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**Supplement to Deliverable 2.2
(Extension of the monaural model)**

**The auditory front-end framework
User manual**



Two!EARS

WP2 *



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1 Executive summary

The goal of the Two!EARS project is to develop an intelligent, active computational model of auditory perception and experience in a multi-modal context. The auditory front-end (AFE) represents the first stage of the system architecture and concerns bottom-up auditory signal processing, which transforms binaural signals into multi-dimensional auditory representations. The output provided by this AFE consists of several transformed versions of ear signals enriched by perception-based descriptors which form the input to the higher model stages. Specific emphasis is given on the modularity of the software framework, making this AFE more than just a collection of models documented in the literature. Bottom-up signal processing is implemented as a collection of processor modules, which are instantiated and routed by a manager object. A variety of processor modules is provided to compute auditory cues such as ratemaps, interaural time and level differences, interaural coherence, onsets and offsets. An object-oriented approach is used throughout, giving benefits of reusability, encapsulation and extensibility. This affords great flexibility, and allows modification of bottom-up processing in response to feedback from higher levels of the system during run time. Such top-down feedback could, for instance, lead to on-the-fly changes in parameter values of peripheral modules, like the filter bandwidths of the basilar-membrane filters. In addition, the object-oriented AFE framework allows direct switching between alternative peripheral filter modules, while keeping all other components unchanged, allowing for a systematic comparison of alternative processors. Finally, the AFE framework supports online processing of the two-channel ear signals.

In the description of work (DOW) for the TWO!EARS project, the deliverable D2.2, documenting the auditory software framework, has the nature of a (public) demonstration. This deliverable has been realized by making the AFE software package publicly available at <https://github.com/TWOEARS/auditory-front-end>. In addition, the consortium members decided to provide a manual for use of the software both within and beyond the consortium. This supplementary document thus serves as a manual to describe the fundamentals of the AFE framework and to help others to make use of its range of capabilities. Chapter 2 gives an overview about the framework functionality. A more technical description of the framework is presented in chapter 3. Chapter 4 gives a detailed overview about the auditory representations that are supported by the framework. To support the concept of “reproducible research”, appendix A comprises a list of all demo files used in the description of the individual modules in Chapter 4. These can be freely accessed and allow all users a replication of the examples in their own software environment.

2 The auditory front-end framework

2.1 Framework functionality

The purpose of the Two!EARS auditory front-end (AFE) is to extract a *subset* of common auditory representations from a binaural recording or from a *stream* of binaural audio data. These representations are to be used later by higher modeling or decision stages. This short description of the role of the AFE highlights its three fundamental properties:

- The framework operates on a request-based mechanism and extracts the *subset* of all available representations which has been requested by the user. Most of the available representations are computed from other representations, i.e., they *depend* on other representations. Because different representations can have a common dependency, the available representations are organized following a “dependency tree”. The framework is built such as to respect this structure and limit redundancy. For example, if a user requests A and B, both depending on a representation C, the software will not compute C twice but will instead reuse it. As will be presented later, to achieve this, the processing is shared among processors. Each processor is responsible for one individual step in the extraction of a given representation. The framework then instantiates only the necessary processors at a given time.
- It can operate on a *stream* of input data. In other words, the framework can operate on consecutive chunks of input signal, each of arbitrary length, while returning the same output(s) as if the whole signal (i.e., the concatenated chunks) was used as input.
- The user request can be modified at *run time*, i.e., during the execution of the framework. New representations can be requested, or the parameters of existing representations can be changed in between two blocks of input signal. This mechanism is particularly designed to allow higher stages of the whole Two!EARS framework to provide feedback, requesting adjustments to the computation of auditory representations. In connection to the first point above, when the user requests such a change, the framework will identify where in the dependency tree the requested change starts affecting the processing and will only compute the steps affected.

2.2 Getting started

The Two!EARS AFE framework was developed entirely using MATLAB version 8.3.0.532 (R2014a). It was tested for backward compatibility down to MATLAB version 8.0.0.783 (R2012b). The source code, test and demo scripts are all available from the public repository at <https://github.com/TWOEARS/auditory-front-end>.

The files for the AFE are divided in three folders, `/doc`, `/src` and `/test` containing respectively the documentation of the framework, the source code, and various test scripts. Once MATLAB opened, the source code (and if needed the other folders) should be added to the MATLAB path. This can be done by executing the script `startAuditoryFrontEnd` in the main folder:

```
>> startAuditoryFrontEnd
```

As will be seen in the following subsection, the framework is request-based: the user places one or more requests, and then informs the framework that it should perform the processing. Each request corresponds to a given auditory representation, which is associated with a short *nametag*. The command `requestList` can be used to get a summary of all supported auditory representations:

```
>> requestList
The following requests for Two!Ears Auditory Front-End processing are
currently valid:
    'time'
    'filterbank'
    'innerhaircell'
    'adaptation'
    'ams_features'
    'crosscorrelation'
    'autocorrelation'
    'ratemap'
    'ild'
    'itd'
    'ic'
    'spectral_features'
    'onset_strength'
    'offset_strength'
    'pitch'
    'onset_map'
    'offset_map'
    'gabor'
```

A detailed description of the individual processors used to obtain these auditory representations will be given in chapter 4.

The implementation of the AFE is object-oriented, and two objects are needed to extract

any representation:

- A *data* object, in which the input signal, the requested representation, and also the dependent representations that were computed in the process are all stored.
- A *manager* object which takes care of creating the necessary processors as well as managing the processing.

In the following sections, examples of increasing complexity are given to demonstrate how to create these two objects, and which functionalities they offer.

2.3 Computation of an auditory representation

The following sections describe how the AFE framework can be used to compute an auditory representation with default parameters of a given input signal. We will start with a simple example, and gradually explain how the user can gain more control over the respective parameters. It is assumed that the entire input signal - for which the auditory representation should be computed - is available. Therefore, this operation is referred to as *batch processing*. As stated before, the framework is also compatible with *chunk-based processing* (i.e., when the input signal is acquired continuously over time, but the auditory representation is computed for smaller signal chunks). The chunk-based processing will be explained in Sec. 2.4.

2.3.1 Using default parameters

As an example, extracting the interaural level difference 'ild' for a stereo signal *sIn* (e.g., obtained from a '.wav' file through MATLAB's *wavread*) sampled at a frequency *fsHz* (in Hz) can be done in the following steps:

```
1 % Instantiation of data and manager objects
2 dataObj = dataObject(sIn,fsHz);
3 managerObj = manager(dataObj);
4
5 % Request the computation of ILDs
6 sOut = managerObj.addProcessor('ild');
7
8 % Request the processing
9 managerObj.processSignal;
```

Line 2 and 3 show the instantiation of the two fundamental objects: the data object and the manager. Note that the data object is always instantiated first, as the manager needs a data object instance as input argument to be constructed. The manager instance in line

3 is however an “empty” instance of the `manager` class, in the sense that it will not perform any processing. Hence a processing needs to be requested, as done in line 6. This particular example will request the computation of the inter-aural level difference ‘ild’. This step is configuring the manager instance `managerObj` to perform that type of processing, but the processing itself is performed at line 9 by calling the `processSignal` method of the manager class.

The request of an auditory representation via the `addProcessor` method of the manager class on line 6 returns as an output argument a handle to the requested signal, here named `sOut`. In the AFE framework, signals are also objects. For example, for the output signal just generated:

```
>> sOut
sOut =
    TimeFrequencySignal with properties:
        cfHz: [1x31 double]
        Label: 'Interaural level difference'
        Name: 'ild'
        Dimensions: 'nSamples x nFilters'
        FsHz: 100
        Channel: 'mono'
        Data: [267x31 circVBufArrayInterface]
```

This shows the various properties of the signal object `sOut`. These properties will be described in detail in chapter 3. To access the computed representation, e.g., for further processing, one can create a copy of the data contained in the signal into a variable, say `myILDs`:

```
>> myILDs = sOut.Data(:);
```

Note the use of the column operator `(:)`. That is because the property `.Data` of signal objects is not a conventional MATLAB array and one needs this syntax to access all the values it stores. The nature of the `.Data` property is further described in Sec. 3.2.1.

2.3.2 Input/output signals dimensionality

The input signal `sIn`, for which a given auditory representation needs be computed, is a simple array. Its first dimension (lines) should span time. Its first column should correspond to the left channel (or mono channel, if it is not a stereo signal) and the second column to

the right channel. This is typically the format returned by MATLAB's embedded functions `audioread` and `wavread`.

The input signal can be either mono or stereo/binaural. The framework can operate on both. However, some representations, such as the interaural level difference (ILD) as requested in the previous example, are based on a comparison between the left and the right ear signals. If a mono signal was provided instead of a binaural signal, the request of computing the ILD representation would produce the following warning and the request would not be computed:

```
Warning: Manager cannot instantiate a binaural cue extractor
         for a single-channel signal
> In manager>manager.addProcessor at 823
Warning: The request was invalid, returning an empty handle
> In manager>manager.addProcessor at 963
```

The dimensionality of the output signal from the `addProcessor` method will depend on the representation requested. In the previous example, the '`ild`' request returns a single output for a stereo input. However, when the request is based on a single channel and the input is stereo, the processing will be performed for left and right channel, and both left and right outputs are returned. In such cases, the output from the method `addProcessor` will be a cell array of dimensions `1 x 2` containing output signals for the left channel (first column) and right channel (second column). For example, the returned `sOut` could take the form:

```
>> sOut
sOut =
    [1x1 TimeFrequencySignal]    [1x1 TimeFrequencySignal]
```

The left-channel output can be accessed using `sOut{1}`, and similarly, `sOut{2}` for the right-channel output.

2.3.3 Change parameters used for computation

For the requested representation

Each individual processor that is supported by the AFE can be controlled by a set of parameters. Each parameter can be accessed by a unique *nametag* and has a default value. A summary of all parameter names and default values for the individual processors can be listed by the command `parameterHelper`:

```
>> parameterHelper
```

2 The auditory front-end framework

Parameter handling in the TWO!EARS auditory front-end (AFE)

The extraction of various auditory representations performed by the TWO!EARS Auditory Front-End software involves many parameters. Each parameter is given a unique name and a default value. When placing a request for TWO!EARS auditory front-end processing that uses one or more non-default parameters, a specific structure of non-default parameters needs to be provided as input. Such structure can be generated from [genParStruct](#), using pairs of parameter name and chosen value as inputs.

Parameters names for each processor are listed below:

[Adaptation loop](#)
[Amplitude modulation spectrogram features](#)
[Auditory filterbank](#)
[Auto-correlation](#)
[Cross-correlation](#)
[Gabor features](#)
[Inner hair-cell extraction](#)
[Interaural coherence](#)
[Interaural level difference](#)
[Interaural time difference](#)
[Offset strength extraction](#)
[Onset strength extraction](#)
[Pitch estimation](#)
[Plot properties](#)
[Pre-processing](#)
[Ratemap extraction](#)
[Spectral features](#)
[Transient mapping](#)

Each element in the list is a hyperlink, which will reveal the list of parameters for a given element, e.g.,

Interaural Level Difference parameters:

Name	Default	Description
---	-----	-----
ild_wname	'hann'	Window name
ild_wSizeSec	0.02	Window duration (s)
ild_hSizeSec	0.01	Window step size (s)

It can be seen that the ILD processor can be controlled by three parameters, namely `ild_wname`, `ild_wSizeSec` and `ild_hSizeSec`. A particular parameter can be changed by creating a parameter structure which contains the parameter name (*nametags*) and the corresponding value. The function `genParStruct` can be used to create such a parameter structure. For instance:

```
>> parameters = genParStruct('ild_wSizeSec',0.04,'ild_hSizeSec',0.02);

parameters =
    ild_wSizeSec: 0.0400
    ild_hSizeSec: 0.0200
```

will generate a suitable parameter structure `parameters` to request the computation of ILD with a window duration of 40 ms and a step size of 20 ms. This parameter structure is then passed as a second input argument in the `addProcessor` method of a manager object. The previous example can be rewritten considering the change in parameter values as follows:

```
1 % Instantiation of data and manager objects
2 dataObj = dataObject(sIn,fsHz);
3 managerObj = manager(dataObj);
4
5 % Non-default parameter values
6 parameters = genParStruct('ild_wSizeSec',0.04,'ild_hSizeSec',0.02);
7
8 % Place a request for the computation of ILDs
9 sOut = managerObj.addProcessor('ild',parameters);
10
11 % Perform processing
12 managerObj.processSignal;
```

For a dependency of the request

The previous example showed that the processor extracting ILDs was accepting three parameters. However, the representation it returns, the ILDs, will depend on more than these three parameters. For instance, it includes a certain number of frequency channels, but there is no parameter to control these in the ILD processor. That is because such parameters are from other processors that were used in intermediate steps to obtain the ILD. Controlling these parameters therefore requires knowledge of the individual steps in the processing.

Most auditory representations will depend on another representation, itself being derived from yet another one. Thus, there is a chain of *dependencies* between different representations, and multiple processing stages will be required to compute a particular output. The list of dependencies for a given request can be visualized using the function `getDependencies('nametag')`, e.g.,

```
>> getDependencies('ild')

ans =
```

```
'innerhaircell'      'filterbank'      'time'
```

shows that the ILD depends on the inner hair-cell representation ('`innerhaircell`'), which itself is obtained from the output of a gammatone filterbank ('`filterbank`'). The filterbank is derived from the time-domain signal, which itself has no further dependency as it is directly derived from the input signal.

When placing a request to the manager, the user can also request a change in parameters of any of the request's dependencies. For example, the number of frequency channels in the ILD representation is a property of the filterbank, controlled by the parameter '`fb_nChannels`'. (which name can be found using `parameterHelper.m`). This parameter can also be requested to have a non-default value, although it is not a parameter of the processor in charge of computing the ILD. This is done in the same way as previously shown:

```
5 % Non-default parameter values
6 parameters = genParStruct('fb_nChannels',16);
7
8 % Place a request for the computation of ILDs
9 sOut = managerObj.addProcessor('ild',parameters);
10
11 % Perform processing
12 managerObj.processSignal;
```

The resulting ILD representation stored in `sOut` will be based on 16 channels, instead of 31.

2.3.4 Compute multiple auditory representations

Place multiple requests

Multiple requests are supported in the framework, and can be carried out by consecutive calls to the `addProcessor` method of an instance of the manager with a single request argument. It is also possible to have a single call to the `addProcessor` method with a cell array of requests, e.g.:

```
8 % Place a request for the computation of ILDs AND autocorrelation
9 [sOut1 sOut2] = managerObj.addProcessor({'ild','autocorrelation'})
```

This way, the manager set up in the previous example will extract an ILD and an auto-correlation representation, and provide a handle to both signals in `sOut1` and `sOut2`.

To use non-default parameter values, three syntax are possible:

- If there are several requests, but all use the same set of parameter values `p`:

```
managerObj.addProcessor({'name1', ... , 'nameN'}, p)
```

- If there is only one request (`name`), but with different sets of parameter values (`p1,...,pN`), e.g., for investigating the influence of a given parameter:

```
managerObj.addProcessor('name', {p1, ... , pN})
```

- If there are several requests and some, or all, of them use a different set of parameter values, then it is necessary to have a set of parameter (`p1,...,pN`) for each request (possibly by duplicating the common ones) and place them in a cell array as follows:

```
managerObj.addProcessor({'name1', ... , 'nameN'}, {p1, ... , pN})
```

Note that in the two examples above, no output is specified for the `addProcessor` method, but the representations will be computed nonetheless. The output of `addProcessor` is there for convenience and the following subsection will explain how to get a hang on the computed signals without an explicit handle from `addProcessor`.

Requests can also be placed directly as optional arguments in the manager constructor, e.g., to reproduce the previous script example:

```
% Instantiation of data and manager objects
dataObj = dataObject(sIn, fsHz);
managerObj = manager(dataObj, {'ild', 'autocorrelation'});
```

The three possibilities described above can also be used in this syntax form.

Computing the signals

This is done in the exact same way as for a single request, by calling the `processSignal` method of the manager:

```
% Perform processing
managerObj.processSignal;
```

Access internal signals

The optional output of the `addProcessor` method is provided for convenience. It is actually a pointer (or handle, in MATLABS terms) to the actual signal object which is hosted by

the data object on which the manager is based. Once the processing is carried out, the properties of the data object can be inspected:

```
>> dataObj

dataObj =
    dataObject with properties:

    bufferSize_s: 10
    isStereo: 1
    gammatone: {[1x1 TimeFrequencySignal] [1x1 TimeFrequencySignal]}
        ratemap: {[1x1 TimeFrequencySignal] [1x1 TimeFrequencySignal]}
    autocorrelation: {[1x1 CorrelationSignal] [1x1 CorrelationSignal]}
        ild: {[1x1 TimeFrequencySignal]}
    input: {[1x1 TimeDomainSignal] [1x1 TimeDomainSignal]}
        time: {[1x1 TimeDomainSignal] [1x1 TimeDomainSignal]}
    innerhaircell: {[1x1 TimeFrequencySignal] [1x1 TimeFrequencySignal]}
```

Apart from the properties `bufferSize_s` and `isStereo` which are inherent properties of the data object (and discussed later in chapter 3), the remaining properties each correspond to one of the representations computed to achieve the user's request(s). They are each arranged in cell arrays, with first column being the left, or mono channel, and the second column the right channel. For instance, to get a handle `sGammaR` to the right channel of the gammatone filterbank output, type:

```
>> sGammaR = dataObj.gammatone{2}

sGammaR =
    TimeFrequencySignal with properties:

    cfHz: [1x31 double]
    scaling: 'magnitude'
    Label: 'Gammatone filterbank output'
    Name: 'gammatone'
    Dimensions: 'nSamples x nFilters'
    FSHz: 44100
    Channel: 'right'
    Data: [118299x31 circVBufArrayInterface]
```

2.3.5 How to plot the result

Plotting auditory representations is made very easy in the AFE framework. As explained before, each representation that was computed during a session is stored as a signal object, which each are individual properties of the data object. Signal objects of each type have a

`plot` method. Called without any input arguments, `signal.plot` will adequately plot the representation stored in `signal` in a new figure, and returns as output a handle to said figure. The plotting method for all signals can accept at least one optional argument, which is a handle to an already existing figure or subplot in a figure. This way the representation can be included in an existing plot. A second optional argument is a structure of non-default plot parameters. The `parameterHelper` script also lists plotting options, and they can be modified in the same way as processor parameters, via the script `genParStruct`. These concepts can be summed up in the following example lines, that follows right after the demo code from the previous subsection:

```

11 % Request the processing
12 managerObj.processSignal;
13
14 % Plot the ILDs in a separate figure
15 sOut.plot;
16
17 % Create an empty figure with subplots
18 figure;
19 h1 = subplot(2,2,1);
20 h2 = subplot(2,2,2);
21 h3 = subplot(2,2,3);
22 h4 = subplot(2,2,4);
23
24 % Change plotting options to remove colorbar and reduce title size
25 p = genParStruct('bColorbar',0,'fsize_title',12);
26
27 % Plot additional representations
28 dataObj.innerhaircell{1}.plot(h1,p);
29 dataObj.innerhaircell{2}.plot(h2,p);
30 dataObj.gammatone{1}.plot(h3,p);
31 dataObj.gammatone{2}.plot(h4,p);

```

This script will produce the two figure windows displayed in Fig. 2.1. Line 22 of the script creates the window “Figure 1”, while lines 35 to 38 populate the window “Figure 2” which was created earlier (in lines 25 to 29).

