

ELEC 207 Part B

Control Theory Lecture 8: System Stability

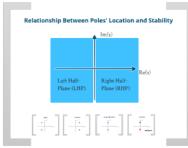
Prof Simon Maskell CHAD-G68 s.maskell@liverpool.ac.uk 0151 794 4573



Stability and Instability

Bounded Input Bounded Output definition of stability:

- · A system is stable if every bounded input yields a bounded output.
- · Systems that aren't stable are unstable (and are pretty useless).

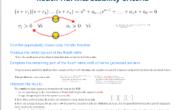


Example 2: Numeric

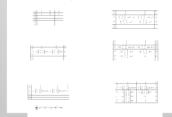
$$s^4 + 6s^3 + 13s^2 + 12s + 4$$

s^4		1			13 12		
83		6			0		
s^2	$-\frac{1}{6}$	1 13 6 12	=11	$-\frac{1}{6}$	1 4 6 0	= 4	0
s^1	$-\frac{1}{11}$	6 12 11 4	$=\frac{108}{11}$		0		0
s^0	$-\frac{11}{108}$		$\begin{vmatrix} 4 \\ 0 \end{vmatrix} = 4$		0		0

Routh-Hurwitz Stability Criteria



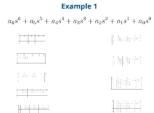
Example 4: Row of Zeros



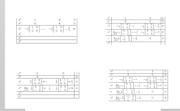
Example 5: Parameterised







Example 3: Zero First Column





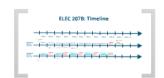
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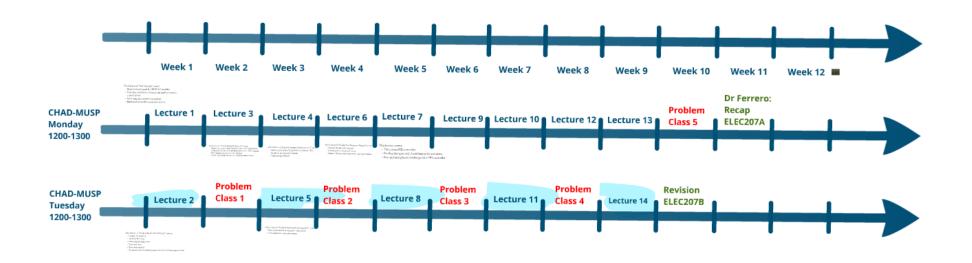


This lecture covers:



- Stability and Instability
- Relationship between poles' location and stability
- · Routh-Hurwitz stability criteria

ELEC 207B: Timeline



This lecture on "Introduction" covers:

- (Introduction to part B of ELEC207 module)
- Function, architecture, history and applications of a control system
- Open-loop and closed-loop systems
- Mathematical model of a control system

Lecture 1

Lecture 2

This lecture on "Control System Modelling (1)" covers:

- Laplace Transforms
- Transfer Function
- Characteristic Equations
- · Poles and zeros
- State-space model
- Transformation between transfer function and state-space model

1 Lecture 3

Lect

This lecture on "Control System Modelling (2)" covers:

- Single-input single-output and multi-input multi-output systems
- · Components and the underpinning mathematics of block diagrams
- Block diagram manipulation and reduction
- · Closed-loop transfer function of a negative feedback system

This lecture on "Control Sy

- How to use Laplace Transcent
 Response of a Dynamic
- Typical Input Signals

3

Lecture 4

This lect

- Gene
- Stead
- Syste

covers:

multi-output systems

atics of block diagrams

feedback system

This lecture on "Control Systems Performance (1)" covers.

- How to use Laplace Transforms to Solve the Time Response of a Dynamic System
- Typical Input Signals

em 1

Lecture 5 Class

This lecture on "Control Systems Performance (2)" covers.

- First-order system and second-order system
- Generalized second-order system

Lecture 6

Lec

This lecture on "Steady-State Response Design" covers:

- General Steady-state response
- Steady-state accuracy and errors
- System Characterization by order and type number

This lecture

- Three ter
- Finding t
- Pole and

6 Lecture 7 Lecture 7

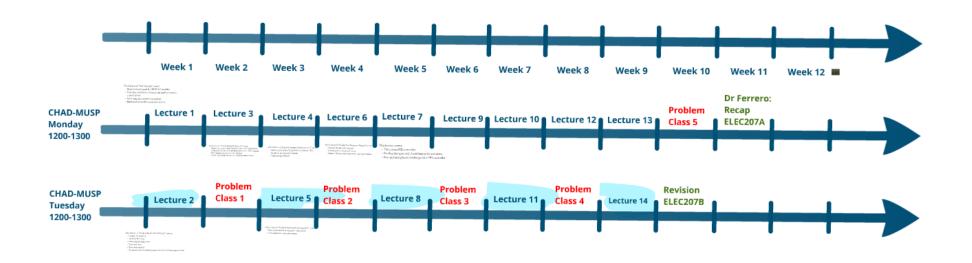
gn" covers:

e number

This lecture covers:

- Three term PID controller
- Finding the open and closed loop poles and zeros
- Pole and zero placement design for a PID controller

ELEC 207B: Timeline

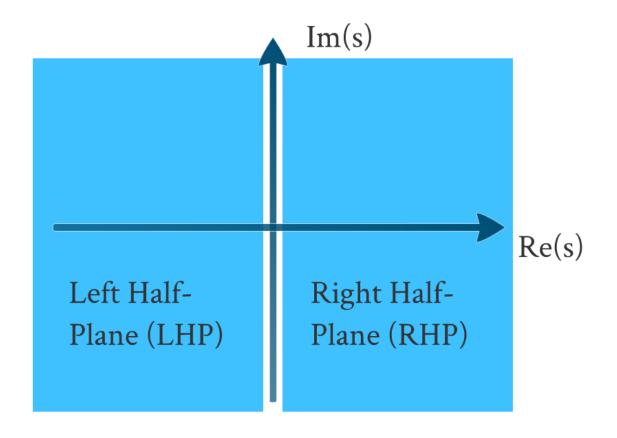


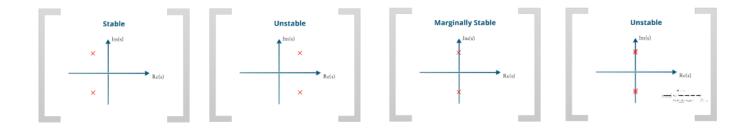
Stability and Instability

Bounded Input Bounded Output definition of stability:

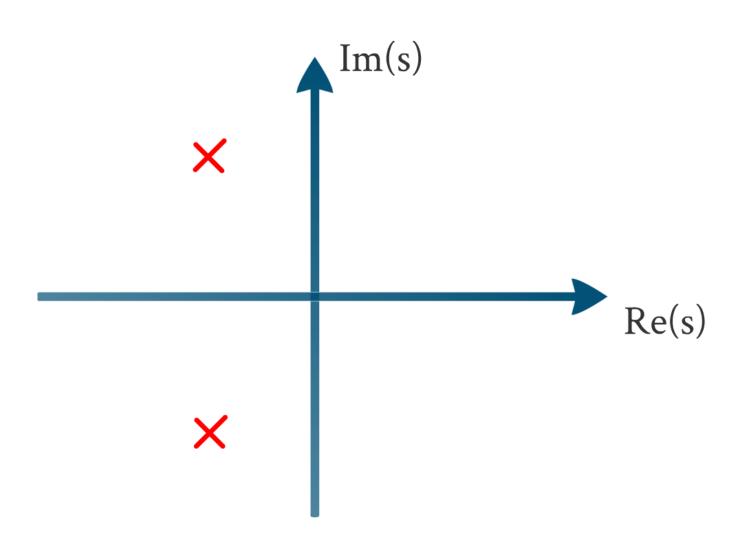
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Relationship Between Poles' Location and Stability

