

3TR4: Communication Systems

Lab 1

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Design and Simulation

Important Information given from the appendix

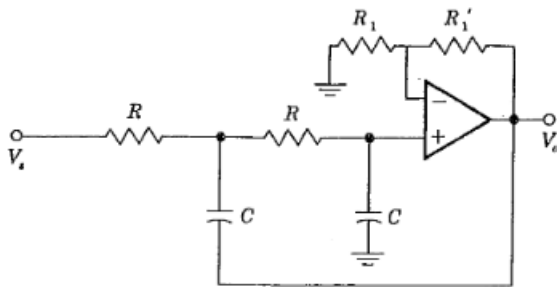


Figure 1. Second-order low-pass filter

TABLE 16-1 Normalized Butterworth polynomials

n	Factors of polynomial $P_n(s)$
1	$(s + 1)$
2	$(s^2 + 1.414s + 1)$

$$\frac{1}{(s/\omega_o)^2 + 2k(s/\omega_o) + 1}$$

Since we chose to create a 2nd order filter we use this equation $= \frac{1}{s^2 + 1.414s + 1} = \frac{1}{s^2 + 2ks + 1}$ From this. We can conclude that $2k = 1.414$. We assume $\omega_o = 1 \frac{\text{rad}}{s}$ which is why it is removed from the equation.

$$A_{Vo} = 3 - 2k$$

$$A_{Vo} = 3 - 1.414 = 1.586$$

We can use the equation below to calculate the transfer function of the filter.

$$A_V(s) = \frac{V_o}{V_i} = \frac{A_{Vo}Z_3Z_4}{Z_3(Z_1 + Z_2 + Z_3) + Z_1Z_2 + Z_1Z_4(1 - A_{Vo})} = A_{Vo} \frac{(1/RC)^2}{s^2 + \left(\frac{3 - A_{Vo}}{RC}\right)s + \left(\frac{1}{RC}\right)^2}$$

The following equation is the transfer function of the filter. This is important to know because it can help us solve the filter frequency:

$$H(s) = 1.586 - \frac{\left(\frac{1}{RC}\right)^2}{s^2 + \frac{1.414}{RC}s + \left(\frac{1}{RC}\right)^2}$$

To calculate the cut-off frequency using a resistor of $1K\Omega$ and a capacitor of $0.1\mu F$:

$$F_c = \frac{1}{2\pi RC} = \frac{1}{2\pi(1)(0.1)} \approx 1.59 \approx 1.6kHz$$

Both capacitors must be the same value and both must have a value of less than $1\mu F$ in order to work for a second-order low pass Butterworth filter.

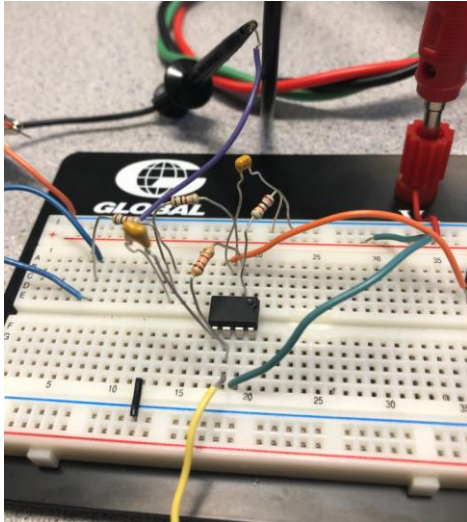


Figure 2. Physical Circuit used in-lab

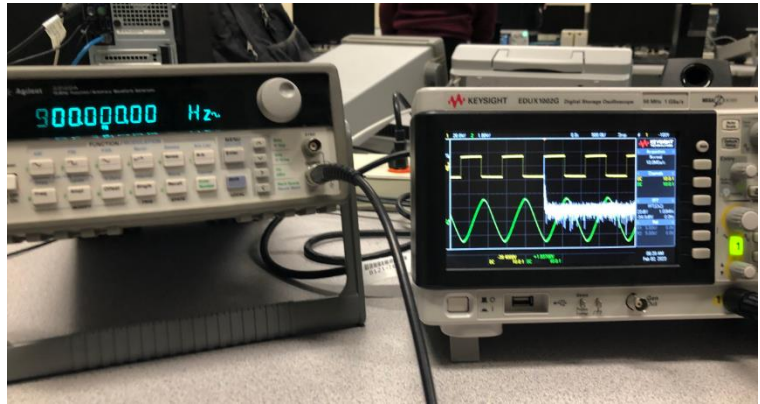
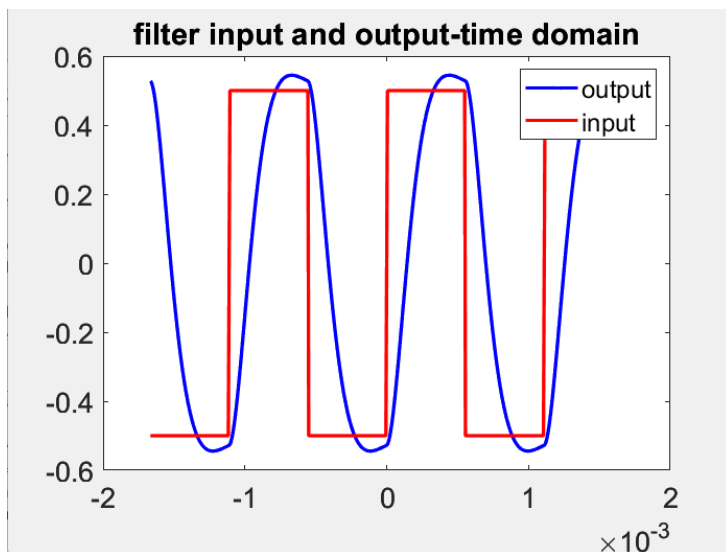


