# Toward single particle reconstruction without particle picking

Tamir Bendory, Nicolas Boumal, William Leeb and Amit Singer
May 15, 2018

#### Abstract

Here comes the abstract

#### 1 Introduction

Single particle reconstruction (SPR) using cryo-electron microscopy (cryo-EM) is an innovative technology for reconstructing the 3-D structure of macromolecules. In recent years, structures of many molecules, previously regarded as insurmountable, are now being reconstructed to near-atomic resolution; see for instance [34, 10, 51]. This technological advancement was recognized by the 2017 Nobel Prize in Chemistry [1].

In a cryo-EM experiment, multiple biological samples of the (ideally) same molecule are rapidly frozen in a thin layer of vitreous ice. Within the ice, the molecules are randomly oriented and positioned. Then, the microscope produces a 2D tomographic projection image, called a *micrograph*, of the multiple samples embedded in the ice. The micrograph is dominated by noise due to the small electron doses that can be applied to the specimen without causing radiation damage. The cryo-EM problem is to estimate the structure of the molecule from the micrograph (or, typically, several micrographs).

Particle picking. All contemporary methods in the field split the reconstruction procedure to several subroutines. The first step in the pipeline is the so-called particle picking, in which one aims to detect the 2-D tomographic projections of the samples from the noisy micrograph. The output of ideal particle picker is a series of 2-D images, each containing one centered tomographic projection associated with an unknown viewing direction and an in-plane rotation. This series of images is then used to estimate the structure.

Many automatic and semi-automatic methods for particle picking have been proposed, based on edge detection, template matching and deep learning; see for instance [26, 44, 55, 23, 47, 29]. However, most of these procedures are prone to model bias. For instance, in the popular framework of RELION [47], the user manually marks hundreds of spots on the micrograph, believed to contain projections. Therefore, the algorithm's performance depends on the prior assumptions of the users about the particle's structure; the same holds true for deep learning based approaches which require constructing labeled sets of data. Other methods use disks or differences of Gaussians as templates [35, 53]. Nowadays, it also still popular to pick particles manually. This method, while it exploits the researcher's experience, is both tedious and subject to model bias.

The significance of the model bias in cryo-EM was stressed in [49, 30] and demonstrated by the "Einstein from noise" example. In this experiment, the image of Einstein is correlated with multiple images of pure noise. The noisy images are aligned to Einstein's image using cross-correlation and then averaged. In Figure 1 we conducted this experiment with different number of noise images. With merely 1000 noise images, Einstein's face is clearly observed in the averaged image. With 10,000 image, the output is remarkably similar to the template image of Einstein rather than to pure noise. Crucially, the more noisy images we use, the more severe the model bias towards the template. In the context of cryo-EM, this example demonstrates how our prior assumptions about the particle may influence the reconstructed structure. This true in general to any method which is based on template matching.

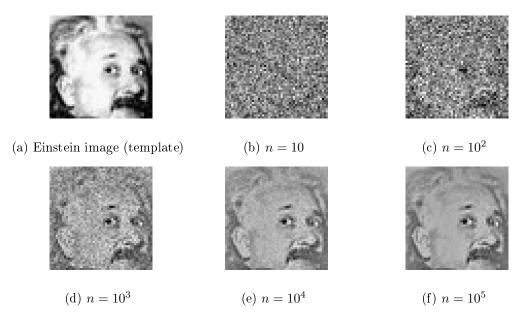


Figure 1: Einstein from noise experiment. As a template, we used Einstein's image of size  $50 \times 50$  pixels. We drawn n images of the same size whose entries are composed of i.i.d. Gaussian entries with the same mean and variance as Einstein's image. Each noise image was aligned with respect to the template by the maximal value of their cross correlation. After the alignment, all the images were averaged. As can be seen, the more pure noise images we use, the closer the averaged image to the template.

Besides the model bias, particle pickers have several limitations due to the high noise level, in particular for small particles and in low contrast. As a result, the projections in the 2D images are typically not centered; including the correct centers as parameters dramatically increases the complexity of reconstruction algorithms.

Simplified model for SPR In this paper, we present a preliminary study that aims to show that estimating a particle directly from the micrograph, without identifying the location of individual particle images in the micrograph, might be possible.

In particular, we consider the toy problem of estimating a set of signals  $x_1, \ldots, x_K$  from their multiple occurrences in unknown locations in a noisy data sequence y. Here, y plays the

role of the micrograph and the K signals can be thought of as K different viewing directions of the particle, each of which appears multiple times in the micrograph. A precise mathematical formulation of the model and the estimation problem is provided in Section 4. To be clear, we do not consider here many prominent features of real SPR experiments and do not aim to reconstruct any 3-D structure; instead, we solve a simpler problem that we believe captures key elements of the true SPR problem. We also mention in passing that similar models emerge in different scientific fields, such as spike sorting [37], passive radar [25] and system identification [43].

Figure 2 shows a simple example of our problem with one signal (K=1). Each panel demonstrates an image of size  $250 \times 250$ , containing 4 occurrences of an Einstein image of size  $50 \times 50$  presented in Figure 1a. The left panel shows the data without noise. In the middle panel, i.i.d. Gaussian noise with standard deviation  $\sigma=0.5$  was added, however, Einstein is still easily identified. In this scenario it is likely that standard detection algorithms can locate the copies of the signal within the micrograph. In the right panel, we show the same data swamped in i.i.d. Gaussian noise with standard deviation  $\sigma=3$ . Clearly, identifying the locations of Einstein in this micrograph is much more challenging. In fact, it can be shown that in the low signal-to-noise (SNR) regime, detection of individual image occurrences is impossible—even if the true image is known—and therefore particle picking is impossible. This phenomenon is explained in detail in Appendix C.