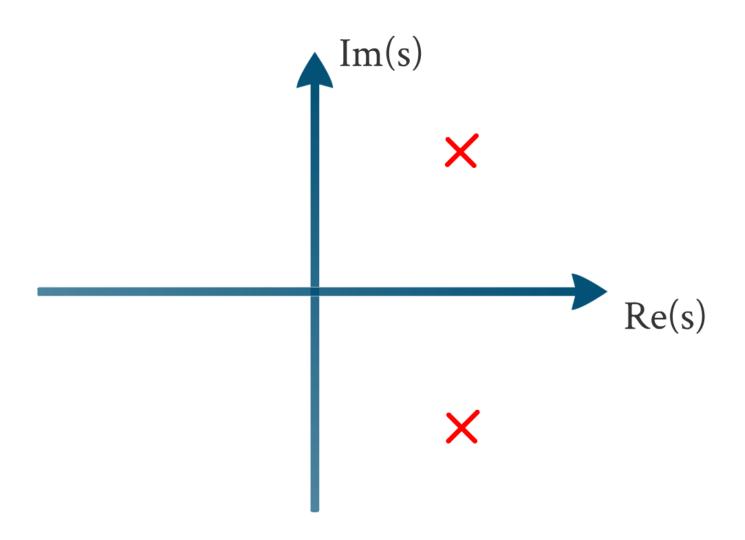




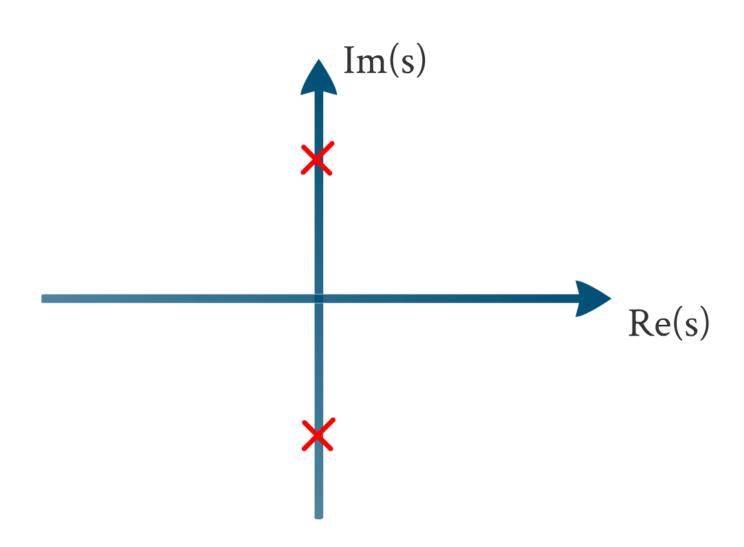
Stable



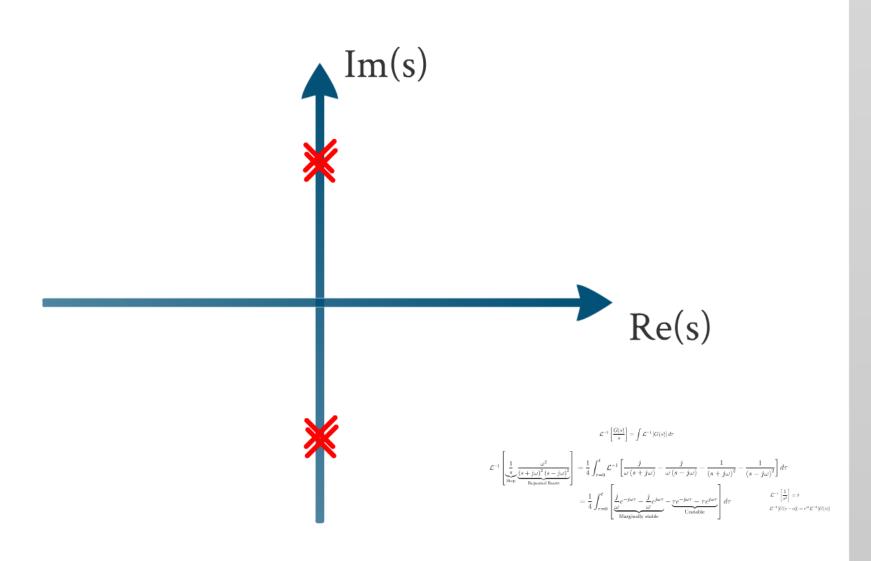
Unstable



Marginally Stable



Unstable



$$\mathcal{L}^{-1}\left[\frac{G(s)}{s}\right] = \int \mathcal{L}^{-1}\left[G(s)\right] d\tau$$

$$\mathcal{L}^{-1} \left[\underbrace{\frac{1}{s}}_{\text{Step}} \underbrace{\frac{\omega^2}{(s+j\omega)^2 (s-j\omega)^2}} \right] = \frac{1}{4} \int_{\tau=0}^t \mathcal{L}^{-1} \left[\frac{j}{\omega (s+j\omega)} - \frac{j}{\omega (s-j\omega)} - \frac{1}{(s+j\omega)^2} - \frac{1}{(s-j\omega)^2} \right] d\tau$$

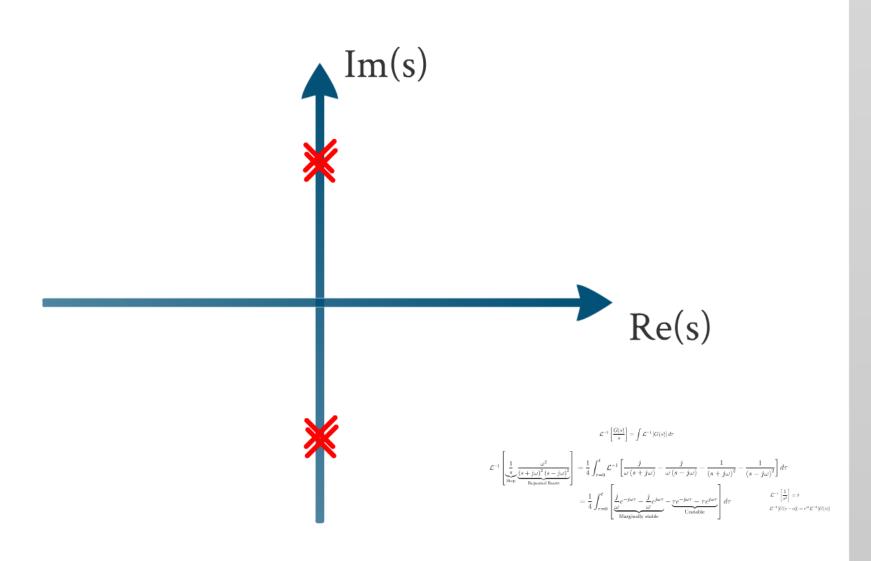
$$= \frac{1}{4} \int_{\tau=0}^t \left[\underbrace{\frac{j}{\omega} e^{-j\omega\tau} - \frac{j}{\omega} e^{j\omega\tau}}_{\text{Marginally stable}} - \underbrace{\tau e^{-j\omega\tau} - \tau e^{j\omega\tau}}_{\text{Unstable}} \right] d\tau$$

