

ELEC 207 Part B

Control Theory Lecture 12: Frequency Response (2)

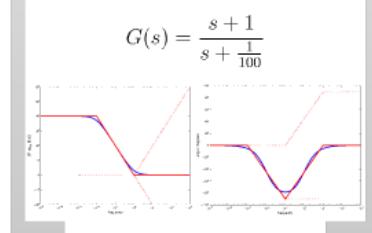
Prof Simon Maskell
CHAD-G68
s.maskell@liverpool.ac.uk
0151 794 4573



Frequency Response Techniques

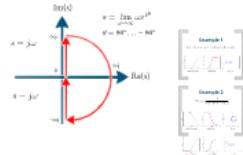
Bode Plot

$$G(s) = \frac{s + 1}{s + \frac{1}{100}}$$



Nyquist Plot

Bode plot shows magnitude and phase of $G(s)$
Nyquist plot shows $G(s)$ on the complex plane



Phase and Gain Margins

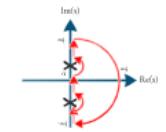


Nyquist Stability Criterion



Nyquist Plot

If there are purely imaginary poles,
the locus takes a little detour around them



This lecture covers:
• Nyquist Stability Criterion
• Phase and gain margins
• Diagrammatic representations using Nyquist plot



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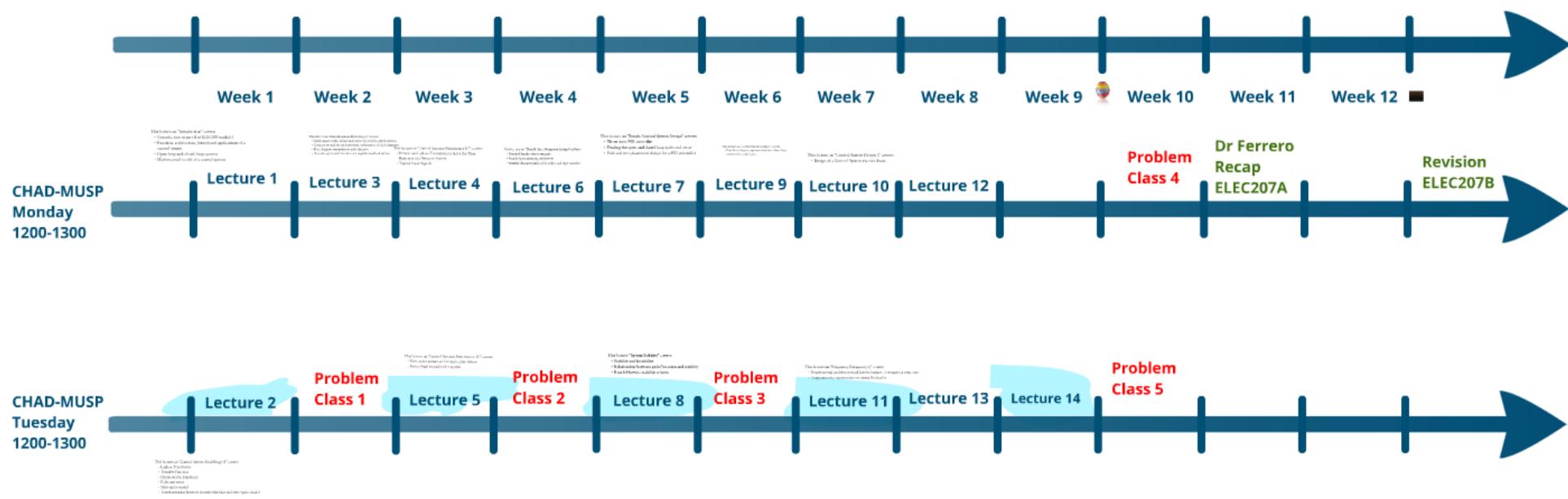
UNIVERSITY OF
LIVERPOOL

This lecture covers:

- Nyquist Stability Criterion
- Phase and gain margins
- Diagrammatic representations using Nyquist plot



ELEC 207B: Timeline





Week 7

Week 8

Week 9



Week 10

Week 11

Week 12



This lecture on "Control System Design 3" covers:

- Design of a Control System via root locus.



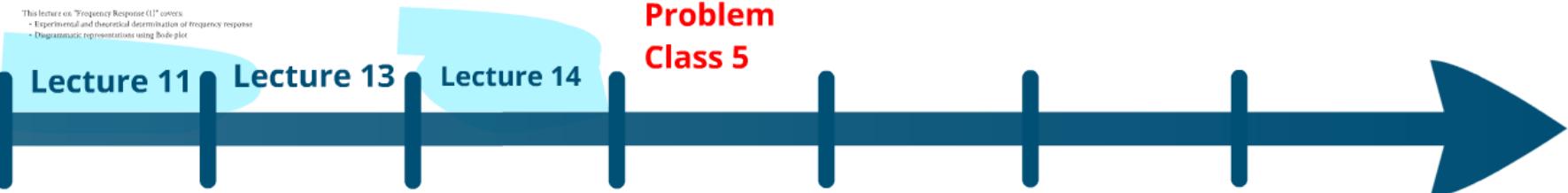
Lecture 10

Lecture 12

Problem
Class 4

Dr Ferrero
Recap
ELEC207A

Revision
ELEC207B



This lecture on "Frequency Response (1)" covers:

- Experimental and theoretical determination of frequency response
- Diagrammatic representations using Bode plot

Lecture 11

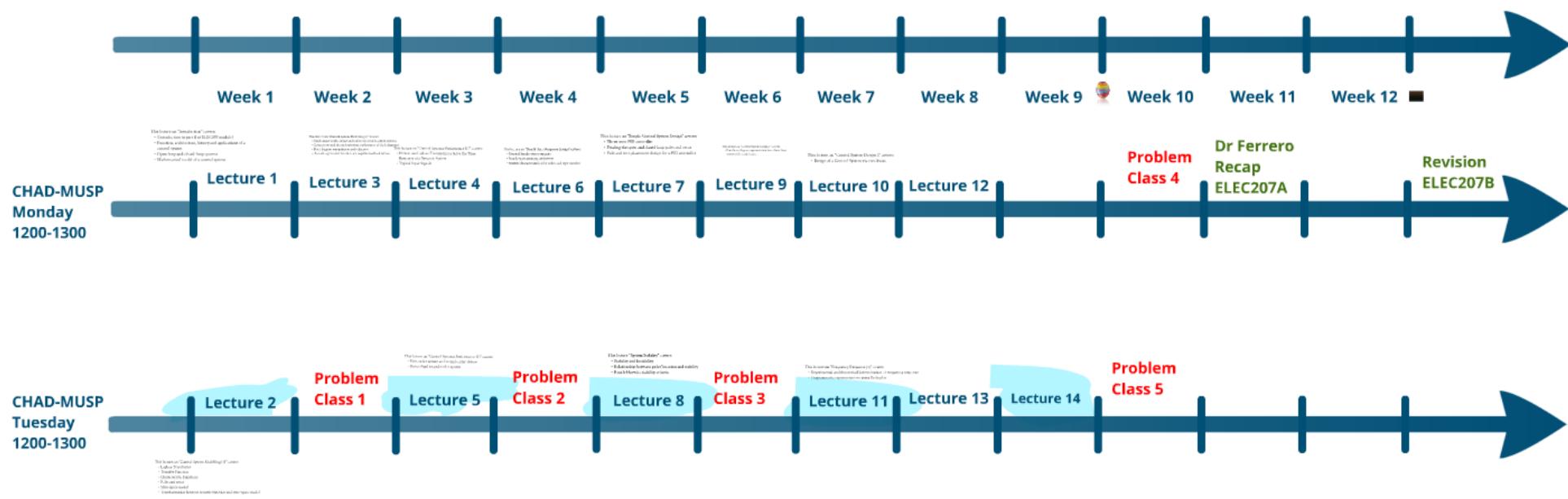
Lecture 13

Lecture 14

Problem
Class 5



ELEC 207B: Timeline



This lecture covers:

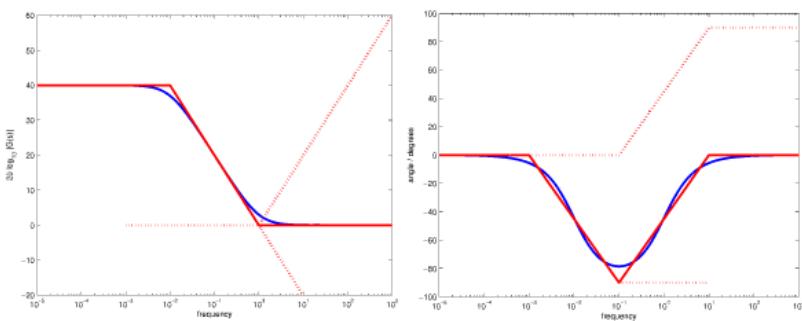
- Nyquist Stability Criterion
- Phase and gain margins
- Diagrammatic representations using Nyquist plot



Frequency Response Techniques

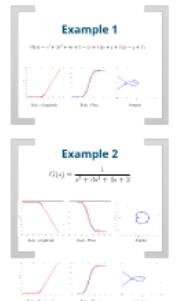
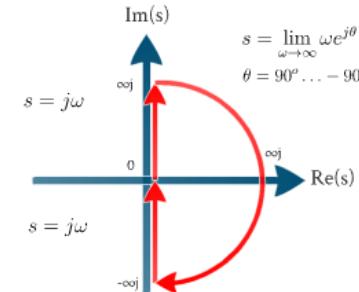
Bode Plot

$$G(s) = \frac{s + 1}{s + \frac{1}{100}}$$



Nyquist Plot

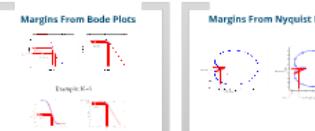
Bode plot shows magnitude and phase of $G(s)$
Nyquist plot shows $G(s)$ on the complex plane



Phase and Gain Margins

We can be interested in **relative stability**, ie how close a system is to going unstable

- **Phase Margin** is the extra phase required to make the phase -180° at a gain of zero dB [or $-20\log_{10}(K)$ if $K=1$];
- **Gain Margin** is the extra gain required to make the gain unity (zero dB) [or $-20\log_{10}(K)$ if $K=1$] at a phase of -180° .



Nyquist Stability Criterion

Relationship Between Closed Loop and Open Loop Poles and Zeros

- Closed loop poles are zeros in open loop zeros.
- Closed loop poles are complicated function of the open loop poles and the open loop zeros.
- Poles of $G(j\omega)$ and $G(j\omega)$ are the same.
- If $G(j\omega)$ is stable, so is $G(j\omega)$.

Loop Around Poles and Zeros

- Consider a closed loop system with poles and zeros.
- If we move the poles around the loop, the closed loop poles will change.
- If we move the zeros around the loop, the closed loop zeros will change.
- If we move both poles and zeros around the loop, the closed loop poles and zeros will change.

Stability Criterion

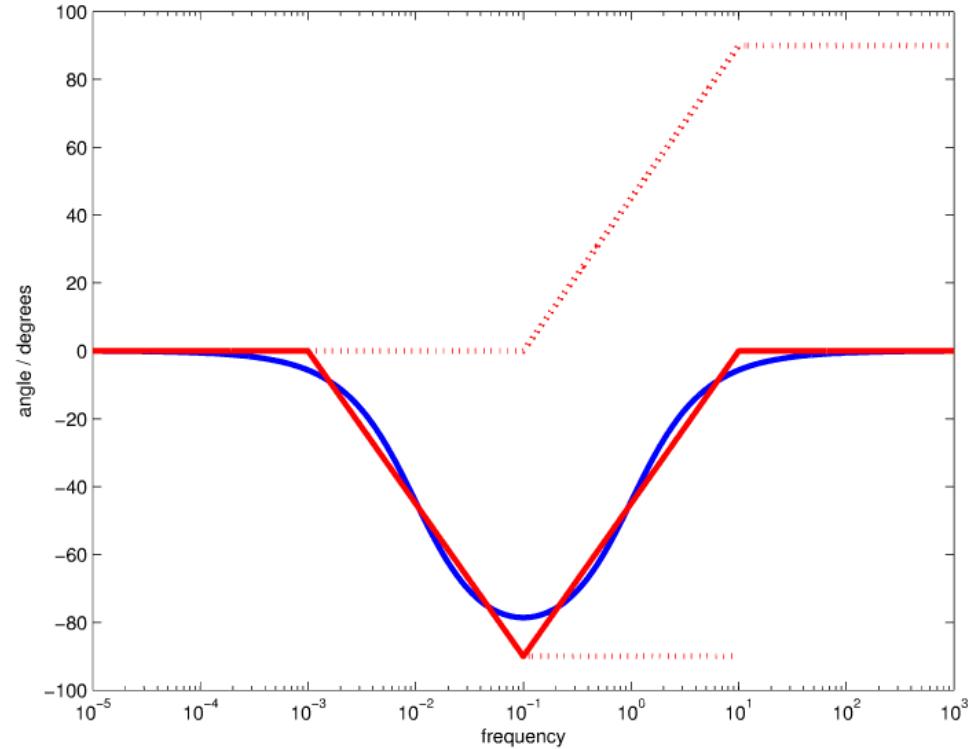
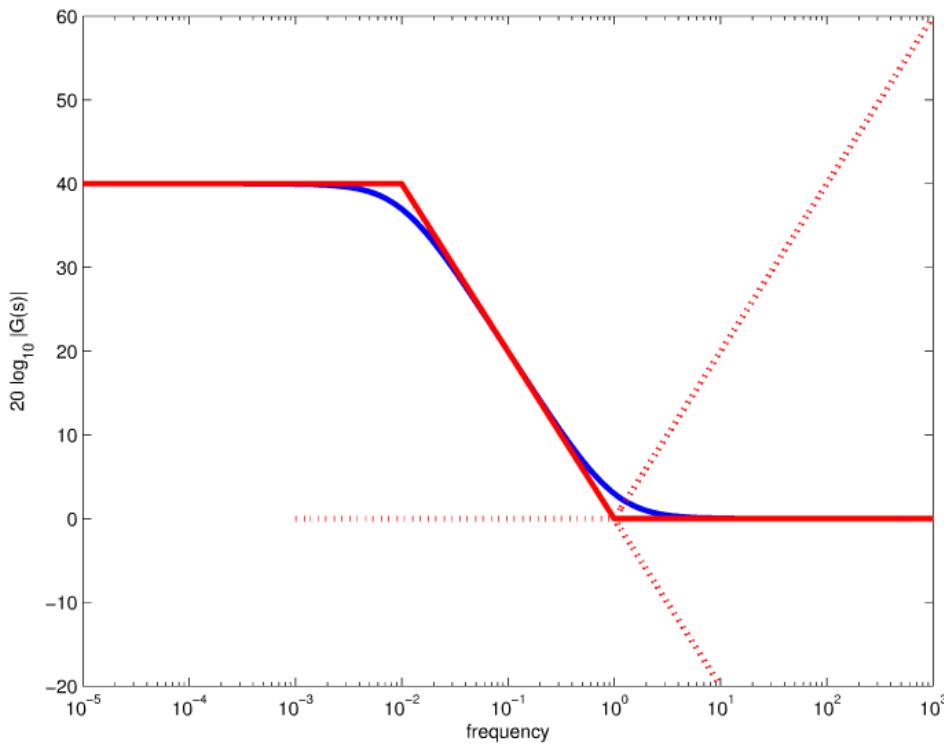
- Closed loop poles in the RHP will cause the closed-loop system to be unstable.