Figure 3. Oscilloscope from in-lab testing



We can compare the MATLAB plot with the experimental plot above. Both plots have similarities when looking at the shapes of both waves. We see a square wave and a sinusoidal wave in both the MATLAB and experimental plots.

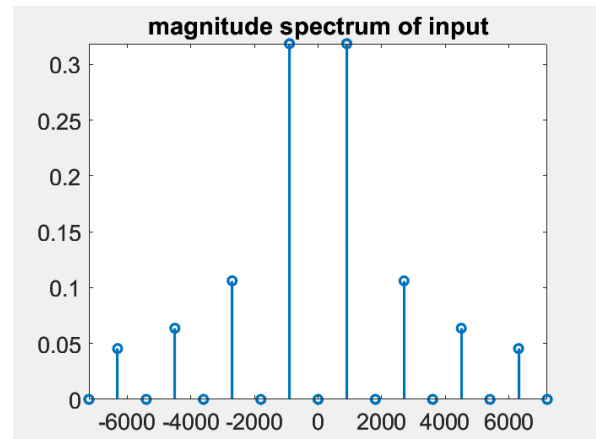
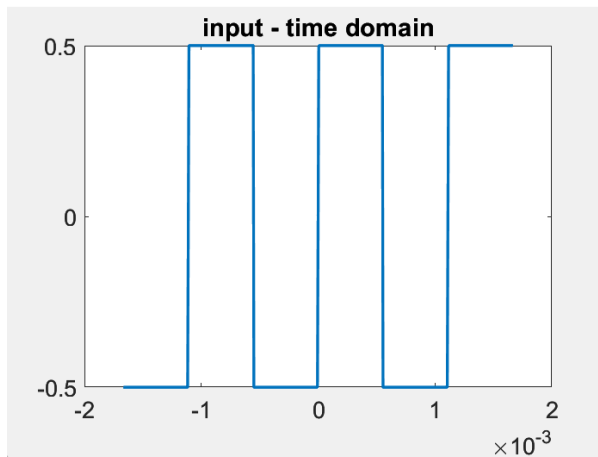
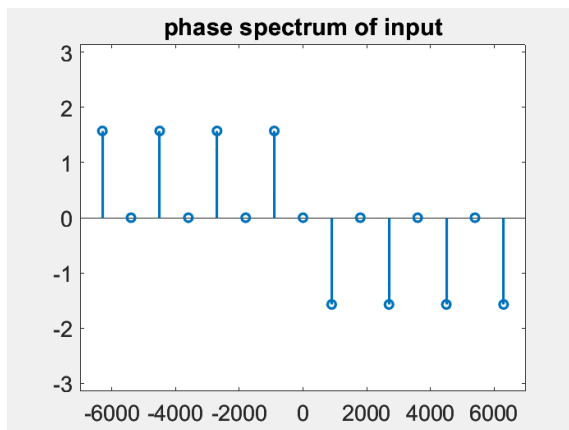


Figure 3. Matlab input with a frequency of 900Hz



Part 1: Butterworth Filter Experiments

We can analyze the details on the testing of the filter by applying a square wave input. Looking at the given notes, we can see that the Fourier series representation is given as

$$C_n = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} g(t) e^{-j2\pi n f_0 t} dt$$

$$C_n = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^0 (-1) e^{-j2\pi n f_0 t} dt + \int_0^{\frac{T_0}{2}} (1) e^{-j2\pi n f_0 t} dt$$

$$C_n = \frac{1}{jn\pi} (1 - \cos n\pi)$$

These values are only important when:

$$C_n = \begin{cases} \frac{2}{jn\pi} & n = \text{odd} \\ 0 & \text{otherwise} \end{cases}$$

Using the information above, we can find $\frac{C_3}{C_1}$

$$= \frac{\left(\frac{1}{j3\pi}\right)}{\frac{1}{jn}} = -10\text{dB (before)}$$

However, the information given in the appendix gives us a better representation of the changes between different frequency components.

