

Abstract

Going to make LIGO the best possible ever.

Adaptive Mode Matching in Advanced LIGO

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Preface

The era of gravitational waves astronomy was ushered in by the LIGO (Laser Interferometer Gravitational-Wave Observatory) collaboration with the detection of a binary black hole collision (Detection paper). The event that shook the foundation of space-time allowed mankind to view the cosmos in a way that had never been done previously.

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Chapter 1

Introduction

Say something profound here.

1.1 Gravitational Waves

In 1915, Albert Einstein published his theory of general relativity 1.1.

The seminal equation in this theory is:

$$G_{\mu\nu} = 8\pi T_{\mu\nu} \tag{1.1}$$

Which is a set of 10 coupled second-order differential equations that are nonlinear. In its complete form, equation 1.1 fully describes the interaction between space-time and mass-energy.

To describe the physics in a highly curved space-time, one would have to fully solve the Einstein field equations numerically. In areas where the curvature is close to flat, the weak field approximation can be applied and the metric is described as

$$g_{\mu\nu} \cong \eta_{\mu\nu} + h_{\mu\nu} \tag{1.2}$$

where $\eta_{\mu\nu}$ is the metric of flat space time and $|h_{\mu\nu}| \ll 1$ is the perturbation due to a gravitational field.

By plugging in equation 1.2 into 1.1 and using empty space we obtain the familiar wave equation

$$\left(\nabla^2 - \frac{1}{c^2} \frac{\partial^2}{\partial t^2}\right) h_{\mu\nu} = 0 \tag{1.3}$$

which has a plane-wave solution of the form $h_{\mu\nu} = A_{\mu\nu} e^{ik_\nu x^\nu}$.

Using the gauge constraint $h^\mu{}_\nu = 0$, it follows that $A_{\mu\nu}k^\mu = 0$ which means that the gravitational wave amplitude is orthogonal to the propagation vector.

Further imposing transverse-traceless gauge and assuming that the wave is traveling in the x^3 direction, it can be shown that the complex amplitude has physical significance expressed in the matrix

$$A_{\mu\nu} = \begin{pmatrix} 0 & 0 & 0 & 0 \\ 0 & A_{xx} & A_{yx} & 0 \\ 0 & A_{xy} & -A_{xx} & 0 \\ 0 & 0 & 0 & 0 \end{pmatrix} \quad (1.4)$$

Oftentimes, the four non-zero components of equation 1.4 can be categorized into two distinct polarizations called plus and cross such that $h_+ = A_{xx} = -A_{yy}$ and $h_\times = A_{xy} = A_{yx}$.

It is natural to attempt to understand the physical interpretation of equation 1.4 as an affect on the position of a free floating particle. Consider the four-velocity, U^α , in the transverse traceless gauge where the coordinate itself is attached the particles and incorporates any small wiggles that would shake the coordinates. Of course, any free particles will follow the geodesic equation

$$\nabla U^\alpha = \frac{d}{d\tau}U^\alpha + \Gamma_{\mu\nu}^\alpha U^\mu U^\nu = 0 \quad (1.5)$$

where $\Gamma_{\mu\nu}^\alpha = \frac{1}{2}g^{\alpha\gamma}(g_{\gamma\mu,\nu} + g_{\gamma\nu,\mu} - g_{\mu\nu,\gamma})$ are the famous Christoffel symbols. By evaluating the first term of the acceleration in equation 1.5,

$$\left(\frac{dU^\alpha}{d\tau}\right)_0 = -\Gamma_{00}^\alpha = \frac{1}{2}\eta_{\mu\nu}(h_{\beta 0,0} + h_{0\beta,0} + h_{00,\beta}) \quad (1.6)$$

However, comparing equation 1.4 and equation 1.6, it is clear that if the particle is initially at rest, then a moment later it is still at rest! The term "at rest" is actually used liberally here since the coordinate system varies along with the gravitational wave.

Alternatively, one can ask if a gravitational wave passed by a pair of particles separated by length L , what would be the effect on the distance between two points? The proper distance is defined as

$$\delta l = \int g_{\mu\nu}dx^\mu dx^\nu = \int_0^L g_{xx}dx \approx |g_{xx}(x=0)|^{1/2} \approx [1 + \frac{1}{2}h_{xx}(x=0)]L \quad (1.7)$$

which shows us two very important points about the nature of gravitational waves. Firstly, the effect is very small since the length variation, h_{xx} , is a small perturbation on flat space-time. Secondly, the effect is proportional to the initial separation between the particles. This means a detector which is large will

have a better chance to measure these small effects, an important point that drove the design of the Laser Interferometer Gravitational-Wave Interferometer (LIGO).

1.1.1 Energy from a Gravitational Wave

1.1.2 Sources of Gravitational Waves

Compact Binary Inspirals

Continuous Waves

Bursts

Stochastic

Even with some of the most energetic events known to humanity such as the merger of neutron stars and black holes, the amount of strain expected is on the order of 10^{-24} .

[1]

1.1.3 Measuring Gravitational Waves with Light

Even with the theoretical formulation of gravitational waves resolved by the 1970s ref[Pirani], detection of GWs by ground-based detectors was still a controversial topic among scientists in the field ref[Collins]. This was due to the incredible accuracy required to measure the strain from even the most dense astrophysical objects ref[Rai]. The earliest attempts at detecting these small signals were famously done by Joseph Weber ref[Weber] using large resonant bars and piezoelectric transducers to extract the energy from gravitational waves at the resonant frequencies of the bars. Picture of bar detectors at LHO. However, these bars are limited by thermal noise and can only detect GWs in very narrow frequency bands ref[Bars].

Interferometers are devices that measure small displacements by using a laser that is split with a partially transmitting mirror (or beamsplitter), which allows 50% of the light to get reflected and 50% to be transmitted. Each of the split beams travel down arms and reflect off of mirrors and return down the arms. Upon reaching the beamsplitter, the beams recombine and by using the superposition of electromagnetic waves the laser will add linearly at the output port (or antisymmetric port). The laser beams will gather phase as they propagate down each individual arm, and when recombining, the intensity of the light will be proportional to the phase differences between each beam. This will correspond to a differential length that is described by

$$L_- = l_x - l_y \quad (1.8)$$

As shown in equation 1.7, the effect of gravitational waves on the proper length between two free falling objects is proportional to the initial separation. From figure[GWparticles], it is intuitively clear that interferometry would be an ideal technique to detect signals from a gravitational wave. However, one can explicitly derive a ground-based interferometer's response to a GW from an astrophysical object.

Consider a gravitational wave source arbitrarily located in the sky with respect to an interferometer on Earth. By denoting the interferometer's Cartesian coordinates as $\{\hat{x}, \hat{y}, \hat{z}\}$ with x and y located along the arms respectively such that the z -axis points directly towards the zenith (i.e. a right-handed system.). Using the well known Euler angles, a relation from the detector frame to the source frame with coordinates $\{\hat{x}', \hat{y}', \hat{z}'\}$ can be seen in Figure[Euler].

