

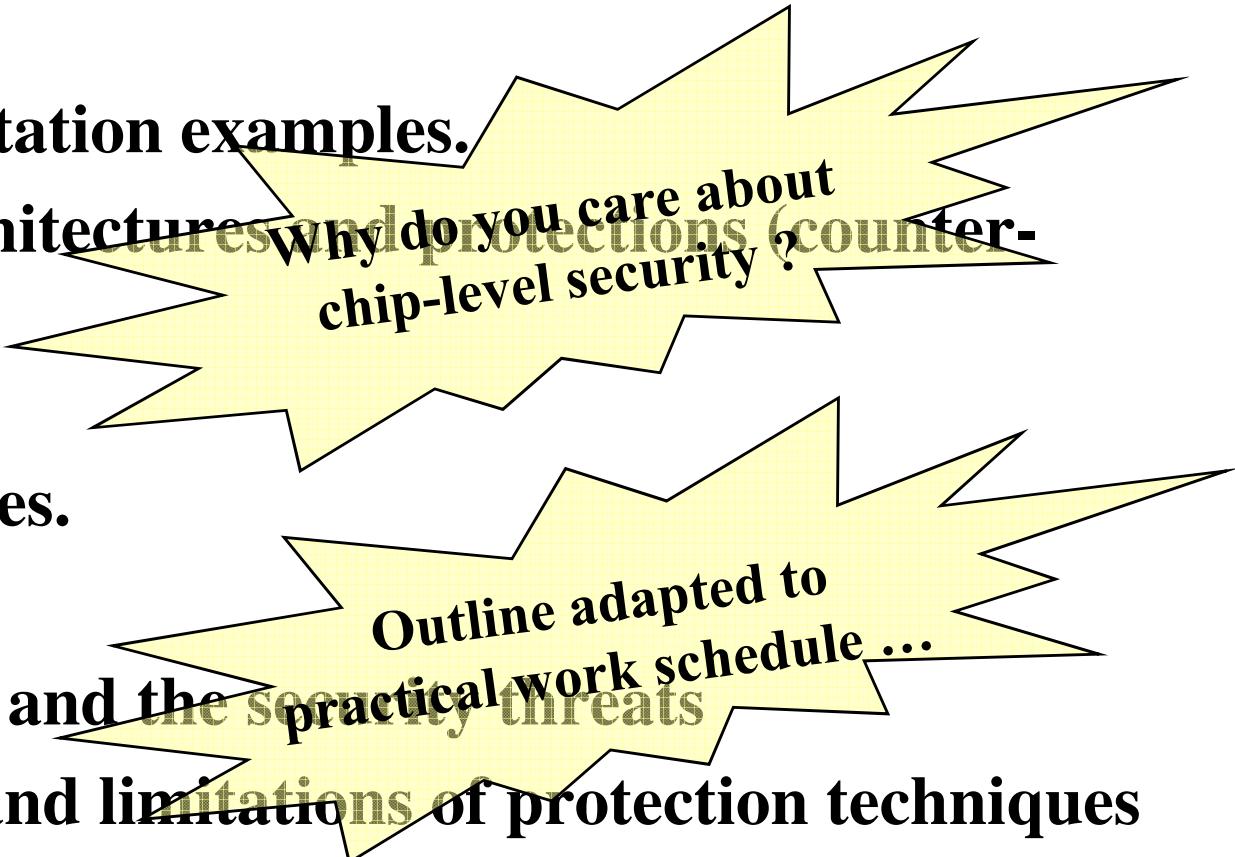
# Part III

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## Hardware-based attacks Types of attacks, exploitation examples

# Part II contents and objectives

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- Types of attacks, exploitation examples.
  - Examples of secure architectures (counter-measures).
  - Fault/error models.
  - Impact on test techniques.
  - Understand the context and the practical work schedule ...  
    ↳ Threats
  - Understand the basics and limitations of protection techniques
  - Prerequisites: basics in cryptography (M1 courses).
- 

# Types of attacks: global taxonomy

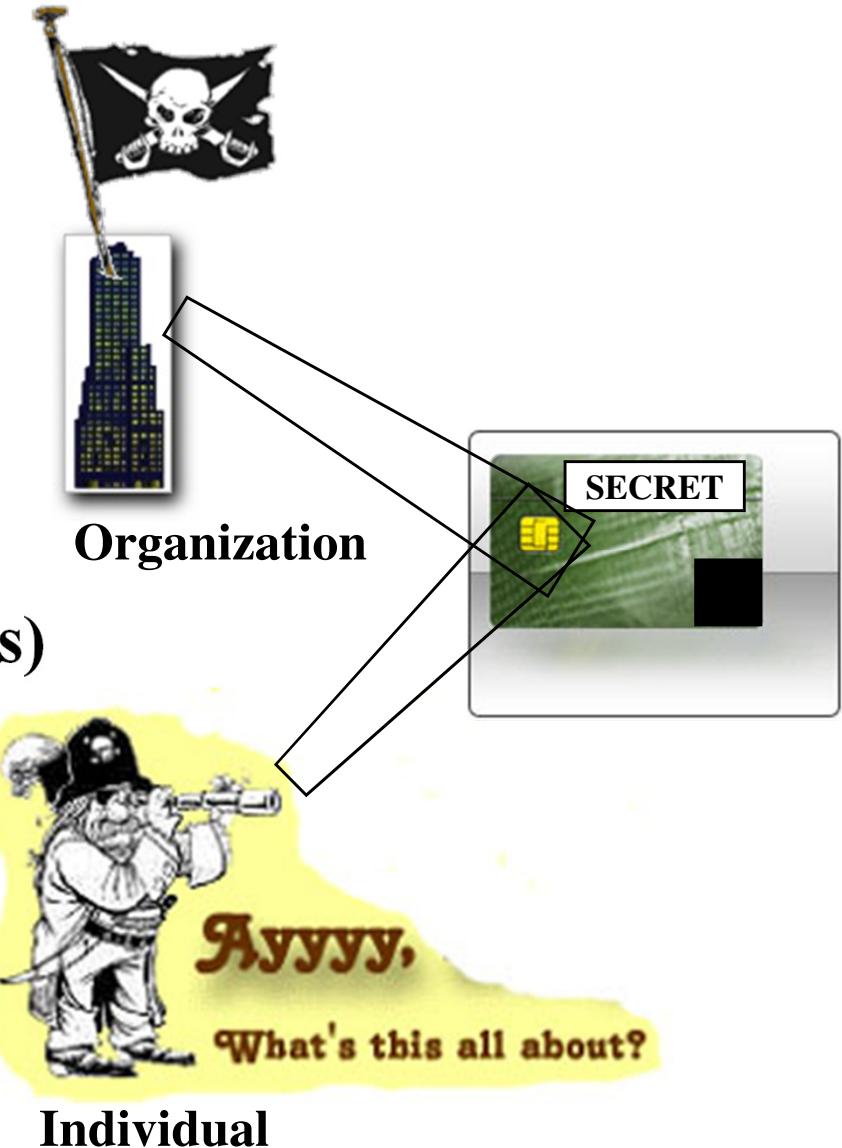
## □ Implementation based attacks

- ◆ Invasive
  - Circuit modifications
  - Probing
- ◆ Semi-invasive
- ◆ Non-invasive

## □ Observation-based (passive, side channels)

## □ Perturbation-based (active, fault based)

## □ Note: we will not discuss here attacks based on software&networks threats (e.g. viruses, malicious applets)



# Limitations of counter-measures

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- Never 100% protection/security – just make the attacks harder
- Still an empirical selection process
  - ◆ No global optimization of overheads
  - ◆ A protection against a given attack can ease another one (e.g. error detecting codes w.r.t. power-based attacks)
- All types of possible attacks must be simultaneously addressed (the hacker will use the easiest way to success)

# Implementation Attacks

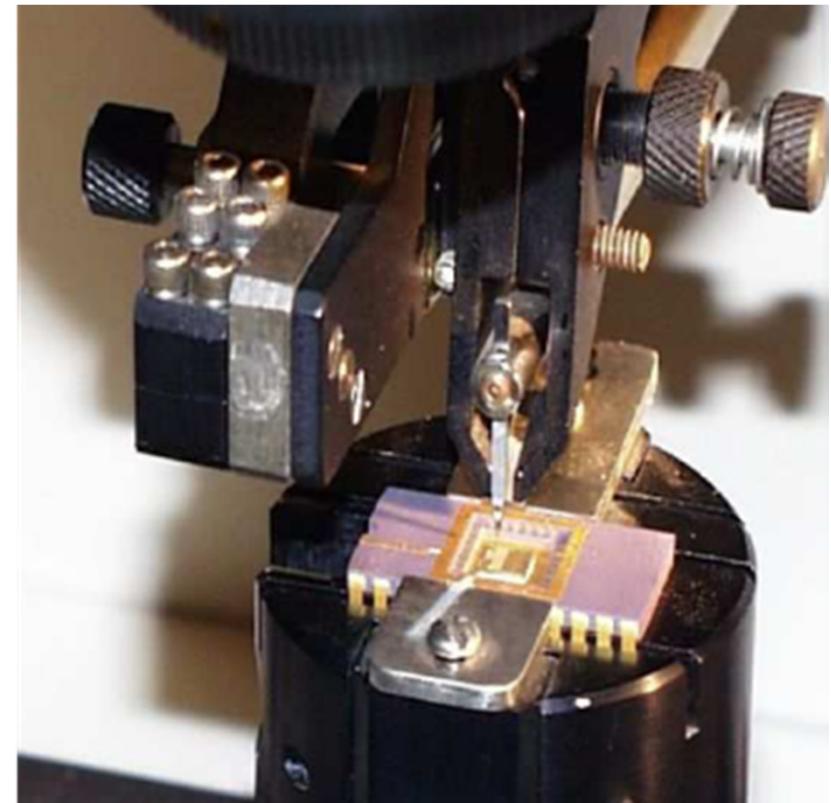
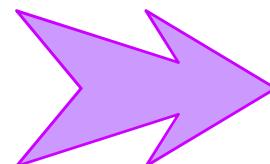
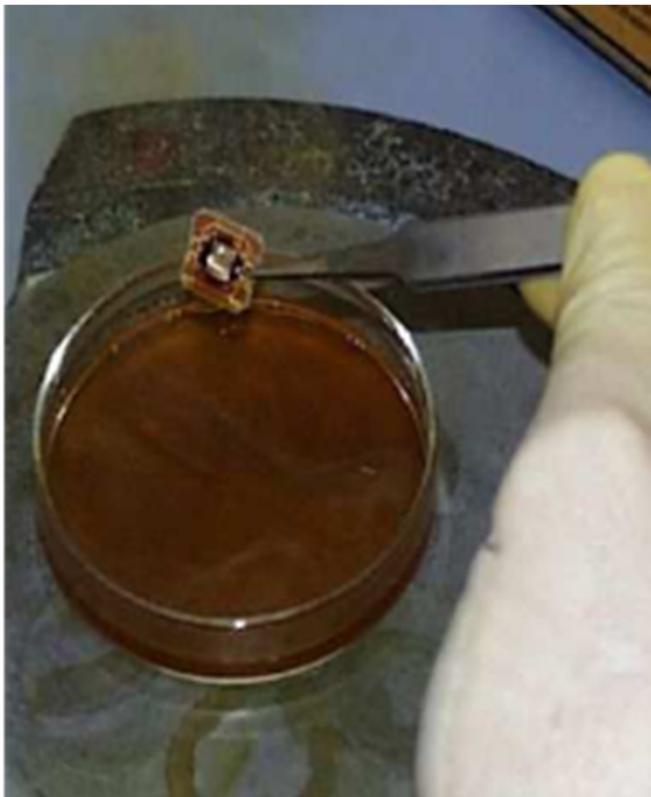
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- Invasive attacks on the circuit
  - Side Channel Analysis
    - ◆ Time Analysis
    - ◆ Power Analysis
    - ◆ EM analysis
    - ◆ ...
  - Test structures
    - ◆ Scan chains,
    - ◆ ...
  - Which countermeasures?
- 

## □ Fault attacks

- ◆ Voltage glitches,
- ◆ Clock glitches,
- ◆ Overvolting,
- ◆ Downvolting,
- ◆ Laser shots,
- ◆ Harmonic EM,
- ◆ Pulsed EM injection,
- ◆ ...

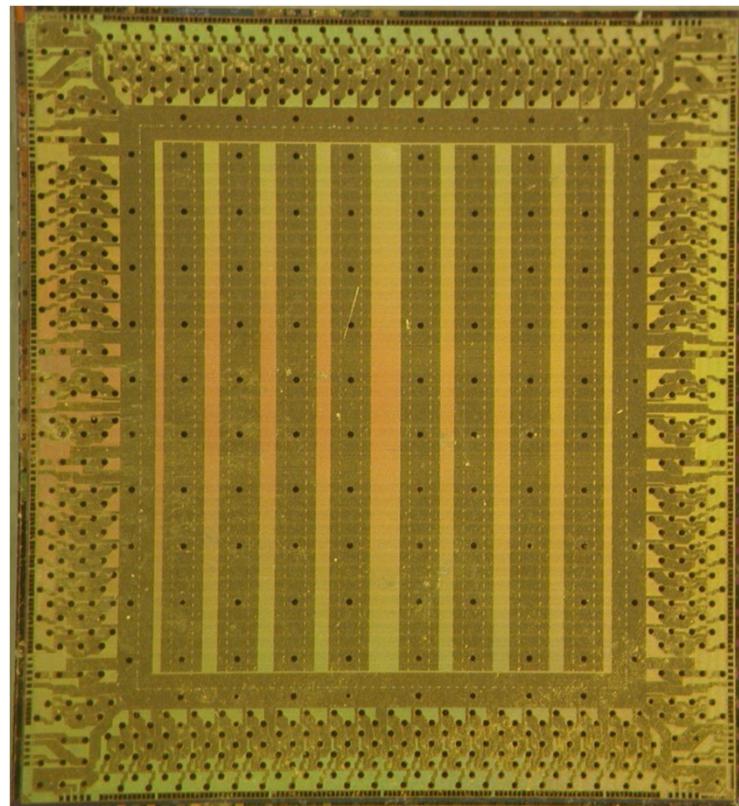
# Depackaging / repackaging



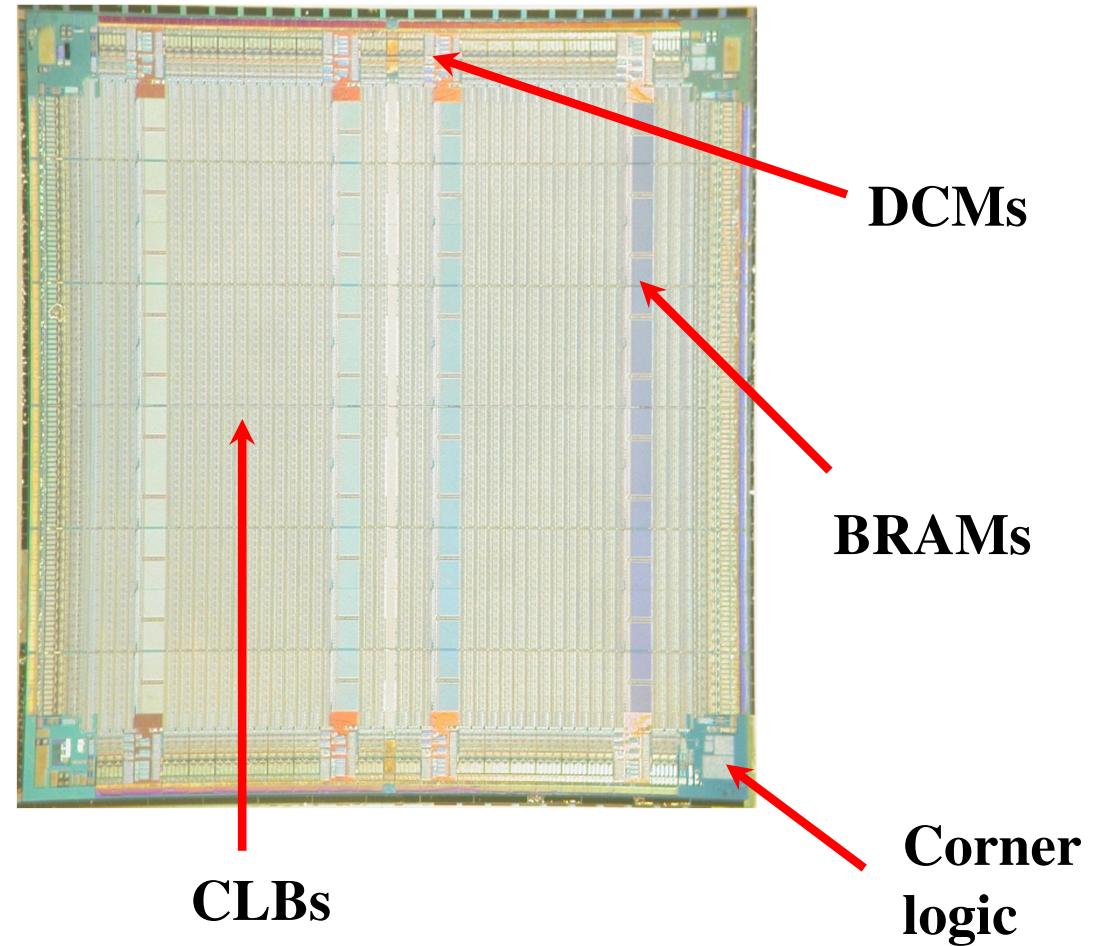
Sensors (light/UV) can be used to detect intrusion and stop functioning  
(or even destroy the critical contents)

# Layer removal: example on a XC2V1000

Before

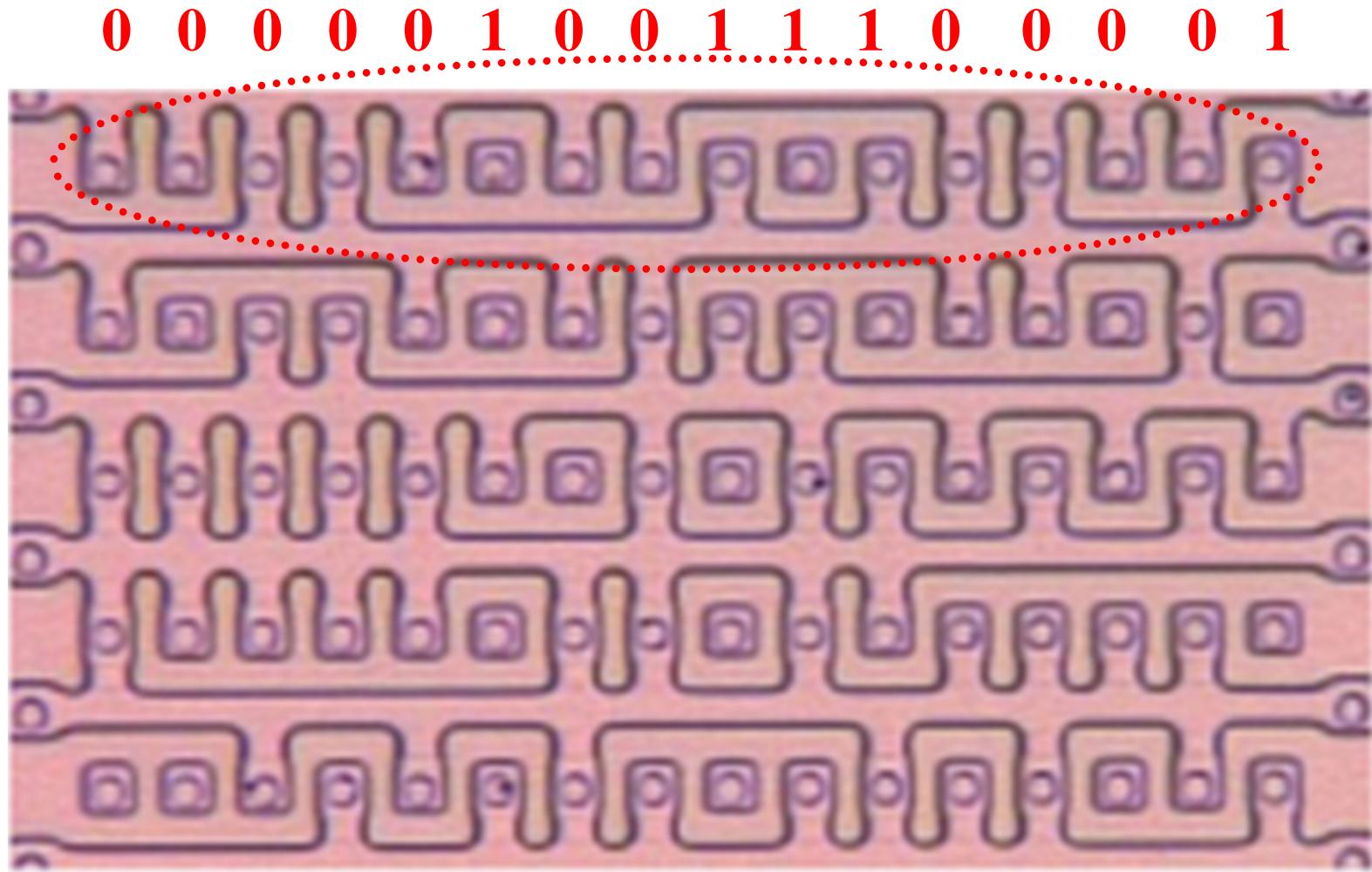


After - Poly-Silicon layer

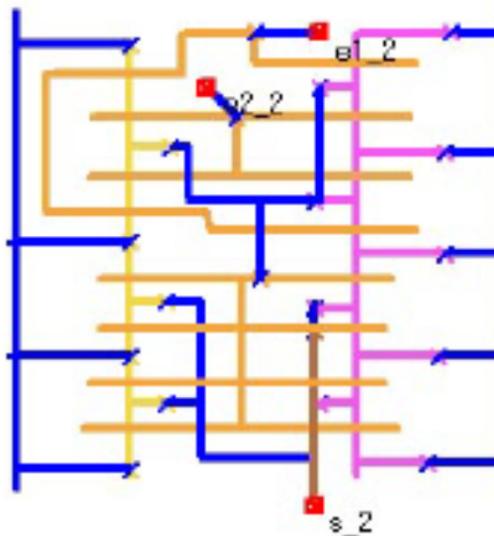
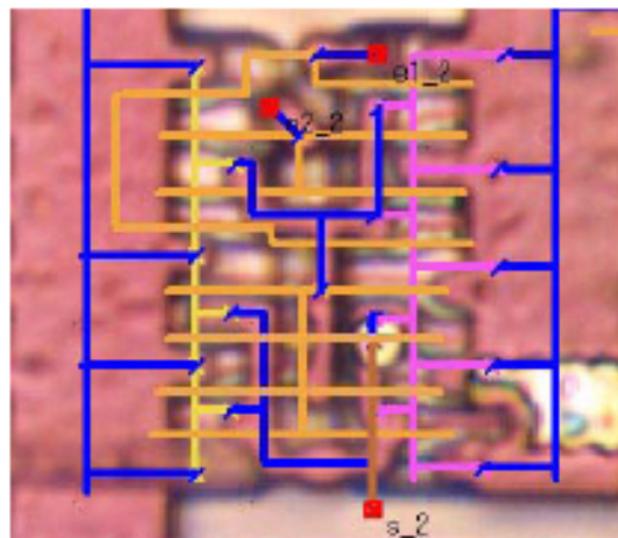
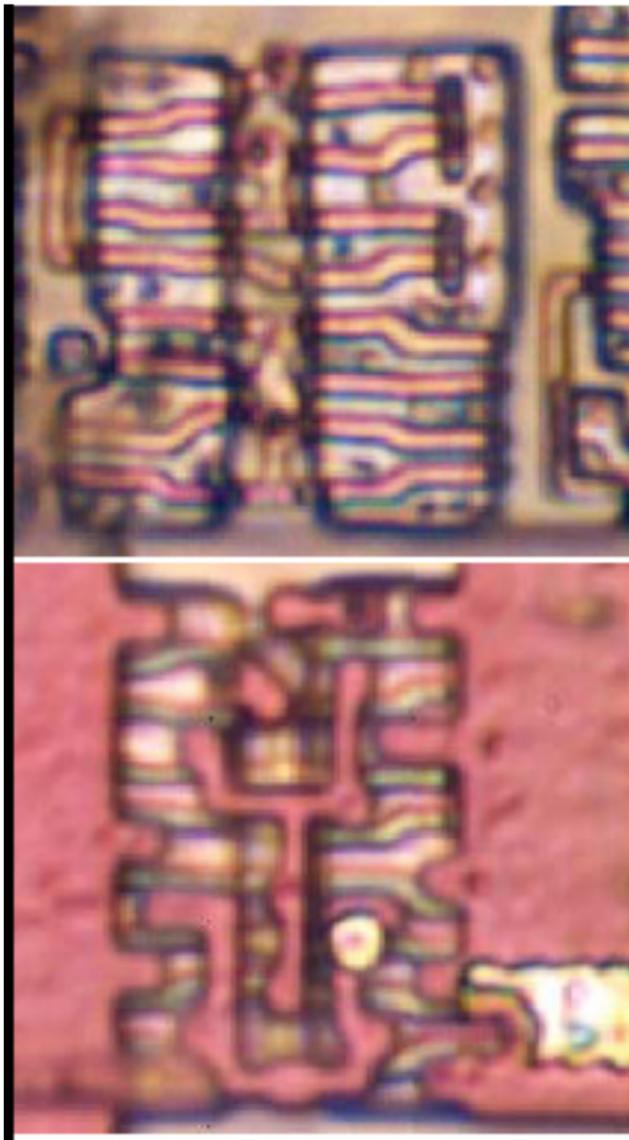


# Reading ROM memory contents ...

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# Reverse engineering



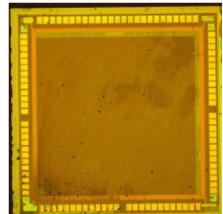
[Merle 2006]

A. Merle, "Security testing for hardware product: the security evaluations practice", Minatec CrossRoad, May 2006

# Layout observation and micro-probing

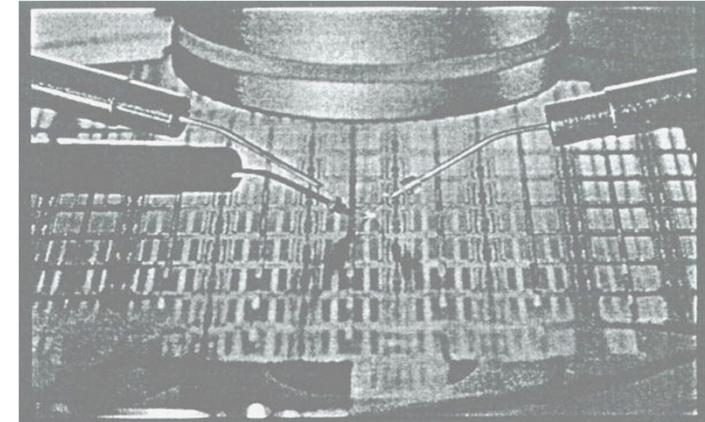
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- Successive layer removal (chemical) => destructive !
- Optical observation - more difficult in recent technologies  
(inherent shielding: dummies as covering layer ...)



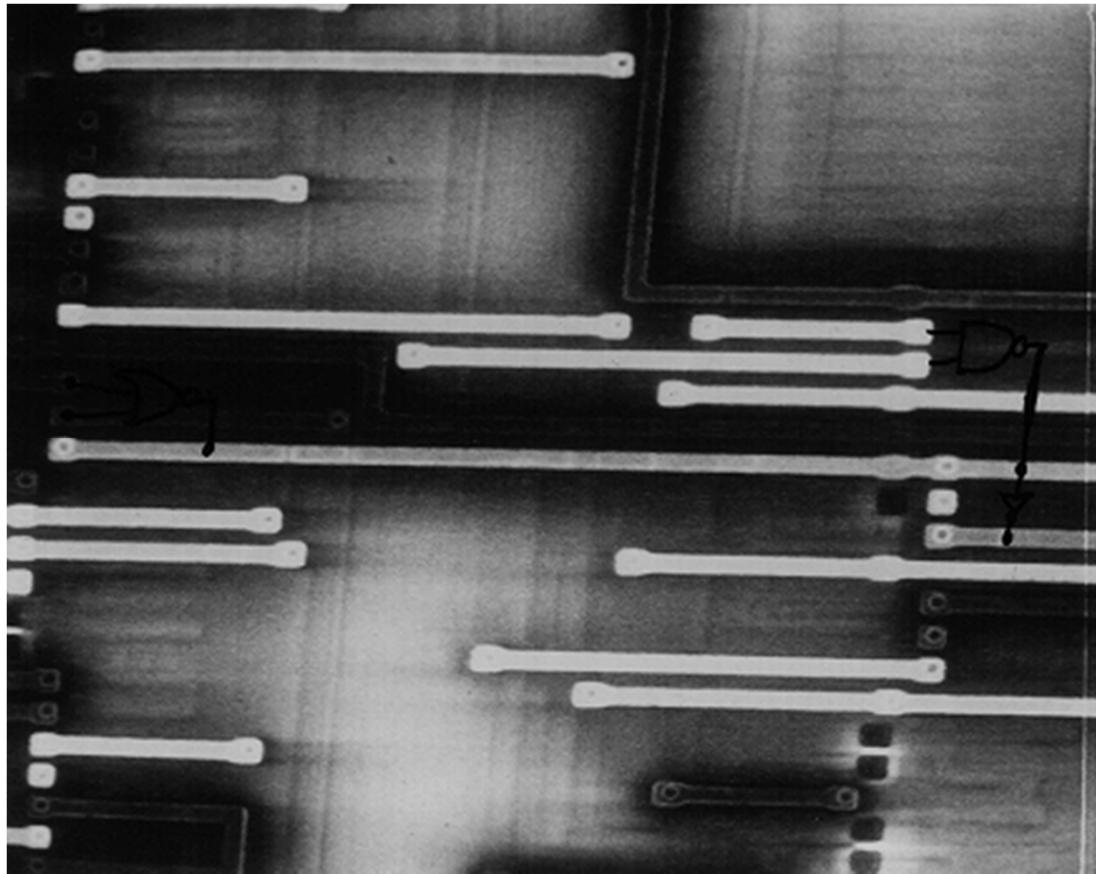
## □ Bus microprobing

- ◆ Bus must be localized ...
- ◆ Number of probes ...
- ◆ Physical access required (partial layer removal)
- ◆ Many data can be read on a single probing point ... but difficulty to identify what those data are (functional meaning)



# Voltage contrast microscopy (SEM)

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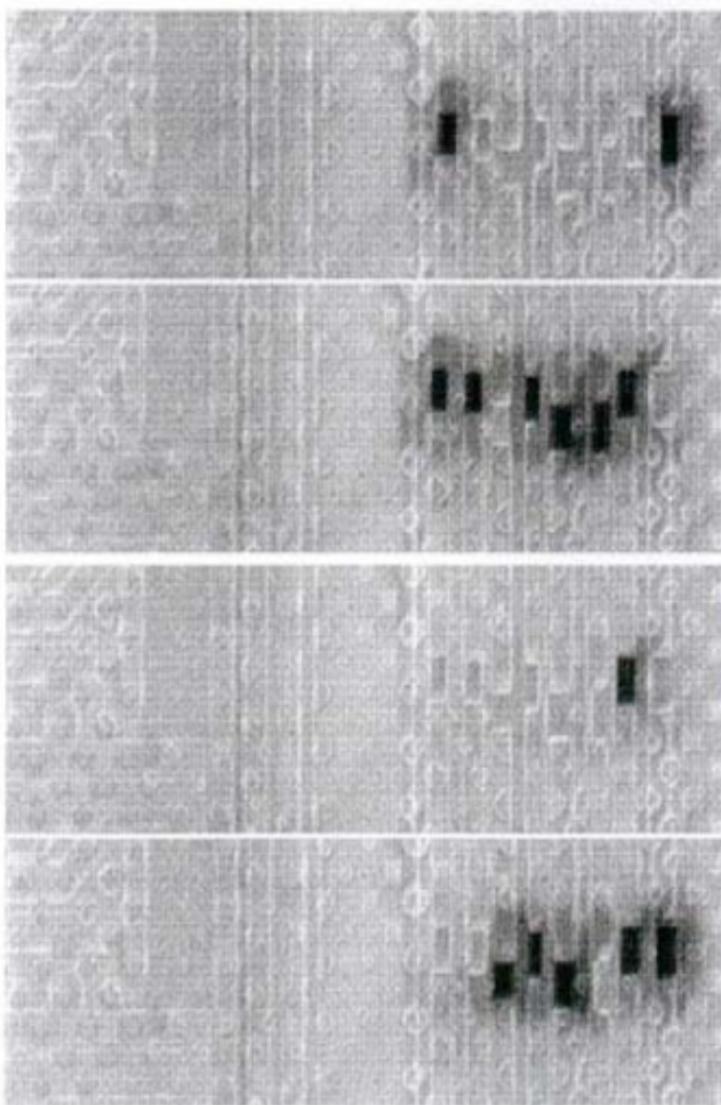


Contactless, damage free probing of DC voltage within the IC

Usage example: could extract information from a Flash ROM storage cell

# Dynamic analysis of data sent on a bus

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t=0 s

10000001

t=1 s

11011110

t=2s

00000010

t=3 s

00111011

[Merle 2006]

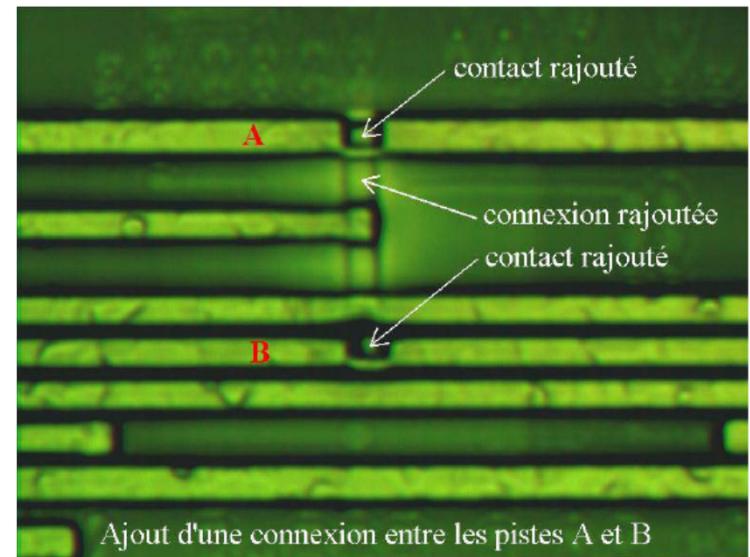
# Other physical interventions

- Cutting internal connections

- ◆ Protection disconnection

- Adding/repairing internal connections

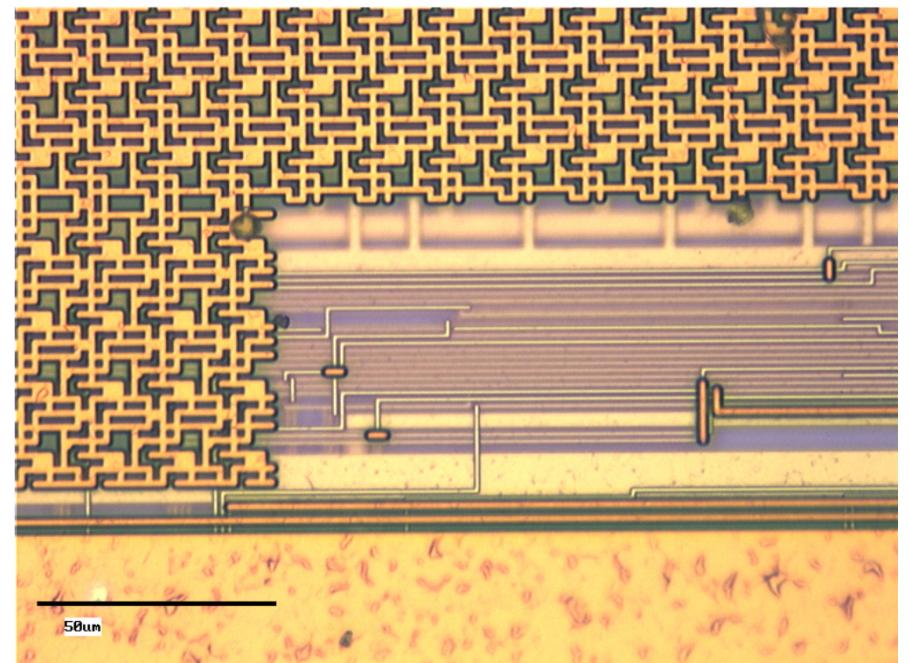
- ◆ Reconnecting a security fuse that prevents reading of the device
  - ◆ Reconnecting a cut scan access by boundary scan



# Counter-measures against invasive attacks

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- Avoid easy layout reading/exploitation
  - ◆ For optical observation after layer removal => e.g. memory scrambling
  - ◆ For probing => e.g. bus scrambling (or encryption)
  
- Protect against local probing
  - ◆ Protection grid (active) for passive probing
  - ◆ Detection signals on extra wires for active probing



# Implementation Attacks

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## □ Invasive attacks on the circuit

## □ Side Channel Analysis

- ◆ Time Analysis
- ◆ Power Analysis
- ◆ EM analysis
- ◆ ...

## □ Test structures

- ◆ Scan chains,
- ◆ ...

## □ Which countermeasures?

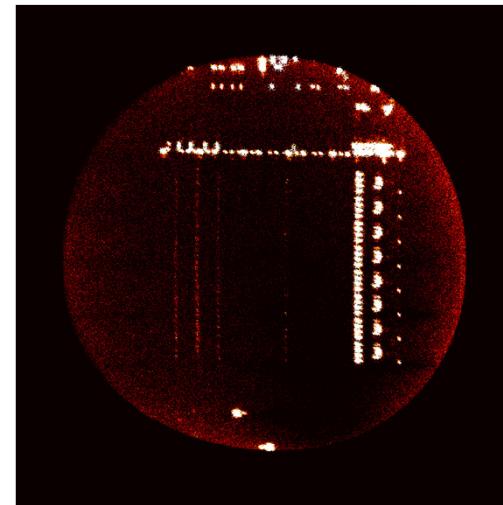
## □ Fault attacks

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- ◆ Harmonic EM,
- ◆ Pulsed EM injection,
- ◆ ...

# Side channels

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- Execution time => Timing Attack (TA)
- Power consumption => Power Attacks (SPA / DPA / CPA)
- Electromagnetic emissions => Electromagnetic Attack (EMA)
- Sound, heat, photons ...
- And many more ?



# Conditions of side channel exploitation

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- Relationship between processed (secret) data and a given physical quantity that can be measured
  - ◆ It must exist ...
  - ◆ ... and a (good) model of the relationship must be available
- Known executed algorithm, so that predictions can be made
  - ◆ e.g. AES, RSA ...
  - ◆ Security must no more be based on secret algorithms ...
- Global secret used in independent pieces, so that exhaustive search can be done on separate parts
  - ◆ Replaces the complexity of a brute force attack by a much less easy search repeated on the several pieces of the secret

# Side channel analysis

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## □ Simple Side Channel Analysis

- ◆ Makes use of characteristics that are directly visible in one measurement trace.
- ◆ The secret key needs to have some simple, exploitable relationship with the operations that are visible in the measurement trace.

## □ Differential Side Channel Analysis

- ◆ Looks for side channel differences that are not directly visible in one measurement trace => statistical methods
- ◆ Targets one specific intermediate result (typically a selection function, i.e., an intermediate result at the beginning or end of the cryptographic algorithm) that shows up in a specific part of the measurement traces.
- ◆ The result of the selection function depends on the known input/output data and a small number of hypotheses on the secret (key) value.
- ◆ The outcome of the selection function leads to a partitioning of the overall measurement data for each hypothesis used.
- ◆ For the correct key hypothesis, different statistical properties of the two partitioning sets are expected at that points in time which depend on the result of the selection function.

# Timing attacks - principles

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- Software level: e.g. If Then Else => execution time may be directly related to the tested condition (e.g. secret key bit !!)
- Hardware level: possible similar effect of condition tests in FSMs or data-related computation time in operators
- Security requires balanced execution paths and avoiding tests on critical (secret) values
  - ◆ Example:

If ( $x=1$ )

then  $a:=a+b;$



$t[0]:=a;$   
 $t[1]:=a+b;$   
 $a:=t[x];$

# A practical example

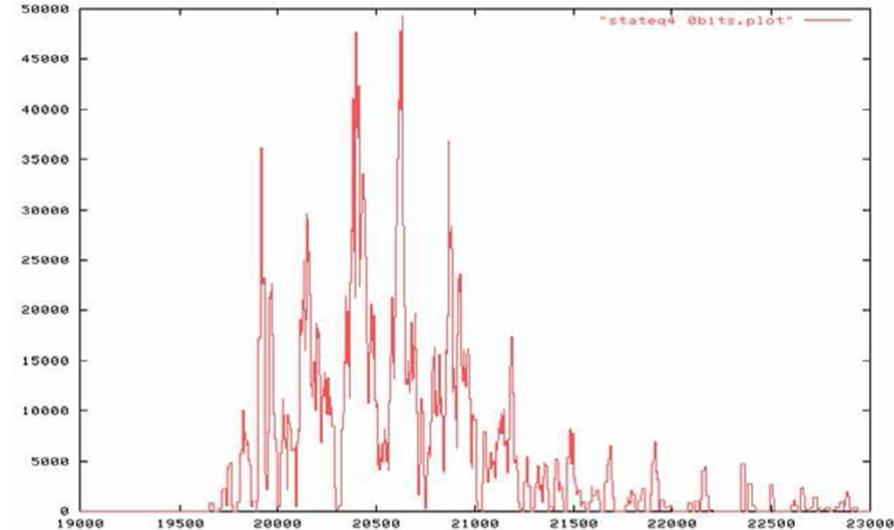
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- Timing attack on a PIN code
- Basic algorithm: successive comparison of the digits
- The more time before PIN rejection, the more digits are right ...
- Successive exhaustive trials on each digit: for each one, the longest rejection time corresponds to the right value (since the next one has also been checked)
- PIN trials limitation necessary (+ better algorithm) !!!