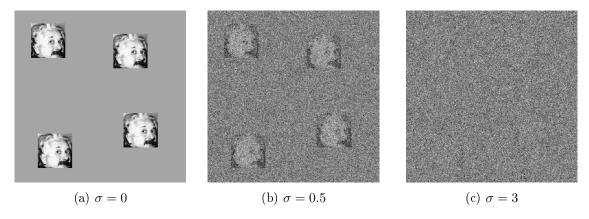


Figure 2: Example of data sequences (micrographs) of size  $250 \times 250$  with additive i.i.d. Gaussian noise with variance  $\sigma^2$ . Each micrograph contains four occurrences of a  $50 \times 50$  image of Einstein; see Figure 1a.

Autocorrelation analysis. We now return to the general problem of estimating  $x_1, \ldots, x_K$  from the noisy observation y. In order to recover the signals  $x_1, \ldots, x_K$  from the data, we use autocorrelation analysis. In a nutshell, the method consists of two stages. First, we estimate a mixture (i.e., linear combination) of the low-order autocorrelation functions of the signals from the data. These quantities can be estimated, to any desired accuracy, if individual occurrences are separated by the maximum signal size and each signal appears sufficiently many times in the data. There is no need to detect individual occurrences. In the second stage, the signals are estimated from the mixed autocorrelations using a nonconvex least-squares (LS) algorithm or a phase retrieval algorithm; see Section 4 for details. This method requires only one pass

over the data and can recover the signals in any SNR, if the signals appear enough times in the data. As a side note, we mention that expectation-maximization—a popular framework in SPR—is intractable for this problem (see Appendix C for more details).

#### 2 Results

In this section we describe some numerical experiments and show results. The experimental details and the algorithms are discussed at length in Section 4.

In the first experiment, we estimated Einstein's image of size  $50 \times 50$  from multiple micrographs of size  $4096 \times 4096$  pixels, contaminated with additive i.i.d. Gaussian noise with standard deviation  $\sigma = 3$  (SNR = 1/20). An illustrative example of a micrograph appears in the right panel of Figure 2. Each micrograph contains multiple occurrences of Einstein's image in random locations, while keeping a separation of 49 pixels between the images in each direction. From the micrographs, we estimated the signal's power spectrum, which is equivalent to the second-order autocorrelation. Then, in order to estimate the signal itself, we applied a standard phase retrieval algorithm called relaxed-reflect-reflect (RRR).

As an initial guess, we picked an image of the physicist Issac Newton; see Figure 3a. If the algorithm was prone to model bias, we would expect to get as an output an image that resembles Newton, similarly to the "Einstein from noise" effect of Figure 1. However, the experiment exhibits the desired phenomenon: the more data we collect, the better the reconstruction quality.

Figure 3 demonstrates several recovery results for different number of recorded micrographs. Figure 4 presents the normalized recovery error and the power spectrum estimation error as a function of the number of micrographs. To measure the error, we use the root mean square error (RMSE) defined as

RMSE := 
$$\frac{\|x - \hat{x}\|_{F}}{\|x\|_{F}}$$
, (2.1)

where x and  $\hat{x}$  are, respectively, the underlying and estimated image, and  $\|\cdot\|_{F}$  stands for the Frobenius norm. As expected, the RMSE of estimating the power spectrum decreases linearly with slope -1/2 in logarithmic scale. [Few comments: 1. We have a movie in the supplementary material. 2. Does the error also have the right slope? 3. We can put more images to exemplify the progress 4. Need to improve the micrograph's generation code ]

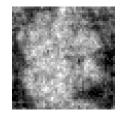
Results from a similar experiment with three (K=3) one-dimensional signals are presented in Figure 5. We define the ratio of the space occupied by the *i*th signal as

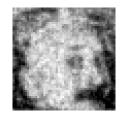
$$\gamma_i = \frac{M_i L}{N},\tag{2.2}$$

where  $M_i$  is the number of signal's occurrences, L is the length of the signal and N is the micrograph length. These ratios present the density of the signals in the data. In the experiment, we do not assume to know these ratios, neither the noise level  $\sigma$ .

In order to estimate the signals, we computed the first three autocorrelation functions of the data and then estimated the signals and their corresponding  $\gamma_i$  using a nonconvex least-squares. As can be seen, given enough signal occurrences, we can estimate accurately the signals. The estimation quality of the triangle signal is poorer than the other two signals,









(a) Newton (model)

(b) P = 1024

(c)  $P = 1024 \times 10$ 

(d)  $P = 1024 \times 100$ 

Figure 3: Recovery of Einstein's image. The RRR algorithm is initialized with the image of the physicist Issac Newton in the left panel. We show estimations of Einstein using  $P=1024,1024\times10$  and  $1024\times100$  number of micrographs, each contains 700 image occurrences on average and the SNR is 1/20. The signal's estimation RMSE is 0.5514, 0.3494 and 0.2112, respectively. The power spectrum estimation RMSE is, respectively, 0.0269, 0.0083 and 0.0027. [Can add more images to exemplify the progress]

a phenomenon that is explained using Proposition C.1. [1. We will replace the figure with a "progress figure" (like we did for Einstien) for each signals and a plot of the recovery error for all three signals. 2. We may want to replace the triangle signal]

#### 3 Discussion

All current algorithmic pipeline for SPR using cryo-EM start with a particle picking algorithm which is prone to model bias. Bypassing the particle picking stage and constructing a 3-D model directly from the data—without assuming prior knowledge on the particle to be estimated—can be used to reconstruct ab initio models to initialize a refinement algorithm. Alternatively, it can be applied to generate templates for a particle picker which does not suffer from model bias.