$$\mathcal{L}^{-1} \left[\frac{1}{s^2} \right] = t$$

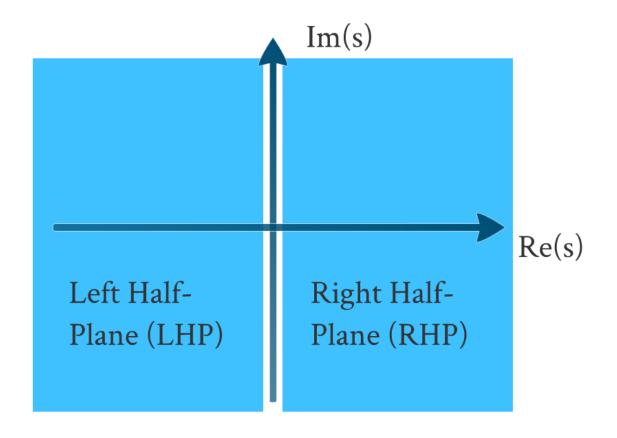
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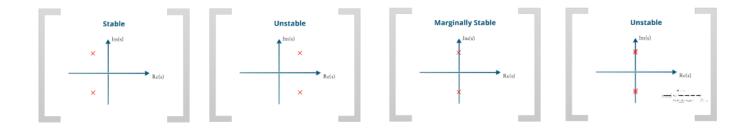
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Unstable



Relationship Between Poles' Location and Stability





Routh-Hurwitz Stability Criteria

$$(s+r_1)(s+r_2)\dots(s+r_n) = s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0 = 0$$

$$r_i > 0 \quad \forall i \qquad \qquad \underbrace{(s-1+3j)(s-1-3j)(s+3) = s^3 + s^2 + 4s + 30}_{\text{Roots in RHP}}$$

Find the (equivalent) closed-loop transfer function

Produce the initial lay-out of the Routh table

Write the coefficients of the characteristic function across the first two rows.

Complete the remaining part of the Routh table until all terms generated are zero

Progress across and down until you have completed the s° row, working out negative of scaled determinant of matrix of previously calculated values.

Two problematic cases can occur:

Case 1: There is a zero in the first column, but the rest of Case 2: There is a zero across an entire row the row includes nonzero numbers.

- We can replace the zero with a small number, e, continue and reinterpret the results.
- Consider the Auxiliary Equation, A(s), formed from
- the coefficients of the row just above the row of zeros.
- · Replace the row of zeros with a row of coefficients from dA(s)/ds and continue Routh's tabulation.

Interpret the results

The number of sign changes is the number of roots in the RHP.

- So, no sign change in the first column of Routh table implies the system is stable.
- A sign change implies that the system is unstable.

If there were a row of zeros, then there are symmetric roots about the origin



If there are no sign changes, but an entire row of zeros was encountered in the tabulation process, then the possibility for repeated purely imaginary roots needs to be considered.

- The roots of the auxiliary equation are also roots of the original equation. So, you can solve each auxiliary equation for the corresponding roots.
- Repeated roots on the imaginary axis make the system unstable.
- Multiple distinct roots on the imaginary axis make the system marginally stable.

 $+ \dots + a_1 s + a_0 = 0$

$$\underbrace{(s-1+3j)(s-1-3j)}_{\text{Roots in RHP}}(s+3) = s^3 + s^2 + 4s + 30$$

Routh-Hurwitz Stability Criteria

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Example 1

$$a_6s^6 + a_5s^5 + a_4s^4 + a_3s^3 + a_2s^2 + a_1s^1 + a_0s^0$$

s ⁶	a_6	a_4	a_2	ra ₀
8 ⁵	a_5	n ₃	a_1	-0
84				
.83				
s2				
s1				
80				

8^{Ω}	- 6	ls.				60;4			- 4	1		- 4
s^k	$\delta_{\parallel} = -\tfrac{1}{4\pi}$	86 85	α_L α_S	5-3	$=-\frac{1}{\epsilon_0}$	Ali Ali	.02 .01	b ₃ :	$=-\frac{1}{\epsilon_1}$	Ali Ali	40 0	4
a^2	$c_1=-\tfrac{1}{k_1}$	δε δη	a_3 b_2	62	$= -\frac{1}{i_1}$	85 81	δ ₁ δ ₂			0		-6
s^2												
s^1												
×n												

s^6	п	6	a_4	a_2	0,
R^{2}	а	5	a_3	a_1	- 0
s^4	$b_1 = -\frac{1}{a_5}$	a ₆ a ₄ a ₅ a ₈			Τ
$\frac{s^3}{s^2}$					\top
s^1					
s^0					

s.e.	a	6			14			12	0
'n.	ős,			ria.			n ₁		
s ⁴	$b_{\parallel}=-\tfrac{1}{a_{1}}$	σ_6 σ_5	α ₄ α ₃	$h_2 = -\frac{1}{\infty}$	σ ₆ σ ₅	α ₃ α ₁	$h_0 = -\frac{1}{\infty}$	n ₆ n ₉ n ₅ 0	(
y ³	$c_1=-\tfrac{1}{b_1}$	a_5 b_1	h_2	$c_2 = -\frac{1}{b_2}$	65 61	h_3		0	ŀ
x ²	$d_1 = -\frac{1}{c_1}$	δ ₁	h2 r2	$d_2 = -\frac{1}{c_1}$	h _i	λ _h 0		n	1
n ³									$\overline{}$

s^n	- 0	6	6	24	02	0.0
s^{α}	- 4	6	6	23	a1	- 0
s^4	$b_1 = -\frac{1}{4n}$	a ₆ a ₄ a ₅ a ₅	$b_2 = -\frac{1}{a_4}$	as as		Т
s^3						
s^2						
s^1						
e ⁿ						

A^{B}	0	iei .			0	4			- 0	2			90
N^2	0	la .			43.			a_1				0	
s^4	$b_1 = -\tfrac{1}{c\varsigma}$	06 03	a_1		$h_2=-\tfrac{1}{\alpha\varsigma}$	α ₆ α ₅	03 01	b3 =	$-\frac{1}{v_S}$	α ₆ α ₅	m) 0		D
s^3	$c_1 = -\frac{1}{b_1}$	03 b1	a_3 b_2		$e_2=-\frac{1}{b_1}$	a_5 b_1	b_8)		T	0
s^2	$d_1 = -\frac{1}{r_1}$	la q	В ₂ Г2		$d_2 = -\tfrac{1}{\epsilon_1}$	h ₁	b ₃ 0		-)			0
s^1	$e_1 = -\frac{1}{v_i}$	cı dı	$\frac{e_2}{d_2}$		(D)			0
s^0		11	11.9	Н								+	