Properties of Shifted Loci

- The locus of $G(j\omega)$ is just locus of $G(j\omega)$ but shifted one unit to the right on the real axis.
- According to the above, the locus of $G(j\omega)$ is the locus of $G(j\omega)$ but the poles are shifted.

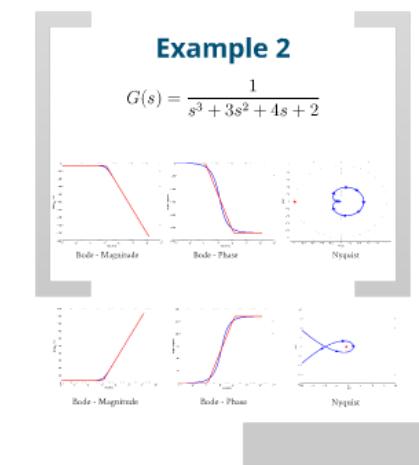
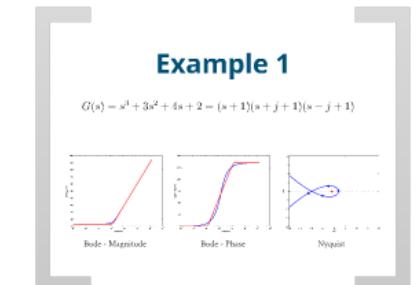
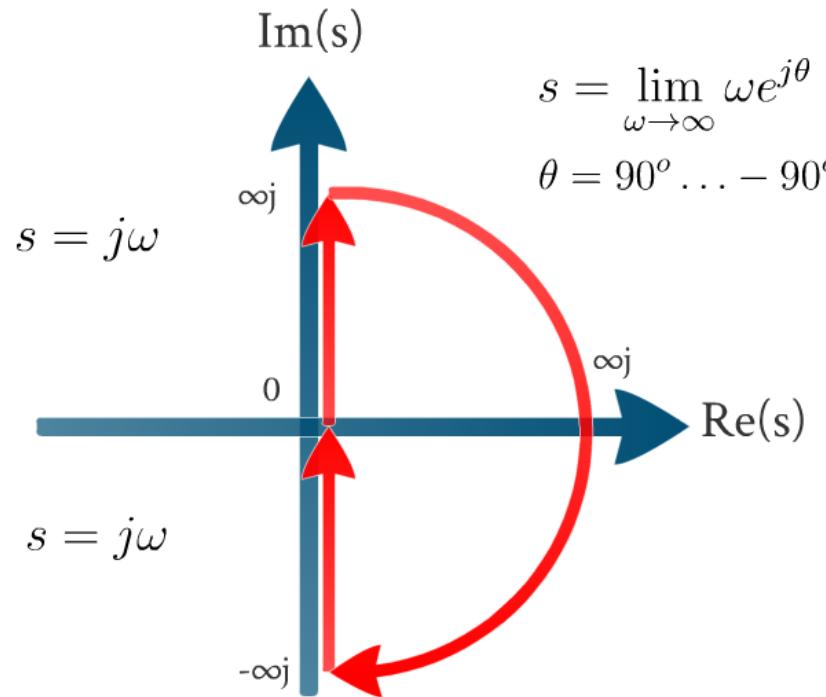
Bode Plot

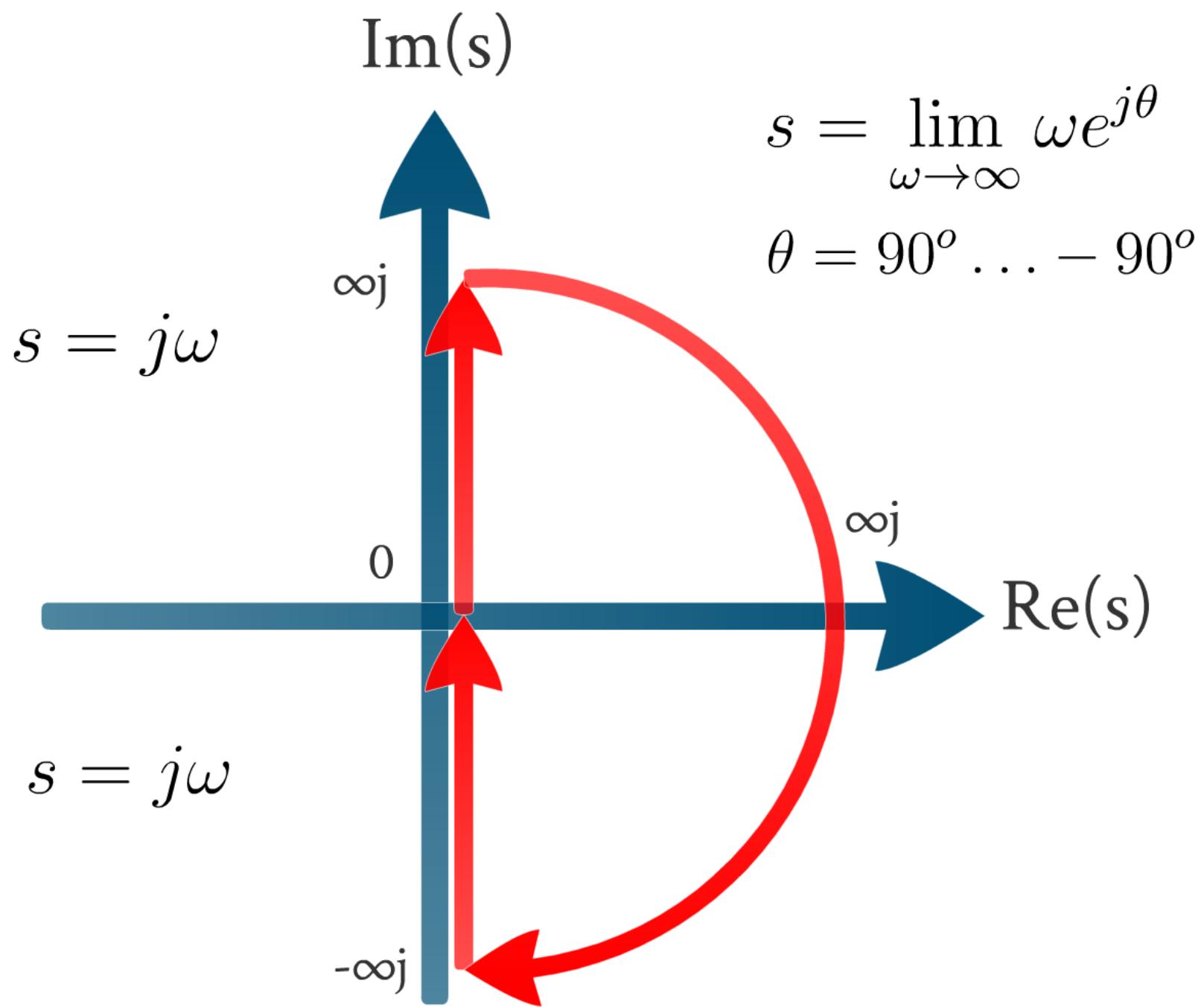
$$G(s) = \frac{s + 1}{s + \frac{1}{100}}$$



Nyquist Plot

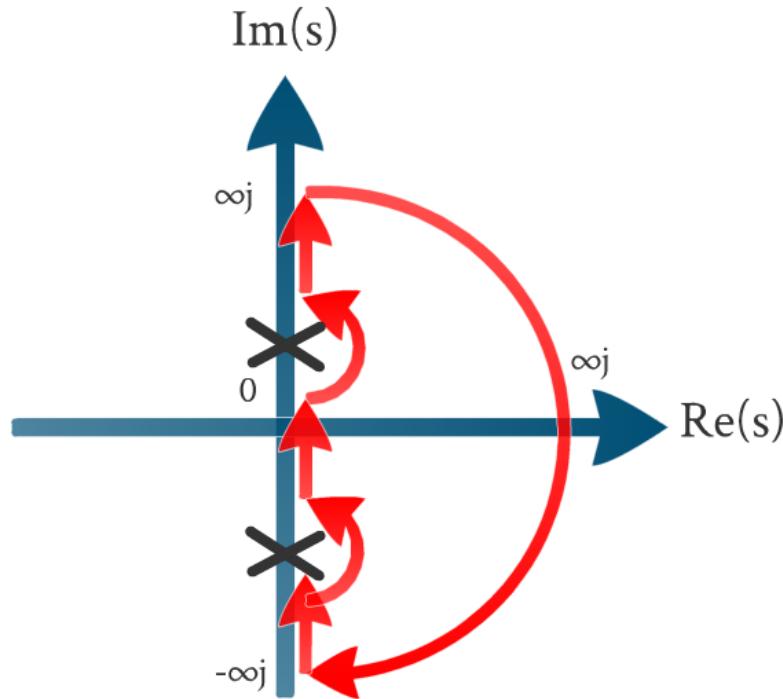
Bode plot shows magnitude and phase of $G(s)$
Nyquist plot shows $G(s)$ on the complex plane