# Architectural features and attacks

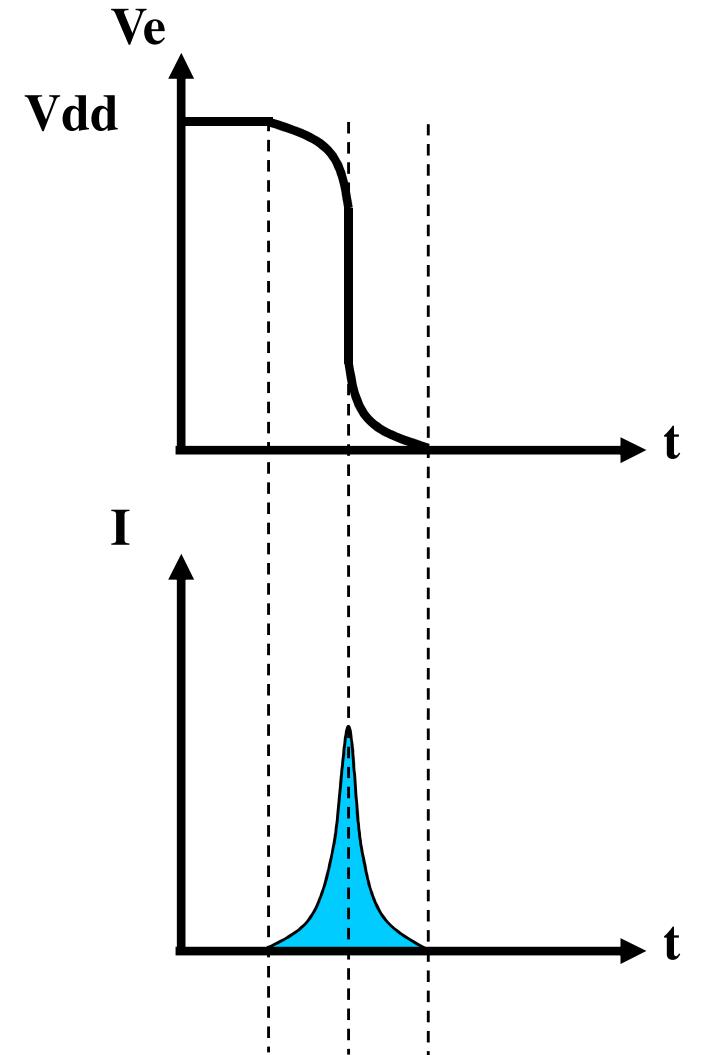
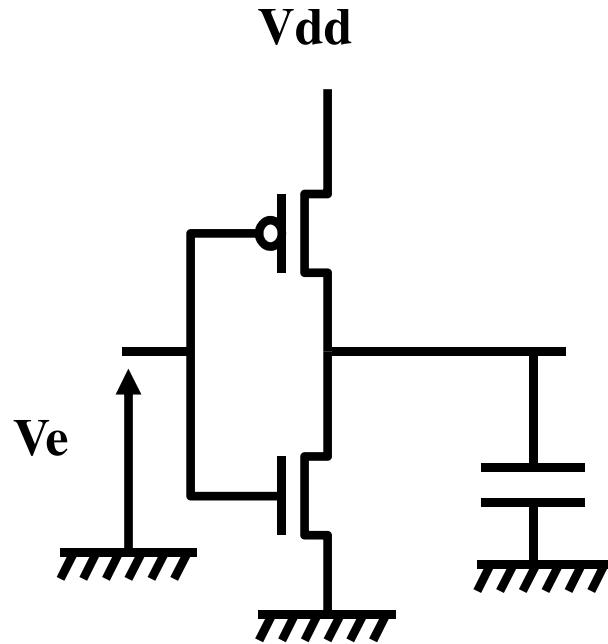
- Performance optimizations in processors can help timing attacks

e.g. effect of HW optimizations in Pentium 4, for constant-time assembly-level implementation:



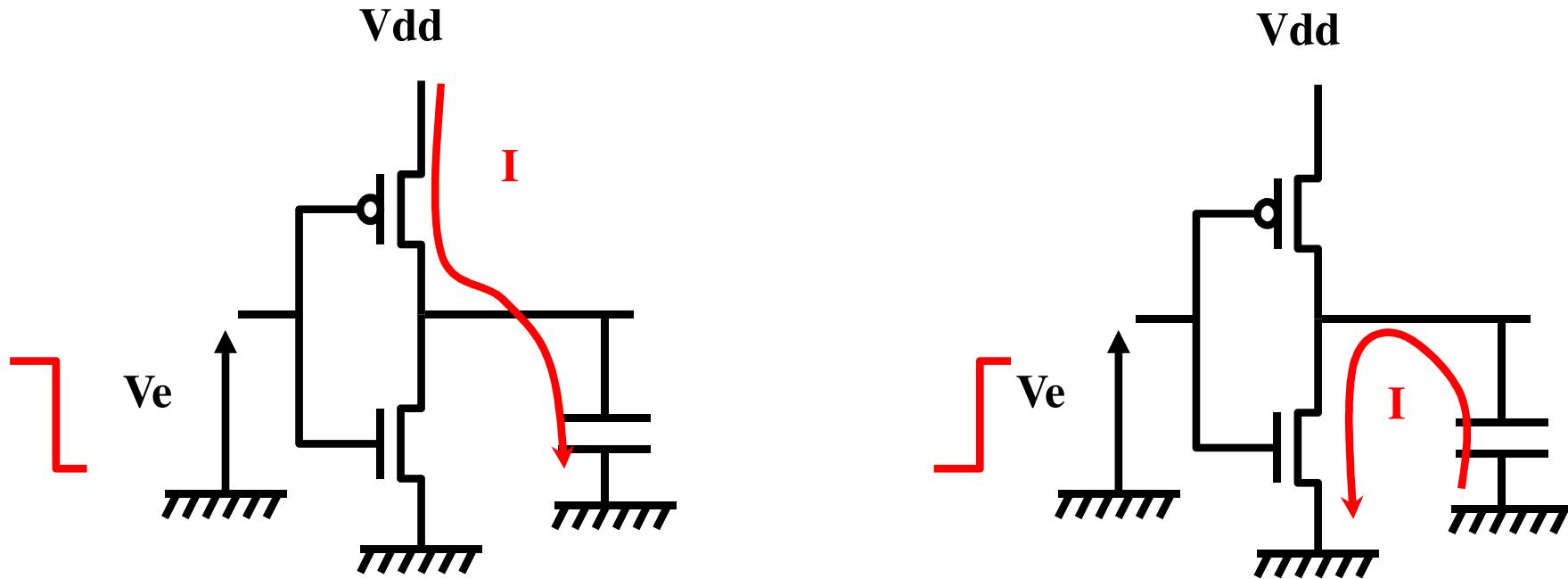
- Branch predictions (already executed branch has not the same execution time as a new branch)
- Cache memories (access to cache has not the same execution time as access to central memory)
- Performance counters provide accurate picture

# Power consumption in CMOS - Basics



- Three components:
  - ◆ Static consumption
  - ◆ Dynamic consumption (logic commutations)
    - Short circuit current
    - Load charge/discharge

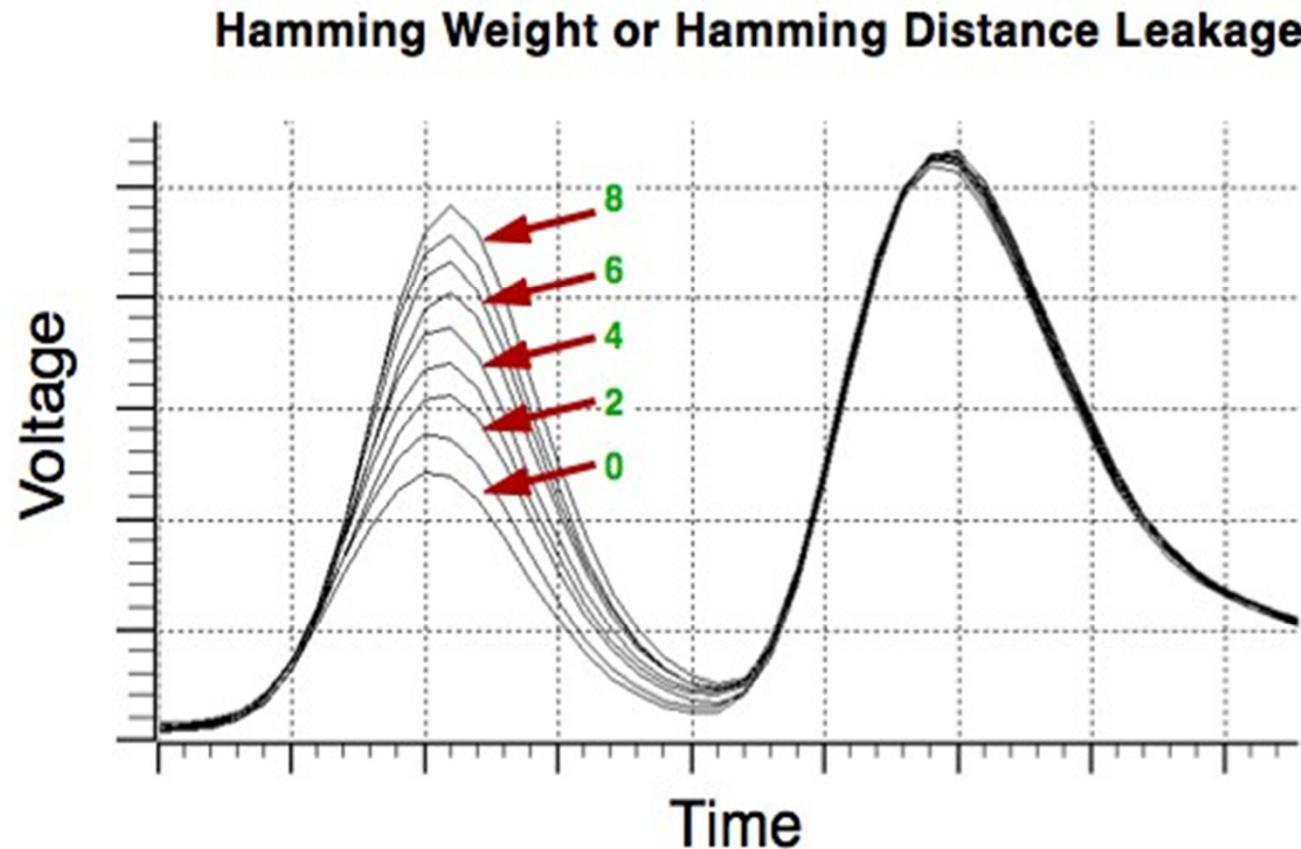
# Power consumption in CMOS - Dissymmetry



- Opposite logic commutation => different current signature
- Not the same current flow on  $V_{dd}$  and  $V_{ss}$
- Not exactly the same shape (value in time) due to either NMOS or PMOS conduction

# Information leakage: power vs. transitions

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# Reading Hamming weight from SPA trace

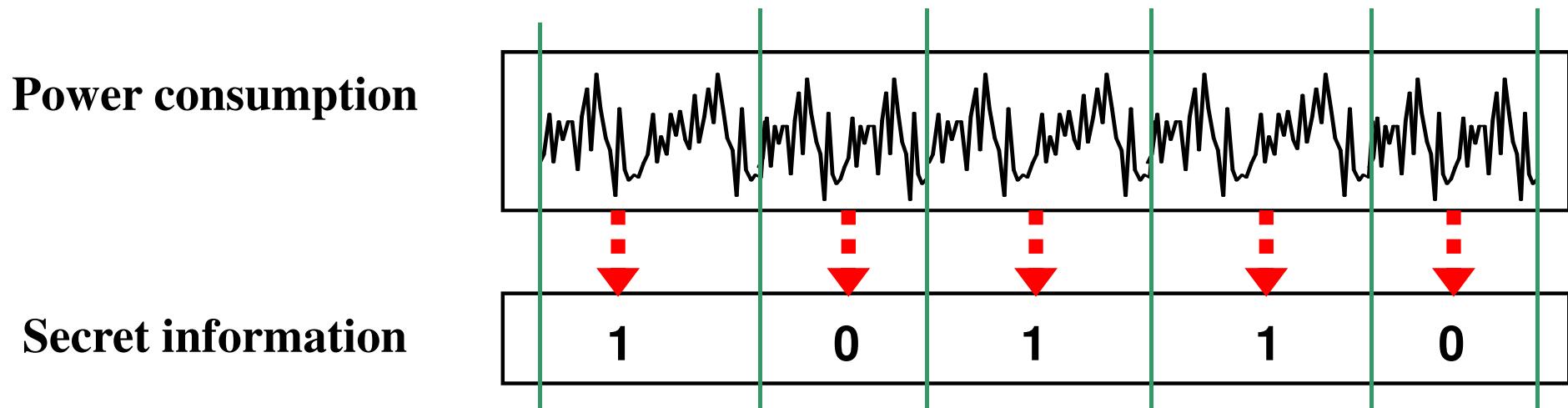


Requires to know at what time significant data must be observed ...

# SPA (Simple Power Analysis)

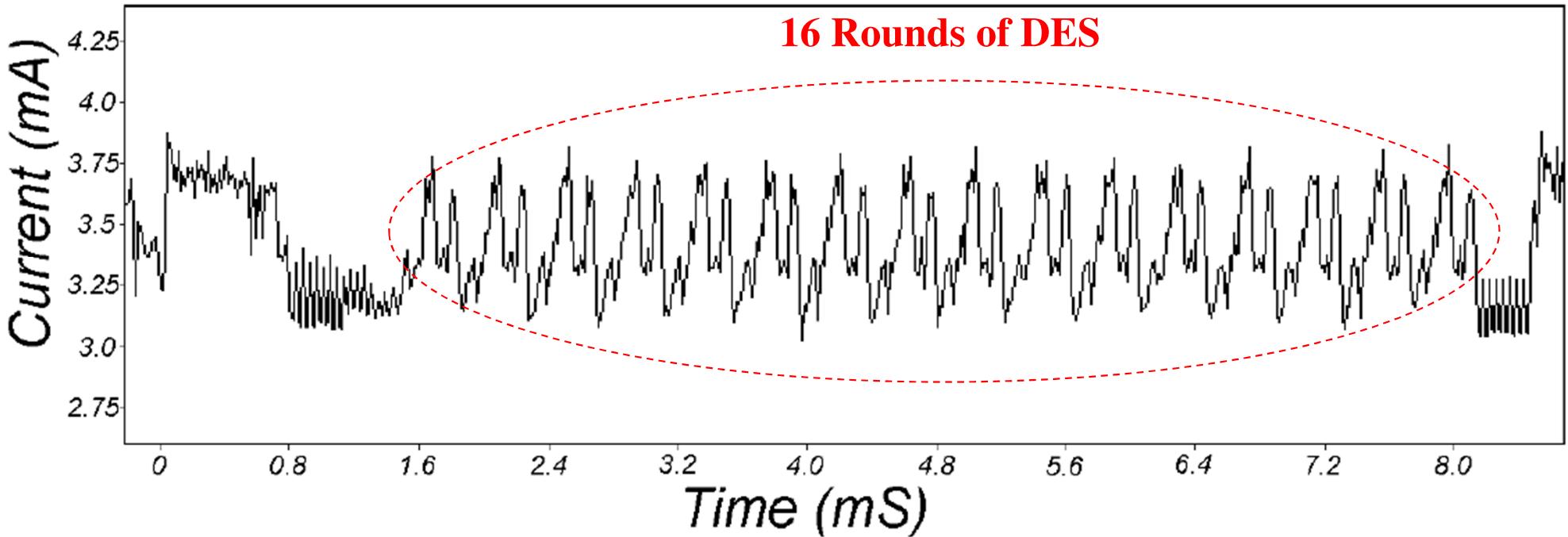
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- Monitor the power consumption trace => signatures
  
- Find translation into
  - ◆ Executed instructions (software)
  - ◆ Manipulated data



# Example of SPA - DES

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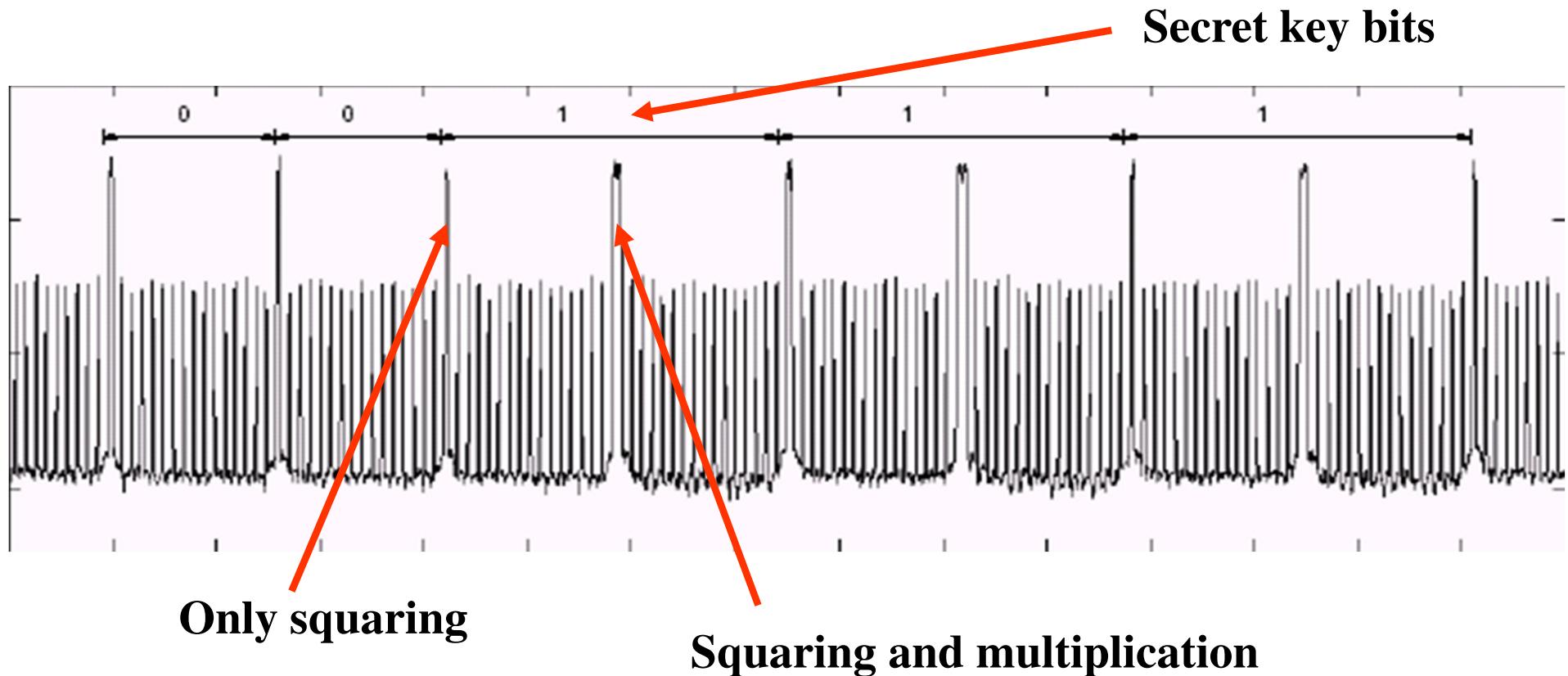


- May also be used to
  - ◆ Synchronize another attack (e.g. laser shot)
  - ◆ Measure the duration of some internal (intermediate) computations (TA)

# Example of SPA - RSA

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## □ RSA computation: supply current trace



# Example of SPA - ECC

## General scalar multiplication algorithm

### Algorithm 1. Double-and-add algorithm

Input : A point  $P$ , and  $d = (d_{n-1}d_{n-2}\dots d_1d_0)_2$ ,  $d_{n-1} = 1$

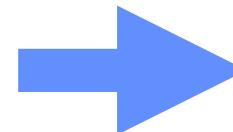
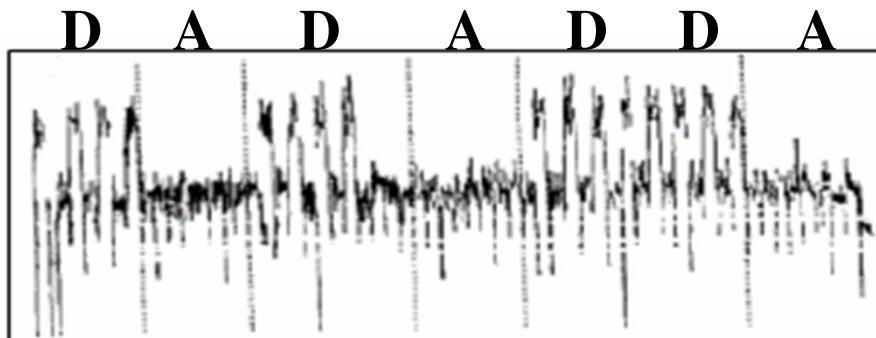
Output :  $dP$

1.  $S = P$
2. For  $i = n - 2$  downto 0
  - 2.1  $S = 2S$
  - 2.2 If  $d_i = 1$ ,  $S = S + P$
3. Return( $S$ )

**d: secret exponent**

**Point Doubling (D):**  
Execution at each round

**Point Addition (A):**  
Execution when bit value is "1"



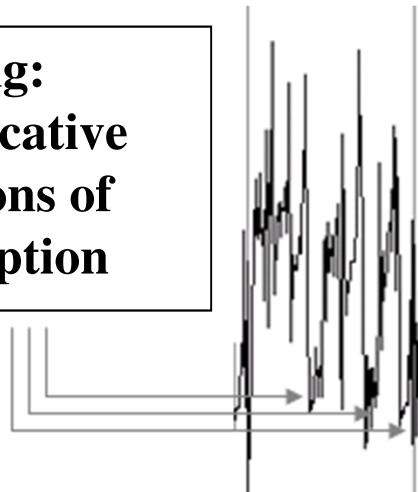
$d = 1101$

# Simple computation

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- Addition / doubling
- Curve with 270 points (private key between 1 and 269)
- Characterization of power consumption

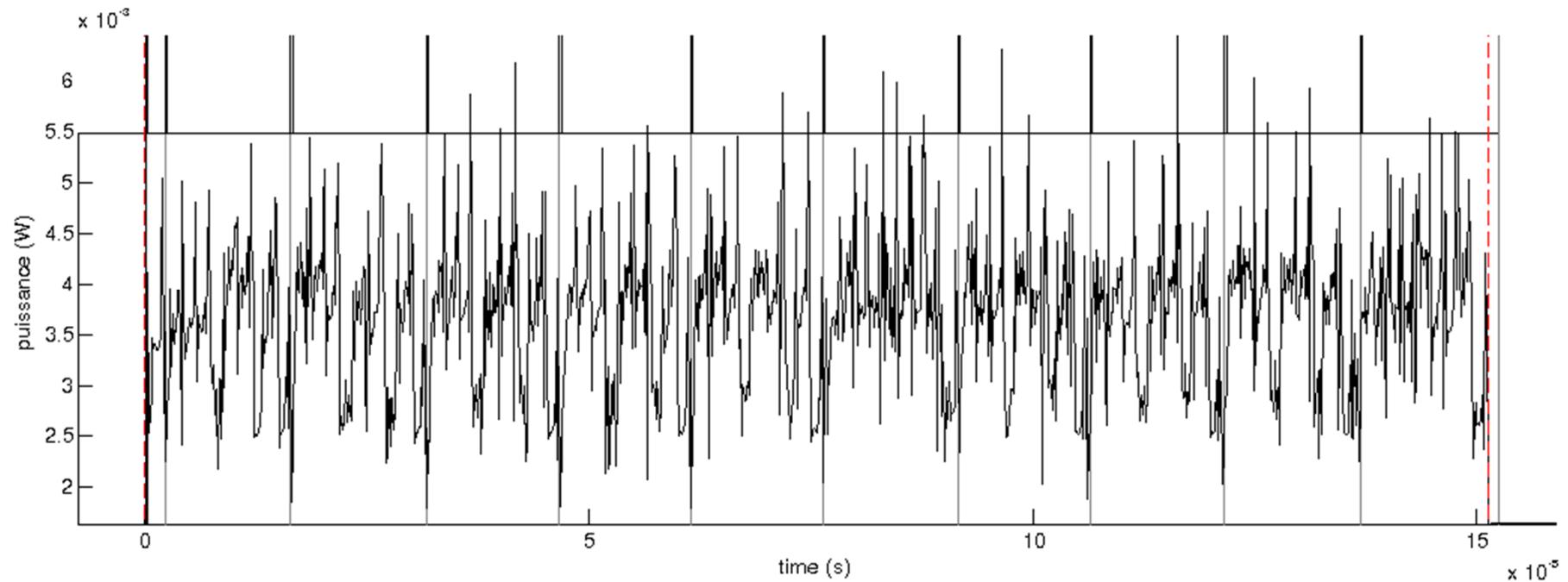
**Doubling:**  
3 significative  
reductions of  
consumption



**Addition:**  
Only 1  
significative  
reduction of  
consumption



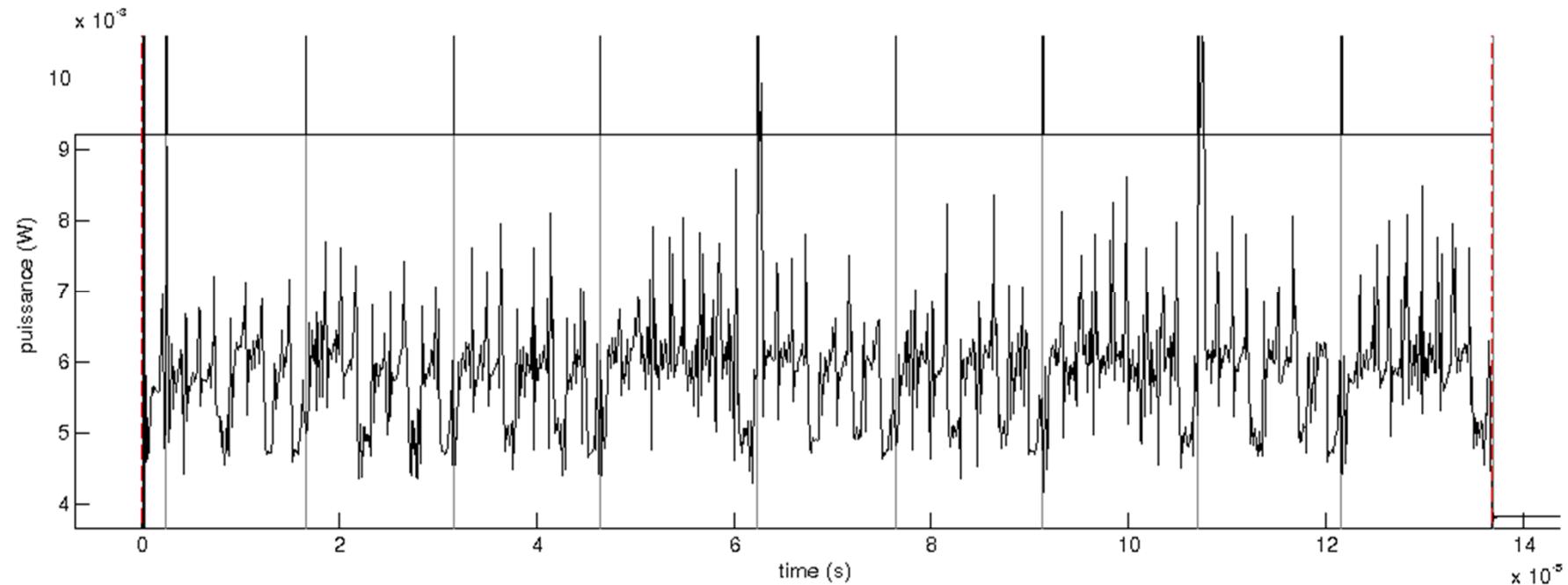
# Which key?



- Add / Doubling? (or accumulator overwriting)
- What about MSBs?
- What is the private key?

# Window-based computation

- Random window size: 2 or 3 bits (when necessary, 1 bit for LSB)
- Pre-computed values for additions on 2 or 3 bits

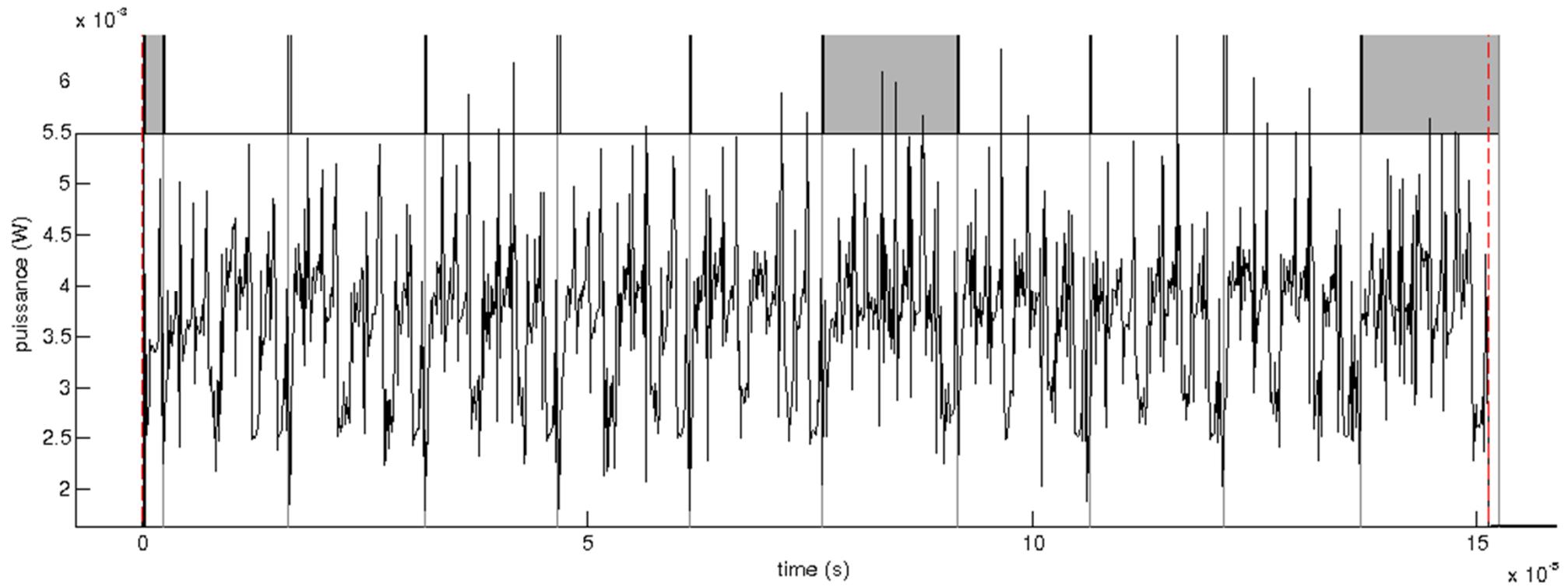


- Add / Doubling? (or accumulator overwriting)
- What about MSBs?
- What is the private key?

# Answers (1) ...

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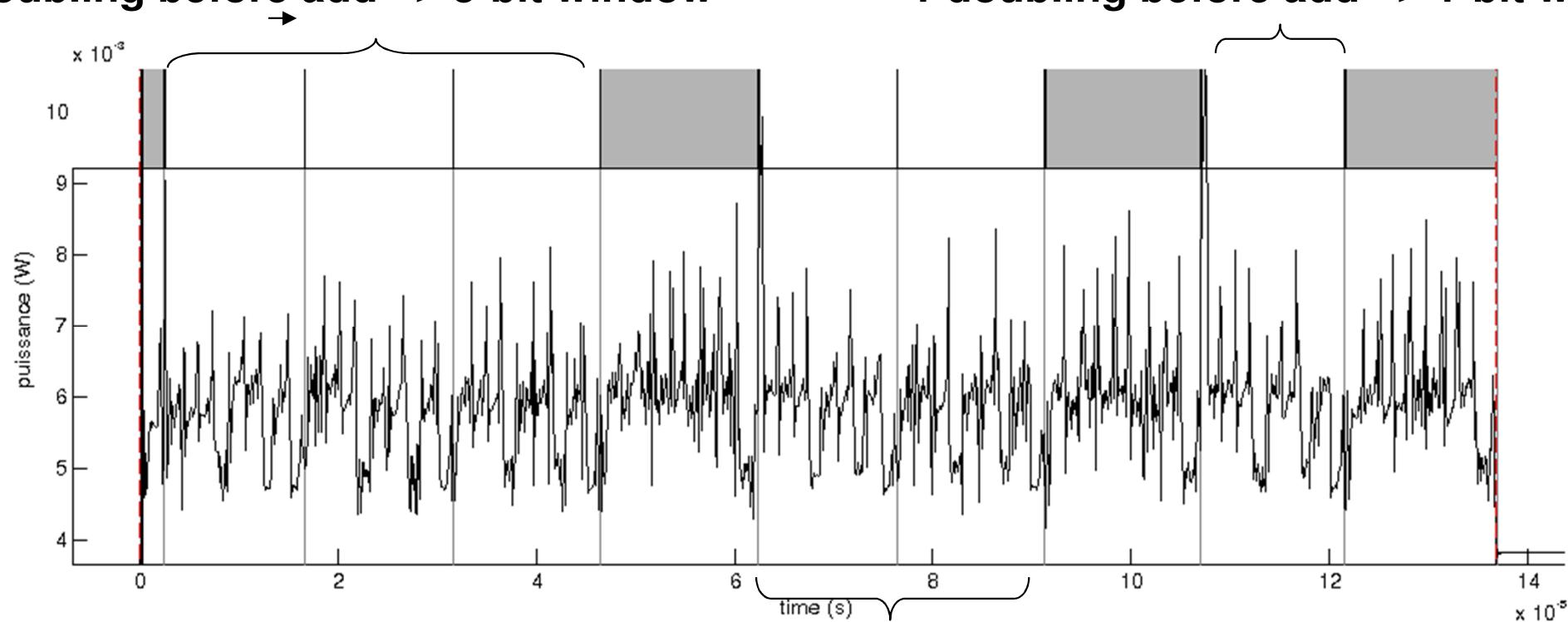
- Key with simple computation: 8 doubling (+1 over-writing),  
key with 9 bits => MSB=1 => reading additions,  
key=100001001 (265)



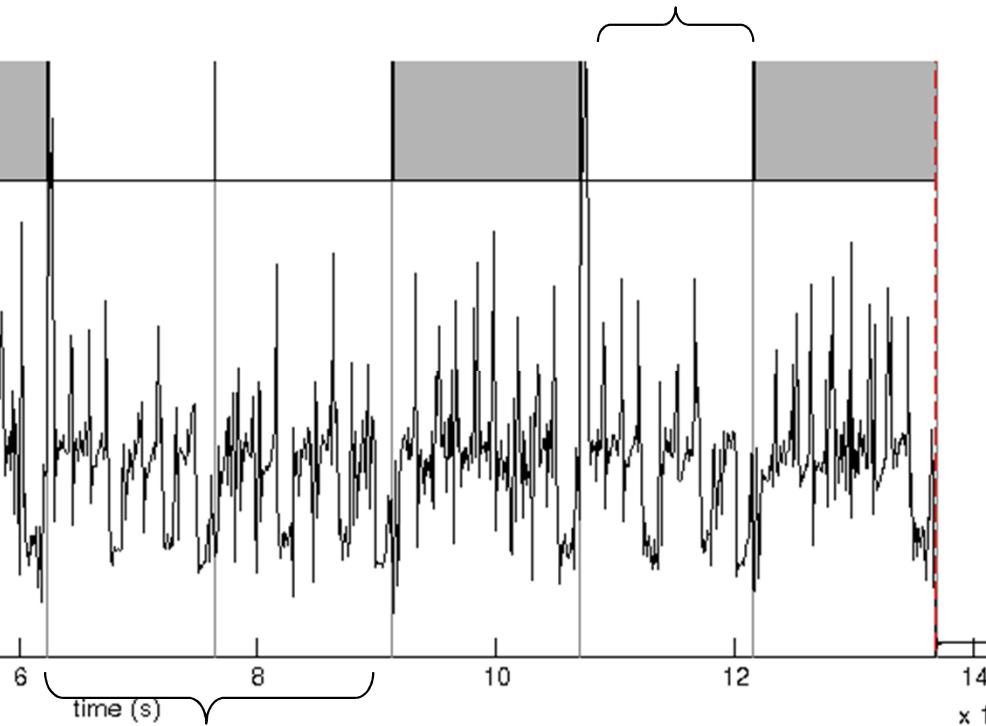
# Answers (2) ...

- Key with windowed computation: 4 windows, LSB=1

**3 doubling before add => 3-bit window**



**1 doubling before add => 1-bit window**



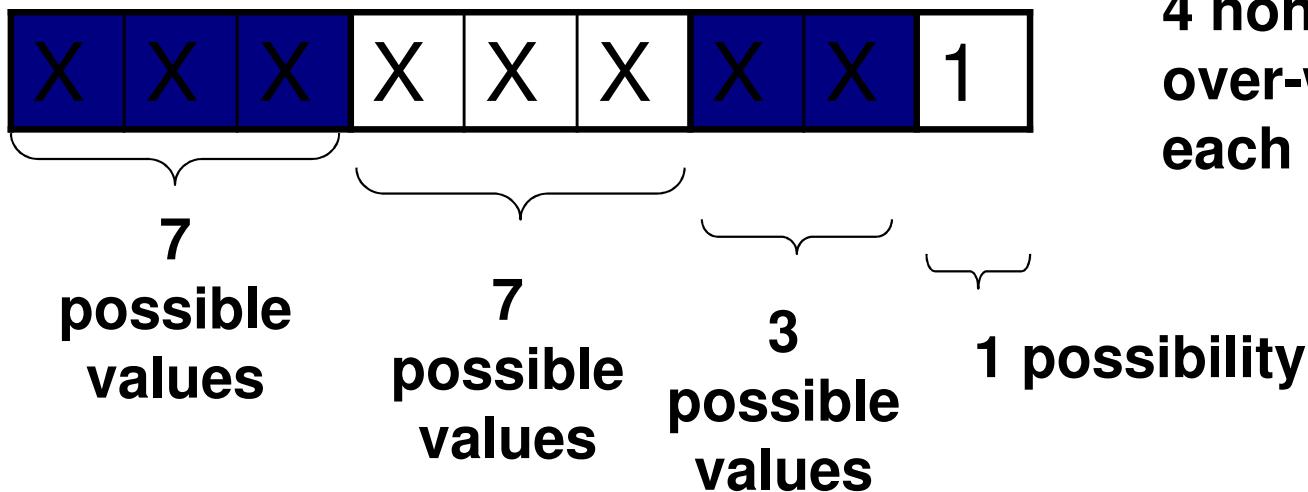
**3 (non-null) bits before over-writing!**

**2 doubling before add => 2-bit window**



# Answers (3) ...

- Key with windowed computation: 4 windows, LSB=1



4 non-null windows since over-writing or addition for each ...

⇒  $7 \times 7 \times 3 \times 1$  possible keys (147 values, so 54% of all keys)

BUT key  $\leq 269$  so ... how many possible keys?

Key was  $(001\ 011\ 11\ 1)_2 = 95$

# DPA (Differential Power Analysis)

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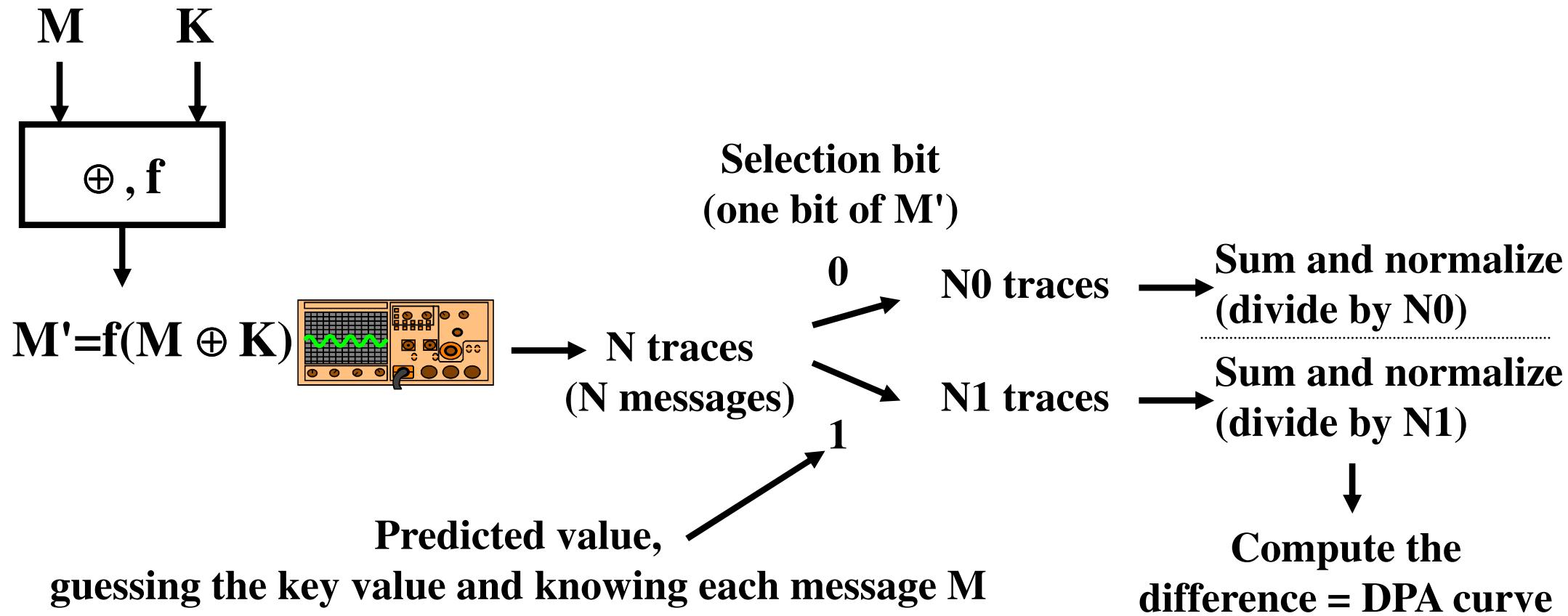
- Introduced by P. Kocher (1998)
- Data collection phase + data analysis phase
- Basic procedure
  - ◆ Gather many power consumption curves
  - ◆ Assume a key value
  - ◆ Divide data into two groups (0 and 1 for chosen bit)
  - ◆ Calculate mean value curve of each group
  - ◆ Correct key assumption → not negligible difference
- Attack in pieces (divide and conquer - small parts of the key)  
– e.g. DES only  $2^6$  choices per Sbox (exhaustive search feasible)

# DPA - Basics

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- Modeling the power consumption: Hamming weight model
  - ◆ Typically measured on a bus,  $Y=aH(X)+b$   
(Y: power consumption; X: data value; H: Hamming weight)
  - ◆ Other possibility: the Hamming distance model  
 $Y=aH(P \oplus X)+b$  - Accounting for the previous value on the bus (P)
- DPA can be performed in any algo that has operation  $M' = f(M \oplus K)$ , where M, f are known and K is the segment key
- The waveforms are captured by a scope (many executions with many messages) - The selection bit will classify the wave
  - ◆ Hypothesis 1: bit is zero / Hypothesis 2: bit is one
  - ◆ The DPA waveform with the highest peak will validate the hypothesis
  - ◆ A differential trace is calculated for each bit
- Typical: 500K+ acquisitions ...