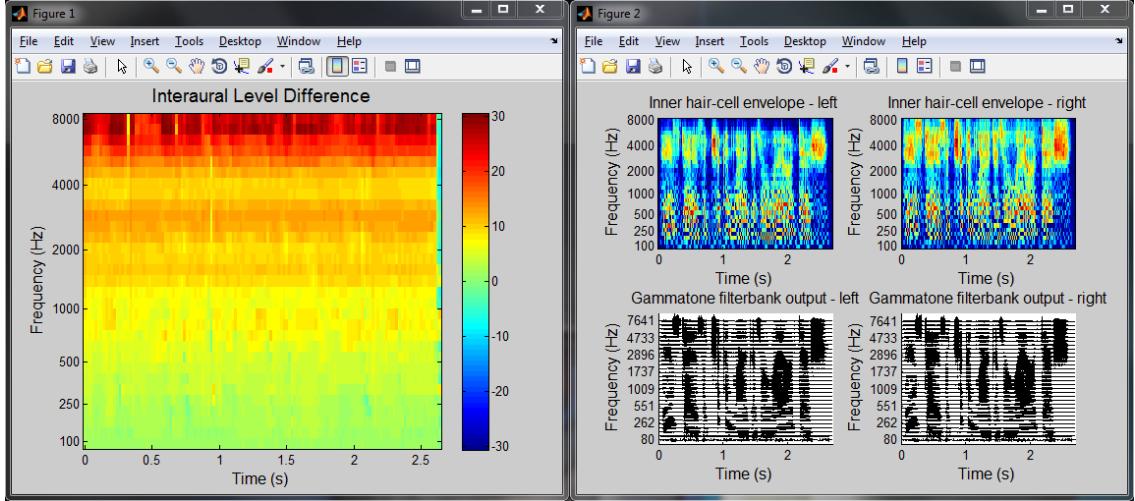


Figure 2.1: The two example figures generated by the demo script.

2.4 Chunk-based processing

As mentioned in the previous section, the framework is designed to be compatible with chunk-based processing. As opposed to “batch processing”, where the entire input signal is known *a priori*, this means working with consecutive chunks of input signals of arbitrary size. In practice the chunk size will often be the same from one chunk to another. However, this is not a requirement here, and the framework can accept input chunks of varying size.

The main constraint behind working with an input that is segmented into chunks is that the returned output should be exactly the same as if the whole input signal (i.e., the concatenated chunks) was used as input. In other terms, the transition from one chunk to the next needs to be taken into account in the processing. For example, concatenating the outputs obtained from a simple filter applied *separately* to two consecutive chunks will not provide the same output as if the concatenated chunks were used as input. To obtain the same output, one should for example use methods such as overlap-add or overlap-save. This is not trivial, particularly in the context of the AFE where more complex operations than simple filtering are involved. A general description of the method used to ensure chunk-based processing is given in Sec. 3.3.2.

Handling segmented input in practice is done mostly the same way as for a whole input signal. The available demo script `DEMO_ChunkBased.m` provides an example of chunk-based processing by simulating a chunk-based acquisition of the input signal with variable chunk size and computing the corresponding ILDs.

In this script, one can note the two differences in using the AFE in a chunk-based scenario, in comparison to a batch scenario:

```
18 % Instantiation of data and manager objects
19 dataObj = dataObject([],fsHz,10,2);
20 managerObj = manager(dataObj);
```

Because the signal is not known before the processing is carried out, the data object cannot be initialized from the input signal. Hence, as is seen on line 19, one needs to instantiate an empty data object, by leaving the first input argument blank. The sampling frequency is still necessary however. The third argument (here set to 10) is a global signal buffer size in seconds. Because in an online scenario, the framework could be operating over a long period of time, internal representations cannot be stored over the whole duration and are instead kept for the duration mentioned there. The last argument (2) indicates the number of channel that the framework should expect from the input (a mono input would have been indicated by 1). Again, it is necessary to know the number of channels in the input signal, to instantiate the necessary objects in the data object and the manager.

```
40 % Request the processing of the chunk
41 managerObj.processChunk(sIn(chunkStart:chunkStop,:),1);
```

The processing is carried out on line 40 by calling the `processChunk` method of the manager. This method takes as input argument the new chunk of input signal. The additional argument, 1, indicates that the results should be appended to the internal representations already computed. This can be set to 0 in cases where keeping track of the output for the previous chunks is unnecessary, for instance if the output of the current chunk is used by a higher-level function. The difference with the `processSignal` method is important. Although `processSignal` actually calls internally `processChunk`, it also resets internal states of the framework (what ensures continuity between chunks) before processing.

The script `DEMO_ChunkBased.m` will also compute the offline result and will plot the difference in output for the two computations. This plot is shown in Fig. 2.2. Note the magnitude on the order of 10^{-15} , which is in the range of MATLAB numerical precision, suggesting that the representations computed online or offline are the same up to some round-off errors.

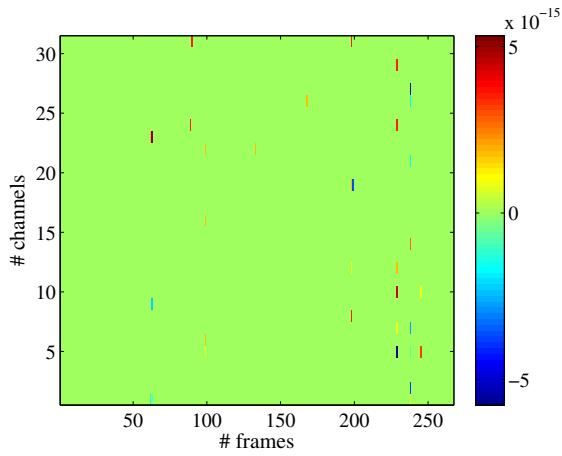


Figure 2.2: Difference in ILD obtained in online and offline processing

2.5 List of commands

This section sums up the commands that could be relevant to a standard user of the AFE framework. It does not describe each action extensively, nor does it give a full list of corresponding parameters. A more detailed description can be obtained through calling the help script of a given method from MATLAB's command window. Note that one can get help on a specific method of a given class. For example

```
>> help manager.processChunk
```

will return help related to the `processChunk` method of the `manager` class. The following aims at being concise, hence optional inputs are marked as “...” and can be reviewed from the specific method help.

Signal objects `sObj`

<code>sObj.Data(:)</code>	Returns all the data in the signal
<code>sObj.Data(n1:n2)</code>	Returns the data in the time interval [n1,n2] (samples)
<code>sObj.findProcessor(mObj)</code>	Finds processor that computed the signal
<code>sObj.getParameters(mObj)</code>	Parameter summary for that signal
<code>sObj.getSignalBlock(T,...)</code>	Returns last T seconds of the signal
<code>sObj.play</code>	Plays back the signal (time-domain signals only)
<code>sObj.plot(...)</code>	Plots the signal

Data objects dObj

<code>dObj = dataObject(s,fs,bufSize,nChannels)</code>	Constructor
<code>dObj.addSignal(sObj)</code>	Adds a signal object
<code>dObj.clearData</code>	Clears all signals in <code>dObj</code>
<code>dObj.getParameterSummary(mObj)</code>	Lists parameter used for each signal
<code>dObj.play</code>	Plays back the containing audio signal

Processors pObj

<code>pObj.Dependencies</code>	Handle to the dependent processor of <code>pObj</code>
<code>pObj.getCurrentParameters</code>	Parameter summary for that processor
<code>pObj.getDependentParameter(parName)</code>	Value of a parameter from <code>pObj</code> or its dependencies
<code>pObj.hasParameters(parStruct)</code>	True if <code>pObj</code> used the exact values in <code>parStruct</code>
<code>pObj.Input</code>	Handle to input signal object
<code>pObj.Output</code>	Handle to output signal object

Manager mObj

<code>mObj = manager(dObj)</code>	Constructor
<code>mObj = manager(dObj, name, param)</code>	Constructor with initial request
<code>mObj.addProcessor(name, param)</code>	Adds a processor (including eventual dependencies)
<code>mObj.Data</code>	Handle to the associated data object
<code>mObj.processChunk(input, ...)</code>	Process a new chunk
<code>mObj.Processors</code>	Lists instantiated processors
<code>mObj.processSignal</code>	Process a signal offline
<code>mObj.reset</code>	Resets all processors

2.6 Acknowledgment

The AFE framework includes the following contributions from publicly available MATLAB toolboxes or classes:

- AMToolbox
- LTFAT
- Voicebox

2 The auditory front-end framework

- circVBuf

3 Technical description

3.1 Overview

Many different auditory models are available that can transform an input signal into an auditory representation. The actual design challenges behind the AFE framework arise from the multiplicity of supported representations, the requirement to process continuous signal in a chunk-based manner, and the ability to change what is being computed at run-time, which will allow the incorporation of *feedback* from higher processing stages. In addition to these three constraints, the framework will be subject to frequent updates in the future of the Two!EARS project (e.g., adding new processors), so the expandability and maintainability of its implementation should be optimal. For these reasons, the framework is implemented using a modular object-oriented approach.

This chapter exposes the architecture and interactions of all the objects involved in the AFE and how the main constraints were tackled conceptually. In an effort to respect encapsulation and the hierarchical organization of the objects, the sections are arranged in a “bottom-up” way: from the most fundamental objects to the more global processing.

All classes involved in the AFE implementation are inheriting the MATLAB `handle` master class. This allows every created object to be of the `handle` type, and simulates a “call-by-reference” when manipulating the objects. Given an object `obj` inheriting the `handle` class, doing `obj2 = obj` will not copy the object, but only obtain a pointer to it. If `obj` is modified, then so is `obj2`. This avoids unnecessary copies of objects, limiting memory use, as well as providing user friendly handles to objects included under many levels of class hierarchy. The user can manipulate a simple short-named handle instead of tediously accessing the object.

3.2 Data handling

3.2.1 Circular buffer

Memory pre-allocation of large arrays in MATLAB is well known to be a critical operation for optimizing computation time. The AFE, particularly in an online scenario, will be confronted with this problem. For each new chunk of the input signal, chunks of output are computed for each internal representation and are appended to the already existing output. Computation time will be strongly affected if the arrays containing the data are not initialized appropriately (i.e., the memory it occupies is pre-allocated) to fit the input signal duration.

The issue in a real-time scenario is that the signal duration is unknown. To overcome this problem, data for each signal is stored in a buffer of fixed duration which is itself pre-allocated. Buffers are updated following a first in, first out (FIFO) rule: once the buffer is full, the oldest samples in the buffer are overwritten by the new signal samples.

The `circVBuf` class

A conceptual way of implementing a FIFO rule is to use circular (or ring) buffers. The inconvenience of a traditional, linear buffer is that once it is full and new input overwrites old samples (i.e., it is in its “steady-state”), reading the data from it implies reaching the end of the buffer and continuing reading from its beginning. The data read will be in two fragments, because of the linear buffer having a physical beginning and end which do not match to the oldest and newest data samples. This is eliminated in circular buffers which do not have a beginning or end, and a contiguous segment is always obtained upon reading. Circular buffers were implemented for the AFE framework based on the third-party class provided by Göbbert (2014), which has been slightly modified to account for multi-dimensional data (instead of vector-only).

Circular buffer interface

The `circVBuf` class provides a buffer that is conceptually circular, in the sense that it allows continuous reading of the data. However in practice it still stores data in a linear array in MATLAB (the size of which is, however, twice the size of the actual data). Accessing stored data requires knowledge about this class and can be tedious to a naive user. To eliminate confusion and make the buffer transparent to the user, the interface `circVBuffArrayInterface` was implemented, with the aim of allowing the buffer to use

most basic array operations.

Given a circular buffer `circBuffer`, the interface is obtained by

```
buffer = circVBufArrayInterface(circBuffer)
```

It will allow the following operations:

- `buffer(n1:n2)` returns stored data between positions `n1` and `n2`, where position 1 is the oldest sample in the buffer (but not necessarily the first one in the actual array storing data, due to circularity). For multiple dimensions, these indices always refer to the first dimension. To return stored data up to the most recent sample, use `buffer(n1:end)`.
- `buffer(:)` returns all data stored in the buffer (ignoring “empty” sections of the buffer, if said buffer was never filled).
- `buffer('new')` returns the latest chunk of data that was added to the buffer.
- `length(buffer)` returns the effective (i.e., ignoring empty sections) buffer length across its first dimension.
- `size(buffer)` returns the effective size of the buffer (including other dimensions).
- `numel(buffer)` returns the total number of elements stored (calculated as product of the effective dimensions).
- `isempty(buffer)` returns `true` when no data is stored, `false` otherwise.

This provides an array behavior to the buffers, simplifying greatly their use. Note that the only limitation is the need of the column operator `:` to access all data, as in `buffer(:)`. Without it, `buffer` will return a handle to the `circVBufArrayInterface` object.

3.2.2 Signal objects

Signals are implemented as objects in the AFE. To avoid code repetition and make better use of object-oriented concepts, signals are grouped according to their dimensionality, as they then share the same properties. The following classes are implemented:

- `TimeDomainSignal` for one-dimensional (time) signals.
- `TimeFrequencySignal` which stores two-dimensional signals where the first dimension

relates to time (but can be, e.g., a frame index) and the second to the frequency channel. These signals include as an additional property a vector of channel center frequencies `cfHz`. Signals of such form are obtained from requesting, for example, `'filterbank'`, `'innerhaircell'`, `'ild'`, ...

- `CorrelationSignal` for three-dimensional signals where the third dimension is a lag position. These include also the `cfHz` property as well as a vector of lags (`lags`).
- `ModulationSignal` for three-dimensional signals where the third dimension is a modulation frequency. These include `cfHz` and `modCfHz` (vector of center modulation frequencies) as properties.
- `FeatureSignal` used to store a collection of time-domain signals, each associated to a specific name. Each feature is a single vector, and all of them are arranged as columns of a same matrix. Hence they include an ordered list of features names `fList` that labels each column.

All these classes inherit the parent `Signal` class. Hence they all share the following common “read-only” properties:

- `Label`, which is a “formal” description of the signal, e.g., `'Inner hair-cell envelope'`, used for example when plotting the signal.
- `Name`, which is a nametag unique to each signal type, e.g., `'innerhaircell'`. This name corresponds to the name used for a request to the manager.
- `Dimensions`, which describes in a short string how dimensions are arranged in the signal, e.g., `'nSamples x nFilters'`
- `FsHz`, the sampling frequency of this specific signal. If the signal is framed or downsampled (e.g., like a ratemap or an ILD) this value will be different from the input signal’s sampling frequency.
- `Channel`, which states `'left'`, `'right'` or `'mono'`, depending on which channel from the input signal this signal was derived.
- `Data`, an interface object (`circVBufArrayInterface` described earlier) to the circular buffer containing all data. The actual buffer, `Buf` is a `circVBuf` object and a protected property of the signal (not visible to the user).

The `Signal` class defines the following methods that are then shared among children objects:

- A super constructor, which sets up the internal buffer according to the signal dimensions. Each children signal class is calling this super constructor before populating

its other properties.

- An `appendChunk` method used to fill the internal buffer.
- A `setData` method used for initializing the internal buffer given some data.
- A `clearData` method for re-initialization.
- The `getSignalBlock` method returning a segment of data of chosen duration, starting from the newest elements.
- The `findProcessor` method which, given a handle to a manager object, will retrieve which processor has computed this specific signal (by comparing it with the `Output` property of each processor, described in Sec. 3.3.1).
- A `getParameters` method which, given a handle to a manager object, will retrieve the list of parameters used in the processing to obtain that signal.

In addition, the `Signal` class defines an abstract `plot` method, which each children should implement. This cannot be defined in the parent class as the plotting routines will be drastically different depending on children signal dimensionality. Children classes therefore only implement their own constructor (which still calls the super-constructor) and their respective plotting routines.

3.2.3 Data objects

Description

Many signal objects are instantiated by the AFE (one per representation involved and per channel). To organize and keep track of them, they are collected in a `dataObject` class. This class inherits the `dynamicprops` MATLAB class (itself inheriting the `handle`) class. This allows to dynamically define properties of the class.

This way, each signal involved in a given session of the AFE framework will be grouped according to its class in a distinct property of the `dataObject`, with name given by the signal `signal.Name` unique nametag. Extra properties of the data object include:

- `bufferSize_s` which is the common duration of all `circVBuf` objects in the signals.
- A flag `isStereo`, which if true will indicate to the data object that all signals come as pairs of left/right channels.

Data objects are constructed by providing an input signal (which can be empty in online scenarios), a mandatory sampling frequency in Hz, a global buffer size (10s by default),

3 Technical description

and the number of channels of the input (1 or 2). This number of channel is not necessary if an input signal is used as argument in the constructor but needs to be provided otherwise.

The `dataObject` definition includes the following, self-explanatory methods:

- `addSignal(signalToAdd)`
- `clearData`
- `getParameterSummary` returning a list of all parameters used for the computation of all included signal (given a handle to the corresponding manager).
- `play`, provided for user convenience.