If a gravitational wave at the source has emitted GWs with plus and cross polarizations as denoted by equation 1.4, then the detector time series can be regarded as ref[Duncan Thesis and AndersonGWs]

$$h(t) = F_+(\theta, \phi, \psi) h_+(t) + F_\times(\theta, \phi, \psi) h_\times(t) \quad (1.9)$$

where $F_+(\theta, \phi, \psi)$ and $F_\times(\theta, \phi, \psi)$ are the antenna pattern functions that project the gravitational wave amplitudes onto the detector frame.

$$F_+(\theta, \phi, \psi) = -\frac{1}{2}[1 + \cos^2(\theta)]\cos(2\phi)\cos(2\psi) - \cos(\theta)\sin(2\phi)\sin(2\psi) \quad (1.10)$$

$$F_\times(\theta, \phi, \psi) = +\frac{1}{2}[1 + \cos^2(\theta)]\sin(2\phi)\cos(2\psi) - \cos(\theta)\sin(2\phi)\sin(2\psi) \quad (1.11)$$

If the gravitational wave is located directly above the interferometer (ie $\theta = 0$) and setting $\psi = 0$, then the magnitude of the antenna pattern is equal to unity. Furthermore, by rotating about the ϕ angle such that the detector arms align with the plus polarization, the null geodesic equation (ie the path of a photon) in the interferometer frame becomes

$$ds^2 = g_{\mu\nu}dx^\mu dx^\nu = -dt^2 + [1 + h_+]dx^2 + [1 - h_+]dy^2 + dz^2 = 0 \quad (1.12)$$

Now if the photon is traveling along the x -arm, this means that $dy^2 = dz^2 = 0$ and the metric equation transforms to

$$\frac{dt}{dx} = \sqrt{1 + h_+} \approx 1 + \frac{1}{2}h_+ \quad (1.13)$$

The amount of time required for the photon to reach the x-end mirror (starting at $t = 0$) is equal to

$$t_1 = \int_0^{L_x} \left[1 + \frac{1}{2}h_+(x)\right]dx \quad (1.14)$$

where t_0 is the start time and L_x is the total length of the x-arm. Upon returning to the beamsplitter, the photon's total time of flight for the x and y arms are, respectively,

$$t_2 = 2L_x + \frac{1}{2} \int_0^{L_x} \left[h_+(x) + h_+(x + L_x) \right] dx \quad (1.15)$$

$$t'_2 = 2L_y - \frac{1}{2} \int_0^{L_y} \left[h_+(y) + h_+(y + L_y) \right] dy \quad (1.16)$$

If the gravitational wave period is much longer than the time of flight, then h_+ does not change much during the measurement, which means $h_+(\eta_i) \approx h_+(\eta_i + L_{\eta_i}) \approx \text{constant}$. By subtracting the flight times of the photons for each arm and setting $L = L_x = L_y$, the difference is proportional to the gravitational wave perturbation multiplied by the sum of arm lengths (with a factor of c to get the units right),

$$\Delta t = t_2 - t'_2 = \frac{2L}{c} h_+ \quad (1.17)$$

By recasting the expression for time of flight in terms of the phase picked up laser light as it travels through space, the differential phase shift is

$$\Delta\Phi = \Phi(t_2) - \Phi(t'_2) = \frac{4\pi}{\lambda} h_+ L \quad (1.18)$$

The equation above is simple, however, it only works for gravitational wave signals that are not frequency dependent and it assumes that the path length can be arbitrarily long. Both points are actually not true but we can alleviate these discrepancies by considering a gravitational wave signal of the form $h(t) = h_0 \exp(i2\pi f_{GW}t)$ and repeating the calculation between equations 1.14 and 1.18:

$$\Delta\Phi(t) = h(t) \tau \frac{2\pi c}{\lambda} \text{sinc}(f_{GW}\tau) e^{i\pi f_{GW}\tau} \quad (1.19)$$

where $\tau_{RT} = 2L/c$. The response for a detector whose arm lengths are blank km have null points at the around blank Hz which means the instrument cannot be arbitrarily long, however, this point is not concerning because it is too expensive and difficult to make a terrestrial detector of this size.

How to practically measure $\Delta\Phi$ with an interferometer is explained in the next section.

1.2 The LIGO Instrument

In the simplest form, the LIGO instrument is an incredibly large Michaelson interferometer. If we imagine the output as a measure of the differential arm length, it becomes a natural way of detecting gravitational waves.

The LIGO instruments are considered dual-recycled Fabry-Perot interferometers.

1.2.1 Simple Michaelson

As show in figure[michaelson], the interferometer readout uses a photodetector that measure the total power which depends on how laser light in the arms constructively or destructively interferes. At the end of section 1.1.3, it was shown that the differential time of flight between photons traveling down each arm carries gravitational wave information. The difference in flight times, Δt , can be easily cast into how much phase, $\Delta\phi$ is accumulated by the photons as they propagate through space. But the question remains how an interferometer actually measures $\Delta\phi$:

If the input electric field of the interferometer is E_0 , the beamsplitter will transmit $E_0/2$ down the x-arm and reflect $E_0/2$ down the y-arm. By setting the beamsplitter to be the origin, the beams traveling down their respective arms will have gathered a phase ϕ_i . Then, upon reflecting off the end mirrors and returning to the beamsplitter, each of the electric fields can be described by these equations

$$\begin{aligned} E_x &= \frac{iE_0}{2} e^{2i\phi_x} \\ E_y &= \frac{iE_0}{2} e^{2i\phi_y} \end{aligned} \tag{1.20}$$

Since the electromagnetic waves are linear, the resultant sum of waves at the output will be $E_{out} = E_x + E_y$. A photodiode (PD) is placed at the output (or antisymmetric) port to read out the integrated power which is related to the total electric field by

$$\begin{aligned} P|_{PD} &= \int_{Area} I \, dA \\ &= \int_{Area} (E_x + E_y)^2 \, dA \\ &= P_{in} \cos^2(\Delta\phi) \end{aligned} \tag{1.21}$$

where $\Delta\phi = \phi_x - \phi_y = k_x L_x - k_y L_y$ and P_{in} is the input power.. By using equation 1.18 and the common (or average) arm length $L_+ = \frac{L_x + L_y}{2}$, the differential phase results in a power shift

$$P|_{PD} \approx P_{max} (1 - 2\Delta\phi) = P_{in} (1 - 2kL_+ h_+) \tag{1.22}$$

There is a large DC term that is dependent on the input power and generally, it is very difficult to measure small changes in a large signal. So the next obvious method would be to shift the arms such that the output port is operating on a dark fringe, normally this is called a null-point operation. However, there are difficulties associated with this method as well. Consider shifting the phase of equation 1.21 by $\pi/2$, which would result in

$$P|_{null} = P_{in} \sin^2(\Delta\phi) \approx P_{in} (kL_+h_+)^2 \quad (1.23)$$

However, this results in second-order dependence on the gravitational-wave signal which is already approximated to be very small.

So a good solution to the issue of how to *read* out a gravitational wave signal from an interferometer is familiar to radio engineers, namely, RF detection methods:

Consider changing the interferometer input by adding an electro-optical modulator (EOM) to sinusoidally modulate the laser frequency and expanding to first order using the Bessel functions,

$$\begin{aligned} E_{in} &= E_0 e^{i(wt + \beta \cos(\Omega t))} \\ &\approx E_0 e^{iwt} [J_0(\beta) + J_1(\beta) e^{i\Omega t} + J_1(\beta) e^{-i\Omega t}] \\ &= E_{C,in} + E_{SB+,in} + E_{SB-,in} \end{aligned} \quad (1.24)$$

where Ω and β are the modulation frequency and depth, respectively. The

It is useful to define an optical transfer functions which transforms the interferometer's input electric field to its output,

$$E_{out} = E_C + E_{SB+} + E_{SB-} = \begin{pmatrix} t_C \\ t_{SB+} \\ t_{SB-} \end{pmatrix} \begin{pmatrix} E_{C,in} & E_{SB+,in} & E_{SB-,in} \end{pmatrix} \quad (1.25)$$

The carrier transfer function, t_C has already been calculated by equations 1.21 - 1.23 and the sideband transfer functions are not much different.