# DPA - Illustration



If function  $f$  only depends on a small number of bits of the key (typically Sbox), exhaustive search of this sub-key becomes easy

- 1 DPA curve per possible sub-key value
- highest peak shows the right guess (contrast or peak height proportional to  $N^{1/2}$ )

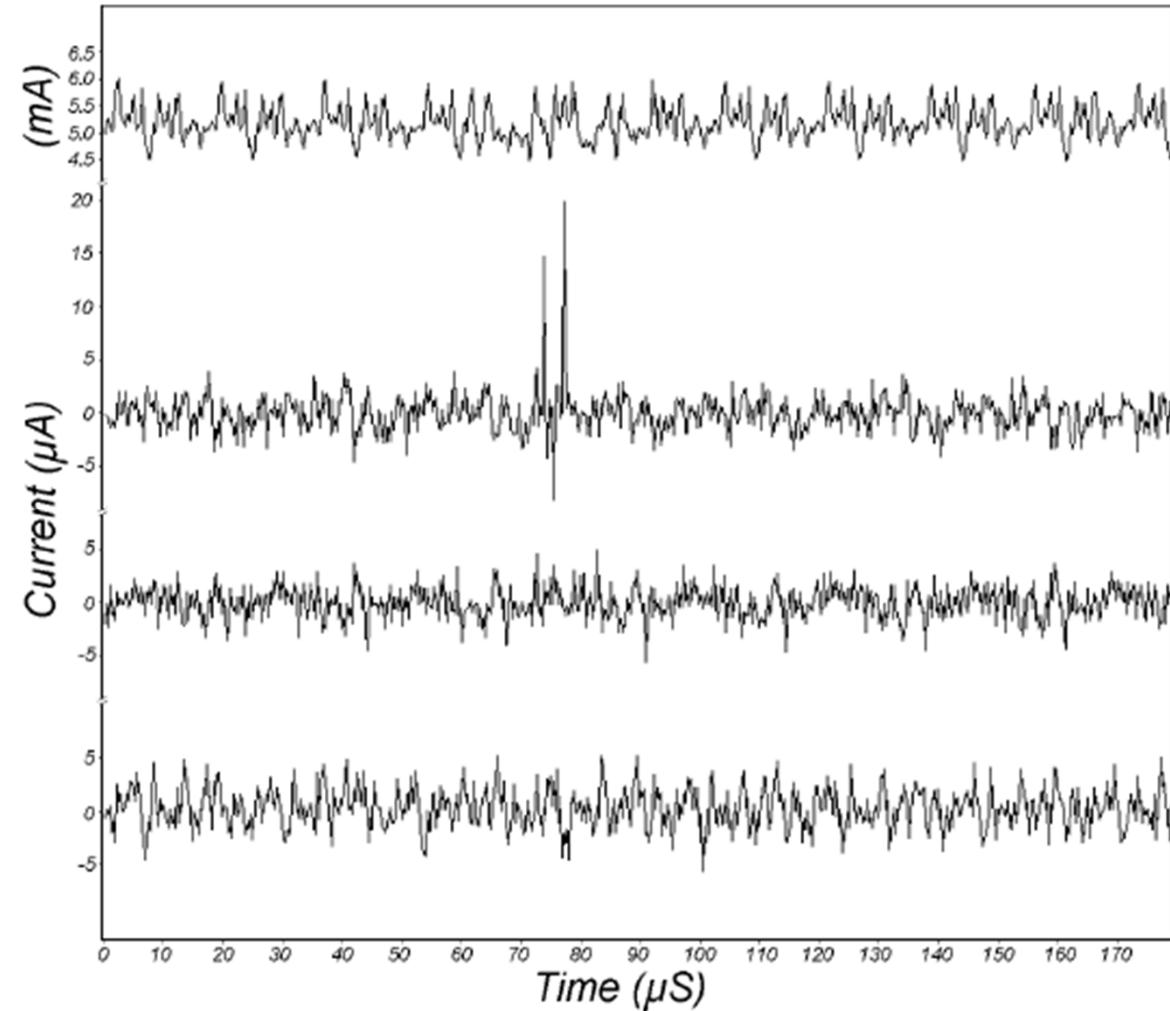
# Example of DPA - DES

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Average power consumption

Power consumption differential curve with correct key guess

Power consumption differential curves with incorrect key guess



# Example of DPA - AES

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- Number of computations ?
- 128-bit key (16 bytes) – Each byte used independently as a 8-bit Sbox input (after message XORing)
- 256 DPA curves to compute per byte, **16\*256 to find the 128-bit key**  
$$2^4 * 2^8 << 2^{128} !!$$
- But
  - ◆ Several peaks can be similar (or even a wrong guess may lead to a higher peak)
    - Algorithmic noise (implementation model)
    - Implementation characteristics
  - ◆ Comparing results for several selection bits may help (but do not always agree !)

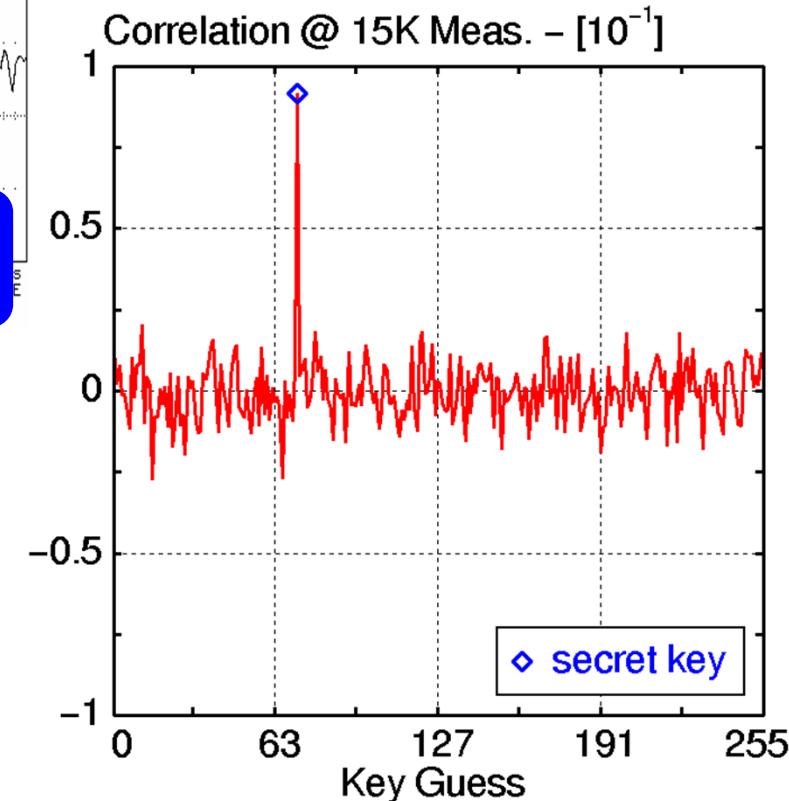
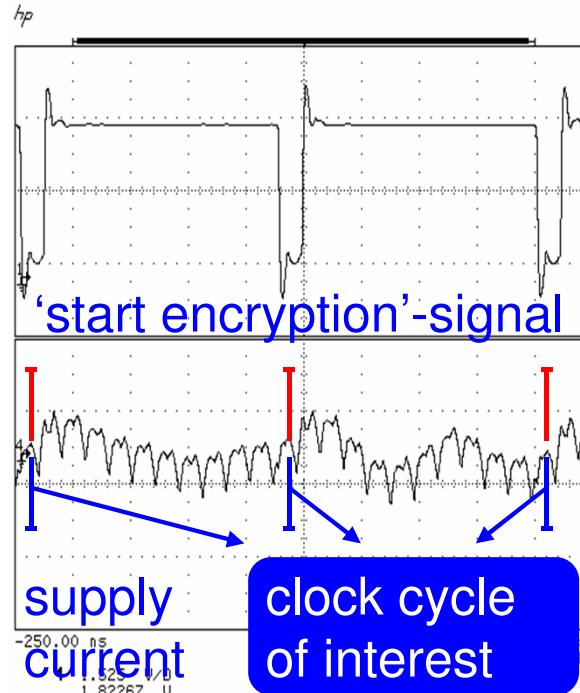
# And attacks are more and more acute ...

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- Mono-bit vs. multi-bit DPA
- Second order power analysis
  - ◆ Rather than observing a single consumption time, focus on correlation between two points in time
  - ◆ Can be extended to 3rd, 4th order ...
- CPA: Correlation Power Analysis
  - ◆ The equation for generating differential waveforms are replaced with correlations
  - ◆ Rather than attacking one bit, the attacker tries predicting the Hamming weight of a word
- Template attacks
  - ◆ A single (a few) sample may be sufficient (suited to stream ciphers, and cases with key re-use limited by system level protocols)

# Power analysis - AES

- Unprotected ASIC AES with 128-bit datapath, key scheduling
- Measurement: Ipeak in round 11
- Estimation: HamDistance of 8 internal bits
- Comparison: correlation
- Key bits easily found despite algorithmic noise
- 128-bit key under 3 min.



Source: K. Tiri, Intel

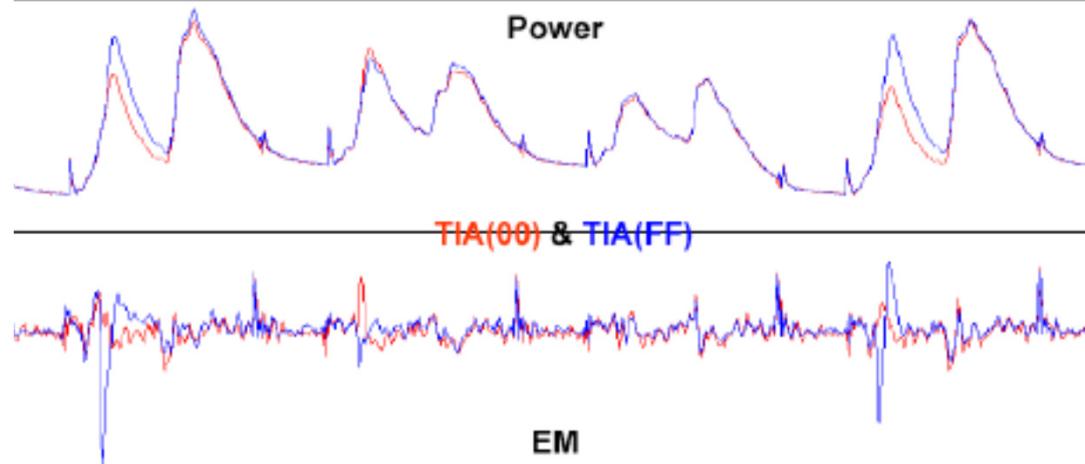
# Electromagnetic attacks (SEMA – DEMA)



**Same principles as power attacks, with a different measurement equipment giving more precise (more localized) internal measures and bypassing current smoothers ... but experimentally more difficult to put in practice.**

# Comparison power/EM

- EM more local (depending on setup and probe), with wider bandwidth (~100 MHz power, ~1 GHz EM) => more data in the recorded signal



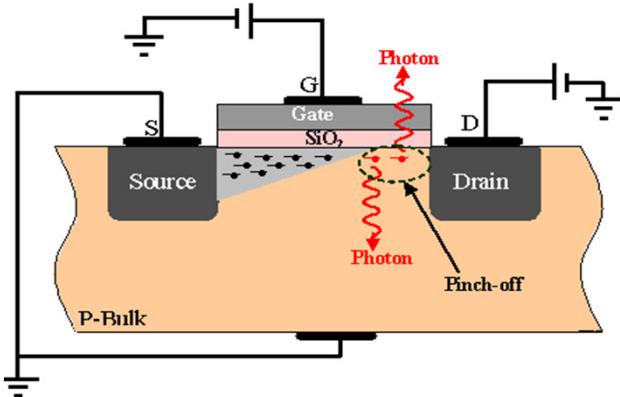
- No direct connection required for EM (e.g. better for FPGA PCB !)
- Magnetic field better for smartcard analysis than Electric field, but small amplitude and SNR => low noise amplification required + trade-off between precision (probe size) and signal amplitude
- Many parameters: probe size and orientation, measurement position on chip, distance to chip, time interval for analysis, ...

# EMA "success story" - example

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- EAL4 product - AVA\_VAN.5 robustness level
- Protected ECC crypto-processor with "add always" implementation or RSA with "multiply always"
- About 2 weeks of (manual/visual) EM measurement analysis by expert
- Broken key – due to difference in storage address (least significant bit) for the useful and useless additions or multiplications

# Attacks based on light emission (photons)

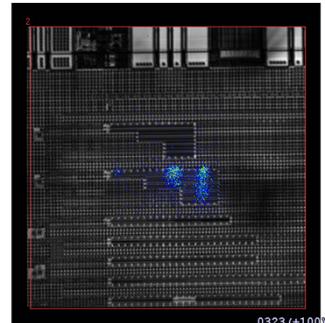


NMOS transistor

Failure analysis equipments  
Hamamatsu – Tri PHEMOS



- Spatial analysis of photon accumulation
- Dynamic analysis techniques (TRE, PICA)
  - ◆ Signal propagation analysis
  - ◆ Functional analysis (function behavior - reverse engineering)



Courtesy J. Di-Battista (PhD defense - April 2011)

# Counter-measures against SCA

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- Attacker always needs an initial guess to compare his measurements to - Protection: make measurements unguessable or constant (data independent)
- Several levels
  - ◆ Algorithmic level
  - ◆ Architecture level
  - ◆ Gate level
  - ◆ ...
- Examples
  - ◆ Avoid conditional execution
  - ◆ Use constant-time (i.e. worst-case time) programs
  - ◆ Use dedicated “non-leaking” logic
  - ◆ Randomize execution flow (SW or HW) – But resynchronization attacks are possible
  - ◆ Add noise... but this will only require more samples for a successful attack

# Protections against power analysis

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## □ Hardware:

- ◆ Main idea: balanced switching (dual rail, or N rails, with systematic simultaneous 1 -> 0 and 0 -> 1 transitions ... on SAME load ...)

**Not so easy ! Specific design techniques/tools (dual functions),  
load balancing (P&R ...), ... but increasing process variations !  
+ must avoid evaluation anticipation (internal Z nodes), ...**

- ◆ Other ideas:

- Reduced synchronization by variable/random jitter on clock or some operations (not just noise !!)
- Masking of actual algorithmic values (manipulated data are not directly correlated to the key value) – but still vulnerable to DPA
- Add dummy operations (possibly randomly) => useless power consumption during sensitive operations
- Random execution order for independent instructions or computations

## □ Software: Specific coding techniques (usually at assembly level)

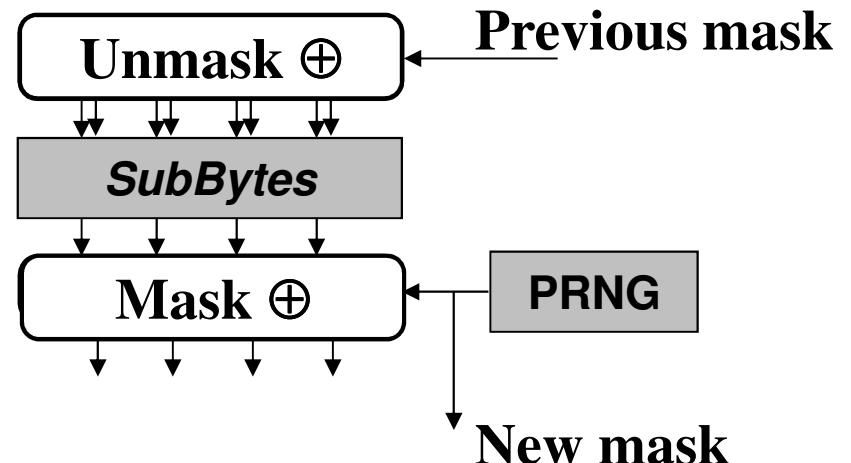
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# RTL Countermeasures – Masking

---

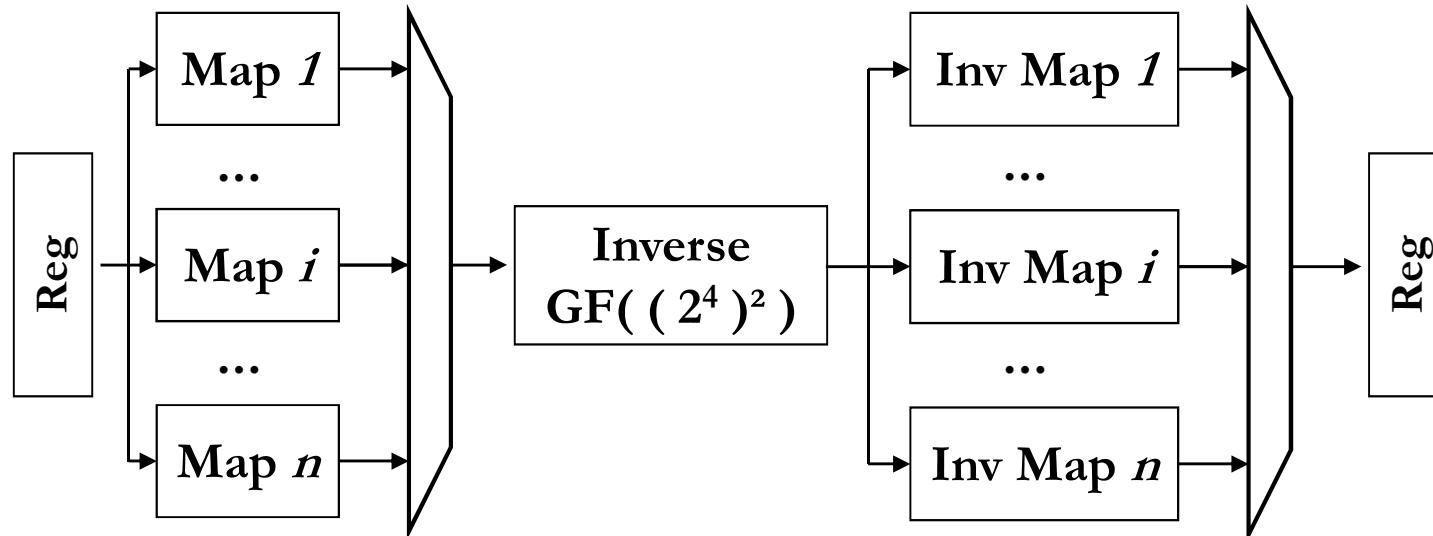
- Linear masking is a common low-cost masking scheme for linear data path (P network)
  - ◆ Masking at S-Box output
  - ◆ Unmasking at S-Box input
  - ◆ Change mask value as often as possible

- S-Box not protected
  - ◆ Use different (multiplicative) masking
  - ◆ Rotating masked S-Boxes



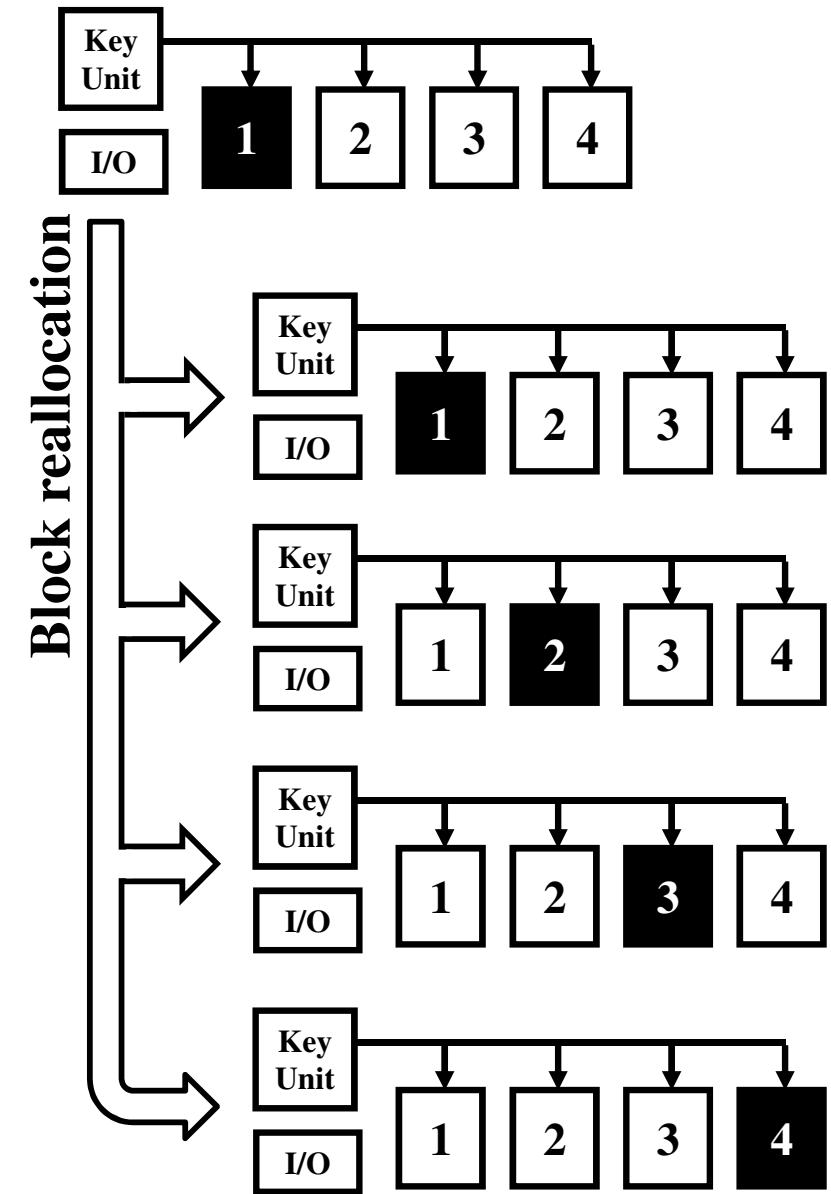
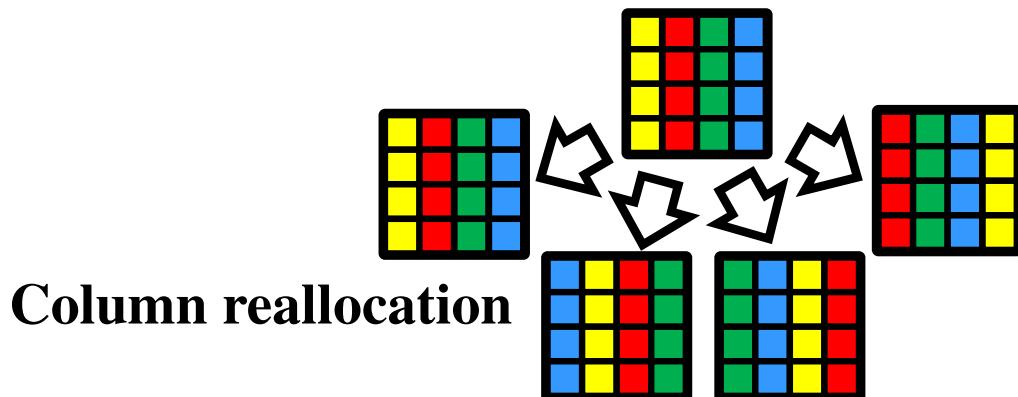
# RTL Countermeasures – Dynamic Representation

- For each S-Box, implement several parallel composite field mappings
  - ◆ From 1 to 8 possible dynamic mappings for any composite field
  - ◆ Choose randomly at runtime (several granularities)
  - ◆ At output, choose the correct inverse mapping to get back the result
- Limited to S-Box data path
- Independence of mappings?
- Vulnerable to Zero- and One-value attacks (as multiplicative masking)



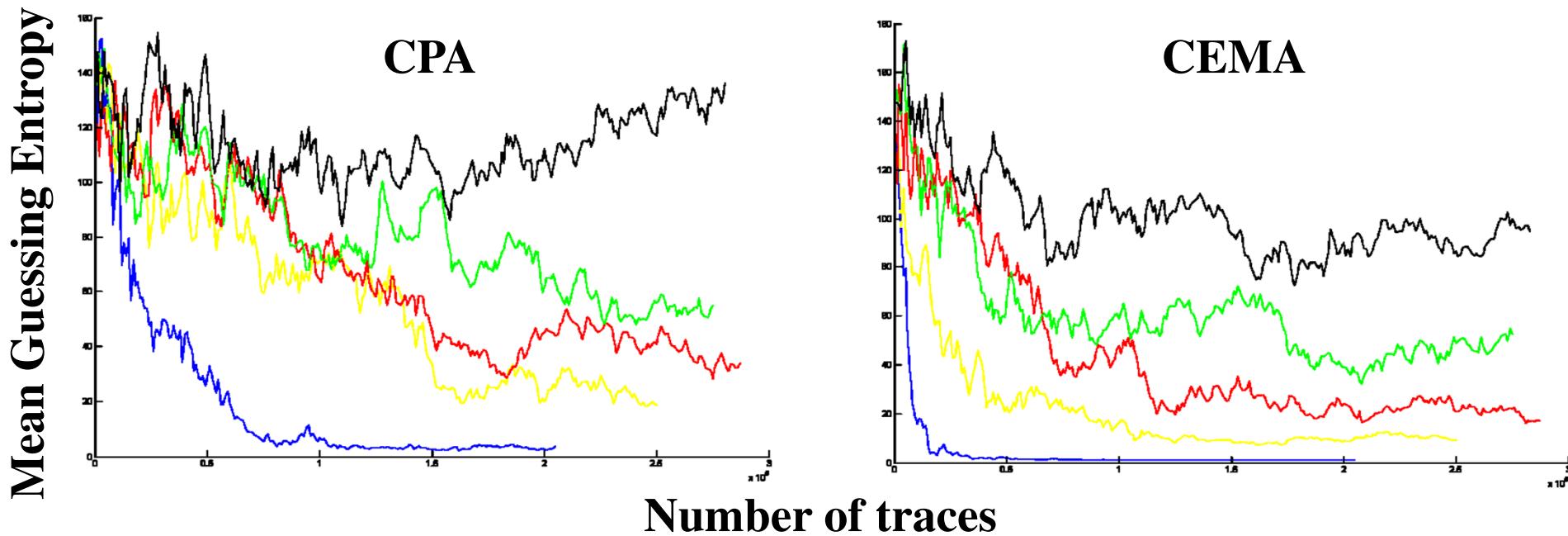
# RTL Countermeasures for AES

- Random register renaming (sw)
- Correct order is restored before output
- Random data allocation
  - ◆ Block-wise
  - ◆ Column-wise
  - ◆ Byte-wise
- Several external constraints due to data dependency



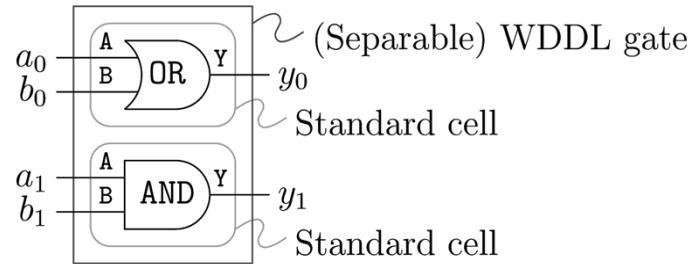
# EM vs Power (again...)

Design	Power Analysis			EM Analysis		
	Key bytes found	Mean Guessing Entropy	# traces (x10 <sup>3</sup> )	Key bytes found	Mean Guessing Entropy	# traces (x10 <sup>3</sup> )
<b>Basic</b>	15	1	205	16	1	155
+ LinMsk	4	54	275	8	52	275
+ DynMap	5	34	287	9	17	287
+ D*R	7	19	250	12	9	250
+ All	0	136	283	0	94	283

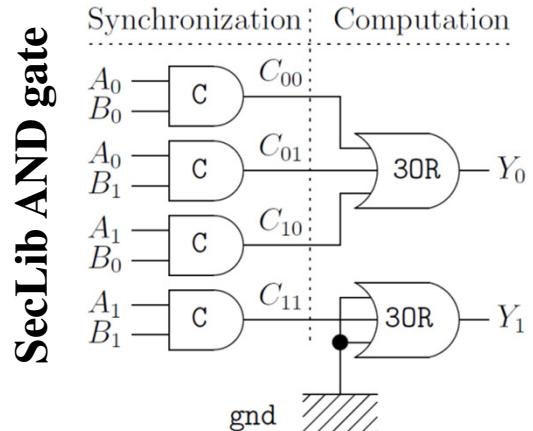


# Dual Rail Logic

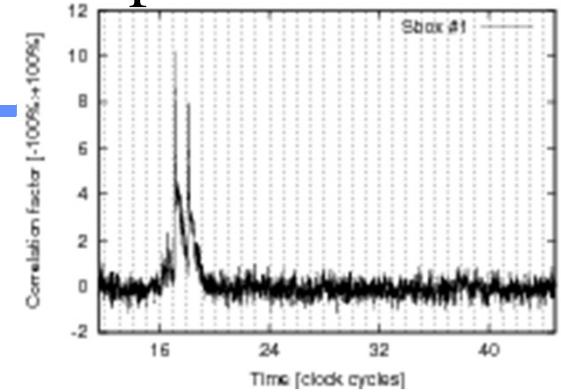
## □ WDDL: Wave Dynamic Differential Logic



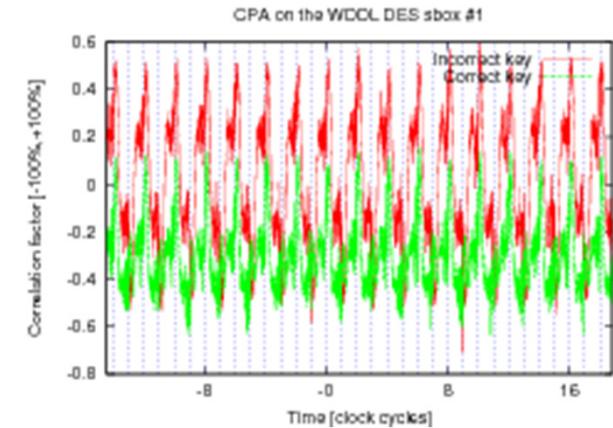
## □ SecLib: full-custom balanced quasi-delay insensitive (QDI) cell library



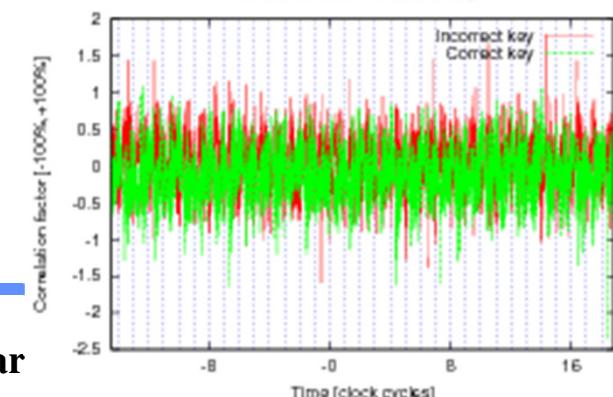
## 1. Unprotected version



## 2. WDDL resistance

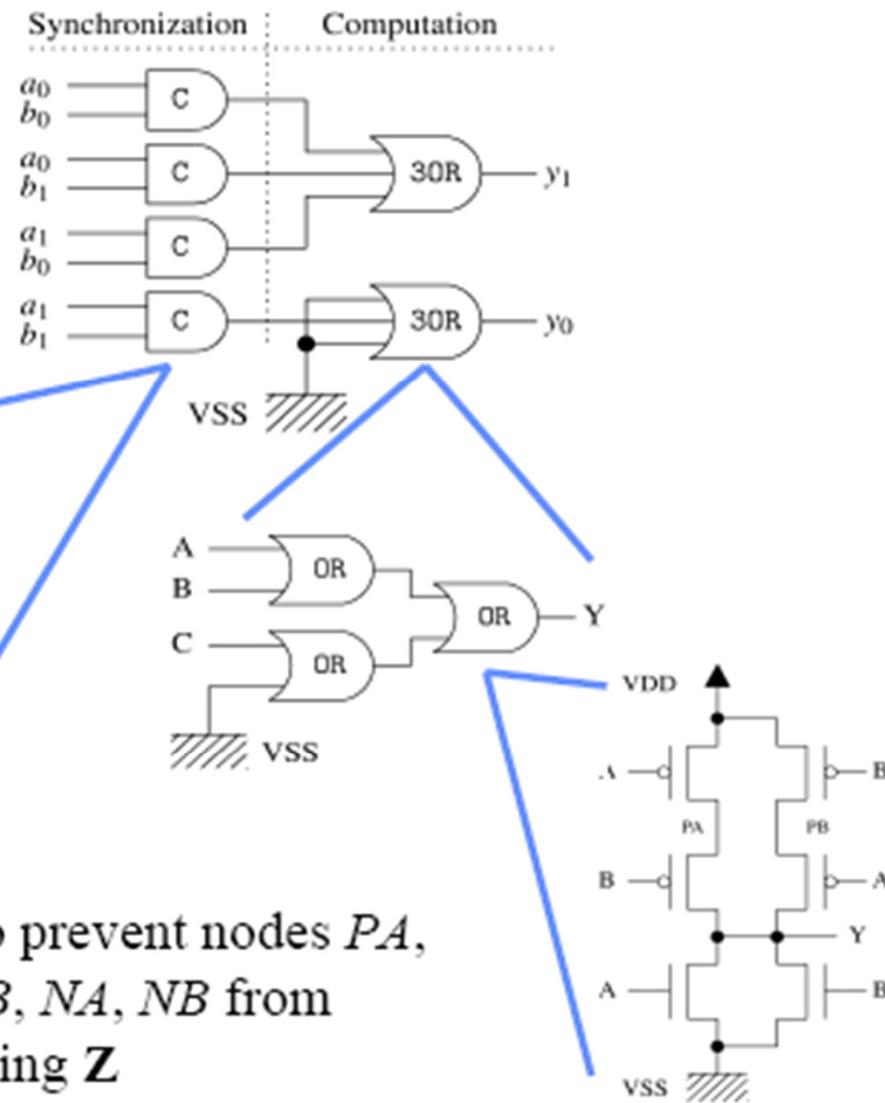
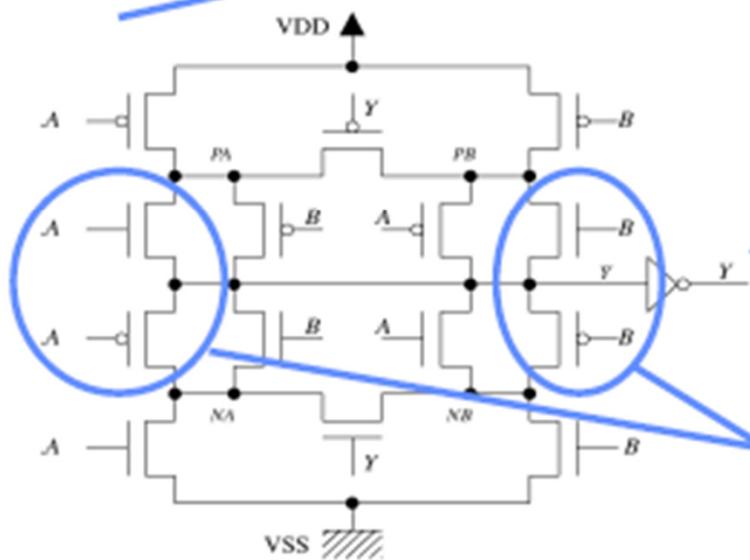


## 3. SecLib resistance



# Example of secured NAND gate

- ◆ Symmetry  $(a_0, a_1) \leftrightarrow (b_0, b_1)$
- ◆ Inputs synchronization
- ◆ No internal Z



To prevent nodes  $PA$ ,  $PB$ ,  $NA$ ,  $NB$  from being Z

From ECoFac 2006 presentation – S. Guilley et al. (GET/ENST)

# Multi-level counter-measures (SCA)

---

Level	Counter-measures against SCA (examples)
Gate (transistor level structure)	<b>Power-constant logic or dual-rail with precharge = DRL). Pwr-cst in average logic = RSL, M-DLP, etc.</b>
P&R	Differential pairs routing
RTL	Constant Hamming weight encoding => <b>DRL</b>
Algorithm	Dynamic masking, substitution box reconstruction: <b>cracked</b>
Architecture	S-RNG ( <b>biased for low N</b> ), bus scrambling ( <b>broken</b> ), public key infrastructure in NoCs ( <b>too expensive</b> )
System	Secret sharing, peripheral access after bus integrity verification

# Implementation Attacks

---

## □ Invasive attacks on the circuit

## □ Side Channel Analysis

- ◆ Time Analysis
- ◆ Power Analysis
- ◆ EM analysis
- ◆ ...

## □ Test structures

- ◆ Scan chains,
- ◆ ...

## □ Which countermeasures?

## □ Fault attacks

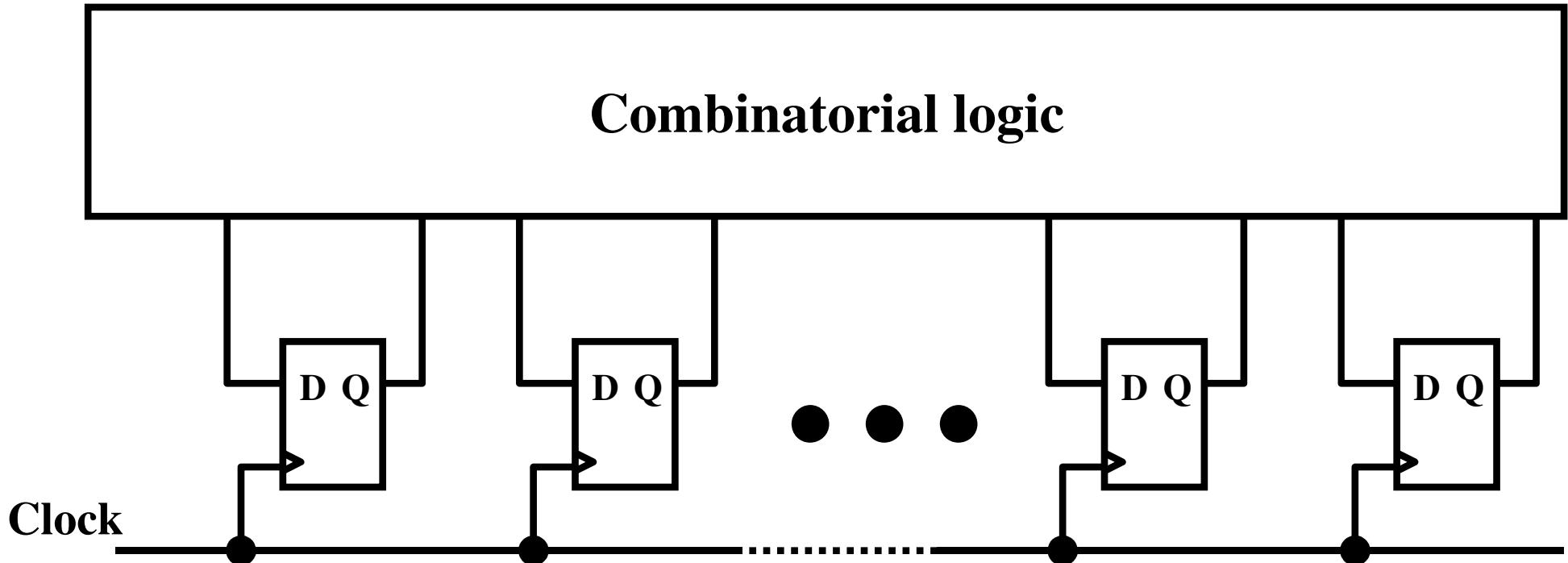
- ◆ Voltage glitches,
- ◆ Clock glitches,
- ◆ Overvolting,
- ◆ Downvolting,
- ◆ Laser shots,
- ◆ Harmonic EM,
- ◆ Pulsed EM injection,
- ◆ ...