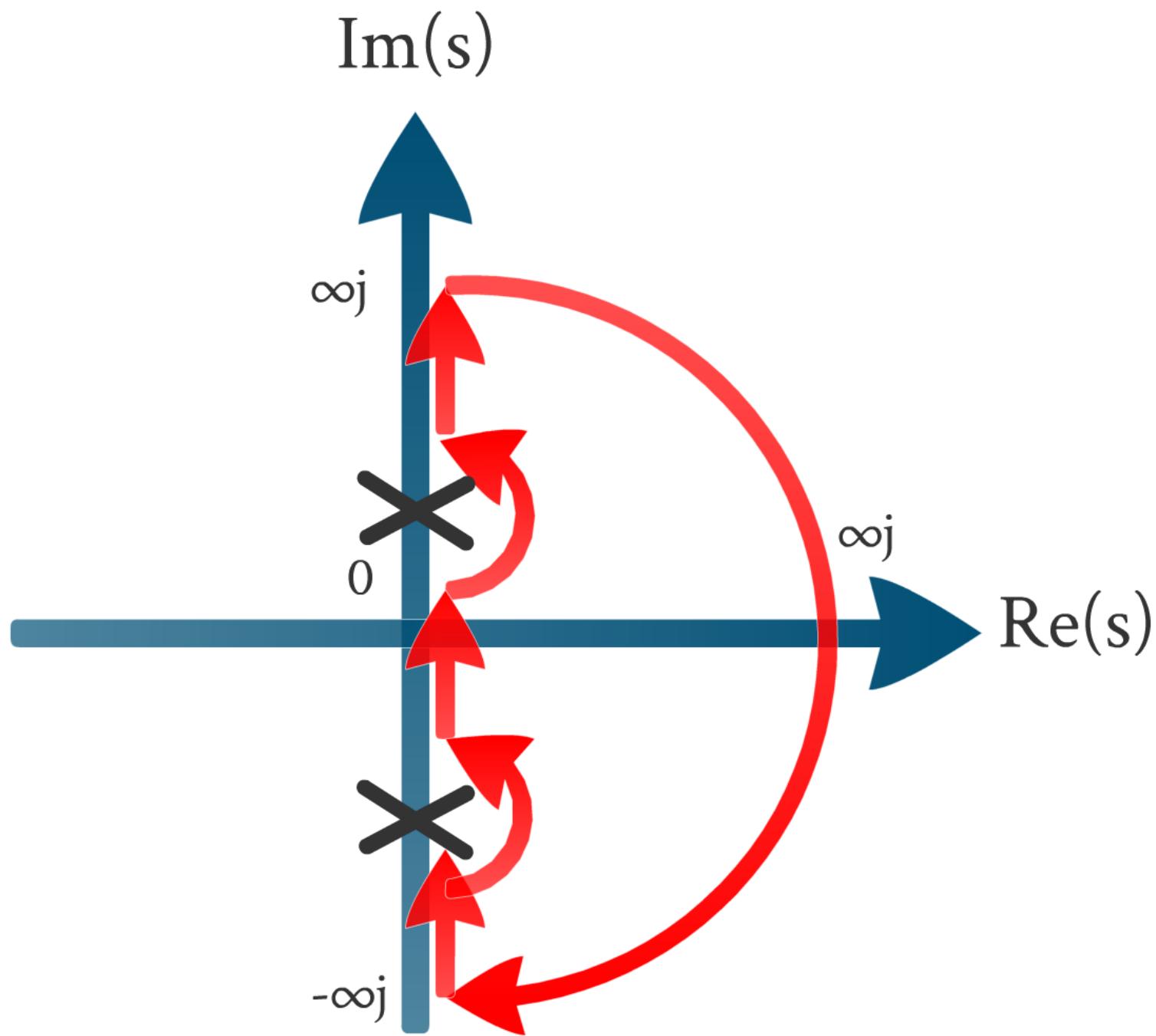




Nyquist Plot

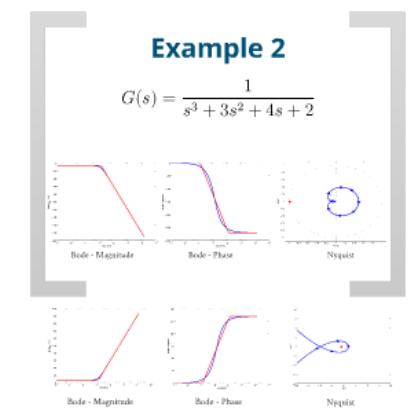
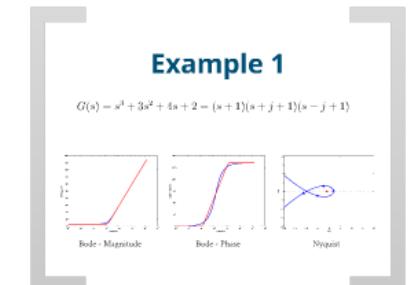
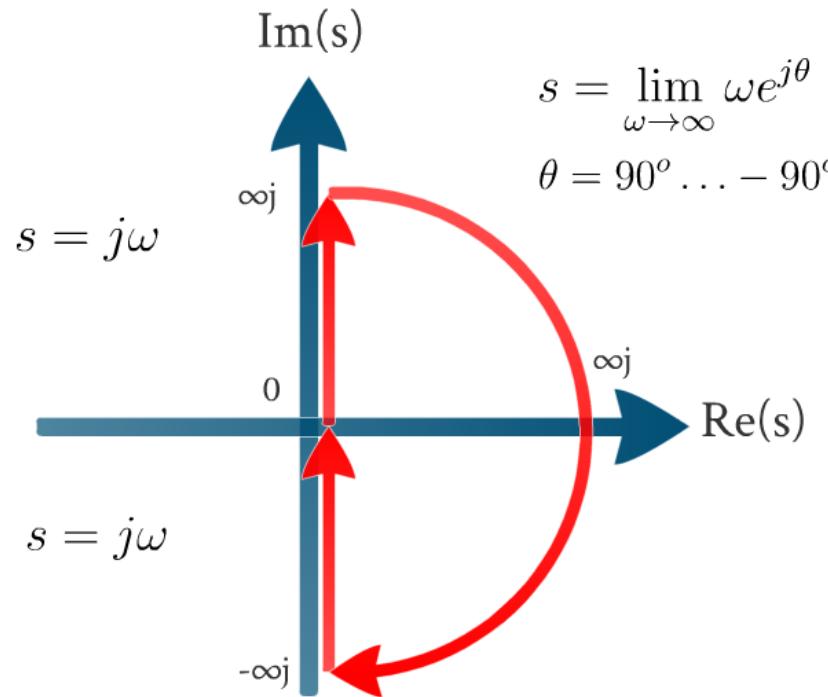
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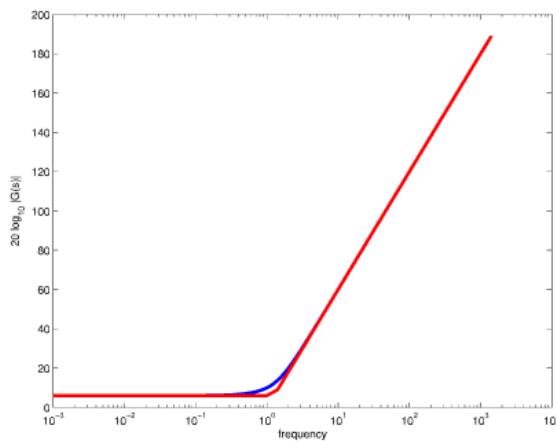
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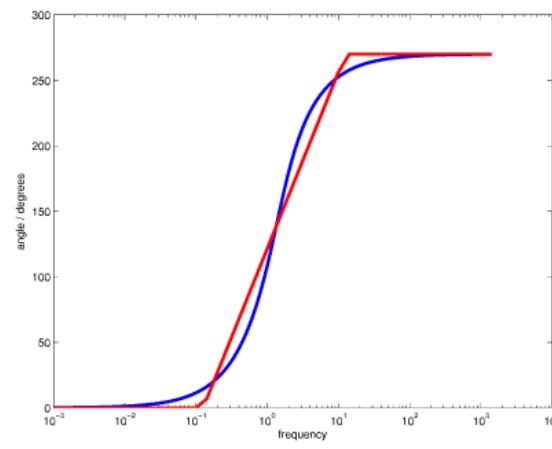


Example 1

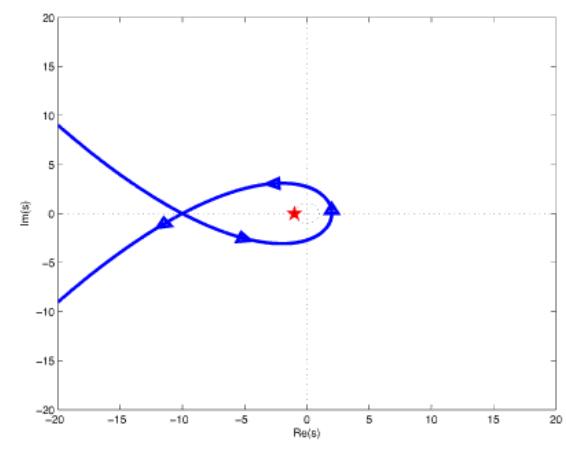
$$G(s) = s^3 + 3s^2 + 4s + 2 = (s + 1)(s + j + 1)(s - j + 1)$$



Bode - Magnitude



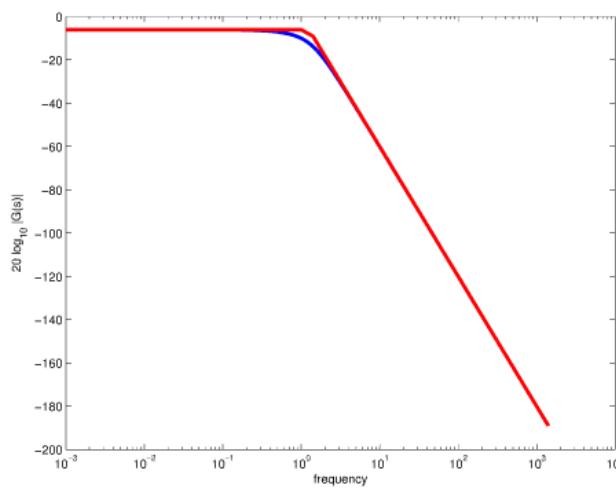
Bode - Phase



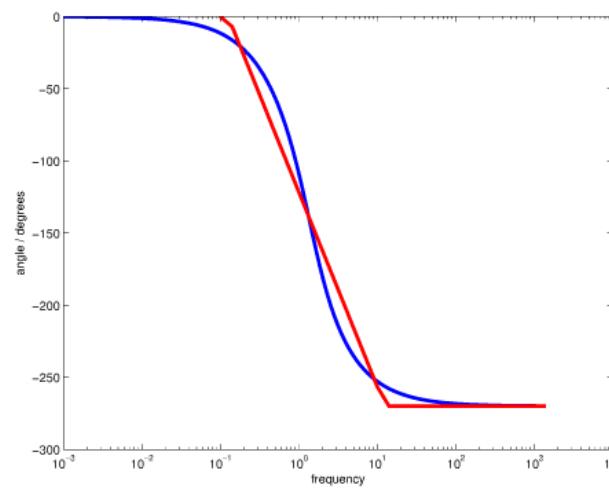
Nyquist

Example 2

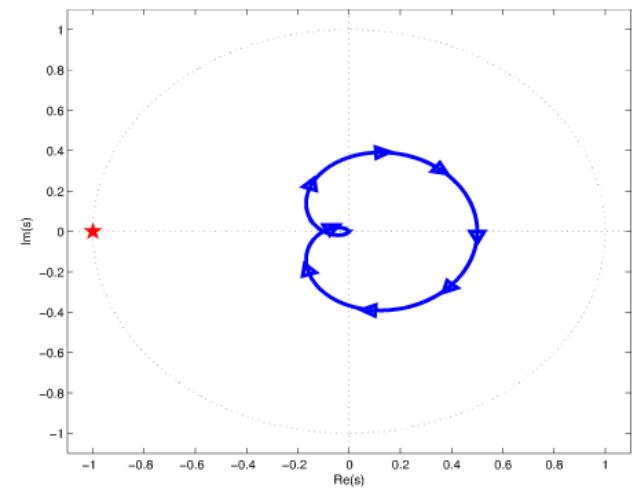
$$G(s) = \frac{1}{s^3 + 3s^2 + 4s + 2}$$



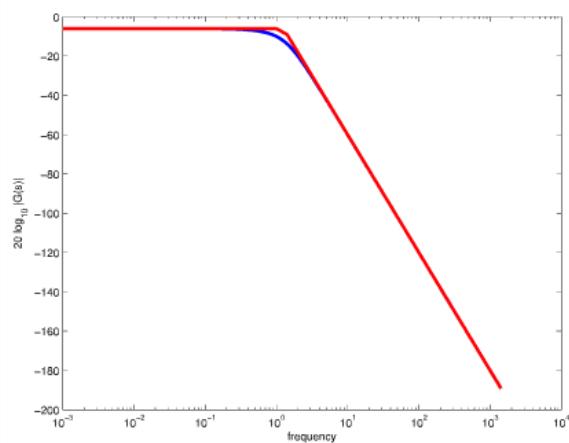
Bode - Magnitude



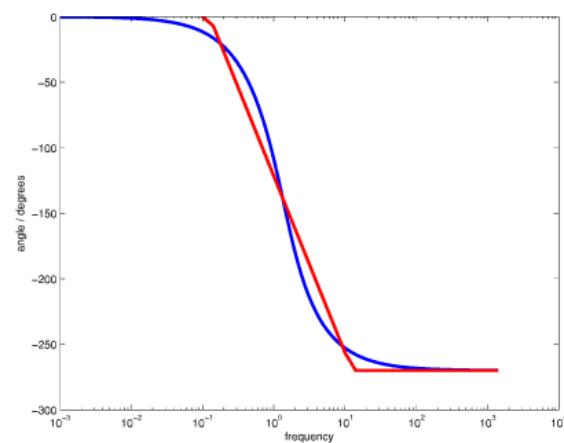
Bode - Phase



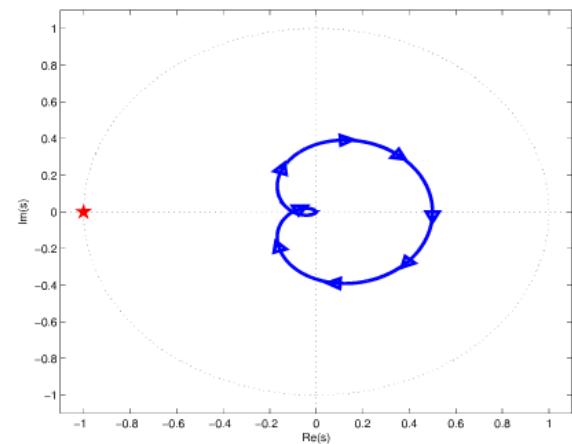
Nyquist



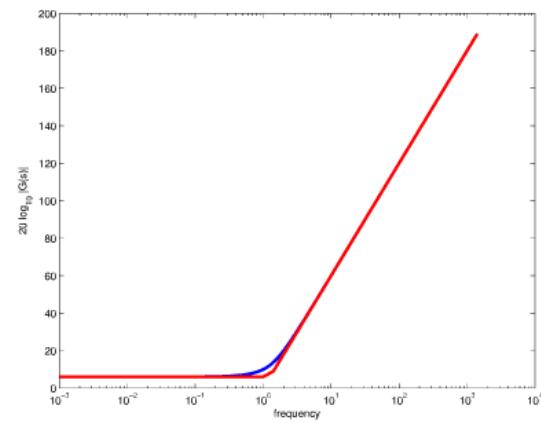
Bode - Magnitude



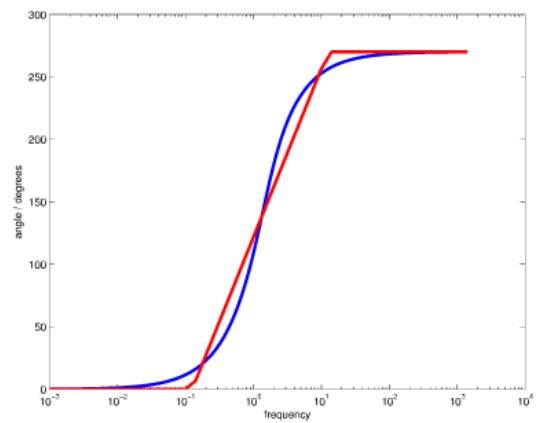
Bode - Phase



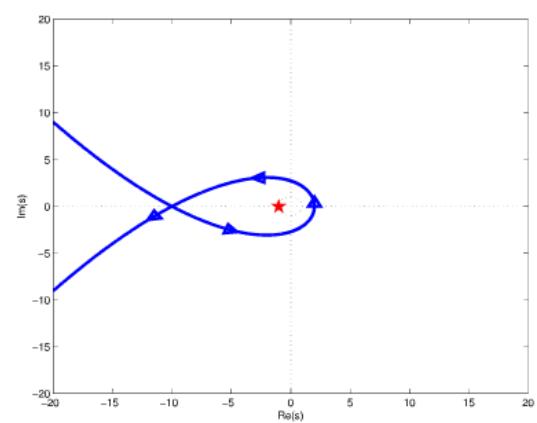
Nyquist



Bode - Magnitude



Bode - Phase

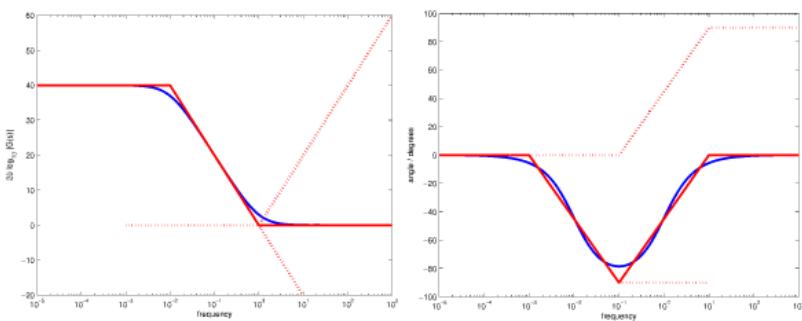


Nyquist

Frequency Response Techniques

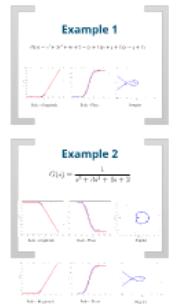
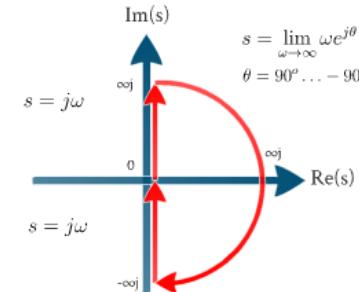
Bode Plot