$$t_{SB\pm} = r_{x,\pm} e^{i\phi_{\pm,x}} - r_{y,\pm} e^{i\phi_{\pm,y}} \quad (1.26)$$

where $\phi_{\pm,i} = (k \pm k_\Omega) \ell_i = (\frac{w \pm \Omega}{c}) \ell_i$. In the current example, there is no difference in reflectivity between the sidebands and carrier fields, however, this will not be true in general when integrating Fabry-Perot cavities. Plugging equation 1.26 into 1.25, the output electric field becomes

$$E_{out} = ie^{i\omega t} [J_0(\beta)k\ell_+h_+ + J_1(\beta)\sin(k\Delta\ell + k_\Omega\Delta\ell)e^{i\Omega t} + J_1(\beta)\sin(k\Delta\ell + k_\Omega\Delta\ell)e^{-i\Omega t}] \quad (1.27)$$

By choosing the carrier signal to be on the dark fringe, $k\Delta\ell = \pi/2$ but the sidebands to be slightly off the null (Figure []) and hence leak into the anti-symmetric port, the electric field reduces to

$$E_{out} = ie^{i\omega t} [J_0(\beta)k\ell_+h_+ + J_1(\beta)\sin(k_\Omega\Delta\ell)e^{i\Omega t} + J_1(\beta)\sin(k_\Omega\Delta\ell)e^{-i\Omega t}] \quad (1.28)$$

Recall that the intensity is equal to the electric field squared,

$$\begin{aligned} I = |E_{out}|^2 &= |E_C|^2 + |E_{SB+}|^2 + |E_{SB-}|^2 \\ &+ 2\text{Re}\{ E_{SB+}E_{SB-}^*e^{2i\Omega t} \} \\ &+ 2\text{Re}\{ (E_C E_{SB-}^* + E_{SB+}E_C^*)e^{i\Omega t} \} \end{aligned} \quad (1.29)$$

The last term is referred to as the *beat note* between the carrier signal and the sidebands. It is possible to extract only the signal at the modulation frequency using a mixer and low-pass filter to get a demodulated signal. The mixer is an analog device which takes in two signals and outputs the product of the two inputs. Using a local oscillator at the modulation frequency, the demodulated intensity is

$$\begin{aligned} I_{Demod} &\propto [4\pi J_0(\beta)J_1(\beta)\frac{\ell}{\lambda}\sin(k_\Omega\Delta\ell)h_+][\cos(\Omega t)\sin(\Omega t + \phi_{Demod})] \\ &= [4\pi J_0(\beta)J_1(\beta)\frac{\ell}{\lambda}\sin(k_\Omega\Delta\ell)h_+][\sin(\phi_{Demod}) + \sin(2\Omega t + \phi_{Demod})] \end{aligned} \quad (1.30)$$

where ϕ_{Demod} is the phase that can be set by the user in order to practically extract the right signal. For the output, the After the mixer, there will be signals at DC, Ω , 2Ω and so on. However, the part that is linear in the gravitational wave amplitude will be at DC so a low-pass filter will allow the final signal to dominate:

$$S = 4\pi J_0(\beta)J_1(\beta)\frac{\ell}{\lambda}\sin(k_\Omega\Delta\ell)\sin(\phi_{Demod})h_+ \quad (1.31)$$

This shows that a RF detection technique will be linear in GW signal with no large DC offset. However, this method depends on the local oscillator's stability. Reference [Fritschel, Hild]

Setting the carrier on a null point means $\Delta\ell = \frac{k_\Omega}{k}\frac{\pi}{2}$ and allows the designer to optimize the Schnupp asymmetry length to get the best signal for some modulation frequency.

This type of readout scheme is called Heterodyne detection, where the local oscillator signal is provided

by an EOM. However, the Advanced LIGO scheme uses Homodyne detection in order to derive the GW signal but

Disadvantages of heterodyne scheme: fluctuations in the local oscillator via the modulation depth changing or the phase oscillation.

Fix: Homodyne scheme, ref[Tobin and Nick Smith], this has the advantage of being mode-matched to the carrier beam and lower quantum noise and easier to do squeezing. see section []

[Black Paper on Signal extraction,]

1.2.2 Fabry-Perot Cavities

There are two ways to make our instrument better: one is to increase the sensitivity to gravitational waves and the other is to make our noise lower. From equation 1.18, the gravitational wave signal is proportional to the optical path length that the photon travels, which means the most straightforward method of increasing the sensitivity is to make the arms as long as possible (up to the null point described by equation 1.19. Generally, there were two methods to do this: a delay line or a Fabry-Perot resonator. At the time of writing this thesis, all modern gravitational wave detectors use the latter method.

A Fabry-Perot cavity is an optical system comprised of two mirrors and one laser input. The very simple condition that once the laser has made a round trip around the optics, it is the same shape and size. Conceptually, this may seem simple but in practice, controlling and sensing any optical cavity comes with many challenges.

Plane wave analysis: Consider a two mirror system separated by a length L with reflection and transmission coefficients: r_1, t_1, r_2, t_2 .

Starting with a plane wave at the input mirror with amplitude E_0 [1],

the reflection coefficient is

$$r_{FP} = -r_1 + \frac{t_1^2 r_2 e^{-i2kL}}{1 - r_1 r_2 e^{-i2kL}} \quad (1.32)$$

the transmission coefficient is

$$t_{FP} = \frac{t_1 t_2 e^{ikL}}{1 - r_1 r_2 e^{-i2kL}} \quad (1.33)$$

the circulating coefficient is

$$c_{FP} = \frac{t_1}{1 - r_1 r_2 e^{-i2kL}} \quad (1.34)$$

Resonance Condition: $L = n\lambda/2$

If you are on resonance, the circulating coefficient in the cavity is maximized so that the gain is

$$\text{Gain} = c_{FP}^2|_{\text{resonating}} = \left(\frac{t_1}{1 - r_1 r_2} \right)^2 \quad (1.35)$$

Frequency response of a single FP (plot): [2] A cavity of fixed length and frequency, the circulating power becomes

which is maximized when

$$\frac{\Delta w}{w} = -\frac{\Delta L}{L} \quad (1.36)$$

Notice that the circulating power depends on the length of the cavity as well as the frequency and one might naively determine that modulating the two parameters independently cause the same effect. However, when both the frequency and the length are changing, the resonant light they are related by a frequency dependent transfer function

$$C(s) \frac{\Delta w}{w} = -\frac{\Delta L}{L} \quad (1.37)$$

where

$$C(s) = \frac{1 - e^{-2sL/c}}{2sL/c} \quad (1.38)$$

in the Laplace domain.

Finesse, or the line width of the resonant peak, function of r_1 and r_2 :

$$\mathbb{F} = \frac{\pi \sqrt{r_1 r_2}}{1 - r_1 r_2} \quad (1.39)$$

Storage Time :

$$\tau_s = \frac{L}{c\pi} \mathbb{F} \quad (1.40)$$

Cavity Pole:

$$f_{pole} = \frac{1}{4\pi\tau_s} \quad (1.41)$$

Free Spectral Range:

$$f_{FSR} = \frac{c}{2L} \quad (1.42)$$

Stability: In order to prove that the Fabry Perot is stable, it is useful to introduce the matrices that describe a periodic optical system which is explicitly derived in appendix A. A Fabry Perot cavity that is separated by distance L with spherical mirrors that have radii of curvature R_1 and R_2 will need to satisfy

$$0 \geq \left(1 + \frac{L}{R_1}\right) \left(1 + \frac{L}{R_2}\right) \geq 1 \quad (1.43)$$

in order to be geometrically stable.

Power circulating as a function of our defined parameters slightly off resonance by a length of δL :

$$P_{cav} = |c_{FP}|^2 = \frac{Gain}{1 + \left(\frac{2\mathbb{F}}{\pi}\right)^2 \sin^2(k\delta L)} \quad (1.44)$$

Modal Contents

Locking a Fabry Perot Cavity

This is described everywhere, reference Black, Drever, Jenne, Evan Hall. Just give the highlights.