Courtesy D. Hely  
[IOLTO6]

# "Scanpath" implementation – basics (1)

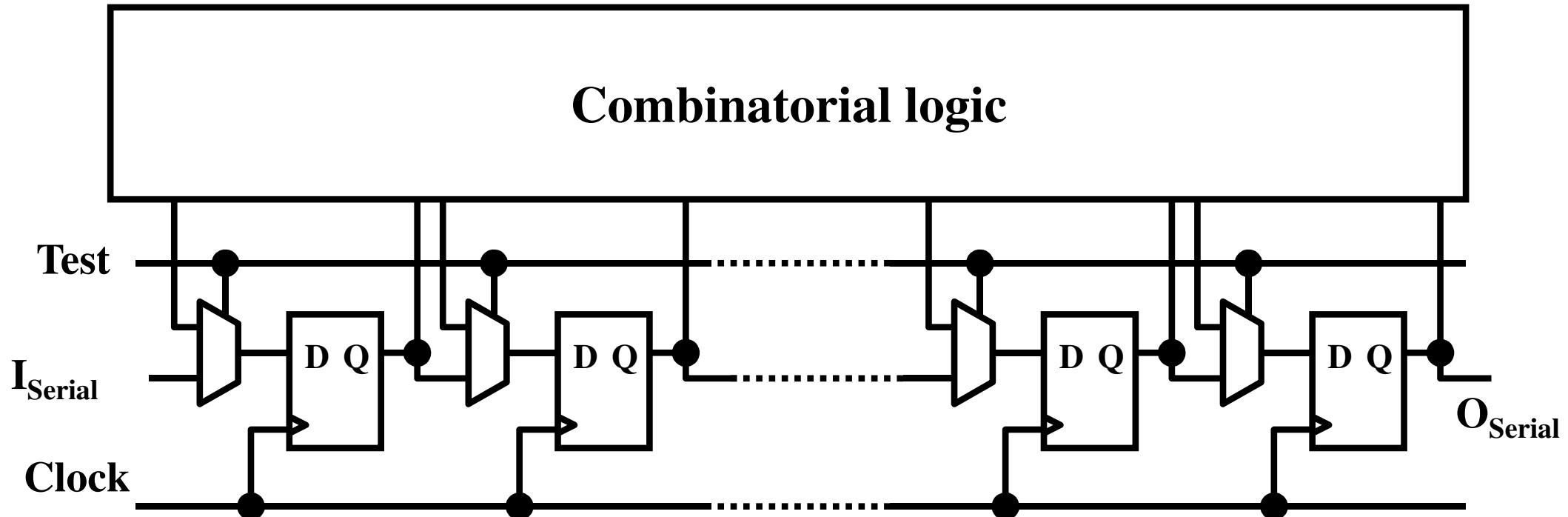
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Initial circuit:



# "Scanpath" implementation – basics (2)

Circuit with a single scan chain:



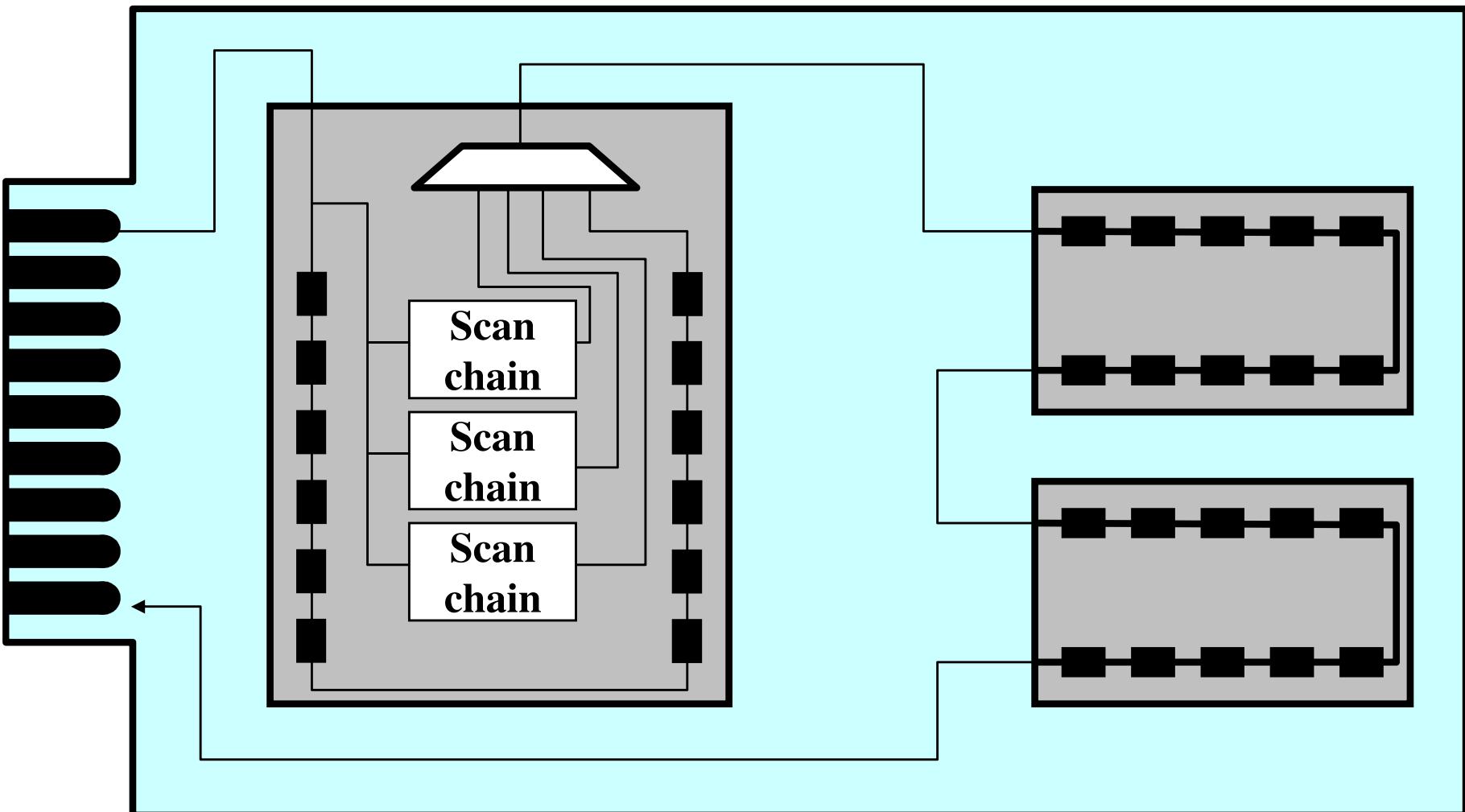
# Boundary scan (JTAG)

Test  
Access  
Port

serial  
input  
TDI

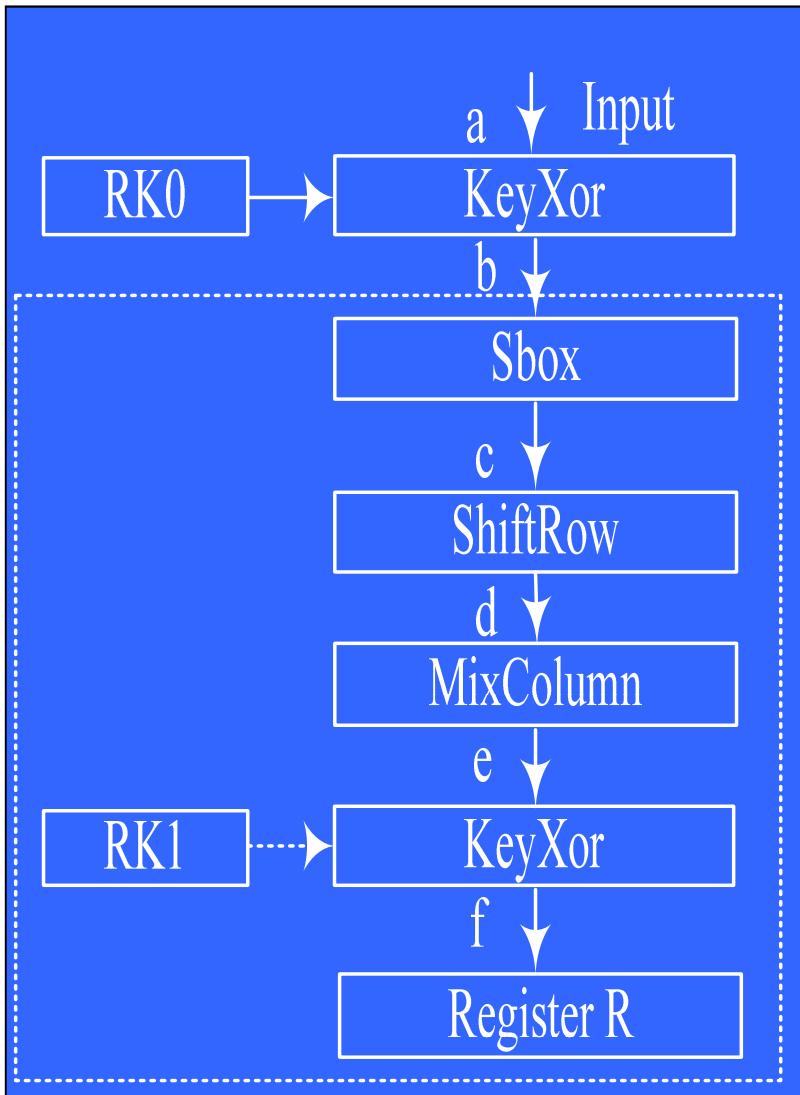
TMS  
TCK  
TRST\*

serial  
output  
TDO



Several usages: board test, in-situ circuit test, (remote) configuration

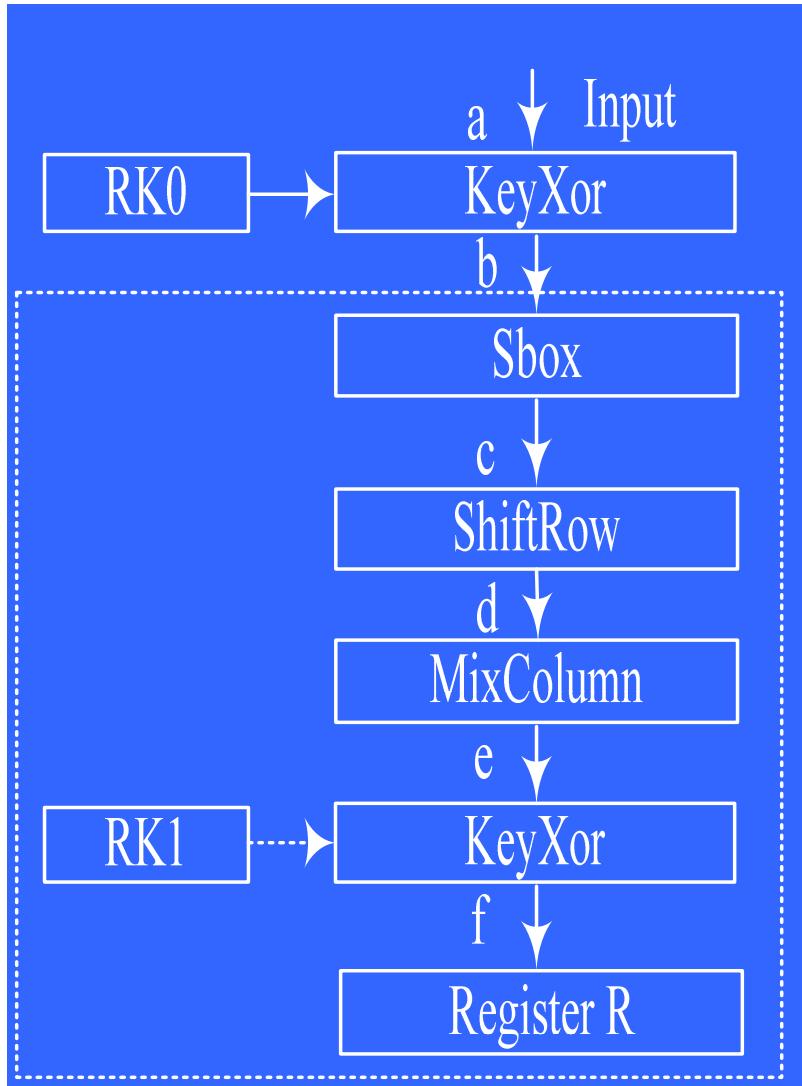
# Scan-based attack on AES – Starting point



- Input is partitioned into 16 bytes  $a_{11}, \dots, a_{14}, a_{21}, \dots, a_{24}, a_{31}, \dots, a_{34}, a_{41}, \dots, a_{44}$
- Register  $R$  is fed back to point  $b$  ten times with  $RK1$  to  $RK10$
- **128-bit Round register  $R$  is in scan chains**
- The complexity of AES is reduced to one round
- Value of  $RK0$  ?

Yang, Wu and Karri, "Secure Scan: A Design for Test Architecture for Crypto-chips", DAC 2005

# Scan-based attack on AES – 2 steps



## □ Determine scan chain structure

- ◆ The locations of flip-flops of R in the scan chains are unknown
- ◆ Change in  $a_{11}$  → change in  $b_{11}$  → change in  $c_{11}$  → change in  $d_{10}$  → change in  $e_{i0}$  → change in  $f_{i0}$  → 4 byte at R
- ◆ On average, 15 patterns are enough applied at  $a_{11}$  to determine all the 32-bit in Register R ( $f_{i0}$ ) by comparing the scanned out bit streams

## □ Recovering Round Key RK0

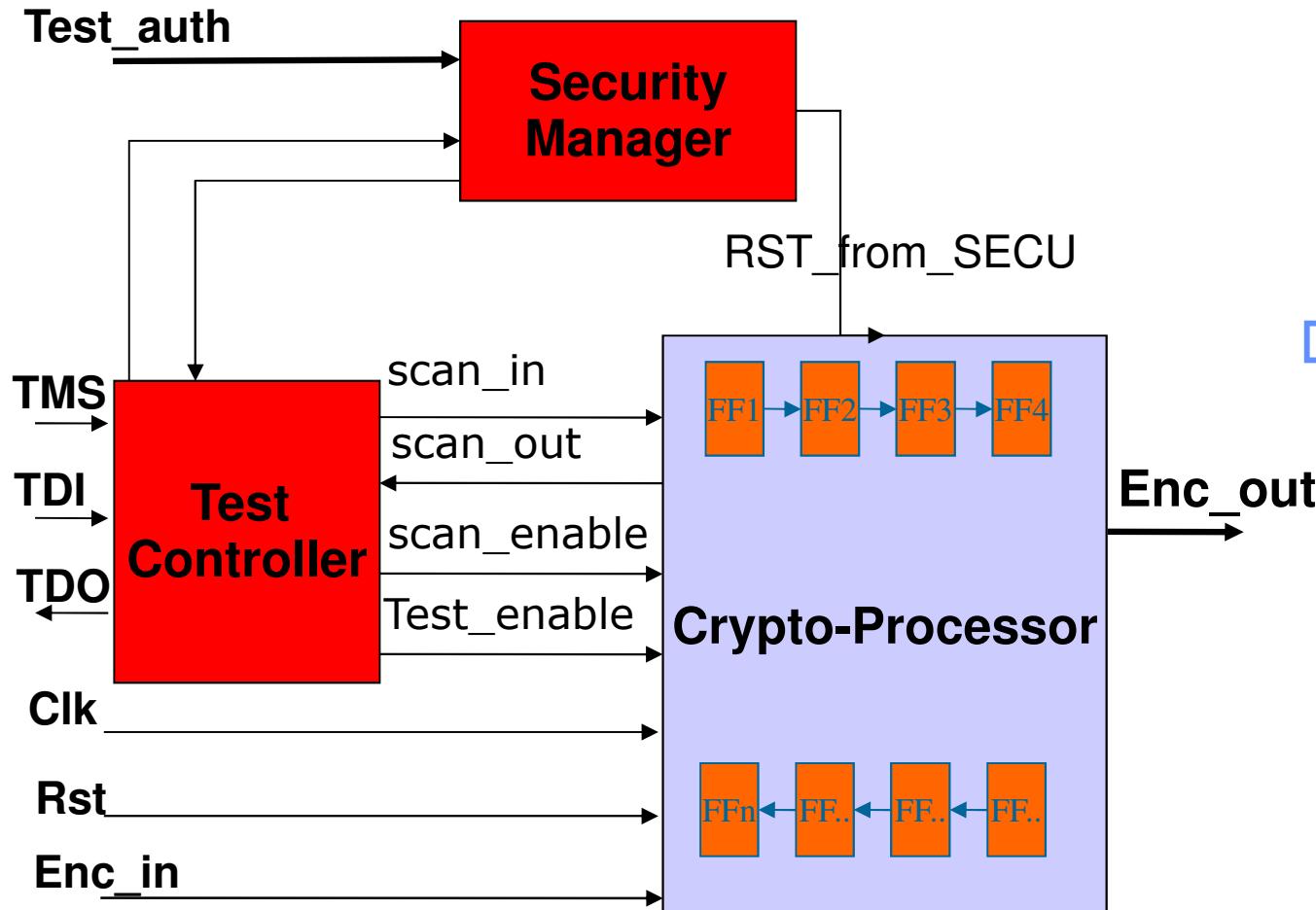
- ◆ Determining bit positions
- ◆ Cancelling RK1 effect by differential analysis
- ◆ RK0 obtained from a XOR b

# Securing test modes ...

---

- Secure test mode activation (secret key, ...)
- Detection of unexpected shifts (due to probing)
- (Variable, random) scrambling of scan chains
- Dummy bits in scan chain (detection of invalid shifted patterns)
- ...
- Avoid scan techniques when possible: self-test (BIST, SBST)

# A typical system



## □ System overview

- Security manager in charge of authentication
- A scan chain driven by a test controller

## □ Attacks

- **Test Mode Attack:** authentication is bypassed
- **System Mode Attack:** Scan\_enable activation, Scan\_out observation

Courtesy D. Hely [IOLT06]

# Securing the scan chain

---

## □ Goal

- ◆ No observation or control of the functional data processed by the secure system

## □ Principle

- ◆ Prevent illegal scan shift operations

## □ Solutions

- ◆ Test mode protection

- Scan protocol

- ◆ System mode protection

- Scan chain scrambling

- Scan enable tree protection

- Spy FFs

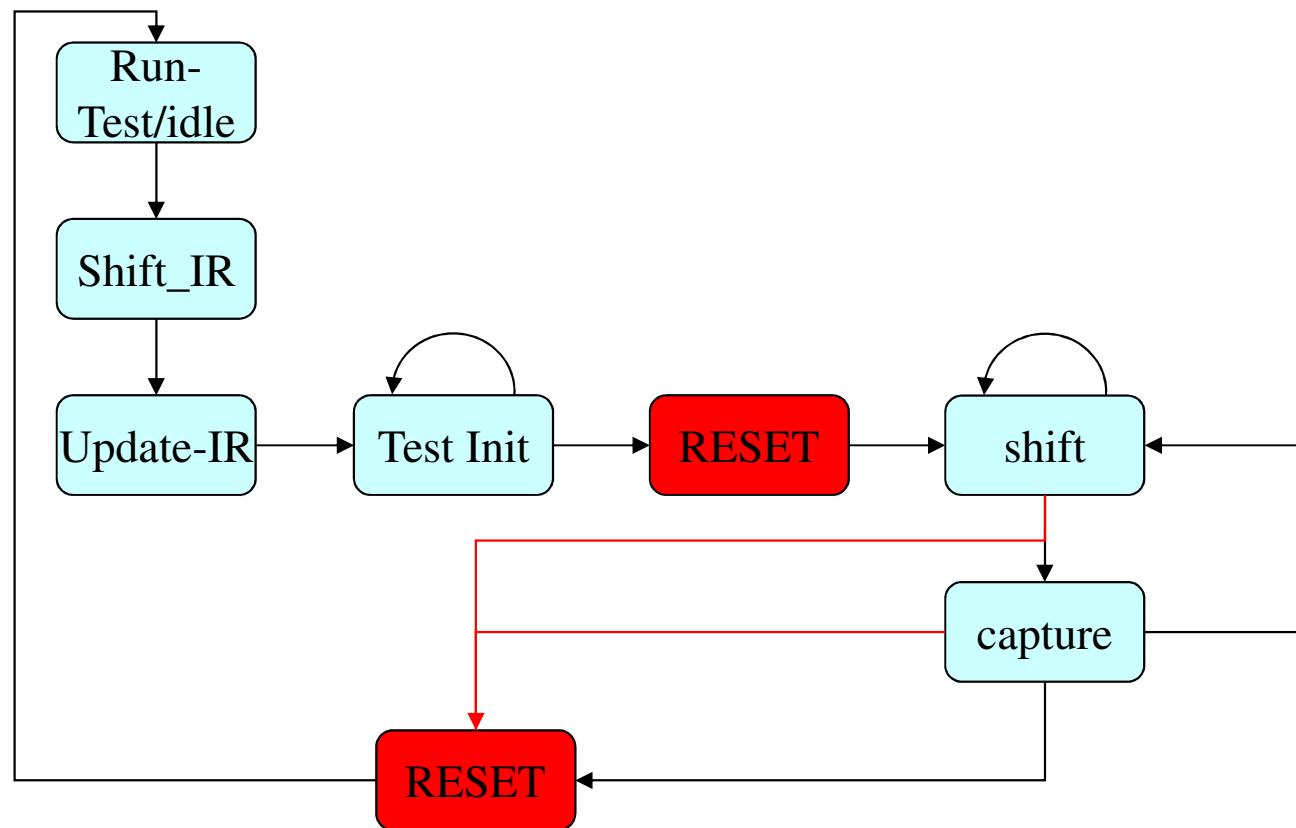
} protection against  
illegal usage of the test mode

} protection against  
scan chain probing attacks

# Test mode protection

## □ Secured Test Controller [He05]

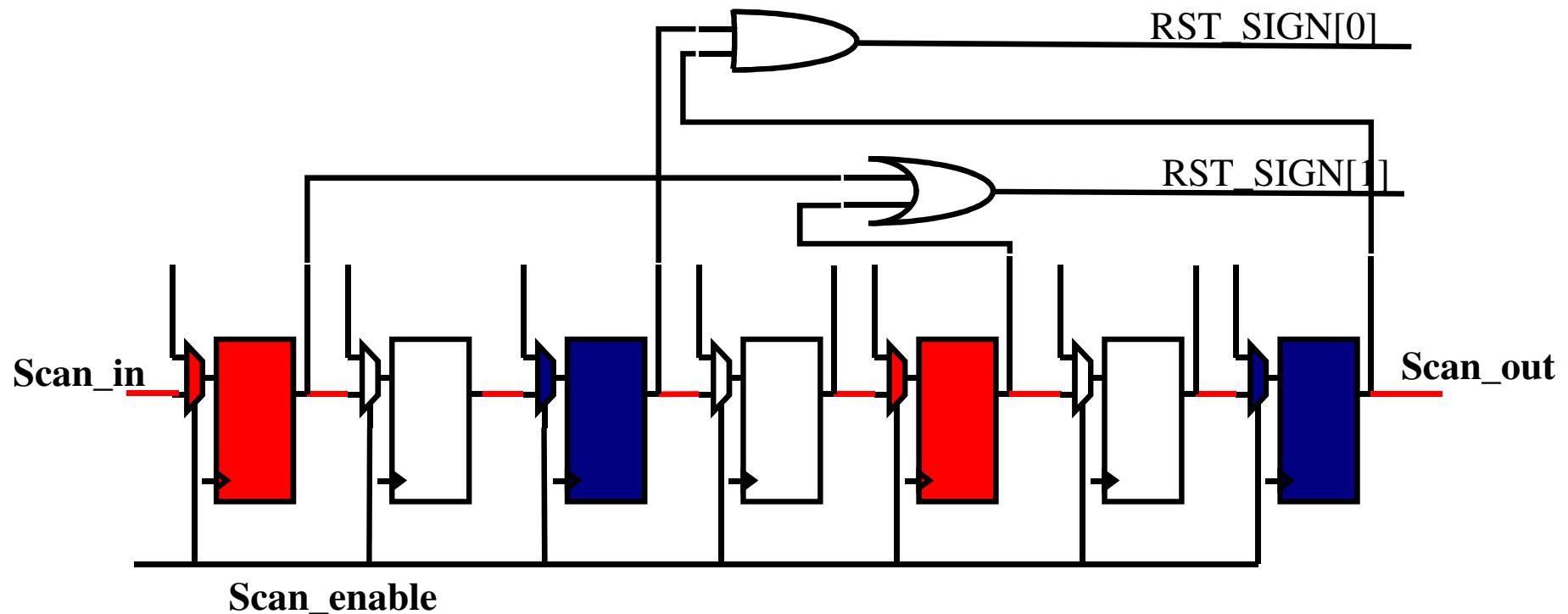
- ◆ Addition of two Reset states
- ◆ Secret data are erased before scan operations
- ◆ Scan data are erased before crypto-operations



# Test mode protection

## □ Reset Verification

- ◆ Static Verification (no scan chain activation)
- ◆ Verification done of a subset of FFs



# System mode protection

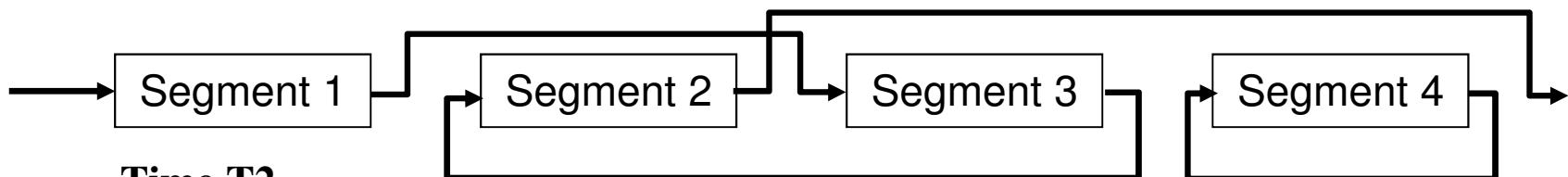
## □ Scrambling method [He04]

- ◆ Scan path with a prefixed segment organization during test mode

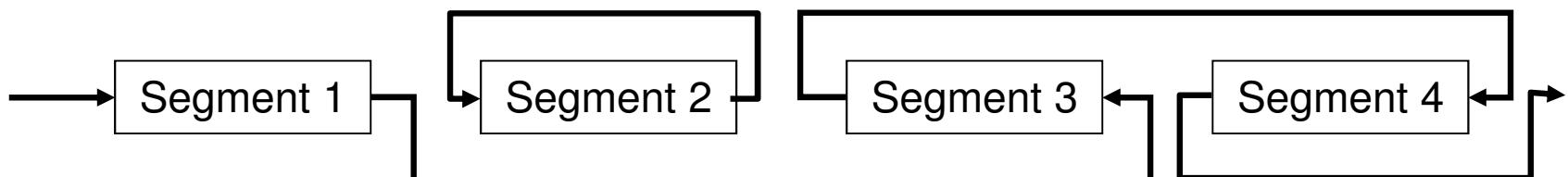


- ◆ Scan path with random segment organization if shift during system mode

— Time T1

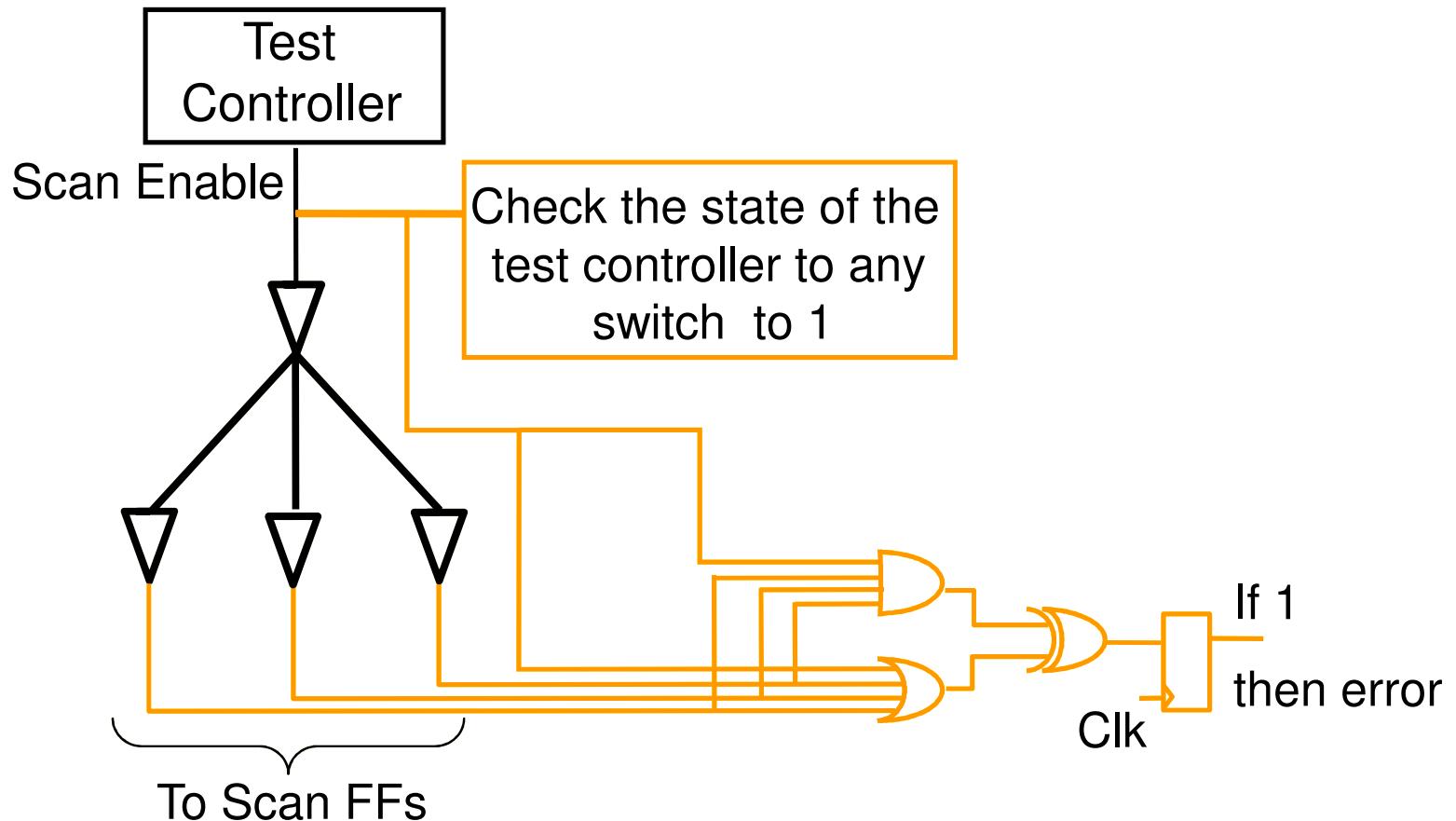


— Time T2



# System mode protection

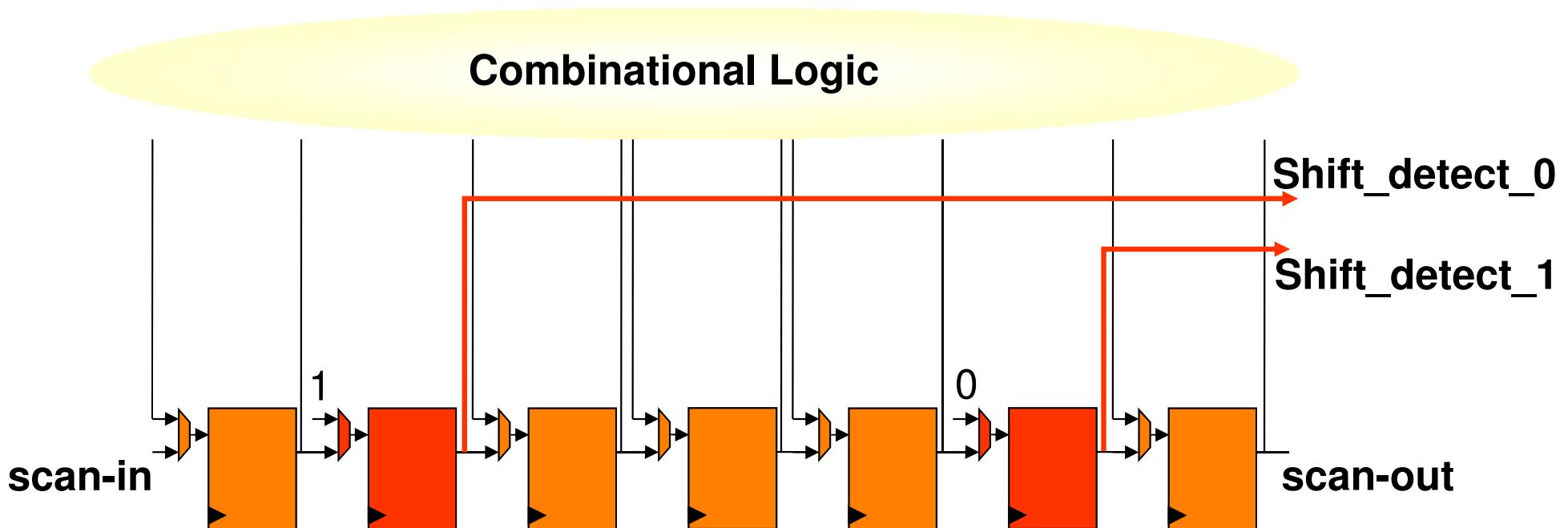
- Scan-Enable Tree Protection [He05]
  - ◆ Compare the scan enable signals at different locations



# System mode protection

## □ Spy Flip-Flops [He06]

- ◆ Include spy cells in the scan chain
- ◆ Control the spy cells to a constant value
- ◆ Observe the spy cells states



# System mode protections - Comparisons

---

## □ Case study:

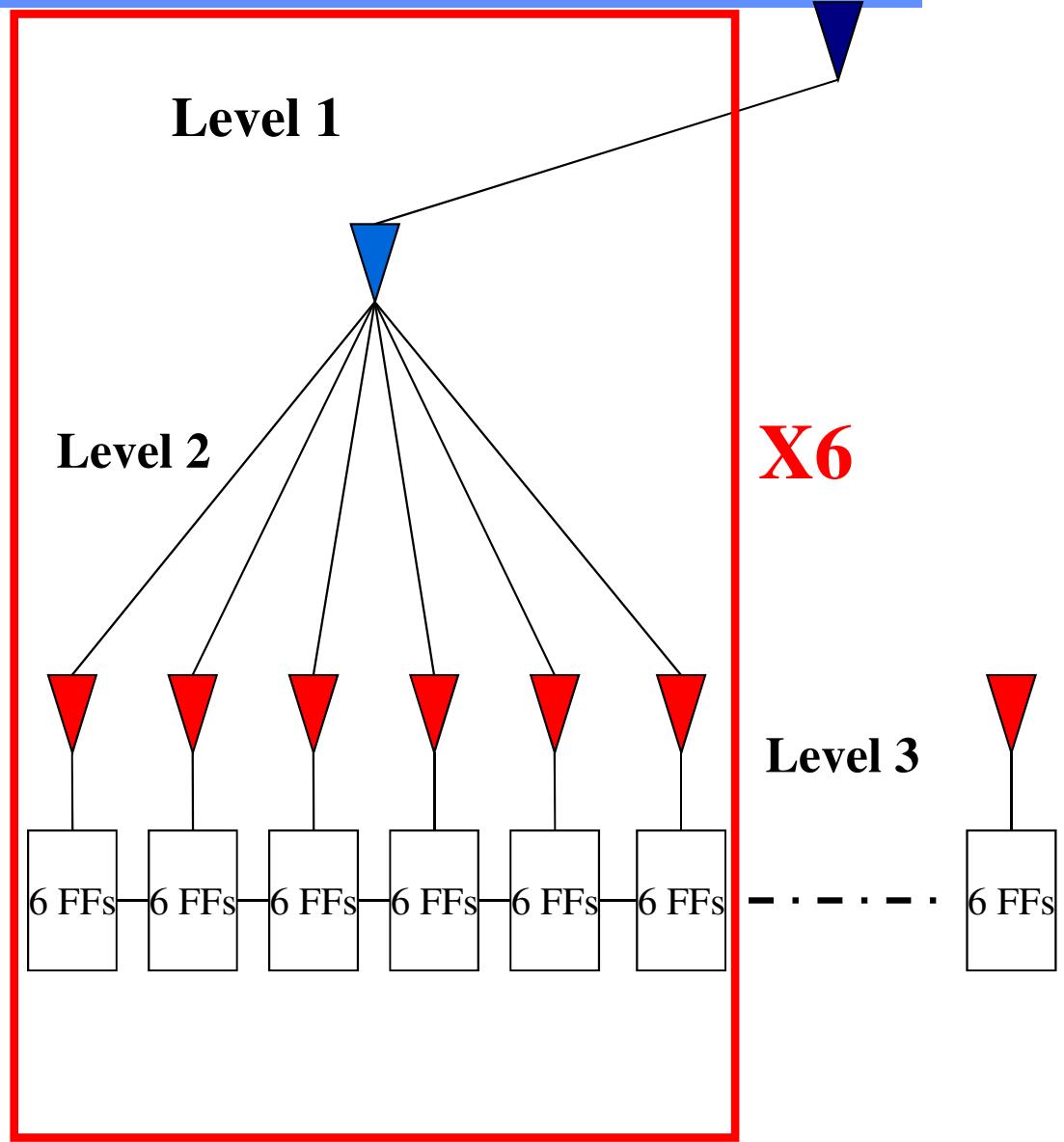
- ◆ DES crypto-processor
- ◆ Scan chain with 198 FFs
- ◆ Attack described in [Bo04]
- ◆ Secure test controller for the 3 architectures
- ◆ Several system mode protections
  - Architecture A (SEI):  
Scan\_enable integrity
  - Architecture B (Spy Cell):  
Shift detection flip-flops
  - Architecture C (Scrambling):  
Scan chain scrambling

## □ Criteria of comparison:

- ◆ Design integration
- ◆ Design reuse, Validation...
- ◆ Test cost
- ◆ Test time, test coverage...
- ◆ Design cost
- ◆ Area, Power, Timing closure...
- ◆ Security
- ◆ Efficiency

# System mode protection - Parameters

- System mode protection parameters depend on s-e tree bufferization
  - ◆ 3 level tree: 1 buffer drives 6 FFs
  - ◆ Parameters are chosen so that all brute attack on level 2 are overcome
  - ◆ At level 2 one buffer drives 36 FFs



# System mode protection - Architectures

---

## □ Architecture A

- ◆ SEI\_INT signal is connected to the 6 branches of level 2 of s-e tree.
- ◆ Connexion is made after synthesis (gate level modification)

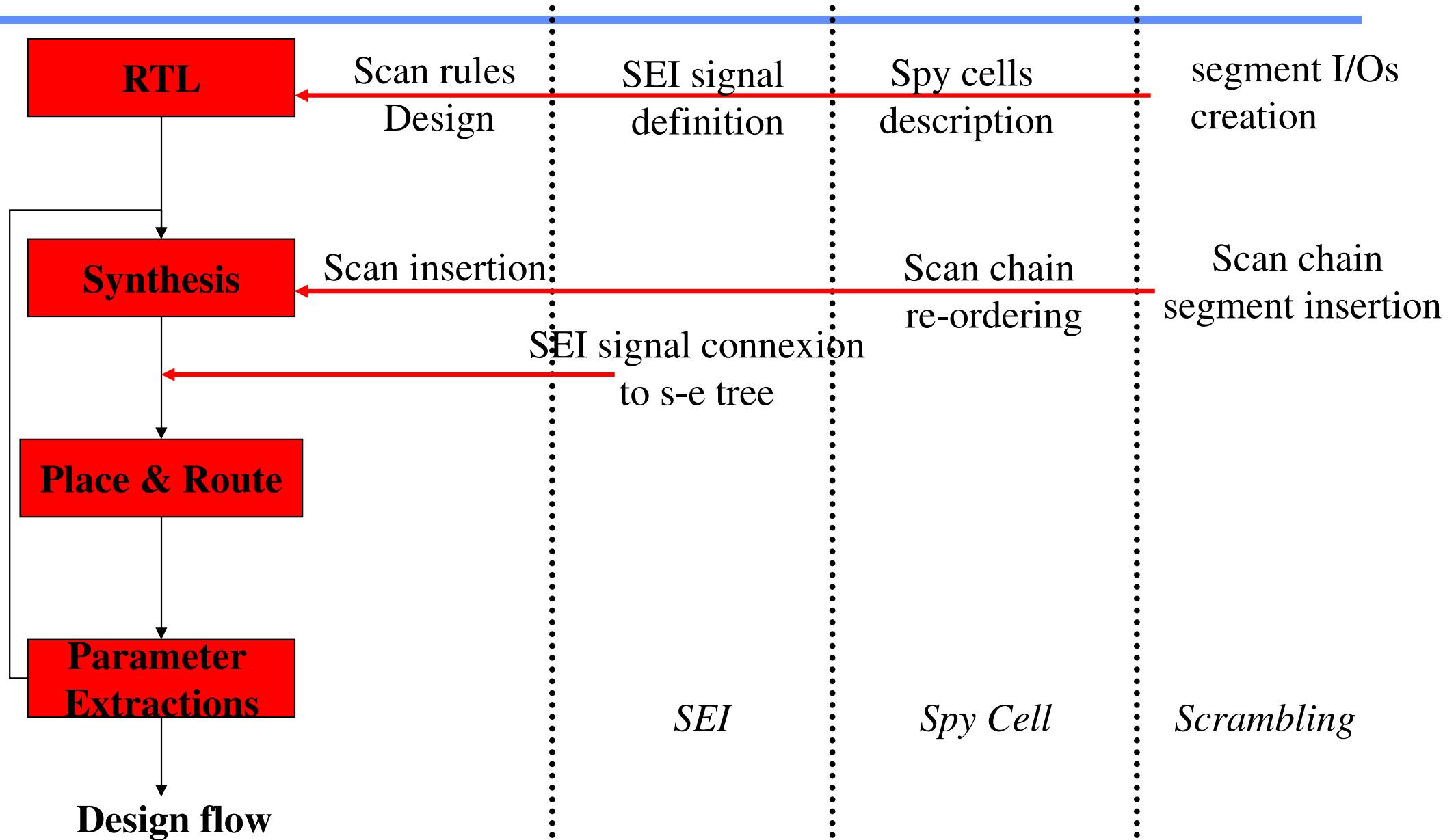
## □ Architecture B

- ◆ 6 spy flip-flops spread in the scan chain
- ◆ Spy flip-flops inserted in VHDL description of the IP
- ◆ Scan re-ordering to correctly spread them

## □ Architecture C

- ◆ Scan chain divided into 6 segments
- ◆ Random number generator is mandatory to provide alea to the scrambler

# Comparison results - integration



# Comparison results – test characteristics

---

□ Test Coverage: idem for the 3 architectures

□ Test Data:

SEI	Spy	Scrambling	Without security
+1.75%	+3.5%	+0%	62 patterns

□ Test Time:

SEI	Spy	Scrambling	Without security
+1.75%	+5.25%	+0%	12536 cycles

# Comparison results – design costs

---

## □ Area overhead

- ◆ IP: almost nul for SEI and Scrambling, very low for spy cell
- ◆ Control mechanism in test controller =>Area increase

Arch.	Area.
SEI	<b>1359</b>
Spy	<b>1269</b>
scrambling	<b>2560</b>
Without security	<b>1100</b>
Area of test controller	

## □ Power Consumption

- ◆ Almost nul for SEI ans Spy cell
- ◆ Sensible for scrambling (+7% during a DES ciphering)

# Comparison results – security

---

## □ Brute attack on scan chain

- ◆ Any attacks on level 2 are detected
- ◆ Immediate detection for SEI, after 2 clock cycles (average) for spy cell
- ◆ DES Attack of [Bo04] simulated with scrambling on, attack fails

## □ Solution robustness

- ◆ Architectures A and B rely on a reset operation after detection
- ◆ Scrambling is more robust

[Bo04] Bo et al. "Scan-based Side-Channel Attack on Dedicated Hardware Implementations on Data Encryption Standard", International Test Conference (ITC 2004), pp 339-344

# Comparison results – summary (circuit level)

		Scrambling	Scan enable	Spy cell
Insertion flow		RTL	RTL + place&route	RTL
Test	Test time	0%	1%	5%
Design	Area	0.2%	0.3%	1.8%
	power c.	7%	0%	0%
Security		+++	++	++

Architectures should be chosen according to application priority:

**Time to design: spy cells**

**Power aware application: SEI and spy cells**

**High Security: scrambling**

...

