MODEL PREDICTIVE CONTROL

HYBRID MODELS FOR MPC

Alberto Bemporad

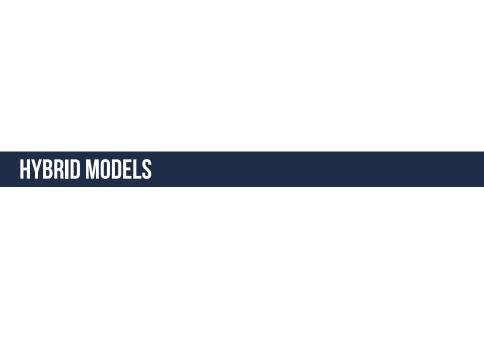
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COURSE STRUCTURE

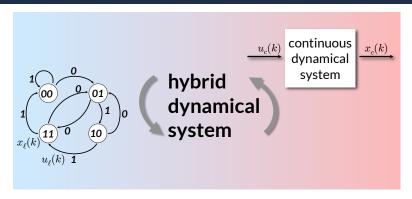
- ✓ Basic concepts of model predictive control (MPC) and linear MPC
- ✓ Linear time-varying and nonlinear MPC
- ✓ Quadratic programming (QP) and explicit MPC
- Hybrid MPC
- Stochastic MPC
- Learning-based MPC

Course page:

http://cse.lab.imtlucca.it/~bemporad/mpc_course.html



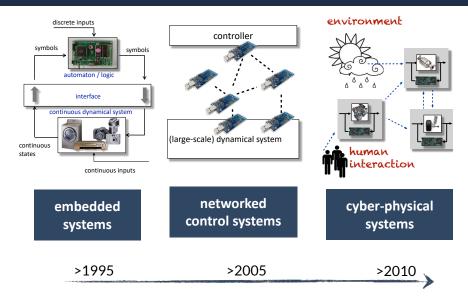
HYBRID DYNAMICAL SYSTEMS



- Variables are binary-valued $x_{\ell} \in \{0,1\}^{n_{\ell}}, u_{\ell} \in \{0,1\}^{m_{\ell}}$
- Dynamics = finite state machine
- Logic constraints

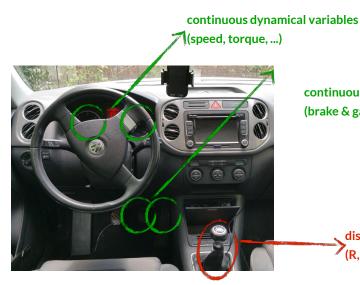
- Variables are real-valued $x_c \in \mathbb{R}^{n_c}, \ u_c \in \mathbb{R}^{m_c}$
- Difference/differential equations
- Linear inequality constraints

TECHNOLOGICAL PUSH FOR STUDYING HYBRID SYSTEMS



AN EXAMPLE OF "INTRINSICALLY HYBRID" SYSTEM

Vehicle



continuous commands (brake & gas pedal)

+

discrete command (R,N,1,2,3,4,5)

KEY REQUIREMENTS FOR HYBRID MODELS

- Descriptive enough to capture the behavior of the system
 - continuous dynamics (physical systems)
 - logic components (switches, automata)
 - interconnection between logic and dynamics
- **Simple** enough for solving analysis and synthesis problems

$$\begin{cases} x' = Ax + Bu \\ y = Cx + Du \end{cases}$$
 linear hybrid systems

$$\begin{cases} x' = f(x, u, t) \\ y = g(x, u, t) \end{cases}$$

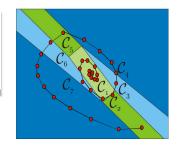
"Perfection is achieved not when there is nothing more to add, but when there is nothing left to take away."