In this paper, we examined a simplified model that, we believe, captures important features of the SPR problem. While the field is currently dominated by Bayesian methods such as EM, they are intractable for such problems. As an alternative, we propose to use autocorrelation analysis technique that shares some common lines (did you get the wordplay?) with Kam's method for ab inito modeling [32, 36, 50]. That being said, the SPR model is far more complicated than the model presented here. In a future research, we hope to bridge this gap.

Our results rely on two core assumptions that are not necessarily met by an SPR experiment. First, we modeled the background information as i.i.d. additive noise. In practice, the background information may be structured or signal dependent. Second, we assumed that the signal occurrences are all separated by the length of the signal, see (4.2). This separation can be induced by careful experimental design [???]. If the signals are not separated, one can introduce a new variable that represents the distribution of the spacing between signal occurrences and then compute explicitly the relation between the autocorrelation functions of the data and the signals. It is yet to be studied under what conditions on this distribution one can estimate the signals.

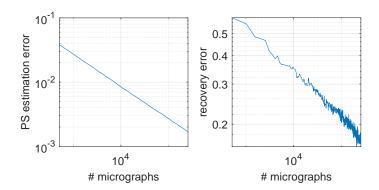


Figure 4: The left panel shows the RMSE of estimating the power spectrum of the signal from the data as a function of the number of collected micrographs. As expected, the curve is linear with slope -1/2 in logarithmic scale. The right panel shows the RMSE (in logarithmic scale) of recovering Einstein's image. [The program still runs so these figures will be extended to the right (more micrographs).]

#### 4 Methods

**Mathematical model** Let  $x_1, \ldots, x_K \in \mathbb{R}^L$  be the sought signals and let  $y \in \mathbb{R}^N$  be the data. The forward model can be posed as a mixture of *blind deconvolution* problems between binary signals and the target signals  $x_i$ :

$$y = \sum_{i=1}^{K} x_i * s_i + \varepsilon, \quad \varepsilon \sim \mathcal{N}(0, \sigma^2 I).$$
 (4.1)

The nonzero values of each  $s_i \in \{0,1\}^N$  determine the position of  $x_i$ 's occurrences. We denote the set of these nonzero values by  $\mathcal{S}_i$  and its cardinality by  $|\mathcal{S}_i| = M_i$ . By assuming that all  $\mathcal{S}_i$ 's are disjoint, we let  $s = \sum_{i=1}^K s_i$ ,  $\mathcal{S} = \bigcup_{i=1}^K \mathcal{S}_i$  and  $|\mathcal{S}| := M = \sum_{i=1}^K M_i$ . Literature survey on blind deconvolution and related problems is given in Appendix B.

In order to estimate the mixture of autocorrelations, we assume that the support of s is not clustered. In particular, we assume that

$$|i-j| \ge L-1$$
, for all  $i, j \in \mathcal{S}$  such that  $i \ne j$ . (4.2)

The goal of the problem is to estimate  $x_1, \ldots, x_K$  from y.

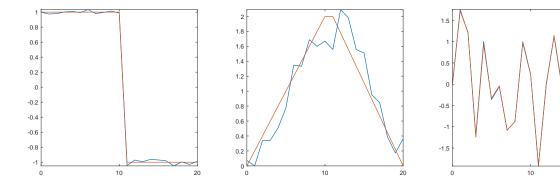


Figure 5: An experiment for estimate three one-dimensional signals simultaneously with noise level  $\sigma=3$ . Red signals are the ground truth (targets) and the blue signals are our estimations. Individual RMSE of the estimates: 0.0239393/0.208925/0.0335956. The estimated  $\gamma$ 's are 0.02574/0.01693/0.00818 and the true ones 0.02561/0.01707/0.00853. The SNR is 1/12. [1. to replace with a "progress" plot 2. replace the triangle signal 3. add error progress figure]

**Aperiodic autocorrelation functions** For  $z \in \mathbb{R}^L$  and  $k \geq 2$ , the autocorrelation of order k is defined for any integer shifts  $\ell_1, \ldots, \ell_{k-1}$  by

$$a_z^k[\ell_1, \dots, \ell_{k-1}] = \sum_{i=-\infty}^{+\infty} z[i]z[i+\ell_1]\dots z[i+\ell_{k-1}],$$
 (4.3)

where indexing of z out of the bounds  $0, \ldots, L-1$  is zero-padded, as usual. Explicitly, the mean (which can be thought of as a first-order autocorrelation) and the second and third autocorrelations are given by

$$a_{z}^{1} = \sum_{i=0}^{L-1} z[i],$$

$$a_{z}^{2}[\ell] = \sum_{i=\max\{0,-\ell\}}^{L-1+\min\{0,-\ell\}} z[i]z[i+\ell],$$

$$a_{z}^{3}[\ell_{1},\ell_{2}] = \sum_{i=\max\{0,-\ell_{1},-\ell_{2}\}}^{L-1+\min\{0,-\ell_{1},-\ell_{2}\}} z[i]z[i+\ell_{1}]z[i+\ell_{2}].$$

$$(4.4)$$

Note that the autocorrelation functions are symmetric so that  $a_z^2[\ell] = a_z^2[-\ell]$  and also

$$a_z^3[\ell_1, \ell_2] = a_z^3[\ell_2, \ell_1] = a_z^3[-\ell_1, \ell_2 - \ell_1].$$

Estimating the autocorrelation function of a single signal We first consider the problem of estimating the autocorrelations of a single signal from the data. The main principles carry through for K > 1 as will be shown next.