# Built in Self Test (BIST)

---

- Avoid scan-based testing
- Symmetric ciphers are based on diffusion and confusion principles
  - ◆ Diffusion: every input bit should influence several output bits
  - ◆ Confusion: relating input, key and output should be as complex as possible
- Large observability ...
  - ◆ Every input affects different outputs
- ... and controllability of the internal state
  - ◆ Every output depends on several inputs
- Block ciphers can be excellent pseudo-random generators and signature analyzers
- 100% fault coverage after a few encryption cycles (~240) and only 3% area overhead

Flottes et al. "AES-based BIST: Self-test, Test Pattern Generation and Signature Analysis", DELTA'08

# Implementation Attacks

---

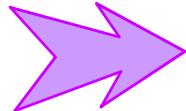
- Invasive attacks on the circuit
  - Side Channel Analysis
    - ◆ Time Analysis
    - ◆ Power Analysis
    - ◆ EM analysis
    - ◆ ...
  - Test structures
    - ◆ Scan chains,
    - ◆ ...
  - Which countermeasures?
-

# Perturbation-based attacks

---

## □ Modifying the nominal working conditions

- ◆ Temperature
- ◆ Power voltage
- ◆ Clock frequency (overclocking)
- ◆ ...



Indirect fault/error induction (e.g. timing errors)

Sensors (temperature/voltage) can be used to detect abnormal conditions

## □ Directly inducing faults/errors

- ◆ External electrical perturbations – glitches (power, clock, ...)

Sensors may be used on some critical lines + filtering (e.g. power line)

- ◆ Optical perturbations – (focused) white flash light, laser, UV, X Ray
- ◆ Electromagnetic sources, particles
- ◆ ...

# Overclocking vs. clock glitch

---

## □ Overclocking

- ◆ Faults may be induced at each cycle – no timing control
- ◆ No spatial control
- ◆ No identification of induced errors

## □ Clock glitch

- ◆ Faults induced only at the attacked cycles – some timing control
- ◆ Indirect spatial control and error identification due to
  - (1) the circuit structure: critical paths activated for a given internal state and given input values
  - (2) the attacked cycle, defining the internal state and the input values
  - (3) the clock shape (aspect ratio - duration of each level), to activate only some critical paths

**But both can be detected (by sensors) and glitches can be filtered (e.g. by PLLs)**

---

# Ideal fault induction technique ...

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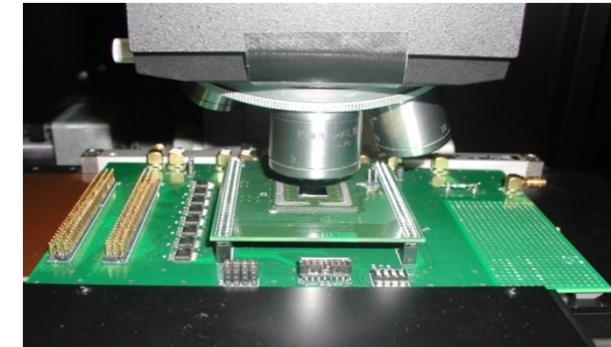
- Location control (x,y)
- Timing control (start, duration)
- Fault type control (stuck at, bit flip, etc.)
- Focalization control (number of faulty bits)
- Reproducibility
- Low cost
- Easy to develop and use

# Laser-based attack: preparation

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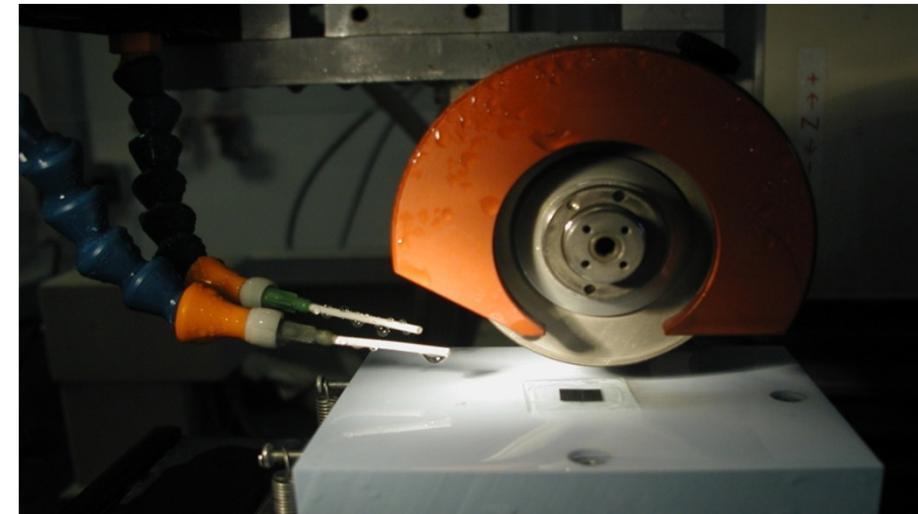
## □ Example on a Virtex chip (just unpackaged)

- ◆ Substrate @ 790  $\mu\text{m}$
- ◆  $\lambda \approx 900 \text{ nm}$
- ◆  $P \approx \text{some Watts}$
- ◆  $\varnothing = 40 \mu\text{m}$  and  $8 \mu\text{m}$
- ◆ Backside attack => No error



## □ Substrate thinning

- ◆ Die thinned by a mechanical process
- ◆ Residual thickness of  $30 \mu\text{m} \pm 1 \mu\text{m}$
- ◆ Successful backside attacks



# Fault-based attacks: basics

---

- In general a combination of perturbation and cryptanalysis  
=> fault creation + (potential) exploitation
- Example of the Bellcore attack on CRT-based RSA
  - ◆ One correct encryption result
  - ◆ One erroneous result from the same encryption (using e.g. a laser)
  - ◆ A simple GCD computation ... and the 1024-bit key is no more secret !
  - ◆ Few constraints on the injection: hit only one of the exponentiations
- Predictive robustness analysis: requires a good knowledge of injected faults
- FA (erroneous result directly exploited) – DFA (erroneous vs. correct result)

# Some insights into Bellcore attack – RSA & CRT

---

- Signing process of a message (assuming the whole message is used to sign)
  - ◆  $p, q$ : two large prime numbers,  $n = p * q$
  - ◆  $d, e$ : two numbers in  $\mathbb{Z}/n\mathbb{Z}$  relatively prime to  $n$  and verifying  $d * e = 1 \text{ mod } \Phi(n)$ , where  $\Phi(n) = (p-1)*(q-1)$  is Euler function
  - ◆ Signing  $s$  is computed by  $s = m^d \text{ mod } n$
  
- Modified computation with CRT algorithm
  - ◆  $d_p, d_q$  defined by  $d_p = d \text{ mod } (p-1)$  and  $d_q = d \text{ mod } q-1$
  - ◆  $s_p, s_q$  computed with  $s_p = m^{dp} \text{ mod } p$  and  $s_q = m^{dq} \text{ mod } q$
  - ◆  $a, b$  defined by  $a = 1 \text{ mod } p, a = 0 \text{ mod } q$  and  $b = 0 \text{ mod } p, b = 1 \text{ mod } q$
  - ◆ Final signing:  $s = s_p * a + s_q * b$

# Some insights into Bellcore attack – DFA

---

- One single fault/error induction => computation of  $s_p$  (respectively  $s_q$ ) is wrong
  - =>  $s_p'$  (respectively  $s_q'$ ) and wrong signature  $s'$  defined by  $s' = s_p'*a + s_q'*b$  (respectively  $s' = s_p'*a + s_q'*b$ )
- $\Delta = s - s' \Rightarrow \Delta = a*(s_p - s_p')$  (respectively  $\Delta = b*(s_q - s_q')$ )
- Given that  $a = 0 \text{ mod } q$  (respectively  $b = 0 \text{ mod } p$ ),  
 $\Delta$  is a multiple of  $q$  (respectively  $p$ ),  
and a simple calculation of a Greatest Common Divisor (GCD) will factor  $n$ :  
 $GCD(\Delta, n) = q$  (respectively  $GCD(\Delta, n) = p$ )

# Advanced Encryption Standard

## □ Symmetric block cipher

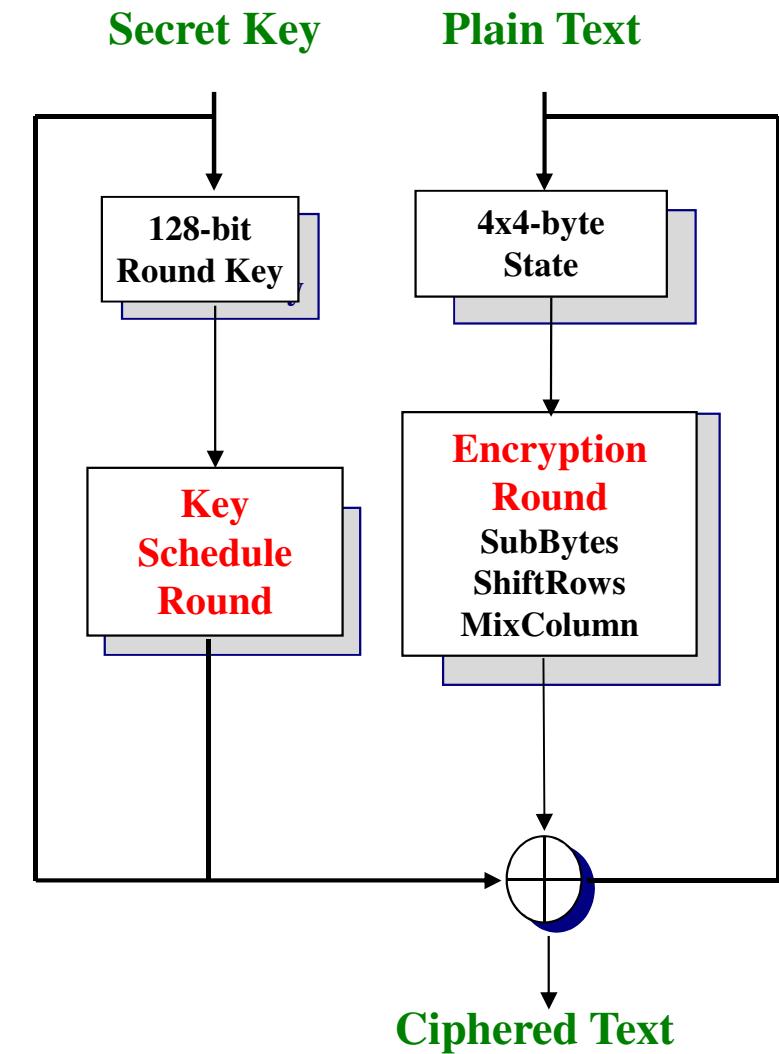
- ◆ Plain Text 128b,
- ◆ Secret Key 128/192/256b

## □ SPN cipher

- ◆ 10/12/14 rounds

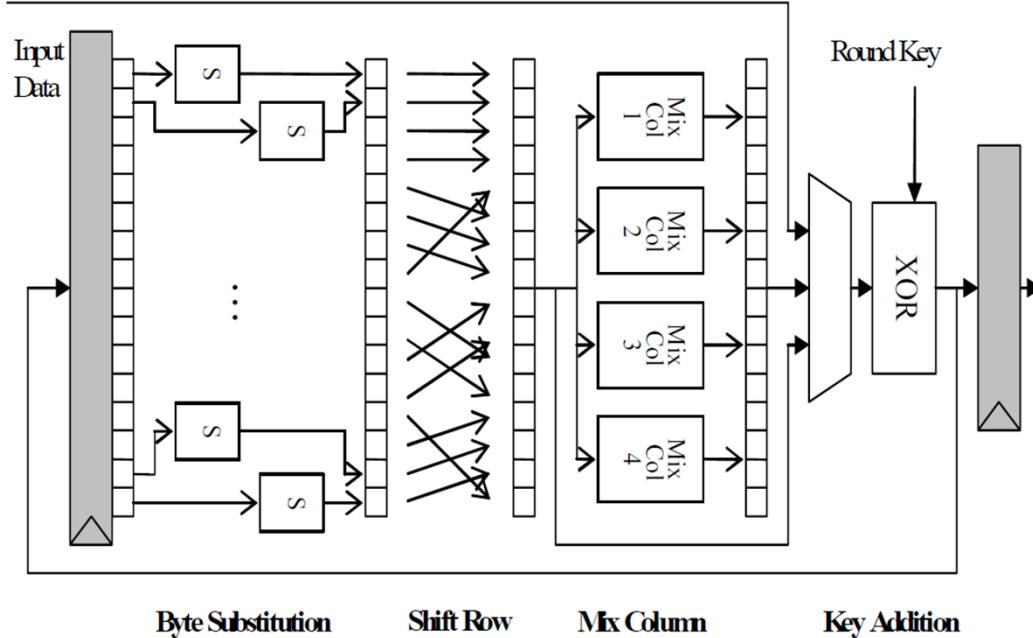
## □ Round operations

- ◆ SubBytes: nonlinear byte substitution
- ◆ ShiftRows: row-wise word rotation
- ◆ MixColumns: column-wise linear multiplication
- ◆ Key Addition: XOR with round subkey

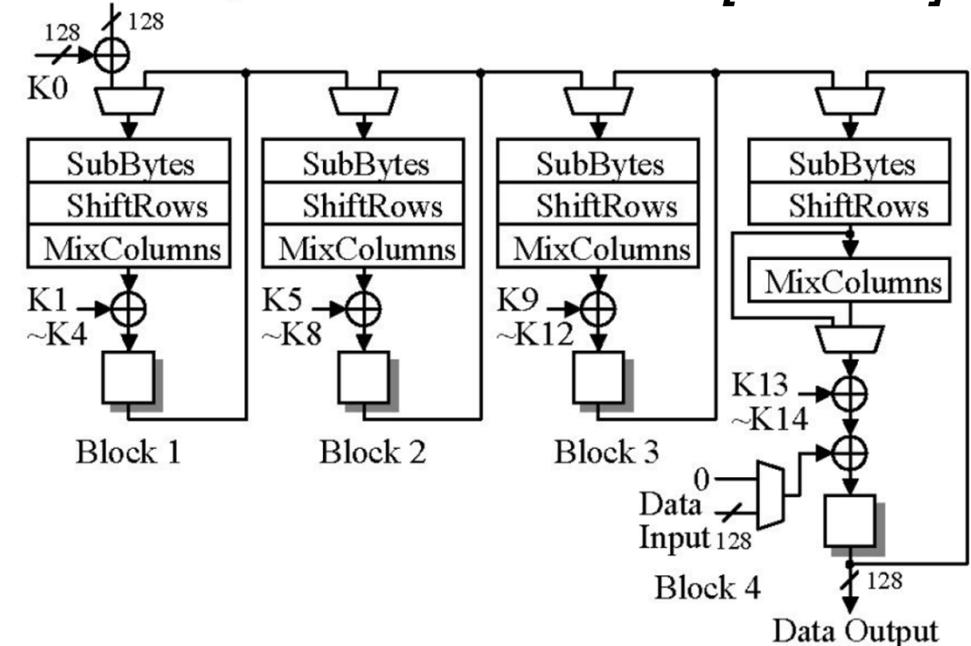


# AES Hardware Implementations

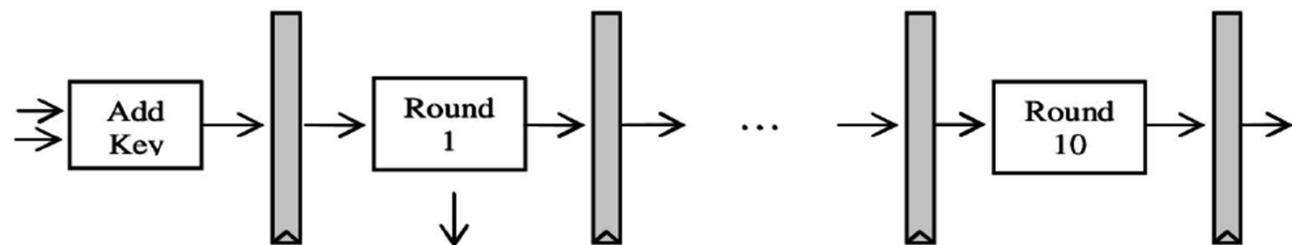
*Round-iterated designs [GLSVLSI05]*



*Partial round instantiation [ISCAS06]*

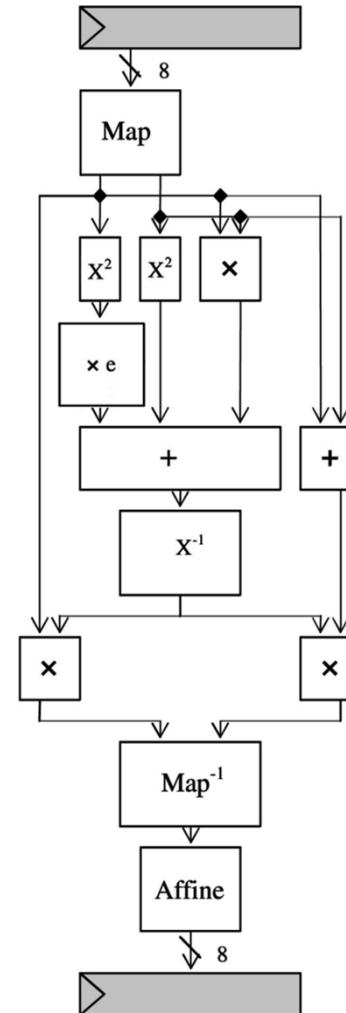


*Fully unrolled  
architectures  
[TC 55/4]*



# S-Box Implementations

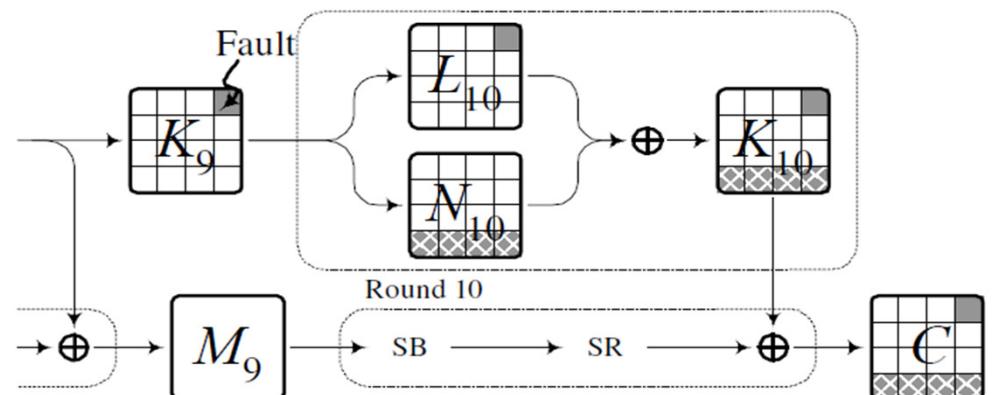
- Look-up table
  - ◆ 256B ROM
  - ◆ Combinational logic network
- Composite field
  - ◆ Mapping from  $\text{GF}(2^8)$  to  $\text{GF}(2^4)^2$
  - ◆ Smaller than LUT,
  - ◆ but slower
  - ◆ Several possible mappings
  - ◆ Easy pipeline



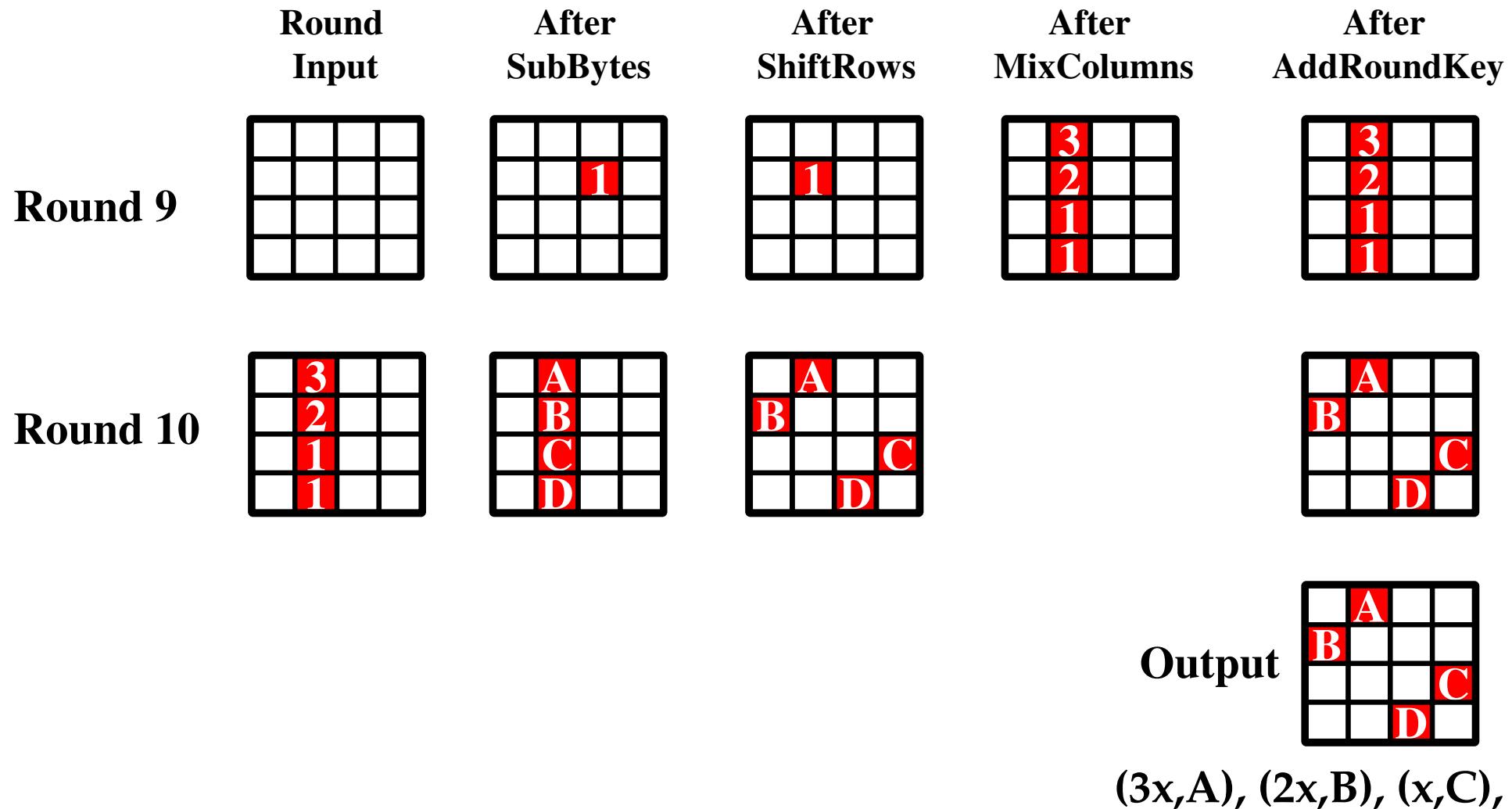
# Fault Attacks on AES

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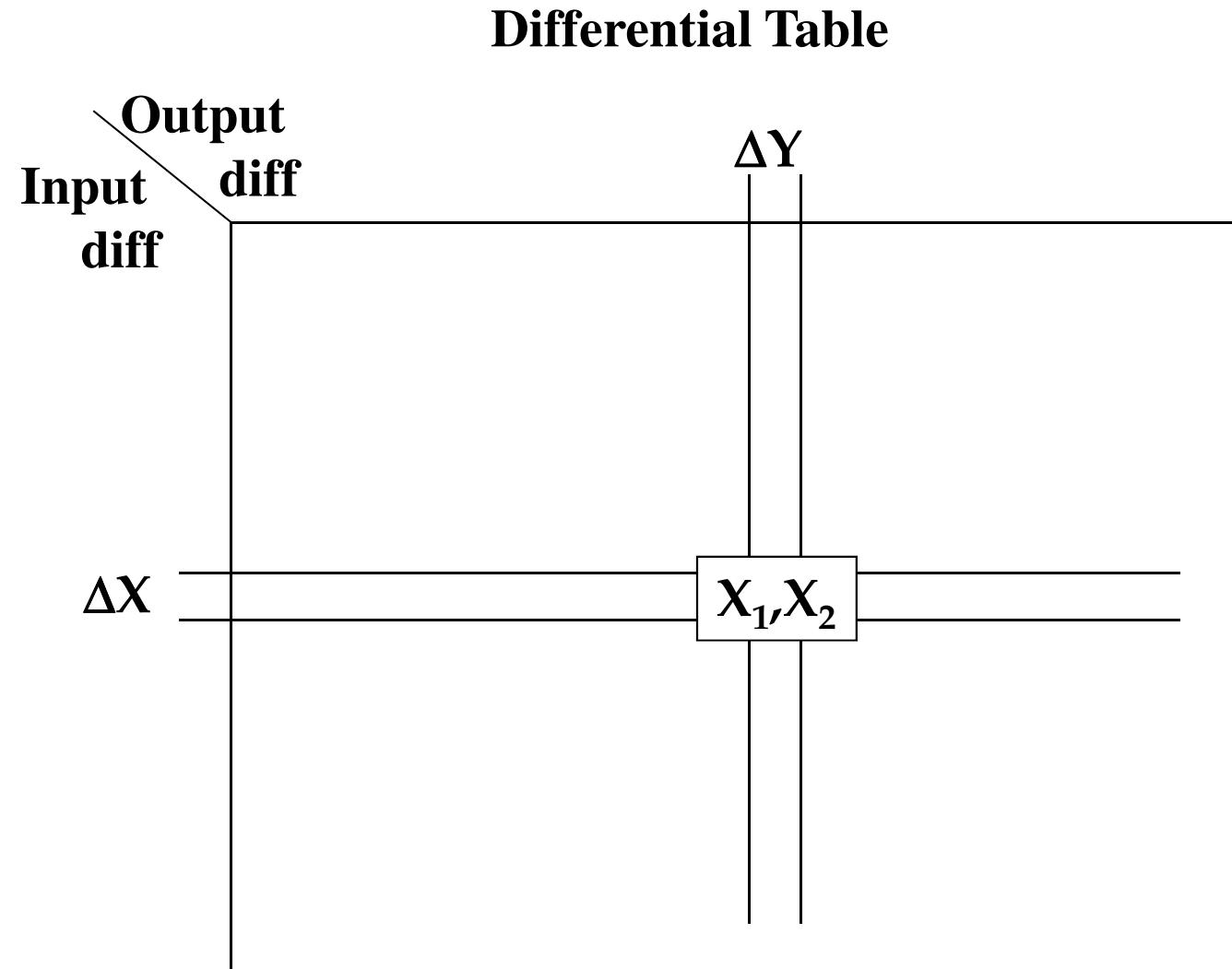
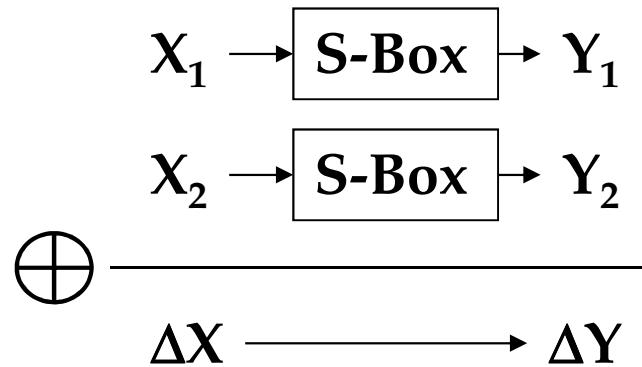
- C-safe error
  - ◆ Specific stuck-at injected at probabilistic location (byte-level)
  - ◆ Corrupted output → Error injected → Known value (~90k injections)
- Exploit S-Box & MixColumn scaling (Dusart et al, ACNS 2003)  
$$s(x) + s(x + c \cdot \varepsilon) = \varepsilon$$
  - ◆ 50 faulty ciphertexts to recover a full 128-bit key
- Injection in last round key
  - ◆ Chen and Yen
  - ◆ About 44 fault injections
- Differential Fault Analysis...
  - ◆ Piret & Quisquater, [CHES 2003], see next slides



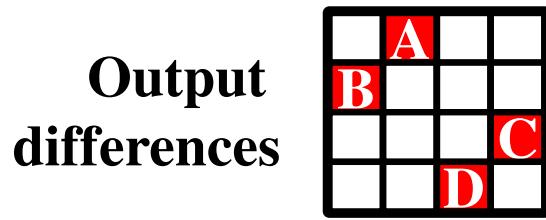
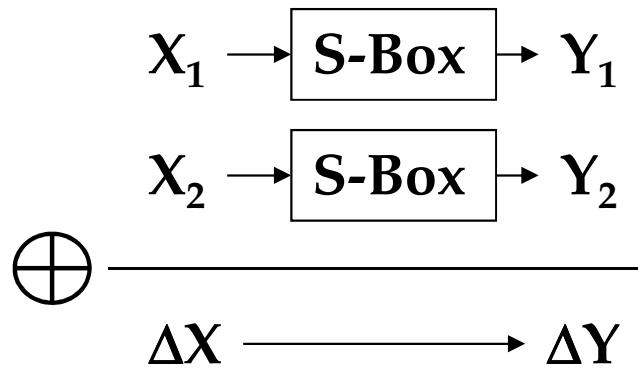
# DFA on AES – Fault Propagation



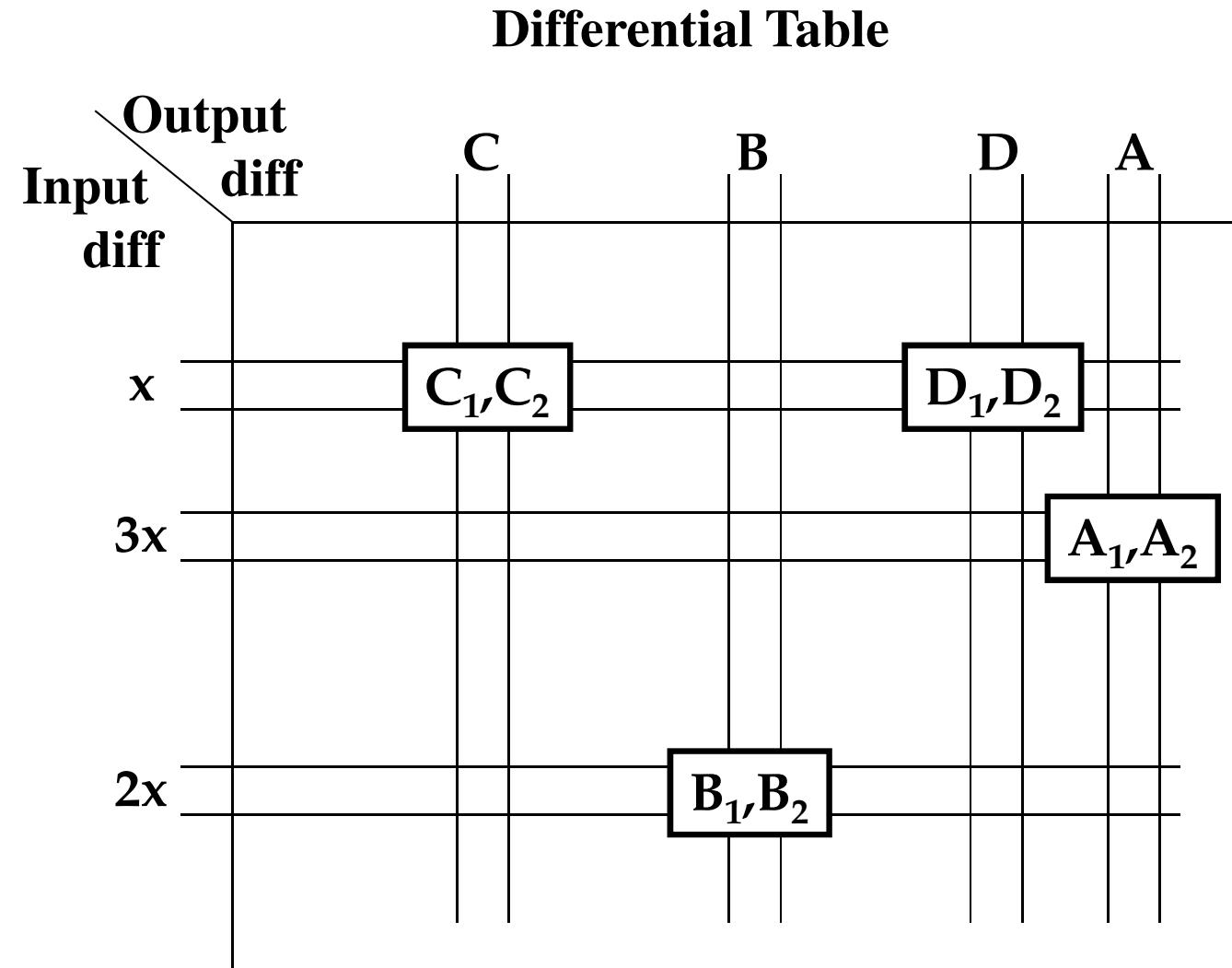
# DFA on AES – Differential Fault Space



# DFA on AES – Finding the Candidates

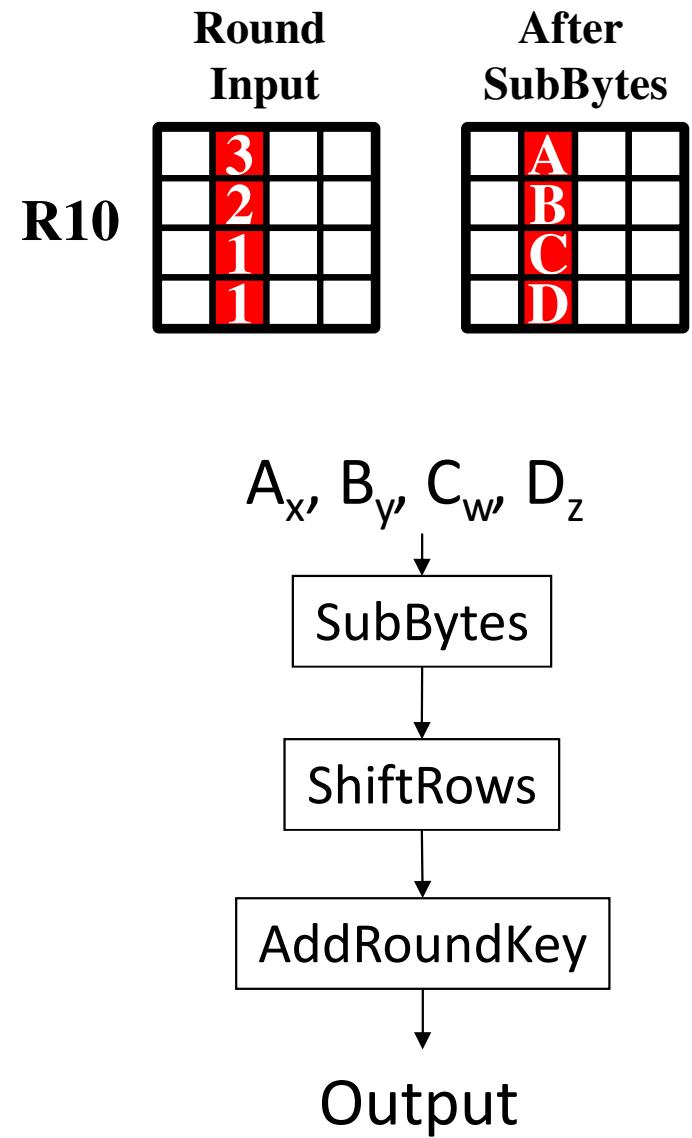


(3x,A), (2x,B), (x,C), (x,D)



# DFA on AES – Finding the Key

- For every possible error value
  - ◆ Add candidates to list
  - ◆ Discard on at least one empty cell
- Perform another fault injection
  - ◆ Build a second set of candidates
- Intersect the two lists
  - ◆ Shared values are last S-Box inputs
- Find last round key
- Compute reverse key schedule
  - ◆ and find the secret key!



# When is the attack successful ?

---

- When the fault is injected and the computation is corrupted ?

**Not yet !**

- When the erroneous result is obtained ?

**Yes !**

- So the erroneous result must

- ◆ Either be hidden (but may give useful indications to the hacker)
  - ◆ Or be corrected
  - ◆ ... Or the hacker may be misled (erroneous result replaced by another that cannot be exploited)
- 
- ◆ Note: delays induced by detection + recovery => information to the hacker

# Use of induced errors: DFA, but not only !

---

- Round reduction, RNG output forcing => easy cryptanalysis
- PIN counter corruption => unlimited trials
- Hardware or software protection bypass ...
- Behavior analysis ("safe-error" attack): if the attack effect is controlled, e.g. stuck-at-0 bit, a normal or abnormal behavior indicates detection or not of the attack, thus the initial value of the attacked bit (**even without output result to analyze**).
- Direct hacking: counter modifications (e.g. money in e-purse)

# Induced errors – logic point of view

---

- The error type does not really depend on the physical perturbation technique => soft errors
  - ◆ Direct induction in memory elements: SEUs/MBUs
  - ◆ Induction in combinatorial logic: SET then propagation
  - ◆ Final effect: erroneous bit(s) in register(s)
- Multiplicity depends on the source and fault location
- Error models assumed in published attack schemes are not always realistic (w.r.t. current fault induction techniques) – e.g. one single particular bit forced at a given value at a given cycle during the computation ...

# A special case: reconfigurable hardware

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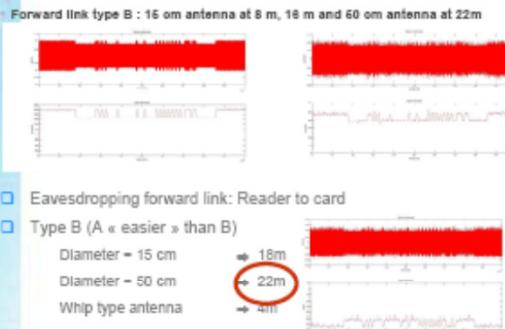
- In case of SRAM-based or Flash-based FPGA
- Specific possible attack by "downgrade", i.e. replacing the FPGA configuration by an older one (correctly encrypted and with correct integrity checks) => may maintain functionality but remove security "patches" correcting previously identified security holes
- Need for accurate versioning !!

# "New" concerns: wireless/long distance attacks

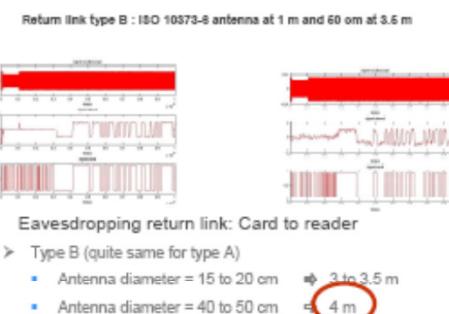


## Contactless: Une problématique spécifique

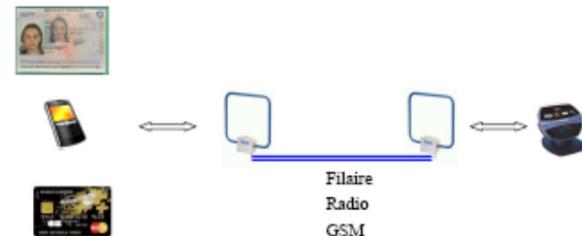
### Eavesdropping



### Eavesdropping



- Ecoute à distance
- Activation non contrôlée
- Relais



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Alain MERLE

| 20

**Other example: "long" distance recording of keyboard activations  
(creating carrier signal modulation)**



M2P SCCI



Hardware and Embedded Systems Security

# RFID and security

---

- Smart phones such as Apple's iPhone and Google's Android are significant (unprotected) targets ... (personal data, zombies ...)
- ... but not only:
  - ◆ One of the most common attacks on wireless networks is “war driving,” in which **hackers drive around a neighborhood, hunting for unsecured wireless nodes**. In the latest twist on war driving, **a security expert armed with a cheap RFID scanner and low-profile antenna managed to clone half a dozen electronic passports in an hour** while cruising around Fisherman’s Wharf in San Francisco.
  - ◆ The researcher who conducted this experiment asserts that the attempt at “war cloning” was successful because the type of RFID in the Homeland Security version of a passport emits a real radio signal, which could conceivably be tracked from a couple of miles away. Although no criminal hacks of passports or e-licenses have been detected to date, this insecure technology poses a strong risk for identity theft and invasion of privacy.[7]
  - ◆ 7. Kelly Jackson Higgins, “Drive-By ‘War Cloning’ Attack Hacks Electronic Passports, Driver’s Licenses,” Dark Reading, February 2, 2009,  
[www.darkreading.com/security/privacy/showArticle.jhtml?articleID=213000321](http://www.darkreading.com/security/privacy/showArticle.jhtml?articleID=213000321)

<http://embedded-computing.com/current-trends-cyber-attacks-mobile-embedded-systems>

---

# Protecting Secure Circuits

- "Classical" protection styles (but adapted), especially against faults/errors

- Specific protections (active)

- ◆ Sensors (voltage levels, glitch, light, temperature, ...)
- ◆ On-chip encryption of data
- ◆ ...