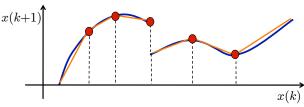
A. de Saint-Exupéry (1900-1944)

PIECEWISE AFFINE SYSTEMS

$$x(k+1) = A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)}$$
$$y(k) = C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)}$$
$$i(k) \text{ s.t. } H_{i(k)}x(k) + J_{i(k)}u(k) \le K_{i(k)}$$

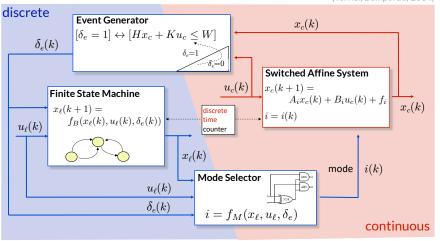


 PWA systems can approximate nonlinear dynamics arbitrarily well (even discontinuous ones)



DISCRETE HYBRID AUTOMATON (DHA)

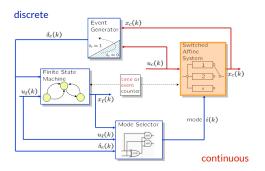
(Torrisi, Bemporad, 2004)



$$x_\ell \in \{0,1\}^{n_\ell} =$$
 binary state $u_\ell \in \{0,1\}^{m_\ell} =$ binary input $\delta_e \in \{0,1\}^{n_e} =$ event variable

 $x_c \in \mathbb{R}^{n_c}$ = real-valued state $u_c \in \mathbb{R}^{m_c}$ = real-valued input $i \in \{1, \dots, s\}$ = current mode

SWITCHED AFFINE SYSTEM

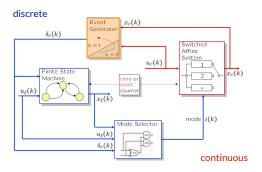


• The affine dynamics depend on the current mode i(k):

$$x_{c}(k+1) = A_{i(k)}x_{c}(k) + B_{i(k)}u_{c}(k) + f_{i(k)}$$

$$x_{c} \in \mathbb{R}^{n_{c}}, u_{c} \in \mathbb{R}^{m_{c}}$$

EVENT GENERATOR



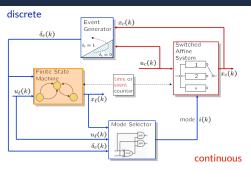
• Event variables are generated by linear threshold conditions over continuous

$$[\delta_e^i(k) = 1] \leftrightarrow [H^i x_c(k) + K^i u_c(k) \le W^i]$$

$$x_c \in \mathbb{R}^{n_c}, \quad u_c \in \mathbb{R}^{m_c}$$
$$\delta_e \in \{0, 1\}^{n_e}$$

• Example: $[\delta_e(k) = 1] \leftrightarrow [x_c(k) \ge 0]$

FINITE STATE MACHINE



• The binary state of the finite state machine evolves according to a Boolean state update function $f_B:\{0,1\}^{n_\ell+m_\ell+n_e}\to\{0,1\}^{n_\ell}$:

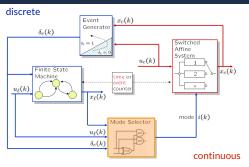
$$x_{\ell}(k+1) = f_B(x_{\ell}(k), u_{\ell}(k), \delta_e(k))$$

$$x_{\ell} \in \{0,1\}^{n_{\ell}}, \quad u_{\ell} \in \{0,1\}^{m_{\ell}}$$

 $\delta_e \in \{0,1\}^{n_e}$

• Example: $x_{\ell}(k+1) = \neg \delta_e(k) \lor (x_{\ell}(k) \land u_{\ell}(k))$

MODE SELECTOR



The mode selector can be seen as the output function of the discrete dynamics

• The active $\operatorname{mode} i(k)$ is selected by a Boolean function of the current binary

