{-D^2(u,v)/2\sigma^2}$$
 or $e^{-D^2(u,v)/2D_0^2}$

 D_0 is the cutoff frequency

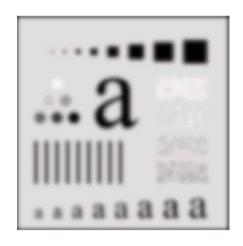


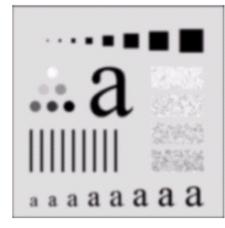
- → Gaussian in frequency domain → Gaussian in spatial domain
- → No ringing artifacts

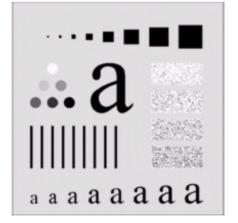
Gaussian low-pass filters

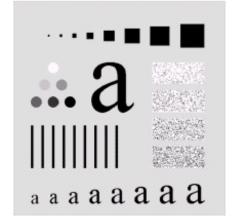












[&]quot;No ringing artifacts"

- Examples
 - Gaussian low-pass filters
 - Connect broken characters

Historically, certain computer programs were written using only two digits rather than four to define the applicable year. Accordingly, the company's software may recognize a date using "00" as 1900 rather than the year 2000.

Historically, certain computer programs were written using only two digits rather than four to define the applicable year. Accordingly, the company's software may recognize a date using "00" as 1900 rather than the year 2000.



- Examples
 - Gaussian low-pass filters
 - Reduction in skin fine lines







Cutoff=100



Cutoff=80

Sharpening frequency-domain filters

$$H_{hp}(u,v) = 1 - H_{lp}(u,v)$$

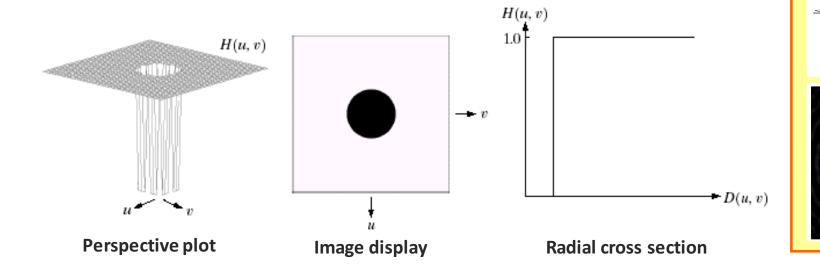
- Ideal high-pass filters
- Butterworth high-pass filters
- Gaussian high-pass filters
- The Laplacian in the frequency domain

Ideal high-pass filters

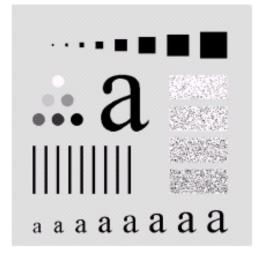
$$H(u,v) = \begin{cases} 0 & \text{if } D(u,v) \le D_0 \\ 1 & \text{if } D(u,v) > D_0 \end{cases}$$

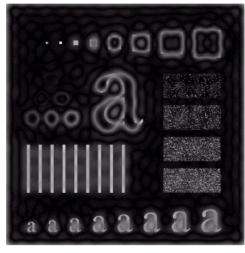
$$D(u,v) = [(u - M/2)^2 + (v - N/2)^2]^{1/2}$$

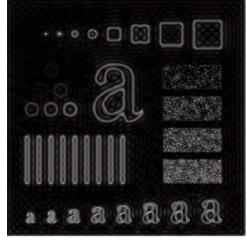
Spatial domain

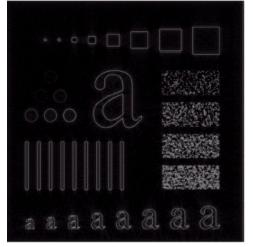


- Ideal high-pass filters
 - Example (with radius = 15, 30 & 80)