Signal organization

As mentioned before, data objects store signal objects. Each class of signal occupies a property in the data object named after the signal `.Name` property. Multiple signals of the same class will be stored as a cell array in that property. In the cell array, the first column is always for the left channel (or mono signal), and the second column for the right channel. If multiple signals of the same type are present (e.g., if the user requested the same representation twice but with a change of parameters), then the corresponding signals are stored in different lines of the array. For instance, for a session where the user requested the inner hair-cell envelope twice, with the second request changing only the way of extracting the envelope (i.e., the parameter '`ihc_method`'), the following data object is created:

```
>> dataObj
dataObj =
    dataObject with properties:
        bufferSize_s: 10
        isStereo: 1
        time: {[1x1 TimeDomainSignal] [1x1 TimeDomainSignal]}
        input: {[1x1 TimeDomainSignal] [1x1 TimeDomainSignal]}
        gammatone: {[1x1 TimeFrequencySignal] [1x1 TimeFrequencySignal]}
        innerhaircell: {2x2 cell}
```

Each signal-related field except `innerhaircell` is a cell array of a single line (one signal), and two columns (for left and right channel). Because the second request from the user included only a change in parameter for the inner hair-cell computation, the same initial `gammatone` signal is used for both, but there are two output `innerhaircell` signals (hence

a cell array of two lines) for each channel (hence two columns).

In that case, to distinguish between the two signals and know which one was computed with which set of parameter, one can call the signal's `getParameters` method. Given a handle to the manager object, it will return a list of all parameters used to obtain that signal (including parameters used in intermediate processing steps).

3.3 Processors

Processors are at the core of the AFE. Each processor is responsible for an individual step in the processing, i.e., going from representation A to representation B. They are adapted from existing models documented in the literature such as to allow for block-based (online) processing. This is made possible by keeping track of the information necessary to transition adequately between two chunks of input. The nature of this “information” varies depending on the processor, and we use in the following the term “internal state” of the processor to refer to it. Internal states and online processing compatibility are then assessed in Sec. 3.3.2.

A detailed overview of all processors, with a list of all parameters they accept, is given in chapter 4. Hence this section will focus on the properties and methods shared among every processors, as well as the techniques employed to make processing compatible with chunk-based inputs.

3.3.1 General considerations

As for signal objects, processors make use of inheritance, with a parent `Processor` class. The parent class defines shared properties of the processor, abstract classes that each children must implement, and a couple of methods shared among children. Each processor shares the properties:

- `Type` - describes formally the processing performed
- `Input` - handle to input signal object
- `Output` - handle to output signal object
- `isBinaural` - Flag indicating the need of left and right channel as input
- `FsHzIn` - Input signal sampling frequency
- `FsHzOut` - Output signal sampling frequency

- **Dependencies** - Handle to the processor that generated the input signal

These properties are populated automatically when using the AFE framework by the manager class which is described later in Sec. 3.4.

The parent **Processor** class defines the following abstract methods. Because these methods are children dependent, each processor sub-class **pObj** should then implement them:

- **out = pObj.processChunk(in)**, the core processing method. Returns an output **out** given the input **in**. It will, if necessary, use the internal states of the processor (derived from previous chunk(s) of input) to calculate the output. These internal states should be accordingly updated in this method after the processing was performed. Next sub-section provides more details regarding these internal states.
- **pObj.reset**, that clears the internal states of the processor. To be used e.g., in an offline scenario in between two different input signals.
- **pObj.hasParameters(parStruct)**, which should return true if and only if the processor has the parameter values described in the structure **parStruct**. This structure is of the same form as the structure used for requests and can be generated using the script **genParStruct**.

Two methods are then identical across all processors and are therefore implemented in the parent **Processor** class:

- **value = pObj.getDependentParameter(parName)** recursively recovers the value of a specific parameter used by **pObj** or by one of its dependencies
- **parStruct = pObj.getCurrentParameters** returns a structure of the parameter values currently used by the processor.

The motivation behind the implementation of these methods is probably not clear at this stage, but should appear in the following sections. Many of these methods are used in the manager object described later for organizing and routing the processing such as to always perform as few operations as needed.

3.3.2 processChunk method and chunk-based compatibility

General approach

As briefly exposed above, exact computation performed by each processors are taken from published models, and are described individually in chapter 4. However, most of the

available implementations are for batch processing, i.e., using one whole input signal at once. To be included in the AFE, these implementations need to be adapted to account for chunk-based processing, i.e., when the input signal is fed to the system in non-overlapping contiguous blocks, or chunks.

Some processors rely on the input only at time t to generate the output at time t . These processors are then compatible as such with chunk-based processing. This is the case for instance for the `itdProc` which given cross-correlation deduces the interaural time differences (ITDs). That is because the processor, at time t , is provided a cross-correlation value as input (which is a function of frequency and lag), and only locates for each frequency the lag value for which the cross-correlation is maximal. There is no influence of past (or future) inputs to provide the output at time t . This is unfortunately not the case for most processors, which output at a given time will be influenced, to different extent, by older input. However, so far, all the processing involved in the AFE is causal, i.e., might depend on past input, but will not depend on future input.

Adapting offline implementations to online is of course case-dependent, and how it was done for each individual processors will not be described here. However the same concept is used for each, and can be related to the *overlap-save* method traditionally used for filtering long signals (or a stream of input signal) with a finite impulse response (FIR) filter. This concept revolves around using an internal buffer to store the input samples of a given chunk that will influence the processing of the next chunk. Because of the causality, these samples will always be at the end of the present chunk. Considering a processor which is in “steady-state” (i.e., has a populated internal buffer) and a new incoming chunk of input signal, the following steps are performed:

1. The buffer is appended in the beginning of the new input chunk. Conceptually, this provides also a chunk of the input signal, but a longer one that starts at an earlier point in time.
2. The input extended in this way is processed following the computations described in literature. If the input is required to have specific dimensions in time (e.g., when windowing is performed), then it is virtually truncated to these dimensions (i.e., input samples falling outside the required dimensions are discarded). The goal is for the output to be as long as possible while still being “valid”, i.e., not being influenced by the boundary with the next input chunks. If additional output was generated due to the appended buffer, it is discarded.
3. The buffer is updated to prepare for the next input chunk. This step can vary between processors but the idea is to store in the buffer the end of the current chunk which did not generate output, or which will influence the output of next chunk.

An example: ratemap

A practical example to better illustrate the concepts described above is given in the following. The ratemap is conceptually a “framed” version of an inner hair-cell (IHC) multichannel envelope. The IHC envelope is a two-dimensional representation (time versus frequency), and the ratemap extraction is the same procedure repeated for every frequency channel. Hence the following is described for a single channel. To extract the ratemap, the envelope is windowed by a set of overlapping windows, and its magnitude averaged in each window. This process is adapted to online processing as illustrated in Fig. 3.1

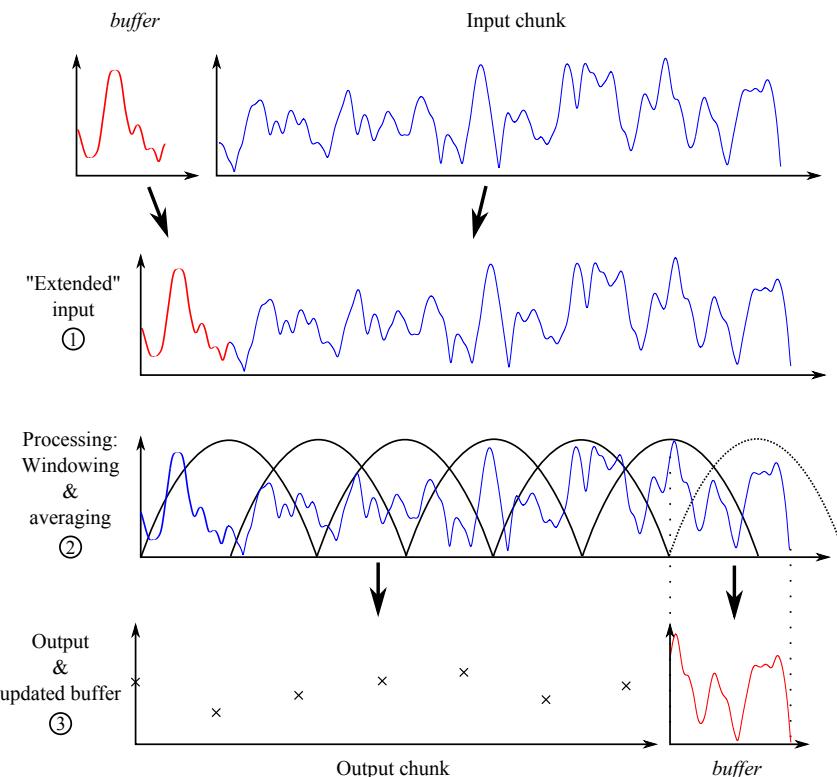


Figure 3.1: Three steps for simple online windowing, given a chunk of input and an internal buffer.

The three above-mentioned steps are followed:

1. The internal buffer (which can be empty, e.g., if first chunk) is appended to the input chunk.
2. This “extended” input is then processed. In that case, it is windowed and the average is taken in each window.

3. The “valid” outputs form the output chunk. Note that the right-most window (dashed line) is not fully covering the signal. Hence the output it would provide is not “valid”, since it would also partly depend on the content of the next input chunk. Therefore the section of the signal corresponding to this incomplete window forms the new buffer.

Note that the output chunk could in theory be empty. If the duration of the “extended” input in step 1 is shorter than the duration of the window, then no valid output is produced for this chunk, and the whole extended input will be transferred to the internal buffer. This is unlikely to happen in practice however.

Particular case for filters

The processing performed by the AFE often involves filtering (e.g., in auditory filterbank processing, inner haircell envelope detection, or amplitude modulation detection). While filtering by FIR filters could in principle be made compatible with chunk-based processing using the principle described above, it will be impractical for filters with long impulse response, and in theory impossible for infinite impulse response (IIR) filters.

For this reason, chunk-based compatibility is managed differently for filtering. In MATLAB’s `filter` function, the user can specify initial conditions and can get as optional output the final conditions of the filter delays. These take the form of a vector, of dimension equal to the filter order.

In the AFE, filters are implemented as objects, and encapsulate a private `states` property. This property simply contains the final conditions of the filter delays, i.e., its internal states after the last processing it performed. If applied to a new input chunk, these states are used as initial condition and are updated after the processing. This will provide a continuous output given a fragmented input.

3.4 Manager

The `manager` class is fundamental in the AFE framework. It is responsible for, from a user request, instantiating the correct processors and signal objects, and linking these signals as inputs/outputs of each processor. In a standard session of the AFE, only a single instance of this class is created. It is with this object that the user interacts.

3.4.1 Processors and signals instantiation

Single request

A standard call to the manager constructor, i.e., with no other argument than a handle to an already created data object `dataObj` will produce an “empty” manager:

```
>> mObj = manager(dataObj)

mObj =
manager with properties:

Processors: []
InputList: []
OutputList: []
Map: []
Data: [1x1 dataObject]
```

Empty properties include a list of processors, of input signals, output signals, and a mapping vector that provides a processing order. The `Data` property is simply a handle to the `dataObj` object provided for convenience.

Populating these properties is made via the `addProcessor` method already described in Sec. 2.3. From a given request and an empty manager, instantiating the adequate processors and signals is done following these steps:

1. Get the list of signals needed to compute the user request, using the `getDependencies` function.
2. Flip this list around such as to have the list starting with '`time`', and ending up with the requested signal. The list then provides the needed signals in the order they should be computed.
3. Loop over the elements of the list. For each signal on the list:
 - a) Instantiate a corresponding processor (two if stereo signal)
 - b) Instantiate the signal that will contain the output of the processor (two if stereo)
 - c) Add the signal(s) to `dataObj`
 - d) A handle to the output signal of the previous processor on the list is stored as the current processor’s input (in `mObj.InputList` as well as in the processor’s `Input` property). If it is the first element of the list, this will link to the original time domain signal.

- e) A handle to the newly instantiated signal is stored similarly as output. This handle is stored further for the next element in the loop.
 - f) A handle to the previously instantiated processor is stored in the current processor's **Dependencies** property (possibly empty if first element of the list).
4. Generate a linear mapping (vector of indexes of the processors ordered in increasing processing order).
 5. Return a handle to the requested signal to the user.

Once **addProcessor** called, the properties of the manager will have been populated, e.g.:

```
>> mObj
mObj =
    manager with properties:

    Processors: {3x2 cell}
    InputList: {3x2 cell}
    OutputList: {3x2 cell}
        Map: [1 2 3]
        Data: [1x1 dataObject]
```

Processors are arranged with the same convention as for signals in a data objects: they are stored in a cell array, where the first column is for left (or mono) channel, and second column for right channel. Different lines are for different processors, e.g.:

```
>> mObj.Processors
ans =
    [1x1 preProc      ]      [1x1 preProc      ]
    [1x1 gammatoneProc]      [1x1 gammatoneProc]
    [1x1 ihcProc       ]      [1x1 ihcProc       ]
```

InputList and **OutputList** are cell arrays of handles to signal objects. An element in one of them will correspond to the input/output of the processor at the same position in the cell array.

Handling of multiple requests

The above-described process gets more complicated when a request is placed in a non-empty manager (i.e., when multiple requests have been placed). The same steps could be used,

and would result in a functioning result. However, this would likely be sub-optimal in terms of computations. If the new request has common elements with representations that are already computed, one need not recompute them.

If correctly implemented, a manager should be able to “branch” the processing, such that only new representations, or representations where a parameter has been changed, are recomputed. Achieving this relies on the `findInitProc` method of the manager, which is described in more details in the next subsection. This method is passed the same arguments as the `addProcessor` method, i.e., a request name and a structure of parameters. It will return a handle to an already existing processor in the manager that is exactly computing one of the steps needed for that request. It will return the “highest” already existing step. In other terms, it finds the point in the already existing ordered list of processors where the processing should “branch out” to obtain the newly requested feature. Knowing the processor to start from and updating accordingly the list of signals/processors that need to be instantiated, the same procedure as before can then be used in the `addProcessor` method.

The `findInitProc` method

To find an initial processor suitable in a request, this method calls the `hasProcessor` method of the manager and the `hasParameters` method of each processor. From a given request, it can obtain a list of necessary processing steps from `getDependencies` and run the list backwards. For each element of the list, `findInitProc` “asks” the manager if it has such a processor via its `hasProcessor` method. If yes, it calls this processor `hasParameters` method to verify that what the processor computes corresponds to the request. If yes, then it found a suitable initial step. If no, it moves on to the next element in the list and repeats.

3.4.2 Carrying out the processing

As of the current AFE implementation, the processing is linear and the `processChunk` methods of each individual processor are called one after the other when asking the manager to start processing (either via its `processChunk` or `processSignal` method). The order in which the processors are called is important, as some will take as input what was other’s output. This order is stored in the property `Map` of the manager. `Map` is a vector of indexes corresponding to the lines in the `Processors` cell array property of the manager. It is constructed at instantiation of the processors. Conceptually, if there are N instantiated processors, the `processChunk` method of the manager `mObj` will call the `processChunk` methods of each processor following this loop:

```
1 | for ii = 1 : N
```

```
2 % Get index of current processor
3 jj = mObj.Map(ii);
4
5 % Getting input signal handle (for code readability)
6 in = mObj.InputList{jj,1};
7
8 % Perform the processing
9 out = mObj.Processors{jj,1}.processChunk(in.Data('new'));
10
11 % Store the result
12 mObj.OutputList{jj,1}.appendChunk(out);
13 end
```

Note the difference between indexes `ii` which relate to the processing order (processing first `ii=1` and last `ii=N`) and `jj = mObj.Map(ii)` which relate the processing order with the actual position of the processors in the cell array `mObj.Processors`.

4 Available processors

This chapter presents a detailed description of all processors that are currently supported by the AFE framework. Each processor can be controlled by a set of parameters, which will be explained and all default settings will be listed. Finally, a demonstration will be given, showing the functionality of each processor. The corresponding MATLAB files are contained in the AFE folder `/test` and can be used to reproduce the individual plots. A summary of all demo scripts can be found in appendix A. A full list of available processors can be displayed by using the command `requestList`. An overview of the commands for instantiating processors is given in Sec. 2.3.

4.1 Pre-processing (`preProc.m`)

Prior to computing any of the supported auditory representations, the input signal stored in the data object can be pre-processed with one of the following elements:

1. Direct current (DC) bias removal
2. Pre-emphasis
3. Root mean square (RMS) normalization using an automatic gain control (AGC)
4. Level scaling to a pre-defined sound pressure level (SPL) reference
5. Middle ear filtering

The order of processing is fixed. However, individual stages can be activated or deactivated, depending on the requirement of the user. The output is a time domain signal representation that is used as input to the next processors. Moreover, a list of adjustable parameters is listed in Tab. 4.1.

The influence of each individual pre-processing stage except for the level scaling is illustrated in Fig. 4.1, which can be reproduced by running the script `DEMO_PreProcessing.m`. Panel 1 shows the left and the right ears signals of two sentences at two different levels. The ear signals are then mixed with a sinusoid at 0.5 Hz to simulate an interfering humming noise. This humming can be effectively removed by the DC removal filter, as shown in

Table 4.1: List of parameters related to the auditory representation ‘time’.

Parameter	Default	Description
pp_bRemoveDC	false	Activate DC removal filter
pp_cutoffHzDC	20	Cut-off frequency in Hz of the high-pass filter
pp_bPreEmphasis	false	Activate pre-emphasis filter
pp_coefPreEmphasis	0.97	Coefficient of first-order high-pass filter
pp_bNormalizeRMS	false	Activate RMS normalization
pp_intTimeSecRMS	2	Time constant in s used for RMS estimation
pp_bBinauralRMS	true	Link RMS normalization across both ear signals
pp_bLevelScaling	false	Apply level scaling to the given reference
pp_refSPLdB	100	Reference dB SPL to correspond to the input RMS of 1
pp_bMiddleEarFiltering	false	Apply middle ear filtering
pp_middleEarModel	‘jepsen’	Middle ear filter model

panel 3. Panel 4 shows the influence of the pre-emphasis stage. The AGC can be used to equalize the long-term RMS level difference between the two sentences. However, if the level difference between both ear signals should be preserved, it is important to synchronize the AGC across both channels, as illustrated in panel 5 and 6. Panel 7 shows the influence of the level scaling when using a reference value of 100 dB SPL. Panel 8 shows the signals after middle ear filtering, as the stapes motion velocity. Each individual pre-processing stage is described in the following subsections.