$$i(k) = f_M(x_{\ell}(k), u_{\ell}(k), \delta_e(k))$$

$$x_{\ell} \in \{0, 1\}^{n_{\ell}}, \quad u_{\ell} \in \{0, 1\}^{m_{\ell}}$$

 $\delta_e \in \{0, 1\}^{n_e}$

$$\begin{array}{c|cccc} \text{Example:} & & u_{\ell}/x_{\ell} & 0 & 1 \\ i(k) = \begin{bmatrix} \neg u_{\ell}(k) \lor x_{\ell}(k) \\ u_{\ell}(k) \land x_{\ell}(k) \end{bmatrix} & & u_{\ell}/x_{\ell} & 0 & i = \begin{bmatrix} 1 \\ 0 \end{bmatrix} & i = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \\ \hline 1 & i = \begin{bmatrix} 0 \\ 0 \end{bmatrix} & i = \begin{bmatrix} 1 \\ 1 \end{bmatrix} \\ \end{array}$$

the system has 3 modes

CONVERSION OF LOGIC FORMULAS TO LINEAR INEQUALITIES

(Glover, 1975) (Williams, 1977) (Hooker, 2000)

• Key observation: $X_1 \vee X_2 = \texttt{true}$ $\delta_1 + \delta_2 \geq 1, \delta_1, \delta_2 \in \{0, 1\}$

• We want to impose the Boolean statement

$$F(X_1,\ldots,X_n)=\mathtt{true}$$

• Convert the formula to Conjunctive Normal Form (CNF)

$$\bigwedge_{j=1}^m \left(\bigvee_{i\in P_j} X_i \bigvee_{i\in N_j} \bar{X}_i\right) = \texttt{true}, \quad P_j \cup N_j \subseteq \{1,\dots,n\}$$

Transform the CNF into the equivalent linear inequalities

$$\left\{ \begin{array}{ccc} \sum_{i \in P_1} \delta_i + \sum_{i \in N_1} (1 - \delta_i) & \geq & 1 \\ & \vdots & \vdots \\ \sum_{i \in P_m} \delta_i + \sum_{i \in N_m} (1 - \delta_i) & \geq & 1 \end{array} \right. \\ \left. \begin{array}{ccc} A\delta \leq b, \ \delta \in \{0, 1\}^n \\ \text{polyhedron} \end{array} \right.$$

Any logic proposition can be translated into integer linear inequalities

$\mathsf{LOGIC} o \mathsf{INEQUALITIES}$: Symbolic approach

• Example:

$$F(X_1, X_2, X_3) = [X_3 \leftrightarrow X_1 \land X_2]$$

• Convert Conjunctive Normal Form (CNF):

(see e.g. http://formal.cs.utah.edu:8080/pbl/PBL.php or just google "CNF + converter" ...)

$$(X_3 \vee \neg X_1 \vee \neg X_2) \wedge (X_1 \vee \neg X_3) \wedge (X_2 \vee \neg X_3)$$

Transform into inequalities:

$$\begin{cases} \delta_3 + (1 - \delta_1) + (1 - \delta_2) & \geq 1 \\ \delta_1 + (1 - \delta_3) & \geq 1 \\ \delta_2 + (1 - \delta_3) & \geq 1 \end{cases}$$

$\mathsf{LOGIC} o \mathsf{INEQUALITIES}$: Geometric approach

• Consider the Boolean statement $F(X_1,\ldots,X_n)=\mathtt{true}$ and collect the rows of the **truth table** T(F) of F

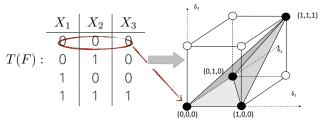
The convex hull $P=\{\delta\in\mathbb{R}^n:A\delta\leq b\}$ of the rows in T(F) is the smallest polytope equivalent to the Boolean statement F

(Mignone, Bemporad, Morari, 1999)

 Convex hull packages: cdd, lrs, qhull, chD, Hull, Porto CDDMEX package by K. Fukuda included in the Hybrid Toolbox

$\mathsf{LOGIC} o \mathsf{INEQUALITIES}$: GEOMETRIC APPROACH

• Example: $F(X_1,X_2,X_3)=[X_3\leftrightarrow X_1\wedge X_2]$ (logic and)



