

SECONG ORDER CONE PROGRAMMING

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REFERENCES:

S. BOYD AND L. VANDENBERGHE, **CONVEX OPTIMIZATION**. CAMBRIDGE
UNIVERSITY PRESS, 2004. (Available from <http://www.stanford.edu/~boyd/cvxbook/>)

1. Second-order Cone

1.1 Euclidean Norm

$$\|\mathbf{x}\|_2 = \sqrt{\mathbf{x}^T \mathbf{x}} = \sqrt{x_1^2 + \cdots + x_N^2}, \quad (1)$$

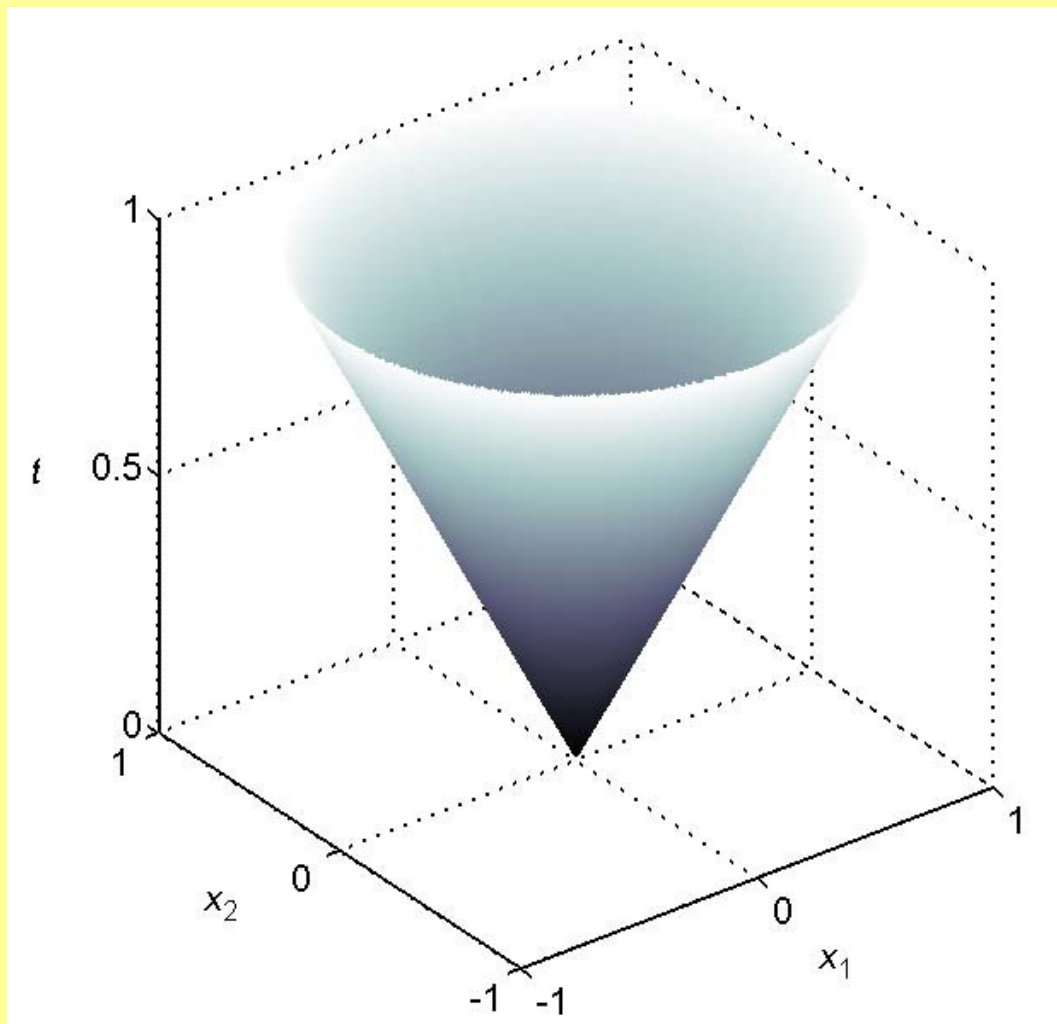
1.2 Second-order Cone

$$\begin{aligned} C &= \{(\mathbf{x}, t) \in \mathbb{R}^{N+1} \mid \|\mathbf{x}\|_2 \leq t\} \\ &= \left\{ \begin{bmatrix} \mathbf{x} \\ t \end{bmatrix} \mid \begin{bmatrix} \mathbf{x} \\ t \end{bmatrix}^T \begin{bmatrix} I & 0 \\ 0 & -1 \end{bmatrix} \begin{bmatrix} \mathbf{x} \\ t \end{bmatrix} \leq 0, \ t \geq 0 \right\} \end{aligned} \quad (2)$$

Example:

Boundary of second-order cone in \mathbb{R}^3 .

$$\{(x_1, x_2, t) \mid \sqrt{x_1^2 + x_2^2} \leq t\}$$



2. Second-order Cone Programming

$$\begin{aligned}
 & \min_{\mathbf{x}} \quad \mathbf{c}^T \mathbf{x} \\
 & \text{subject to} \quad \|\mathbf{A}_i \mathbf{x} + \mathbf{b}_i\|_2 \leq \mathbf{c}_i^T \mathbf{x} + \mathbf{d}_i, \quad i = 1, \dots, M, \\
 & \quad \quad \quad \mathbf{F} \mathbf{x} = \mathbf{g},
 \end{aligned} \tag{3}$$

where $\mathbf{x} \in \mathbb{R}^N$ is the optimization variable, $\mathbf{A}_i \in \mathbb{R}^{N_i \times N}$, $\mathbf{F} \in \mathbb{R}^{P \times N}$, and $\|\mathbf{A}_i \mathbf{x} + \mathbf{b}_i\|_2 \leq \mathbf{c}_i^T \mathbf{x} + \mathbf{d}_i$ for $n = 1, \dots, N$ are second-order cone constraints.