Described above are the theoretical constructs of a FP cavity, but the question remains, how does one practically construct a resonant optical cavity? The answer comes from using a heterodyne sensing scheme similar to the one described in section []. Except the optical system is not a Michelson interferometer but rather a two mirror cavity, however, this can apply to a number of different geometries such as triangular or bow-tie cavities shown in Figure []. All of which are used in LIGO for various reasons.

Starting with a input laser and an EOM (electro-optical modulator) that imparts upper and lower sidebands at a modulation frequency Ω , the user injects three beams into the optical system exactly as in equation 1.24. When placing a photodetector on the cavity reflection port, one should see the effects of the cavity on each of the three electric fields.

$$E_{FP,out} = E_C + E_{SB+} + E_{SB-} = \begin{pmatrix} r_C \\ r_{SB+} \\ r_{SB-} \end{pmatrix} \begin{pmatrix} E_{C,in} & E_{SB+,in} & E_{SB-,in} \end{pmatrix} \quad (1.45)$$

where the reflection coefficients follow equation 1.32. It is obvious that the phase accumulated by the cavity differ between the carrier and sidebands so that the beat frequency will be proportional to the phase difference. To make a good reference for the resonant carrier beam, the sidebands should be anti-resonant in the cavity.

$$r_C = -r_1 + \frac{t_1^2 r_2 e^{-i2kL}}{1 - r_1 r_2 e^{-i2kL}} \quad (1.46)$$

and

$$r_{SB\pm} = -r_1 + \frac{t_1^2 r_2 e^{-i2(k+k_\Omega)L}}{1 - r_1 r_2 e^{-i2(k+k_\Omega)L}} \quad (1.47)$$

This time,

Application to LIGO

Frequency response of a FP Michaelson: Just write it down. References:[Black, Rick's Paper]

1.2.3 Power-Recycled Fabry-Perot Interferometers

Power Recycling If the interferometer is operating such that the 4 km arms are exactly different in arms a π over two times the wavelength, then the intensity of the light at antisymmetric port will be close to null. This means the power from the arms will reflect back towards the input laser. Ref[] shows the effect of adding a partially reflecting mirror to increase the optical gain of the Michaelson for the sidebands and carrier fields.

References: [Meers, Kiwamu]

1.2.4 Dual-Recycled Fabry-Perot Interferometers

One of the biggest changes made between initial and advanced LIGO was the addition of a signal recycling mirror at the antisymmetric port shown in Figure []. This extra mirror allows an extra degree of freedom in shaping the sensitivity curve such.

Figure comparing the three cases of increased sensitivity.

References: [Kiwamu, Buonanno and Chen]

1.2.5 Fundamental Noise Sources

Ref: Evan Hall, GWINC

Noise budget:

Quantum Noise

One fundamental noise source that is limiting LIGO's sensitivity arises from the fluctuation of quantum vacuum entering the anti-symmetric port and coupling to the electric fields.

Seismic Noise

Thermal Noise

Newtonian Noise

Although there are some noise sources that can be reduced using increasingly complicated techniques such as more levels of seismic isolation or quantum nondemolition devices. Newtonian noise that is caused by

fluctuating gravitational fields arising from the wave motion of the earth is one that cannot easily be reduced.

1.3 Squeezed States of Light

The Quantum Noise is a fundamental source that can be improved by modifying the quantum vacuum using correlated photons and injecting the states of light into the antisymmetric port of the interferometer. Within the LIGO community, this procedure of modifying quantum vacuum is called squeezing. Caves analytically derived the effects of quantum noise on the interferometer as well as the improvement due to i

References: [Caves, Dwyer, Kwee, Miao]

Chapter 2

The Fundamentals of Mode-Matching

Theory section of modematching. Gaussian beams, define relevant quantities, Gouy Phase!

In order to properly explain the effects of electromagnetic waves in the LIGO interferometers, it is useful to review the

References: [Kogonik and Li]

2.1 Gaussian Beam Optics

When dealing with length sensing degrees of freedoms such as section 1.2.2, the simple plane wave approximation is sufficient in describing the dynamics, however, when trying to understand the misalignment and mode-mismatch signals, it is necessary to incorporate Gaussian beams and associated their higher order modes.

Consider the famous Maxwell's equations in vacuum:

$$\begin{aligned}\nabla \times \mathbf{E} &= -\frac{\partial \mathbf{B}}{\partial t}, \\ \nabla \cdot \mathbf{B} &= 0, \\ \nabla \times \mathbf{B} &= \mu \mathbf{J} + \frac{1}{c^2} \frac{\partial \mathbf{E}}{\partial t} \\ \nabla \cdot \mathbf{E} &= \frac{\rho}{\epsilon}\end{aligned}\tag{2.1}$$

Concentrating on the electric field in vacuum, we arrive at the Helmholtz Equation

$$(\nabla^2 + k^2)\mathbf{U}(\mathbf{r}, t) = 0\tag{2.2}$$

where $k = \frac{2\pi\nu}{c}$ is the wave number and $\mathbf{U}(\mathbf{r}, t)$ is the complex amplitude which can describe either the

electric or magnetic fields.

It is possible to express the solution to equation 2.2 as a plane wave with a modulated complex envelope

$$\mathbf{U}(\mathbf{r}) = \mathbf{A}(\mathbf{r})e^{-ikz} \quad (2.3)$$

By imposing the constraints which force the envelope to vary slowly with respect to the z-axis within the distance of one wavelength $\lambda = 2\pi/k$,

$$\left| \frac{\partial^2 \mathbf{A}}{\partial z^2} \right| \ll \left| k \frac{\partial \mathbf{A}}{\partial z} \right| \quad (2.4a)$$

$$\left| \frac{\partial^2 \mathbf{A}}{\partial z^2} \right| \ll \left| k \frac{\partial^2 \mathbf{A}}{\partial x^2} \right| \quad (2.4b)$$

$$\left| \frac{\partial^2 \mathbf{A}}{\partial z^2} \right| \ll \left| k \frac{\partial^2 \mathbf{A}}{\partial y^2} \right| \quad (2.4c)$$

the partial differential equation which arises is called the Paraxial Helmholtz Equation:

$$\nabla_T^2 A(r) - i2k \frac{\partial A(r)}{\partial z} = 0 \quad (2.5)$$

where $\nabla_T^2 = \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2}$ is the transverse Laplacian. A simple solution for equation 2.5 is the complex paraboloidal wave

$$A(\mathbf{r}) = \frac{A_0}{q(z)} e^{\frac{-ikr^2}{2q(z)}}, \quad q(z) = z + iz_0 \quad (2.6)$$

where z_0 is the Rayleigh range and is directly proportional to the square of the waist size. In order to separate the amplitude and phase portions of the wave, it is useful to rewrite $q(z)$ as

$$\frac{1}{q(z)} = \frac{1}{R(z)} - i \frac{\lambda}{\pi W^2(z)} \quad (2.7)$$

Plugging equation 2.7 into 2.6 leads directly to the complex amplitude for a Gaussian Beam

$$U(r) = A_0 \frac{W_0}{W(z)} e^{-\frac{r^2}{W^2(z)}} e^{-ikz - ik \frac{r^2}{2R(z)} + i\phi(z)} \quad (2.8)$$

where

$$W(z) = W_0 \sqrt{1 + \left(\frac{z}{z_0} \right)^2} \quad (2.9a)$$

$$R(z) = z \left[1 + \left(\frac{z}{z_0} \right)^2 \right] \quad (2.9b)$$

$$\phi(z) = \tan^{-1}\left(\frac{z}{z_0}\right) \quad (2.9c)$$

$$W_0 = \sqrt{\frac{\lambda z_0}{\pi}} \quad (2.9d)$$

Gouy Phase

The Gouy phase is - Gouy Phase + π

Intensity

- Intensity and Power

Hermite-Gauss Modes

The fundamental Gaussian beam is not the only solution which can be used to solve equation 2.5. In fact, there exists a complete set of solutions that can solve the paraxial Helmholtz Equation, which are referred to as the higher order modes (HOMs)