Key idea: white points cannot be inside the convex hull of black points

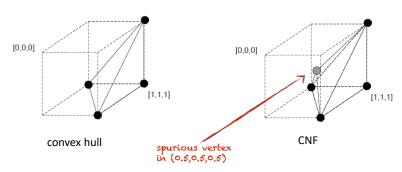
$$\operatorname{conv}\left(\left[\begin{smallmatrix}0\\0\\0\end{smallmatrix}\right],\left[\begin{smallmatrix}1\\0\\0\end{smallmatrix}\right],\left[\begin{smallmatrix}1\\0\\0\end{smallmatrix}\right],\left[\begin{smallmatrix}1\\1\\1\\1\end{smallmatrix}\right]\right) = \left\{\delta \in \mathbb{R}^3: \begin{array}{ccc} -\delta_1 + \delta_3 & \leq & 0\\ -\delta_2 + \delta_3 & \leq & 0\\ \delta_1 + \delta_2 - \delta_3 & \leq & 1 \end{array}\right\}$$

- >> V=struct('V',[0 0 0;0 1 0;1 0 0;1 1 1]);
- >> H=cddmex('hull',V);A=H.A,b=H.B

GEOMETRIC VS SYMBOLIC APPROACH

- The polyhedron obtained via convex hull is the smallest one
- The one obtained via CNF may be larger. Example:

$$(X_1 \lor X_2) \land (X_1 \lor X_3) \land (X_2 \lor X_3) = \mathtt{true}$$



Note: no other example with 3 vars but
 (X₁ ∨ X₂) ∧ (X₁ ∨ X₃) ∧ (X₂ ∨ X₃) ∧ (¬X₁ ∨ ¬X₂ ∨ ¬X₃) = true

BIG-M TECHNIQUE (IFF)

Consider the if-and-only-if condition

$$[\delta = 1] \leftrightarrow [a'x_c - b \le 0] \qquad x_c \in \mathcal{X}$$
$$\delta \in \{0, 1\}$$

• Assume $\mathcal{X} \subset \mathbb{R}^{n_c}$ bounded. Let M and m such that $\forall x_c \in \mathcal{X}$

$$\begin{array}{ccc} M & > & a'x_c - b \\ m & < & a'x_c - b \end{array}$$

• The if-and-only-if condition is equivalent to

$$\begin{cases} a'x_c - b & \leq M(1 - \delta) \\ a'x_c - b & > m\delta \end{cases}$$

• We can replace the second constraint with $a'x_c-b\geq \epsilon+(m-\epsilon)\delta$ to avoid strict inequalities, where $\epsilon>0$ is a small number (e.g., the machine precision)

COMPUTING THE BIG-M

• If ${\mathcal X}$ is a polyhedron, we can use linear programming

$$\begin{array}{lcl} m & < & \min_{x_c \in \mathcal{X}} \{a'x_c\} - b \\ \\ M & > & -\min_{x_c \in \mathcal{X}} \{-a'x_c\} - b \end{array}$$

• If $\mathcal X$ is a box, $\mathcal X=[\ell,u]$, we can use the following simpler method (Lee, Kouvaritakis, 2000) (Bemporad, 2022)

$$\begin{array}{rcl} a_{+} & = & \max\{a,0\} \\ \\ a_{-} & = & a_{+} - a = \max\{-a,0\} \\ \\ m & < & a'_{+}\ell - a'_{-}u - b \\ \\ M & > & a'_{+}u - a'_{-}\ell - b \end{array}$$

BIG-M TECHNIQUE (IF-THEN-ELSE)

Consider the if-then-else condition

$$z = \left\{ \begin{array}{ll} a_1' x_c - b_1 & \text{if } \delta = 1 \\ a_2' x_c - b_2 & \text{otherwise} \end{array} \right. \quad \left. \begin{array}{ll} x_c \in \mathcal{X} \\ \delta \in \{0, 1\} \\ z \in \mathbb{R} \end{array} \right.$$

• Assume $\mathcal{X}\subset\mathbb{R}^{n_c}$ bounded. Let M_1,M_2 and m_1,m_2 such that $\forall x_c\in\mathcal{X}$

$$\begin{array}{llll} M_1 & > & a_1' x_c - b_1 & > & m_1 \\ M_2 & > & a_2' x_c - b_2 & > & m_2 \end{array}$$