$$G(s) = \frac{s + 1}{s + \frac{1}{100}}$$



Nyquist Plot

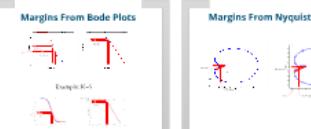
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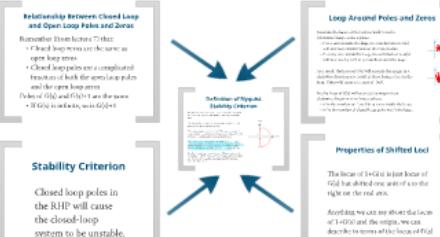
Phase and Gain Margins

We can be interested in **relative stability**, ie how close a system is to going unstable

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Nyquist Stability Criterion



Nyquist Stability Criterion

Relationship Between Closed Loop and Open Loop Poles and Zeros

Remember (from lecture 7) that:

- Closed loop zeros are the same as open loop zeros
- Closed loop poles are a complicated function of both the open loop poles and the open loop zeros

Poles of $G(s)$ and $G(s)+1$ are the same

- If $G(s)$ is infinite, so is $G(s)+1$

Stability Criterion

Closed loop poles in the RHP will cause the closed-loop system to be unstable.

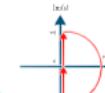
Definition of Nyquist Stability Criterion

We define a loop that goes around the RHP. Note that the loop will enclose all open loop zeros and (possibly) open loop poles in the RHP.

The Nyquist plot is the locus of the open loop transfer function, $G(j\omega)$, as ω goes round the loop.

The number of poles of $G(j\omega)$ in the RHP is equal to the number of zeros of $G(j\omega)$ in the RHP. We know this is the number of times that the Nyquist plot of $G(j\omega)$ intersects $s = -1$ anti-clockwise.

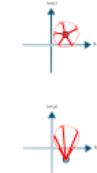
Note that if there is a parameterized plot of K , we are interested in properties of $1+KG(s)$. We know this is the same as the properties of $G(j\omega)$. So we can look back at any trace from $G(j\omega)$ and add $s = -1$.



Loop Around Poles and Zeros

Consider the locus of $G(s)$ as you walk round a (clockwise) loop on the s -plane.

- For a zero outside the loop, its contribution to $G(s)$ will increase and decrease to the initial value.
- For any zero inside the loop, its contribution to $G(s)$ will increase by 360° as you walk around the loop.



As a result, the locus of $G(s)$ will encircle the origin in a clockwise direction as a result of there being a zero in the loop. Poles will cause rotations of -360° .

So, the locus of $G(s)$ will encircle the origin in an clockwise direction $n' - m'$ times, where:

- n' is the number of closed-loop zeros inside the loop;
- m' is the number of closed-loop poles inside the loop.

Properties of Shifted Loci

The locus of $1+G(s)$ is just locus of $G(s)$ but shifted one unit of s to the right on the real axis.

Anything we can say about the locus of $1+G(s)$ and the origin, we can describe in terms of the locus of $G(s)$ but in relation to $s = -1$.

Relationship Between Closed Loop and Open Loop Poles and Zeros

Remember (from lecture 7) that:

- Closed loop zeros are the same as open loop zeros
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Poles of $G(s)$ and $G(s)+1$ are the same

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Stability Criterion

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Loop Around Poles and Zeros

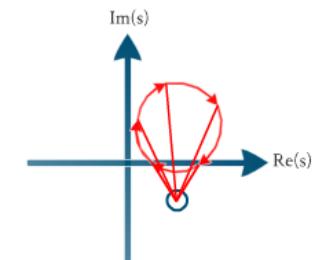
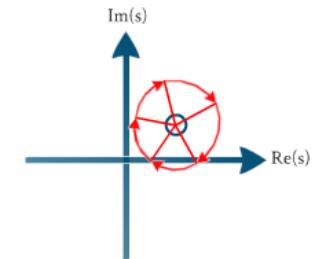
Consider the locus of $G(s)$ as you walk round a (clockwise) loop on the s -plane.

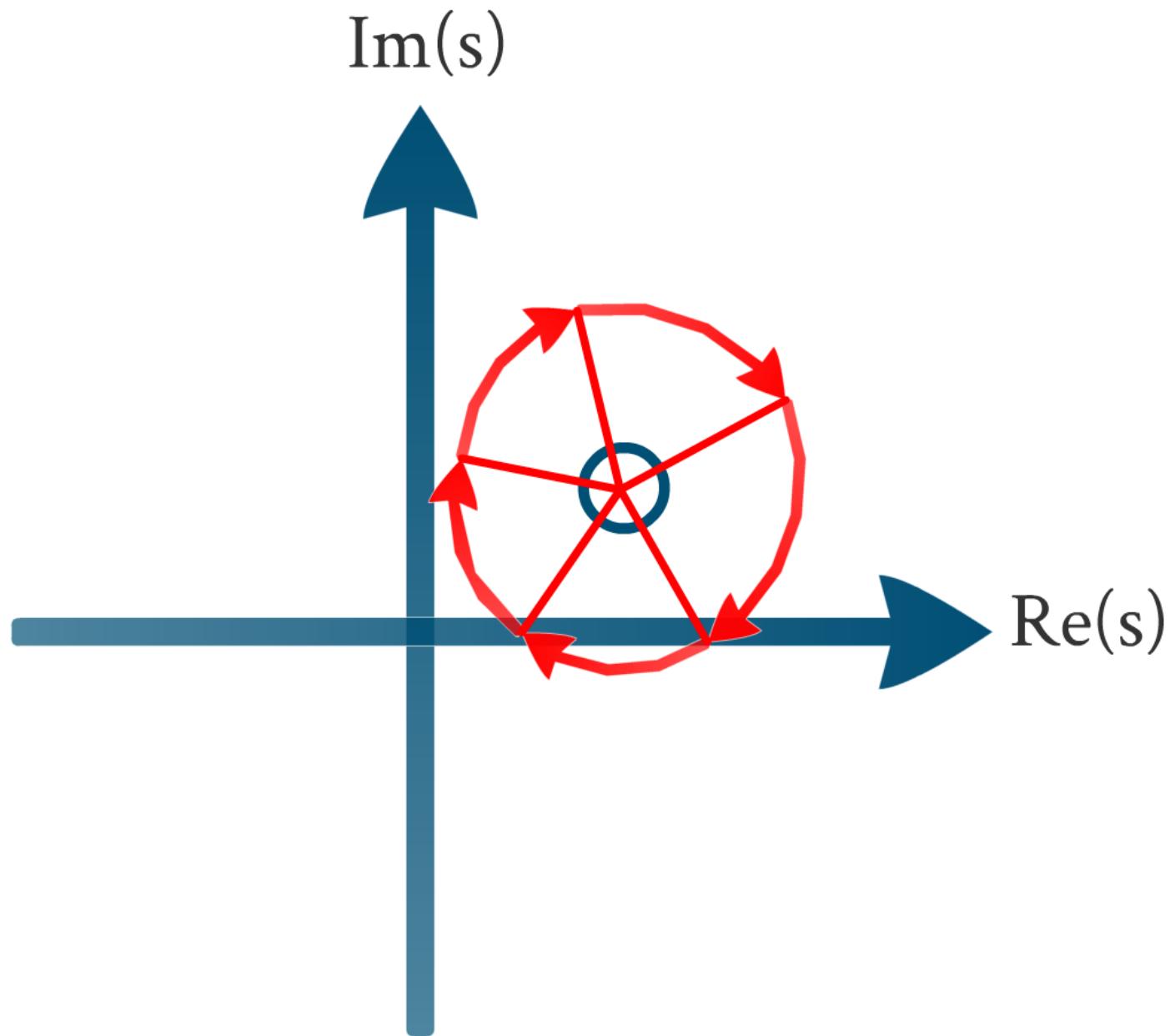
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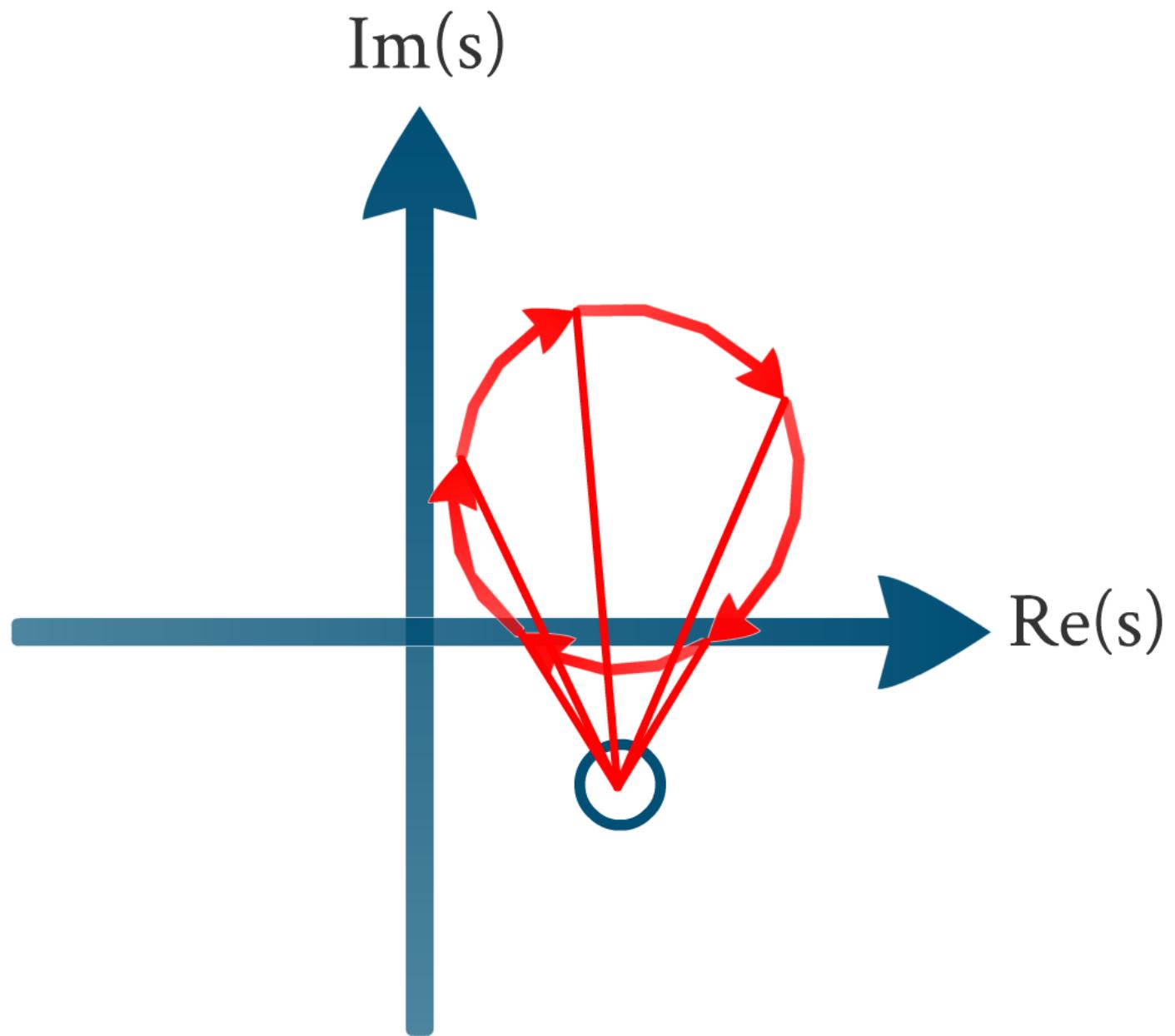
As a result, the locus of $G(s)$ will encircle the origin in a clockwise direction as a result of there being a zero in the loop. Poles will cause rotations of -360° .

So, the locus of $G(s)$ will encircle the origin in an clockwise direction $n'-m'$ times, where:

- n' is the number of closed-loop zeros inside the loop;
- m' is the number of closed-loop poles inside the loop.







Loop Around Poles and Zeros

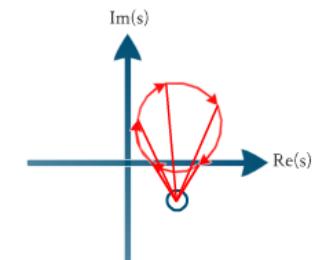
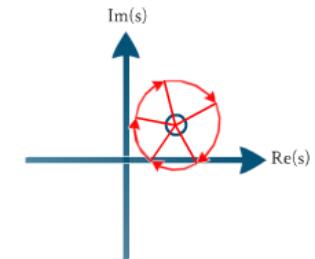
Consider the locus of $G(s)$ as you walk round a (clockwise) loop on the s -plane.