Original

Cutoff=15

Cutoff=30

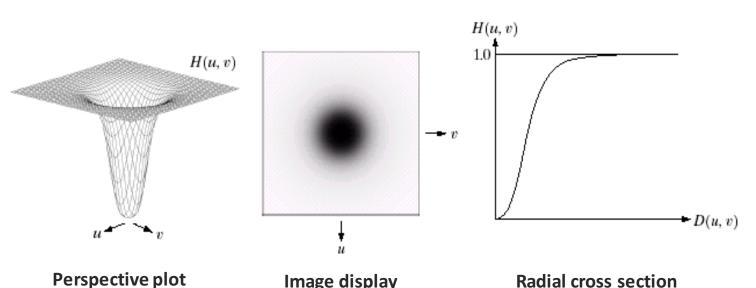
Cutoff=80

Butterworth high-pass filters

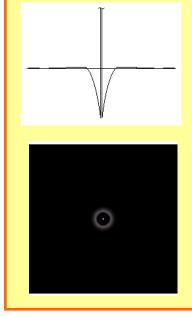
$$H(u,v) = \frac{1}{1 + [D_0/D(u,v)]^{2n}}$$

n: order (must be an integer)

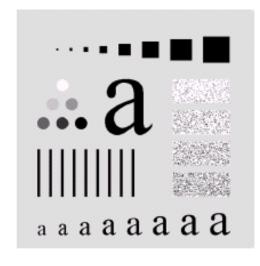
$$D(u,v) = [(u - M/2)^2 + (v - N/2)^2]^{1/2}$$

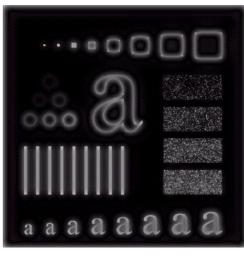


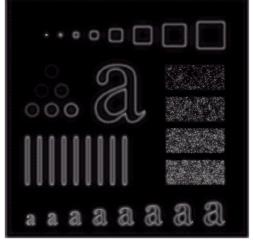
Spatial domain

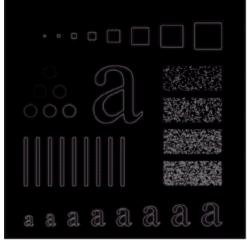


- Butterworth high-pass filters (n=2)
 - Example (with radius = 15, 30 & 80)