Multi-level:  
- Software  
- OS  
- Hardware

- Specific protections (design methodology)

- ◆ Internal clock generation (problem with testing requirements !)
- ◆ Restrictions in design styles (dynamic logic, ...)
- ◆ Shielding
- ◆ Memory scrambling
- ◆ ...

# Towards a dependable system (w.r.t. errors)

- Many approaches/techniques, a few major categories
    - ◆ Spatial redundancy, replication (massive redundancy)
    - ◆ Information redundancy (coding)
    - ◆ Timing redundancy
  - +
    - ◆ Assertion-based on-line checking
    - ◆ Control-flow checking
    - ◆ etc. ...
- 
- Hardware  
and/or  
software
- Choice depends on application constraints (overheads, ...) and on dependability specification
    - ◆ Detection only
    - ◆ Tolerance by detection + recovery (retry, checkpointing, ...)
    - ◆ Tolerance by masking (voting) ... taking care of fault accumulation ...

# Multi-level counter-measures (FA)

---

Level	Counter-measures against FA (examples)
Gate (transistor level structure)	Various sensors, robust structures against SEUs
P&R	Coupling reduction
RTL	<b>Error detection codes: limitations</b> , timing redundancy
Algorithm	Precautions, e.g. CRT use for RSA
Architecture	Function fusions to mask intermediate results
System	Depends on security policy: memory erasing, process or OS-level error recovery

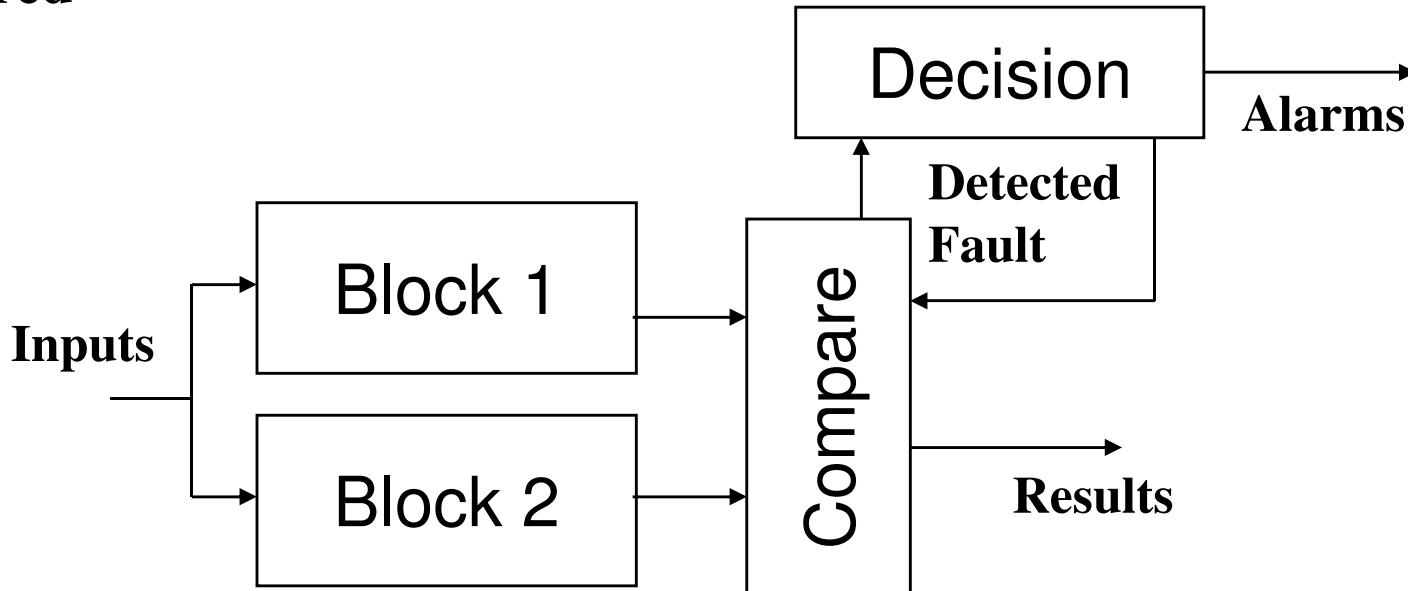
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# Hardware Redundancy

# Basic Redundancy

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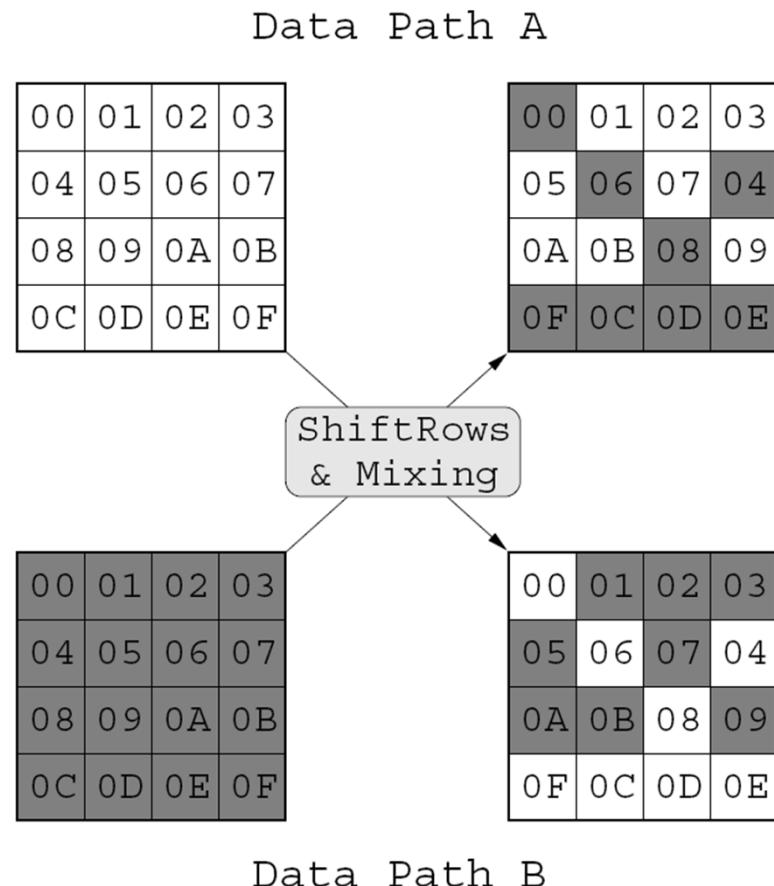
- Multiple [different] functional blocks are implemented and outputs are compared



- ☺ Effective against transient or destructive faults
- ☹ Overhead >100% in area and power
- ☹ Even greater correlation between data and power consumption
  - ◆ Easier DPA !!!

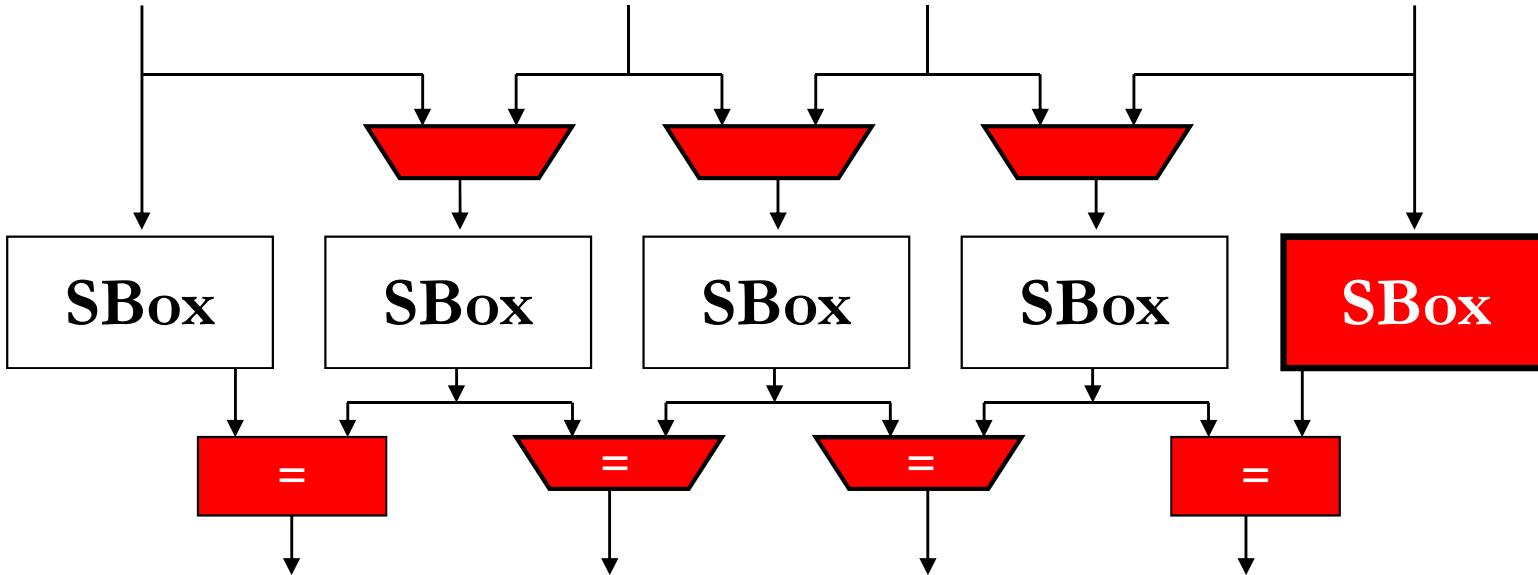
# Redundancy and Diffusion

- Hardware redundancy for fault detection
- Data contamination to avoid fault exploitation
- Same advantages and same problems than simple duplication
  - ◆ Greater security if final validation step is compromised



[Joye, Manet, Rigaud. IET Inf Sec 2007]

# Partial Redundancy (1/2)



- Dynamic comparison with additional component
- ☺ Limited overhead and good detection capability of permanent faults
- ☹ Limited detection capability of transient faults

[Di Natale, Flottes, Rouzeyre. WDSN 2007]

# Partial Redundancy (2/2)

---

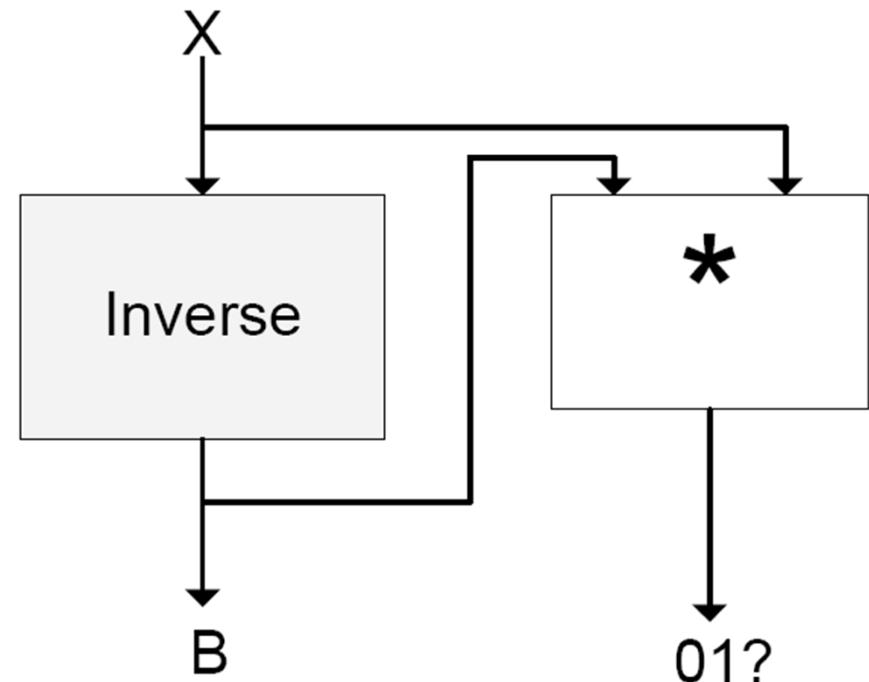
- AES S-Box:

- ◆ Most complex and expensive component
- ◆ Non linear
- ◆ Based on computation of the multiplicative inverse

- Detection based on the computation of the multiplicative inverse

- ◆ Add a GF multiplier
- ◆ Configurable width

- (?) Coverage limited to S-Box
- (?) Coverage dependent on redundancy and overhead
- (?) Linear code required for the rest of the circuit



[Kulikowski, Karpovsky, Taubin. CARDIS 2004]

---

# Hardware Redundancy - Summary

---

- Two (or more) independent functional blocks are used and the outputs are compared
  - ◆ Protections can be limited to a specific unit
  - ◆ Data paths can be mixed
- Effective against transient and permanent faults
- **Very expensive in terms of area**
- No throughput reduction
- Particularly vulnerable to side-channel analysis

---

# Information Redundancy

# Using error detecting codes (EDCs)

---

- May be an answer to fault-based attack detection, before security policy activation
- Choice of code: efficiency (w.r.t. a characterized threat) vs. overheads
- Algorithm impact: codes may be or not well suited (complexity of code prediction vs. number of checkers)
  - ◆ Identify the common components in (symmetric) block ciphers
  - ◆ Associate EDCs to data block and develop a code prediction rule for (possibly) each operation
  - ◆ Evaluate the suitability of a code to the whole cipher (i.e., overhead and error coverage)

A few following slides from L. Breveglieri, I. Koren, P. Maistri, "Detecting Faults in Integer and Finite Field Arithmetic Operations for Cryptography", FDTC: workshop on Fault Diagnosis and Tolerance in Cryptography, Supplemental Volume Proc. of the Conference on Dependable Systems and Networks, pages 361-367, June 2004

---

# Why prediction rules ?

---

- High coverage with low-order errors
  - ◆ Codes often provide 100% coverage of single bit errors
- With high-order errors, coverage depends on redundancy
  - ◆ Output code and data match randomly
- Expected hardware overhead smaller than duplication
  - ◆ EDCs need a code *generator*, a (single) *comparator* and *propagation units* implementing the prediction rules
- EDCs are cheaper when simple prediction rules are available for the whole encryption process
  - ◆ The check bits are generated at the beginning and validated at the end of the process
  - ◆ Checkpoint frequency can be increased for higher coverage

# Operations in symmetric ciphers

Ciphers	XOR	AND,OR	+, -	$\times$	Sbox	Rot	Shift	Perm	$\times \text{ mod } G(x)$
Camellia	✓	✓			✓	✓		✓	
DES	✓				✓			✓	
IDEA	✓			✓				✓	
MARS	✓			✓	✓	✓		✓	
RC5	✓			✓		✓			
RC6	✓			✓		✓		✓	
Rijndael	✓				✓			✓	✓
Serpent	✓				✓	✓	✓		
Twofish	✓			✓	✓	✓		✓	✓

- XOR: Every cipher
- AND, OR: Camellia only
- +: often used
  - ◆ -: in encryption, only MARS
- $\times$ : slow and area-consuming
  - ◆ IDEA uses uncommon modulus

- Rotations: even data-dependent
- Shift: Serpent only
- Permutation: provides confusion
- Polynomial  $x$ : Rijndael and TwoFish, over  $GF(2^8)$
- S-Box: non-linear

# EDC granularity

---

- Symmetric ciphers operate on different word size (8, 16, 32 bits)
- Code granularity should not be larger than operand size
  - ◆ The code should be validated and regenerated with each operation (e.g. substitution tables)
- Finer code adds further complexity and overhead
  - ◆ Detection rate improves
  - ◆ Prediction rule may become more complex

# Matching EDCs to operations (1/3)

---

- Parity is more suited to logical operations; the prediction rules are...
  - ◆ eXclusive OR: ...the XOR of the input parities
  - ◆ Rotation: ...parity is unchanged, if the code is at the same level of the operation
  - ◆ Shifting: ...must consider bits leaving and entering the word
  - ◆ Polynomial multiplication: ...if defined over  $GF(2^n)$ , it is easily predictable when one of the operands is known a priori
  - ◆ Data-dependent operations (RC5, RC6) are obviously more complex
- Addition and multiplication must consider all the carries that are required to compute the result

# Matching EDCs to operations (2/3)

---

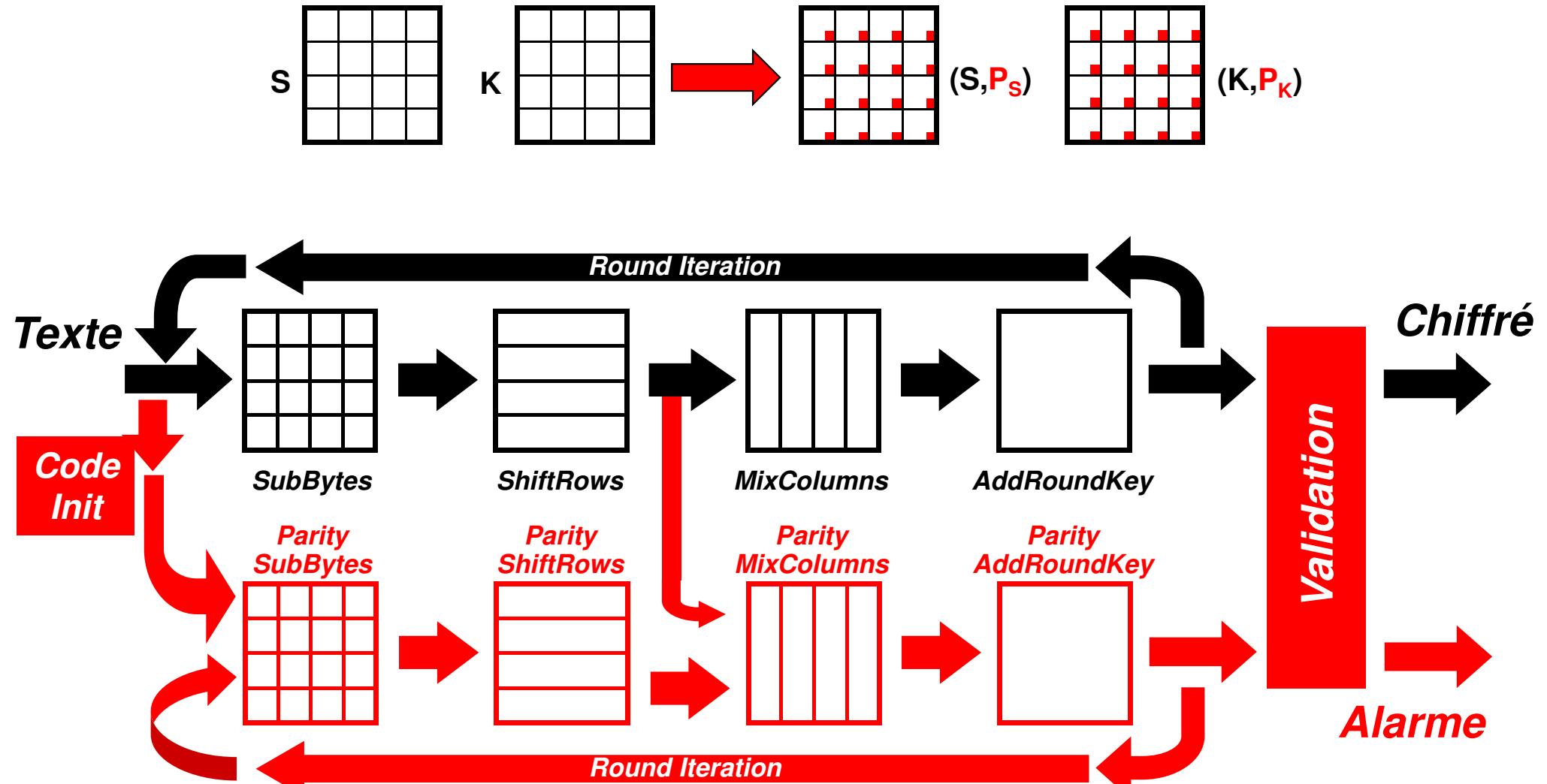
- Residues are more suited to arithmetic operations; the prediction rules are...
  - ◆ Addition: ...the sum of the input residues
    - but overflow needs correction!
  - ◆ Multiplication: ...the product of the input residues
    - but most significant (and neglected) bits need a corrective term
  - ◆ Shifting: ...it can be seen as a multiplication by a power of 2
  - ◆ eXclusive OR: ...the sum of the input residues, but a correction term is needed
- Prediction of the code after polynomial multiplication over  $GF(2^n)$  is expensive

# Matching EDCs to operations (3/3)

---

- Some operations are not suited to parity codes:
  - ◆ Logical AND and OR: prediction is much more expensive than duplication
  - ◆ Validate the code, protect by duplication and generate the code from scratch
  
- Some operations MAY BE suited both to residue and parity:
  - ◆ Substitution boxes: the output code is stored together with the result (if ROM-based or constant table implementation); the input code is used for implicit validation
  - ◆ Address protection by concatenating check bits introduces a large overhead (1 additional bit doubles the table size)
    - Use custom address decoding unit to reduce the area overhead

# An example: Multiple parity for AES



[Bertoni et al. TC 2003]

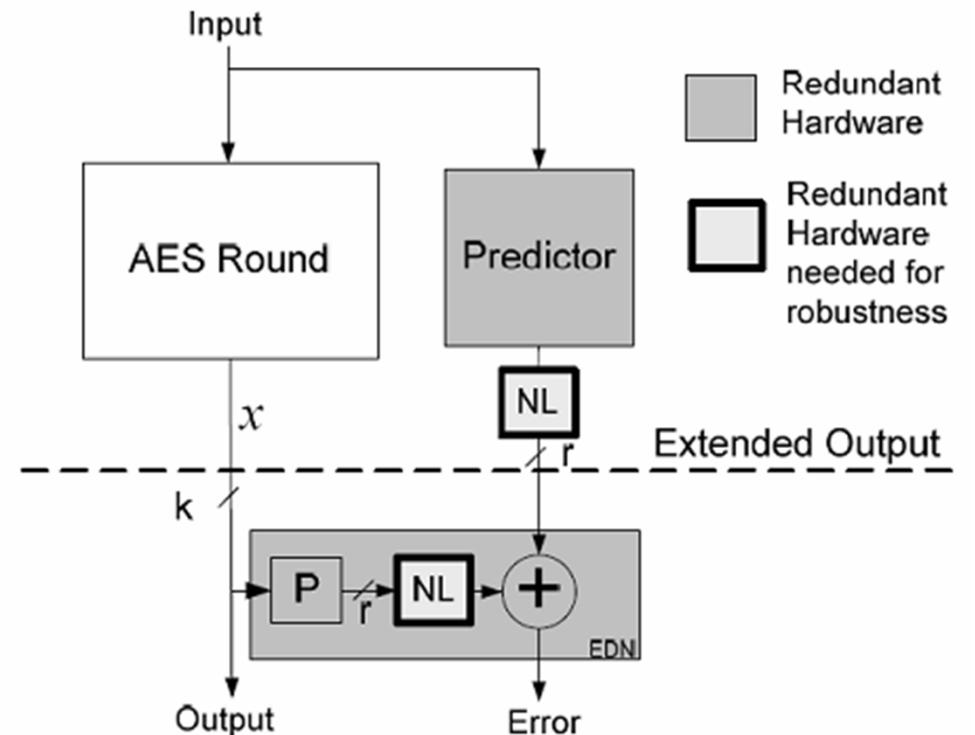
# Choosing the proper EDC

<b>Cipher</b>	$\oplus$	$\wedge, \vee$	$+, -$	$\times$	<b>Sbox</b>	<b>Rot</b>	<b>Sh</b>	<b>Perm</b>	$\times \text{ mod } G(x)$
Camellia	✓	✓			✓	✓		✓	
DES	✓				✓			✓	
IDEA	✓		✓	✓				✓	
MARS	✓		✓	✓	✓	✓		✓	
RC5	✓		✓			✓		✓	
RC6	✓		✓	✓		✓		✓	
Rijndael	✓				✓			✓	✓
Serpent	✓				✓	✓	✓		
Twofish	✓		✓		✓	✓		✓	✓

<b>Cipher</b>	<b>Suggested code</b>	<b>Cipher</b>	<b>Suggested code</b>
AES	Parity, per byte	RC5	Parity or residue
DES	Parity	RC6	Residue
IDEA	Residue, but expensive	Serpent	Parity, per byte
MARS	Residue, but expensive	Twofish	Parity, per byte

# Nonlinear Codes

- Detection by linear codes depends only on error value
  - ◆ Potential vulnerability
- Detection by **nonlinear** codes depends on error and **data** values
  - ◆ More uniform detection rate
  - ◆ Attacker cannot choose a specific error value to inject
- ☺ Very high detection rate
- ☺ Very difficult injection of controlled faults
- ☹ Significant overhead (+70%)



# Theme Variations

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Authors	Publication	Notes
Yen, Wu	TC 2006	AES data path protected by CRC redundancy code
Kermani, Masoleh	DFT 2006	Parity prediction of S-box with GF decomposition
Cota et al.	ISCAS 2008	Hamming and Reed Solomon codes
Di Natale et al.	IOLTS 2007	Parity prediction of S-box taking into account both input and output
Kermani, Masoleh	CHES 2008	Parity prediction of S-box with decomposition with normal basis
...		

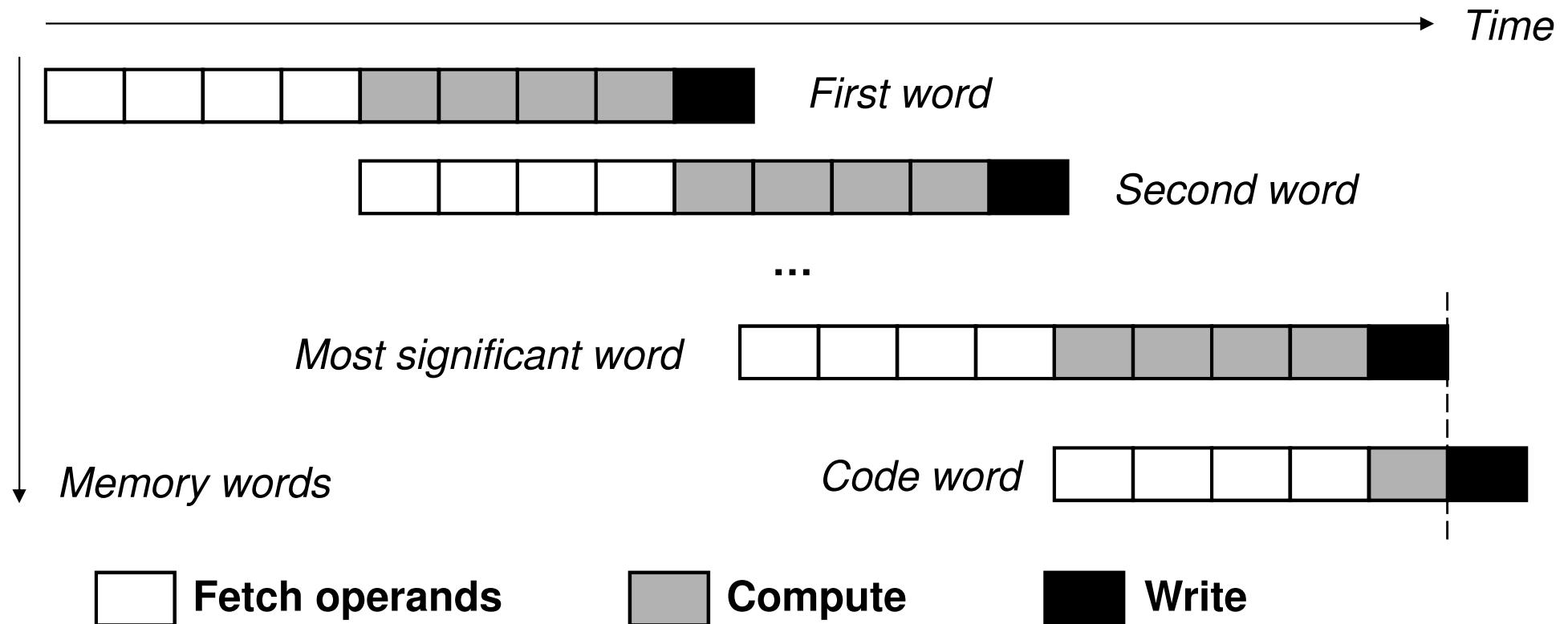
# Public-Key Cryptosystems

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- Fault attacks may affect also asymmetric cryptosystems
  - ◆ CRT-RSA can be broken with just one fault
- Verification of encrypted message needs a new encryption process (exponentiation)
  - ◆ Too slow
  - ◆ It must be done within the chip: no software solution, because the faulty output cannot be made available!
- Asymmetric systems are based on modular arithmetic
  - ◆ Computation errors can be easily identified by means of EDCs
  - ◆ Nonlinear codes can also be used [Gaubatz et al., FDTC 2006]

# An example: RSA

- *Detection and correction of transient errors in a hardware implementation of the RSA cryptosystem [...] can be done efficiently and reliably with acceptable time and area costs equivalent to an increase in the size of the modulus by one digit or less [C. Walter, CHES 2000]*



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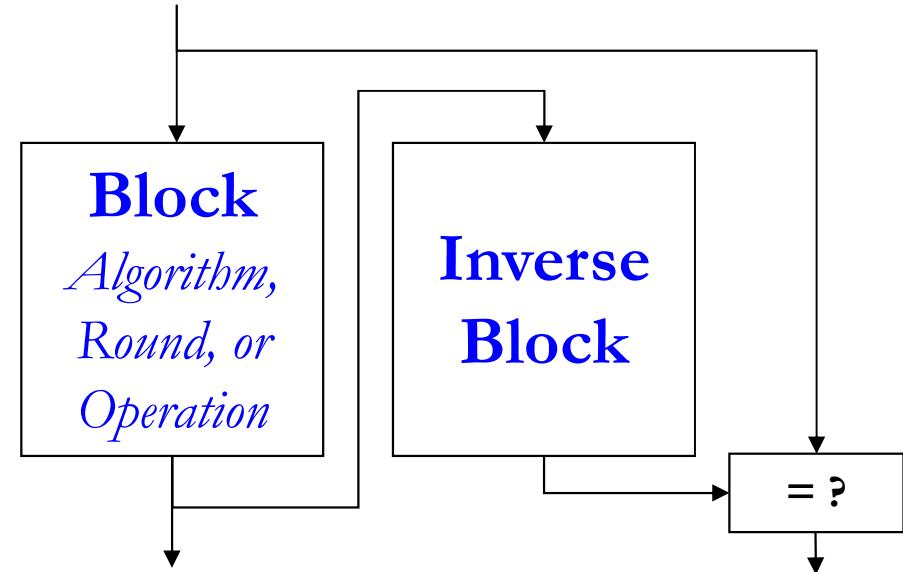
# Timing Redundancy

# Inverse Computation

---

- Based on the different paths for encryption or decryption
- 3 levels of granularity: algorithm, round, operation
  - ◆ Overhead, throughput reduction and detection latency
- Inverse function is computed and the result is compared to the initial value

- (⌚) Decryption needed
- (😊) Transient and permanent faults
- (😢) Latency and overheads



[Karri et al. DAC 2001]

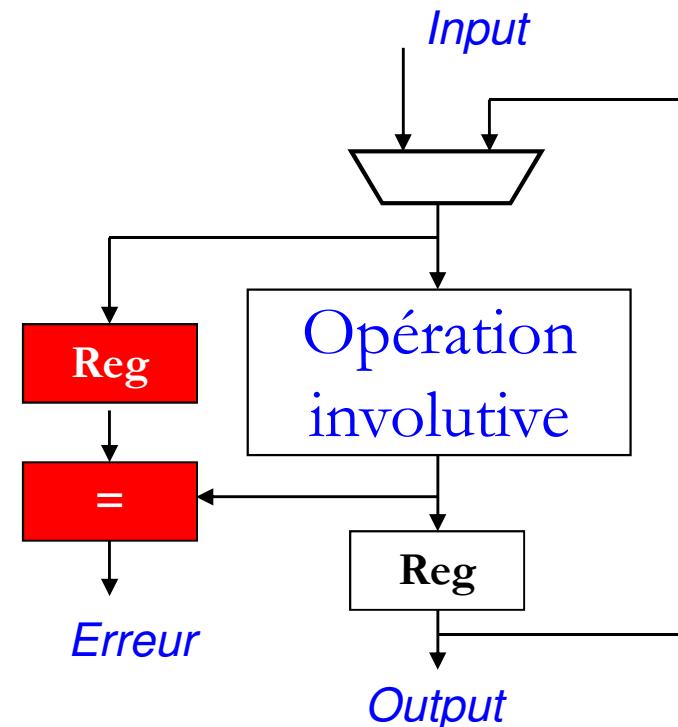
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# Involution Ciphers

- Involution cipher: encryption and decryption functions are the same
- Same idea than inverse computation...

- ☺ Decryption logic is already there
- ☺ Transient and permanent faults
- ☹ 100 % computation time overhead

It may be improved by pipelining...

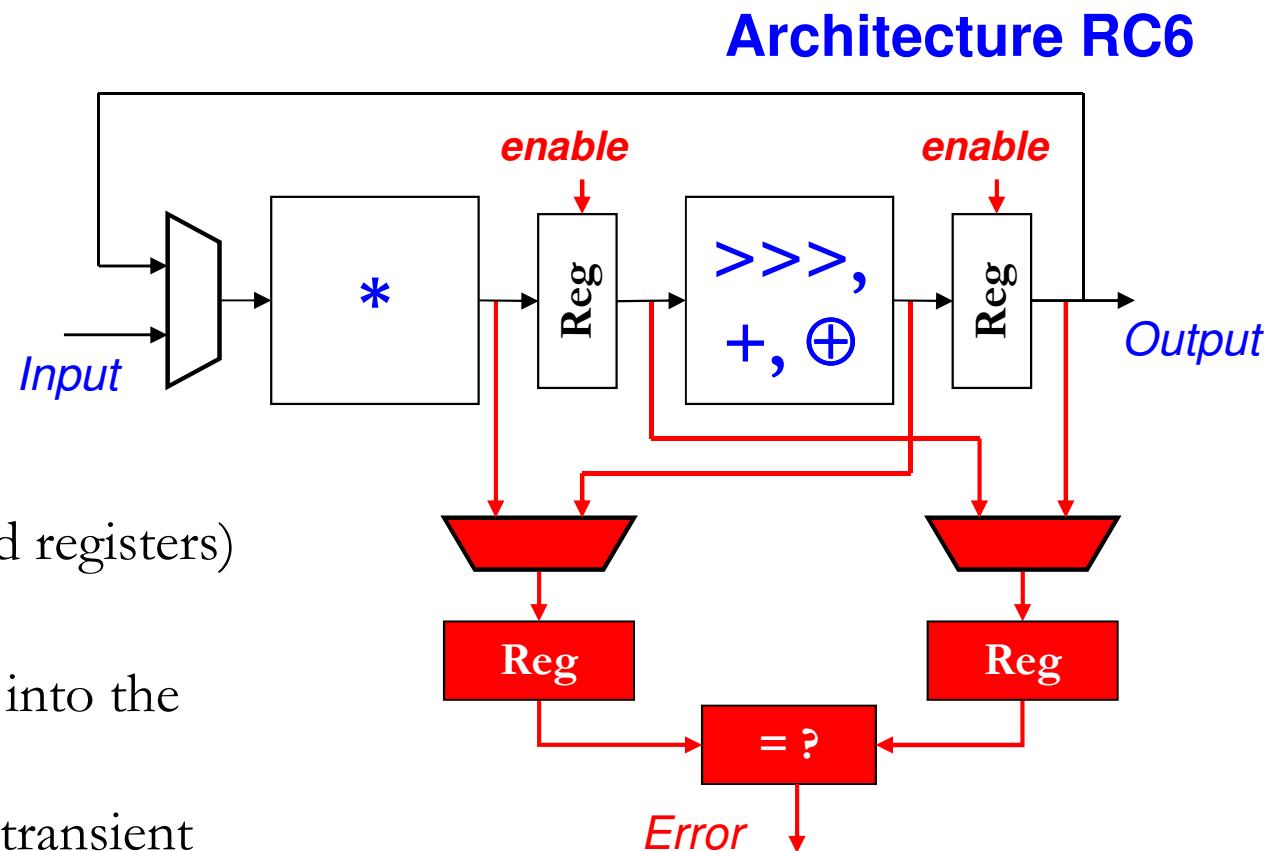


[Joshi et al., CHES 2004]

# Pipeline Redundancy

- Unused pipeline stages are used to redo the same computation twice

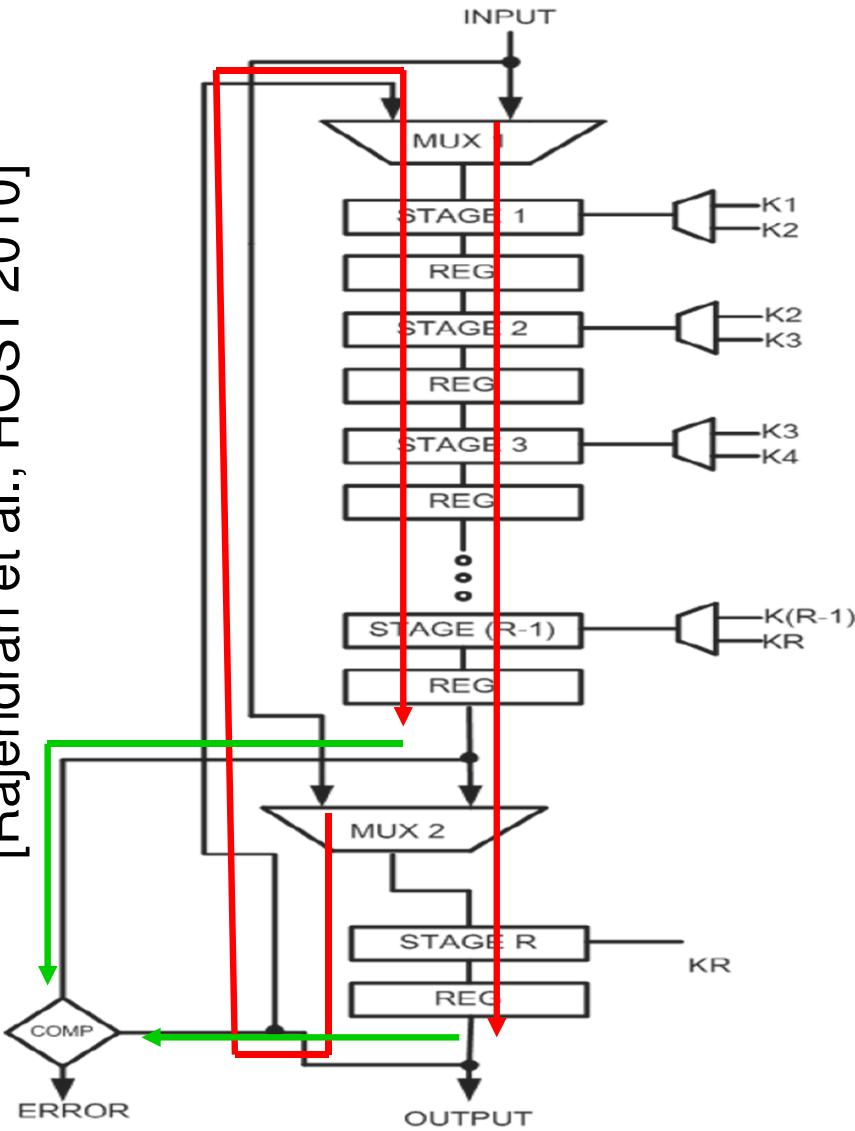
- Detection of transient faults
- Limited overhead (control and registers)
- Reduced detection latency
- Easy to integrate the pipeline into the design flow
- Neither permanent, nor long transient faults (>1 cycle)
- Partially unused pipeline is needed



[Wu, Karri. DFT 2001]

# SLICED

[Rajendran et al., HOST 2010]



Stage Cycle	1	2	3	4	5	6
1	R(1,1)					R(1,1)
2	R(2,1)	R(2,1)				
3	R(1,2)	R(3,1)	R(3,1)			R(1,2)
4	R(2,2)	R(2,2)	R(4,1)	R(4,1)		
5	R(1,3)	R(3,2)	R(3,2)	R(5,1)	R(5,1)	R(1,3)
6	R(2,3)	R(2,3)	R(4,2)	R(4,2)	R(6,1)	R(6,1)
7	R(1,4)	R(3,3)	R(3,3)	R(5,2)	R(5,2)	R(1,4)
8	R(2,4)	R(2,4)	R(4,3)	R(4,3)	R(6,2)	R(6,2)
9	R(1,5)	R(3,4)	R(3,4)	R(5,3)	R(5,3)	R(1,5)
10	R(2,5)	R(2,5)	R(4,4)	R(4,4)	R(6,3)	R(6,3)

- ☺ Transient faults
- ☺ Simple permanent faults
- ☺ Limited overhead (multiplexers)
- ☹ Frequency and throughput reduction
- ☹ Iterative unrolled implementation needed

# What was not so good so far...

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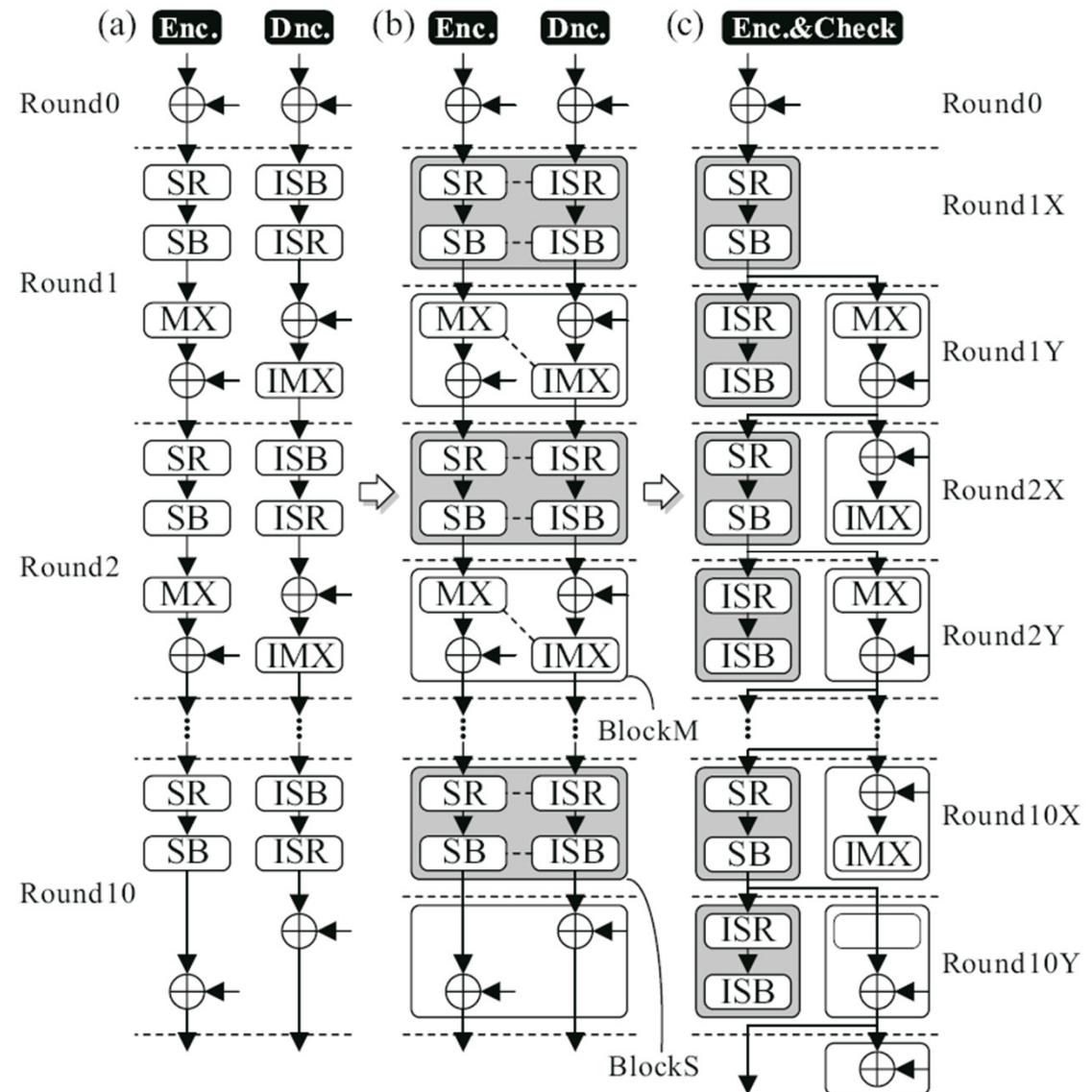
- Error detection codes for AES are either expensive (non-linear networks) or inefficient against malicious faults (parity)
- Spatial/information redundancy may increase correlation with power consumption and EM emissions, thus favoring side-channel attacks
- Temporal redundancy:
  - ◆ Process repetition involves high performance overhead
  - ◆ Pipeline implementation requires fast system clock and significant area overhead (+50%), but ...
  - ◆ ... the global system may work at reduced frequency, thus affecting the global throughput

# Inverse Computation with Pipeline

- Inverse computation combined with pipeline redundancy

- ☺ Detection of transient and permanent faults
- ☺ Limited overhead
- ☺ Low detection latency
- ☺ Quite easy to integrate into the design flow
- ☺ Decryption logic needed
- ☺ Pipeline needed
- ☺ Adaptability to other ciphers to be verified

[Satoh et al. CHES 2008]



# Part III

---

## Protection limitations Ex. DFA vs. DPA on registers

# Actually protecting a circuit

---

- Counter-measures usually proposed to protect against one type of attacks: Is it sufficient?
- Example of evaluation practice against faults:
  - ◆ Theoretical computation of the fault coverage
  - ◆ Simulation/emulation campaigns of fault injections
  - ◆ What about the impact of the added countermeasures on other types of attacks ?
- Issue: evaluate the global security of the protected device
  - ◆ Against Faults
  - ◆ Against Side Channels
  - ◆ Against ...