×6	a_{ij}				£g .			α_2			
80	45				ĉs .		α1			0	
\mathbf{s}^{d}	$b_1 = -\frac{1}{a_n}$	$a_6 - a_4 = a_5 - a_5$	bj	$=-\frac{1}{\kappa_h}$	a_6 a_5	α_2 α_1	$b_8 = -\frac{1}{\kappa_8}$	as as	α _α 0		
83			_								
s ²											
R _I											
s0											

89		ic .			n ₄				82	48.0
A^{2}		5			d ₃				a _L	-0
n ⁴	$b_1=-\tfrac{1}{\alpha_0}$	α ₆	a_4 a_3		$h_2 = -\tfrac{1}{\epsilon_1}$	rtg rtg	a_2 a_1		$b_3 = -\frac{1}{4\epsilon}\begin{bmatrix} a_6 & a_6 \\ a_5 & 0 \end{bmatrix}$	0
κ^3	$c_1 = -\tfrac{1}{\delta_1}$	h_1	$\frac{\sigma_3}{b_2}$		$c_2 = -\tfrac{1}{b_1}$	$\frac{a_5}{b_1}$	а1 Ла	Г	0	0
s^2	$d_1 = -\tfrac{1}{c_1}$	b ₁	b ₂ c ₂		$d_2 = -\frac{1}{c_1}$	b ₁ e ₁	λ ₀ 0		0	0
nº	$c_1=-\tfrac{1}{d_1}$	$\frac{c_1}{d_1}$	$\frac{c_2}{d_2}$		(9			0	0
κ^0	$f_1 = -\frac{1}{\epsilon_1}$	d1 61	d ₂ 0	Π		0			0	0

g ^B	n ₆	04	n ₂	n ₀
8,2	n ₅	63	a ₁	0
s^4	$b_1 = -\frac{1}{a_0} \begin{bmatrix} a_0 & a_4 \\ a_5 & a_3 \end{bmatrix}$	$b_2 = -\frac{1}{a_0} \begin{bmatrix} a_6 & a_2 \\ a_5 & a_1 \end{bmatrix}$	$h_3 = -\frac{1}{4q} \begin{bmatrix} a_6 & a_0 \\ a_5 & 0 \end{bmatrix}$	0
43				
в2				
81				
s ⁰				

s^6	a_6	a_4	a_2	a_0
s^5	a_5	a_3	a_1	0
s^4				
s^3				
s^2				
s^1				
s^0				



You Tube

s^6	a_6	a_4	a_2	a_0
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s^2												
s^1												
×n												

s^6	п	6	a_4	a_2	0,
R^{2}	а	5	a_3	a_1	- 0
s^4	$b_1 = -\frac{1}{a_5}$	a ₆ a ₄ a ₅ a ₈			Τ
$\frac{s^3}{s^2}$					\top
s^1					
s^0					

s.e.	a	6			14			12	0
'n.		5		rta.			n ₁		
s ⁴	$b_{\parallel}=-\tfrac{1}{a_{1}}$	σ_6 σ_5	α ₄ α ₃	$h_2 = -\frac{1}{\infty}$	σ ₆ σ ₅	α ₃ α ₁	$h_0 = -\frac{1}{\infty}$	n ₆ n ₉ n ₅ 0	(
y ³	$c_1=-\tfrac{1}{b_1}$	a_5 b_1	h_2	$c_2 = -\frac{1}{b_2}$	65 61	h_3	0		
x ²	$d_1 = -\frac{1}{c_1}$	δ ₁	h2 r2	$d_2 = -\frac{1}{c_1}$	h _i	λ _h 0		n	1
n ³									$\overline{}$

s^n	- 0	6	6	24	02	0.6
s^{α}	- 4	6	6	23	a1	- 0
s^4	$b_1 = -\frac{1}{4n}$	a ₆ a ₄ a ₅ a ₅	$b_2 = -\frac{1}{a_4}$	as as		Т
s^3						
s^2						
s^1						
e ⁿ						

A^{B}	0	iei .			0	4			- 0	2			90
N^2	0	la .			0		#1				0		
s^4	$b_1 = -\tfrac{1}{c\varsigma}$	06 03	a_1		$h_2=-\tfrac{1}{\alpha\varsigma}$	α ₆ α ₅	03 01	b3 =	$-\frac{1}{v_S}$	α ₆ α ₅	m) 0		D
s^3	$c_1 = -\frac{1}{b_1}$	03 b1	a_3 b_2		$e_2=-\frac{1}{b_1}$	$a_5 \\ b_1$	b_8)		T	0
s^2	$d_1 = -\frac{1}{r_1}$	la q	В ₂ Г2		$d_2 = -\tfrac{1}{\epsilon_1}$	h ₁	b ₃ 0	0				0	
s^1	$e_1 = -\frac{1}{v_i}$	cı dı	$\frac{e_2}{d_2}$		0)			0
s^0		11	11.9	Н								+	

×6	a ₀				£g .			a ₂			
80	425				ĉs .			0			
\mathbf{s}^{d}	$b_1 = -\frac{1}{a_n}$	$a_6 - a_4 = a_5 - a_5$	bj	$=-\frac{1}{\kappa_h}$	a_6 a_5	α_2 α_1	$b_8 = -\frac{1}{\kappa_8}$	as as	α _α 0		
83			_								
s ²											
R _I											
s0											

89		ic .				4			82	48.0
A^{2}		5			d ₃				a _L	-0
n ⁴	$b_1=-\tfrac{1}{\alpha_0}$	α ₆	a_4 a_3		$h_2 = -\tfrac{1}{\epsilon_1}$	rtg rtg	a_2 a_1		$b_3 = -\frac{1}{4\epsilon}\begin{bmatrix} a_6 & a_6 \\ a_5 & 0 \end{bmatrix}$	0
κ^3	$c_1 = -\tfrac{1}{\delta_1}$	h_1	$\frac{\sigma_3}{b_2}$		$c_2 = -\tfrac{1}{b_1}$	$\frac{a_5}{b_1}$	а1 Ла	Г	0	0
s^2	$d_1 = -\tfrac{1}{c_1}$	b ₁	b ₂ c ₂		$d_2 = -\frac{1}{c_1}$. , b ₁ b ₂			0	0
nº	$c_1=-\tfrac{1}{d_1}$	$\frac{c_1}{d_1}$	$\frac{c_2}{d_2}$		0				0	0
κ^0	$f_1 = -\frac{1}{\epsilon_1}$	d1	d ₂ 0	Π	0			Ī	0	0

g ^B	n ₆	04	n ₂	n ₀
8,2	n ₅	63	a ₁	0
s^4	$b_1 = -\frac{1}{a_0} \begin{bmatrix} a_0 & a_4 \\ a_5 & a_3 \end{bmatrix}$	$b_2 = -\frac{1}{a_0} \begin{bmatrix} a_6 & a_2 \\ a_5 & a_1 \end{bmatrix}$	$h_3 = -\frac{1}{4q} \begin{bmatrix} a_6 & a_0 \\ a_5 & 0 \end{bmatrix}$	0
43				
в2				
81				
s ⁰				

s^6	a	6	a_4	a_2	a_0
s^5	a		a_3	a_1	0
s^4	$b_1 = -\frac{1}{a_5}$	$egin{array}{ccc} a_6 & a_4 \ a_5 & a_3 \ \end{array}$			
s^3					
s^2					
s^1					
s^0					