Original

Cutoff=15

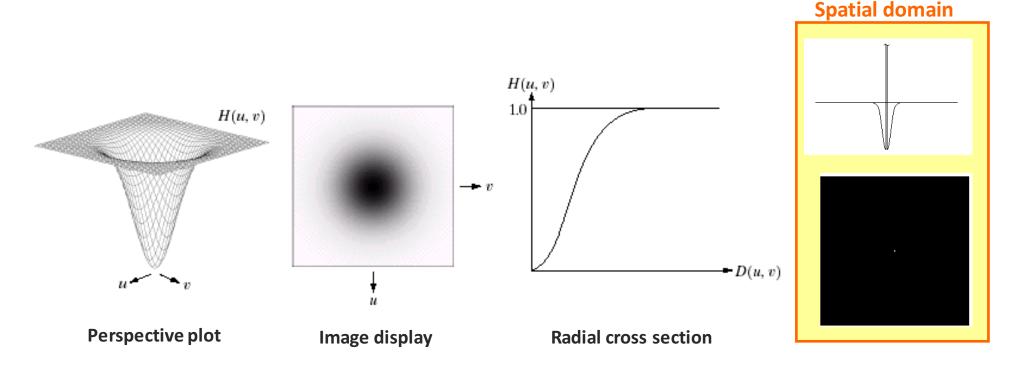
Cutoff=30

Cutoff=80

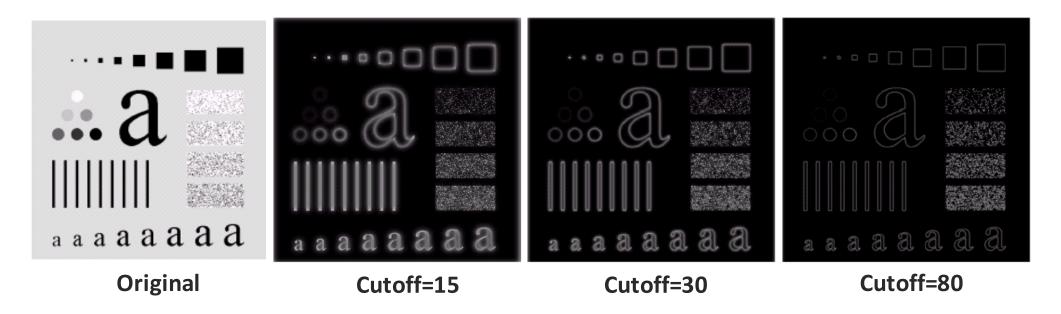
Gaussian high-pass filters

$$H(u, v) = 1 - e^{-D^2(u, v)/2D_0^2}$$

 D_0 is the cutoff frequency



- Gaussian high-pass filters
 - Example (with radius = 15, 30 & 80)



//Note// High-pass filters can be constructed by the difference of Gaussian low-pass filters

→ more parameters → more control over the filter shape

The Laplacian in the Frequency domain

$$\Im\left[\frac{d^n f(x)}{dx^n}\right] = (ju)^n F(u)$$

$$\Im\left[\frac{\partial^2 f(x,y)}{\partial x^2} + \frac{\partial^2 f(x,y)}{\partial y^2}\right] = (ju)^2 F(u,v) + (jv)^2 F(u,v)$$
$$= -(u^2 + v^2) F(u,v)$$

