

Operational Amplifiers

Kevin "Yama" Keyser^{1, a)}

University of Missouri-Kansas City

In Laboratory 5, we focus on operational amplifiers. From general usage with a bipolar supply (and null measurements), voltage comparator, voltage follower, a follower with gain, inverting amplifier, integrator, differentiator, and logarithmic amplifier, we step through the many uses of operational amplifiers, and try to gain insight on their practical uses.

Keywords: Operational Amplifier

I. BACKGROUND

The operational amplifier is a configuration of resistors, capacitors, and transistors that has been abstracted away into a single component (See FIG. 1). This idea of higher level abstractions exist all over in electronic circuits. A computer programmer is not required to understand the inner workings of the machine he is developing on in order to write programs that run on it. In a similar sense, operational amplifiers are chips that simplify an electronic circuit, so we can focus on the fewest amount of inputs and outputs. This kind of abstraction makes it easier for anyone who needs the functionality that the operational amplifier offers without having to build one themselves.

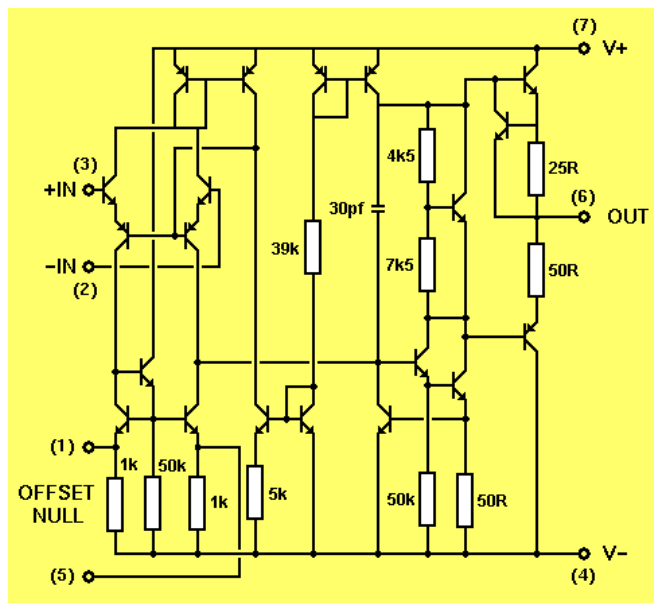


FIG. 1. An internal view of a 741 operational amplifier.

II. PROCEDURE

For each of the following sections, we will be using a +12V and -12V power supply on the VCC+ and VCC-

pins on our operational amplifiers. These are our voltage rails (our high rail, and our low rail, making this a bipolar supply. If we grounded the VCC- pin, we would have a unipolar supply, and our low rail would be 0V). We will also only be connecting to side one of our TL082 operational amplifier. With the TL082, which side you use determines what your high rail and your low rail is. For side one, our low rail is positive, and our high rail is negative. On side two, it is the opposite way around. All of our experiments, except for 5-2, will be using the TL082 operational amplifier. For 5-2, we will be using the LM311 operational amplifier.

All configurations and lab related images will be attached as the last page to the lab write up.

A. 5-1: Null Measurements

For the set up of this experiment, we will be connecting a +20V variable power supply to our inverting input pin, and see what this does to our output as we vary the voltage. We will then do the same with a -20V variable power supply to see what output we get as we change our input voltage. We will be doing our readings with a multimeter

B. 5-2: Voltage Comparator

This experiment requires us to change our operational amplifier to the LM311. We will connect the +20V variable power supply to our non-inverting input, and a +5V power supply to the inverting input pin. We will also use a 100k Ω resistor for both power supplies, and a 20pF capacitor in parallel before connecting to the inverting and non-inverting pins (see attached lab handout). We also have another +5V line with a 470 Ω resistor in parallel with the output pin that connects to our multimeter.

Afterward, we will be using a function generator to replace to +20V variable power supply on the non-inverting pin, with a $\pm 5V$ peak-to-peak, and set to output a triangle wave. Starting from 100kHz, we reduce the frequency until our oscilloscope readings become erratic for the output frequency. We then go back to 100kHz and increase the frequency until the readings on the oscilloscope becomes erratic again for the output frequency.

^{a)}Electronic mail: kk8r8@mail.umck.edu

C. 5-3: Voltage Follower

For the voltage follower, we will connect either the +20V or -20V variable power supply to the non-inverting input pin, and the output connects to the inverting input pin. We will then vary the voltage and monitor both input and output voltages.

D. 5-4: Follower with Gain

Similar to experiment 5-3, we will put a $10k\Omega$ resistor to ground, and connect that to the inverting input pin, and a resistor for our feedback voltage coming from our output before it connects to the inverting input as well. We will then vary the voltage to specific values to see what our the gain is for our input vs. our output. We will use a $100k\Omega$, $1M\Omega$, and $10M\Omega$ resistors for our feedback resistors, to see what this does to our gain (it has occurred to me in the writing of this lab report, that I read the $10k\Omega R_{in}$ as $100k\Omega$, and did not start my measurements at $10k\Omega$. This was my own fault for not reading the directions correctly.)