s^6	a_6	a_4	a_2	a_0
e 5	0-	α{\circ}	0.	\cap

s^6	a	6		а	$\sqrt{4}$			a_2	a_0
s^5	a	[,] 5		а	3	a_1	0		
s^4	$b_1 = -\frac{1}{a_5}$	$egin{array}{ccc} a_6 & a_6 & a_5 & a_6 \end{array}$	$b_2 = -\frac{1}{a_5}$	$a_6 \ a_5$	a_2 a_1				
s^3			•						
s^2									
s^1									
s^0									

s^6	a_6	a_4	a_2	a_0
s^5	a_5	a_3	a_1	0
l	l l			

s^{2}		
s^1		
s^0		

s^6	a_6		a	4	a_2	a_0
s^5	a_5		a	3	a_1	0
s^4	$b_1 = -\frac{1}{a_5} \begin{vmatrix} a_6 \\ a_5 \end{vmatrix}$	$\begin{bmatrix} a_4 \\ a_3 \end{bmatrix}$	$b_2 = -\frac{1}{a_5}$	$\begin{bmatrix} a_6 & a_2 \\ a_5 & a_1 \end{bmatrix}$	$b_3 = -\frac{1}{a_5} \begin{vmatrix} a_6 & a_0 \\ a_5 & 0 \end{vmatrix}$	
s^3		'				
s^2						
s^1						
s^0						

s^6	a_6		а	b_4	a	a_0	
s^5	a_5		a	43	a	0	
_c 4	$\begin{vmatrix} b_1 & - & 1 \end{vmatrix} a_6$	a_4	$h_2 - \frac{1}{2}$	a_6 a_2	$h_{2} = -\frac{1}{2}$	a_6 a_0	0

3		
s^1		
s^0		

s^6	a_6	3	а	4	a_2	a_0	
s^5	a_5	5	а	3	a_1	0	
s^4	$b_1 = -\frac{1}{a_5}$	$\begin{bmatrix} a_6 & a_4 \\ a_5 & a_3 \end{bmatrix}$	$b_2 = -\frac{1}{a_5}$	$\begin{array}{c cc} a_6 & a_2 \\ a_5 & a_1 \end{array}$	$b_3 = -\frac{1}{a_5} \begin{vmatrix} a_6 \\ a_5 \end{vmatrix}$	$\begin{bmatrix} a_0 \\ 0 \end{bmatrix}$	0
s^3		•			·	•	
s^2							
s^1							
s^0							

s^6	a_6 a_5			a_4			a	a_0			
s^5					a_3			a_1			0
s^4	$b_1 = -\frac{1}{a_5}$	$a_6 \ a_5$	$\begin{bmatrix} a_4 \\ a_3 \end{bmatrix}$		$b_2 = -\frac{1}{a_5}$	$a_6 \ a_5$	$\begin{bmatrix} a_2 \\ a_1 \end{bmatrix}$	$b_3 = -\frac{1}{a_5}$	$a_6 \ a_5$	$\begin{bmatrix} a_0 \\ 0 \end{bmatrix}$	0
s^3	$c_1 = -\frac{1}{b_1} \begin{vmatrix} a_5 & a_3 \\ b_1 & b_2 \end{vmatrix}$				$c_2 = -\frac{1}{b_1}$	$a_5 \ b_1$	$egin{array}{c c} a_1 \ b_3 \end{array}$	0			0
s^2							·				
s^1											
s^0											

s^6	a_6 a_5				a_4 a_3			a_2				$\overline{a_0}$	
s^5								a_1				0	
s^4	$b_1 = -\frac{1}{a_5}$	$a_6 \ a_5$	$\begin{bmatrix} a_4 \\ a_3 \end{bmatrix}$		$b_2 = -\frac{1}{a_5}$	$a_6 \\ a_5$	$\begin{bmatrix} a_2 \\ a_1 \end{bmatrix}$		$b_3 = -\frac{1}{a_5}$	$\begin{vmatrix} a_6 \\ a_5 \end{vmatrix}$	$\begin{bmatrix} a_0 \\ 0 \end{bmatrix}$		0
s^3	$c_1 = -\frac{1}{b_1}$	$egin{array}{c} a_5 \ b_1 \end{array}$	$\begin{vmatrix} a_3 \\ b_2 \end{vmatrix}$		$c_2 = -\frac{1}{b_1}$	$a_5 \\ b_1$	$\begin{bmatrix} a_1 \\ b_3 \end{bmatrix}$		0				0
s^2	$d_1 = -\frac{1}{c_1}$	$egin{array}{c} b_1 \\ c_1 \end{array}$	$\begin{array}{c c} b_2 \\ c_2 \end{array}$		$d_2 = -\frac{1}{c_1}$	b_1 c_1	$\begin{bmatrix} b_3 \\ 0 \end{bmatrix}$		0				0
s^1			•				·						
s^0													

s^6	a_6			\overline{a}	4		a	b_2	a_0
s^5	a_5			a_3			a	0	
s^4	$b_1 = -\frac{1}{a_5}$	$a_6 \ a_5$	$\begin{bmatrix} a_4 \\ a_3 \end{bmatrix}$	$b_2 = -\frac{1}{a_5}$	$a_6 \ a_5$	$egin{array}{c c} a_2 & \\ a_1 & \end{array}$	$b_3 = -\frac{1}{a_5}$	$\begin{bmatrix} a_6 & a_0 \\ a_5 & 0 \end{bmatrix}$	0
s^3	$c_1 = -\frac{1}{b_1}$	a_5 b_1	$\begin{vmatrix} a_3 \\ b_2 \end{vmatrix}$	$c_2 = -\frac{1}{b_1}$	$egin{array}{c} a_5 \ b_1 \end{array}$	$\begin{bmatrix} a_1 \\ b_3 \end{bmatrix}$)	0
s^2	$d_1 = -\frac{1}{c_1}$	$egin{array}{c} b_1 \\ c_1 \end{array}$	$\begin{array}{c c} b_2 \\ c_2 \end{array}$	$d_2 = -\frac{1}{c_1}$	$\begin{vmatrix} b_1 \\ c_1 \end{vmatrix}$	$\begin{bmatrix} b_3 \\ 0 \end{bmatrix}$)	0
s^1	$e_1 = -\frac{1}{d_1}$	$egin{array}{c} c_1 \ d_1 \end{array}$	$egin{array}{c} c_2 \ d_2 \end{array}$	0				0	
s^0									

s^6	a	6		a	4		\overline{a}	2		a_0
s^5	a_5			a_3			a_1			0
s^4	$b_1 = -\frac{1}{a_5}$	$a_6 \ a_5$	$egin{array}{c} a_4 \ a_3 \end{array}$	$b_2 = -\frac{1}{a_5}$	$a_6 \ a_5$	$\begin{bmatrix} a_2 \\ a_1 \end{bmatrix}$	$b_3 = -\frac{1}{a_5}$	$a_6 \\ a_5$	$\begin{bmatrix} a_0 \\ 0 \end{bmatrix}$	0
s^3	$c_1 = -\frac{1}{b_1}$	$egin{array}{c} a_5 \ b_1 \end{array}$	$egin{array}{c} a_3 \ b_2 \end{array}$	$c_2 = -\frac{1}{b_1}$	$egin{array}{c} a_5 \ b_1 \end{array}$	$\begin{bmatrix} a_1 \\ b_3 \end{bmatrix}$	()	·	0
s^2	$d_1 = -\frac{1}{c_1}$	$egin{array}{c} b_1 \\ c_1 \end{array}$	$\begin{array}{c c} b_2 \\ c_2 \end{array}$	$d_2 = -\frac{1}{c_1}$	$egin{array}{c} b_1 \\ c_1 \end{array}$	$\begin{bmatrix} b_3 \\ 0 \end{bmatrix}$	0			0
s^1	$e_1 = -\frac{1}{d_1}$	$egin{array}{c} c_1 \ d_1 \end{array}$	d_2	0			()		0
s^0	$f_1 = -\frac{1}{e_1}$	$egin{array}{c} d_1 \ e_1 \end{array}$	$\begin{pmatrix} d_2 \\ 0 \end{pmatrix}$	()		()		0

$$(s+r_1)(s+r_2)\dots(s+r_n) = s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0 = 0$$

$$r_i > 0 \quad \forall i \qquad \qquad \underbrace{(s-1+3j)(s-1-3j)(s+3) = s^3 + s^2 + 4s + 30}_{\text{Roots in RHP}}$$

