EE3015 & EE3025 Presentation

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October 3, 2021

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Given
$$x(n) = \{1, 2, 3, 4, 2, 1\}$$
 (1)

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2)$$
 (2)

$$\implies h(n) = \left(\frac{-1}{2}\right)^n u(n) + \left(\frac{-1}{2}\right)^{n-2} u(n-2) \tag{3}$$

Question

Compute X(k), H(k) and y(n) using FFT and IFFT methods.

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Solution

Computing y (for N samples) using FFT and IFFT:

$$X = FFT(x) \tag{4}$$

$$H = FFT(h) \tag{5}$$

$$Y = X.H \tag{6}$$

$$y = IFFT(Y) = \frac{1}{N} * FFT(Y^*) \{ \text{only if y is real} \}$$
 (7)

where Y^* = complex conjugate of Y



Recursive N-point FFT Algorithm

An N-point DFT can be written as:

$$X_{k} = \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$

$$= \sum_{m=0}^{N/2-1} x_{2m}e^{-\frac{j2\pi km}{N/2}} + W_{N}^{k} \sum_{m=0}^{N/2-1} x_{2m+1}e^{-\frac{j2\pi k(m)}{N/2}}$$
N/2 DFT with even inputs

where
$$W_N=e^{rac{-j2\pi}{N}}$$

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Recursive N-point FFT Algorithm

While exploiting symmetry of W_N as:

$$W_N^{k+N/2} = -W_N^k \tag{8}$$

We transformed the iterative DFT problem to a Divide-Conquer algorithm as :

$$X_{0 \to \frac{N}{2} - 1} = X_{even} + \overline{W}_{N/2} * X_{odd}$$
(9)

$$X_{\frac{N}{2} \to N-1} = X_{even} - \overline{W}_{N/2} * X_{odd}$$
 (10)

$$\overline{W}_{N/2}(i) = W_N^i$$

; for $i = 0,1,2 \dots (N/2) -1$

Matrix representation of the FFT algorithm

An 8-point DFT can be represented as a Matrix product as follows:

$$\overline{X} = \overline{W} \, \overline{x}$$

$$\begin{bmatrix} X(0) \\ X(1) \\ X(2) \\ X(3) \\ X(4) \end{bmatrix} = \begin{bmatrix} W^0 & W^0 & W^0 & W^0 & W^0 & W^0 & W^0 \\ W^0 & W^1 & W^2 & W^3 & W^4 & W^5 & W^6 & W^7 \\ W^0 & W^2 & W^4 & W^6 & W^0 & W^2 & W^4 & W^6 \\ W^0 & W^3 & W^6 & W^1 & W^4 & W^7 & W^2 & W^5 \\ W^0 & W^4 & W^0 & W^4 & W^0 & W^4 & W^0 & W^4 \\ X(5) \\ X(6) \\ X(7) \end{bmatrix} = \begin{bmatrix} W^0 & W^5 & W^2 & W^7 & W^4 & W^1 & W^6 & W^3 \\ W^0 & W^6 & W^4 & W^2 & W^0 & W^6 & W^4 & W^2 \\ W^0 & W^7 & W^6 & W^5 & W^4 & W^3 & W^2 & W^1 \end{bmatrix} \begin{bmatrix} x(0) \\ x(1) \\ x(2) \\ x(3) \\ x(4) \\ x(5) \\ x(6) \\ x(7) \end{bmatrix}$$

$$(11)$$

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The FFT algorithm decompose \overline{W} into sparse matrices by permuting \overline{x} in a bit-reversed fashion resulting in $\overline{x_p}$

$$\overline{X} = \overline{W3} \ \overline{W2} \ \overline{W1} \ \overline{x_p} \tag{12}$$

$$\begin{bmatrix} \chi(0) \\ \chi(1) \\ \chi(2) \\ \chi(3) \\ \chi(4) \\ \chi(5) \\ \chi(6) \\ \chi(6) \\ \chi(7) \end{bmatrix} = \begin{bmatrix} 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . & . \\ . & 1 & . & . & . & . \\ . & 1 & . & . & .$$

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Assignment 1

Relation with the butterfly diagram

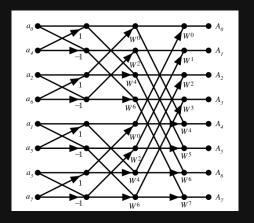


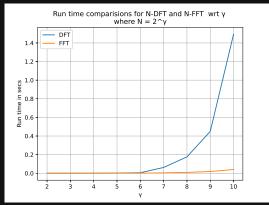
Figure: 8-point FFT butterfly diagram

Ref: Heckbert, P., 1995. Fourier transforms and the fast fourier transform (fft) algorithm. Computer Graphics, 2, pp.15-463.