4.1.1 DC removal filter

To remove low-frequency humming, a DC removal filter can be activated by using the flag `pp_bRemoveDC = true`. The DC removal filter is based on a fourth-order IIR butterworth filter with a cut-off frequency of 20 Hz, as specified by the parameter `pp_cutoffHzDC = 20`.

4.1.2 Pre-emphasis

A common pre-processing stage in the context of automatic speech recognition (ASR) includes a signal whitening. The goal of this pre-processing stage is to roughly compensate for the decreased energy at higher frequencies (e.g. due to lip radiation). Therefore, a first-order FIR high-pass filter is employed, where the filter coefficient `pp_coefPreEmphasis` determines the amount of pre-emphasis and is typically selected from the range between 0.9

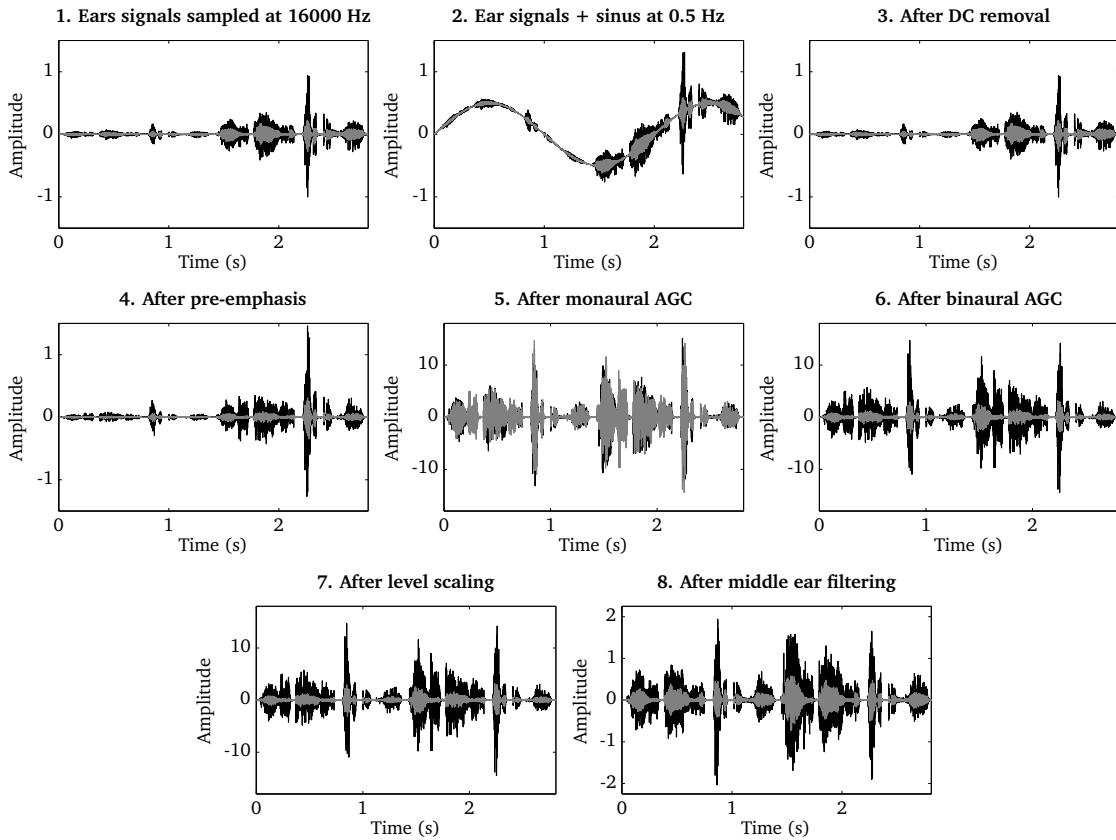


Figure 4.1: Illustration of the individual pre-processing steps. 1) Ear signals consisting of two sentences recorded at different levels, 2) ear signals mixed with a 0.5 Hz humming 3), ear signals after DC removal filter, 4) influence of pre-emphasis filter, 5) monaural RMS normalization, 6) binaural RMS normalization, 7) level scaling and 8) middle ear filtering.

and 1. Here, we set the coefficient to `pp_coefPreEmphasis = 0.97` by default according to (Young *et al.*, 2006). This pre-emphasis filter can be activated by setting the flag `pp_bPreEmphasis = true`.

4.1.3 RMS normalization

A signal level normalization stage is available which can be used to equalize long-term level differences (e.g. when recording two speakers at two different distances). For some applications, such as ASR and speaker identification systems, it can be advantageous to maintain a constant signal power, such that the features extracted by subsequent processors are invariant to the overall signal level. To achieve this, the input signal is normalized by its RMS value that has been estimated by a first-order low-pass filter with a time

constant of `pp_intTimeSecRMS = 2`. Such a normalization stage has also been suggested in the context of amplitude modulation spectrogram (AMS) feature extraction (Tchorz and Kollmeier, 2003), which are described in Sec. 4.12. The choice of the time constant is a balance between maintaining the level fluctuations across individual words and allowing the normalization stage to follow sudden level changes.

The normalization can be either applied independently for the left and the right ear signal by setting the parameter `pp_bBinauralRMS = false`, or the processing can be linked across ear signals by setting `pp_bBinauralRMS = true`. When being used in the binaural mode, the larger RMS value of both ear signals is used for normalization, which will preserve the binaural cues (e.g. ITDs and ILDs) that are encoded in the signal. The RMS normalization can be activated by the parameter `pp_bNormalizeRMS = true`.

4.1.4 Level reference and scaling

This stage is designed to implement the effect of calibration, in which the amplitude of the incoming digital signal is matched to sound pressure in the physical domain. This operation is necessary when any of the AFE models requires the input to be represented in physical units (such as pascals, see the middle ear filtering stage below). Within the current AFE framework, the dual-resonance non-linear (DRNL) filterbank model requires this signal representation (see Sec. 4.2.2). The request for this is given by setting `pp_bApplyLevelScaling = true`, with a reference value `pp_refSPLdB` in dB SPL which should correspond to the input RMS of 1. Then the input signal is scaled accordingly, if it had been calibrated to a different reference. The default value of `pp_refSPLdB` is 100, which corresponds to the convention used in the work of Jepsen *et al.* (2008). The implementation is adopted from the AMToolbox (Søndergaard and Majdak, 2013).

4.1.5 Middle ear filtering

This stage corresponds to the operation of the middle ear where the vibration from the eardrum is transformed into the stapes motion. The filter model is based on the findings from the measurement of human stapes displacement by Goode *et al.* (1994). Its implementation is adopted from the AMToolbox (Søndergaard and Majdak, 2013), which derives the stapes velocity as the output (Lopez-Poveda and Meddis, 2001, Jepsen *et al.*, 2008). The input is assumed to be the eardrum pressure represented in pascals which in turn assumes prior calibration. This input-output representation in physical units is required particularly when the DRNL filterbank model is used for the basilar membrane (BM) operation, because of its level-dependent nonlinearity, designed based on that representation (see Sec. 4.2.2). When including the middle-ear filtering in combination with the linear gammatone filter, only the simple band-pass characteristic of this model is

needed without the need for input calibration or consideration of the input/output units. The middle ear filtering can be applied by setting `pp_bMiddleEarFiltering = true`. The filter data from Lopez-Poveda and Meddis (2001) or from Jepsen *et al.* (2008) can be used for the processing, by specifying the model `pp_middleEarModel = 'lopezpoveda'` or `pp_middleEarModel = 'jepsen'` respectively.

4.2 Auditory filterbank

One central processing element of the AFE is the separation of incoming acoustic signals into different spectral bands, as it happens in the human inner ear. In psychoacoustic modeling, two different approaches have been followed over the years. One is the simulation of this stage by a *linear* filterbank composed of gammatone filters. This linear gammatone filterbank can be considered a standard element for auditory models and has therefore been included in the TWO!EARS framework. A computationally more challenging, but at the same time physiologically more plausible simulation of this process can be realized by a *nonlinear* BM model, and we have implemented the DRNL model, as developed by Meddis *et al.* (2001). The filterbank representation is requested by using the nametag '`filterbank`'. The filterbank type can be controlled by the parameter `fb_type`. To select a gammatone filterbank, `fb_type` should be set to '`gammatone`' (which is the default), whereas the DRNL filterbank is used when setting `fb_type = 'drnl'`. Some of the parameters are common to the two filterbank, while some are specific, in which case their value is disregarded if the other type of filterbank was requested. Table 4.2 summarizes all parameters corresponding to the '`filterbank`' request. Parameters specific to a filterbank type are separated by a horizontal line. The two filterbank implementations are described in detail in the following two subsections, along with their corresponding parameters.

4.2.1 Gammatone (`gammatoneProc.m`)

The time domain signal can be processed by a bank of gammatone filters that simulates the frequency selective properties of the human BM. The corresponding MATLAB function is adopted from the AMToolbox (Søndergaard and Majdak, 2013). The gammatone filters cover a frequency range between `fb_lowFreqHz` and `fb_highFreqHz` and are linearly spaced on the ERB scale (Glasberg and Moore, 1990). In addition, the distance between adjacent filter center frequencies on the ERB scale can be specified by `fb_nERBs`, which effectively controls the frequency resolution of the gammatone filterbank. There are three different ways to control the center frequencies of the individual gammatone filters:

1. Define a vector with center frequencies, e.g. `fb_cfhz = [100 200 500 ...]`. In this

Table 4.2: List of parameters related to the auditory representation 'filterbank'.

Parameter	Default	Description
<code>fb_type</code>	'gammatone'	Filterbank type, 'gammatone' or 'drnl'
<code>fb_lowFreqHz</code>	80	Lowest characteristic frequency in Hz
<code>fb_highFreqHz</code>	8000	Highest characteristic frequency in Hz
<code>fb_nERBs</code>	1	Distance between adjacent filters in ERB
<code>fb_nChannels</code>	[]	Number of frequency channels
<code>fb_cfHz</code>	[]	Vector of characteristic frequencies in Hz
<code>fb_nGamma</code>	4	Filter order, 'gammatone'-only
<code>fb_bwERBs</code>	1.01859	Filter bandwidth in ERB, 'gammatone'-only
<code>fb_lowFreqHz</code>	80	Lowest characteristic frequency in Hz, 'gammatone'-only
<code>fb_mocIpsi</code>	1	Ipsilateral MOC factor (0 to 1). Given as a scalar (across all frequency channels) or a vector (individual per frequency channel), 'drnl'-only
<code>fb_mocContra</code>	1	Contralateral MOC factor (0 to 1). Same format as <code>fb_mocIpsi</code> , 'drnl'-only
<code>fb_model</code>	'CASP'	DRNL model (reserved for future extension), 'drnl'-only

case, the parameters `fb_lowFreqHz`, `fb_highFreqHz`, `fb_nERBs` and `fb_nChannels` are ignored.

2. Specify `fb_lowFreqHz`, `fb_highFreqHz` and `fb_nChannels`. The requested number of filters `fb_nChannels` will be spaced between `fb_lowFreqHz` and `fb_highFreqHz`. The center frequencies of the first and the last filter will match with `fb_lowFreqHz` and `fb_highFreqHz`, respectively. To accommodate an arbitrary number of filters, the spacing between adjacent filters `fb_nERBs` will be automatically adjusted. Note that this changes the overlap between neighboring filters.
3. It is also possible to specify `fb_lowFreqHz`, `fb_highFreqHz` and `fb_nERBs`. Starting at `fb_lowFreqHz`, the center frequencies will be spaced at a distance of `fb_nERBs` on the ERB scale until the specified frequency range is covered. The center frequency of the last filter will not necessarily match with `fb_highFreqHz`.

The filter order, which determines the slope of the filter skirts, is set to `fb_nGamma` = 4 by default. The bandwidths of the gammatone filters depend on the filter order and the center frequency, and the default scaling factor for a forth-order filter is approximately `fb_bwERBs` = 1.01859. When adjusting the parameter `fb_bwERBs`, it should be noted that the resulting filter shape will deviate from the original gammatone filter as measured by Glasberg and Moore (1990). For instance, increasing `fb_bwERBs` leads to a broader

filter shape. A full list of parameters is shown in Tab. 4.2.

The gammatone filterbank is illustrated in Fig. 4.2, which has been produced by the script `DEMO_Gammatone.m`. The speech signal shown in the left panel is passed through a bank of 16 gammatone filters spaced between 80 and 8000 Hz. The output of each individual filter is shown in the right panel.

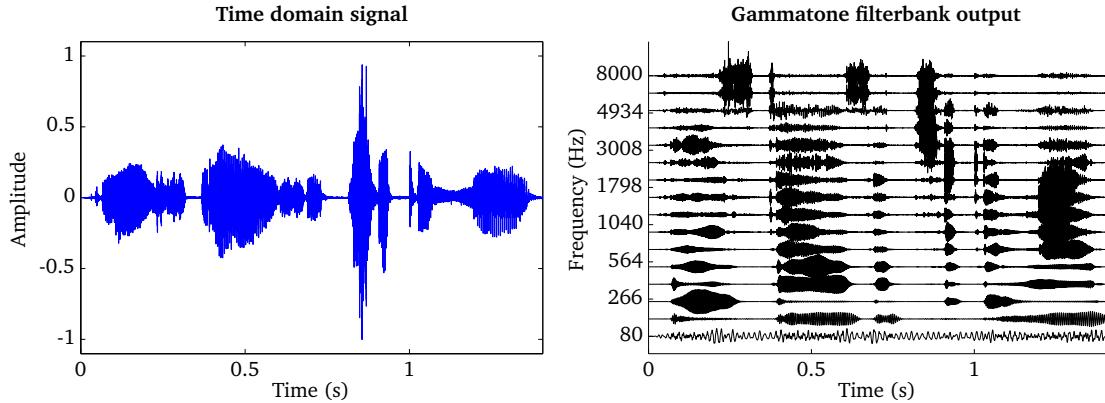


Figure 4.2: Time domain signal (left panel) and the corresponding output of the gammatone processor consisting of 16 auditory filters spaced between 80 and 8000 Hz (right panel).

4.2.2 Dual-resonance non-linear filterbank (`drnlProc.m`)

The DRNL filterbank models the nonlinear operation of the cochlear, in addition to the frequency selective feature of the BM. The DRNL processor was motivated by attempts to better represent the nonlinear operation of the BM in the modelling, and allows for testing the performance of peripheral models with the BM nonlinearity and medial olivo-cochlear (MOC) feedback in comparison to that with the conventional linear BM model. All the internal representations that depend on the BM output can be extracted using the DRNL processor in the dependency chain in place of the gammatone filterbank. This can reveal the implication of the BM nonlinearity and MOC feedback for activities such as speech perception in noise (see Brown *et al.* (2010) for example) or source localisation. It is expected that the use of a nonlinear model, together with the adaptation loops (see Sec. 4.4), will reduce the influence of overall level on the internal representations and extracted features. In this sense, the use of the DRNL model is a physiologically motivated alternative for a linear BM model where the influence of level is typically removed by the use of a level normalization stage (see AGC in Sec. 4.1 for example).

The structure of DRNL filterbank is based on the work of Meddis *et al.* (2001). The frequencies corresponding to the places along the BM, over which the responses are to be derived and observed, are specified as a list of characteristic frequencies `fb_cfHz`. For each

characteristic frequency channel, the time domain input signal is passed through linear and nonlinear paths, as seen in Fig. 4.3. Currently the implementation follows the model defined as computational auditory signal-processing and perception (CASP) by Jepsen *et al.* (2008), in terms of the detailed structure and operation, which is specified by the default argument 'CASP' for `fb_model`.

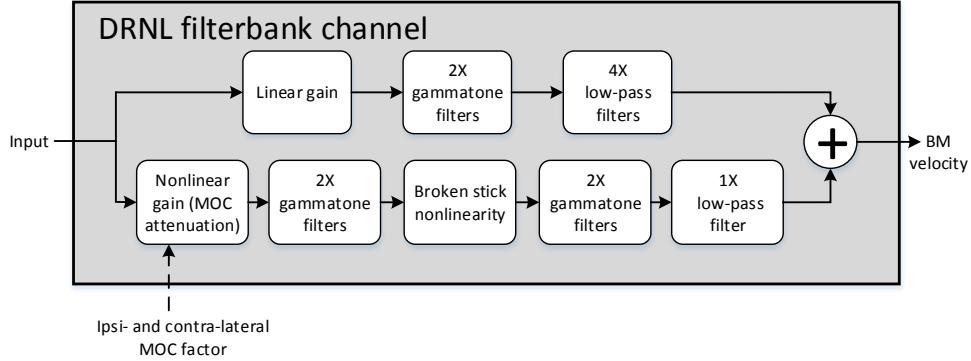


Figure 4.3: DRNL filterbank channel structure, following the CASP model specification (Jepsen *et al.*, 2008) as default, with an additional nonlinear gain stage to receive feedback.

In the CASP model, the linear path consists of a gain stage, two cascaded gammatone filters, and four cascaded low-pass filters; the nonlinear path consists of a gain (attenuation) stage, two cascaded gammatone filters, a 'broken stick' nonlinearity stage, two more cascaded gammatone filters, and a low-pass filter. The outputs at the two paths are then summed as the BM output motion. These sub-modules and their individual parameters (e.g., gammatone filter center frequencies) are specific to the model and hidden to the users. Details regarding the original idea behind the parameter derivation can be found in (Lopez-Poveda and Meddis, 2001), which the CASP model slightly modified to provide a better fit of the output to physiological findings from human cochlear research works.

The MOC feedback is implemented in an open-loop structure within the DRNL filterbank model as the gain factor to be applied to the nonlinear path. This approach is used by Ferry and Meddis (2007), where the attenuation caused by the MOC feedback at each of the filterbank channels is controlled externally by the user. Two additional input arguments are introduced for this feature: `fb_mocIpsi` and `fb_mocContra`. These represent the amount of reflexive feedback through the ipsilateral and contralateral paths, in the form of a factor from 0 to 1 that the nonlinear path input signal is multiplied by in conjunction. Conceptually, `fb_mocIpsi = 1` and `fb_mocContra = 1` would mean that no attenuation is applied to the nonlinear path input, and `fb_mocIpsi = 0` and `fb_mocContra = 0` would mean that the nonlinear path is totally eliminated. Tab. 4.2 summarizes the parameters for the DRNL processor that can be controlled by the user. Note that `fb_cfHz` corresponds to the *characteristic* frequencies and not the *centre* frequencies as used in the

gammatone filterbank, although they can have the same values for comparison. Otherwise, the characteristic frequencies can be generated in the same way as the center frequencies for the gammatone filterbank.