$$U_{mn}(x, y, z) = A_{mn} \left[\frac{W_0}{W(z)} \right] \mathbb{G}_m \left(\frac{\sqrt{2}x}{W(z)} \right) \mathbb{G}_n \left(\frac{\sqrt{2}y}{W(z)} \right) \times \exp \left\{ -ikz - \frac{ik(x^2 + y^2)}{2R(z)} + i(m + n + 1)\phi(z) \right\} \quad (2.10)$$

where,

$$\mathbb{G}(u) = \mathbb{H}(u) \exp(-u^2/2) \quad (2.11)$$

and $\mathbb{H}(u)$ are the well known Hermite polynomials. It is important to mention that the Gouy phase of the complex amplitude is different than the fundamental Gaussian beam by a factor of $(m + n + 1)$. Also, the intensity distribution of these higher order modes are much different. Both of these facts will become extremely important in the following wavefront sensing discussion.

It is useful to normalize the Hermite-Gauss modes with respect to the overall power, which are derived in Appendix[].

Laguerre Modes

Another complete set of alternative solutions to equation 2.5 exists which are called the Laguerre-Gauss modes

$$\begin{aligned}
V_{\mu\nu}(\rho, \theta, z) &= A_{\mu\nu} \left[\frac{W_0}{W(z)} \right] \mathbb{L}_{\nu}^{\mu} \left(\frac{\sqrt{2}x}{W(z)} \right) \\
&\times \exp \left\{ -ikz - \frac{ik\rho^2}{2R(z)} + i(\mu + 2\nu + 1)\phi(z) \right\}
\end{aligned} \tag{2.12}$$

where $\mathbb{L}_{\nu}^{\mu} \left(\frac{\sqrt{2}x}{W(z)} \right)$ is the Laguerre polynomial function. Both equations 2.10 and 2.12 are able to fully describe any complex electromagnetic amplitude; and because they both form complete sets, there is a rotation which can map from one basis to the other [ref Bond and Biejergasern]

$$U_{\mu\nu}^{LG}(x, y, z) = \sum_k^N i^k b(n, m, k) U_{N-k, k}^{HG}(x, y, z) \tag{2.13}$$

where

$$b(n, m, k) = \sqrt{\left(\frac{(N-k)!k!}{2^N n!m!} \right)} \frac{1}{k!} \frac{d^k}{dt^k} [(1-t)^m (1+t)^m] |_{t=0} \tag{2.14}$$

2.1.1 Misalignment and Higher Order Modes

Morrison and Anderson derived a simplistic way of how small misalignments and mismodematched cavities can couple the fundamental Gaussian beam into various higher order modes. This is done by taking a linear cavity and using its perfectly matched Gaussian beam as a reference, and then varying the input electric field with small perturbations and expanding in terms of the cavity modes. As long as the mismatches are small, it is possible to consider only the first few terms of the expansion which have gained power from the fundamental mode.

Consider the first three modes of equation 2.10 in one dimension and normalized to set the total optical power to unity (derived in Appendix[]):

$$\begin{aligned}
U_0(r) &= \left(\frac{2}{\pi w^2(z)} \right)^{1/4} e^{-r^2/w^2(z)} \\
U_1(r) &= \left(\frac{2}{\pi w^2(z)} \right)^{1/4} \frac{2r}{w(z)} e^{-r^2/w^2(z)} \\
U_2(r) &= \left(\frac{2}{\pi w^2(z)} \right)^{1/4} \frac{1}{\sqrt{2}} \left(\frac{4r^2}{w^2(z)} - 1 \right) e^{-r^2/w^2(z)}
\end{aligned} \tag{2.15}$$

Beam Axis Tilted

If the input beam into an optical cavity is tilted by an angle α with respect to the nominal cavity axis as shown in Figure [], the wave front of the input beam will have an extra phase propagation relative to the cavity that is approximately proportional to $e^{ik\alpha r}$. By implementing the small angle approximation, which

is valid if the misalignment is much smaller than the divergence angle of the fundamental mode $k\alpha r \ll 1$, the resultant input beam is

$$\Psi \approx U_0(r)e^{ik\alpha r} \approx U_0(r)(1 + ik\alpha r) = U_0(r) + \frac{ik\alpha w(z)}{\sqrt{2\pi}}U_1(r) \quad (2.16)$$

Here the factor associated with the first higher order mode is complex, indicating there is a 90 degree phase difference between the fundamental and off axis mode.

Beam Axis Displaced

If the input beam is displaced in the transverse direction by a quantity Δr , the resultant waveform will be

$$\begin{aligned} \Psi &= U_0(r + \Delta r) \\ &= \left(\frac{2}{\pi w^2(z)} \right)^{1/4} e^{-(r+\Delta r)^2/w^2(z)} \\ &= \left(\frac{2}{\pi w^2(z)} \right)^{1/4} e^{-(r^2+2r\Delta r+\Delta r^2)/w^2(z)} \\ &\approx \left(\frac{2}{\pi w^2(z)} \right)^{1/4} e^{-r^2/w^2(z)} e^{-2r\Delta r/w^2(z)} \\ &\approx \left(\frac{2}{\pi w^2(z)} \right)^{1/4} e^{-r^2/w^2(z)} \left(1 - \frac{2r\Delta r}{w^2(z)} \right) \\ &= \left(U_0(r) - \sqrt{\frac{2}{\pi}} \frac{\Delta r}{w(z)} U_1(r) \right) \end{aligned} \quad (2.17)$$

Similarly to a tilted input beam axis, the displaced beam axis couples power to the first higher order mode, however, the latter does not have a 90 degree phase difference previously seen in the former. This point is of extreme importance when trying to discern between the two effects as shown in Section []. Although comparing the two cases in Figure, one can already see the difference between the wavefronts in the near field, $z \ll z_R$, and the far field, $z \gg z_R$.

In the near field, there is no phase difference due to a displaced beam, but there is one for a tilted beam. Conversely, in the far field, there is no phase difference due to a tilted beam, but there is one from a displaced beam. In order to implement a closed loop feedback system, the wavefront sensors discussed in Section 2.2 will use this precise logic to extract an error signal.

2.1.2 Mode Mismatch and Higher Order Modes

Waist Size Shifted

By considering the effect of evaluating the fundamental mode at the waist position, $z = 0$, but changing the waist size by a small amount ϵ , it is possible to see coupling into higher order modes by expanding to first order.

$$\begin{aligned}
 \Psi &= U_0(r, w(z) = w_0/(1 + \epsilon)) \\
 &= \left(\frac{2}{\pi w_0^2} \right)^{1/4} \sqrt{1 + \epsilon} \ e^{-r^2(1+\epsilon)^2/w_0^2} \\
 &\approx \left(\frac{2}{\pi w_0^2} \right)^{1/4} (1 + \epsilon/2) \ e^{-r^2/w_0^2} \ e^{-2r^2\epsilon/w_0^2} \\
 &\approx \left(\frac{2}{\pi w_0^2} \right)^{1/4} (1 + \epsilon/2) \ e^{-r^2/w_0^2} \ (1 - 2r^2\epsilon/w_0^2) \\
 &\approx \left(\frac{2}{\pi w_0^2} \right)^{1/4} \left(1 + 2\epsilon \left(\frac{1}{4} - \frac{r^2}{w_0^2} \right) \right) \ e^{-r^2/w_0^2} \\
 &= U_0(r) - \frac{\epsilon}{\sqrt{2}} U_2(r)
 \end{aligned} \tag{2.18}$$

Changing the waist size by a small amount will couple the fundamental mode to the in-phase second order Hermite Gauss mode.