- When $\mathbf{c}_i = 0$ for $i = 1, \dots, M$, the SOCP reduces to quadratically constrained quadratic programming (QCQP).
- When $\mathbf{A}_i = 0$ for $i = 1, \dots, M$, the SOCP reduces to linear programming (LP).

3. Design Example: FIR Filter Design

- **Frequency response of a general digital FIR filter:**

$$H(e^{j\omega}) = \sum_{n=0}^{N-1} h(n)e^{-j\omega n} = \mathbf{h}^T (\mathbf{c}(\omega) - js(\omega)),$$

where $\mathbf{h} = [h(0), \dots, h(N-1)]^T$, $\mathbf{c}(\omega) = [1, \dots, \cos((N-1)\omega)]^T$ and $\mathbf{s}(\omega) = [0, \dots, \sin((N-1)\omega)]^T$.

- **Minimax Design Criterion:**

$$\min_{\mathbf{h}} \max_{\omega \in \Omega} W(\omega) |H(e^{j\omega}) - H_d(\omega)|,$$

where Ω is the frequency interval of interest; $W(\omega)$ is a positive weighting function; $H_d(\omega)$ is the desired frequency response.

- **Discretizing** $\omega \in \Omega$ **into** M **equally spaced samples, gives**

$$\begin{aligned} \min_h \quad & \delta \\ \text{subject to} \quad & \delta - \{\alpha_R^2(\omega_i) + \alpha_I^2(\omega_i)\}^{1/2} \geq 0, \quad i = 1, \dots, M, \end{aligned}$$

where $\alpha_R(\omega) = W(\omega) \cdot \{\mathbf{h}^T \mathbf{c}(\omega) - \text{Re}[H_d(\omega)]\},$

$$\alpha_I(\omega) = W(\omega) \cdot \{\mathbf{h}^T \mathbf{s}(\omega) + \text{Im}[H_d(\omega)]\}.$$