The if-then-else condition is equivalent to

$$\begin{cases}
(m_1 - M_2)(1 - \delta) + z & \leq a_1' x_c - b_1 \\
(m_2 - M_1)(1 - \delta) - z & \leq -(a_1' x_c - b_1) \\
(m_2 - M_1)\delta + z & \leq a_2' x_c - b_2 \\
(m_1 - M_2)\delta - z & \leq -(a_2' x_c - b_2)
\end{cases}$$

SWITCHED AFFINE SYSTEM

• The state-update equation of a SAS can be rewritten as

$$x_c(k+1) = \sum_{i=1}^s z_i(k) \quad z_i(k) \in \mathbb{R}^{n_c}$$

Switched Affine System

with

$$z_1(k) = \begin{cases} A_1 x_c(k) + B_1 u_c(k) + f_1 & \text{if } \delta_1(k) = 1\\ 0 & \text{otherwise} \end{cases}$$

:

$$z_s(k) \quad = \quad \left\{ \begin{array}{ll} A_s x_c(k) + B_s u_c(k) + f_s & \text{if } \delta_s(k) = 1 \\ 0 & \text{otherwise} \end{array} \right.$$

and with $\delta_i(k) \in \{0,1\}$ subject to the **exclusive or** condition

$$\sum_{i=1}^s \delta_i(k) = 1 \text{ or equivalently } \left\{ \begin{array}{ll} \sum_{i=1}^s \delta_i(k) & \geq & 1 \\ \sum_{i=1}^s \delta_i(k) & \leq & 1 \end{array} \right.$$

• Output eq. $y_c(k) = C_i x_c(k) + D_i u_c(k) + g_i$ admits a similar transformation

TRANSFORMATION OF A DHA INTO LINEAR (IN)EQUALITIES

$$X_1 \vee X_2 = \mathsf{TRUE} \qquad \qquad \delta_1 + \delta_2 \geq 1, \qquad \delta_1, \delta_2 \in \{0,1\}$$
 Any logic statement
$$f(X) = \mathsf{TRUE} \qquad \qquad \begin{cases} 1 \leq \sum\limits_{i \in P_1} \delta_i + \sum\limits_{i \in N_1} (1 - \delta_i) \\ \vdots \\ 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \end{cases}$$

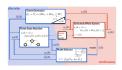
$$\vdots \\ 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \\ \vdots \\ 1 \leq \sum\limits_{i \in P_m} \delta_i + \sum\limits_{i \in N_m} (1 - \delta_i) \end{cases}$$

$$[\delta_e^i(k) = 1] \leftrightarrow [H^i x_c(k) \leq W^i] \qquad \begin{cases} H^i x_c(k) - W^i \leq M^i (1 - \delta_e^i(k)) \\ H^i x_c(k) - W^i > m^i \delta_e^i(k) \end{cases}$$

$$\mathsf{IF} \ [\delta = 1] \ \mathsf{THEN} \ = a_1^T x + b_1^T u + f_1 \\ \mathsf{ELSE} \ z = a_2^T x + b_2^T u + f_2 \end{cases} \qquad \begin{cases} (m_1 - M_2)(1 - \delta) + z \leq a_1 x + b_1 u + f_1 \\ (m_2 - M_1)(1 - \delta) - z \leq -a_1 x - b_1 u - f_1 \\ (m_2 - M_1)(1 - \delta) - z \leq -a_2 x + b_2 u + f_2 \\ (m_1 - M_2)\delta - z \leq -a_2 x - b_2 u - f_2 \end{cases}$$
 Finite State Machine

(Bemporad, Morari, 1999)