- For a zero outside the loop, its contribution to $G(s)$ will increase and decrease to the initial value.
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As a result, the locus of $G(s)$ will encircle the origin in a clockwise direction as a result of there being a zero in the loop. Poles will cause rotations of -360° .

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- m' is the number of closed-loop poles inside the loop.



Properties of Shifted Loci

The locus of $1+G(s)$ is just locus of $G(s)$ but shifted one unit of s to the right on the real axis.

Anything we can say about the locus of $1+G(s)$ and the origin, we can describe in terms of the locus of $G(s)$ but in relation to $s=-1$.

Nyquist Stability Criterion

Relationship Between Closed Loop and Open Loop Poles and Zeros

Remember (from lecture 7) that:

- Closed loop zeros are the same as open loop zeros
- Closed loop poles are a complicated function of both the open loop poles and the open loop zeros

Poles of $G(s)$ and $G(s)+1$ are the same

- If $G(s)$ is infinite, so is $G(s)+1$

Stability Criterion

Closed loop poles in the RHP will cause the closed-loop system to be unstable.

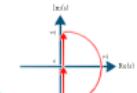
Definition of Nyquist Stability Criterion

We define a loop that goes around the RHP. Note that the loop will enclose all open loop zeros and (possibly) open loop poles in the RHP.

The Nyquist plot is the locus of the open loop transfer function, $G(j\omega)$, as ω goes round the loop.

The number of poles of $G(j\omega)$ in the RHP is equal to the number of zeros of $G(j\omega)$ in the RHP. We know this is the number of times that the Nyquist plot of $G(j\omega)$ intersects $s = -1$ anti-clockwise.

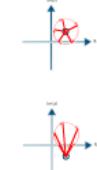
Note that if there is a parameterized plot of $G(s)$, we are interested in properties of $1+G(s)$. We know this is the same as the properties of $G(s)+1$. So we can look back at any trace from $G(s)$ and add $s = -1$.



Loop Around Poles and Zeros

Consider the locus of $G(s)$ as you walk round a (clockwise) loop on the s -plane.

- For a zero outside the loop, its contribution to $G(s)$ will increase and decrease to the initial value.
- For any zero inside the loop, its contribution to $G(s)$ will increase by 360° as you walk around the loop.



As a result, the locus of $G(s)$ will encircle the origin in a clockwise direction as a result of there being a zero in the loop. Poles will cause rotations of -360° .

So, the locus of $G(s)$ will encircle the origin in an clockwise direction $n' - m'$ times, where:

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Properties of Shifted Loci

The locus of $1+G(s)$ is just locus of $G(s)$ but shifted one unit of s to the right on the real axis.

Anything we can say about the locus of $1+G(s)$ and the origin, we can describe in terms of the locus of $G(s)$ but in relation to $s = -1$.

Definition of Nyquist Stability Criterion

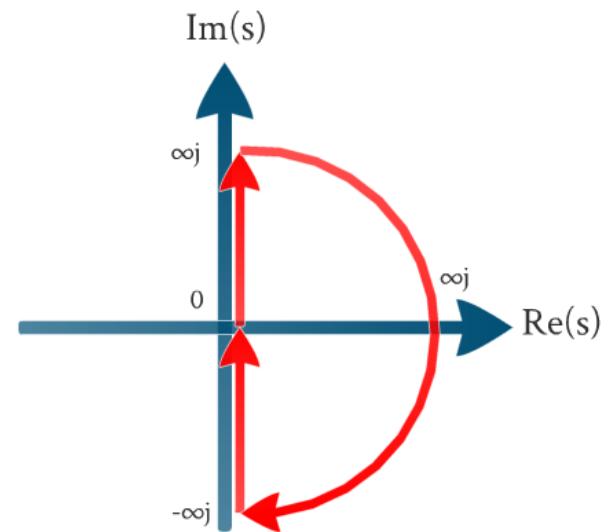
We define a loop that goes around the RHP. Note that the loop will enclose all open loop zeros and (unstable) open loop poles in the RHP.

The Nyquist plot is exactly the locus of the open loop transfer function, $G(s)$, as s goes round this loop.

The number of poles of $G(s)+1$ in the RHP minus the number of zeros of $G(s)+1$ in the RHP is equal to the number of times that $G(s)$ encircles $s = -1$ anticlockwise as s goes round the loop.

The number of **open-loop poles** in the RHP minus the number of **closed-loop poles** in the RHP is equal to the **number of times that the Nyquist plot of $G(s)$ encircles $s = -1$ anticlockwise**.

Note that if there is a parameterised gain of K , we are interested in properties of $1 + KG(s)$. We know this is the same as the properties of $1/K + G(s)$. So, we can look how many times $G(s)$ encircles $s = -1/K$, rather than $s = -1$.



Nyquist Stability Criterion

Relationship Between Closed Loop and Open Loop Poles and Zeros

Remember (from lecture 7) that:

- Closed loop zeros are the same as open loop zeros
- Closed loop poles are a complicated function of both the open loop poles and the open loop zeros

Poles of $G(s)$ and $G(s)+1$ are the same

- If $G(s)$ is infinite, so is $G(s)+1$

Stability Criterion

Closed loop poles in the RHP will cause the closed-loop system to be unstable.

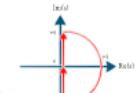
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We define a loop that goes around the RHP. Note that the loop will enclose all open loop zeros and (possibly) open loop poles in the RHP.

The Nyquist plot is the locus of the open loop transfer function, $G(j\omega)$, as ω goes round the loop.

The number of poles of $G(j\omega)$ in the RHP is equal to the number of zeros of $G(j\omega)$ in the RHP. We know this is the number of times that the Nyquist plot of $G(j\omega)$ intersects $s = -1$ anti-clockwise.

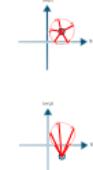
Note that if there is a parameterized plot of K , we are interested in properties of $1+KG(s)$. We know this is the same as the properties of $G(j\omega)$. So we can look back at any trace from $G(j\omega)$ and add $s = -1$.



Loop Around Poles and Zeros

Consider the locus of $G(s)$ as you walk round a (clockwise) loop on the s -plane.

- For a zero outside the loop, its contribution to $G(s)$ will increase and decrease to the initial value.
- For any zero inside the loop, its contribution to $G(s)$ will increase by 360° as you walk around the loop.



As a result, the locus of $G(s)$ will encircle the origin in a clockwise direction as a result of there being a zero in the loop. Poles will cause rotations of -360° .

So, the locus of $G(s)$ will encircle the origin in an clockwise direction $n' - m'$ times, where:

- n' is the number of closed-loop zeros inside the loop;
- m' is the number of closed-loop poles inside the loop.

Properties of Shifted Loci

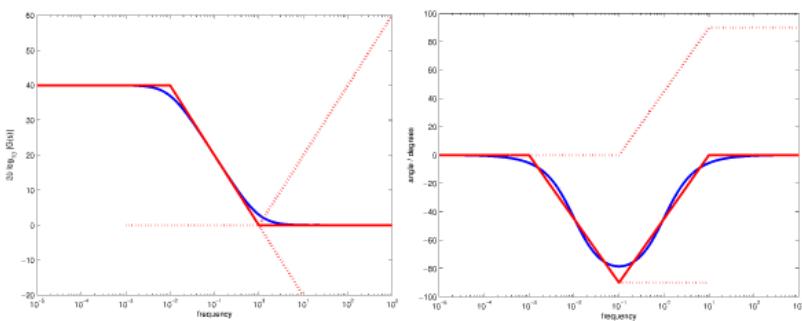
The locus of $1+G(s)$ is just locus of $G(s)$ but shifted one unit of s to the right on the real axis.

Anything we can say about the locus of $1+G(s)$ and the origin, we can describe in terms of the locus of $G(s)$ but in relation to $s = -1$.

Frequency Response Techniques

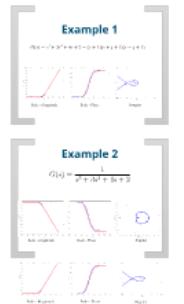
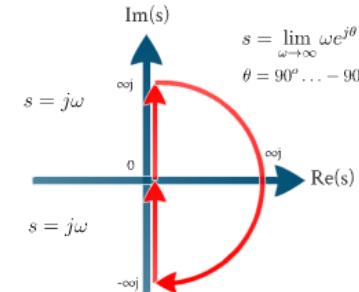
Bode Plot

$$G(s) = \frac{s + 1}{s + \frac{1}{100}}$$



Nyquist Plot

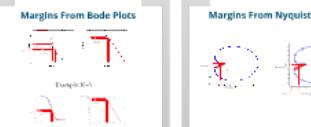
Bode plot shows magnitude and phase of $G(s)$
Nyquist plot shows $G(s)$ on the complex plane



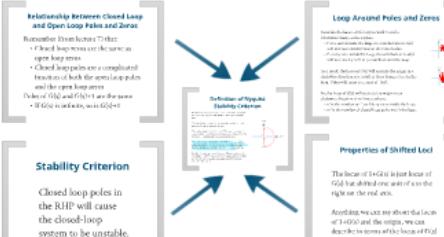
Phase and Gain Margins

We can be interested in **relative stability**, ie how close a system is to going unstable

- **Phase Margin** is the extra phase required to make the phase -180° at a gain of zero dB [or $-20\log_{10}(K)$ if $K=1$];
- **Gain Margin** is the extra gain required to make the gain unity (zero dB) [or $-20\log_{10}(K)$ if $K=1$] at a phase of -180° .



Nyquist Stability Criterion

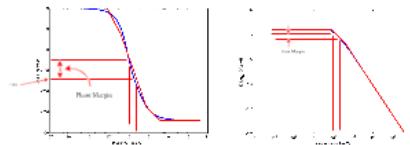


Phase and Gain Margins

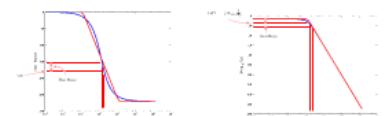
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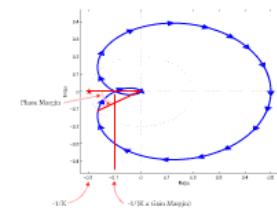
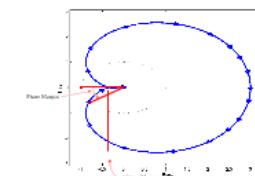
Margins From Bode Plots



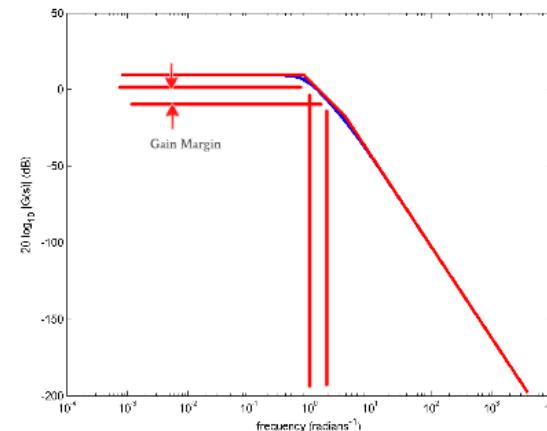
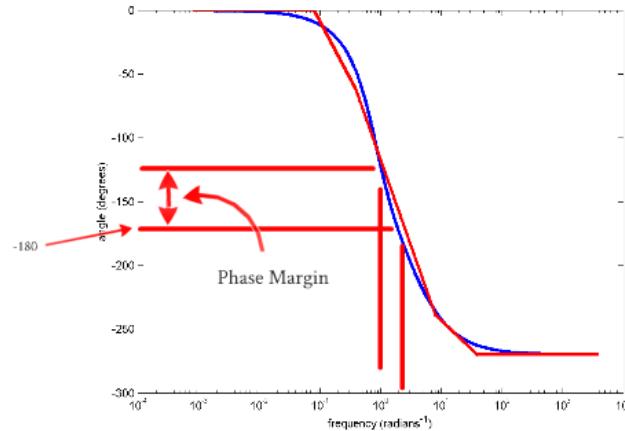
Example: $K=5$



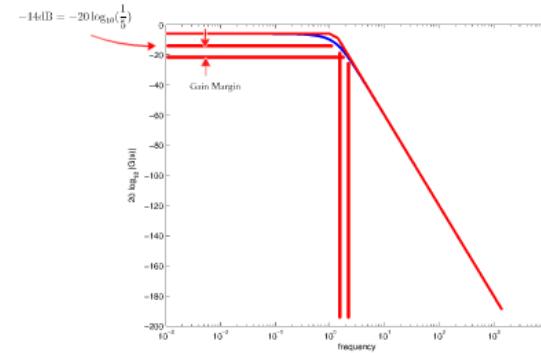
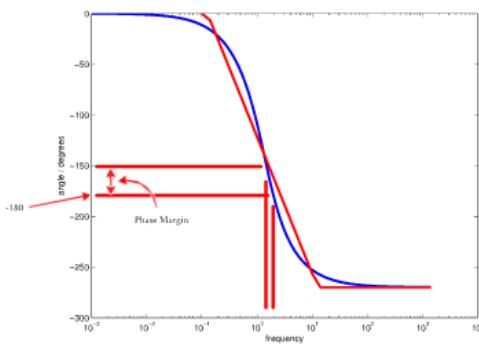
Margins From Nyquist Plots