Figure 4.4 shows the BM stage output at 1 kHz characteristic frequency using the DRNL processor (on the right hand side), compared to that using the gammatone filterbank (left hand side), based on the right ear input signal shown in panel 1 of Fig. 4.1 (speech excerpt repeated twice with a level difference). The plots can be generated by running the script DEMO_DRNL.m. It should be noted that the CASP model of DRNL filterbank expects the input signal to be transformed to the middle ear *stapes velocity* before processing. Therefore, for direct comparison of the outputs in this example, the same pre-processing was applied for the gammatone filterbank (stapes velocity was used as the input, through the level scaling and middle ear filtering). It is seen that the level difference between the initial speech component and its repetition is reduced with the nonlinearity incorporated, compared to the gammatone filterbank output, which shows the compressive nature of the nonlinear model responding to input level changes as described earlier.

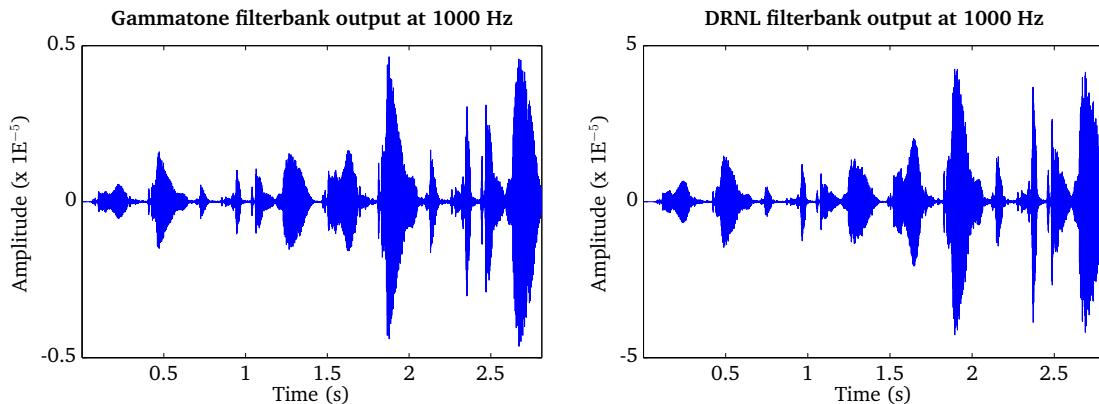


Figure 4.4: The gammatone processor output (left panel) compared to the output of the DRNL processor (right panel), based on the right ear signal shown in panel 1 of Fig. 4.1, at 1 kHz center or characteristic frequency. Note that the input signal is converted to the stapes velocity before entering both processors for direct comparison. The level difference between the two speech excerpts is reduced in the DRNL response, showing its compressive nature to input level variations.

4.3 Inner hair-cell (ihcProc.m)

The IHC functionality is simulated by extracting the envelope of the output of individual gammatone filters. The corresponding IHC function is adopted from the AMToolbox (Søndergaard and Majdak, 2013). Typically, the envelope is extracted by combining half-wave rectification and low-pass filtering. The low-pass filter is motivated by the loss of phase-

locking in the auditory nerve at higher frequencies (Bernstein and Trahiotis, 1996, Bernstein *et al.*, 1999). Depending on the cut-off frequency of the IHC models, it is possible to control the amount of fine-structure information that is present in higher frequency channels. The cut-off frequency and the order of the corresponding low-pass filter vary across methods and a complete overview of supported IHC models is given in Tab. 4.3. A particular model can be selected by using the parameter `ihc_method`.

Table 4.3: List of supported IHC models related to the auditory representation 'innerhaircell'.

<code>ihc_method</code>	Description
'hilbert'	Hilbert transform
'halfwave'	Half-wave rectification
'fullwave'	Full-wave rectification
'square'	Squared
'dau'	Half-wave rectification and low-pass filtering at 1000 Hz (Dau <i>et al.</i> , 1996)
'joergensen'	Hilbert transform and low-pass filtering at 150 Hz (Jørgensen and Dau, 2011)
'breebart'	Half-wave rectification and low-pass filtering at 770 Hz (Breebaart <i>et al.</i> , 2001)
'bernstein'	Half-wave rectification, compression and low-pass filtering at 425 Hz (Bernstein <i>et al.</i> , 1999)

The effect of the IHC processor is demonstrated in Fig. 4.5, where the output of the gammatone filterbank is compared with the output of an IHC model by running the script `DEMO_IHC.m`. Whereas individual peaks are resolved in the lowest channel of the IHC output, only the envelope is retained at higher frequencies.

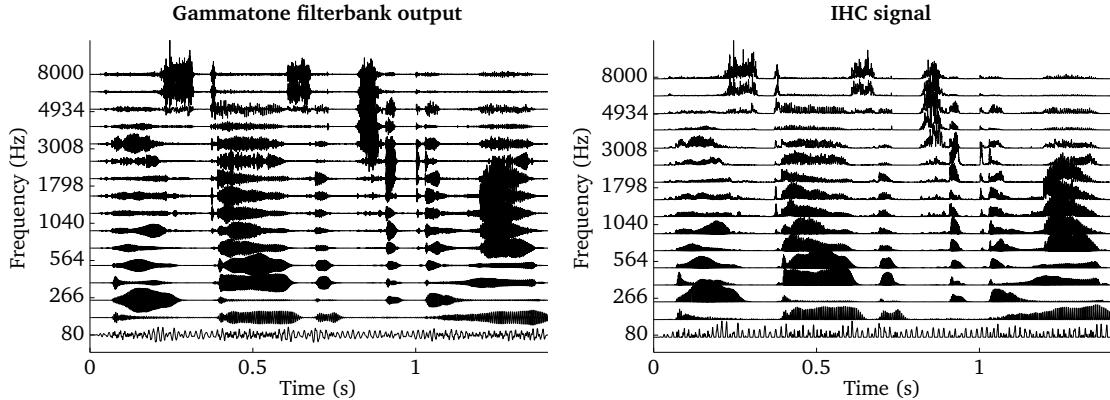


Figure 4.5: Illustration of the envelope extraction processor. BM output (left panel) and the corresponding IHC model output using `ihc_method = 'dau'` (right panel).

4.4 Adaptation (adaptationProc.m)

This processor corresponds to the adaptive response of the auditory nerve fibers, in which abrupt changes in the input result in emphasised overshoots followed by gradual decay to compressed steady-state level (Smith, 1977, Smith *et al.*, 1983). The function is adopted from the AMToolbox (Søndergaard and Majdak, 2013). The adaptation stage is modelled as a chain of five feedback loops in series. Each of the loops consists of a low-pass filter with its own time constant, and a division operator (Püschen, 1988, Dau *et al.*, 1996, 1997a). At each stage, the input is divided by its low-pass filtered version. The time constant affects the charging / releasing state of the filter output at a given moment, and thus affects the amount of attenuation caused by the division. This implementation realises the characteristics of the process that input variations which are rapid compared to the time constants are linearly transformed, whereas stationary input signals go through logarithmic compression.

The adaptation processor uses three parameters to generate the output from the IHC representation: `adpt_lim` determines the maximum ratio of the onset response amplitude against the steady-state response, which sets a limit to the overshoot caused by the loops. `adpt_mindB` sets the lowest audible threshold of the input signal. `adpt_tau` are the time constants of the loops. Though the default model uses five loops and thus five time constants, variable number of elements of `adpt_tau` is supported which can vary the number of loops. Some specific sets of these parameters, as used in related studies, are also supported optionally with the `adpt_model` parameter. This can be given instead of the other three parameters, which will set them as used by the respective researchers. Tab. 4.4 lists the parameters and their default values, and Tab. 4.5 lists the supported models. The output signal is expressed in model units (MU) which deviates the input-output relation from a perfect logarithmic transform, such that the input level increment

Table 4.4: List of parameters related to the auditory representation 'adaptation'.

Parameter	Default	Description
<code>adpt_lim</code>	10	Overshoot limiting ratio
<code>adpt_mindB</code>	0	Lowest audible threshold of the signal (in dB SPL)
<code>adpt_tau</code>	[0.005 0.050 0.129 0.253 0.500]	Time constants of feedback loops
<code>adpt_model</code>	'adt_dau'	Implementation model ('adt_dau', 'adt_puschel', or 'adt_breebart'): can be used instead of the above three parameters (See Tab. 4.5)

at low level range results in a smaller output level increment than the input increment at higher level range. This corresponds to a smaller just-noticeable level change at high levels than at low levels (Dau *et al.*, 1996). Jepsen *et al.* (2008), with the use of DRNL model for the BM stage, introduces an additional squaring expansion process between the IHC output and the adaptation stage, which transforms the input that comes through the DRNL - IHC processors into an intensity-like representation to be compatible with the adaptation implementation originally designed based on the use of gammatone filterbank. The adaptation processor recognises whether DRNL or gammatone processor is used in the chain and adjusts the input signal accordingly.

Table 4.5: List of supported models related to the auditory representation 'adaptation'.

adpt_model	Description
'adt_dau'	Choose the parameters as in the models of Dau <i>et al.</i> (1996, 1997a). This consists of 5 adaptation loops with an overshoot limit of 10 and a minimum level of 0 dB. This is a correction in regard to the model described in Dau <i>et al.</i> (1996), which did not use overshoot limiting. The adaptation loops have an exponentially spaced time constants (<code>adpt_tau=[0.005 0.050 0.129 0.253 0.500]</code>)
'adt_puschel'	Choose the parameters as in the original model (Püschel, 1988). This consists of 5 adaptation loops without overshoot limiting (<code>adpt_lim=0</code>). The adaptation loops have a linearly spaced time constants (<code>adpt_tau=[0.0050 0.1288 0.2525 0.3762 0.5000]</code>).
'adt_breebaart'	As 'adt_puschel', but with overshoot limiting

The effect of the adaptation processor - the exaggeration of rapid variations - is demonstrated in Fig. 4.6, where the output of the IHC model from the same input as used in the example of Sec. 4.3 (the right panel of Fig. 4.5) is compared to the adaptation output by running the script `DEMO_Adaptation.m`.

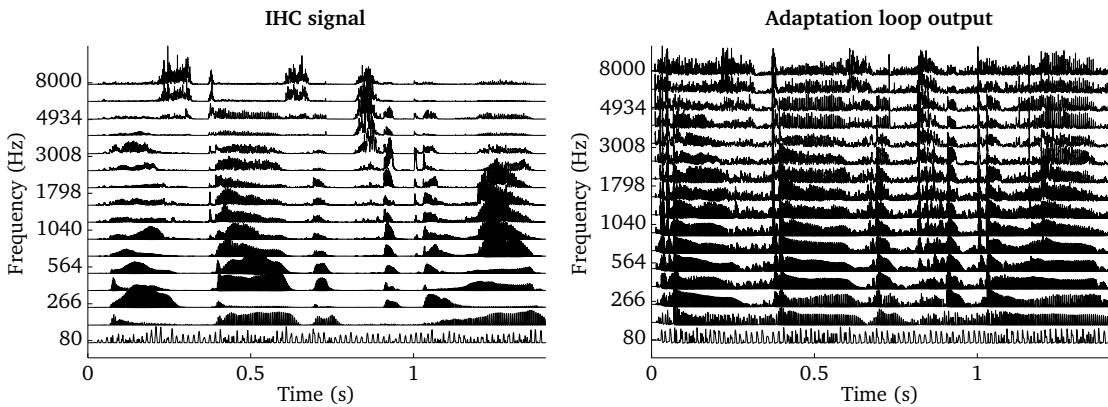


Figure 4.6: Illustration of the adaptation processor. IHC output (left panel) as the input to the adaptation processor and the corresponding output using `adpt_model='adt_dau'` (right panel).

4.5 Auto-correlation (autocorrelationProc.m)

Auto-correlation is an important computational concept that has been extensively studied in the context of predicting human pitch perception (Licklider, 1951, Meddis and Hewitt, 1991). To measure the amount of periodicity that is present in individual frequency channels, the auto-correlation function (ACF) is computed in the fast Fourier transform (FFT) domain for short time frames based on the IHC representation. The *unbiased* ACF scaling is used to account for the fact that fewer terms contribute to the ACF at longer time lags. The resulting ACF is normalized by the ACF at lag zero to ensure values between minus one and one. The window size `ac_wSizeSec` determines how well low-frequency pitch signals can be reliably estimated and common choices are within the range of 10 ms – 30 ms.

For the purpose of pitch estimation, it has been suggested to modify the signal prior to correlation analysis in order to reduce the influence of the formant structure on the resulting ACF (Rabiner, 1977). This pre-processing can be activated by the flag `ac_bCenterClip` and the following nonlinear operations can be selected for `ac_ccMethod`: center clip and compress '`c1c`', center clip '`cc`', and combined center and peak clip '`sgn`'. The percentage of center clipping is controlled by the flag `ac_ccAlpha`, which sets the clipping level to a fixed percentage of the frame-based maximum signal level.

A generalized ACF has been suggested by Tolonen and Karjalainen (2000), where the exponent `ac_K` can be used to control the amount of compression that is applied to the ACF. The conventional ACF function is computed using a value of `ac_K = 2`, whereas the function is compressed when a smaller value than 2 is used. The choice of this parameter is a trade-off between sharpening the peaks in the resulting ACF function and amplifying the noise floor. A value of `ac_K = 2/3` has been suggested as a good

compromise (Tolonen and Karjalainen, 2000). A list of all ACF-related parameters is given in Tab. 4.6. Note that these parameters will influence the pitch processor, which is described in Sec. 4.11.

Table 4.6: List of parameters related to the auditory representation ‘autocorrelation’.

Parameter	Default	Description
ac_wname	‘hann’	Window type
ac_wSizeSec	0.02	Window duration (s)
ac_hSizeSec	0.01	Window step size (s)
ac_bCenterClip	false	Activate center clipping
ac_clipMethod	‘clp’	Center clipping method (‘clc’, ‘clp’, or ‘sgn’)
ac_clipAlpha	0.6	Center clipping threshold within [0, 1]
ac_K	2	Exponent in ACF

A demonstration of the ACF processor is shown in Fig. 4.7, which has been produced by the scrip `DEMO_ACF.m`. It shows the IHC output in response to a 20 ms speech signal for 16 frequency channels (left panel). The corresponding ACF is presented in the upper right panel, whereas the summary auto-correlation function (SACF) is shown in the bottom right panel. Prominent peaks in the SACF indicate lag periods which correspond to integer multiples of the fundamental frequency of the analyzed speech signal. This relationship is exploited by the pitch processor, which is described in Sec. 4.11.

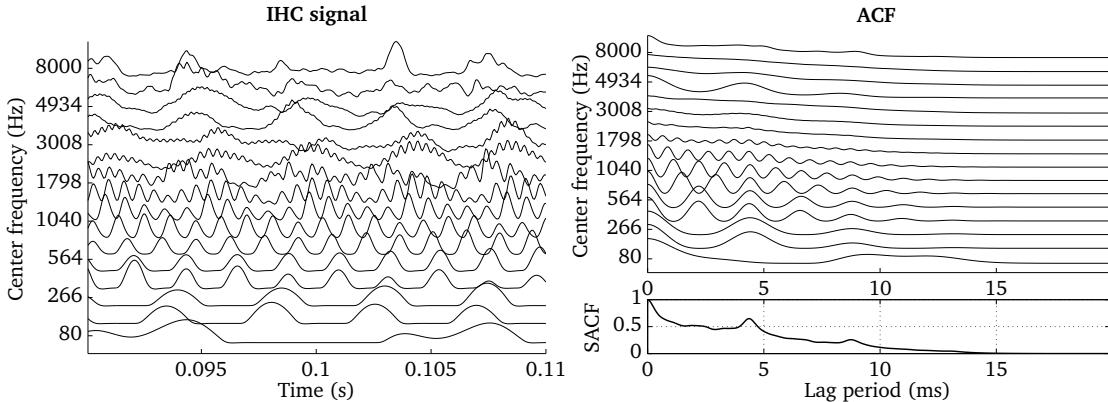


Figure 4.7: IHC representation of a speech signal shown for one time frame of 20 ms duration (left panel) and the corresponding ACF (right panel). The SACF summarizes the ACF across all frequency channels (bottom right panel).

4.6 Ratemap (`ratemapProc.m`)

The ratemap represents a map of auditory nerve firing rates (Brown and Cooke, 1994) and is frequently employed as a spectral feature in computational auditory scene analysis (CASA) systems (Wang and Brown, 2006), ASR (Cooke *et al.*, 2001) and speaker identification systems (May *et al.*, 2012). The ratemap is computed for individual frequency channels by smoothing the IHC signal representation with a leaky integrator that has a time constant of typically `rm_decaySec` = 8 ms. Then, the smoothed IHC signal is averaged across all samples within a time frame and thus the ratemap can be interpreted as an auditory spectrogram. Depending on whether the ratemap scaling `rm_scaling` has been set to '`magnitude`' or '`power`', either the magnitude or the squared samples are averaged within each time frame. The temporal resolution can be adjusted by the window size `rm_wSizeSec` and the step size `rm_hSizeSec`. Moreover, it is possible to control the shape of the window function `rm_wname`, which is used to weight the individual samples within a frame prior to averaging. The default ratemap parameters are listed in Tab. 4.7.

Table 4.7: List of parameters related to the auditory representation '`ratemap`'.

Parameter	Default	Description
<code>rm_wname</code>	'hann'	Window type
<code>rm_wSizeSec</code>	0.02	Window duration (s)
<code>rm_hSizeSec</code>	0.01	Window step size (s)
<code>rm_scaling</code>	'power'	Ratemap scaling (' <code>magnitude</code> ' or ' <code>power</code> ')
<code>rm_decaySec</code>	0.008	Leaky integrator time constant (s)

The ratemap is demonstrated by the script `DEMO_Ratemap` and the corresponding plots are presented in Fig. 4.8. The IHC representation of a speech signal is shown in the left panel, using a bank of 64 gammatone filters spaced between 80 and 8000 Hz. The corresponding ratemap representation scaled in dB is presented in the right panel.