E. 5-5: Inverting Amplifier

We will keep the circuit design as we did in 5-4, but we will ground our non-inverting input pin, and will put our variable power supply on the inverting input pin. The resistor on our voltage input will be $10k\Omega$, and the feedback resistor will be $100k\Omega$. We will check our input vs. output at different voltages between $\pm 0.7V$. We will then repeat the experiment with our input resistor now at $100k\Omega$, and our feedback resistor at $10M\Omega$.

F. 5-6: Integrator

Keeping the general design of the 5-5 experiment, we will replace our feedback resistor with a capacitor. We will be looking for a time constant of milliseconds for our circuit (which is a simple formula of $Resistance \times Capacitance$, so we pick a resistor of $10k\Omega$, and a capacitor of $100pF$.) Our input voltage will be supplied by a function generator (with a $\pm 5V$ input). We will switch between sinusoidal, square, and triangle waves for our input and compare their shape to our output.

G. 5-7: Differentiator

The circuit design for this experiment requires us to switch our capacitor and resistor. We will then run the same experiment with the function generator as our voltage input and compare its shape to the shape of our output.

H. 5-8: Logarithmic Amplifier

Going back to experiment 5-6, we will replace the capacitor with a diode. Our input voltage goes back to the +20V variable power supply. We then vary our voltage and record our input voltage vs our output voltage.

III. PRESENTATION OF DATA

A. 5-1: Null Measurements

No tabular data. Will discuss experimental results in the Discussion section.

B. 5-2: Voltage Comparator

No tabular data. Will discuss experimental results in the Discussion section.

C. 5-3: Voltage Follower

No tabular data, but the general formula for our gain is estimated to come out to $G = \frac{V_{out}}{V_{in}} = 1$

D. 5-4: Follower with Gain

100kΩ		1MΩ		10MΩ	
V_{in}	V_{out}	V_{in}	V_{out}	V_{in}	V_{out}
0.176	0.344	0.184	1.760	0.192	11.40
0.408	0.808	0.408	4.360	0.408	11.40
0.528	1.080	0.480	5.040	0.480	11.40
1.040	2.000	1.020	10.80	1.040	11.40
2.080	4.080	1.960	11.40	2.040	11.40
5.000	9.800	5.200	11.40	5.100	11.40
Gain (Approximations)					
100kΩ		1MΩ		10MΩ	
2		11		101	
Offset					
100kΩ		1MΩ		10MΩ	
16.00mV		56.00mV		340.00mV	

E. 5-5: Inverting Amplifier

100k Ω		10M Ω	
V_{in}	V_{out}	V_{in}	V_{out}
0.088	-0.656	0.088	-6.600
0.416	-3.800	0.424	-10.80
0.688	-10.08	0.688	-10.80
-0.256	2.680	-0.264	11.40
-0.528	5.600	-0.504	11.40

F. 5-6: Integrator

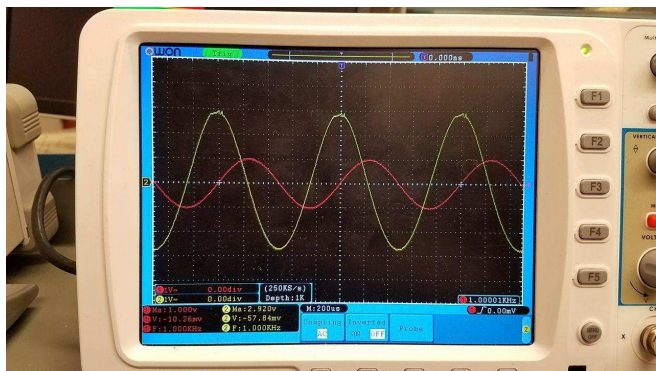


FIG. 2. Integrator circuit with sinusoidal input.

G. 5-7: Differentiator

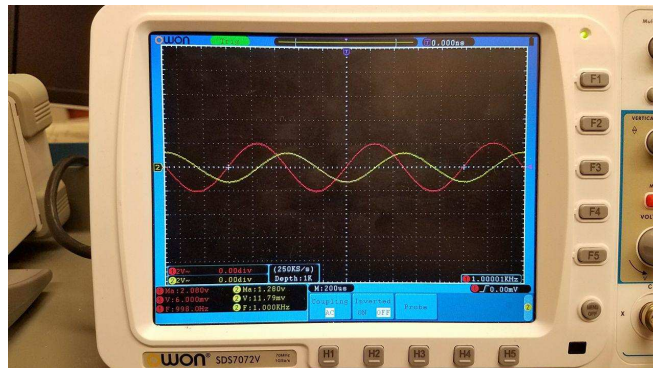


FIG. 5. Differentiator circuit with sinusoidal input.