Waist Position Shifted

To repeat the process from above with a waist position shift, it is useful to start with a more general equation that includes the phase that is gained from including the radius of curvature,

$$\Psi = \left(\frac{2}{\pi w^2(z)} \right)^{1/4} e^{-r^2/w^2(z)} e^{-ikr^2/2R(z)} \tag{2.19}$$

where $R(z)$ is from equation 2.9b. It is also useful to approximate the shift in waist position along the longitudinal axis is small compared to the Rayleigh range of the beam, $\Delta z \ll z_0$, which leaves the waist size approximately the same and the radius of curvature inversely proportional to the shift.

$$w^2(\Delta z) = w_0^2 \left[1 + \left(\frac{\Delta z}{z_0} \right)^2 \right] \approx w_0^2 \tag{2.20a}$$

$$R(\Delta z) = \Delta z \left(1 + \left(\frac{z_0}{\Delta z} \right)^2 \right) \approx \frac{z_0^2}{\Delta z} \tag{2.20b}$$

Plugging the equations above into 2.19,

$$\begin{aligned}
\Psi &\approx \left(\frac{2}{\pi w_0^2}\right)^{1/4} e^{-r^2/w_0^2} e^{-ikr^2\Delta z/2z_0^2} \\
&\approx \left(\frac{2}{\pi w_0^2}\right)^{1/4} e^{-r^2/w_0^2} \left(1 - \frac{ikr^2\Delta z}{2z_0^2}\right) \\
&= U_0(r) - \left(\frac{2}{\pi w_0^2}\right)^{1/4} e^{-r^2/w_0^2} \frac{ikr^2\Delta z}{2z_0^2} \\
&= U_0(r) - i\frac{\Delta z}{2kw_0^2} \left(4U_2(r) + U_0(r)\right)
\end{aligned} \tag{2.21}$$

The equations above show that a fundamental Gaussian mode that is shifted in waist position will couple power to the second order Hermite Gauss mode. Although changes in the waist size or position couple power to the same mode, they differ by a 90 degrees in phase as denoted by the extra factor of i in the coupling coefficient. By recognizing the two effects are in different quadrature phases will allow a user to design a system to distinguish between the different types of physical couplings, this is shown in Section 2.2.

In order to be physically valid one would need to consider the full two dimensional space so that the equation would encapsulate the full transverse mode, however, the x and y components would follow the exact same derivation. On that point, it is important to note that only the mode mismatch couplings from either a varying waist position or size has higher order modes that are circularly symmetric.

2.2 Wavefront Sensing

Heterodyne detection via modal decomposition of the full electric field allows the use of wavefront sensors to extract an error signal from the optical system. Hefetz et.al Ref[Sigg and Nergis] created a formalism to describe the use of wavefront sensors by creating frequency sidebands which accumulate a different Gouy phase than the electric field at the carrier frequency when passed through the optical system. By observing the demodulated signal of the intensity, it is possible to obtain a linear signal that corresponds to a physical misalignment or mode mismatch.

Fundamentally, the purpose of wavefront sensing is to detect the content of higher order modes due to physical disturbances of the optical cavity (ie. mode mismatch or misalignment). In other words, it is examining the difference of basis sets between the incoming eigenmodes and the cavity eigenmodes.

Consider a general equation for an electric field which is a linear combination of all higher order modes of the complex amplitude

$$E(x, y, z) = \sum_{m,n}^{\infty} a_{mn} U_{mn}(x, y, z) \tag{2.22}$$

where $U_{mn}(x, y, z)$ are the eigenmodes described in equation 2.10 (or 2.12) and a_{mn} is the complex

amplitude. It is also convenient in the following analysis to use vectors when describing the composition of the electric fields.

$$|E(x, y, z)\rangle = \begin{pmatrix} E_{00} \\ E_{01} \\ E_{10} \\ E_{20} \\ E_{02} \end{pmatrix} \quad (2.23)$$

When creating a theory that involves laser beams, it is useful to define operators that are important in describing physical situations. For example, laser beams propagate through space and pick up phase according to equation 2.10 which can be represented by the spatial propagation operator,

$$\hat{P}_{mn,kl} = \delta_{mn}\delta_{kl} \exp[-ik(z_2 - z_1)]\exp[i(m + n + 1)\phi(z)] \quad (2.24)$$

However, it is useful to compare how the fundamental Gaussian mode propagates compared to the higher order modes,

$$\hat{\eta}_{\mu\nu} = \begin{pmatrix} e^{i\phi} & 0 & 0 & 0 & 0 \\ 0 & e^{2i\phi} & 0 & 0 & 0 \\ 0 & 0 & e^{2i\phi} & 0 & 0 \\ 0 & 0 & 0 & e^{3i\phi} & 0 \\ 0 & 0 & 0 & 0 & e^{3i\phi} \end{pmatrix} \quad (2.25)$$

From the above diagonal elements, it is clear that the higher order modes have an extra phase compared to the fundamental 00 mode, this effect will be extremely important on how an error signal can be derived from the optical system.

$$|E(x, y, z_2)\rangle = \hat{M}(x, y, z_1, z_2) |E(x, y, z_1)\rangle \quad (2.26)$$

where $\hat{M}(x, y, z_1, z_2)$ is the misalignment operator. Since we are using the paraxial approximation, the z-components of the misalignment operator are small so we can approximate $\hat{M}(x, y, z_1, z_2) \approx \hat{M}(x, y)$ and the expectation value is

$$M_{mn,kl} = \langle U_{mn}(x, y, z_1) | M(x, y) | U_{kl}(x, y, z_2) \rangle \quad (2.27)$$

where the product is an integral over the transverse space $\iint_{D(x,y)} dx dy$

$$\hat{\Theta}_{\mu\nu} = \begin{pmatrix} 1 & 2i\theta_x & 2i\theta_y & 0 & 0 \\ 2i\theta_x & 1 & 0 & 0 & 0 \\ 2i\theta_y & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \end{pmatrix} \quad (2.28)$$

$$\hat{\mathbb{D}}_{\mu\nu} = \begin{pmatrix} 1 & \alpha_x/\omega_0 & \alpha_y/\omega_0 & 0 & 0 \\ \alpha_x/\omega_0 & 1 & 0 & 0 & 0 \\ \alpha_y/\omega_0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \end{pmatrix} \quad (2.29)$$

$$\hat{\mathbb{Z}}_{\mu\nu} = \begin{pmatrix} 1 & 0 & 0 & \Delta z_x & \Delta z_y \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \Delta z_x & 0 & 0 & 1 & 0 \\ \Delta z_y & 0 & 0 & 0 & 1 \end{pmatrix} \quad (2.30)$$

where $\Delta z_{(x,y)} = \frac{i}{\sqrt{2}} \frac{\lambda b}{2\pi\omega_0}$

$$\hat{\mathbb{Z}}_{0,\mu\nu} = \begin{pmatrix} 1 & 0 & 0 & \Delta z_{0,x} & \Delta z_{0,y} \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \Delta z_{0,x} & 0 & 0 & 1 & 0 \\ \Delta z_{0,y} & 0 & 0 & 0 & 1 \end{pmatrix} \quad (2.31)$$

where $\Delta z_{0,(x,y)} = \frac{1}{\sqrt{2}} \frac{\omega' - \omega_0}{\omega_0}$

Figure: Beat frequency between two modes.