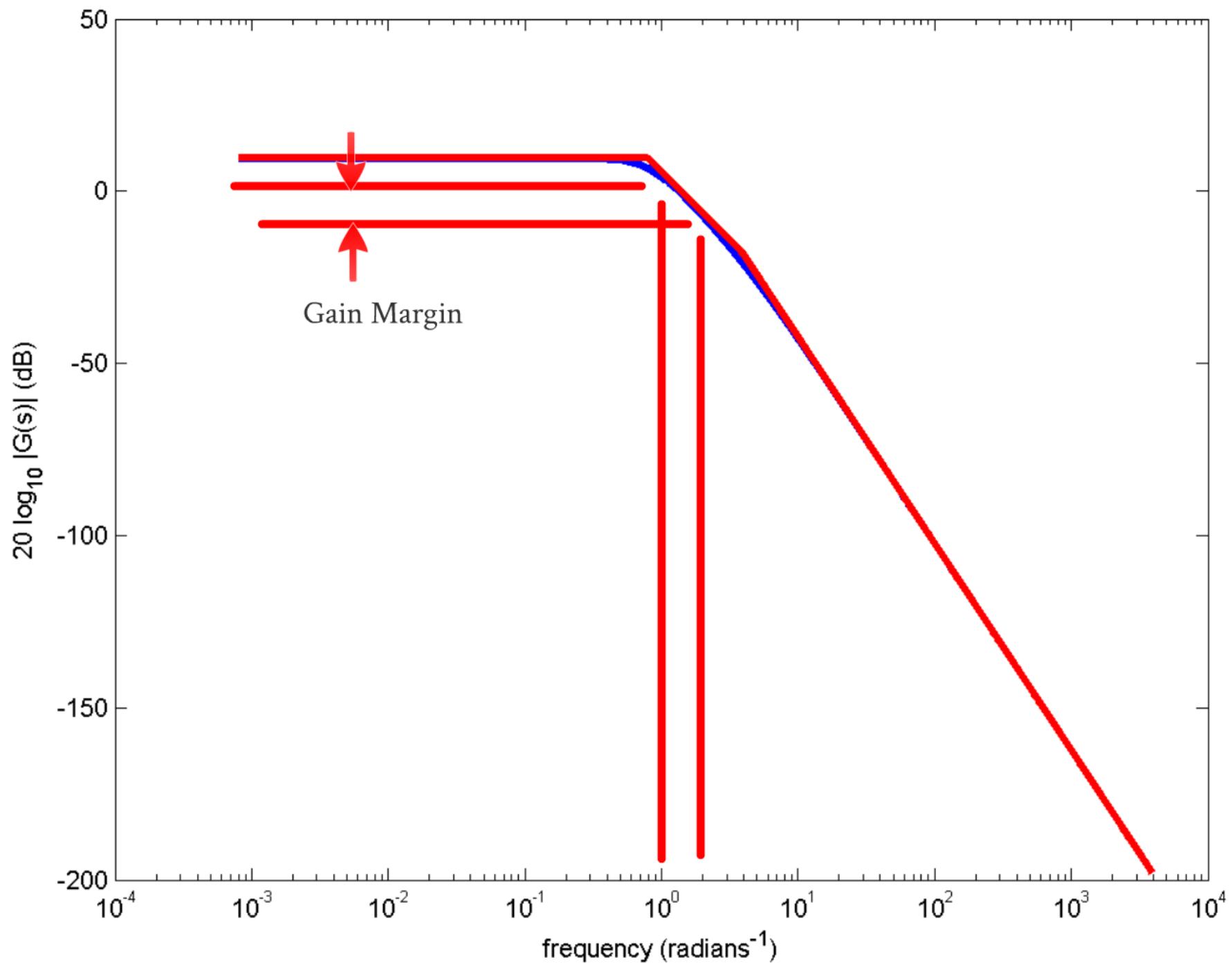


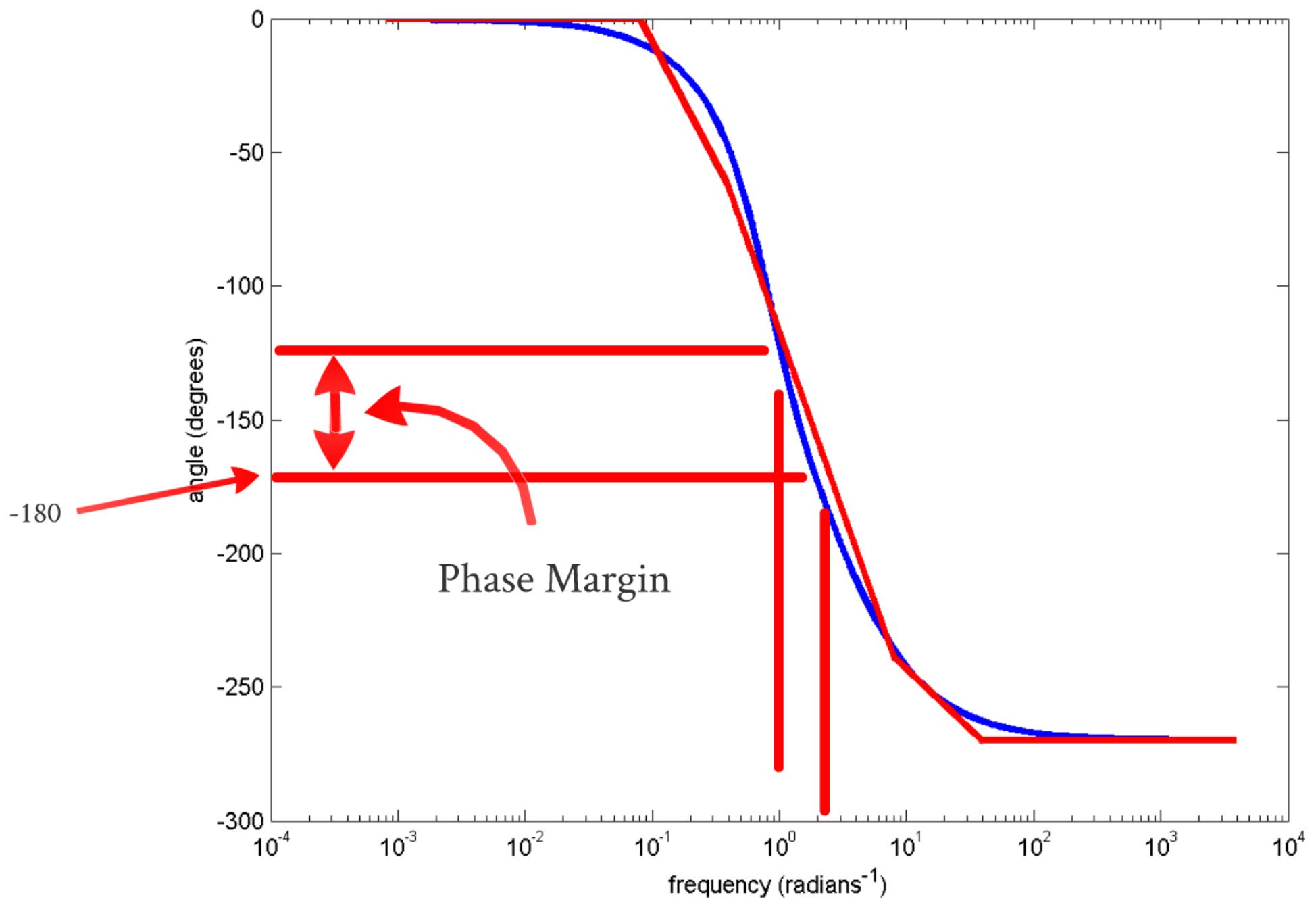
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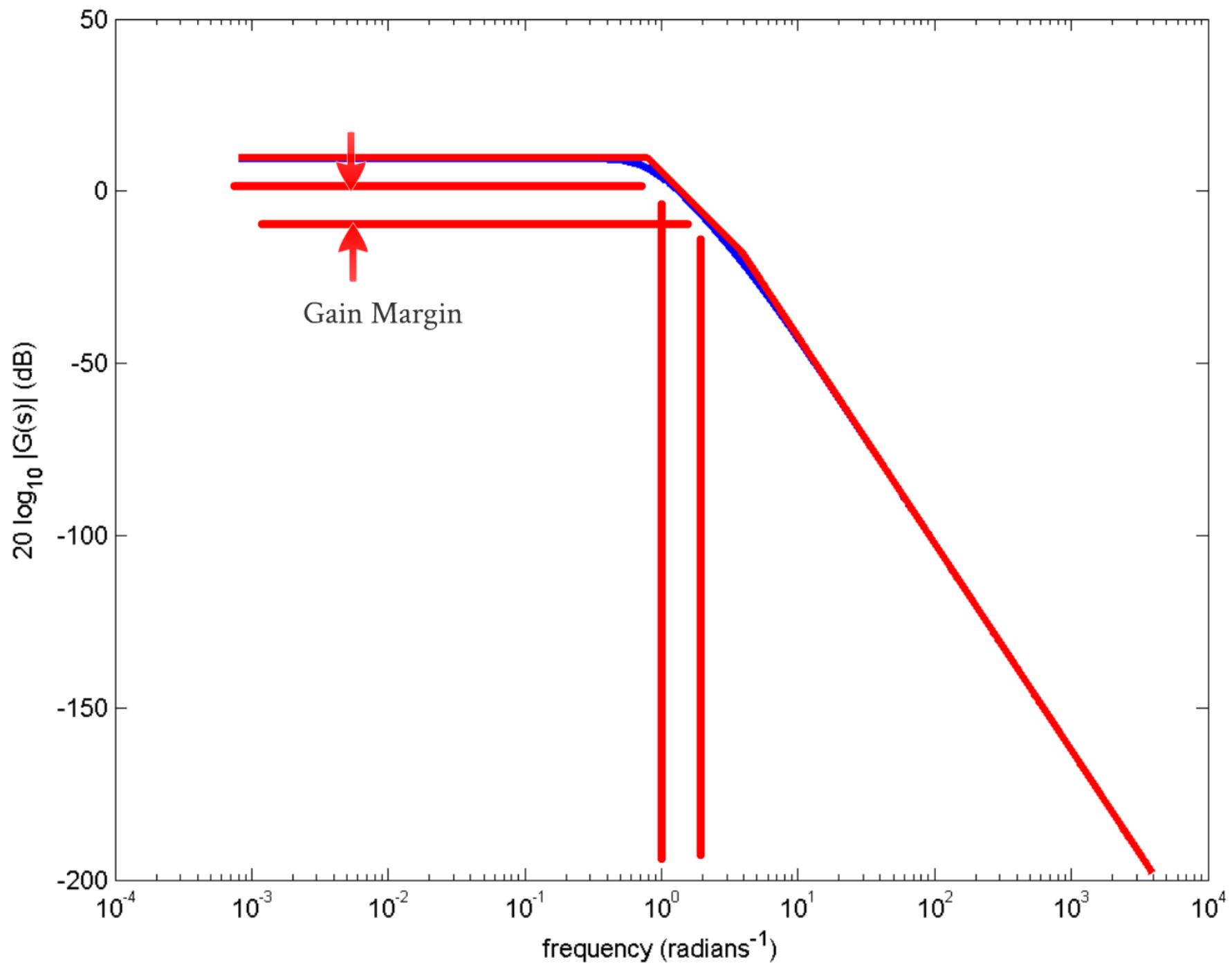