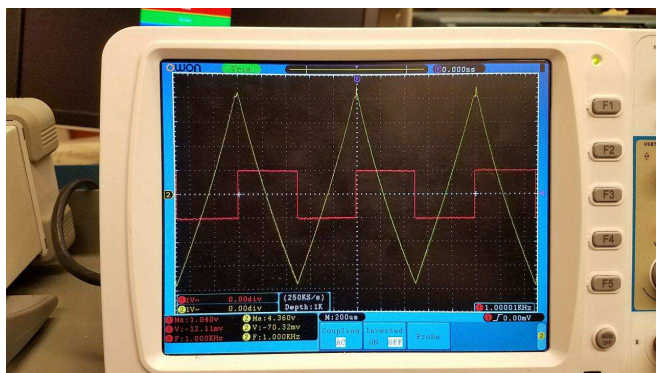


FIG. 3. Integrator circuit with square input.

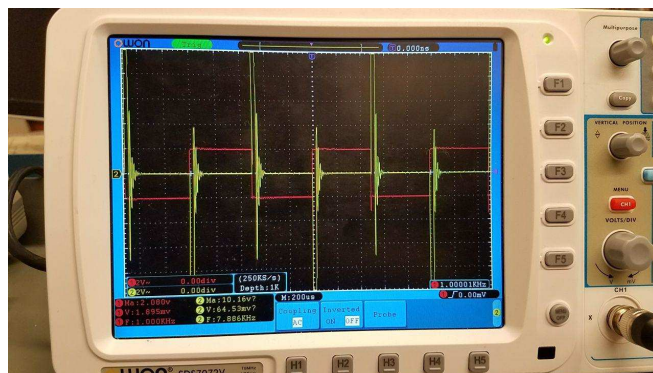


FIG. 6. Differentiator circuit with square input.

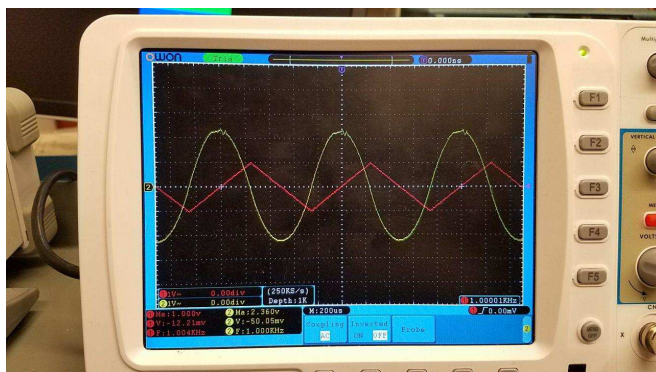


FIG. 4. Integrator circuit with triangle input.



FIG. 7. Differentiator circuit with triangle input.

H. 5-8: Logarithmic Amplifier

V_{in}	$\ln(V_{in})$	V_{out}
0.016	-4.135	-0.540
0.360	-1.022	-0.600
0.664	-0.409	-0.620
0.960	-0.041	-0.620
1.760	0.565	-0.657
5.000	1.609	-0.683
7.053	1.953	-0.693
9.903	2.293	-0.701
12.00	2.485	-0.706
Slope		y-intercept
-0.0249		-0.630

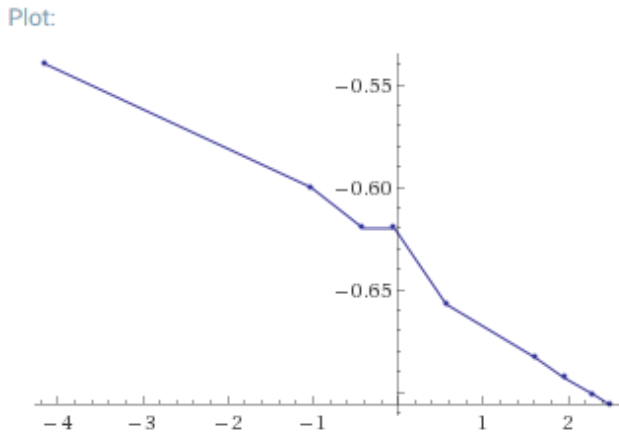


FIG. 8. plot of $\ln(V_{in})$ vs V_{out} .

IV. DISCUSSION

A. 5-1: Null Measurements

For experiment 5-1, there were a few things to notice. First is that depending on what side (side one or side two) of the TL082 we were using, we had a different high/low rail. Side one's low rail was positive. Since we were using the inverted input, if you have a negative variable voltage input, your output will always be positive at full rail. When the circuit is at zero input, you get the full low rail. When you put a negative voltage into the inverted input, it tries to invert it (to positive) and gives you the full low rail. So you don't see much of a change.

