MATLAB Exercise 2 – Discrete Fourier Transform and Block Processing

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The objective in this exercise is to develop an understanding of digital filter implementation through DFT-based block processing. Section 1 focuses on the circular convolution property of the DFT and the relationship between the linear convolution and the circular convolution. Section 2 focuses on how to employ DFT to implement linear time—invariant filtering.

1 Circular convolution

The DFT domain filtering uses the fact that the circular convolution of u[n] and h[n] corresponds to the multiplication of their DFTs, i.e., U[k] H[k]. However, some samples of the circular convolution output may be not same as the linear convolution output, because of the time-aliasing inherent in circular convolution. It is therefore necessary to identify those samples, particularity for the block processing (overlap-save algorithm).

1. Write a MATLAB function myCircConv(u,h,M) to implement a circular convolution in DFT domain for an input signal. The function inputs should include an input signal u[n], a filter h[n] of length N and a DFT size of length M > N. Hints: i) Using the MATALB function fft instead of DFT matrix W in the below stated notation is more computationally efficient, ii) Proper consideration of the dimensions of the employed matrices and vectors is crucial!

To remind you, the input signal u[n] is split into blocks of length M with no overlap between blocks so that the l^{th} block of time samples can be expressed as

$$\mathbf{u}_{l} = \left[u\left[lM\right], \dots, u\left[lM + M - 1\right]\right]^{T}$$

The block circular convolution output in the DFT domain is computed as

$$\mathbf{y}_{l}[k] = \mathbf{U}_{l}[k] \mathbf{h}_{zp}[k]$$

where $\mathbf{y}_{l}[k] = [Y_{l}[0], \dots, Y_{l}[M-1]]^{T}$ is the vector of stacked frequency bins $Y_{l}[k]$. $\mathbf{U}_{l}[k]$ denotes a $M \times M$ frequency representation diagonal matrix of $\mathbf{u}_{l}[k]$, i.e.,

$$\mathbf{U}_{l}[k] = \operatorname{diag} \{\mathbf{u}_{l}[k]\} = \operatorname{diag} \{\mathbf{W}\mathbf{u}_{l}\}\$$

with the $M \times M$ dimensional DFT matrix **W**, and $\mathbf{h}_{zp}[k]$ denotes the frequency representation vector of the zero-padded filter (after appending zeroes, filter has length M), i.e.

$$\mathbf{h}_{\mathrm{zp}}[k] = \mathbf{W} \left[\mathbf{h}^T[n], \mathbf{0}^T \right]^T$$

with $\mathbf{h}[n] = [h[0], \dots, h[N-1]]$, and $\mathbf{0}$ the vector of length M-N containing only zeroes. The block circular convolution output in the time domain is then computed as

$$\mathbf{y}_{l} = \mathbf{W}^{-1} \mathbf{y}_{l} \left[k \right].$$

2. Consider the following input signal and filter,

$$u[n] = \begin{cases} (0.9)^n & 0 \le n \le 12\\ 0 & \text{otherwise,} \end{cases}$$
 (1)

$$h[n] = \begin{cases} 1 & 0 \le n \le 11\\ 0 & \text{otherwise,} \end{cases}$$
 (2)

- (a) Use your implemented code to compute the circular convolution output using a 16–point DFT
- (b) Try a linear convolution in the time domain using MATLAB function conv. Determine the output indices where the circular convolution output derived in (a) are the same as the linear convolution output.
- (c) Try a a 16-point circular convolution in the time domain using MATLAB function cconv. Compare this output with the output in (a)
- (d) How to obtain less samples corrupted by the time–aliasing in the circular convolution output?
- 3. Consider 'speech.wav' as the input signal. This signal was recorded with a sampling frequency of $f_s = 16000$ Hz. Generate a random time-domain filter h[n] with a length of 16 ms via MATLAB function randn. Normalize h[n] such that its energy would be 1. Save the filter since it will be required for the next implementation (overlap—save algorithm).
 - Use your implemented code to compute the circular convolution output using a DFT size of 16 ms.

2 Overlap-save Implementation

In this part, implementation of the overlap—save algorithm is aimed.

1. Write a MATLAB function to implement the overlap-save algorithm. The function inputs should include an input signal u[n], a filter h[n] of length N and a DFT size M=2N. To remind you, unlike the previous implementation, \mathbf{u}_l includes now two consecutive blocks (old and new blocks), each block of length N, i.e.

$$\mathbf{u}_{l} = \left[u\left[lN-N\right], \dots, u\left[lN+N-1\right]\right]^{T}.$$

In the DFT domain, the circular convolution output is computed as

$$\mathbf{y}_{l}\left[k\right] = \mathbf{U}_{l}\left[k\right]\mathbf{h}_{\mathrm{zp}}\left[k\right]$$

where $\mathbf{y}_{l}[k] = [Y_{l}[0], \dots, Y_{l}[M-1]]^{T}$ is the vector of stacked frequency bins $Y_{l}[k]$. $\mathbf{U}_{l}[k]$ is a $M \times M$ frequency representation diagonal matrix of \mathbf{u}_{l} , i.e.,

$$\mathbf{U}_{l}[k] = \operatorname{diag} \{\mathbf{u}_{l}[k]\} = \operatorname{diag} \{\mathbf{W}\mathbf{u}_{l}\}$$

with the $M \times M$ dimensional DFT matrix **W**, and $\mathbf{h}_{zp}[k]$ denotes the frequency representation vector of the zero-padded filter (after appending zeroes, filter has length M), i.e.

$$\mathbf{h}_{\mathrm{zp}}[k] = \mathbf{W} \left[\mathbf{h}^{T}[n], \mathbf{0}^{T} \right]^{T}$$

with $\mathbf{h}[n] = [h[0], \dots, h[N-1]]$, and $\mathbf{0}$ the vector of length N containing only zeroes. In the time domain, the filtered output is then computed as

$$\mathbf{y}_l = \mathbf{K}\mathbf{W}^{-1}\mathbf{y}_l[k]; \quad \mathbf{K} = [\mathbf{0}_N, \mathbf{I}_N]$$

with $\mathbf{0}_N$ the $N \times N$ -dimensional matrix containing only zeroes and \mathbf{I}_N the $N \times N$ -dimensional identity matrix. The matrix \mathbf{K} contributes to the time-domain output signal by skipping the output first half (which has been corrupted by the time-aliasing) and retaining the output last half (which equals to the linear filtering output).

- 2. Use the input signal from Eq. (1) and the filter from Eq. (2) to compute the filter output using a 32-point DFT. Compare the output with the output of task 2 from section 1. Is there a difference between the outputs? Explain why?
- 3. Use 'speech.wav' as the input signal and the filter you generated in task 3 from the previous section. Compute the filter output using a DFT size of 32 ms. Compare this output with the output of task 3 from the previous section. Is there a difference between the outputs? Explain why?