Time complexity

Convolution/DFT requires N^2 operations $\approx O(N^2)$. While the same output can be achieved using FFT and IFFT within:

$$\underbrace{O(NlogN)}_{X \to X} + \underbrace{O(N)}_{Y = X * H} + \underbrace{O(NlogN)}_{Y \to y} \approx O(NlogN)$$
(14)



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C vs Python Implementaion

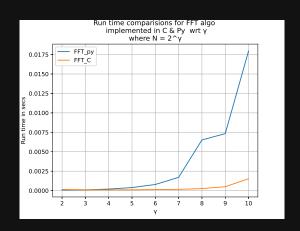


Figure: Comparing C vs Py implementation

Assignment 02

Design equivalent FIR realizations for filter number 114 with the following specifications.

Pass band specifications

$$F_s=48 kHz$$
 $F_{p1}=7.2 kHz$; $F_{p2}=6 kHz$ $\omega_{p1}=0.3\pi$; $\omega_{p2}=0.25\pi$ $\omega_{c}=0.275\pi$

$$\omega_l = \frac{\omega_{p1} - \omega_{p2}}{2} = 0.025\pi$$

Stop band specifications

Tolerance:
$$\delta_1=\delta_2=\delta=0.15$$
 $\Delta F=0.3kHz$; $\Delta\omega=0.0125\pi$
 $F_{s1}=7.5Hz$; $F_{s2}=5.7kHz$
 $\omega_{s1}=0.3125\pi$; $\omega_{s2}=0.2375\pi$

Designing a low pass equivalent

$$h_l(n) = \frac{\sin(n\omega_l)}{n\pi} w(n) \tag{15}$$

The Kaiser window is defined and computed as:

$$w(n) = \frac{I_0 \left[\beta N \sqrt{1 - \left(\frac{n}{N}\right)^2}\right]}{I_0(\beta N)}, -N \le n \le N, \beta > 0$$
 $= 0$ otherwise,

Parameters of Kaiser window are:

$$A = -20 \log_{10} \delta = 16.478 (< 21) \implies \beta N = 1$$
 $N \ge \frac{A-8}{4.57\Delta\omega} = 47.2 \implies chosen N = 100$

$$w(n) = 1, -100 \le n \le 100$$

= 0 otherwise

Desired lowpass filter impulse response is

$$h_{lp}(n) = \frac{\sin(\frac{n\pi}{40})}{n\pi} \; ; -100 \leq n \leq 100$$

= 0, otherwise

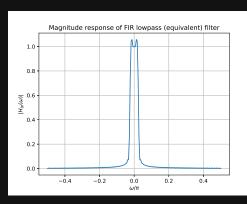


Figure: Magnitude response of the FIR lowpass digital filter

Converting into a causal FIR Bandpass Filter

A band pass filter can be obtained from a low pass equivalent through the following transformation ...

$$H_{bp}(j\omega) = H_{lp}(j(\omega - \omega_c)) + H_{lp}(j(\omega + \omega_c))$$
(16)

$$\implies h_{bp}(n) = h_{lp}(n) * 2cos(n\omega_c)$$
 (17)

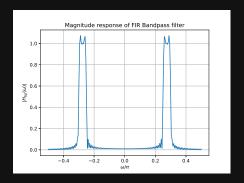


Figure: Magnitude response of the FIR bandpass digital filter

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References

Heckbert, P., 1995. Fourier transforms and the fast fourier transform (fft) algorithm. Computer Graphics, 2, pp.15-463

Code & Reports

- ► Assignment-01 Link
- Assignment-02 Link
- ► Beamer template Link

Thank You !!