For the purpose of the analysis, we consider the asymptotic regime where  $M, N \to \infty$ , while preserving fixed ratio. Specifically, we define the ratio of the measurement occupied by the signal as

$$\gamma = \frac{ML}{N}.\tag{4.5}$$

Under the spacing constraint (4.2), we have  $\gamma \leq \frac{L}{2L-1} \approx 1/2$ .

The main pillar of this work is the following simple observation. If the support signal s satisfies the spacing constraint (4.2), then the first L entries of the data autocorrelations converge to a scaled, biased, version of the signal's autocorrelation:

$$\lim_{N \to \infty} a_y^1 = \gamma a_x^1,$$

$$\lim_{N \to \infty} a_y^2[\ell] = \gamma a_x^2[\ell] + \sigma^2 \delta[\ell],$$

$$\lim_{N \to \infty} a_y^3[\ell_1, \ell_2] = \gamma a_x^3[\ell_1, \ell_2] + \sigma^2 \gamma a_x^1 (\delta[\ell_1, 0] + \delta[0, \ell_2] + \delta[\ell_1, \ell_2]),$$
(4.6)

for  $\ell, \ell_1, \ell_2 = 0, \dots L - 1$ , and where  $\delta$  denotes the Kronecker delta function.

Estimating the autocorrelation function of a multiple signals As before, we consider the asymptomatic regime where  $M_1, \ldots, M_K, N \to \infty$ , while preserving fixed ratios

$$\gamma_k = \frac{M_k L}{N}, \quad \gamma = \sum_{k=1}^K \gamma_k. \tag{4.7}$$

If the support s satisfies the spacing constraint (4.2), then similarly to (4.6) one can estimate the mixture of the K signals' autocorrelations:

$$\lim_{N \to \infty} a_y^1 = \sum_{k=1}^K \gamma_k a_{x_k}^1,$$

$$\lim_{N \to \infty} a_y^2[\ell] = \sum_{k=1}^K \gamma_k a_{x_k}^2[\ell] + \sigma^2 \delta[\ell],$$

$$\lim_{N \to \infty} a_y^3[\ell_1, \ell_2] = \sum_{k=1}^K \gamma_k a_{x_k}^3[\ell_1, \ell_2] + \sigma^2 \left(\sum_{k=1}^K \gamma_k a_{x_k}^1\right) (\delta[\ell_1, 0] + \delta[0, \ell_2] + \delta[\ell_1, \ell_2]),$$
(4.8)

This relation is proven in Appendix A.

Numerical experiments details For the 2-D experiment shown in Figures 3 and 4, we generated P micrographs of size  $4096 \times 4096$  pixels. In each micrograph, we placed Einstein's image of size  $50 \times 50$  in random locations (see Figure 1a), while keeping the separation condition (4.2). This is done by randomly selecting 4000 placements in the micrograph, one at a time with an accept/reject rule based on the separation constraint and locations picked so far. In average, 700 images were placed in each micrograph. Then, an i.i.d. Gaussian noise with standard deviation  $\sigma = 3$  was added inducing SNR of approximately 1/20. An example of a micrograph's excerpt is presented in the right panel of Figure 2.

In this experiment, we assume to know the number of Einstein's occurrences and the noise level. In this situation, the second-order autocorrelation (or, equivalently, the Fourier magnitudes of the signal) determines almost all images uniquely up to reflection through the origin [27, 13]. Therefore, we estimate the signal's Fourier magnitude from the Fourier magnitude of the micrographs, in the cost of one FFT per micrograph. To estimate the signal, we use a phase retrieval algorithm called relaxed-reflect-reflect (RRR) that iterates by

$$z \to z + \beta (P_2(2P_1(z) - z) - P_1(z)),$$
 (4.9)

where we set  $\beta=1$ . RRR estimates a  $2L\times 2L$  image. Exact solution contains Einstein's image in the upper-left corner and zero else where. The operator  $P_2(z)$  combines the Fourier phases of the current estimation z with the Fourier magnitudes of the signal (estimated from the data). The operator  $P_1(z)$  zeros out all entries of z outside the  $L\times L$  upper-left corner. In order to compare the performance in multiple cases and different noise levels, the algorithm stopped after a fixed number of 1000 iterations and the iteration with the smallest error compared to the ground truth (also taking into account the reflection ambiguity) was chosen as the solution. While this cannot be done in practice since we do not have access to the ground truth, this procedure enables us to compare a large number of instances in different noisy environments [Note the last two sentences!].

For the 1-D experiment, we worked with three signals of length L=21 and generated the data in the same way as in the 2-D example. The only difference is that here, for each placement, one of the three signals is picked at random proportionally to the desired number of occurrences of each. In this experiment, we computed the first three autocorrelation functions. We do not assume to know the number of occurrences of each signal  $\gamma_i$  ahead and we removed the biased terms (see (4.6)) so we do not need to know  $\sigma$  either. In Appendix C we provide an argument on the number of equations we get from the first three autocorrelation functions.