Butterworth Filter⁶ A common approximation of Eq. (16-17) uses the Butterworth polynomials $B_n(s)$, where

$$A_V(s) = \frac{A_{Vo}}{B_n(s)} \quad (16-18)$$

and with $s = j\omega$,

$$|A_V(s)|^2 = |A_V(s)| |A_V(-s)| = \frac{A_{Vo}^2}{1 + (\omega/\omega_c)^{2n}} \quad (16-19)$$

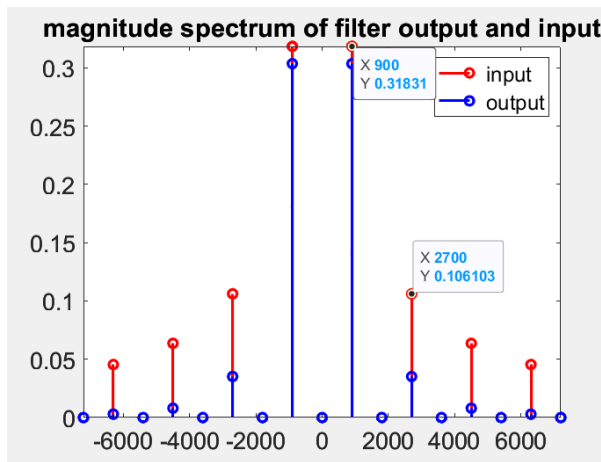
From Eqs. (16-18) and (16-19) we note that the magnitude of $B_n(\omega)$ is given by

$$|B_n(\omega)| = \sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^{2n}} \quad (16-20)$$

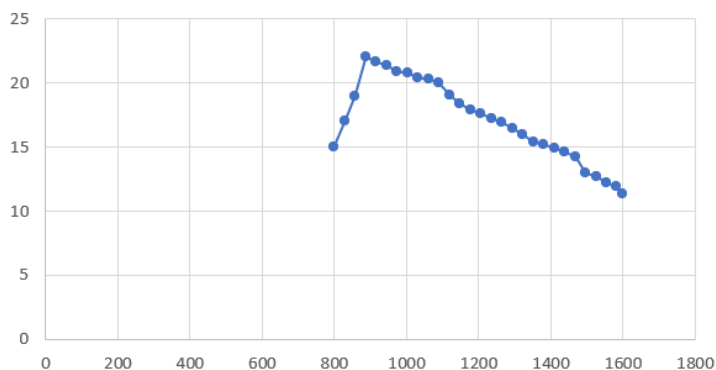
Using these equations above, we can try out the same ratio above. $N = 2$ as we are in the 2nd order. Also, we consider the input frequency at $\frac{1}{2}$ of the cut off frequency as shown earlier.

$$\left(\frac{1}{3}\right) \left(\frac{\left(\frac{A_v^2}{\sqrt{1 + \left(\frac{3}{2}\right)^4}} \right)}{\left(\frac{A_v^2}{\sqrt{1 + \left(\frac{1}{2}\right)^4}} \right)} \right) = \left(\frac{1}{3}\right) \left(\frac{\sqrt{1 + \left(\frac{1}{2}\right)^4}}{\sqrt{1 + \left(\frac{3}{2}\right)^4}} \right) \approx 0.1395$$

$20 \log(0.1395) = -17.1dB$ (after)



The graph above is the most important graph as we can see the input and output magnitudes. As we can see, the 3rd harmonic is 0.106103 and the 1st harmonic is 0.31831. The ratio between these two values are $\frac{C_3}{C_1} = \frac{0.106103}{0.31831} \approx 0.33$. When looking at the 3rd harmonic. We can do the same equation as earlier to compare those values. $20 \log(0.106103) = -19.485dB$. The value above was -17.1dB



This graph shows the shape that resembles the low-pass filter. There are some weird data points, however, that is expected as there will always be some type of error in labs if they are not virtual. These data points were found by changing the frequency from 800Hz to our cut off (1.6kHz) and seeing the peak of the graph.

Part 2: Audio Signal Filtering

Low-Pass Filter:

The low pass filter sounds quieter than the original sound while also giving a muffled sound. It is still audible, but it is close to being hard to understand what is being said in this audio.

Band-Pass Filter:

The band pass filter sounds quieter than the original sound while also giving a muffled sound. It is harder to understand than the low-pass filter, but it is still audible. I realized that the sound also changes. At first it is more muffled, but it becomes less muffled and easier to understand near the end.