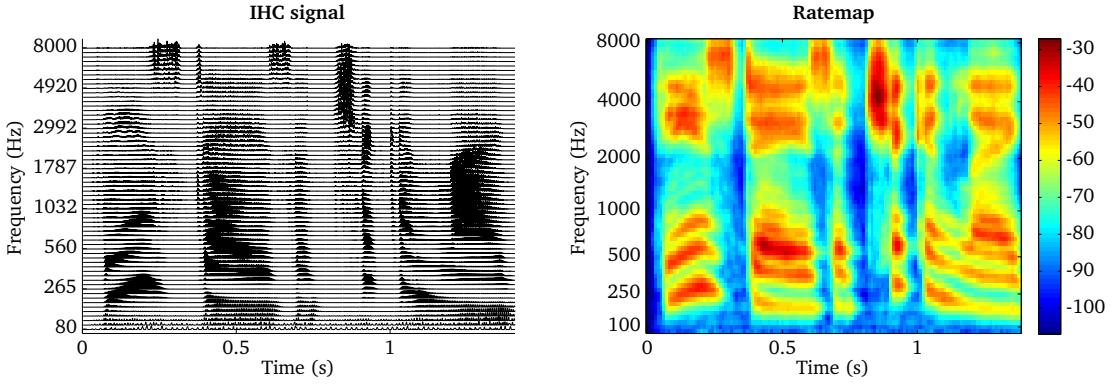


Figure 4.8: IHC representation of s speech signal using 64 auditory filters (left panel) and the corresponding ratemap representation (right panel).

4.7 Spectral features (spectralFeaturesProc.m)

In order to characterize the spectral content of the ear signals, a set of spectral features is available that can serve as a physical correlate to perceptual attributes, such as timbre and coloration (Peeters *et al.*, 2011). All spectral features summarize the spectral content of the ratemap representation across auditory filters and are computed for individual time frames. The following 14 spectral features are available:

1. **'centroid'** : The spectral centroid represents the center of gravity of the ratemap and is one of the most frequently-used timbre parameters (Tzanetakis and Cook, 2002, Jensen and Andersen, 2004, Peeters *et al.*, 2011). The centroid is normalized by the highest ratemap center frequency to reduce the influence of the gammatone parameters.
2. **'spread'** : The spectral spread describes the average deviation of the ratemap around its centroid, which is commonly associated with the bandwidth of the signal. Noise-like signals have usually a large spectral spread, while individual tonal sounds with isolated peaks will result in a low spectral spread. Similar to the centroid, the spectral spread is normalized by the highest ratemap center frequency, such that the feature value ranges between zero and one.
3. **'brightness'** : The brightness reflects the amount of high frequency information and is measured by relating the energy above a pre-defined cutoff frequency to the total energy. This cutoff frequency is set to `sf_br_cf = 1500 Hz` by default (Jensen and Andersen, 2004, Peeters *et al.*, 2011). This feature might be used to quantify the sensation of sharpness.
4. **'high-frequency content'** : The high-frequency content is another metric that

measures the energy associated with high frequencies. It is derived by weighting each channel in the ratemap by its squared center frequency and integrating this representation across all frequency channels (Jensen and Andersen, 2004). To reduce the sensitivity of this feature to the overall signal level, the high-frequency content feature is normalized by the ratemap integrated across-frequency.

5. **'crest'** : The spectral crest measure (SCM) is defined as the ratio between the maximum value and the arithmetic mean and can be used to characterize the peakiness of the ratemap. The feature value is low for signals with a flat spectrum and high for a ratemap with a distinct spectral peak (Peeters *et al.*, 2011, Lerch, 2012).
6. **'decrease'** : The spectral decrease describes the average spectral slope of the ratemap representation, putting a stronger emphasis on the low frequencies. (Peeters *et al.*, 2011)
7. **'entropy'** : The entropy can be used to capture the peakiness of the spectral representation (Misra *et al.*, 2004). The resulting feature is low for a ratemap with many distinct spectral peaks and high for a flat ratemap spectrum.
8. **'flatness'** : The spectral flatness measure (SFM) is defined as the ratio of the geometric mean to the arithmetic mean and can be used to distinguish between harmonic (SFM is close to zero) and a noisy signals (SFM is close to one) (Peeters *et al.*, 2011).
9. **'irregularity'** : The spectral irregularity quantifies the variations of the logarithmically-scaled ratemap across frequencies (Jensen and Andersen, 2004).
10. **'kurtosis'** : The excess kurtosis measures whether the spectrum can be characterized by a Gaussian distribution (Lerch, 2012). This feature will be zero for a Gaussian distribution.
11. **'skewness'** : The spectral skewness measures the symmetry of the spectrum around its arithmetic mean (Lerch, 2012). The feature will be zero for silent segments and high for voiced speech where substantial energy is present around the fundamental frequency.
12. **'roll-off'** : Determines the frequency in Hz below which a pre-defined percentage `sf_ro_perc` of the total spectral energy is concentrated. Common values for this threshold are between `sf_ro_perc` = 0.85 (Tzanetakis and Cook, 2002) and `sf_ro_perc` = 0.95 (Scheirer and Slaney, 1997, Peeters *et al.*, 2011). The roll-off feature is normalized by the highest ratemap center frequency and ranges between zero and one. This feature can be useful to distinguish voiced from unvoiced signals.
13. **'flux'** : The spectral flux evaluates the temporal variation of the logarithmically-

scaled ratemap across adjacent frames (Lerch, 2012). It has been suggested to be useful for the distinction of music and speech signals, since music has a higher rate of change (Scheirer and Slaney, 1997).

14. '**variation**' : The spectral variation is defined as one minus the normalized correlation between two adjacent time frames of the ratemap (Peeters *et al.*, 2011).

A list of all parameters is presented in Tab. 4.8.

Table 4.8: List of parameters related to the auditory representation '**spectral_features**'.

Parameter	Default	Description
sf_requests	'all'	List of requested spectral features (e.g. 'flux'). Type <code>help spectralFeaturesProc</code> in the MATLAB command window to display the full list of supported spectral features.)
sf_br_cf	1500	Cut-off frequency in Hz for brightness feature
sf_ro_perc	0.85	Threshold (re. 1) for spectral roll-off feature

The extraction of spectral features is demonstrated by the script `Demo_SpectralFeatures.m`, which produces the plots shown in Fig. 4.9. The complete set of 14 spectral features is computed for the speech signal shown in the top left panel. Whenever the unit of the spectral feature was given in frequency, the feature is shown in black in combination with the corresponding ratemap representation.

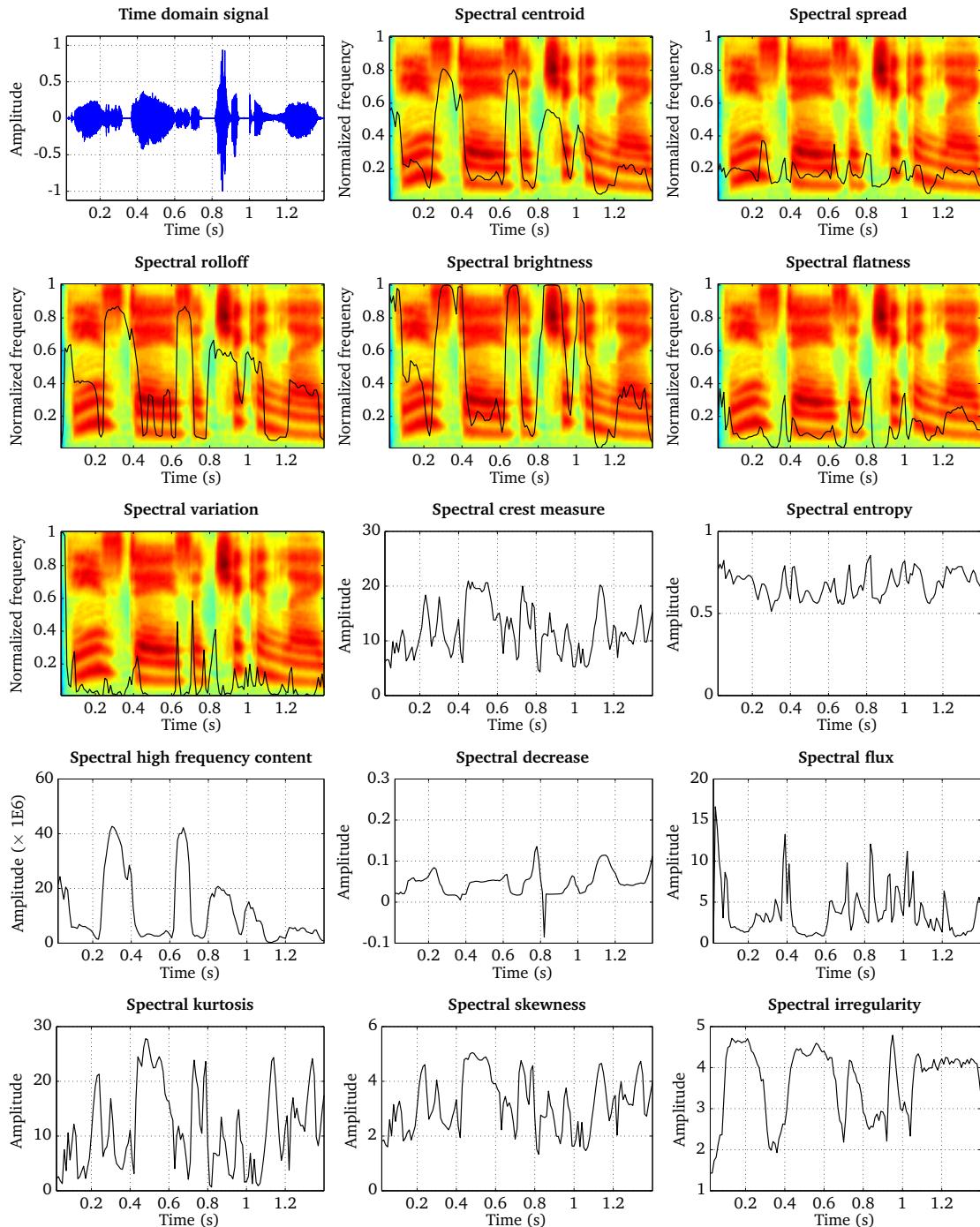


Figure 4.9: Speech signal and 14 spectral features that were extracted based on the ratemap representation.

4.8 Onset strength (`onsetProc.m`)

According to Bregman (1990), common onsets and offsets across frequency are important grouping cues that are utilized by the human auditory system to organize and integrate sounds originating from the same source. The onset processor is based on the ratemap representation, and therefore, the choice of the ratemap parameters, as listed in Tab. 4.7, will influence the output of the onset processor. The temporal resolution is controlled by the window size `rm_wSizeSec` and the step size `rm_hSizeSec`, respectively. The amount of temporal smoothing can be adjusted by the leaky integrator time constant `rm_decaySec`, which reduces the amount of temporal fluctuations in the ratemap. Onsets are detected by measuring the frame-based increase in energy of the ratemap representation. This detection is performed based on the logarithmically-scaled energy, as suggested by Kläppuri (1999). It is possible to limit the strength of individual onsets to an upper limit, which is by default set to `ons_maxOnsetdB = 30`. A list of all parameters is presented in Tab. 4.9.

Table 4.9: List of parameters related to the auditory representation '`onset_strength`'.

Parameter	Default	Description
<code>ons_maxOnsetdB</code>	30	Upper limit for onset strength in dB

The resulting onset strength expressed in dB, which is a function of time frame and frequency channel, is shown in Fig. 4.10. The two figures can be replicated by running the script `DEMO_OnsetStrength.m`. When considering speech as an input signal, it can be seen that onsets appear simultaneously across a broad frequency range and typically mark the beginning of an auditory event.

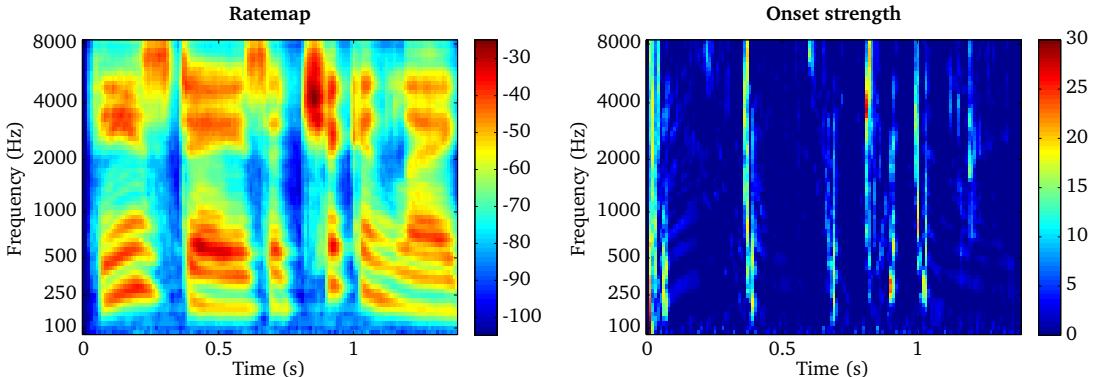


Figure 4.10: Ratemap representation (left panel) of speech and the corresponding onset strength in dB (right panel).

4.9 Offset strength (`offsetProc.m`)

Similarly to onsets, the strength of offsets can be estimated by measuring the frame-based decrease in logarithmically-scaled energy. As discussed in the previous section, the selected ratemap parameters as listed in Tab. 4.7 will influence the offset processor. Similar to the onset strength, the offset strength can be constrained to a maximum value of `ons_maxOffsetdB` = 30. A list of all parameters is presented in Tab. 4.10.

Table 4.10: List of parameters related to the auditory representation '`offset_strength`'.

Parameter	Default	Description
<code>ofs_maxOffsetdB</code>	30	Upper limit for offset strength in dB

The offset strength is demonstrated by the script `DEMO_OffsetStrength.m` and the corresponding figures are depicted in Fig. 4.11. It can be seen that the overall magnitude of the offset strength is lower compared to the onset strength. Moreover, the detected offsets are less synchronized across frequency.

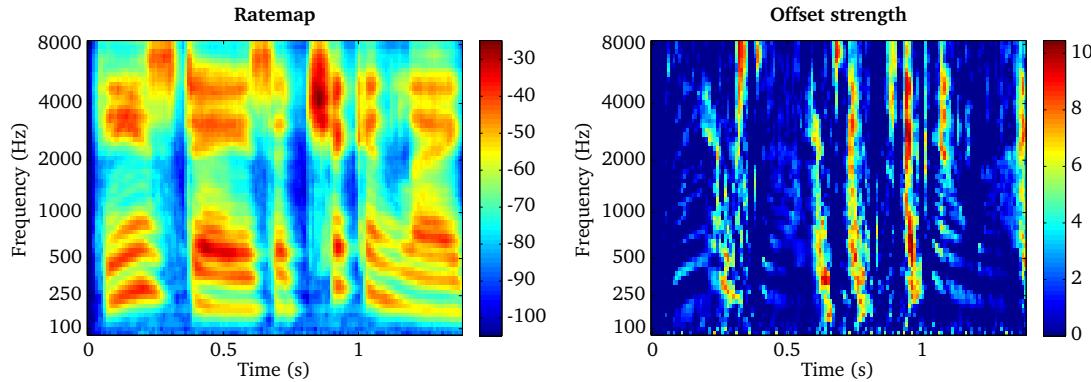


Figure 4.11: Ratemap representation (left panel) of speech and the corresponding offset strength in dB (right panel).

4.10 Binary onset and offset maps (`transientMapProc.m`)

The information about sudden intensity changes, as represented by onsets or offsets, can be combined in order to organize and group the acoustic input according to individual auditory events. The required processing is similar for both onsets and offsets, and is summarized by the term *transient detection*. To apply this transient detection based on the onset strength or offset strength, the user should use the request name '`onset_map`' or '`offset_map`', respectively. Based on the transient strength which is derived from the

corresponding onset strength and offset strength processor (described in Sec. 4.8 and 4.9, a binary decision about transient activity is formed, where only the most salient information is retained. To achieve this, temporal and across-frequency constraints are imposed for the transient information. Motivated by the observation that two sounds are perceived as separated auditory events when the difference in terms of their onset time is in the range of 20 ms – 40 ms (Turgeon *et al.*, 2002), transients are fused if they appear within a pre-defined *time context*. If two transients appear within this time context, only the stronger one will be considered. This time context can be adjusted by `trm_fuseWithinSec`. Moreover, the minimum across-frequency context can be controlled by the parameters `trm_minSpread`. To allow for this selection, individual transients which are connected across multiple time-frequency (T-F) units are extracted using MATLAB’s image labeling tool `bwlabeled`. The binary transient map will only retain those transients which consists of at least `trm_minSpread` connected T-F units. The salience of the cue can be specified by the detection thresholds `trm_minStrengthdB`. Whereas this thresholds control the required relative change, a global threshold excludes transient activity if the corresponding ratemap level is below a pre-defined threshold, as determined by `trm_minValuedB`. A summary of all parameters is given in Tab. 4.11.

Table 4.11: List of parameters related to the auditory representation ‘`onset_map`’ and ‘`offset_map`’.

Parameter	Default	Description
<code>trm_fuseWithinSec</code>	<code>30E-3</code>	Time constant below which transients are fused
<code>trm_minSpread</code>	5	Minimum number of connected T-F units
<code>trm_minStrengthdB</code>	3	Minimum onset strength in dB
<code>trm_minValuedB</code>	-80	Minimum ratemap level in dB

To illustrate the benefit of selecting onset and offset information, a ratemap representation is shown in Fig. 4.12 (left panel), where the corresponding onsets and offsets detected by the `transientMapProc`, through two individual requests ‘`onset_map`’ and ‘`offset_map`’, and without applying any temporal or across-frequency constraints are overlaid (respectively in black and white). It can be seen that the onset and offset information is quite noisy. When only retaining the most salient onsets and offsets by applying temporal and across-frequency constraints (right panel), the remaining onsets and offsets can be used as temporal markers, which clearly mark the beginning and the end of individual auditory events.