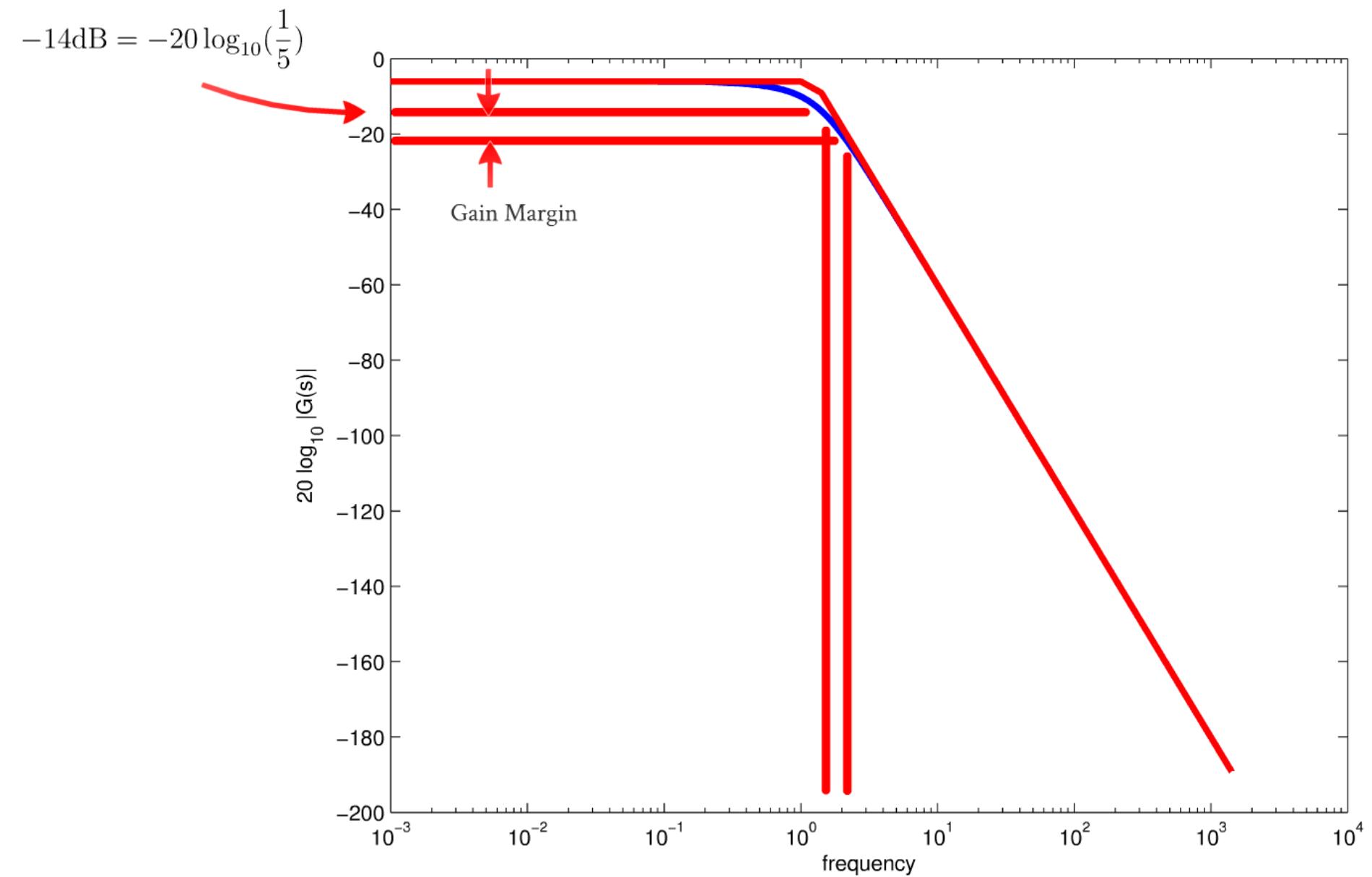
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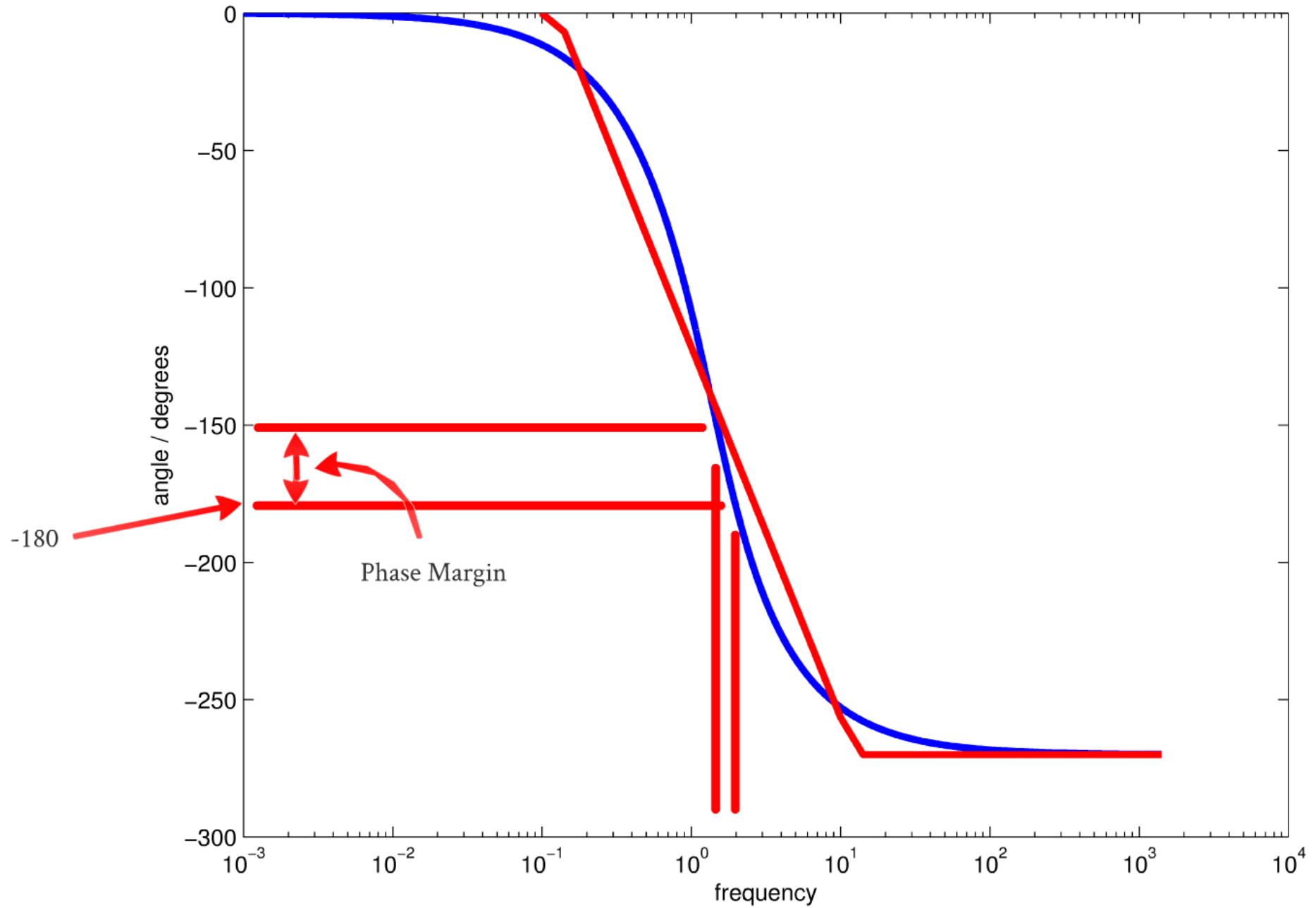


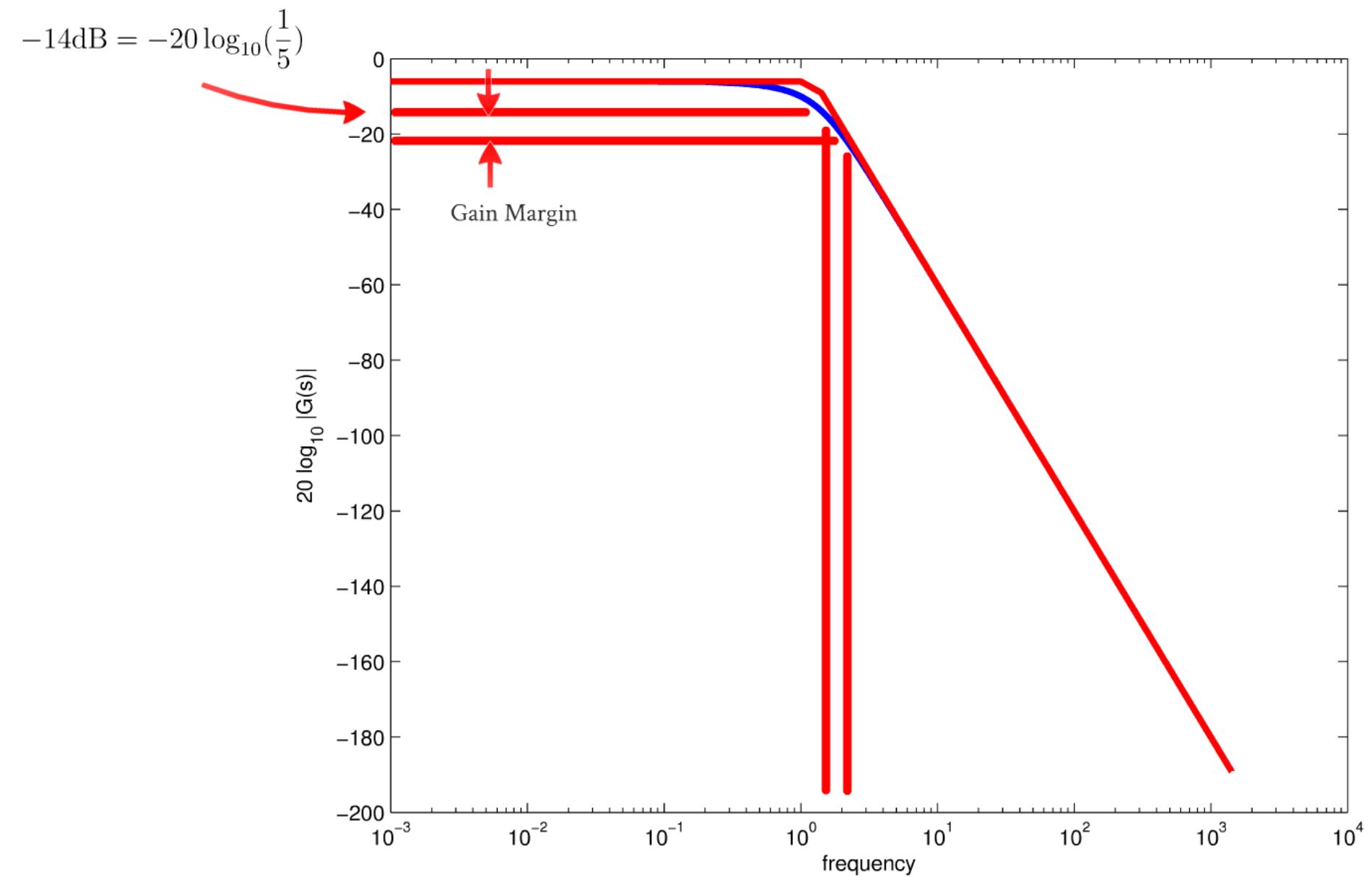




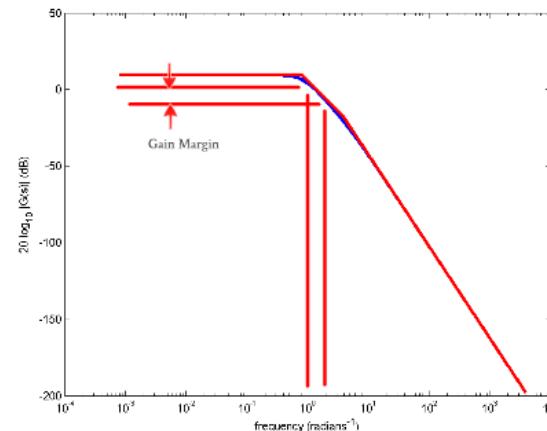
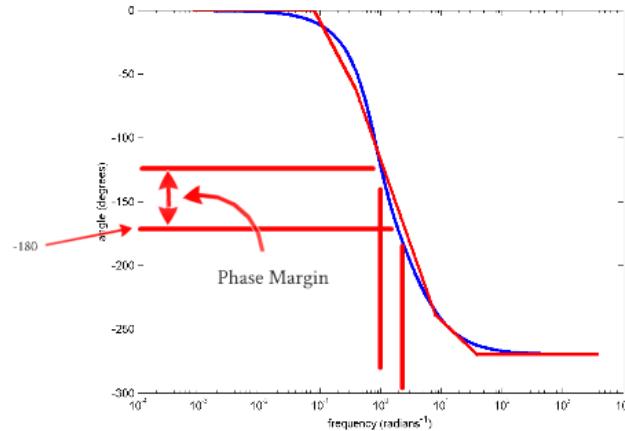




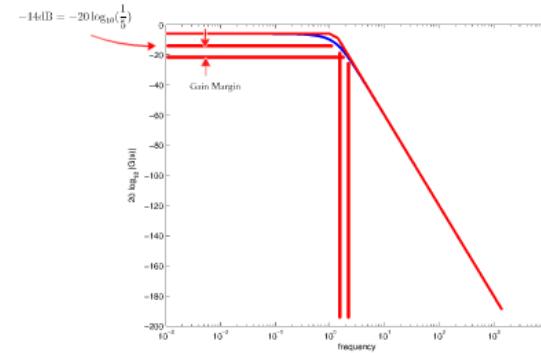
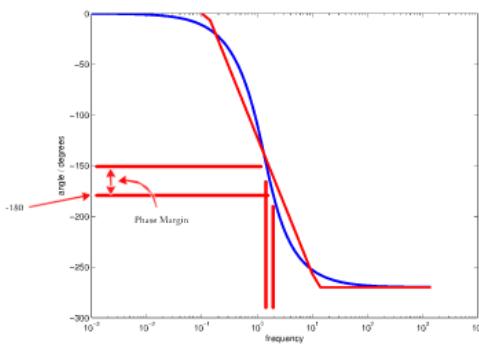




