

PART 17: Input and output resistances of feedback amplifiers

We have frequently stated that

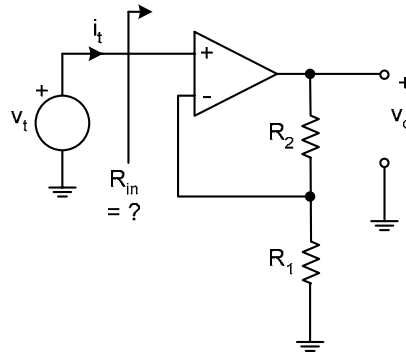
“Negative Feedback derives R_{in} and R_o towards ∞ or 0”

Justification – using op-amp as an example

1. Input Resistance of Non Inverting Amp

If we can calculate i_t due to the applied test voltage, v_t then:

$$R_{in} \equiv \frac{v_t}{i_t}$$



Need to use equivalent circuit of an op-amp

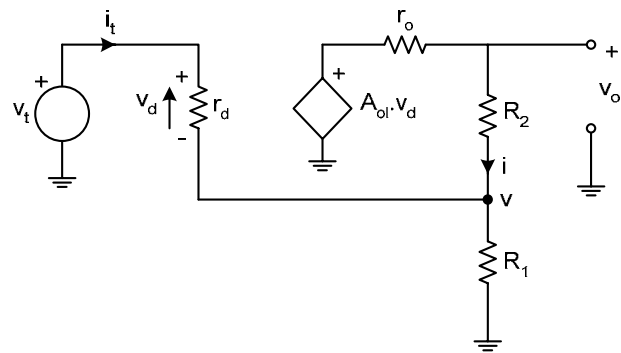
We can write:

$$\begin{aligned} v_t &= i_t r_d + v_d, & v_d &= i_t r_d \\ i_t + i &= v_d / R_1, & i &= [A_{ol} v_d - v] / (r_o + R_2) \end{aligned}$$

Some long-winded algebra should give us

$$R_{in} = \frac{v_t}{i_t} = r_d \left[1 + \frac{A_{ol}}{1 + \frac{R_2 + r_o}{R_1}} \right] + R_1 \parallel (R_2 + r_o)$$

Too complicated !



Make sensible **approximations**:

- 1) $A_{ol} = \text{BIG}$
- 2) 2nd term \ll 1st term
- 3) $r_o \ll R_2$ normally

$$\text{But } \frac{A_{ol}}{1 + \frac{R_2}{R_1}} = \frac{R_1}{R_1 + R_2} A_{ol} = \beta A_{ol} \equiv T (\text{loop gain})$$

$$\text{Then } R_{in} \approx r_d \left[1 + \frac{A_{ol}}{1 + \frac{R_2}{R_1}} \right]$$

$$\therefore R_{in} = r_d [1 + T]$$

Conclusion

Without feedback, input resistance of op-amp was r_d .

Application of negative feedback has increased R_{in} by a factor of $(1 + T)$ - this can be very big!

Another picture of effect of feedback on R_{in}

v_n tracks v_p so $v_p - v_n$ is very small.

$\therefore i_t = \frac{v_p - v_n}{r_d}$ is also very small

$$\text{But } v_n = \beta \cdot v_o = \beta \frac{A_{ol}}{1 + A_{ol}\beta} v_t = \frac{T}{1 + T} v_t$$

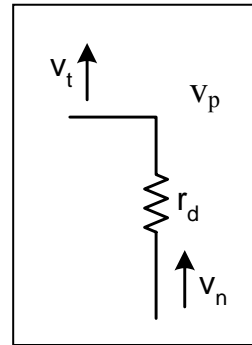
$$\therefore i_t = \frac{v_t - \frac{T}{1 + T} v_t}{r_d} \quad \text{since } v_p = v_t$$

$\rightarrow \frac{v_t}{i_t} = R_{in} = r_d [1 + T]$ as before r_d has been “**bootstrapped up**” by feedback

Input voltage applied to top of r_d

Causes **output** of op-amp to increase,

Causes v_n to increase – circuit stops voltage drops across r_d increasing



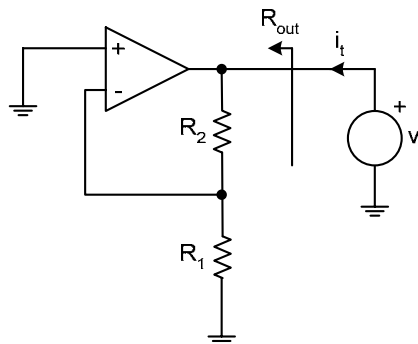
Output Resistance

Suppress input v_g ; apply a test voltage at output

$$\text{Then } R_o \equiv \frac{v_t}{i_t}$$

Use equivalent cct \rightarrow

$$R_o = \frac{r_o}{1 + \frac{A_{ol} + r_o/R_1 + r_o/r_d}{1 + \frac{R_2}{R_1} + \frac{R_2}{r_d}}}$$



With same approximations as before,

$$R_o \approx \frac{r_o}{1 + T}$$

Summary

Output impedance of op-amp **without** feedback is r_o . Feedback **reduces** the output resistance by a factor of $1 + T$

Conclusion

We said before that neg. feedback made R_{in} and R_o tend towards ∞ or 0 (open circuit or short circuit) depending on way feedback is connected.

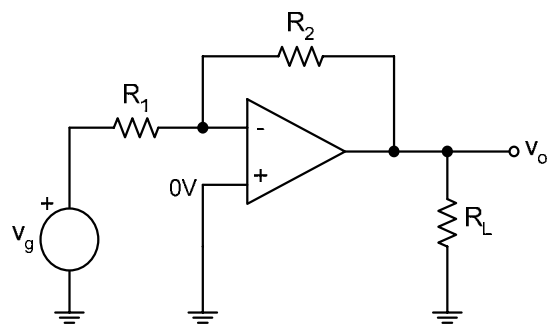
Now we can be more quantitative and give a better estimate for R_{in} and R_o

Generally, the **original** values of the input and output resistances of the **amplifier without feedback** are scaled up or down by the factor $(1 + T)$ when feedback is applied. Whether up or down depends on the topology.

Example: What is the input resistance of the inverting amplifier

- 1) looking into the negative terminal (R_{in}')
- 2) looking from the source, v_g

Solution: The equivalent circuit for the amplifier is shown below.



1) Now $R_i' \equiv \frac{v_i}{i_i} = -\frac{v_d}{i_i}$ also, $i_i = i_{rd} + i_{R2}$

Applying Ohm's Law,

$$i_i = \frac{v_i}{r_d} + \frac{v_i - (-A_{ol}v_i)}{R_2 + r_o}$$

That is,

$$i_i = \frac{v_i}{r_d} + \frac{v_i(1 + A_{ol})}{R_2 + r_o}$$

Which can be written as:

$$\frac{i_i}{v_i} = \frac{1}{r_d} + \frac{1 + A_{ol}}{R_2 + r_o}$$

which can be expressed as: $R_i' \equiv \frac{v_i}{i_i} = r_d \parallel \frac{R_2 + r_o}{1 + A_{ol}}$

Take representative values: $R_2 = 10k$, $R_1 = 1k$ (gain of 10); $r_o \sim 50R$, $A_{ol} = 10^5$, $r_d = 1M\Omega$

Then $R_i' \sim \frac{R_2}{A_{ol}} = \frac{10k}{10^5}$ **$R_i = 0.1\Omega$ which is small - as expected!**

- 2) looking from the source, we see $R_1 + R_i' \sim R_1 = 1k$
that is, the input resistance of this **trans-resistance** amplifier is set by R_1 !