- **Equivalent SOCP problem (with variable** $\mathbf{x} = [\delta \quad \mathbf{h}^T]^T$ **):**

$$\begin{aligned} \min_{\mathbf{x}} \quad & \mathbf{c}^T \mathbf{x} \\ \text{subject to} \quad & \mathbf{c}^T \mathbf{x} \geq \|A_i \mathbf{x} - \mathbf{b}_i\|_2, \quad i = 1, \dots, M, \end{aligned}$$

where $\mathbf{c} = [1 \quad \mathbf{0}_N^T]^T$, $A_i = W(\omega_i) \begin{bmatrix} 0 & \mathbf{c}(\omega_i)^T \\ 0 & \mathbf{s}(\omega_i)^T \end{bmatrix}$, $\mathbf{b}_i = W(\omega_i) \begin{bmatrix} -\text{Re}[H_d(\omega_i)] \\ \text{Im}[H_d(\omega_i)] \end{bmatrix}$

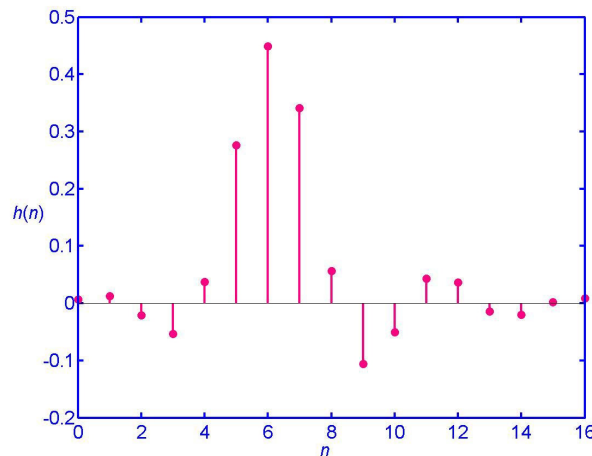
$\mathbf{0}_N$ **is an** N **row zero vector.**

Example (low-delay lowpass FIR filter)

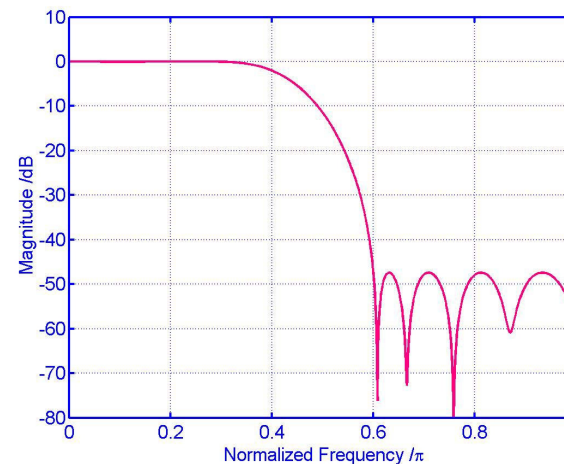
■ **Desired response:** $H_d(e^{j\omega}) = \begin{cases} e^{-j\omega\tau}, & 0 < \omega < \omega_p \\ 0, & \omega_s < \omega < \pi \end{cases}$

where group delay $\tau = (N - 1) / 2 - D$; D is delay reduction parameter.

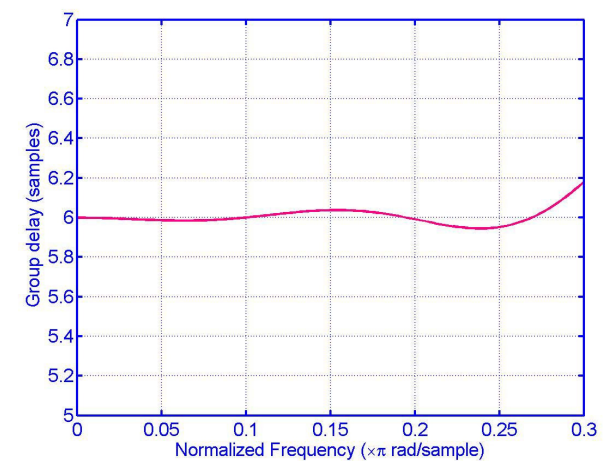
■ **Specifications:** $N = 17$, $D = 2$, $\omega_p = 0.3\pi$, $\omega_s = 0.6\pi$, $M = 200$ and $W(\omega) = 1, \forall \omega$.



Impulse Response



Frequency Response



Group Delay Response