Example: Fabry Perot Cavity

Example: Simple Michaelson

2.3 Effects of Mode-matching on Squeezing

Still not clear to me. In the most simple sense, a loss is akin to coupling quantum vacuum into a squeezed state. References: [Miao, Sheon, Kimble, Dwyer]

Chapter 3

Simulating Mode-Matching with Finesse

3.1 How it works

Summary of how Finesse works (input output matrix), how it handles HOMs

3.2 Finesse Simulations

3.2.1 ALIGO Design with FC and Squeezer

3.2.2 Looking at just Modal Change

3.2.3 QM Limited Sensitivity

3.3 Results

- * Signal recycling cavity mismatches

- * Mismatches before the OMC

- * Mismatch contour graph: Comparing all of ALIGO cavities

- * Optical Spring pops up at 7.4 Hz in the Signal-to-Darm TF, re-run with varying SRM Trans which should.

Chapter 4

Experimental Mode Matching Cavities at Syracuse

In conjunction with Sandoval et al, the adaptive modematching table top experiment was able to show the feasibility of a fully dynamical system.

4.1 Adaptive Mode Matching

Real time digital system and model.

Signal chain.

4.2 Actuators

4.2.1 Thermal Lenses

Fabian's work and UFL paper.

4.2.2 Translation Stages

4.3 Sensors

4.3.1 Bullseye Photodiodes

As stated in section [], we saw that the error signal for a mode-mismatched cavity has cylindrical symmetry due to the beating between the LG-01 and the LG-00 mode. This means that the quadrant photodiodes would not have a way to detect the modal content of a cavity. One way to solve this is to introduce a type of detector that can sense the power outside versus the power inside (ugh re write this).

In [Rana's IO final design document]

Why do we need bullseye photodiodes, mention the geometry of the error signal.

Derivations in the appendix.

Picture of BPD

Pitch and Yaw sensing matrix

Demodulation phase

Explain why we need $\omega_0 = \sqrt{2}r_0$

The ratio of out over in will give:

$$\text{Power Ratio} = \frac{P_2 + P_3 + P_4}{P_1} = \frac{e^{-2r_0^2/\omega_0^2}}{1 - e^{-2r_0^2/\omega_0^2}} \approx 0.582 \quad (4.1)$$

4.3.2 Mode Converters

4.3.3 Scanning Gaussian Beams

Chapter 5

Mode Matching Cavities at LIGO Hanford

5.1 Active Wavefront Control System

5.2 SRC

The importance of mode-matching actually goes beyond reducing the amount of losses in coupled cavities. It also is important for cavity stability. If we look at the g-factor of a cavity, it is required through ABCD transformations that the values lay between 0 and 1. For the signal recycling cavity, if the round trip gouy phase is off by a few millimeters, the stability of the cavity can be compromised.

5.3 Beam Jitter

5.4 Contrast Defect

5.5 Single Bounce Method

Chapter 6

Solutions for Detector Upgrades

- * A full modal picture, sensors and actuators
 - * SR3 Heater
 - * SRM Heater
 - * Bullseye photodetectors
 - * Operation: range (in terms of watts and
 - * Translation stages
 - * Mechanical description (Solidworks designs)
 - * Constraints (range, vacuum, alignment, integration)
 - * Electronics
 - * Software

Appendices

Appendix A

Resonator Formulas

Equation 1.43 describes the stability condition for a two mirror Fabry-Perot cavity. It is worthwhile to derive the criterion for geometric stability from the ray matrix tools commonly used in optics.

Consider two plane waves traveling in space shown by figure [], they differ by two quantities: the axial and angular separations, y and θ , respectively. These two quantities can be transformed via these optical matrices:

Lens

$$\hat{F}_i = \begin{pmatrix} 1 & 0 \\ -\frac{1}{f_i} & 1 \end{pmatrix} \quad (\text{A.1})$$

Curved Mirror

$$\hat{M}_i = \begin{pmatrix} 1 & 0 \\ \frac{2}{R_i} & 1 \end{pmatrix} \quad (\text{A.2})$$

Space

$$\hat{D}_i = \begin{pmatrix} 1 & d_i \\ 0 & 1 \end{pmatrix} \quad (\text{A.3})$$

Periodic Fabry-Perot with two mirrors is

$$\begin{pmatrix} y_{m+1} \\ \theta_{m+1} \end{pmatrix} = \hat{M}_{FP} \begin{pmatrix} y_m \\ \theta_m \end{pmatrix} \quad (\text{A.4})$$

where

$$\hat{M}_{FP} = \hat{M}_i \hat{D}_i \hat{M}_i \hat{D}_i \quad (\text{A.5})$$

is the optical transfer matrix. The goal is to find a geometric condition that is dependent on the optical transfer matrix which keeps the axial displacement from diverging.

$$\begin{pmatrix} y_{m+1} \\ \theta_{m+1} \end{pmatrix} = \begin{pmatrix} A & B \\ C & D \end{pmatrix} \begin{pmatrix} y_m \\ \theta_m \end{pmatrix} \quad (\text{A.6})$$

which means

$$\theta_m = \frac{y_{m+1} - A y_m}{B} \quad (\text{A.7a})$$

$$\theta_{m+1} = \frac{y_{m+2} - A y_{m+1}}{B} \quad (\text{A.7b})$$

Solving for y_{m+2}

$$y_{m+2} = (A + D) y_{m+1} - \det(\hat{M}_{FP}) y_m \quad (\text{A.8})$$

Assuming a geometrical solution where $y_m = y_o h^m$ and plugging into the equation above,

$$h^2 = (A + D) h - \det(\hat{M}_{FP}) \quad (\text{A.9})$$

which is a quadratic equation that has two solutions and can be further simplified if the index of refraction for the entire system remains constant such that $\det(\hat{M}_{FP}) = 1$. Plugging back into y_m and doing some algebra

$$y_m \propto \sin(m\phi) \quad (\text{A.10})$$

where $\phi = \cos^{-1}(\frac{A+D}{2})$, which is also referred to as the round trip Gouy phase of the cavity. In order for y_m to be harmonic instead of hyperbolic and hence confined, this condition must be met

$$\frac{|A + D|}{2} \leq 1 \quad (\text{A.11})$$

By actually calculating the terms of \hat{M}_{FP} and doing even more algebra, it is clear that

$$0 \geq \left(1 + \frac{L}{R_1}\right) \left(1 + \frac{L}{R_2}\right) \geq 1 \quad (\text{A.12})$$

which is what was stated in equation 1.43.

There is a simpler and less algebraic way to reach the same conclusion by looking at the Rayleigh range

of a finite Gaussian beam for a simple cavity. In Table II of Kogelnik and Li ref[Kogelnik], there is an expression for the Rayleigh range

$$\begin{aligned} z_R^2 &= \frac{L(R_1 - L)(R_2 - L)(R_1 + R_2 - L)}{(R_1 + R_2 - 2L)^2} \\ &= \frac{g_1 g_2 (1 - g_1 g_2)}{(g_1 - g_2 - 2g_1 g_2)^2} \end{aligned} \tag{A.13}$$

If the Rayleigh range is a real number, then once again, equation 1.43 must be true.

Effect of higher order modes into the cavity, mode scanning. All comes from round trip Gouy phase. -
RT Gouy Phase - HOM Coupling

Appendix B

Hermite Gauss Normalization

According to equation 2.10, the higher order modes in the Hermite Gauss basis has the intensity profile,

$$I_{mn}(x, y, z) = |A_{mn}|^2 \left[\frac{W_0}{W(z)} \right]^2 \mathbb{G}_n^2 \left(\frac{\sqrt{2}x}{W(z)} \right) \mathbb{G}_n^2 \left(\frac{\sqrt{2}y}{W(z)} \right) \quad (\text{B.1})$$

It is useful to normalize the first few lowest order modes with respect to the total optical power since the Gaussian beam will couple to them the most due either misalignment or mode mismatch as seen in section [3](#).