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The number of sign changes is the number of roots in the RHP.

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- The roots of the auxiliary equation are also roots of the original equation. So, you can solve each auxiliary equation for the corresponding roots.
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If there are no sign changes, but an entire row of zeros was encountered in the tabulation process, then the possibility for repeated purely imaginary roots needs to be considered.

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Example 4: Row of Zeros

s^5	1	2	1
s^4	1	2	1
s^3			
s^2			
s^1			
s^0			

s^5	1	2	1
s^4	1	2	1
s^3	$ \begin{vmatrix} -\frac{1}{1} & 1 & 2 \\ 1 & 2 & 2 \end{vmatrix} = 0 \to 4 $	$ \begin{vmatrix} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{vmatrix} = 0 \to 4 $	0
s^2	$ -\frac{1}{4} \left \begin{array}{cc} 1 & 2 \\ 4 & 4 \end{array} \right = 1 $	$\begin{bmatrix} -\frac{1}{4} & 1 & 1\\ 4 & 0 \end{bmatrix} = 1$	0
s^1	·	·	

s^6	a	6		a	4		\overline{a}	2		a_0
s^5	a_5			a_3			a_1			0
s^4	$b_1 = -\frac{1}{a_5}$	$a_6 \ a_5$	$egin{array}{c} a_4 \ a_3 \end{array}$	$b_2 = -\frac{1}{a_5}$	$a_6 \ a_5$	$\begin{bmatrix} a_2 \\ a_1 \end{bmatrix}$	$b_3 = -\frac{1}{a_5}$	$a_6 \\ a_5$	$\begin{bmatrix} a_0 \\ 0 \end{bmatrix}$	0
s^3	$c_1 = -\frac{1}{b_1}$	$egin{array}{c} a_5 \ b_1 \end{array}$	$egin{array}{c} a_3 \ b_2 \end{array}$	$c_2 = -\frac{1}{b_1}$	$egin{array}{c} a_5 \ b_1 \end{array}$	$\begin{bmatrix} a_1 \\ b_3 \end{bmatrix}$	()	·	0
s^2	$d_1 = -\frac{1}{c_1}$	$egin{array}{c} b_1 \\ c_1 \end{array}$	$\begin{array}{c c} b_2 \\ c_2 \end{array}$	$d_2 = -\frac{1}{c_1}$	$egin{array}{c} b_1 \\ c_1 \end{array}$	$\begin{bmatrix} b_3 \\ 0 \end{bmatrix}$	0			0
s^1	$e_1 = -\frac{1}{d_1}$	$egin{array}{c} c_1 \ d_1 \end{array}$	d_2	0			()		0
s^0	$f_1 = -\frac{1}{e_1}$	$egin{array}{c} d_1 \ e_1 \end{array}$	$\begin{pmatrix} d_2 \\ 0 \end{pmatrix}$	()		()		0

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Example 2: Numeric

$$s^4 + 6s^3 + 13s^2 + 12s + 4$$

s^4	1		4			
s^3	6			0		
s^2	$-\frac{1}{6} \begin{vmatrix} 1 & 13 \\ 6 & 12 \end{vmatrix} =$:11	$-\frac{1}{6}$	1 4 6 0	=4	0
s^1	$\begin{vmatrix} -\frac{1}{11} & 6 & 12 \\ 11 & 4 \end{vmatrix} =$	$=\frac{108}{11}$		0		0
s^0	$-\frac{11}{108} \begin{bmatrix} 11 & 4 \\ \frac{108}{11} & 0 \end{bmatrix}$	=4		0		0

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Case 1: There is a zero in the first column, but the rest of the row includes nonzero numbers.

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er of roots in the RHP.

n of Routh table implies the system

Example 3: Zero First Column

s^5	1	3	5
s^4	2	6	3
s^3	$\begin{vmatrix} -\frac{1}{2} & 1 & 3\\ 2 & 6 \end{vmatrix} = \epsilon$	$\begin{vmatrix} -\frac{1}{2} & \frac{1}{2} & \frac{5}{3} \\ 2 & 3 \end{vmatrix} = \frac{7}{2}$	0
s^2			
s^1			
s^0			

s^5		1						3				
s^4		2				3						
s^3	$-\frac{1}{2}$	$\frac{1}{2}$	3 6	$=\epsilon$	$-\frac{1}{2}$	$\frac{1}{2}$	5 3	$=\frac{7}{2}$	0			
s^2	$\lim_{\epsilon \to 0} - \tfrac{1}{\epsilon}$	$\frac{2}{\epsilon}$	$\frac{6}{\frac{7}{2}}$	$=-rac{7}{\epsilon}$	$-\frac{1}{\epsilon}$	$\frac{2}{\epsilon}$	3 0	= 3	0			
s^1	$\lim_{\epsilon \to 0} \frac{\epsilon}{7}$	$\frac{\epsilon}{-\frac{7}{\epsilon}}$	$\frac{7}{2}$	$=\frac{7}{2}$			0		0			
s^0												

s^5]	L			5			
s^4		2	2		6				3
s^3	$-\frac{1}{2}$	1 2	3 6	$=\epsilon$	$-\frac{1}{2}$	1 2	5 3	$=\frac{7}{2}$	0
s^2	$\lim_{\epsilon \to 0} -\frac{1}{\epsilon}$	$\begin{array}{ c c }\hline 2\\ \epsilon \end{array}$	$\frac{6}{\frac{7}{2}}$	$=-rac{7}{\epsilon}$	$-\frac{1}{\epsilon}$	$\frac{2}{\epsilon}$	3 0	= 3	0
s^1									
s^0									

s^5	1	3	5
s^4	2	6	3
s^3	$-\frac{1}{2} \begin{vmatrix} 1 & 3 \\ 2 & 6 \end{vmatrix} = \epsilon$		0
s^2	$\lim_{\epsilon \to 0} -\frac{1}{\epsilon} \begin{vmatrix} 2 & 6 \\ \epsilon & \frac{7}{2} \end{vmatrix} = -\frac{7}{\epsilon}$	$\begin{vmatrix} -\frac{1}{\epsilon} & 2 & 3\\ \epsilon & 0 \end{vmatrix} = 3$	0
s^1	$\lim_{\epsilon \to 0} \frac{\epsilon}{7} \begin{vmatrix} \epsilon & \frac{7}{2} \\ -\frac{7}{\epsilon} & 3 \end{vmatrix} = \frac{7}{2}$	0	0
s^0	$\lim_{\epsilon \to 0} -\frac{2}{7} \begin{vmatrix} -\frac{7}{\epsilon} & 3\\ \frac{7}{2} & 0 \end{vmatrix} = 3$	0	0