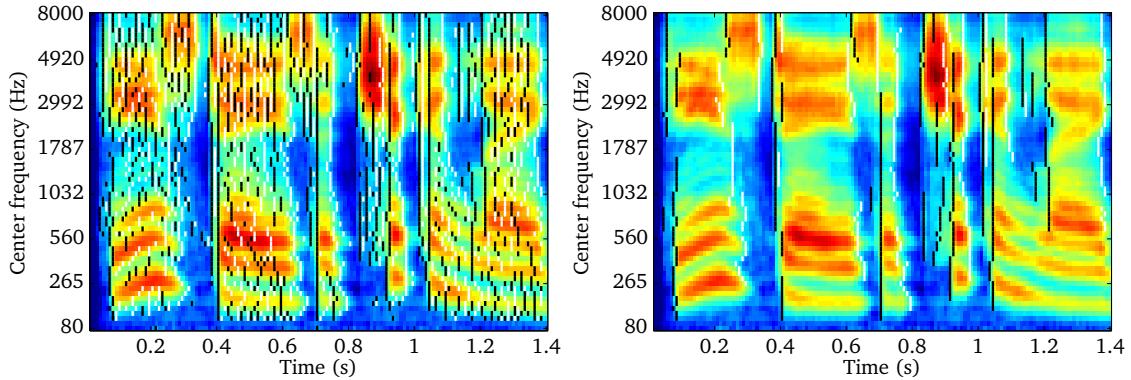


Figure 4.12: Detected onsets and offsets indicated by the black and white vertical bars. The left panels shows all onset and offset events, whereas the right panel applies temporal and across-frequency constraints in order to retain the most salient onset and offset events.

4.11 Pitch (pitchProc.m)

Following Slaney and Lyon (1990), Meddis *et al.* (2001), Meddis and O'Mard (1997), the subband periodicity analysis obtained by the ACF can be integrated across frequency by giving equal weight to each frequency channel. The resulting SACF reflects the strength of periodicity as a function of the lag period for a given time frame, as illustrated in Fig. 4.7. Based on the SACF representation, the most salient peak within the plausible pitch frequency range `p_pitchRangeHz` is detected for each frame in order to obtain an estimation of the fundamental frequency. In addition to the peak position, the corresponding amplitude of the SACF is used to reflect the confidence of the underlying pitch estimation. More specifically, if the SACF magnitude drops below a pre-defined percentage `p_confThresPerc` of its global maximum, the corresponding pitch estimate is considered unreliable and set to zero. The estimated pitch contour is smoothed across time frames by a median filter of order `p_orderMedFilt`, which aims at reducing the amount of octave errors. A list of all parameters is presented in Tab. 4.12. In the context of pitch estimation, it will be useful to experiment with the settings related to the non-linear pre-processing of the ACF, as described in Sec. 4.5.

Table 4.12: List of parameters related to the auditory representation 'pitch'.

Parameter	Default	Description
<code>p_pitchRangeHz</code>	[80 400]	Plausible pitch frequency range in Hz
<code>p_confThresPerc</code>	0.7	Confidence threshold related to the SACF magnitude
<code>p_orderMedFilt</code>	3	Order of the median filter

The task of pitch estimation is demonstrated by the script `DEMO_Pitch` and the correspond-

ing plots are presented in Fig. 4.13. The pitch is estimated for an anechoic speech signal (top left panel). The corresponding SACF is presented in the top right panel, where each black cross represents the most salient lag period per time frame. The plausible pitch range is indicated by the two white dashed lines. The confidence measure of each individual pitch estimates is shown in the bottom left panel, which is used to set the estimated pitch to zero if the magnitude of the SACF is below the threshold. The final pitch contour is post-processed with a median filter and shown in the bottom right panel. Unvoiced frames, where no pitch frequency was detected, are indicated by NaN's.

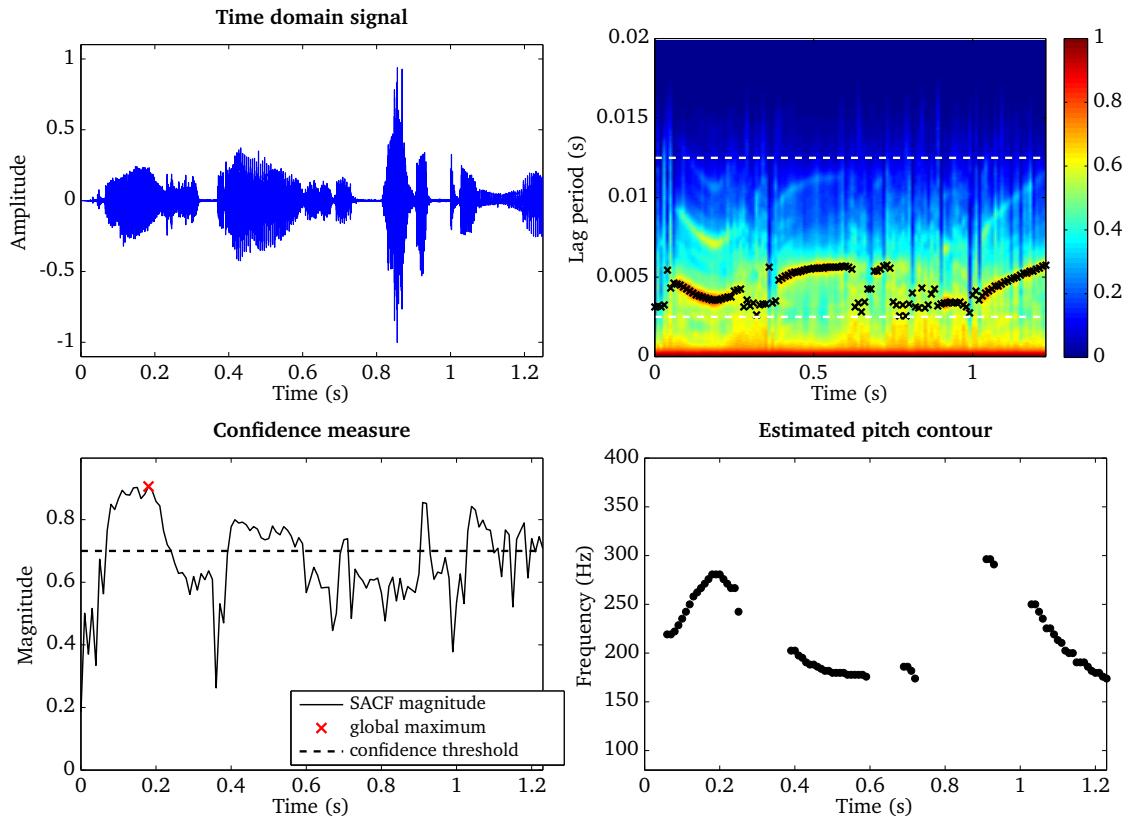


Figure 4.13: Time domain signal (top left panel) and the corresponding SACF (top right panel). The confidence measure based on the SACF magnitude is used to select reliable pitch estimates (bottom left panel). The final pitch estimate is post-processed by a median filter (bottom right panel).

4.12 Amplitude modulation spectrogram (modulationProc.m)

The detection of envelope fluctuations is a very fundamental ability of the human auditory system which plays a major role in speech perception. Consequently, computational models have tried to exploit speech- and noise specific characteristics of amplitude modulations by extracting so-called amplitude modulation spectrogram (AMS) features with linearly-scaled modulation filters (Kollmeier and Koch, 1994, Tchorz and Kollmeier, 2003, Kim *et al.*, 2009, May and Dau, 2013, May and Gerkmann, 2014, May and Dau, 2014). The use of linearly-scaled modulation filters is, however, not consistent with psychoacoustic data on modulation detection and masking in humans (Bacon and Grantham, 1989, Houtgast, 1989, Dau *et al.*, 1997a,b, Ewert and Dau, 2000). As demonstrated by Ewert and Dau (2000), the processing of envelope fluctuations can be described effectively by a second-order band-pass filterbank with logarithmically-spaced center frequencies. Moreover, it has been shown that an AMS feature representation based on an auditory-inspired modulation filterbank with logarithmically-scaled modulation filters substantially improved the performance of computational speech segregation in the presence of stationary and fluctuating interferers (May and Dau, in press). In addition, such a processing based on auditory-inspired modulation filters has recently also been successful in speech intelligibility prediction studies (Jørgensen and Dau, 2011, Jørgensen *et al.*, 2013). To investigate the contribution of both AMS feature representations, the amplitude modulation processor can be used to extract linearly- and logarithmically-scaled AMS features. Therefore, each frequency channel of the IHC representation is analyzed by a bank of modulation filters. The type of modulation filters can be controlled by setting the parameter `ams_fbType` to either '`lin`' or '`log`'. To illustrate the difference between linear linearly-scaled and logarithmically-scaled modulation filters, the corresponding filterbank responses are shown in Fig. 4.14. The linear modulation filterbank is implemented in the frequency domain, whereas the logarithmically-scaled filterbank is realized by a band of second-order IIR butterworth filters with a constant-Q factor of 1. The modulation filter with the lowest center frequency is always implemented as a low-pass filter, as illustrated in the right panel of Fig. 4.14.

Similarly to the gammatone processor described in Sect 4.2.1, there are different ways to control the center frequencies of the individual modulation filters, which depend on the type of modulation filters

- `ams_fbType = 'lin'`
 1. Specify `ams_lowFreqHz`, `ams_highFreqHz` and `ams_nFilter`. The requested number of filters `ams_nFilter` will be linearly-spaced between `ams_lowFreqHz` and `ams_highFreqHz`. If `ams_nFilter` is omitted, the number of filters will be set to 15 by default.

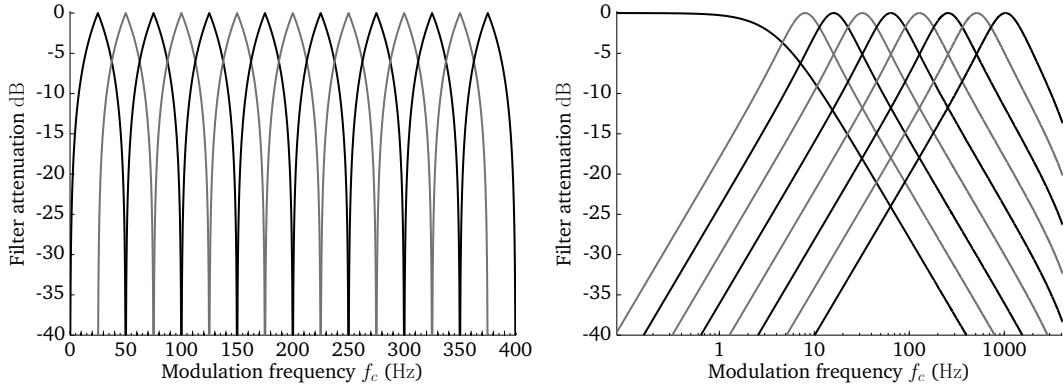


Figure 4.14: Transfer functions of 15 linearly-scaled (left panel) and 9 logarithmically-scaled (right panel) modulation filters.

- `ams_fbType = 'log'`
 1. Directly define a vector of center frequencies, e.g. `ams_cfHz = [4 8 16 ...]`. In this case, the parameters `ams_lowFreqHz`, `ams_highFreqHz`, and `ams_nFilter` are ignored.
 2. Specify `ams_lowFreqHz` and `ams_highFreqHz`. Starting at `ams_lowFreqHz`, the center frequencies will be logarithmically-spaced at integer powers of two, e.g. $2^2, 2^3, 2^4 \dots$ until the higher frequency limit `ams_highFreqHz` is reached.
 3. Specify `ams_lowFreqHz`, `ams_highFreqHz` and `ams_nFilter`. The requested number of filters `ams_nFilter` will be spaced logarithmically as power of two between `ams_lowFreqHz` and `ams_highFreqHz`.

The temporal resolution at which the AMS features are computed is specified by the window size `ams_wSizeSec` and the step size `ams_hSizeSec`. The window size is an important parameter, because it determines how many periods of the lowest modulation frequencies can be resolved within one individual time frame. Moreover, the window shape can be adjusted by `ams_wname`. Finally, the IHC representation can be downsampled prior to modulation analysis by selecting a downsampling ratio `ams_dsRatio` larger than 1. A full list of AMS feature parameters is shown in Tab. 4.13.

The functionality of the AMS feature processor is demonstrated by the script `DEMO_AMS` and the corresponding four plots are presented in Fig. 4.15. The time domain speech signal (top left panel) is transformed into a IHC representation (top right panel) using 23 frequency channels spaced between 80 and 8000 Hz. The linear and the logarithmic AMS feature representations are shown in the bottom panels. The response of the modulation filters are stacked on top of each other for each IHC frequency channel, such that the

Table 4.13: List of parameters related to the auditory representation '`ams_features`'.

Parameter	Default	Description
<code>ams_fbType</code>	'log'	Filterbank type ('lin' or 'log')
<code>ams_nFilter</code>	[]	Number of modulation filters (integer)
<code>ams_lowFreqHz</code>	4	Lowest modulation filter center frequency in Hz
<code>ams_highFreqHz</code>	1024	Highest modulation filter center frequency in Hz
<code>ams_cfHz</code>	[]	Vector of modulation filter center frequencies in Hz
<code>ams_dsRatio</code>	4	Downsampling ratio of the IHC representation
<code>ams_wSizeSec</code>	32E-3	Window duration in s
<code>ams_hSizeSec</code>	16E-3	Window step size in s
<code>ams_wname</code>	'rectwin'	Window name

AMS feature representations can be read like spectrograms. It can be seen that the linear AMS feature representation is more noisy in comparison to the logarithmically-scaled AMS features. Moreover, the logarithmically-scaled modulation pattern shows a much higher correlation with the activity reflected in the IHC representation.

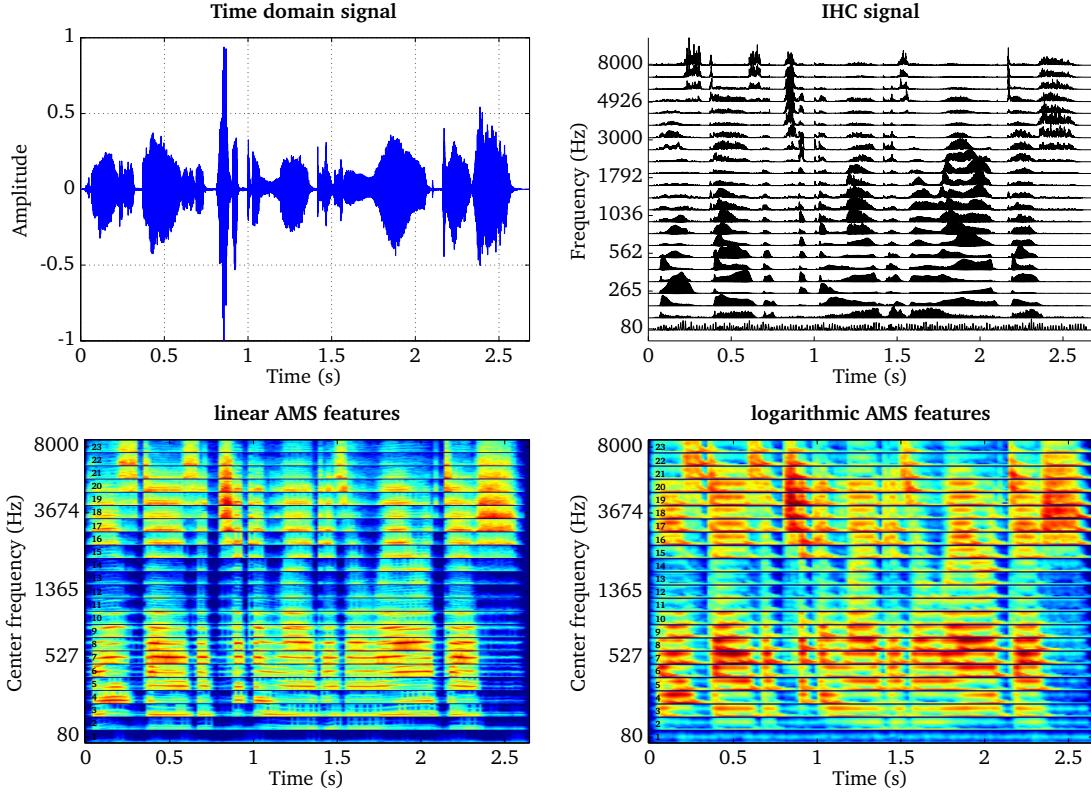


Figure 4.15: Speech signal (top left panel) and the corresponding IHC representation (top right panel) using 23 frequency channels spaced between 80 and 8000 Hz. Linear AMS features (bottom left panel) and logarithmic AMS features (bottom right panel). The response of the modulation filters are stacked on top of each other for each IHC frequency channel, and each frequency channel is visually separated by a horizontal black line. The individual frequency channels, ranging from 1 to 23, are labels at the left hand side.

4.13 Spectro-temporal modulation spectrogram

Neuro-physiological studies suggest that the response of neurons in the primary auditory cortex of mammals are tuned to specific spectro-temporal patterns (Theunissen *et al.*, 2001, Qiu *et al.*, 2003). This response characteristic of neurons can be described by the so-called spectro-temporal receptive field (STRF). As suggested by Qiu *et al.* (2003), the STRF can be effectively modeled by two-dimensional (2D) Gabor functions. Based on these findings, a spectro-temporal filterbank consisting of 41 Gabor filters has been designed by Schädler *et al.* (2012). This filterbank has been optimized for the task of ASR, and the respective real parts of the 41 Gabor filters is shown in Fig. 4.16.

The input is a log-compressed ratemap with a required resolution of 100 Hz, which cor-

responds to a step size of 10 ms. To reduce the correlation between individual Gabor features and to limit the dimensionality of the resulting Gabor feature space, a selection of representative ratemap frequency channels will be automatically performed for each Gabor filter (Schädler *et al.*, 2012). For instance, the reference implementation based on 23 frequency channels produces a 311 dimensional Gabor feature space.

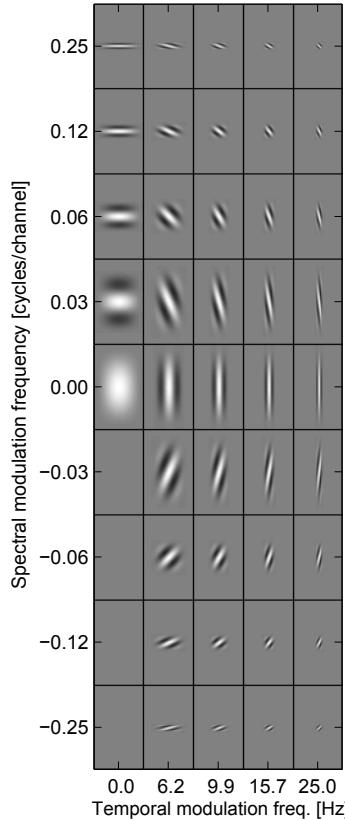


Figure 4.16: Real part of 41 spectro-temporal Gabor filters.