$$\Im\left[\nabla^2 f(x,y)\right] = -(u^2 + v^2)F(u,v)$$

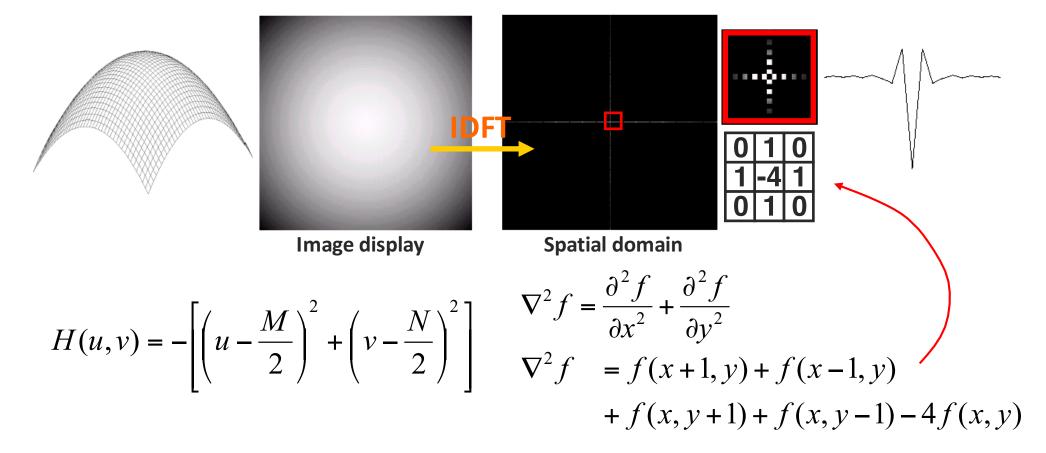
$$G(u,v) = H(u,v)F(u,v)$$

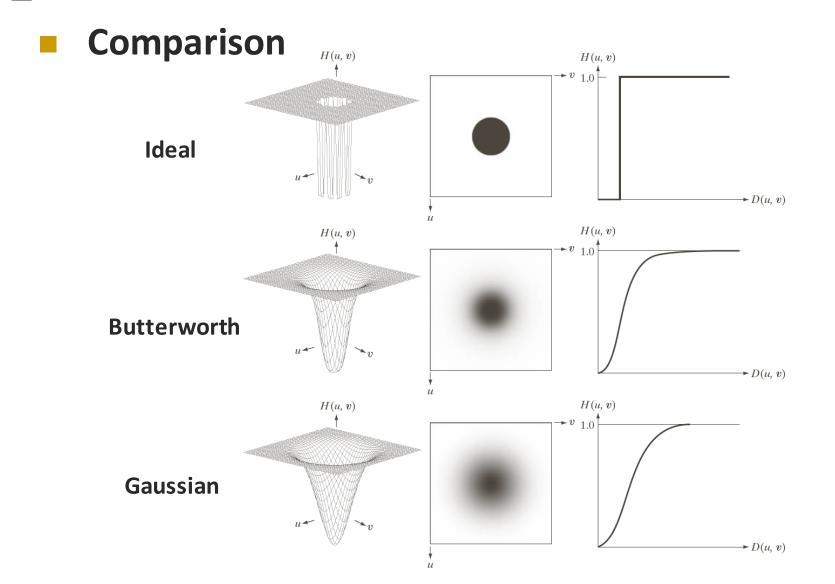
$$\Rightarrow H(u,v) = -(u^2 + v^2)$$

$$\Rightarrow H(u,v) = -\left[\left(u - \frac{M}{2}\right)^2 + \left(v - \frac{N}{2}\right)^2\right]$$

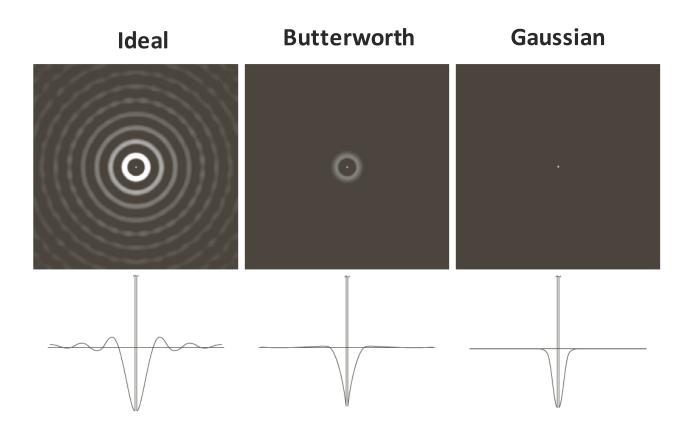
The Laplacian in the Frequency domain

$$\Im \left[\nabla^2 f(x,y) \right] = -((u - M/2)^2 + (v - N/2)^2)F(u,v)$$

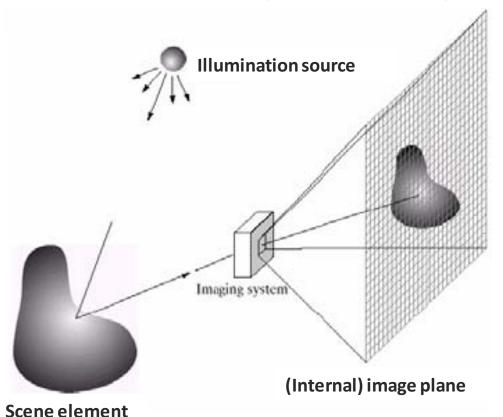




Comparison – spatial representation



- Homomorphic filtering
 - Recall: (Lecture #2)



