

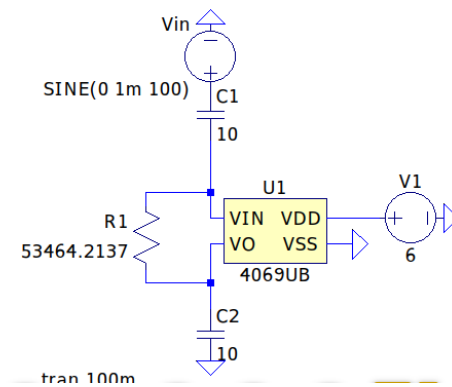
g_m -C Filter with CMOS Inverters

TADIPATRI UDAY KIRAN REDDY
EE19BTECH11038

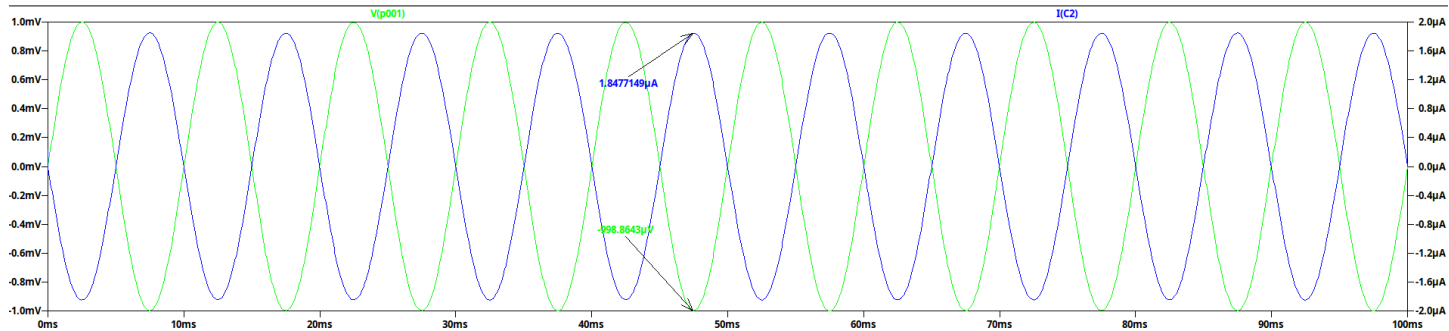
January 29, 2021

1 Problem 1

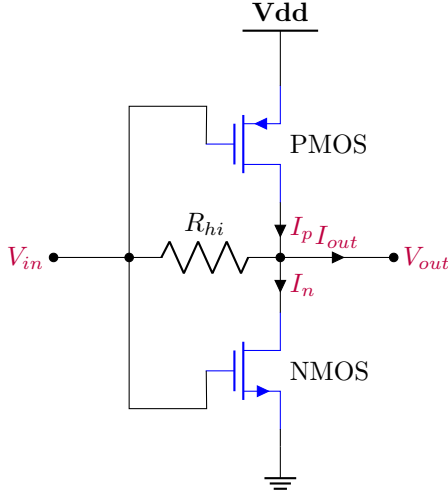
Testbench



Simulation



Observe that we obtained $\frac{i_{out}}{V_{out}}$ as **1.8498m** which is **0.98898** times g_m .



Observe that by applying KCL at node V_{out} we get,

$$I_{out} = I_p - I_n - \frac{V_{out} - V_{in}}{R_{hi}}$$

$$\Rightarrow \frac{\Delta I_{out}}{\Delta V_{in}} = -g_m + \frac{1}{R_{hi}} \Rightarrow \boxed{Transconductance = g_m - \frac{1}{R_{hi}}}$$

From previous experiment we found out that g_m is 0.00187041, Thus inorder to obtain transconductance of $0.99g_m$. We need R_{hi} of $\frac{100}{g_m}$ which is 53464.2137Ω .

Since we are applying AC signal of 1mV at 100Hz at input, and we want our perpetration to be about self bias voltage we introduce a capacitor after the source and before sink so that this self-bias voltage which is DC component doesn't flow through which makes this circuit more like a transconductor. Thus our capacitor should have capacitance such that its time constant is far above than time period of signal.