s^5	1	3	5
s^4	2	6	3
s^3	$\begin{vmatrix} -\frac{1}{2} & 1 & 3 \\ 2 & 6 \end{vmatrix} = \epsilon$	$\begin{vmatrix} -\frac{1}{2} & \frac{1}{2} & \frac{5}{3} \\ 2 & 3 \end{vmatrix} = \frac{7}{2}$	0
s^2			
s^1			
s^0			

s^5		1	-		3				5
s^4		2	2		6				3
s^3	$-rac{1}{2}$	1 2	3 6	$=\epsilon$	$-\frac{1}{2}$	1 2	5 3	$=\frac{7}{2}$	0
s^2	$\lim_{\epsilon \to 0} -\frac{1}{\epsilon}$	$\begin{array}{ c c }\hline 2 \\ \epsilon \end{array}$	$\frac{6}{\frac{7}{2}}$	$=-rac{7}{\epsilon}$	$-rac{1}{\epsilon}$	$\begin{array}{ c c }\hline 2 \\ \epsilon \end{array}$	3	=3	0
s^1									
s^0									

s^5	1	3	5
s^4	2	6	3
s^3	$-\frac{1}{2} \begin{vmatrix} 1 & 3 \\ 2 & 6 \end{vmatrix} = \epsilon$	$\begin{vmatrix} -\frac{1}{2} & \frac{1}{2} & \frac{5}{3} \\ 2 & 3 \end{vmatrix} = \frac{7}{2}$	0
s^2	$\lim_{\epsilon \to 0} -\frac{1}{\epsilon} \begin{vmatrix} 2 & 6 \\ \epsilon & \frac{7}{2} \end{vmatrix} = -\frac{7}{\epsilon}$	$\begin{vmatrix} -\frac{1}{\epsilon} & 2 & 3 \\ \epsilon & 0 \end{vmatrix} = 3$	0
s^1	$\lim_{\epsilon o 0} rac{\epsilon}{7} \left egin{array}{cc} \epsilon & rac{7}{2} \ -rac{7}{\epsilon} & 3 \end{array} ight = rac{7}{2}$	0	0
s^0			

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s^3	$-\frac{1}{2} \left \begin{array}{cc} 1 & 3 \\ 2 & 6 \end{array} \right = \epsilon$	$\left \begin{array}{c c} -\frac{1}{2} & 1 & 5 \\ 2 & 3 \end{array} \right = \frac{7}{2}$	0
s^2	$\lim_{\epsilon \to 0} -\frac{1}{\epsilon} \begin{vmatrix} 2 & 6 \\ \epsilon & \frac{7}{2} \end{vmatrix} = -\frac{7}{\epsilon}$	$\begin{vmatrix} -\frac{1}{\epsilon} & 2 & 3 \\ \epsilon & 0 \end{vmatrix} = 3$	0
s^1	$\lim_{\epsilon \to 0} \frac{\epsilon}{7} \begin{vmatrix} \epsilon & \frac{7}{2} \\ -\frac{7}{\epsilon} & 3 \end{vmatrix} = \frac{7}{2}$	0	0
s^0	$\lim_{\epsilon \to 0} -\frac{2}{7} \begin{vmatrix} -\frac{7}{\epsilon} & 3\\ \frac{7}{2} & 0 \end{vmatrix} = 3$	0	0

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Interpret the results

The number of sign changes is the number of roots in the RHP.

- So, no sign change in the first column of Routh table implies the system is stable.
- A sign change implies that the system is unstable.

If there were a row of zeros, then there are symmetric roots about the origin

If there are no sign changes, but an entire row of zeros was encountered in the tabulation process, then the possibility for repeated purely imaginary roots needs to be considered.

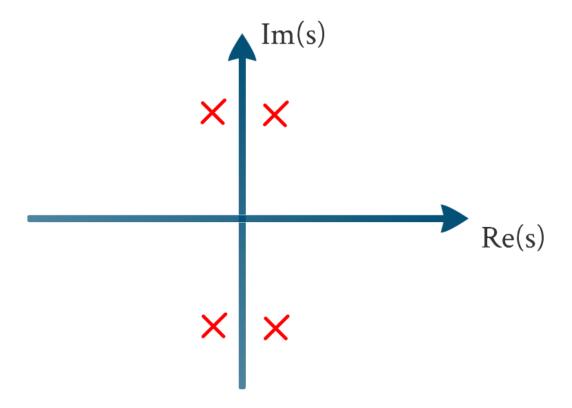
- The roots of the auxiliary equation are also roots of the original equation. So, you can solve each auxiliary equation for the corresponding roots.
- Repeated roots on the imaginary axis make the system unstable.
- Multiple distinct roots on the imaginary axis make the system marginally stable.

Example 4: Row of Zeros

s^5	1	2	1
s^4	1	2	1
s^3			
s^2			
s^1			
s^0			

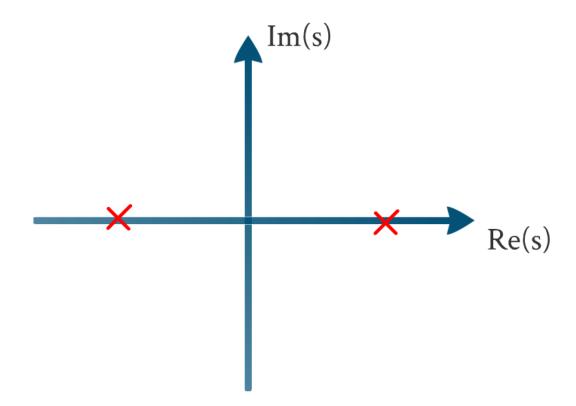
s^5	1	2	1
s^4	1	2	1
s^3	$ \begin{vmatrix} -\frac{1}{1} & 1 & 2 \\ 1 & 2 & 2 \end{vmatrix} = 0 \to 4 $	$ \begin{vmatrix} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{vmatrix} = 0 \to 4 $	0
s^2	$ -\frac{1}{4} \left \begin{array}{cc} 1 & 2 \\ 4 & 4 \end{array} \right = 1 $	$\begin{bmatrix} -\frac{1}{4} & 1 & 1\\ 4 & 0 \end{bmatrix} = 1$	0
s^1	·	·	

Quadrantal symmetric roots



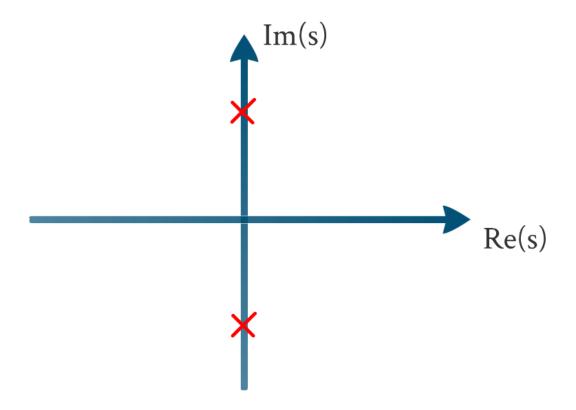
Includes unstable roots, so causes a sign change