In one dimension, the total optical power for the first 3 modes are

$$\begin{aligned} P_0(x, y, z) &= \int_{-\infty}^{\infty} |A_0|^2 \left[\frac{W_0}{W(z)} \right]^2 e^{-2x^2/w^2(z)} dx \\ P_1(x, y, z) &= \int_{-\infty}^{\infty} |A_1|^2 \left[\frac{W_0}{W(z)} \right]^2 \frac{8x^2}{w^2(z)} e^{-2x^2/w^2(z)} dx \\ P_2(x, y, z) &= \int_{-\infty}^{\infty} |A_2|^2 \left[\frac{W_0}{W(z)} \right]^2 \left(\frac{8x^2}{w^2(z)} - 2 \right)^2 e^{-2x^2/w^2(z)} dx \end{aligned} \quad (\text{B.2})$$

In two dimensions, the total optical power for the first 3 modes are

$$\begin{aligned} P_{00}(x, y, z) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} |A_{00}|^2 \left[\frac{W_0}{W(z)} \right]^2 e^{-2x^2/w^2(z)} e^{-2y^2/w^2(z)} dx dy \\ P_{10}(x, y, z) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} |A_{10}|^2 \left[\frac{W_0}{W(z)} \right]^2 \frac{8x}{w^2(z)} e^{-2x^2/w^2(z)} e^{-2y^2/w^2(z)} dx dy \\ P_{20}(x, y, z) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} |A_{20}|^2 \left[\frac{W_0}{W(z)} \right]^2 \left(\frac{8x^2}{w^2(z)} - 2 \right)^2 e^{-2x^2/w^2(z)} e^{-2y^2/w^2(z)} dx dy \end{aligned} \quad (\text{B.3})$$

By setting the equations above to unity, the normalization factors become

$$\begin{aligned}
A_0 &= \left(\frac{2}{\pi w_0^2} \right)^{1/4} \\
A_1 &= \left(\frac{2}{\pi w_0^2} \right)^{1/4} \frac{1}{\sqrt{2}} \\
A_2 &= \left(\frac{2}{\pi w_0^2} \right)^{1/4} \frac{1}{\sqrt{8}}
\end{aligned} \tag{B.4}$$

$$\begin{aligned}
A_{00} &= \sqrt{\frac{2}{\pi w_0^2}} \\
A_{10} &= \sqrt{\frac{1}{\pi w_0^2}} \\
A_{20} &= \sqrt{\frac{1}{4\pi w_0^2}}
\end{aligned} \tag{B.5}$$

Therefore the normalized modes are

Appendix C

Bullseye Photodiode Characterization

C.1 DC

$$\begin{aligned} \text{Power} &= \int_A^B |A_{00}|^2 e^{\frac{-2r^2}{\omega_0^2}} 2\pi r dr \\ &= -|A_{00}|^2 \frac{\pi \omega_0^2}{2} e^{\frac{-2r^2}{\omega_0^2}} \Big|_A^B \end{aligned} \quad (\text{C.1})$$

$$P_{in} = \text{Power} \Big|_0^{r_0} = |A_{00}|^2 \frac{\pi \omega_0^2}{2} [1 - e^{\frac{-2r_0^2}{\omega_0^2}}] \quad (\text{C.2})$$

$$P_{out} = \text{Power} \Big|_{r_0}^{\infty} = |A_{00}|^2 \frac{\pi \omega_0^2}{2} [e^{\frac{-2r_0^2}{\omega_0^2}}] \quad (\text{C.3})$$

$$P_{total} = P_{in} + P_{out} \quad (\text{C.4})$$

$$\omega = \sqrt{\frac{P_{total}}{|A_{00}|^2 \pi / 2}} \quad (\text{C.5})$$

$$\text{DC Power Ratio} = \frac{P_{out}}{P_{in}} = \frac{e^{-2r_0^2/\omega_0^2}}{1 - e^{-2r_0^2/\omega_0^2}} \approx 0.582 \quad (\text{C.6})$$

C.2 RF

$$\begin{aligned} P_{RF} &= \int_A^B |A_{01}|^2 \left(1 - \frac{2r^2}{\omega_0^2}\right) e^{\frac{-2r^2}{\omega_0^2}} 2\pi r dr \\ &= -|A_{00}|^2 \frac{\pi}{2} \omega_0^2 e^{\frac{-2r^2}{\omega_0^2}} \left(1 + \frac{4r^2}{\omega_0^2}\right) \Big|_A^B \end{aligned} \quad (\text{C.7})$$

$$P_{in} = P_{RF} \Big|_0^{r_0} = -|A_{01}|^2 \frac{\pi}{2} \omega_0^2 \left(e^{\frac{-2r^2}{\omega_0^2}} \left(1 + \frac{4r_4}{\omega_0^4} \right) - 1 \right) \quad (C.8)$$

$$P_{out} = P_{RF} \Big|_{r_0}^{\infty} = -|A_{01}|^2 \frac{\pi}{2} \omega_0^2 e^{\frac{-2r^2}{\omega_0^2}} \left(1 + \frac{4r_4}{\omega_0^4} \right) \quad (C.9)$$

$$\text{RF Power Ratio} = \frac{P_{out}}{P_{in}} = \frac{e^{-2r_0^2/\omega_0^2}}{1 - e^{-2r_0^2/\omega_0^2}} \approx 2.7844 \quad (C.10)$$

Appendix D

Overlap of Gaussian Beams

Referenced in section

The full Gaussian beam overlap is important in quantitatively defining the amount of power loss obtained when a cavity is mismatched a incoming laser field.

First we define an arbitrary Gaussian beam in cylindrical coordinates:

$$\begin{aligned} |A(r)\rangle &= \frac{A_0}{q(z)} e^{\frac{-ikr^2}{2q(z)}} \\ &= \frac{A_0}{q(z)} e^{\frac{-ikr^2(z-iz_0)}{2|q(z)|^2}} \end{aligned} \quad (\text{D.1})$$

where A_0 is a real amplitude, $q(z) = z + iz_0$ is the complex beam parameter, k is the wave number, and r is the radial variable in the transverse direction.

First we normalize the overlap integral to unity:

$$\langle A(r)|A(r)\rangle = \frac{|A_0|^2}{z^2 + z_0^2} \int_0^\infty e^{\frac{-kr^2 z_0}{|q(z)|^2}} 2\pi r dr = 1 \quad (\text{D.2})$$

Normalization factor is this:

$$A_0 = \sqrt{\frac{kz_0}{\pi}} \quad (\text{D.3})$$

For two Gaussian beams with arbitrary q-parameters

$$|A_i\rangle = \frac{A_{0,i}}{q_i} e^{\frac{-ikr^2(z-iz_0)}{2|q_i|^2}} \quad (\text{D.4})$$

where $z_{0,i}$ is the waist size of one particular beam.

The amplitude overlap is:

$$\langle A_1 | A_2 \rangle = 2i \frac{z_{0,1} z_{0,2}}{q_1 - q_2^*} \quad (\text{D.5})$$

So the power overlap is:

$$\text{Power Overlap} = |\langle A_1 | A_2 \rangle|^2 = 4 \frac{z_{0,1} z_{0,2}}{|q_1 - q_2^*|^2} \quad (\text{D.6})$$

List of Figures

List of Tables

Bibliography

- [1] Peter Saulson. *Fundamentals of Gravitational Wave Detectors*. World Scientific, 2016. 6, 12
- [2] M. J. Lawrence, B. Willke, M. E. Husman, E. K. Gustafson, and R. L. Byer. Dynamic response of a fabry-perot interferometer. *J. Opt. Soc. Am. B*, 16(4):523–532, Apr 1999. 13