To estimate the signal, we employ an optimization algorithm on the following nonlinear least-squares problem that estimate the signals and their number of occurrences simultaneously:

$$\min_{\substack{\hat{x}_1, \dots, \hat{x}_K \in \mathbb{R}^W \\ \hat{\gamma}_1, \dots, \hat{\gamma}_K > 0}} w_1 \left( a_y^1 - \sum_{k=1}^K \hat{\gamma}_k a_{\hat{x}_k}^1 \right)^2 + w_2 \sum_{\ell=1}^{L-1} \left( a_y^2[\ell] - \sum_{k=1}^K \hat{\gamma}_k a_{\hat{x}_k}^2[\ell] \right)^2 + w_3 \sum_{\substack{2 \le \ell_1 \le L-1 \\ 1 \le \ell_2 \le \ell_1 - 1}} \left( a_y^3[\ell_1, \ell_2] - \sum_{k=1}^K \hat{\gamma}_k a_{\hat{x}_k}^3[\ell_1, \ell_2] \right)^2. \quad (4.10)$$

where  $W \ge L$  is the length of the sought signals and  $w_1 = \frac{1}{2}, w_2 = \frac{1}{2}(L-1), w_3 = \frac{1}{(L-1)(L-2)};$  see [19] see for discussion on how to choose the weights properly.

Setting W = L (as is a priori desired) is problematic because the above optimization problems appears to have numerous poor local optimizers. Thus, we first run the optimization with W = 2L - 1. This problem appears to have fewer poor local optima, perhaps because the additional degrees of freedom allow for more escape directions. Since we hope the signals estimated this way correspond to the true signals zero-padded to length W, we extract from each one a subsignal of length L whose autocorrelation functions are the closest the measured ones in the sense of (4.10). This estimator is then used as initial iterate for (4.10), this time with W = L. We find that this procedure is reliable for a wide range of experimental parameters. To solve (4.10), we run the trust-region method implemented in Manopt [18], which allows to treat the positivity constraints [reference] on coefficients  $\hat{\gamma}_k$ . Notice that the cost function is a polynomial in the variables, so that it is straightforward to compute it and its derivatives.

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#### A Autocorrelation estimations

Throughout the proof, we consider the case of one signal K = 1. The extension to K > 1 is straightforward by averaging the contributions of all signal with appropriates weights, see [19]. Let us define

$$\gamma = \lim_{N \to \infty} \frac{M_N L}{N} < 1. \tag{A.1}$$

With a bit abuse of notation,  $M_N$  stresses that M is a function of N. Indeed, by assuming  $M_N = \Omega(N)$ , we deduce  $\gamma > 0$ . We start by considering the first autocorrelation of the data

$$a_y^1 = \sum_{i=0}^{N-1} y[i] = \frac{1}{N/L} \sum_{j=0}^{M_N-1} \frac{1}{L} \sum_{i=0}^{L-1} x[i] + \underbrace{\frac{1}{N} \sum_{i=0}^{N-1} \varepsilon[i]}_{\text{noise term}} \xrightarrow{a.s.} \gamma a_x^1, \tag{A.2}$$

where the noise term converges to zero almost surely (a.s.) by the law of large numbers.

We proceed with the second autocorrelation for fixed  $\ell \in [0, \ldots, L-1]$ . We can compute:

$$a_y^2[\ell] = \frac{1}{N} \sum_{i=0}^{N-1-\ell} y[i]y[i+\ell]$$

$$\underbrace{\frac{1}{N} \sum_{j=1}^{M_N} \sum_{i=0}^{L-\ell-1} x[i]x[i+\ell]}_{\text{signal term}} + \underbrace{\frac{1}{N} \sum_{i=0}^{N-1-\ell} \varepsilon[i]\varepsilon[i+\ell]}_{\text{noise term}},$$
(A.3)

where the cross terms between the signal and the noise vanish almost surely in the limit  $N \to \infty$ .

We treat the signal and noise terms separately. We first break the signal term into  $M_N$  different sums, each contains one copy of the signal, and get

$$\frac{1}{N} \sum_{i=1}^{M_N} \sum_{i=0}^{L-\ell-1} x[i]x[i+\ell] = \frac{M_N L}{N} \frac{1}{L} \sum_{i=0}^{L-\ell-1} x[i]x[i+\ell] = \gamma a_x^2[\ell]. \tag{A.4}$$

Similarly, for  $\ell \neq 0$ , we can break the noise term into a sum of independent terms

$$\frac{1}{N} \sum_{i=0}^{N-1-\ell} \varepsilon[i] \varepsilon[i+\ell] = \frac{1}{\ell} \sum_{i=0}^{\ell-1} \frac{1}{N/\ell} \sum_{j=0}^{N/\ell-1} \varepsilon[j\ell+i] \varepsilon[(j+1)\ell+i]. \tag{A.5}$$

Each term of  $\frac{1}{N/\ell} \sum_{j=0}^{N/\ell-1} \varepsilon[j\ell+i]\varepsilon[(j+1)\ell+i]$  is an average of  $N/\ell$  independent terms with expectation zero, and thus converges to zero almost surely as  $N \to \infty$ . If  $\ell = 0$ ,

$$\frac{1}{N} \sum_{i=0}^{N-1} \varepsilon^2[i] \xrightarrow{a.s.} \sigma^2. \tag{A.6}$$

We are now moving to analyze the third-order autocorrelation. Let us fix  $\ell_1 \geq \ell_2$  and recall that

$$a_y^3[\ell_1, \ell_2] = \sum_{i=0}^{N-1-\ell_1} y[i]y[i+\ell_1]y[i+\ell_2].$$

Writing explicitly in terms of signal and noise, the sum can be broken into eight terms. The first contains only signal terms (does not see noise) and converges to  $\gamma a_x^3$  from the same reasons as (A.4). Three other terms contain the product of two signal entries and one noise term. Since the noise is independent of the signal and has zero mean, these terms go to zero almost surely.