Purely Real Roots



Includes unstable root, so causes a sign change

Purely Imaginary Roots



Purely imaginary roots do not cause a sign change

Example 4: Row of Zeros

s^5	1	2	1
$\frac{s^4}{s^3}$	1	2	1
s^3			
s^2			
s^1			
s^0			

s^5	1					2		1	
s^4	1					2		1	
s^3	$-\frac{1}{1}$	1 1	2	= 0	$-\frac{1}{1}$	1 1	1 1	= 0	0
s^2									
s^1									
s^0									

s^5		1					2		1
s^4			1				2		1
s^3	$-\frac{1}{1}$	1 1	2	$= 0 \rightarrow 4$	$-\frac{1}{1}$	1	1	$=0 \rightarrow 4$	0
s^2									
s^1									
s^0									

$$\frac{d}{ds}\left(s^4 + 2s^2 + 1\right) = 4s^3 + 4s$$

s^5	1	2	1
s^4	1	2	1
s^3	$\begin{vmatrix} -\frac{1}{1} & 1 & 2\\ 1 & 2 \end{vmatrix} = 0 \to 4$	$\begin{vmatrix} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{vmatrix} = 0 \to 4$	0
s^2	$-\frac{1}{4} \left \begin{array}{cc} 1 & 2 \\ 4 & 4 \end{array} \right = 1$	$\begin{vmatrix} -\frac{1}{4} & 1 & 1 \\ 4 & 0 \end{vmatrix} = 1$	0
s^1			
s^0			

s^5	1	2	1
s^4	1	2	1
s^3	$ \begin{vmatrix} -\frac{1}{1} & 1 & 2 \\ 1 & 2 & 2 \end{vmatrix} = 0 \to 4 $	$ \begin{vmatrix} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{vmatrix} = 0 \to 4 $	0
s^2	$\begin{bmatrix} -\frac{1}{4} & 1 & 2\\ 4 & 4 \end{bmatrix} = 1$	$ \begin{vmatrix} -\frac{1}{4} & 1 & 1 \\ 4 & 0 & 1 \end{vmatrix} = 1 $	0
s^1	$\begin{bmatrix} -\frac{1}{1} & 4 & 4 \\ 1 & 1 \end{bmatrix} = \epsilon$	0	0
s^0			

s^5	1		2	1
s^4	1		2	1
s^3	$-\frac{1}{1}\begin{bmatrix} 1 & 2 \\ 1 & 2 \end{bmatrix} =$	$= 0 \rightarrow 4$	$\begin{bmatrix} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{bmatrix} = 0 \rightarrow 4$	0
s^2	$-\frac{1}{4} \begin{vmatrix} 1 & 2 \\ 4 & 4 \end{vmatrix}$	= 1	$\begin{bmatrix} -\frac{1}{4} & 1 & 1 \\ 4 & 0 \end{bmatrix} = 1$	0
s^1	$-\frac{1}{1} \begin{vmatrix} 4 & 4 \\ 1 & 1 \end{vmatrix}$	$=\epsilon$	0	0
s^0	$-\frac{1}{\epsilon}$ $\begin{vmatrix} 1 & 1 \\ \epsilon & 0 \end{vmatrix}$	= 1	0	0

s^5	1	2	1
$oxed{s^4} \ s^3$	1	2	1
s^3			
s^2			
s^1			
s^0			

_ _ _

s^5	1	2	1
s^4	1	2	1
s^3	$\begin{vmatrix} -\frac{1}{1} & 1 & 2 \\ 1 & 2 \end{vmatrix} = 0$	$ \begin{array}{c cccc} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{array} = 0 $	0
s^2			
s^1			
s^0			

s^5	1	2	1
s^4	1	2	
s^3	$\begin{vmatrix} -\frac{1}{1} & 1 & 2\\ 1 & 2 \end{vmatrix} = 0 \to 4$	$\left \begin{array}{c c} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{array} \right = 0 \to 4$	0
s^2			
s^1			
s^0			

$$\frac{d}{ds}\left(s^4 + 2s^2 + 1\right) = 4s^3 + 4s$$

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s^5	1	2	$\boxed{1}$
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s^3	$\begin{vmatrix} -\frac{1}{1} & 1 & 2 \\ 1 & 2 \end{vmatrix} = 0 \to 4$	$\left \begin{array}{c c} -\frac{1}{1} & 1 & 1 \\ 1 & 1 & 1 \end{array} \right = 0 \to 4$	0
s^2	$-\frac{1}{4} \left \begin{array}{cc} 1 & 2 \\ 4 & 4 \end{array} \right = 1$	$\left \begin{array}{c c} -\frac{1}{4} & 1 & 1\\ 4 & 0 \end{array} \right = 1$	0
s^1			
s^0			

s^5	1	2	$\lfloor 1 \rfloor$
1 1		_	

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s^2	$ \begin{array}{c cccc} -\frac{1}{4} & 1 & 2 \\ 4 & 4 & 4 \end{array} = 1 $	$ \begin{vmatrix} -\frac{1}{4} & 1 & 1 \\ 4 & 0 & = 1 \end{vmatrix} = 1 $	0
s^1	$\begin{vmatrix} -\frac{1}{1} & 4 & 4 \\ 1 & 1 \end{vmatrix} = \epsilon$	0	0
s^0			

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s^2		$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	0
s^1	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	0	0
s^0	$ \begin{vmatrix} -\frac{1}{\epsilon} & 1 & 1 \\ \epsilon & 0 \end{vmatrix} = 1 $	0	0

$$(s+r_1)(s+r_2)\dots(s+r_n) = s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0 = 0$$

$$r_i > 0 \quad \forall i \qquad \qquad \underbrace{(s-1+3j)(s-1-3j)(s+3) = s^3 + s^2 + 4s + 30}_{\text{Roots in RHP}}$$

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Example 5: Parameterised

s^2	$s^2 \mid 1 \mid K$	
s^1	$0 \rightarrow 2$	$0 \rightarrow 0$
s^0		

s^2	1				K
s^1	$0 \rightarrow 2$			$0 \rightarrow 0$	
s^0	$-\frac{1}{2}$	$\frac{1}{2}$	<i>K</i> 0	=K	0

s^2	1	$1 \qquad K$	
s^1	$0 \rightarrow 2$	$0 \rightarrow 0$	
s^0			

s^2	1			K	
s^1	$0 \rightarrow 2$			$0 \rightarrow 0$	
s^0	$-\frac{1}{2}$	1 2	K = 0	=K	0

Example 5: Parameterised

s^2	$1 \qquad K$	
s^1	$0 \rightarrow 2$	$0 \rightarrow 0$
s^0		

s^2	1				K
s^1	$0 \rightarrow 2$			$0 \rightarrow 0$	
s^0	$-\frac{1}{2}$	$\frac{1}{2}$	<i>K</i> 0	=K	0

$$(s+r_1)(s+r_2)\dots(s+r_n) = s^n + a_{n-1}s^{n-1} + \dots + a_1s + a_0 = 0$$

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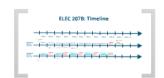
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This lecture covers:



- Stability and Instability
- Relationship between poles' location and stability
- · Routh-Hurwitz stability criteria