$$f(x,y) = i(x,y)r(x,y)$$

- Illumination

$$0 < i(x, y) < \infty$$

- Reflectance

Homomorphic filtering

$$f(x, y) = i(x, y)r(x, y)$$

 The Fourier transform of the product of two functions is NOT separable

$$\Im\{f(x,y)\} \neq \Im[\{i(x,y)]\Im[\{r(x,y)\}\}$$

Define

$$z(x, y) = \ln f(x, y) = \ln i(x, y) + \ln r(x, y)$$

$$\Rightarrow \Im\{z(x, y)\} = \Im\{\ln f(x, y)\}$$

$$= \Im\{\ln i(x, y)\} + \Im\{\ln r(x, y)\}$$

or
$$Z(u, v) = F_i(u, v) + F_r(u, v)$$

Homomorphic filtering

$$S(u,v) = H(u,v)Z(u,v)$$

$$= H(u,v)F_{i}(u,v) + H(u,v)F_{r}(u,v)$$

$$\Rightarrow s(x,y) = \Im \{S(u,v)\}$$

$$= \Im \{H(u,v)F_{i}(u,v)\} + \Im \{H(u,v)F_{r}(u,v)\}$$

let

$$i'(x,y) = \Im^{-1} \{ H(u,v) F_i(u,v) \}$$

$$r'(x,y) = \Im^{-1} \{ H(u,v) F_r(u,v) \}$$

$$\Rightarrow s(x,y) = i'(x,y) + r'(x,y)$$

Homomorphic filtering

$$s(x,y) = i'(x,y) + r'(x,y)$$

$$g(x,y) = e^{s(x,y)}$$

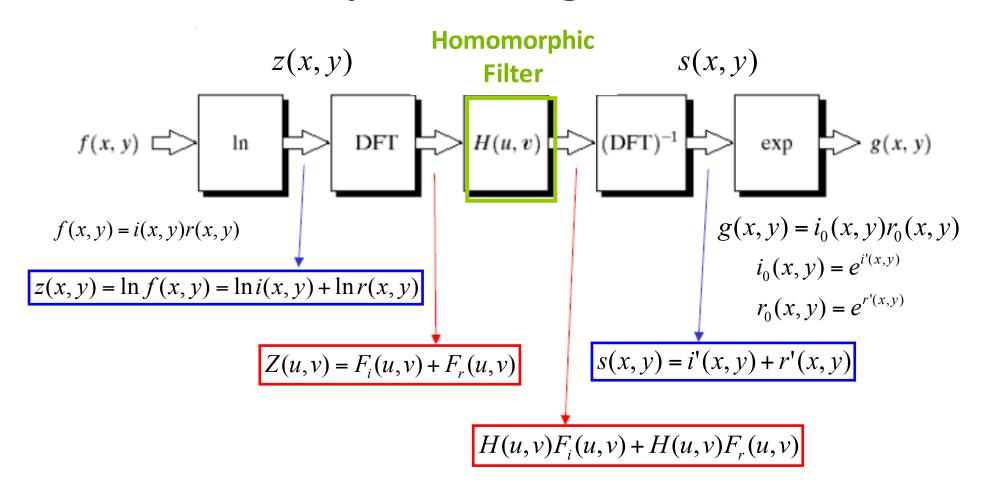
$$= e^{i'(x,y)} \cdot e^{r'(x,y)} = i_0(x,y)r_0(x,y)$$

where

$$i_0(x, y) = e^{i'(x, y)}$$
 and $r_0(x, y) = e^{r'(x, y)}$

are the illumination and reflectance components of the output image

Homomorphic filtering



Two components

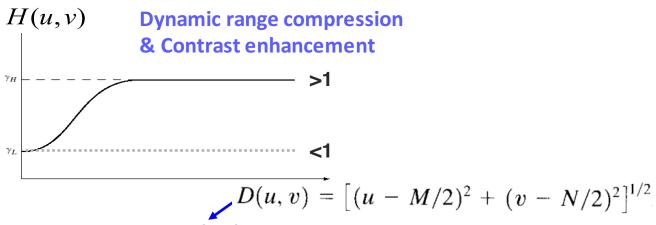
- Illumination component
 - Vary slowly (slow spatial variations)
 - Associate low frequencies
- Reflectance component
 - Vary abruptly
 - Associate high frequencies

Homomorphic filter

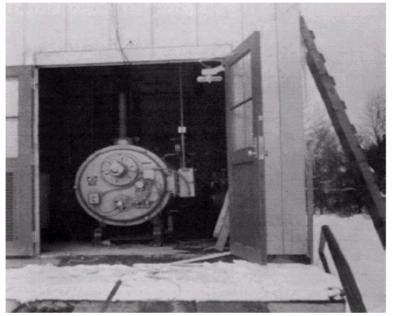
- Separation of illumination & reflectance components
- Affect the low- and high-frequency components of the Fourier transform in different ways

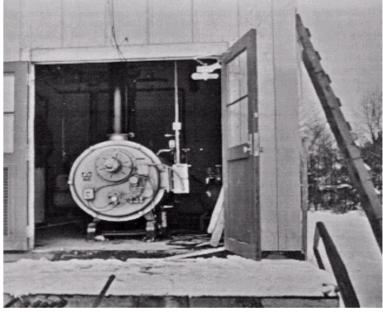
Example

"circularly" symmetric



The distance from point (u,v) to the origin of the centered transform





- Another example of homomorphic filtering
 - PET image