We next analyze the contribution of the term composed of triple products of noise terms. For  $\ell_1 \neq 0$ , this sum can be formulate as follows:

$$\sum_{i=0}^{N-1-\ell_1} \varepsilon[i] \varepsilon[i+\ell_1] \varepsilon[i+\ell_2] = \frac{1}{\ell_1} \sum_{i=0}^{\ell_1-1} \frac{1}{N/\ell_1} \sum_{j=0}^{N/\ell_1-1} \varepsilon[j\ell_1+i] \varepsilon[(j+1)\ell_1+i] \varepsilon[j\ell_1+i+\ell_2].$$

For each fixed i, we sum of over  $N/\ell_1$  independent variables that goes to zero almost surely. For  $\ell_1 = \ell_2 = 0$ , we get a some of N independent variables, each one is a triple product of Gaussian variables with zero mean and therefore has zero expectation.

To complete the analysis, we consider the three terms composed of the product of two noise terms and one signal entry. Most of these terms converge to zero almost surely because of independency between the noise entries. For  $\ell_1 = 0$ ,  $\ell_2 = 0$  and  $\ell_1 = \ell_2$ , a simple computation shows that the sum converges to  $\gamma \sigma^2 a_x^1$ ; c.f. [19].

#### B Related literature

Blind deconvolution. Blind deconvolution is a longstanding problem, arising in a variety of engineering and scientific applications, such as astronomy, communication, image deblurring, system identification and optics; see [31, 48, 7, 3], just to name a few. To make the problem well-posed, we must assume some prior knowledge or structure. In our case, the prior information is that s is a binary signal that satisfies the separation constraint (4.2). Other settings of blind deconvolution problems have been analyzed recently, see for instance [5, 39, 38, 41, 42, 21] where the focus is on high SNR regimes.

An important feature of the problem under consideration is that while both  $x_i$ 's and  $s_i$ 's are unknown, the goal is merely to estimate the  $x_i$ 's. The  $s_i$ 's are referred to as nuisance variables. Indeed, in many blind deconvolution applications the sole purpose is to recover one of the unknown signals. For instance, in image deblurring, both the blurring kernel and the high-resolution image are unknown, but the primary goal is only to sharpen the image.

The impossibility of detection in low SNR. If x is known and K=1, then a sparse signal can be estimated by linear programming in the high SNR regime, e.g., [8, 22, 15, 12, 17]. However, in the low SNR regime, estimating the binary sparse signal s is impossible. To see that, suppose that an oracle provides us M windows of length W > L, each contains one copy of x. That is to say, we get a series of windows of length W, each one contains a signal at an unknown location. Estimating the position of the known signal within each window is an easier problem than detecting the support of s. Nevertheless, even this problem is impossible in the low SNR regime [4]. Therefore, we conclude that detecting the nonzero values of s is impossible in low SNR. As aforementioned, this work focuses on this regime and examines under what conditions we can estimate the signals, despite the impossibility of detecting their individual occurrences.

**System identification.** For K = 1, our problem can be also interpreted as a special case of the system identification problem. Similarly to (4.1), the system identification forward model takes the form

$$y = x * w + \varepsilon, \tag{B.1}$$

where x is the unknown signal ("system"), w is an unknown, random, input sequence and  $\varepsilon$  is an additive noise. The goal of this problem is to estimate x, usually referred to as "identifying the system." The question of identifiability of x under this observation model is addressed for certain Gaussian and non-Gaussian w in [16, 33]. In the special case where  $w \in \{0,1\}^N$ , satisfying the spacing requirement (4.2), we obtain our model in the case of a single signal (K=1).

Likelihood-based methods seek to maximize the likelihood function for x, given the observed signal y. Solving this optimization exactly is typically intractable, and thus heuristic methods are used instead. One proposed technique is to use Markov Chain Monte Carlo (MCMC); in special cases, including the case where w is binary, EM has been used [20]. The EM method is based upon a certain "forward-backward" procedure used in hidden Markov models [46]. However, the complexity of this procedure is still nonlinear in N, and therefore its usage is limited for big data sets. Another paper considers parameterized models for multiple distinct signals, as in our heterogeneity framework (K > 1) [6]. Their proposed solution is an MCMC algorithm tailored for their specific parametrized problem.

Because likelihood methods are computationally expensive, methods based on recovery from moments, which are akin to our method, have also been previously used for system identification. Methods based on the third- and fourth-order moments are described and analyzed in [40, 24, 52].

## C Theory

A one-dimensional signal is determined uniquely and stably by its third-order autocorrelation as proven in the following simple proposition.

**Proposition C.1.** Let  $z \in \mathbb{R}^L$  and suppose that z[0] and z[L-1] are nonzero. Then:

- Uniqueness: z is determined uniquely from  $a_z^2$  and  $a_z^3$ .
- Finite sensitivity: Suppose we can only measure  $\tilde{a_z^3}[k,L-1] = a_z^3[k,L-1] + v$  and that  $|z[0]z[L-1]| \ge \delta > 0$ . Then,  $\hat{z}[k] = \frac{a_z^3[k,L-1]}{a_z^2[L-1]}$  satisfies  $|\hat{z}[k] z[k]| \le \frac{|v|}{\delta}$ .