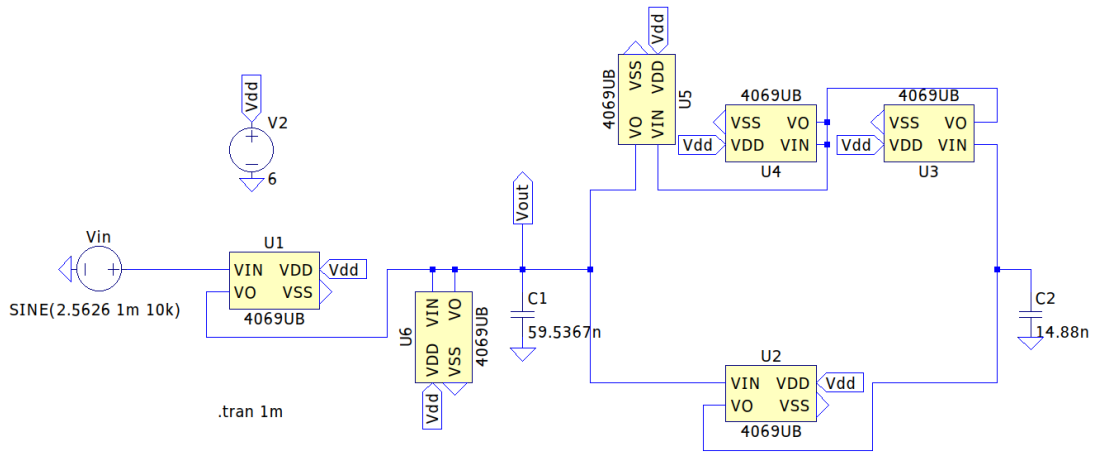
$$R_{eff}C \gg \frac{1}{f}$$

$$C \gg \frac{G_m}{f} \Rightarrow C \gg 18.49810\mu F$$

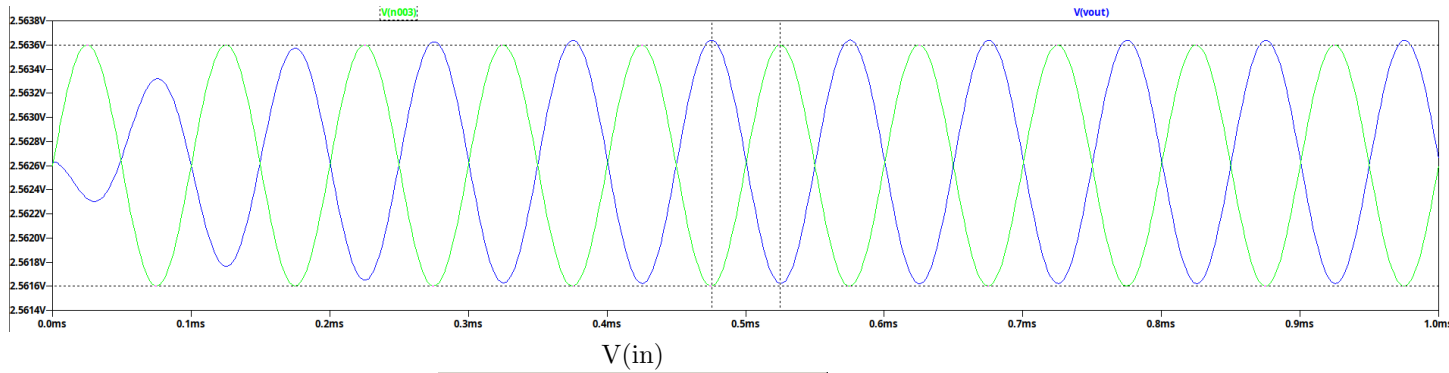
So our capacitance is high enough than required,thus there should not be any problem with it.

2 Problem 2

Testbench



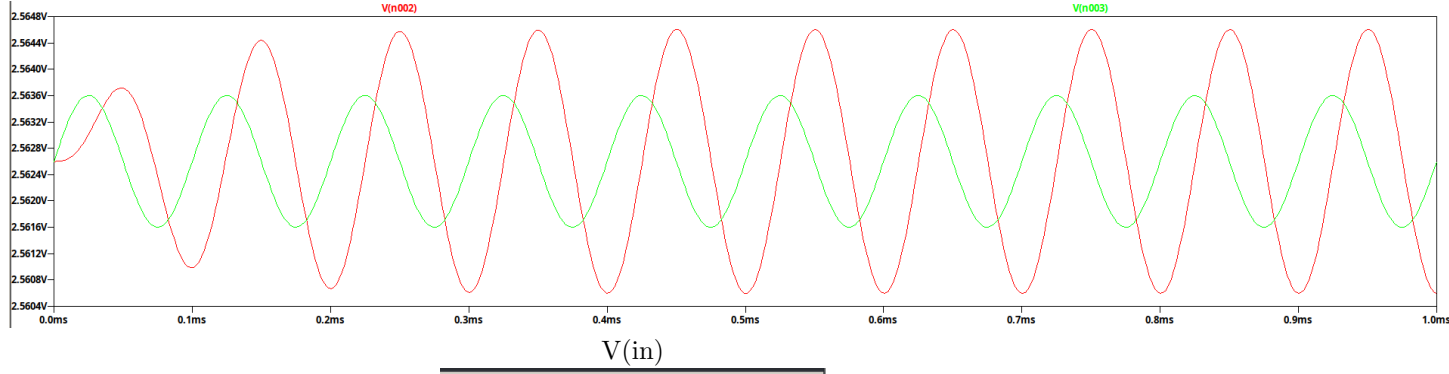
Simulations Resonant Frequency



Cursor 1	
V(n003)	
Horz:	475.88937μs
Vert:	2.5616016V
Cursor 2	
V(n003)	
Horz:	524.7175μs
Vert:	2.5635998V
Diff (Cursor2 - Cursor1)	
Horz:	48.828125μs
Vert:	1.9981861mV
Freq:	20.48KHz
Slope:	40.9229