The Gabor feature processor is demonstrated by the script `DEMO_GaborFeatures.m`, which produces the two plots shown in Fig. 4.17. A log-compressed ratemap with 25 ms time frames and 23 frequency channels spaced between 124 and 3657 Hz is shown in the left panel for a speech signal. These ratemap parameters have been adjusted to meet the specifications as recommended in the European telecommunications standards institute (ETSI) standard (ETSI ES 201 108 v1.1.3, 2003). The corresponding Gabor feature space with 311 dimension is presented in the right panel, where vowel transition (e.g. at time frames around 0.2 s) are well captured. This aspect might be particularly relevant for the task of ASR.

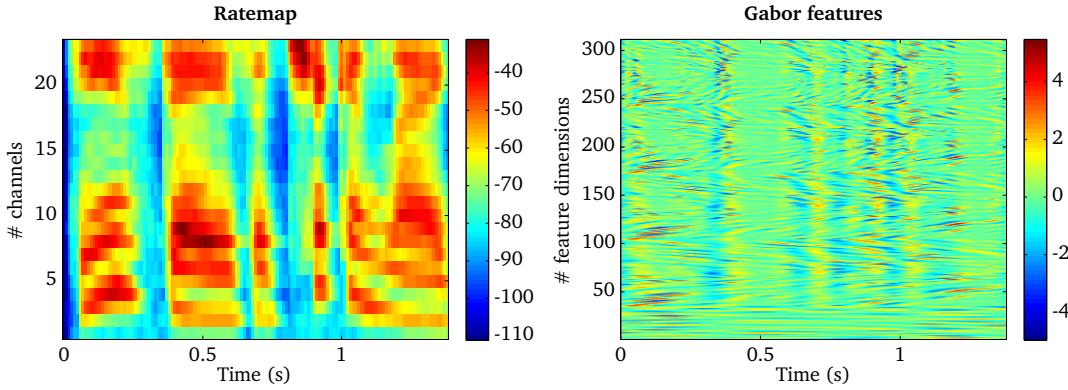


Figure 4.17: Ratemap representation of a speech signal (left panel) and the corresponding output of the Gabor feature processor (right panel).

4.14 Cross-correlation (crosscorrelationProc.m)

The IHC representations of the left and the right ear signals is used to compute the normalized cross-correlation function (CCF) in the FFT domain for short time frames of `cc_wSizeSec` duration with a step size of `cc_hSizeSec`. The CCF is normalized by the auto-correlation sequence at lag zero. This normalized CCF is then evaluated for time lags within `cc_maxDelaySec` (e.g., $[-1 \text{ ms}, 1 \text{ ms}]$) and is thus a three-dimensional function of time frame, frequency channel and lag time. An overview of all CCF parameters is given in Tab. 4.14. Note that the choice of these parameters will influence the computation of the ITD and the interaural coherence (IC) processors, which are described in Sec. 4.15 and Sec. 4.17, respectively.

Table 4.14: List of parameters related to the auditory representation 'crosscorrelation'.

Parameter	Default	Description
<code>cc_wname</code>	'hann'	Window type
<code>cc_wSizeSec</code>	0.02	Window duration (s)
<code>cc_hSizeSec</code>	0.01	Window step size (s)
<code>cc_maxDelaySec</code>	0.0011	Maximum delay (s) considered in CCF computation

The script `DEMO_Crosscorrelation.m` demonstrates the functionality of the CCF function and the resulting plots are shown in Fig. 4.18. The left panel shows the ear signals for a speech source that is located closer to the right ear. As result, the left ear signal is smaller in amplitude and is delayed in comparison to the right ear signal. The corresponding CCF is shown in the right panel for 32 auditory channels, where peaks are centered around positive time lags, indicating that the source is closer to the right ear. This is even more evident by looking at the summary cross-correlation function (SCCF), as shown in the

bottom right panel.

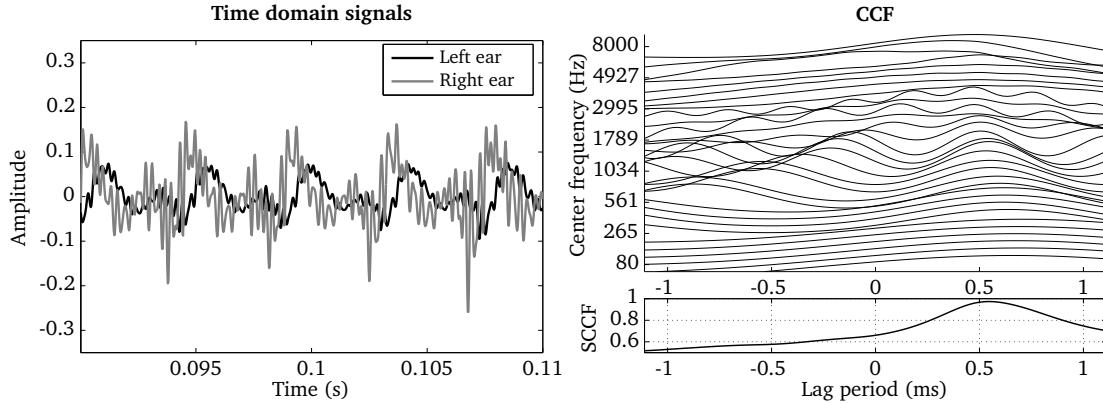


Figure 4.18: Left and right ear signals shown for one time frame of 20 ms duration (left panel) and the corresponding CCF (right panel). The SCCF summarizes the CCF across all auditory channels (bottom right panel).

4.15 Interaural time differences (itdProc.m)

The ITD between the left and the right ear signal is estimated for individual frequency channels and time frames by locating the time lag that corresponds to the most prominent peak in the normalized CCF. This estimation is further refined by a parabolic interpolation stage (May *et al.*, 2011, 2013). The ITD processor does not have any adjustable parameters, but it relies on the CCF described in Sec. 4.14 and its corresponding parameters (see Tab. 4.14). The ITD representation is computed by using the request entry 'itd'.

The ITD processor is demonstrated by the script DEMO_ITD.m, which produces two plots as shown in Fig. 4.19. The ear signals for a speech source that is located closer to the right ear are shown in the left panel. The corresponding ITD estimation is presented for each individual T-F unit (right panel). Apart from a few estimation errors, the estimated ITD between both ears is in the range of 0.5 ms for the majority of T-F units.

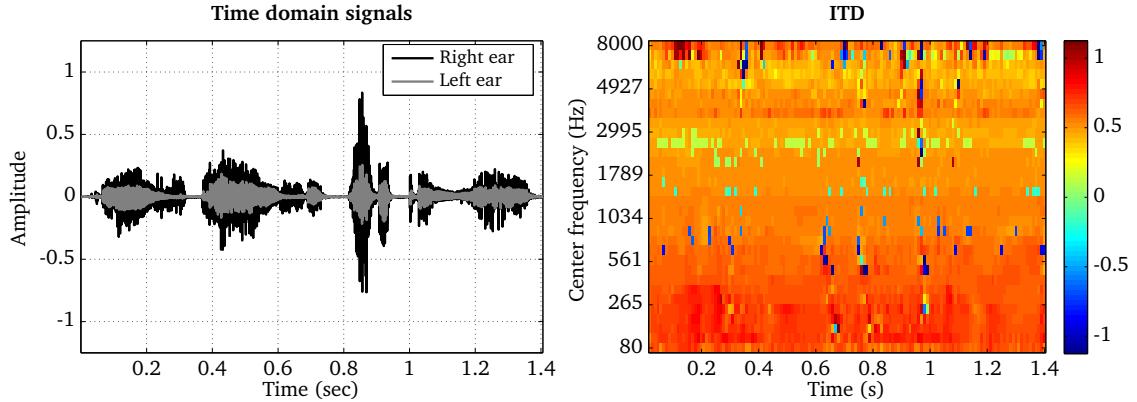


Figure 4.19: Binaural speech signal (left panel) and the estimated ITD in ms shown as a function of time frames and frequency channels.

4.16 Interaural level differences (ildProc.m)

The ILD is estimated for individual frequency channels by comparing the frame-based energy of the left and the right-ear IHC representations. The temporal resolution can be controlled by the frame size `ild_wSizeSec` and the step size `ild_hSizeSec`. Moreover, the window shape can be adjusted by the parameter `ild_wname`. The resulting ILD is expressed in dB and negative values indicate a sound source positioned at the left-hand side, whereas a positive ILD corresponds to a source located at the right-hand side. A full list of parameters is shown in Tab. 4.15.

Table 4.15: List of parameters related to the auditory representation 'ild'.

Parameter	Default	Description
<code>ild_wSizeSec</code>	'20E-3'	Window duration in s
<code>ild_hSizeSec</code>	10E-3	Window step size in s
<code>ild_wname</code>	'hann'	Window name

The ILD processor is demonstrated by the script `DEMO_ILD.m` and the resulting plots are presented in Fig. 4.20. The ear signals are shown for a speech source that is more closely located to the right ear (left panel). The corresponding ILD estimates are presented for individual T-F units. It is apparent that the ILDs change considerably as a function of the center frequency. Whereas hardly any ILDs are observed for low frequencies, a strong influence can be seen at higher frequencies where ILDs can be as high as 30 dB.

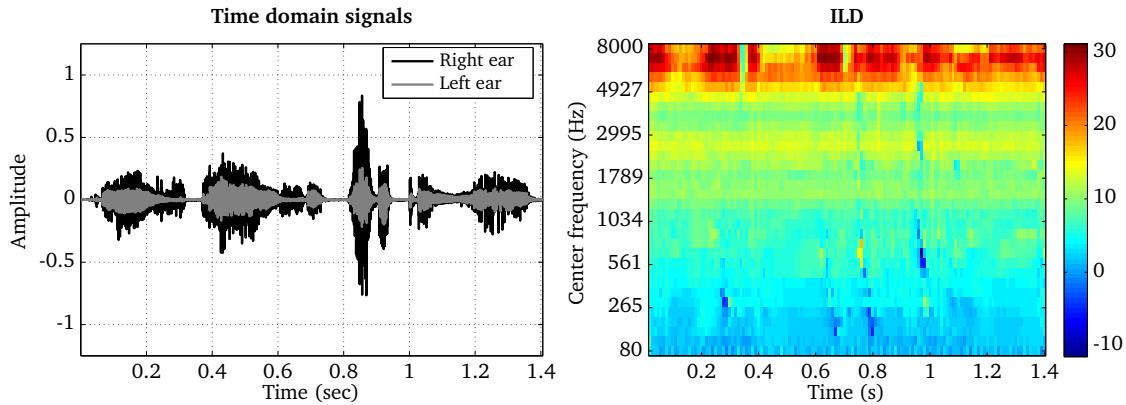


Figure 4.20: Binaural speech signal (left panel) and the estimated ILD in dB shown as a function of time frames and frequency channels.

4.17 Interaural coherence (icProc.m)

The IC is estimated by determining the maximum value of the normalized CCF. It has been suggested that the IC can be used to select T-F units where the binaural cues (ITDs and ILDs) are dominated by the direct sound of an individual sound source, and thus, are likely to reflect the true location of one of the active sources (Faller and Merimaa, 2004). The IC processor does not have any controllable parameters itself, but it depends on the settings of the CCF processor, which is described in Sec. 4.14. The IC representation is computed by using the request entry 'ic'.

The application of the IC processor is demonstrated by the script DEMO_IC, which produces the following four plots shown in Fig. 4.21. The top left and bottom left panels show the anechoic and reverberant speech signal, respectively. It can be seen that the time domain signal is smeared due to the influence of the reverberation. The IC for the anechoic signal is close to one for most of the individual T-F units, which indicates that the corresponding binaural cues are reliable. In contrast, the IC for the reverberant signal is substantially lower for many T-F units, suggesting that the corresponding binaural cues might be unreliable due to the impact of the reverberation.

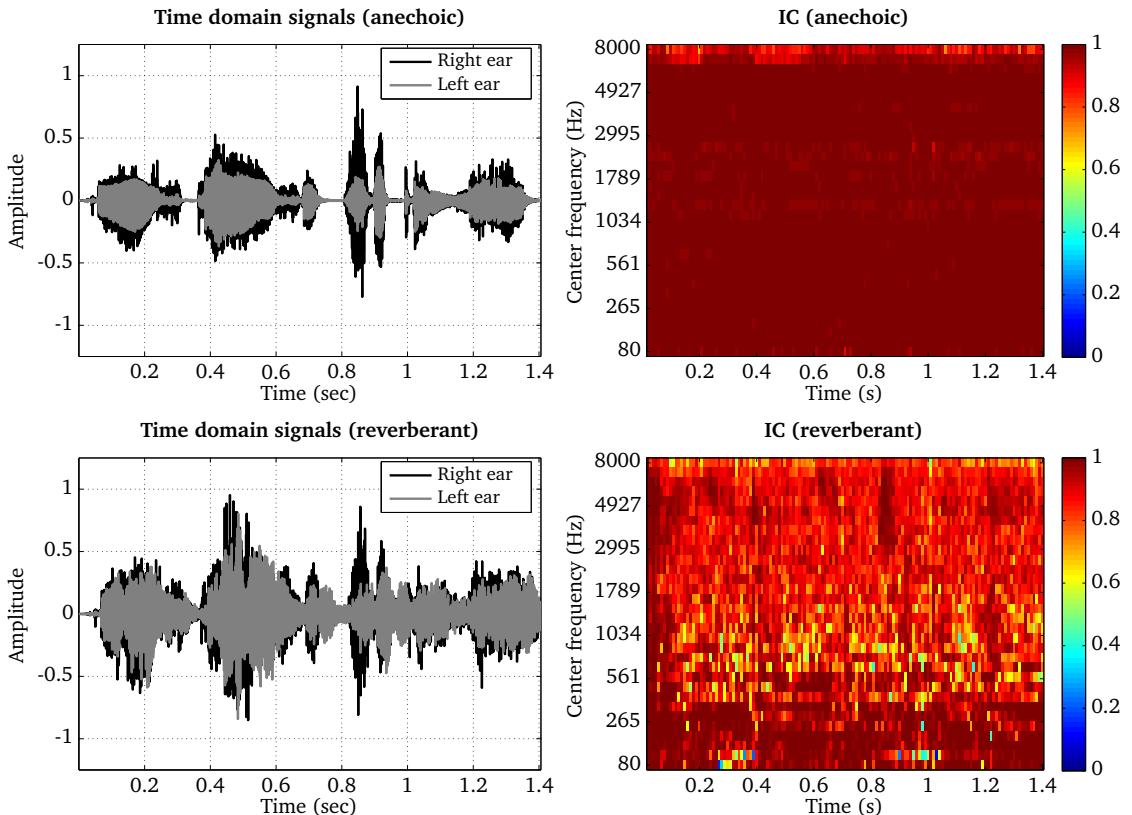


Figure 4.21: Time domain signals and the corresponding interaural coherence as a function of time frames and frequency channels estimated for a speech signal in anechoic and reverberant conditions. Anechoic speech (top left panel) and the corresponding IC (top right panel). Reverberant speech (bottom left panel) and the corresponding IC (bottom right panel).

A List of DEMO files

For the sake of reproducible research, source code to reproduce all the plots in the present document is provided as individual *demo* scripts. Table A.1 lists all the individual demo scripts, with a reference to the corresponding section in the report and a short description. These scripts can be found in the `/test` folder of the AFE software package.

Table A.1: List of demo scripts reproducing all the Matlab plots in this document.

Filename	Section	Description
<code>DEMO_Adaptation.m</code>	4.4	Adaptation loops
<code>DEMO_AMS.m</code>	4.12	Amplitude modulation spectrogram
<code>DEMO_Autocorrelation.m</code>	4.5	Autocorrelation
<code>DEMO_ChunkBased.m</code>	2.4	Chunk-based use of the AFE
<code>DEMO_Crosscorrelation.m</code>	4.14	Cross-correlation
<code>DEMO_DRNL.m</code>	4.2.2	Dual-resonance non-linear auditory filterbank
<code>DEMO_GaborFeatures.m</code>	4.13	Spectro-temporal modulation spectrogram
<code>DEMO_Gammatone.m</code>	4.2.1	Gammatone auditory filterbank
<code>DEMO_IC.m</code>	4.17	Interaural coherence
<code>DEMO_IHC.m</code>	4.3	Inner hair-cell modeling
<code>DEMOILD.m</code>	4.16	Interaural level difference
<code>DEMO_ITD.m</code>	4.15	Interaural time difference
<code>DEMO_OffsetStrength.m</code>	4.9	Offset strength detection
<code>DEMO_OnsetOffsetMaps.m</code>	4.10	Onset and offset mapping
<code>DEMO_OnsetStrength.m</code>	4.8	Onset strength detection
<code>DEMO_Pitch.m</code>	4.11	Pitch estimation
<code>DEMO_PreProcessing.m</code>	4.1	Pre-processing of input signal
<code>DEMO_Ratemap.m</code>	4.6	Ratemap extraction
<code>DEMO_SpectralFeatures.m</code>	4.7	Spectral features extraction

List of Acronyms

Acronyms

2D two-dimensional

ACF auto-correlation function

AFE auditory front-end

AGC automatic gain control

AMS amplitude modulation spectrogram

ASR automatic speech recognition

BM basilar membrane

CASA computational auditory scene analysis

CASP computational auditory signal-processing and perception

CCF cross-correlation function

DC direct current

DOW description of work

DRNL dual-resonance non-linear

ETSI European telecommunications standards institute

FFT fast Fourier transform

FIFO first in, first out

FIR finite impulse response

IC interaural coherence

List of Acronyms

IHC inner hair-cell

IIR infinite impulse response

ILD interaural level difference

ITD interaural time difference

MOC medial olivo-cochlear

MU model units

RMS root mean square

SACF summary auto-correlation function

SCCF summary cross-correlation function

SCM spectral crest measure

SFM spectral flatness measure

SPL sound pressure level

STRF spectro-temporal receptive field

T-F time-frequency

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