*Proof.* By assumption  $a_z^2[L-1] = z[0]z[L-1] \neq 0$ . Then, the uniqueness results, for all  $k = 0, \ldots L - 1$ , follows from:

$$a_z^3[k, L-1] = z[0]z[k]z[L-1].$$

In addition,

$$\hat{z}[k] = \frac{\tilde{a_z^3}[k,L-1]}{a_z^2[L-1]} = z[k] + \frac{\upsilon}{a_z^2[L-1]} \quad \Rightarrow \quad |\hat{z}[k] - z[k]| \leq \frac{|\upsilon|}{\delta}.$$

A few remarks are in order. First, the second result of Proposition C.1 shows that there exists a very simple estimator that has finite sensitivity. In the next numerical experiments we use an estimator based on nonconvex LS that shows empirical robustness to additive noise, in accordance with related problems [14, 19]. Second, these results carry through to signals

of any dimension. Third, if the spacing condition (4.2) holds, then the length of the signal can be determined from the autocorrelations and therefore the assumption that the first and last entries are nonzero is met. In particular, if (4.2) holds for some spacing  $W \geq L$ , then  $a_z^2[i] = 0$  for all i > L - 1. Finally, computing the dth autocorrelation amplifies the variance of the noise by a factor d in the low SNR regime. Therefore, if we can estimate  $a_z^3$  up to small perturbation, it implies that we can estimate  $a_z^2$  accurately as the proposition assumes.

Considering the third-order autocorrelation is also a necessary condition to determine a signal from its autocorrelations. Indeed, the second-order autocorrelation of a one-dimensional signal does not determine a signals uniquely [11, 13]. On the other hand, for dimensions greater than one, almost all signals are determined uniquely, up to sign (phase for the complex signals) and reflection through the origin (with conjugation in the complex case) [27, 28]. The sign ambiguity can be resolved by the mean of the signal if it is not zero. However, in order to determine the reflection symmetry, one needs to use additional information.

If the noise level  $\sigma^2$  is known, one can estimate the ratio M/N from the first two moments.

**Proposition C.2.** Let  $K=1, N\to\infty$  and  $\sigma$  fixed. If the mean of x is nonzero, then

$$\frac{M}{N} = \frac{1}{L} \frac{(a_y^1)^2}{\sum_{j=0}^{L-1} a_y^2[j] - \sigma^2}.$$

*Proof.* The proof follows from plugging the explicit expressions of (4.6) into the right hand side of the equality.

If we use third-order autocorrelation information, then it is possible to estimate both the ratio M/N and  $\sigma$  simultaneously.

**Proposition C.3.** Let  $K=1,\ N\to\infty$  and  $\sigma$  fixed. Then,  $a_y^1,a_y^2$  and  $a_y^3$  determine the ratio M/N and  $\sigma$  uniquely for a generic signal x. If  $\frac{M}{N}\geq \frac{1}{4(L-1)}$ , then it holds for any signal with nonzero mean.

Proof. See Appendix D. 
$$\Box$$

From Propositions C.1 and C.3 we can directly deduce the following:

**Corollary C.4.** Let  $K=1, N \to \infty$  and  $\sigma$  is fixed. Then, the signal, the ratio M/N and  $\sigma$  can be recovered from the first three autocorrelation functions if:

- $\bullet$  x is generic;
- $x[0], x[L-1] \neq 0$ , x has nonzero mean and  $\frac{M}{N} \geq \frac{1}{4(L-1)}$ .

**Expectation-maximization.** Interestingly, expectation-maximization (EM)—a popular algorithm for similar estimation problems, such as Gaussian mixture models and multireference alignment—is intractable for this problem. This is true even if K=1 and the number of signal occurrences M is known. In particular, in each iteration, EM assigns a probability to any feasible combination of positioning the current signal estimate in M locations on the grid  $\{1,\ldots,N\}$ . In total, even when excluding forbidden combinations due to the spacing constraint, there are  $O(N^M)$  such combinations.

Open theoretical questions. The minimal order of data statistics used to get an accurate estimation of a signal is important to understand, in the asymptotic SNR regime, the sample complexity of the problem. In methods which are based on detection and averaging, the number of signals occurrences must scale like 1/SNR. On the contrary, the required number signal occurrences using the autocorrelation analysis should scale like  $1/SNR^d$  to retain a constant estimation error. Accordingly, in our method, M must scale like  $1/SNR^3$ . We believe that similarly to the closely-related problem of multirefrence alignment [45, 2], this estimation rate is optimal in the low SNR regime.

Another interesting question is how many signals can be demixed from the autocorrelation functions. For second-order moments, notice from equation (4.8) that  $a_y^2[\ell]$  suffers no bias for  $\ell$  in 1 to L-1. Thus, we omit  $\ell=0$ , which has the practical effect that we need not know  $\sigma$  to estimate the moments. Likewise, for third-order moments,  $a_y^3[\ell_1,\ell_2]$  for  $0 \le \ell_1,\ell_2 \le L-1$  such that  $\ell_2 \le \ell_1$  includes all relevant moments for our purpose, and we further exclude any such that  $\ell_1,\ell_2$  or  $\ell_1-\ell_2$  are zero to avoid biased elements—there are  $\frac{(L-1)(L-2)}{2}$  remaining moments. As a result, it is unnecessary to estimate  $\sigma$ . We have

$$1 + (L-1) + \frac{(L-1)(L-2)}{2} = \frac{1}{2}L(L-1) + 1 \approx L/2,$$

moments in total.

In the related problem of demixing signals from their periodic autocorrelation functions, there are strong evidences that the number of signals that, theoretically, can be demixed is equal to the number of different equations [9]. Based on these evidence, we conjecture the same here, namely, L/2 signals can be estimated simultaneously. That being said, we believe the number of classed that can actually be demixed by efficient algorithm is significantly smaller and scale like  $\sqrt{L}$ ; see [19, 54].