Cursor 1	
V(vout)	
Horz:	475.88937μs
Vert:	2.5636387V
Cursor 2	
V(vout)	
Horz:	524.7175μs
Vert:	2.5616217V
Diff (Cursor2 - Cursor1)	
Horz:	48.828125μs
Vert:	-2.0170212mV
Freq:	20.48KHz
Slope:	-41.3086

Quality factor



Cursor 1	
V(n003)	
Horz: 526.67062μs	Vert: 2.5635946V
Cursor 2	
V(n003)	
Horz: 475.88937μs	Vert: 2.5616016V
Diff (Cursor2 - Cursor1)	
Horz: -50.78125μs	Vert: -1.9929409mV
Freq: 19.692308KHz	Slope: 39.2456

Cursor 1	
V(n002)	
Horz: 500μs	Vert: 2.5606007V
Cursor 2	
V(n002)	
Horz: 450.49875μs	Vert: 2.5646029V
Diff (Cursor2 - Cursor1)	
Horz: -49.50125μs	Vert: 4.0021926mV
Freq: 20.20151KHz	Slope: -80.8503

The below is the Transfer function for the given g_m -C filter in Experiment

$$\frac{V_{out}}{V_{in}} = \frac{\frac{-sC_2}{g_m}}{\frac{s^2C_1C_2}{g_m^2} + \frac{sC_2}{Rg_m^2} + 1}$$

The following equations are the capacitances and Gain derived from the above transfer function.

$$R = \frac{1}{g_m}$$

$$\omega_n = \frac{g_m}{\sqrt{C_1C_2}}$$

$$Q = R\sqrt{\frac{C_1}{C_2}}g_m$$

$$\Rightarrow C_1 = \frac{Q}{R\omega_n}$$

$$\Rightarrow C_2 = \frac{Rg_m^2}{Q\omega_n} = \frac{Rg_m}{Q\omega_n}$$

$$\frac{V_{out}}{V_{in}}_{s=j\omega_n} = -g_mR = -1$$

Now for verifying Quality factor, We know that at node V_{out} we get Band pass filter response and at C_2 we get 2nd order low pass filter and this has gain of Q and phase shift of -90° at resonant frequency. Now observe the transients of the voltage across C_2 we find that amplitude is 2.001mV and it has phase shift of -90° w.r.t input signal. Thus we can say that ratio is approximately matching specifications.

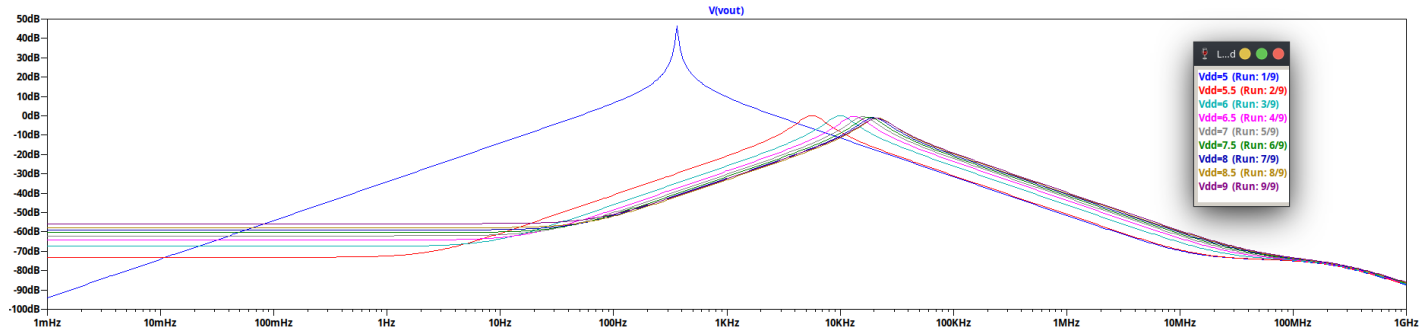
1. Even simulation is not on-par with theoretical results because the gain(g_m) is not exactly same as measured it changes with perpetration. And it is assumed that V_{out} of transistor is constant w.r.t V_{in} but it is not in simulations. Thus these assumptions are creating mismatches with simulations.
2. Inputs/Outputs are biased at $V_B=2.5626V$ or self-bias voltage because there is a constraint of not using resistors in circuits other in CMOS inverter. If there was any load at output or input we could have coupled it with capacitor in-order to obtain desired unbiased output.

When Vdd is 5V it is evident that from previous experiment that at this conditions it cannot be used as analog device or Transconductor rather only as digital logic device. Thus any node the voltage just stays at Vdd or 0. Thus this is not appropriate supply voltage for filter.

$$\frac{\Delta I_{out}}{\Delta V_{in}} = -g_m = -\left(\frac{\beta_p}{2}(V_{dd} - V_{in} - |V_{TH_p}|) + \frac{\beta_n}{2}(V_{in} - V_{TH_n})\right)$$

Testbench





Now this can be observed from the simulation result that Peaking frequency for increment in Vdd increases. Now we can use this property so that by tuning Vdd we can change resonant frequency for the filter. Now this might be good tool for communication devices or any portable filters.