High-Pass Filter:

The high-pass filter sounds a bit louder than the band-pass and low-pass filter however it is much harder to understand. There is a very faint buzzing sound which make it hard to understand If I didn't the original sound. I would not understand what the words said other than the "it's easy to". The sentence "tell the depth of the whale" is very hard to understand.

Notch Filter:

The original sound had a very high tone on top of the audio. The notch filter does a very good job in removing that high tone pitch and keeping the words sounding very clear. However, I hear a weird bubbling effect like the sound is going into a cup.

Matlab

```
1 %% Square wave generator
2 clc
3 clear all
4 hold off
5
6 f0=900; %Fundamental freq of input triangular wave. |
7 T0 = 1/f0; %period
8 tstep = 0.005*T0;
9 no_sample = 3*T0/tstep + 1; %no. of samples within 3*T0
10 no_sample1 = T0/tstep + 1; %no. of samples within T0
11 tt = -1.5*T0:tstep:1.5*T0;
12
13 tt1 = -0.5*T0:tstep:0.5*T0; % time vector for the period -0.5T0 to 0.5T0
14 gp1 = 0.5*sign(2*pi*f0*tt1);
15 gp_in = [gp1 gp1(2:no_sample1-1) gp1]; %3 cycles of the triangular wave
16 figure(1)
17 Hp1 = plot(tt, gp_in);
18 set(Hp1, 'LineWidth', 2)
19 Ha = gca;
20 set(Ha, 'FontSize', 16)
21 title('input - time domain')
22 pause
23
24 %% Fourier series representation of signal (Amplitude Spectrum)
25 K=1/(2);
26 N=100; %no. of harmonics
27 nvec = -N:N;
28 c_in = zeros(size(nvec));
29 for n = nvec
30     m = n+N+1;
31     %% This equation was found during in lab
32     c_in(m) = 0.5*(1-cos(n*pi))/(1i*n*pi);
33
34     if (n == 0)
35         c_in(m) = 0.0;
36     end
37 end
38
39 f = nvec*f0; %frequency vector
40 figure(2)
41 Hp1=stem(f, abs(c_in));
42 axis([-8*f0 8*f0 0 max(abs(c_in))])
43 set(Hp1, 'LineWidth', 2)
44 Ha = gca;
45 set(Ha, 'FontSize', 16)
46 title('magnitude spectrum of input')
47 pause
48
49 %% Fourier series representation of signal (Phase Spectrum)
50 figure(3)
51 Hp1=stem(f, angle(c_in));
52 set(Hp1, 'LineWidth', 2)
53 Ha = gca;
54 set(Ha, 'FontSize', 16)
55 axis([-0.7e4 0.7e4 -pi pi])
56 title('phase spectrum of input')
57 pause
58
59 %% Designing the 2nd order Butterworth filter
60
61 %R=1e3;
62 %C=0.1e-6;
63 %fc=1/(2*pi*R*C) %cutoff freq of filter
64 fc = 1600;
65
66 Hf = 1 ./ (1 + 1.414*(1i*f/fc) + (1i*f/fc).^2); %filter transfer function
67 c_out = c_in .* Hf; %Fourier coefficients of the filter output
68
69 figure(4)
70 stem(f, abs(c_in), 'r', 'LineWidth', 2);
71 hold on
72 stem(f, abs(c_out), 'b', 'LineWidth', 2);
```



```

73 hold off
74 axis([-8*f0 8*f0 0 max(abs(c_in))])
75 Ha = gca;
76 set(Ha,'FontSize',16)
77 title('magnitude spectrum of filter output and input')
78 Ha = gca;
79 set(Ha,'FontSize',16)
80 legend('input','output')
81 pause
82
83 % hold off
84 % Hp1=plot(f,angle(c_out))
85 % set(Hp1,'LineWidth',2)
86 % Ha = gca;
87 % set(Ha,'FontSize',16)
88 % title('phase spectrum of output')
89 % axis([-0.1e4 0.1e4 -pi pi])
90 % pause
91 % hold on
92 % Hp1=plot(f,angle(c_in),'r')
93 % set(Hp1,'LineWidth',2)
94 % Ha = gca;
95 % set(Ha,'FontSize',16)
96 % pause
97 % hold off
98
99 %% Construct the output signal from the Cout Fourier coefficients
100
101 A = zeros(2*N+1,ceil(no_sample));
102 for n = nvec
103     m=n+N+1;
104     A(m,:) = c_out(m) .* exp(1i*2*pi*n*f0*tt);
105 end
106 gp_out = sum(A);
107 figure(5)
108
109 Hp1 = plot(tt,real(gp_out),'b',tt,gp_in,'r');
110 set(Hp1,'LineWidth',2)
111 Ha = gca;
112 set(Ha,'FontSize',16)
113 title('filter input and output-time domain')
114 set(Ha,'FontSize',16)
115 legend('output','input')

```