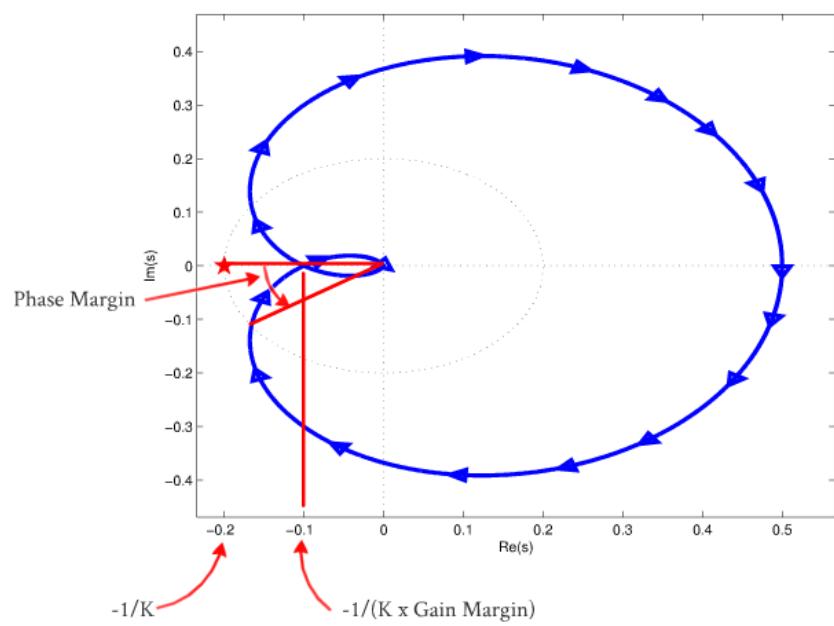
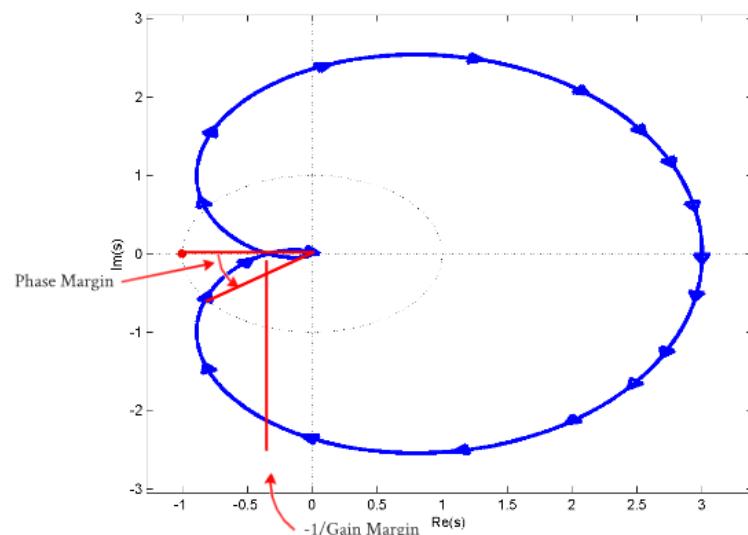
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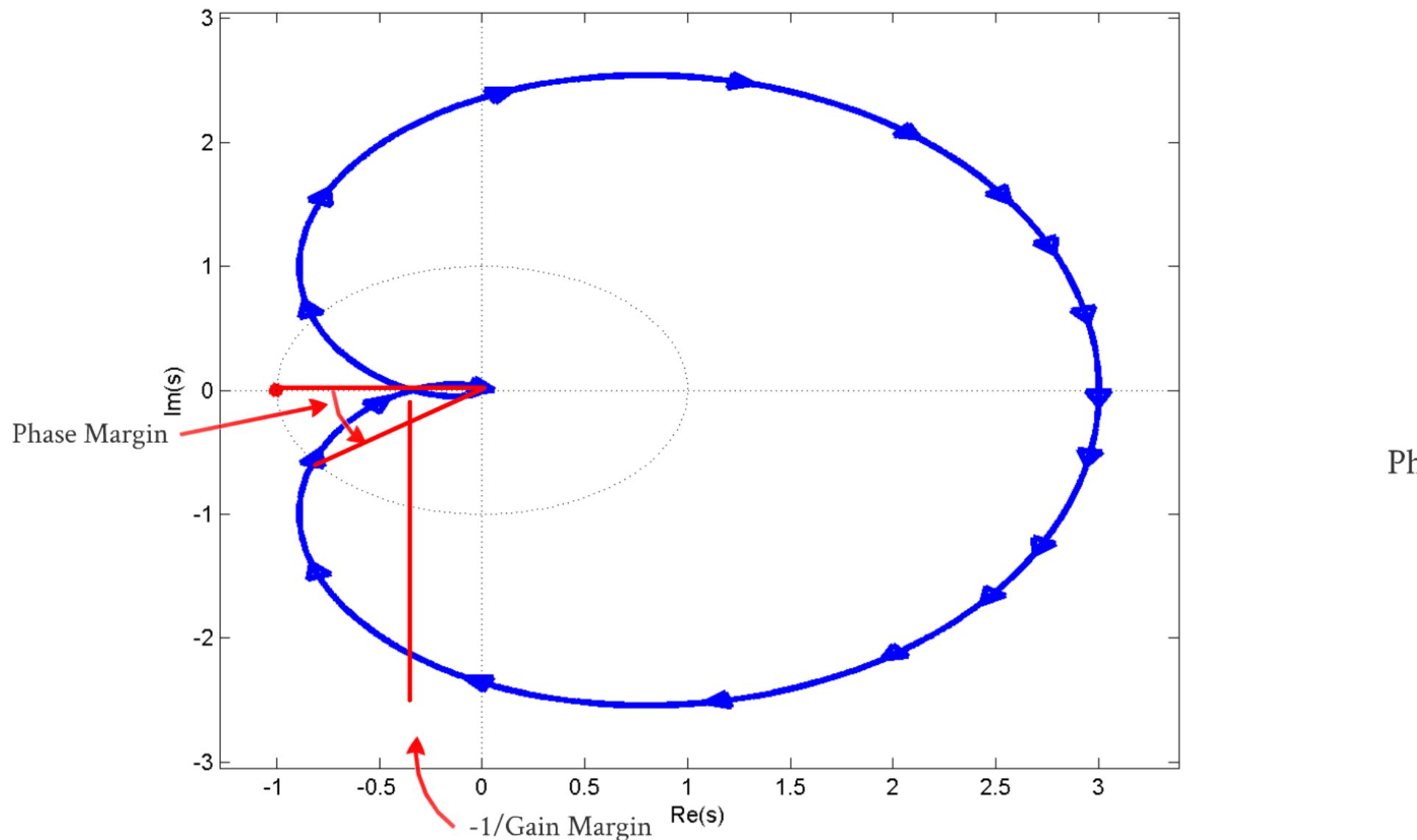


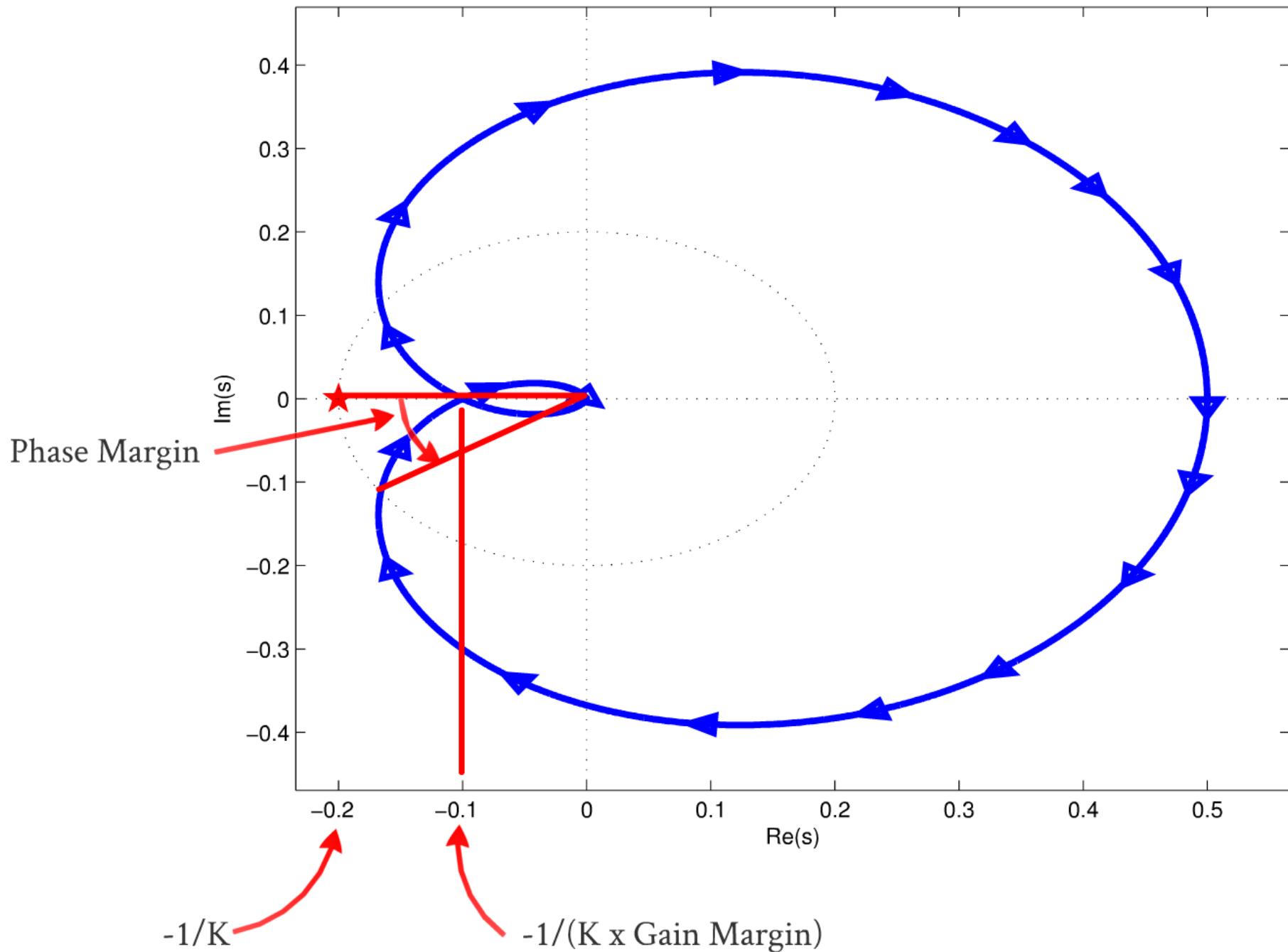
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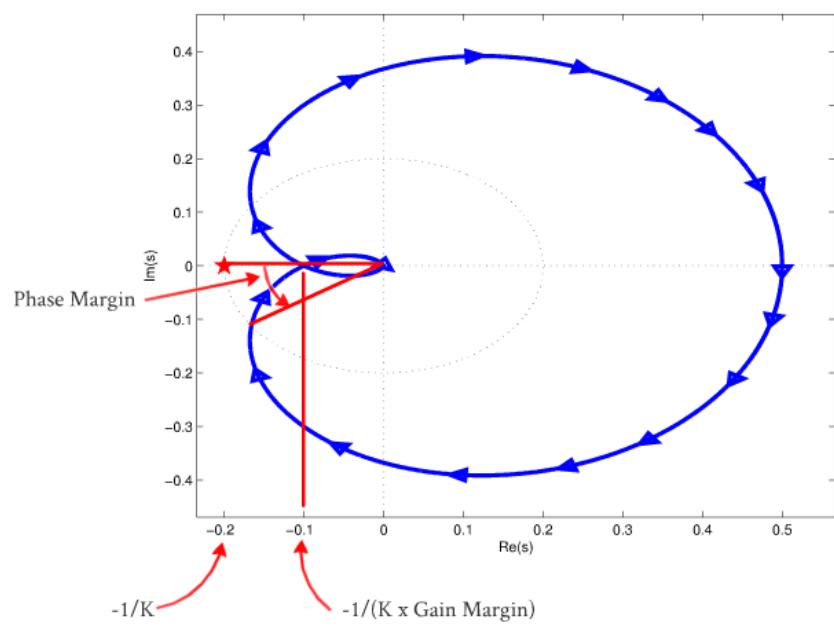
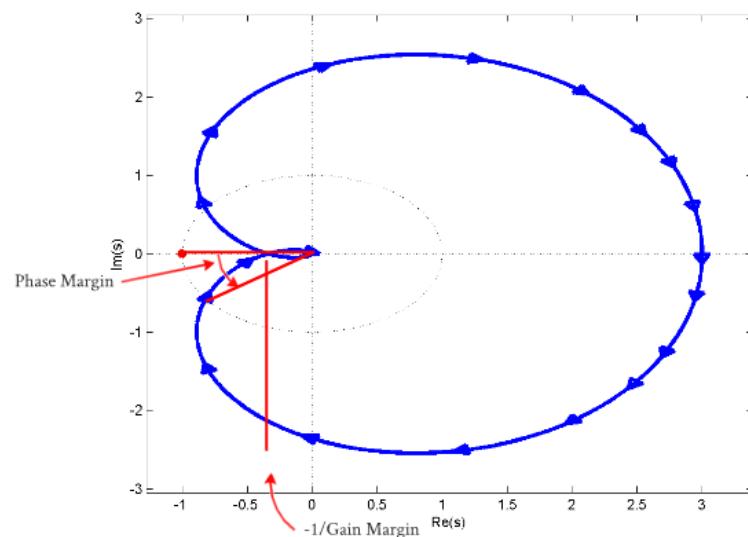
Margins From Nyquist Plots







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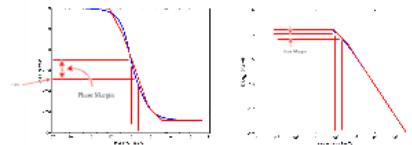


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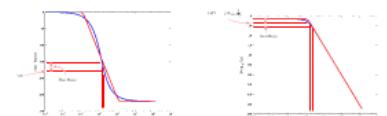
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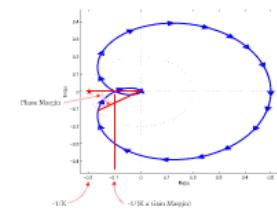
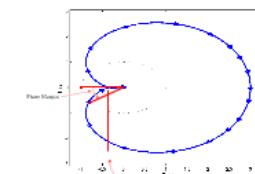
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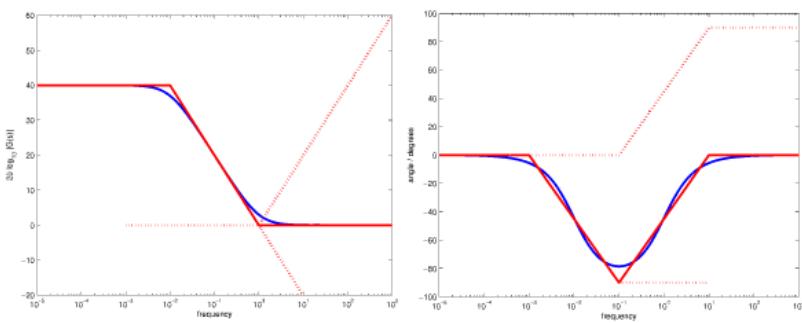
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Frequency Response Techniques

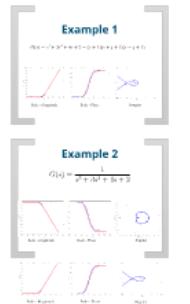
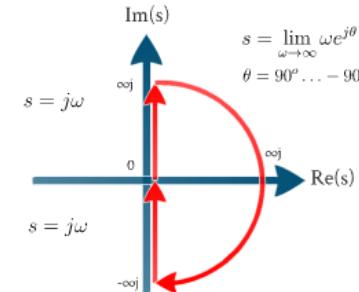
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Nyquist Plot

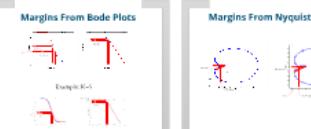
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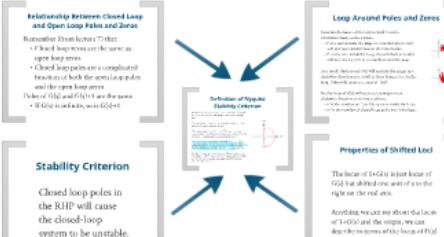
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Nyquist Stability Criterion



This lecture covers:

- Nyquist Stability Criterion
- Phase and gain margins
- Diagrammatic representations using Nyquist plot