# Using error correcting/detecting codes

---

- One of the typical methods of protection
  - ◆ Well established theory – many codes available, protection efficiency against errors is theoretically well known
  - ◆ Timing attack: no extra sensitivity due to protection in synchronous circuits (at least for many classical codes)
  
- Concerns
  - ◆ Error assumptions – choice of code, and consequences on implementation costs
  - ◆ Power attacks: effect of the added checking bits ?
  - ◆ Electromagnetic attacks: derived from the power consumption

# Error detection: secured circuit ?

Weak link of the chain ? Open door ?

Timing attack

Fault-based attack using vdd/gnd/clk glitches

Fault-based attack using lasers

Differential Power Attack + EMA

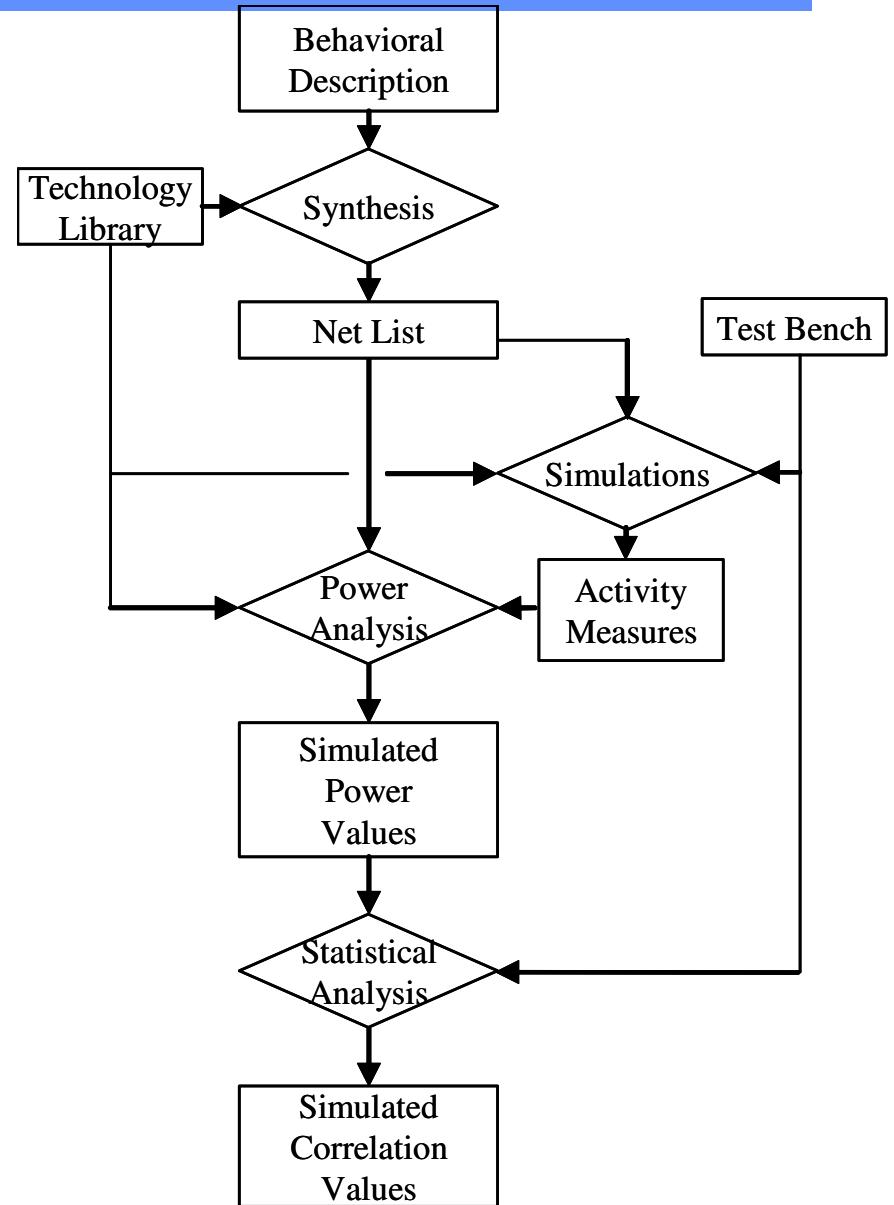
Error  
detecting/  
correcting  
codes  
...  
Hopefully !

Protected  
Circuit

?

# Power attack - sensitivity analysis flow

- **Activity measures**
  - ◆ Transistor-level: precision
  - ◆ Gate-level: simulation time
- **Power estimation**
  - ◆ Using activity measure
  - ◆ After/before P&R (precision)
- **Statistical post processing**
  - ◆ Formatting of the raw data
  - ◆ Statistical analysis (Matlab)

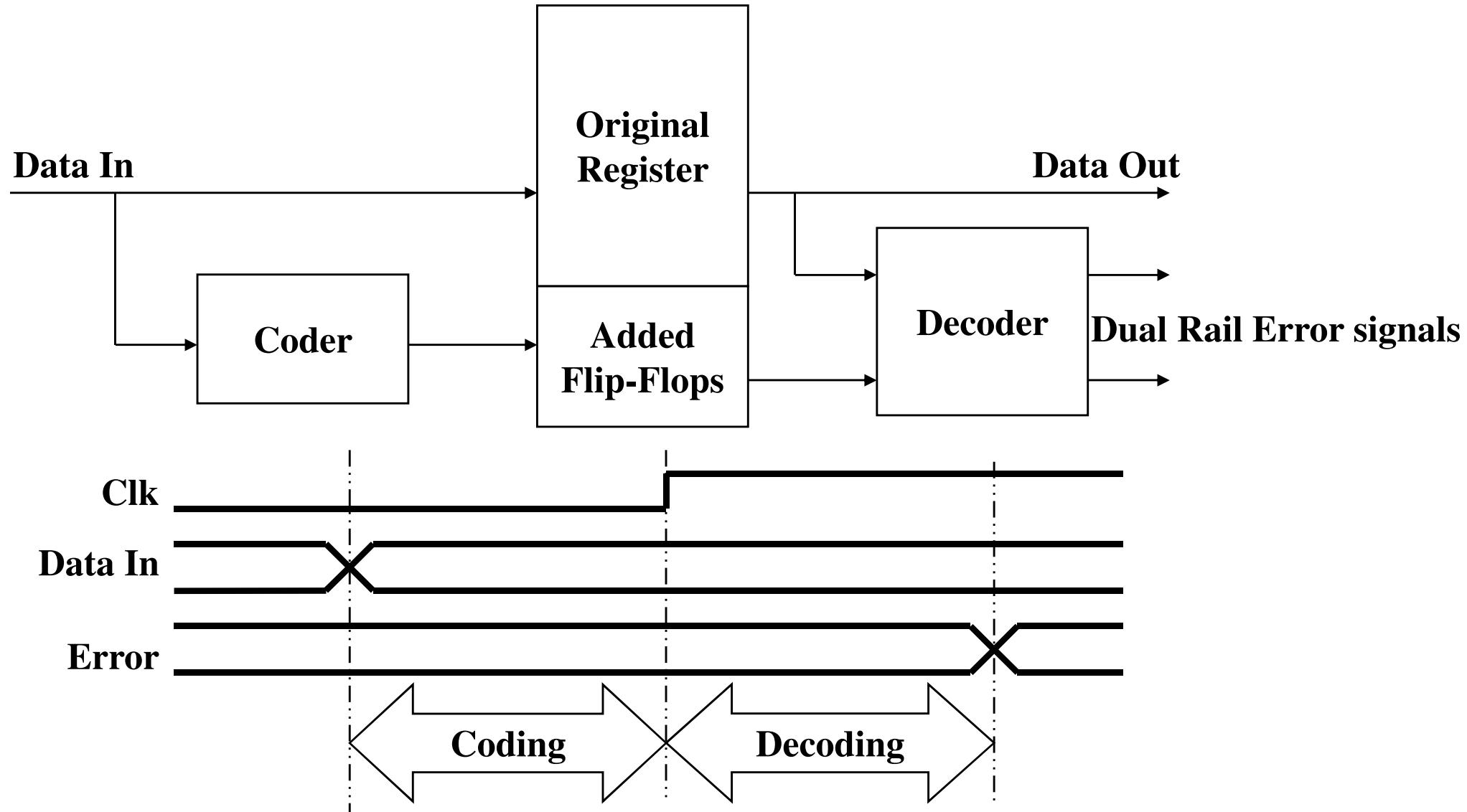


# Error detecting codes - a case study

---

- 8/16-bit registers
- 5 versions of the registers
  - ◆ No code
  - ◆ Simple Parity (one parity bit)
  - ◆ Double Parity (one parity bit for odd index bits)
  - ◆ Complementary Parity (both odd and even parity)
  - ◆ Berger Code (counts how many 0s)
- Technology: AMS C35 Corelib
- Exhaustive simulations for a null starting value (gate level)
- Correlation between data Hamming weight and the power consumption

# Hardened register



# Case study: cost and efficiency

---

- Quite high overheads
  - ◆ Decoder in dual rail
- Detection probability
  - ◆ Unknown multiplicity
  - ◆ Not only unidirectional errors
  - ◆ Uniform distribution of the errors
- Trend to have lower cost for bigger registers
  - ◆ Except Berger

Code	Area overhead 8-bit / 16-bit	Detection probability 8-bit
No code	+0% / +0%	0%
Simple parity	+120% / +105%	50%
Double parity	+159% / +127%	74%
Complementary parity	+230% / +207%	75%
Berger	+358% / +1727%	93%

- Best trade-off cost/detection:
  - ◆ Double parity
  - ◆ Complementary parity

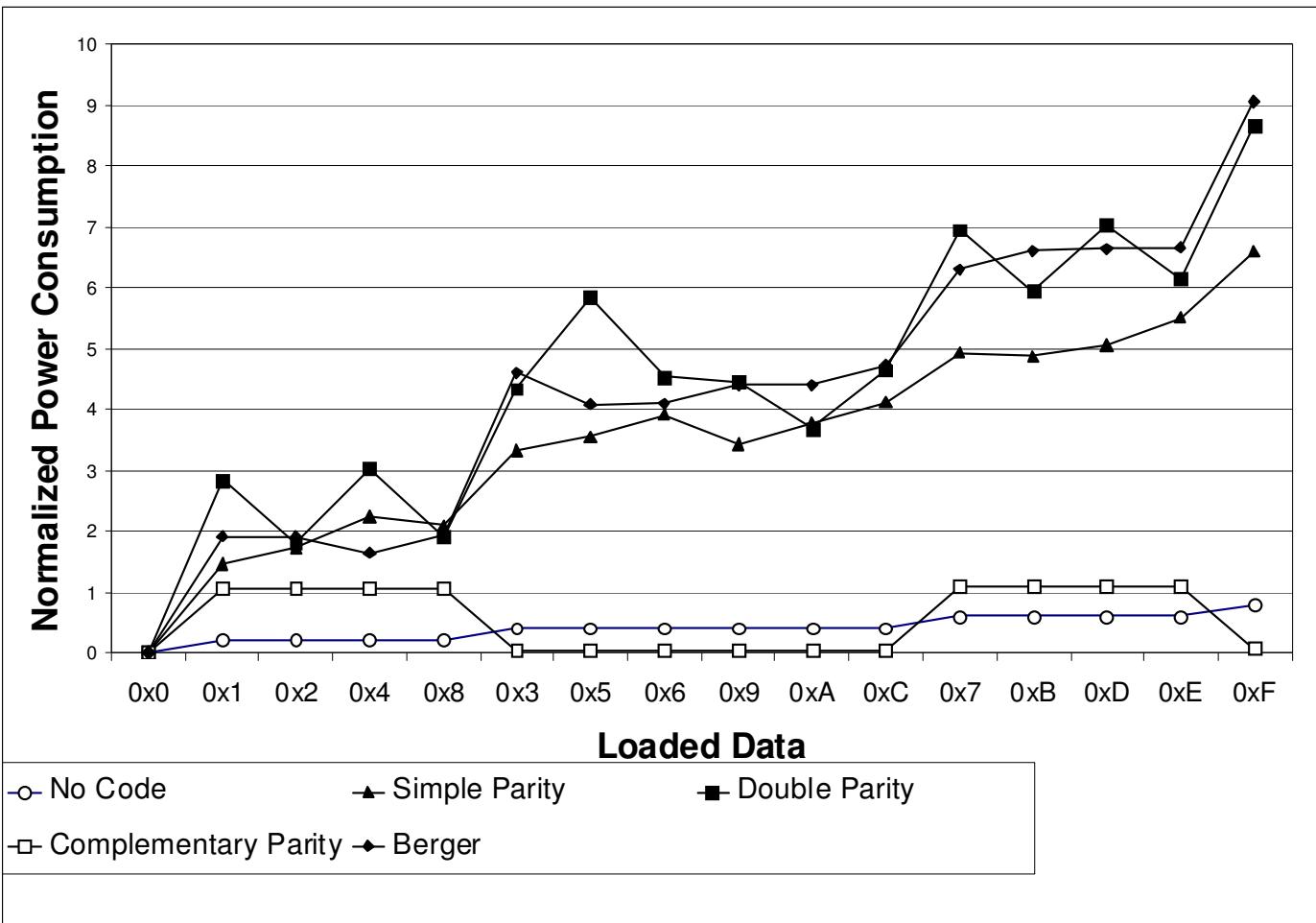
# Case study: consumed power

## □ Normalized consumed power

- ◆ Not very easy to use

## □ Almost linear dependence

- ◆ Except for complementary parity



# Case study: correlation values

Code	Correlation (Encoder) 8-bit / 16-bit	Correlation (Register) 8-bit / 16-bit	Correlation (Decoder) 8-bit / 16-bit	Correlation (Global) 8-bit / 16-bit
No code	n/a	1 / 1	n/a	1 / 1
Simple parity	0.995 / 0.965	0.927 / 0.945	0.955 / 0.949	0.977 / 0.970
Double parity	0.998 / 0.995	0.799 / 0.819	0.909 / 0.926	0.952 / 0.956
Complementary parity	<b>0.012</b> / 0.030	<b>0.334</b> / 0.537	<b>0.005</b> / 0.024	<b>0.056</b> / 0.162
Berger	0.989 / 0.849	0.999 / 0.970	0.993 / 0.929	0.994 / 0.935

## □ Majority of the codes

- ◆ Very high correlation
- ◆ Almost linear dependence between data weight and power consumed

## □ Correlation increase with the size of the register

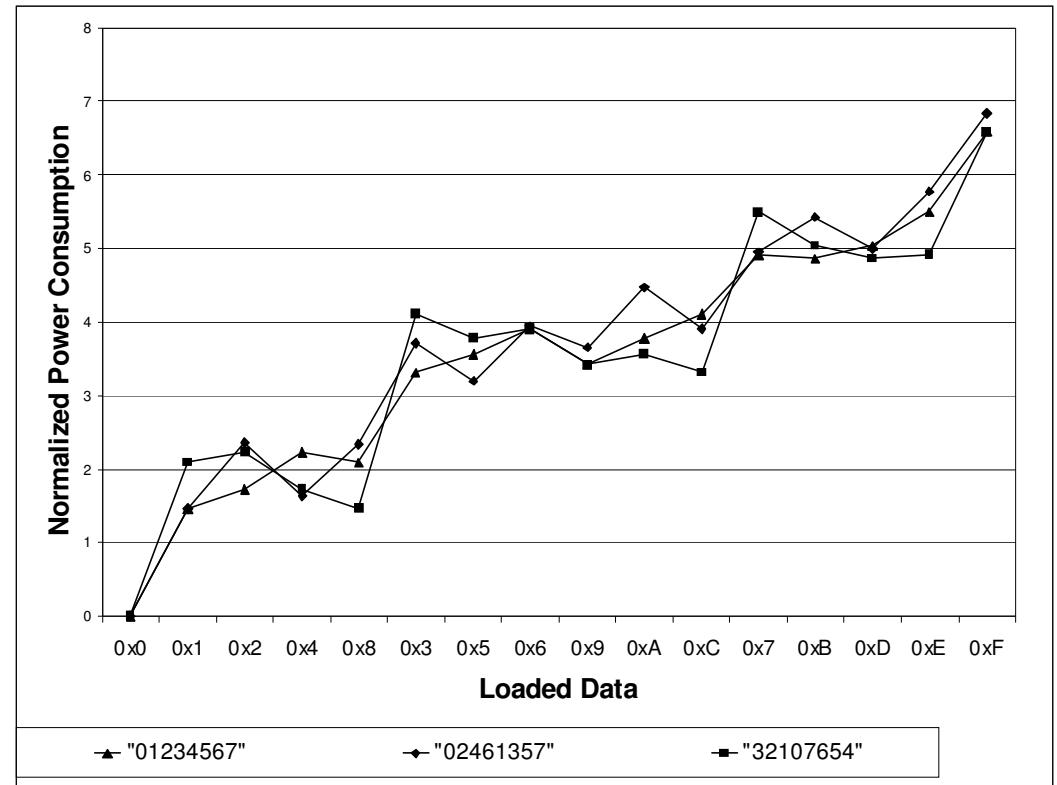
- ◆ Less redundancy bits in proportion

## □ Complementary parity

- ◆ Very low correlation (almost constant consumption)
- ◆ Implementation with dual XOR/XNOR trees

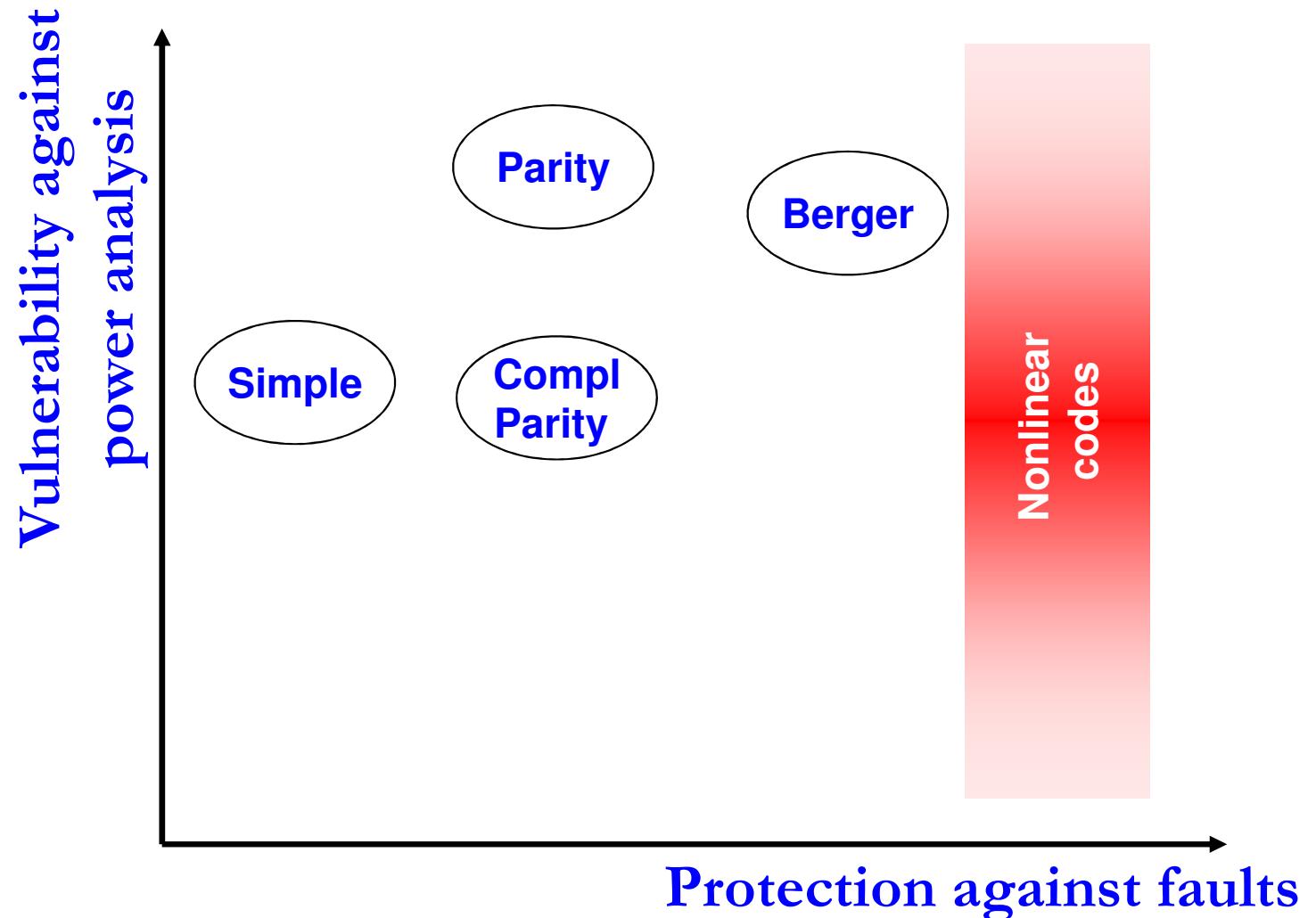
# Case study: implementation influence

- Power for 3 versions of a simple parity code
  - ◆ Permutation of the bits on the decoder XOR/XNOR trees
- All inputs are not equivalent
  - ◆ Different values of power
  - ◆ Small changes in correlation
- Implementation must be done carefully to limit this effect



# Code comparisons

- Complementary parity: best trade-off ?
- Attention must be given to both the design of the protection and its implementation



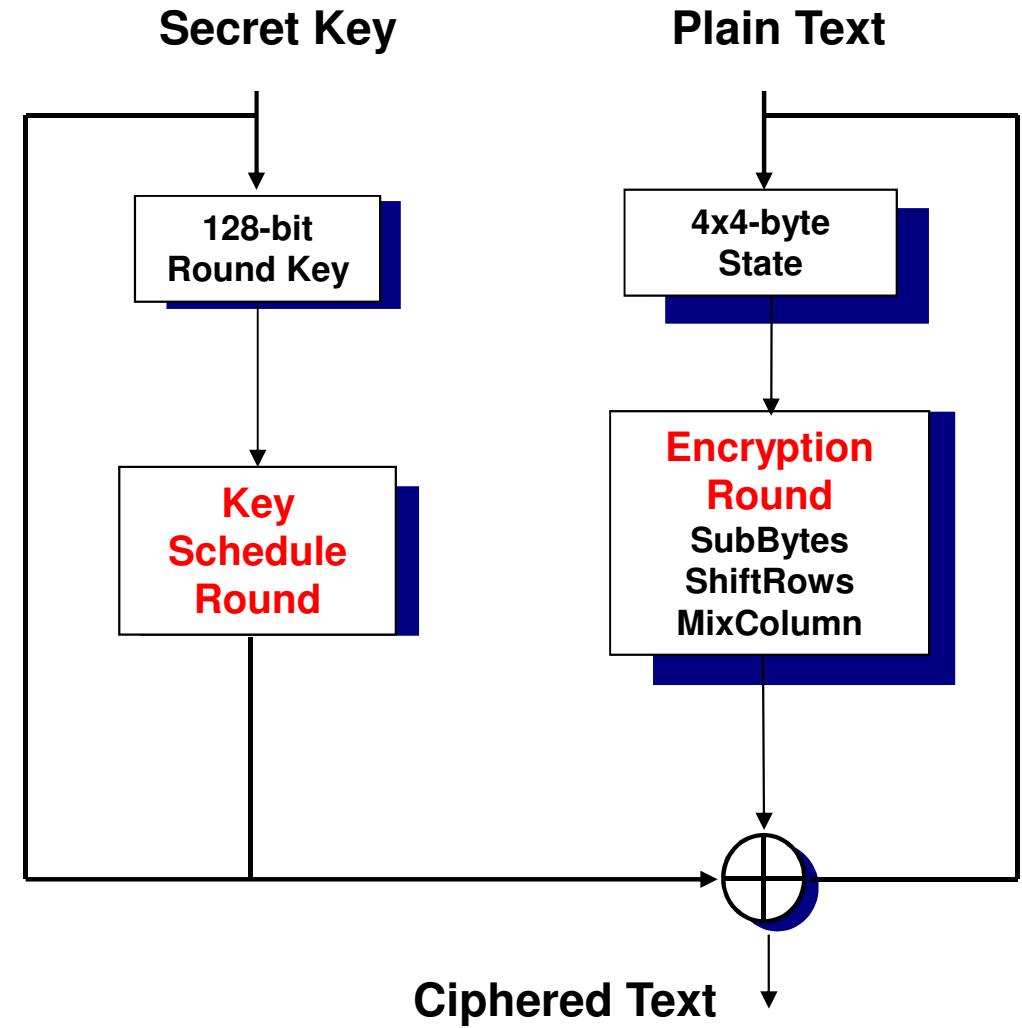
## Part II

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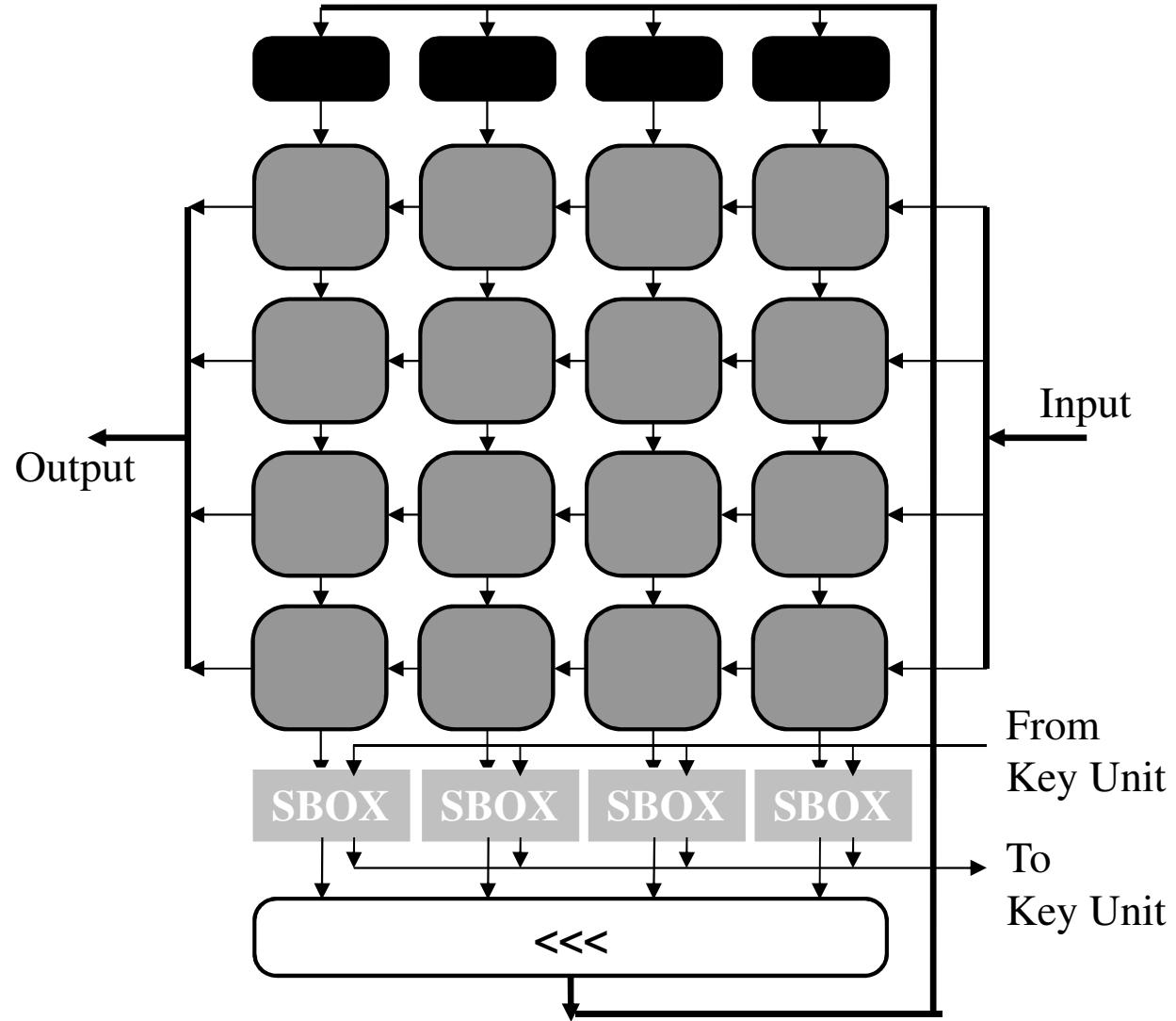
**Protection optimization**  
**Ex. An optimized AES architecture  
with timing redundancy**

# AES – the algorithm

- 128-bit data input
- 128/192/256-bit secret key
- Round-based (10 rounds for 128):
  - ◆ SubBytes: Non-linear byte substitution
  - ◆ ShiftRow: Word rotation
  - ◆ MixColumn: linear word transformation
  - ◆ AddRoundKey: modulus-2 addition with round key
- Encryption and Key Schedule use the same basic operations
  - ◆ Decryption uses inverse operations of encryption

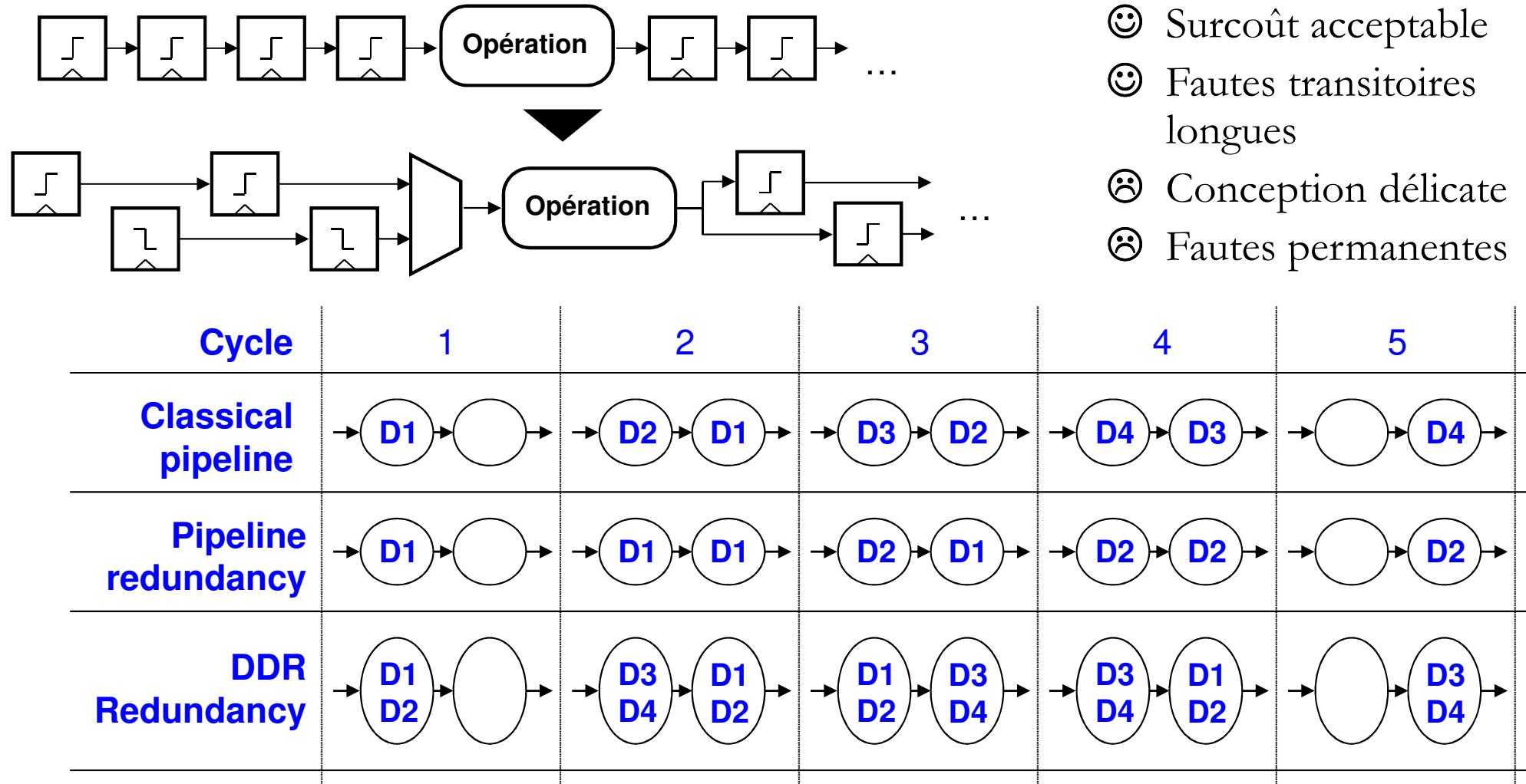


# AES – a data-path architecture



- 32-bit data-path
  - 4 Substitution Boxes
  - 16 GF Multipliers for MixColumns
  - 6 clock cycles per round
  - On-the-fly key unrolling (using shared S-Boxes)
- From Key Unit      To Key Unit
- MixColumns, AddRoundKey, State
  - 2-stage SBox
  - Register layer
  - Combinatorial logic
  - 8-bit signal
  - 32-bit signal

# Double Data Rate (DDR)



[Maistri, Leveugle. TC 2008]

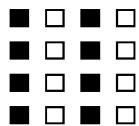
- ✓ 2x the throughput at same frequency
- ✓ Small area overhead for DDR logic

- ✗ More complex routing
- ✗ Overhead for error detection

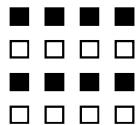
# Data alignment in AES

---

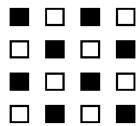
- The data alignment phase partitions the register space into two classes:
  - ◆ Registers triggered by ascending clock edge
  - ◆ Registers triggered by descending clock edge
- Alignment can be done:



**By columns: registers in the same columns share the clock alignment**



**By rows: registers in the same rows share the clock alignment**



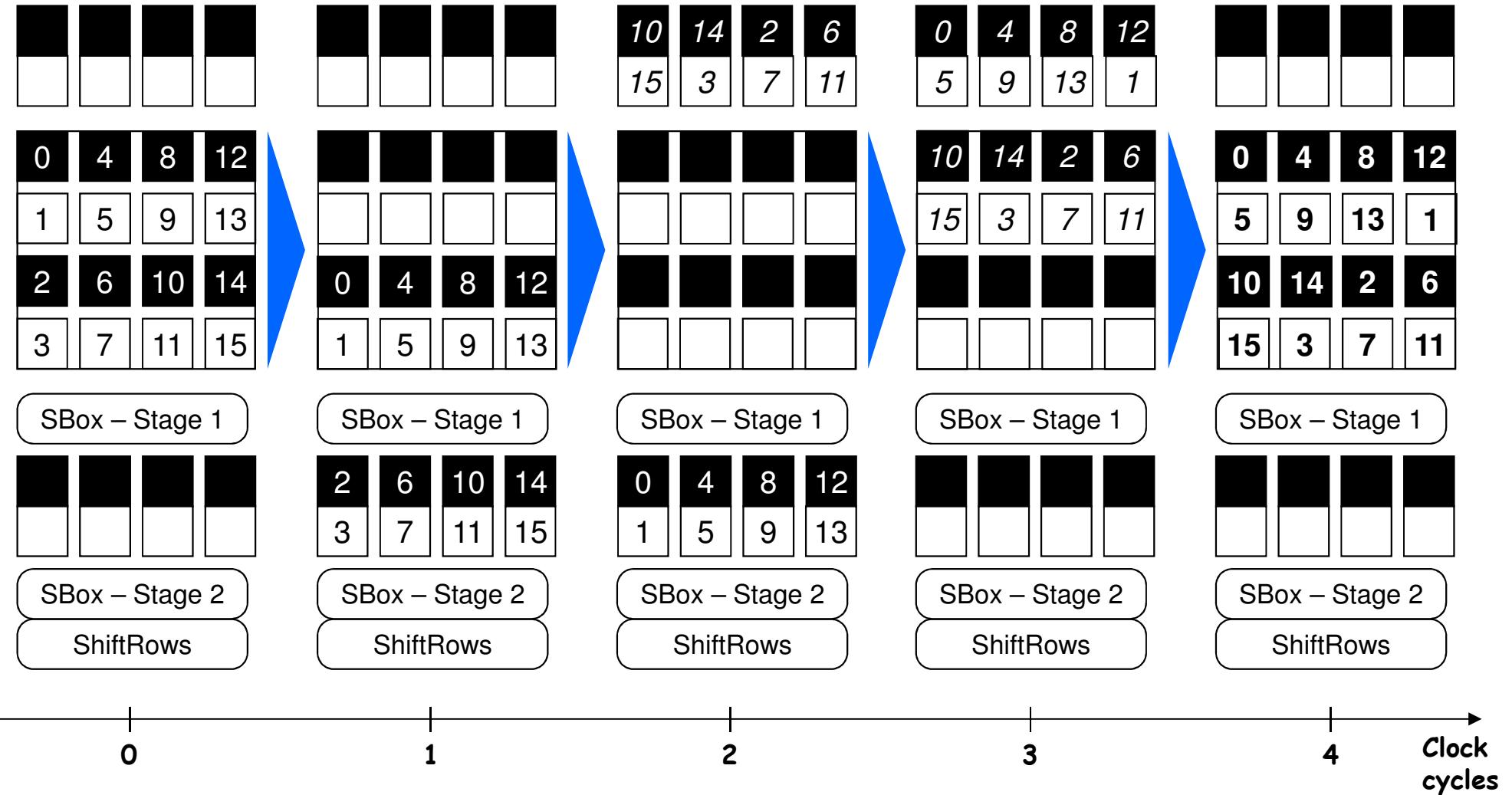
**By checkers: elements of the partitions are interleaved both in columns and rows, like a chess board**