# D Proof of Proposition C.3

We aim to prove that one can estimate both  $\sigma$  and  $\beta=M/N$  from the observed first three moments. To this end, we construct two quadratic equations of  $\beta$  from the observed (measured) quantities, independent of  $\sigma$ . Then, we show that these equations are independent and therefore  $\beta$  is uniquely defined. Given  $\beta$ , we can estimate  $\sigma$  using Proposition C.2. Throughout the proof, it is important to distinguish between observed and unobserved values. to this end, we denote the observed values by  $E_i$  or  $a_y^1, a_y^2, a_y^3$ , while using  $F_i$  for functions of the signal's autocorrelations.

Recall that  $a_y^1 = \beta(\mathbf{1}^T x)$  and and  $a_y^2[0] = \beta ||x||^2 + \sigma^2$ , where  $\mathbf{1} \in \mathbb{R}^L$  stands for vector of ones. Taking the product:

$$E_1 := a_y^1 a_y^2 [0] = (\beta(\mathbf{1}^T x))(\beta ||x||^2 + \sigma^2)$$
  
=  $\sigma^2 a_y^1 + \beta^2 F_1$ , (D.1)

where  $F_1 := a_x^3[0,0] + \sum_{j=1}^{L-1} (a_x^3[j,j] + a_x^3[0,j])$ . The terms of  $F_1$  can be also estimated from  $a_y^3$ , while taking the scaling and bias terms into account:

$$E_2 := \beta F_1 + (2L+1)\sigma^2 a_n^1. \tag{D.2}$$

Therefore, from (D.1) and (D.2) we get

$$E_2\beta - (2L+1)\sigma^2\beta a_y^1 = E_1 - \sigma^2 a_y^1.$$
 (D.3)

Let  $a_y^2 := \sum_{j=0}^{L-1} a_y^2[j]$  and recall from Proposition C.2:

$$\sigma^2 = a_y^2 - (a_y^1)^2 / (\beta L). \tag{D.4}$$

Plugging into (D.3) and rearranging we get

$$\mathcal{A}\beta^2 + \mathcal{B}\beta + \mathcal{C} = 0,\tag{D.5}$$

where

$$\mathcal{A} = E_2 - (2L+1)a_y^1 a_y^2,$$

$$\mathcal{B} = -E_1 + \frac{2L+1}{L} (a_y^1)^3 + a_y^1 a_y^2,$$

$$\mathcal{C} = -(a_y^1)^3 / L.$$

Importantly, these coefficients are observable quantities.

We are now proceeding to derive the second quadratic equation. We notice that

$$E_3 = \frac{1}{L} (a_y^1)^3 = \frac{1}{L} \beta^3 (\mathbf{1}^T x)^3 = \frac{1}{L} \beta^3 F_2,$$
 (D.6)

where

$$F_2 = a_x^3[0,0] + 3\sum_{j=1}^{L-1} a_x^3[j,j] + 3\sum_{j=1}^{L-1} a_x^3[0,j] + 6\sum_{1 \le i < j \le L-1} a_x^3[i,j].$$

On the other hand, from  $a_y^3$  we can directly estimate  $F_2$  up to scale and bias

$$E_4 = \beta F_2 + (6L - 3)\sigma^2 a_y^1. \tag{D.7}$$

Taking the ratio:

$$\frac{E_4}{E_3} = \frac{L}{\beta^2} + \frac{(6L - 3)L\sigma^2 a_y^1}{E_3},$$

we conclude:

$$\sigma^2 = \frac{E_4}{a_v^1 L(6L-3)} - \frac{E_3}{\beta^2 a_v^1 (6L-3)}.$$

Using (D.4) and rearranging we get the second quadratic:

$$\mathcal{D}\beta^2 + \mathcal{E}\beta + \mathcal{F} = 0, \tag{D.8}$$

where

$$\mathcal{D} = a_y^2 - \frac{E_4}{a_y^1 L (6L - 3)},$$
 
$$\mathcal{E} = -(a_y^1)^2 / L,$$
 
$$\mathcal{F} = \frac{E_3}{a_y^1 (6L - 3)}.$$

To complete the proof, we need to show that the two quadratic equations (D.5) and (D.8) are independent. To this end, it is enough to show that the ratio between the coefficients is not the same. From (D.5) and (D.1), we have

$$\frac{\mathcal{B}}{\mathcal{C}} = \frac{LE_1 - (2L+1)(a_y^1)^3 - La_y^1 a_y^2}{(a_y^1)^3}$$
$$= \frac{La_y^2[0] - (2L+1)(a_y^1)^2 - La_y^2}{(a_y^1)^2}.$$

In addition, using (D.6)

$$\frac{\mathcal{E}}{\mathcal{F}} = \frac{(3 - 6L)(a_y^1)^3}{LE_3} = 3 - 6L.$$

Now, suppose that the quadratics are dependent. Then,  $\frac{\mathcal{B}}{\mathcal{C}} = \frac{\mathcal{E}}{\mathcal{F}}$ , or,

$$La_y^2[0] - (2L+1)(a_y^1)^2 - La_y^2 = (a_y^1)^2(3-6L)$$

Rearranging the equation and writing in terms of x we get

$$4(L-1)\beta(a_x^1)^2 - \sum_{i=1}^{L-1} a_x^2[i] = 0.$$
 (D.9)

For generic x, this polynomial equation is not satisfied. Therefore, the equations are independent. More than that, for any nonzero x,  $(a_x^1)^2 > \sum_{i=1}^{L-1} a_x^2[i]$ . Therefore, if  $4(L-1)\beta \geq 1$ , or,

$$\beta \ge \frac{1}{4(L-1)},$$

the condition (D.9) cannot be satisfied for any signal.