When using a positive variable voltage supply, it doesn't take much to make the circuit output the value of the negative rail at full. This makes sense when you consider later the behavior of the voltage comparator. We just need to get above a 0V value, and then this gives us our full rail. Because the lab equipment is not very sensitive, it was difficult to get an output that was less than 5mV. So the moment we touched the dial for the

variable voltage supply, we would get negative full rail right away.

B. 5-2: Voltage Comparator

The voltage comparator on the other hand has a lot more practical applications. This circuit gave us a cutoff voltage at which point we get the full value of our +5V input once we've reached a threshold. This value for our circuit and components came out to a +5.36V minimum to get an output of +5.045V. When the input voltage of our variable voltage supply was below this value, we ended up getting 0.225V as a base output.

As an aside, using a negative voltage input for our variable voltage supply on the non-inverting pin gave us nothing. Since the voltage never gets above +5.0V, the value never changed.

When we connected the input voltage to a function generator, we started to see inconsistencies with the output frequencies when we went below 39kHz and above 226kHz.

C. 5-3: Voltage Follower

Due to there being no resistance in the circuit (and it seems that the resistivity of the wires themselves had no substantial value), what we ended up with, is a circuit that followed our input voltage with its output voltage, until it hit the maximum of what your VCC+ input was. For us, this value was always 11.40V

D. 5-4: Follower with Gain

I stated in the Procedure section above that I failed to follow the directions of this lab, but I can at least comment on the results. We did get the gain values that we were expecting ($Gain = \frac{R_f}{R_{in}} + 1$). There was not much deviation from the expected values. A lot of the deviations we saw were due to the lack of stability in our voltage source, especially at low voltages. Anything below $\pm 1V$, the input would fluctuate a bit, and since there was gain involved (and output lag) this would cause small deviations in our output.

The percentage error of the offset voltage for the gain of 100 was negligible. In fact, since the output voltage was limited to 11.40V, if our input voltage was 0.2V (our lowest value of measure) our output voltage was stuck at 11.40V. This meant that for all measurements of the 100 gain circuit, our output was always 11.40V, due to the restrictions of our bipolar supply.

E. 5-5: Inverting Amplifier

The expected output for gain on the Inverting Amplifier is a little bit more simple ($Gain = -\frac{R_f}{R_{in}}$). Our error rate for the 100k Ω was roughly 0.8%. This difference was higher when we went to lower voltages (on the order of 10 millivolts). For the 10M Ω circuit, it was hard to gauge the error rate, as anything above 100mV would give us our full rail value (-10.80V to 11.40V).

F. 5-6: Integrator

When you integrate a linear function, you get a quadratic function, so the output looks almost sinusoidal. One thing that was noted as well was the lag that gave a phase angle to the output. Looking at FIG. 2, FIG. 3, and FIG. 4, we saw a half wave length lag which is easiest to notice with the square wave. When the values are positive, the output should be increasing, but it is decreasing. This leads me to believe that the lag in the circuit gives a phase angle of $-\pi$.

G. 5-7: Differentiator

Much like the integrator circuit, we also see a phase angle of $-\pi$. What's interesting is not the sinusoidal waves, but the triangular and square waves. These functions are not differentiable. You can not take the derivative of a vertical line, or for a corner. What we see here in these circuits is almost a discontinuity with the outputs. For the square wave, we got a huge spike where the square wave changes. The capacitor acts as a spring, or a restoring force in these situations, which is why we saw it takes a moment to bounce back to a stable value. The triangle

wave does this as well, but the magnitude of this spike is far less than the square wave.

H. 5-8: Logarithmic Amplifier

The logarithmic amplifier does exactly what it says. It amplifies the circuit using a natural log. What we see when we use different input values, and look at the output values, is we get a function that is linear when we plot the natural log of the input versus the output. The equation we get for this is $y = -0.0249x - 0.630$.

V. CONCLUSION

The operational amplifier is a versatile, abstracted circuit that lets you harness the power of many mathematical functions simply based on components added to the circuit, and changing our inputs. From a minimum value comparator, to an amplifier, negative amplifier, integrator, differentiator, and natural log amplifier, we have many mathematical functions we can use to meet our needs. When we consider the power behind this concept (a simple circuit design letting us get the area underneath the curve of our input, or letting us see the rate of change of our input), we begin to realize how powerful this small, higher level abstraction chip is.

The biggest drawback to the lab is the limitations of our input devices (the Elenco variable power supply is not sensitive enough to really give us accurate readings when we're in the millivolts range). Beyond that, is our VCC+ and VCC- being low for our 100 Gain circuits. Almost any input you have on these circuits, you're limited to the full rail values of the VCC+ and VCC-, so the different measurements become a moot point.