# Synchronization

---

- DDR computation can be employed when we have scarce resources, high parallelism and no data dependency
  - ◆ In our design, SBoxes are the scarce resources
  - ◆ Row rotation is performed while moving data during non-linear substitution (collateral data-dependence)
  - ◆ Row-wise DDR alignment is thus chosen
- In AES, all operations are independent on each byte, but the *MixColumns* operation
  - ◆ MixColumns are not a scarce resource (each byte is computed locally), but values have to be stable (i.e., a latch is used)

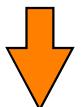
# DDR Round Encryption



# Operation modes

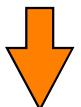
---

$(P_1, P_2)$



$(C_1, C_2)$

$(P_3, P_4)$



$(C_3, C_4)$

**Single:** the unit uses the DDR computation to improve its throughput and no check is performed on data

$(P_1, P_1)$



$(C_1, C_1)$

$(P_2, P_2)$



$(C_2, C_2)$

**Double:** the unit uses the DDR computation to compute each round twice, checking for inconsistencies

(dummy,  $P_1$ )



(dummy,  $C_1$ )

$(P_1, P_2)$



**Interleaved:** like the *Double* mode, but the first and second repetition are processed together with two different (consecutive) blocks in ECB mode, which share the encryption key

# Parity vs. Pipeline vs. DDR

	Parity	Pipeline*	DDR
Area	+ 23%	+ 50%	+ 36%
Throughput reduction	- 3%	- 18%	- 15% Single - 55% Double
Notes	One parity bit per byte	Temporal redundancy w/ unused pipeline stage	Temporal redundancy w/ both clock edges
Pros	High throughput and covers all single faults	(Almost?) all 1-cycle faults should be detectable	Same computations are very far in time; 2x Thr. @ same freq.
Cons	No even-order faults Poor on fault attacks	No guarantees if faults last more than one cycle, altering more bytes	Lower frequency; more complex design than pipeline

\* RC6 implementation

# Protections beyond DDR: controller

---

**Reduce the number of possible targets**

**Simplify the controller removing redundant registers, which store signals that could be instead computed on the fly**

**Protect sensitive targets**

**Specific registers are duplicated to ensure correct behavior (e.g., counters, state registers)**

**Validate state encoding**

**If any FSM falls into an unused state encoding, computation stops and they both return to the reset state**

**Verify state transitions**

**If an FSM performs an erroneous transition (e.g., Idle > Output), the error signal is raised and the machines go back to their reset state**

# Implemented DDR architectures

---

- Medium (represented in previous figures)
  - ◆ 4 S-Boxes, 16 linear multipliers
  - ◆ Synchronization driven by state rotation
  - ◆ Horizontal data alignment: synchronization layer required for column-wise linear multiplication
  - ◆ 64 cycles/block
  
- Small
  - ◆ 4 S-Boxes, 4 linear multipliers
  - ◆ Synchronization driven both by state rotation and linear multiplication
  - ◆ Vertical data alignment: synchronization layer required for row-wise state rotation
  - ◆ 100 cycles/block

# Software simulations

---

- Parity code was previously validated by algorithmic simulations
  - ◆ Computation model was used to further simplify the robustness analysis
  - ◆ Confirmed expected results: only those faults injecting an odd number of bit errors in at least one byte were detectable
- Fast and simple but...
  - ◆ Errors were injected only at the beginning of a round (actually a negligible limitation)
  - ◆ Only computation variables could be altered, no possible intervention on control flow
  - ◆ Software algorithm can not be an accurate model of a hardware implementation

# Injection campaign

---

- **Location**
  - ◆ Linear multiplication layer
  - ◆ Barrel shifter (row rotation)
  - ◆ Non-linear substitution layer: inner and outer stage
  - ◆ Control unit
  - ◆ AES is highly regular: only one target element for each data path location
- **Timing**
  - ◆ Every computation clock cycle
  - ◆ Input or output phase were not considered
  - ◆ From 1 up to several clock cycles (twice the length of a round)
- **Value**
  - ◆ According to the laser-injected fault model, the error value is not controllable
  - ◆ Exhaustive search of all error values is carried on for each targeted area (e.g., all byte values for a byte target)

# Implementation and injection results

Architecture	Area Overhead	Throughput Reduction	Coverage Byte Errors in Datapath
<b>Multiple Parity Bits</b> Bertoni et al., TC '03	33%	3%	~67%
<b>Inverse Process</b> Karri et al., DAC '01	19% - 38%	23% - 61%	100%
<b>Pipeline Recomputation</b> Wu and Karri, DFT '01	50%	18%	~100%
<b>Single Parity Bit</b> Karri et al., CHES '03	18% - 24%	NA	~67%
<b>Non-linear Code</b> Karpovsky et al., DSN '04	77%	13%	~100%
<b>DDR</b>	36%	15% - 55%	~100%

# Protection in the control unit

One-Hot Encoding		Result Class [%]			
Target	Size (bits)	Silent	Undetected	False Pos	Detected
<b>Low FSM</b>	<b>9</b>	<b>4.36</b>	<b>0.20</b>	<b>4.92</b>	<b>90.52</b>
<b>High FSM</b>	<b>19</b>	<b>0.00</b>	<b>16.30</b>	<b>1.87</b>	<b>81.63</b>
<b>Inner Signals</b>	<b>6</b>	<b>15.74</b>	<b>84.26</b>	<b>0.00</b>	<b>0.00</b>

**Reduce the number of possible targets**

**Protect sensitive targets**

**Validate state encoding**

**Verify state transitions**

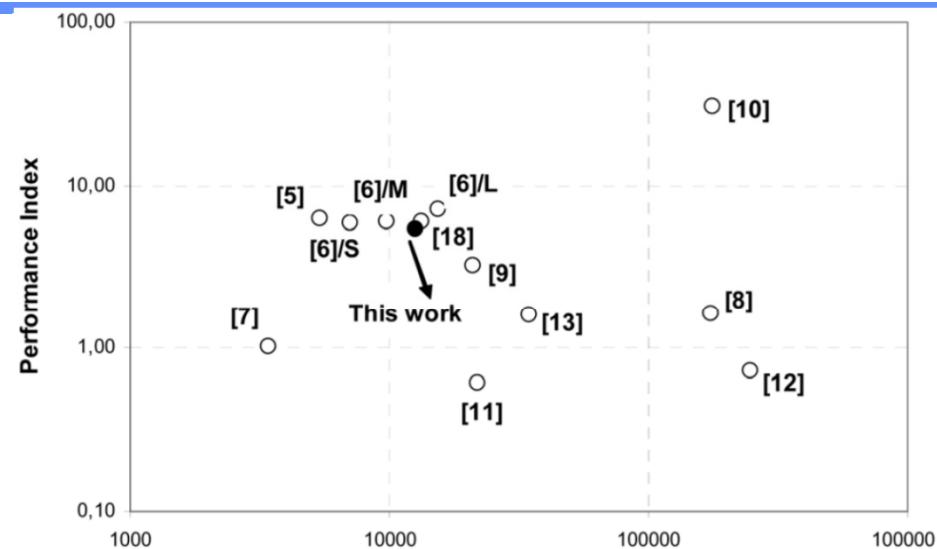
**The controller has been simplified removing superfluous registers**

**Specific registers are duplicated to ensure correct behavior (e.g., counters, state registers)**

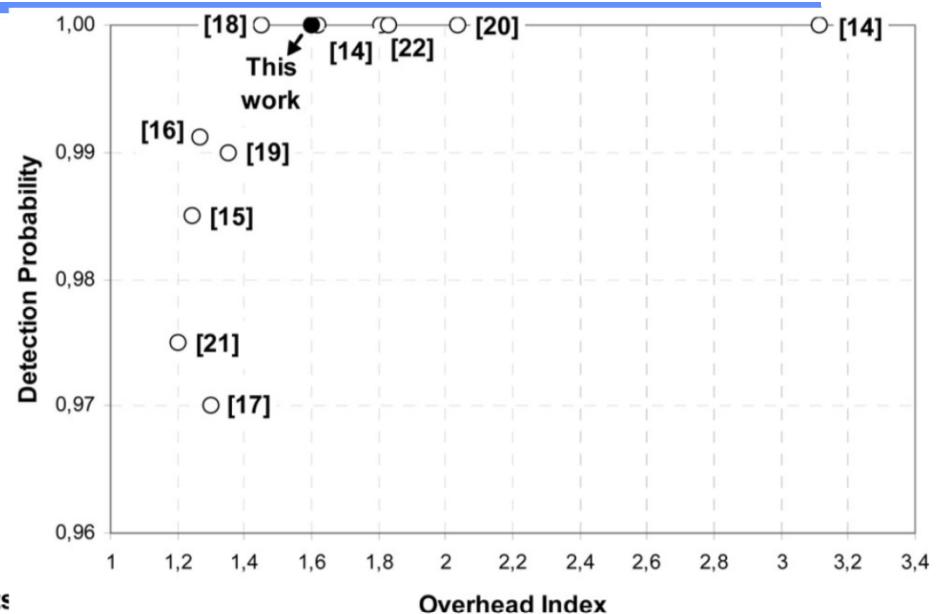
**If the FSMs fall into a non-existing state encoding, they return to the reset state**

**If an FSM performs an erroneous transition (e.g., Idle > Output), the error signal is raised**

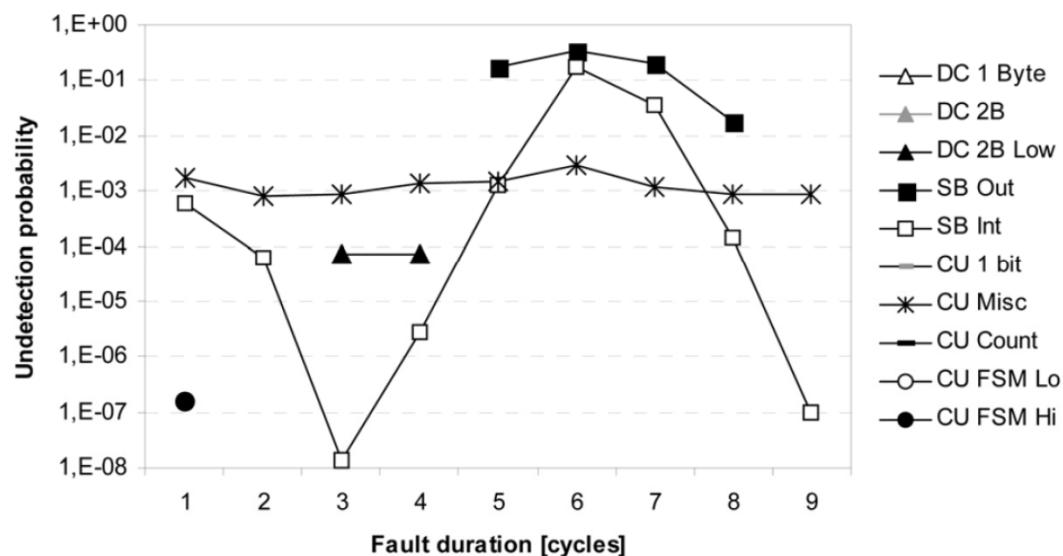
# AES DDR – Results [TC'08]



Performance/area characteristics

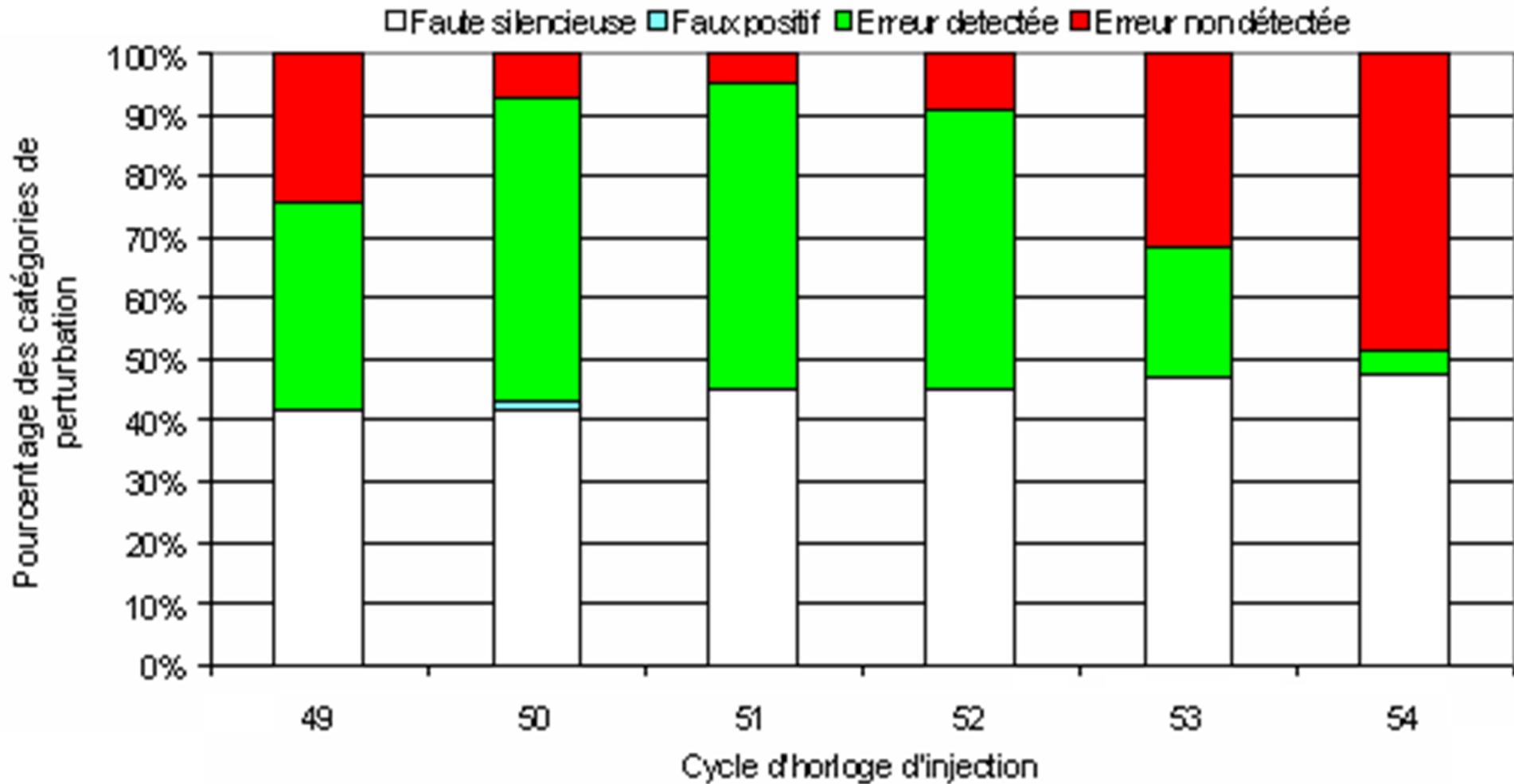


Undetected faults vs. duration



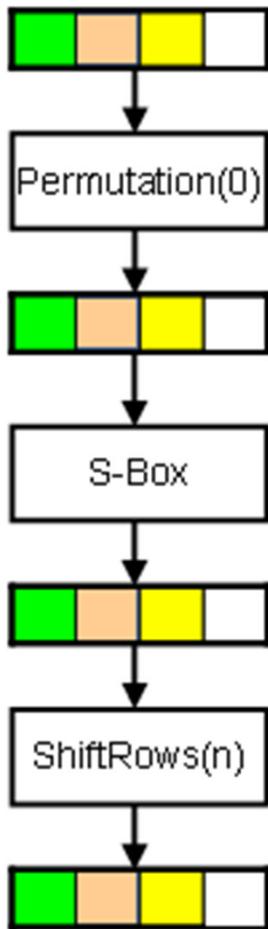
Detection vs. overheads

# DDR on FPGA – Laser Injections

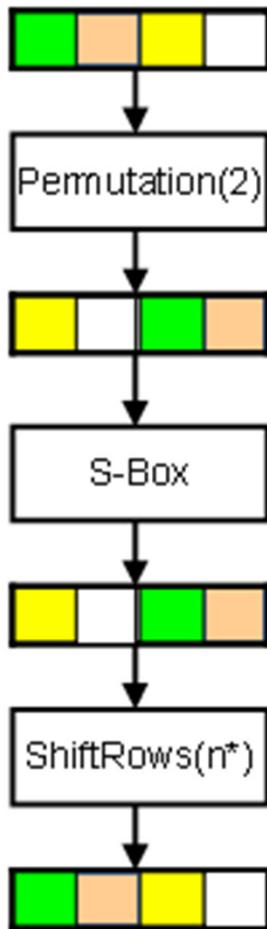


# DDR and Resource Reallocation

Normal round



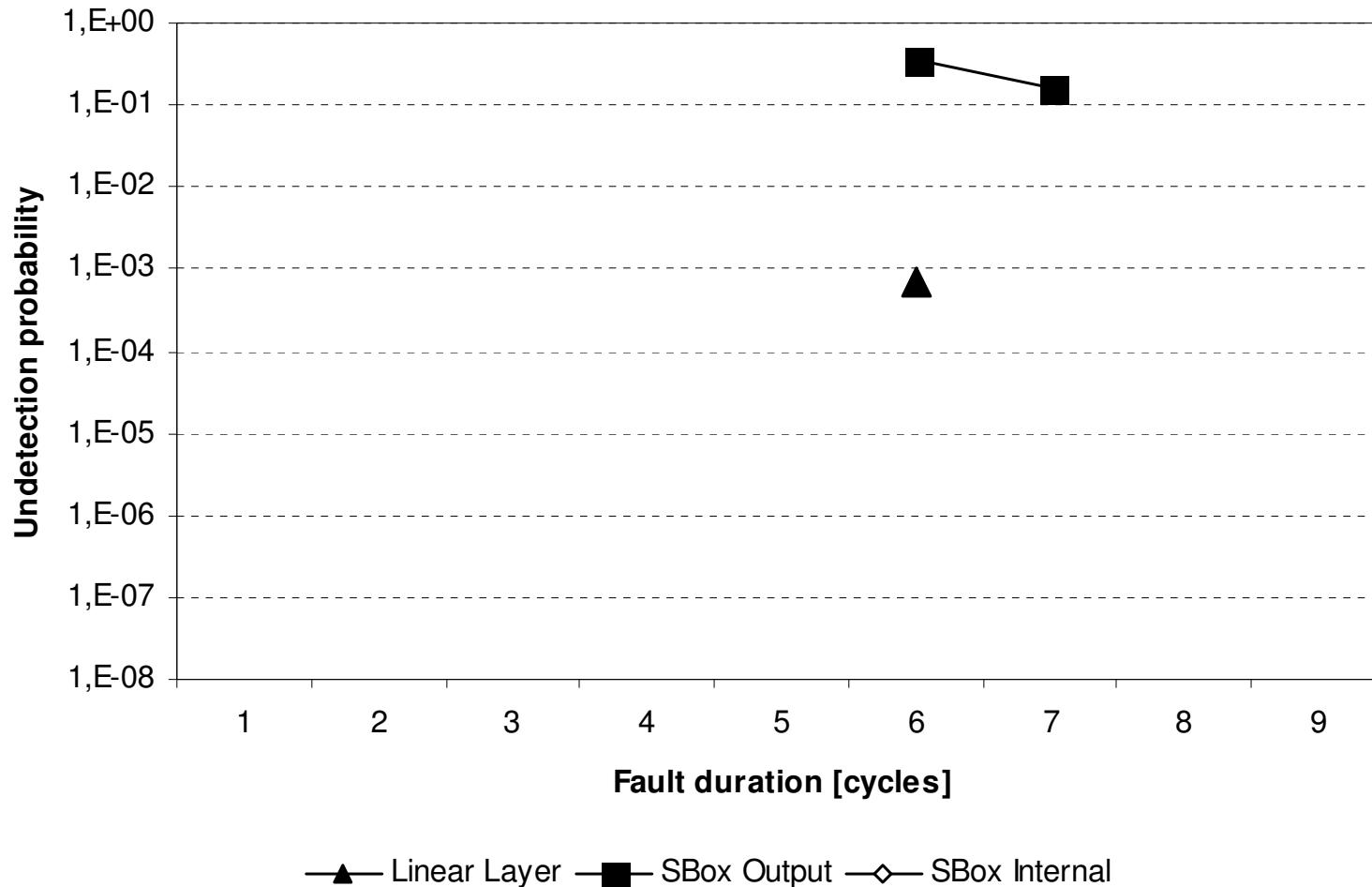
Verification round



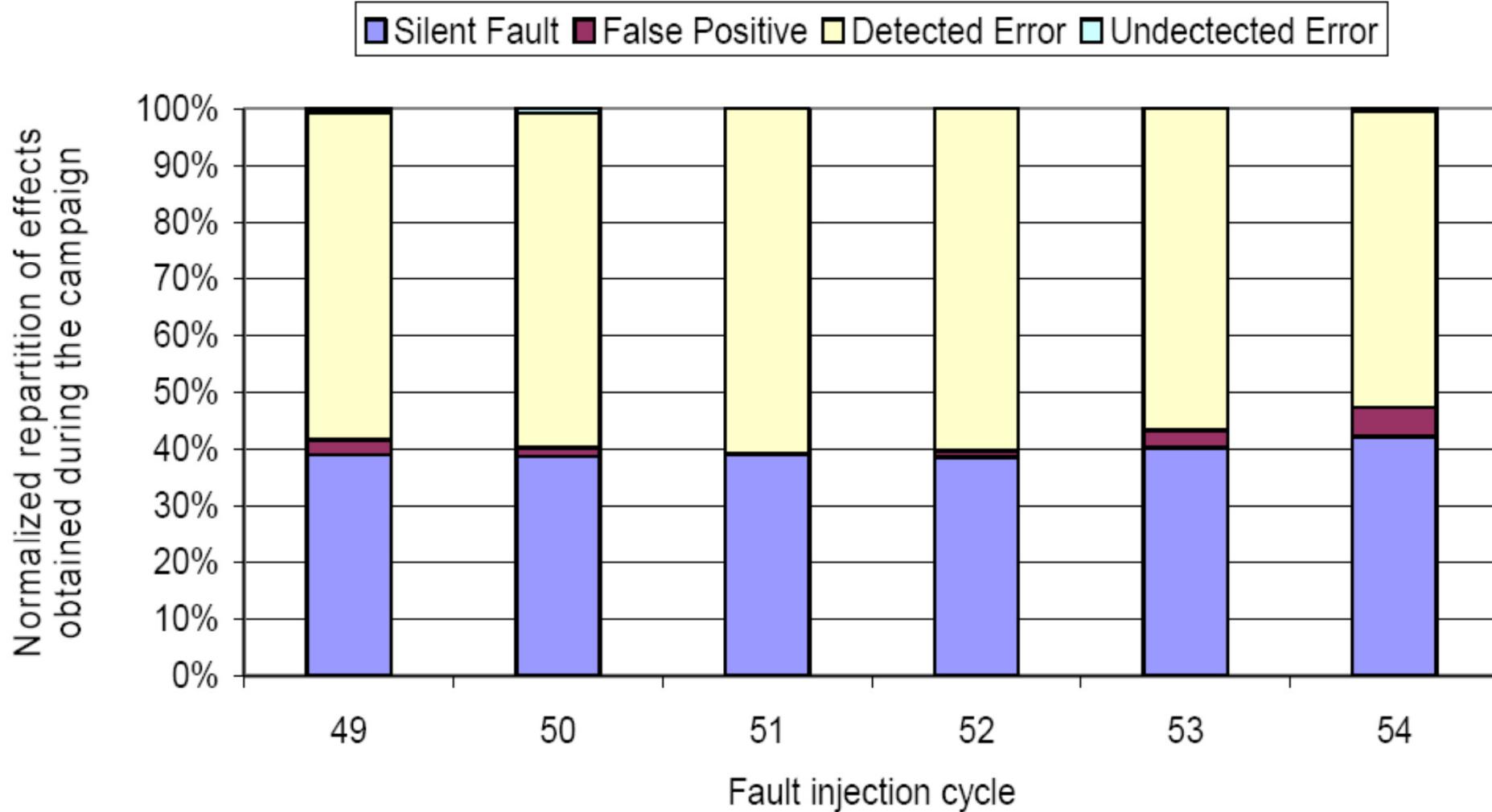
Results	Percentage
Shots with no effect	54,6 %
Modified configuration	45,4 %
<i>Silent Faults</i>	39,6 %
<i>False Positives</i>	2,3 %
<i>Detected Errors</i>	57,9 %
<i>Undetected Errors</i>	0,3 %

# An improved counter-measure

- Taking into account remanent (or permanent) faults  
=> resource reallocation between normal and verification rounds



# Laser attacks on last DDR protected version



# Overview of efficiency vs. overheads [P. Maistri, IOLTS 2011]

Type	Description	Permanent Faults		Multiple transient	Area overhead	Throughput reduction
		Single	Multiple			
Hardware	Duplication	100%	😊	😊	> 100 %	-
	Partial red	😊	😊	😢	25 – 35 %	-
Information	Byte/Word Codes	100 %	> 99 %	50 – 99 %	20 – 30 %	3 %
	Block Codes	98 %	😢	😢	18 – 24 %	~ 0 %
	Nonlinear code	1 – 2 <sup>-56</sup>	1 – 2 <sup>-56</sup>	😊	77 %	13 %
Time	Process repetition	0 %	0 %	😊	~ 0 %	50 %
	Involution ciphers	100 %	😊	😊	~ 0 %	50 %
	SLICED	100 %	😊	😊	~ 0 %	64 %
	Pipeline Red	0 %	😢	😊	50 %	18 %
	DDR Red	0 %	😢	😊	36 %	15 – 55 %
	DDR++	> 90 %	😊	😊	~ 36 %	~ 15 – 55 %
Mixed	Inverse Operation	100 %	😊	😊	19 – 38 %	23 – 61 %
	Inv Pipeline	100 %	😊	😊	-21 – 24 %	~ 25 %

# Temporal Redundancy - Summary

---

- Repetition of the same or of the inverse computation
- Detection of transient faults
  - ◆ Permanent faults detected by inverse computation
- Area overhead and throughput reduction acceptable when implemented with a pipeline
- No significant vulnerability against side channel attacks
  - ◆ if the basic architecture is protected itself, of course

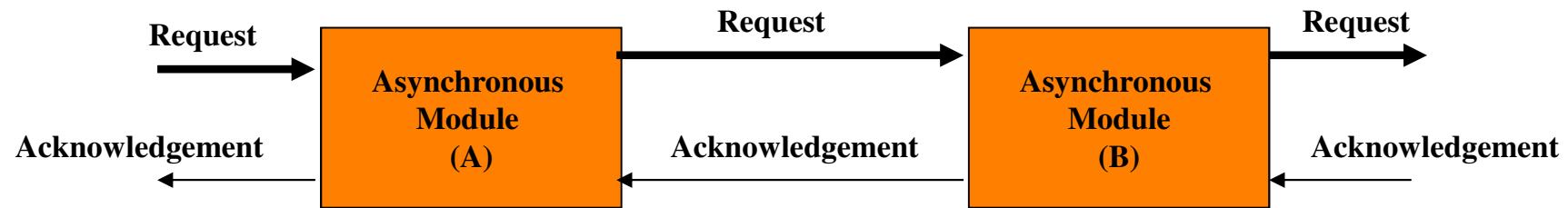
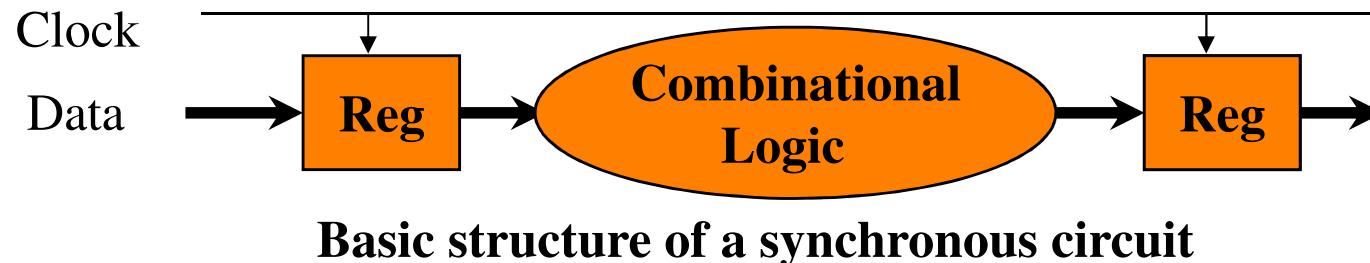
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# Influence of design style

## Example:

## synchronous vs. asynchronous logic

# Asynchronous Logic: Quasi Delay Insensitive circuits

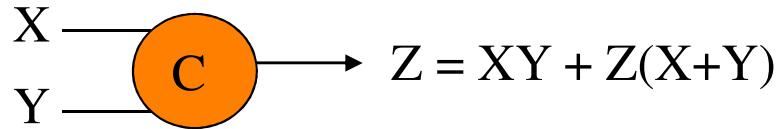


Handshake-based communication between modules (QDI)

Intrinsically more robust to perturbations (delay faults)  
and SCA (better current profile)

# QDI memory cell: the Muller gate (C element)

---



**Truth table:**

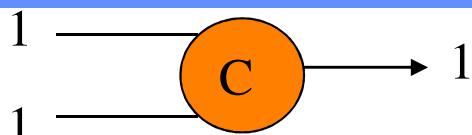
X	Y	Z
0	0	0
0	1	$Z^{-1}$
1	0	$Z^{-1}$
1	1	1

This gate is used to implement both the data paths and the protocol specification.

**QDI circuits sensitivity criterion =>**  
**Ability of Muller gates to memorize/filter a transient fault**

# QDI circuits: specific sensitivity criterion

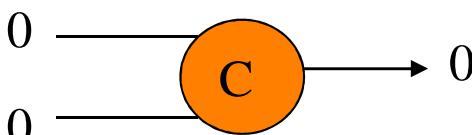
'Set' state:  
(non sensitive)



Single fault at input

→ 0 Fault is filtered !

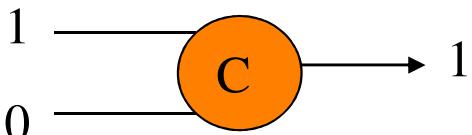
'Reset' state:  
(non sensitive)



Single fault at input

→ 1 Fault is filtered !

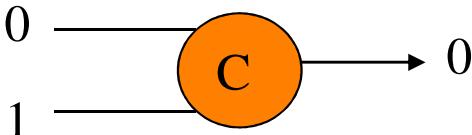
Sensitive to '1':



Single fault at input

→ 0 Fault is memorized !

Sensitive to '0':



Single fault at input

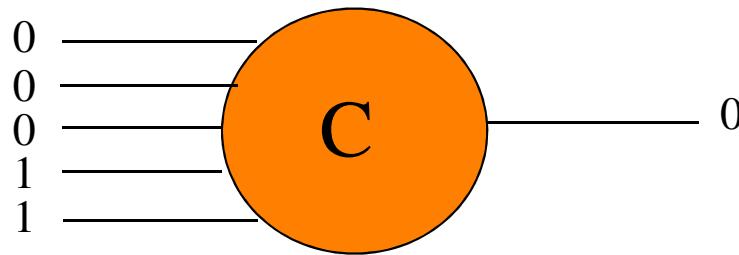
→ 1 Fault is memorized !

Y. Monnet, M. Renaudin, R. Leveugle, "Asynchronous circuits sensitivity to fault injection",  
10th IEEE International On-Line Testing symposium, Funchal, Madeira, Portugal, July 12-14, 2004, pp. 121-126

# Generalization of the sensitivity criterion

---

N-input Muller gate: sensitivity to [1..M] erroneous inputs



A “3-sensitive to 0” Muller gate

# QDI circuits: specific protections

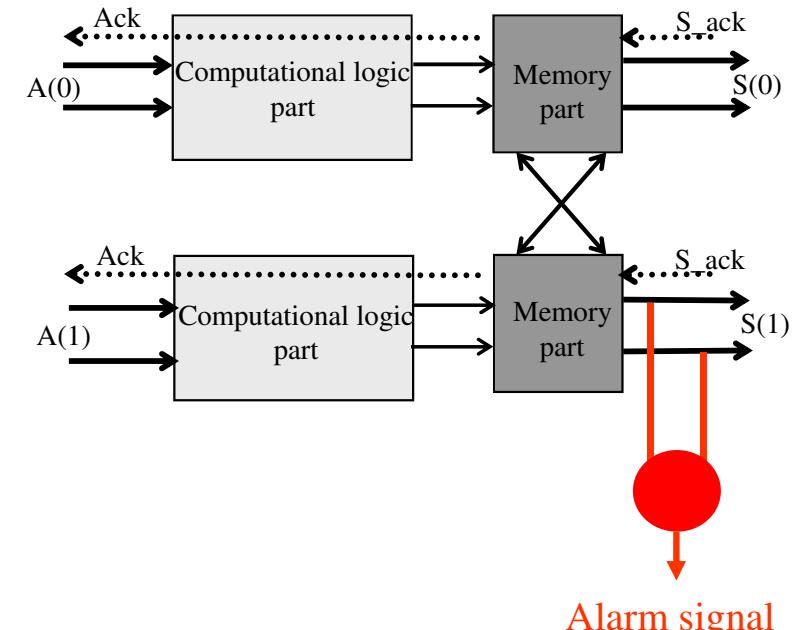
## □ Objective :

- ◆ Protecting logic blocs against transient faults
- ◆ Detecting wrong code generation

## □ Rail synchronization technique

- ◆ Improves the tolerance against transient faults
- ◆ Improves the wrong code detection when combined with alarm cells

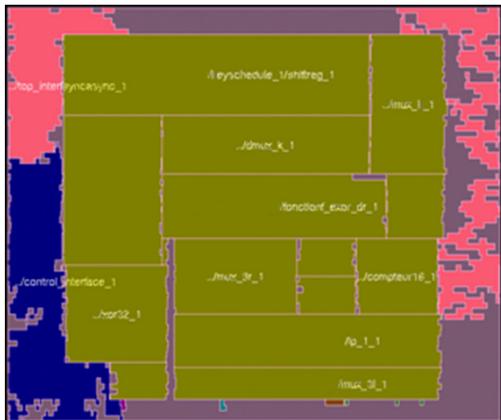
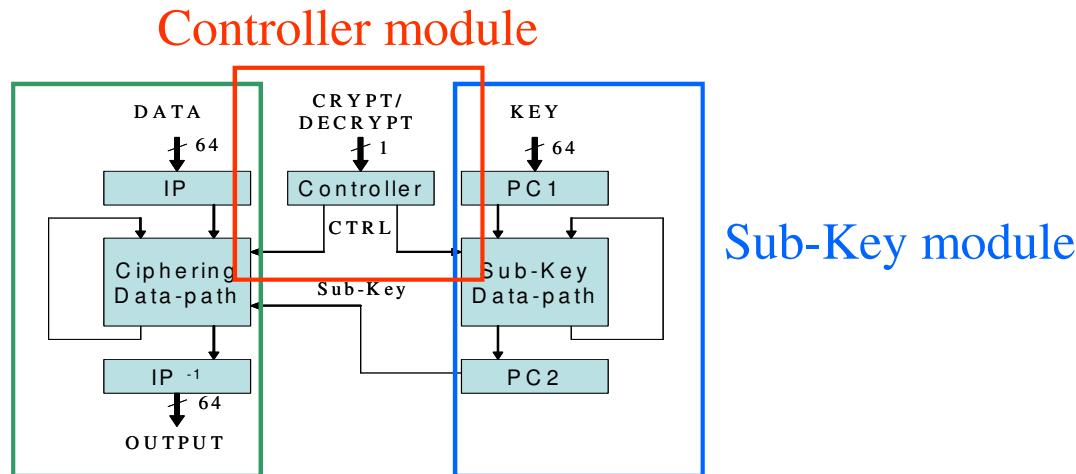
## □ Alarms cells



Y. Monnet, M. Renaudin, R. Leveugle, N. Feyt, P. Moitrel, F. M'Buwa Nzenguet, "Practical evaluation of fault countermeasures on an asynchronous DES cryptoprocessor", 12th IEEE International On-Line Testing symposium, Como, Italy, July 10-12, 2006, pp. 125-130

# Asynchronous DES crypto-processors

Data path module



130 nm STmicroelectronics CMOS process

**Low cost of hardening techniques:  
about 10% overhead in terms of speed and area,  
and negligible overhead in terms of power consumption**

# Practical attacks: laser setup

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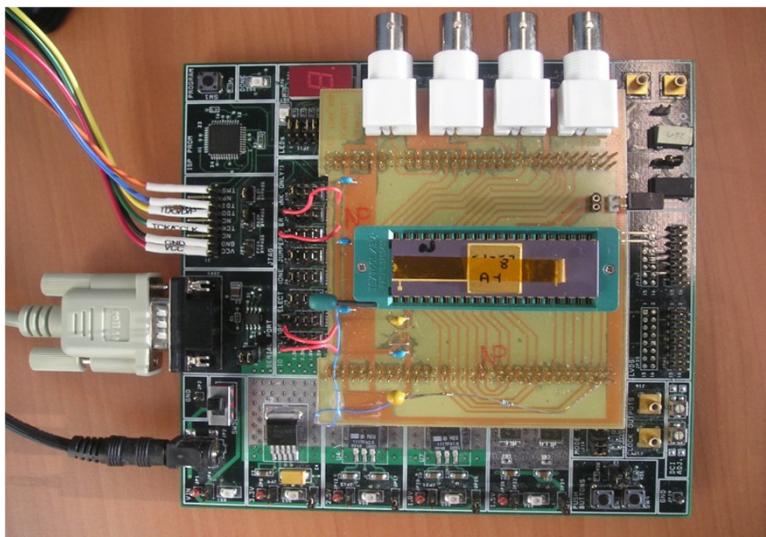
Laser characteristics:

Green Laser

6 ns pulse

Tunable spot size ( $220 \mu\text{m}^2$ )

Tunable energy (0.8 pJ/ $\mu\text{m}^2$ )



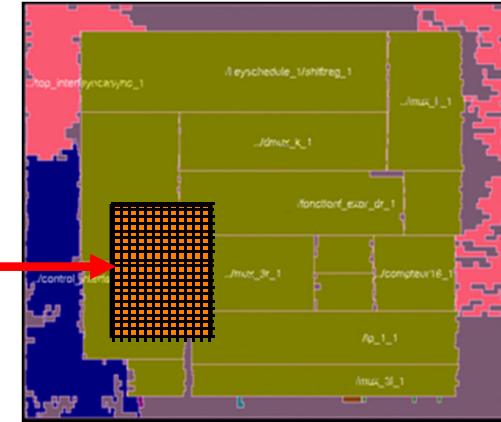
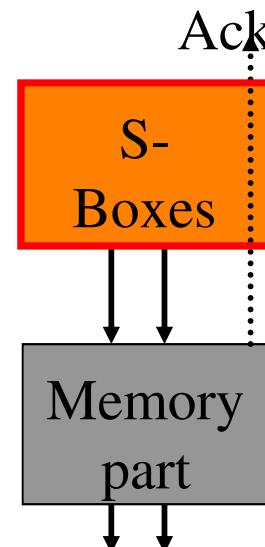
Laser Board for the DES



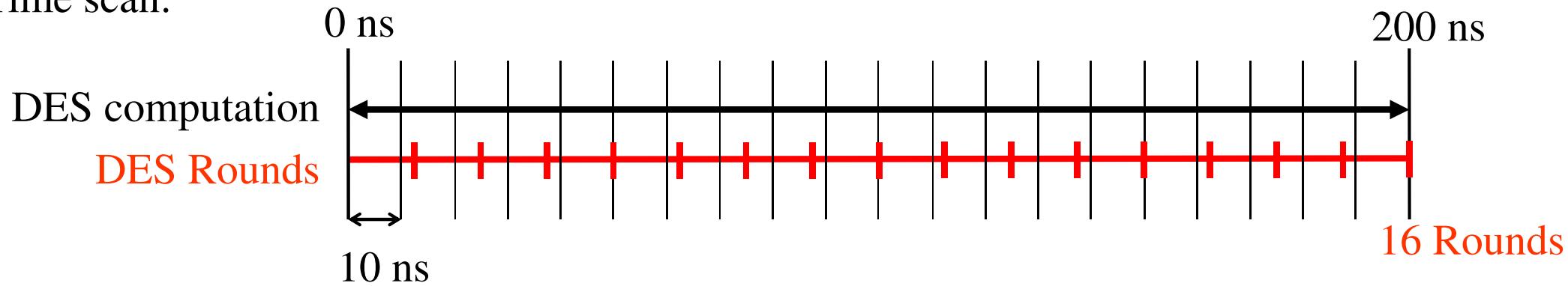
Gemalto laser platform

# Campaign specification (example)

# Spatial scan in S-Boxes (non-linear substitution tables)



## Time scan:



# Results on Sbox

---

## □ Asynchronous logic characterization:

Only 17% of the shots lead  
to an observable error

- > Delay insensitivity ?
- > Asynchronous circuit properties ?

SBOX	Number of shots	Number of errors
Reference	5600	955 (17%)
Hardened	5600	377 (7%)

## □ Fault tolerance of the hardened DES is improved by 2.5

## □ Detection:

- ◆ in unhardened version means deadlock (no end signal)
- ◆ in hardened version means deadlock or alarm

SBOX	Detected errors	Undetected errors
Reference	642 (67%)	313 (33%)
Hardened	377	0 (0%)

# Prevention is better than cure

---

- Instead of detecting the error, avoid the injection:
  - ◆ Filters on power supply
  - ◆ Sensors for glitches and overvolting
  - ◆ Temperature sensors
  - ◆ Light sensors
  - ◆ Metallic shielding
  - ◆ Memory encryption
  - ◆ Address scrambling
  - ◆ ...