By converting logic relations into mixed-integer linear inequalities
 a DHA can be rewritten as the Mixed Logical Dynamical (MLD) system



$$\begin{cases} x(k+1) &= Ax(k) + B_1 u(k) + B_2 \delta(k) + B_3 z(k) + B_5 \\ y(k) &= Cx(k) + D_1 u(k) + D_2 \delta(k) + D_3 z(k) + D_5 \\ E_2 \delta(k) &+ E_3 z(k) \le E_4 x(k) + E_1 u(k) + E_5 \end{cases}$$



$$x \in \mathbb{R}^{n_c} \times \{0, 1\}^{n_b}, \ u \in \mathbb{R}^{m_c} \times \{0, 1\}^{m_b}$$

$$y \in \mathbb{R}^{p_c} \times \{0, 1\}^{p_b}, \ \delta \in \{0, 1\}^{r_b}, \ z \in \mathbb{R}^{r_c}$$

- The translation from DHA to MLD can be automatized, see e.g. the language HYSDEL (HYbrid Systems DEscription Language) (Torrisi, Bemporad, 2004)
- MLD models allow solving MPC, verification, state estimation, and fault detection problems via mixed-integer programming

A SIMPLE EXAMPLE OF MLD SYSTEM

- $\bullet \ \ \mathsf{PWA} \, \mathsf{system^1:} \qquad x(k+1) = \left\{ \begin{array}{rrr} 0.8x(k) + u(k) & \mathsf{if} & x(k) \geq 0 \\ -0.8x(k) + u(k) & \mathsf{if} & x(k) < 0 \end{array} \right.$
- Introduce event variable $[\delta(k)=1] \leftrightarrow [x(k)\geq 0]$ and use big-M technique:

$$x(k) \geq m(1-\delta(k)) \qquad \qquad M = -m = 10$$

$$x(k) \leq -\epsilon + (M+\epsilon)\delta(k) \qquad \qquad \epsilon > 0 \text{ "small"}$$

- Since $x(k+1) = 1.6\delta(k)x(k) - 0.8x(k) + u(k)$, introduce the aux variable

$$z(k) = \delta(k)x(k)$$

$$z(k) \leq M\delta(k)$$

$$z(k) \geq m\delta(k)$$

$$z(k) \leq x(k) - m(1 - \delta(k))$$

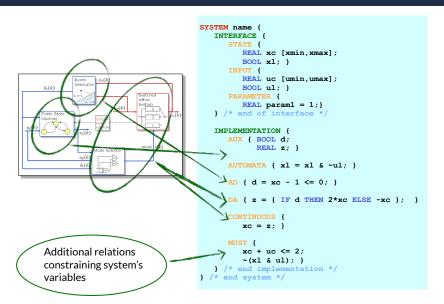
$$z(k) \geq x(k) - M(1 - \delta(k))$$

• Linear state update: x(k+1) = -0.8x(k) + 1.6z(k) + u(k)

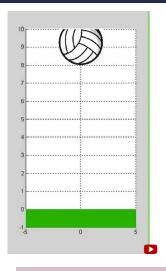
 $^{^{1}\}mbox{This}$ is the nonlinear system x(k+1)=0.8|x(k)|+u(k)

[&]quot;Model Predictive Control" - © 2023 A. Bemporad. All rights reserved.

DHA AND HYSDEL MODELS



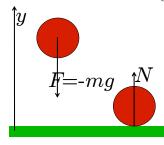
BOUNCING BALL EXAMPLE



$$\ddot{y} = -g$$

$$y \le 0 \Rightarrow \dot{y}(t^{+}) = -(1 - \alpha)\dot{y}(t^{-})$$

$$\alpha \in [0, 1]$$



How to model the bouncing ball as a discrete-time hybrid system?

BOUNCING BALL — TIME DISCRETIZATION

 $\bullet \ \, \mathsf{Case}\, y(k) > 0 \, \mathsf{(ball falling):} \ \, \begin{array}{c} v(k) & \approx & \frac{g(k) - g(k-1)}{T_s} \\ -g & \approx & \frac{v(k) - v(k-1)}{T_s} \end{array}$

$$\begin{cases} v(k+1) &= v(k) - T_s g \\ y(k+1) &= y(k) + T_s v(k+1) \\ &= y(k) + T_s v(k) - T_s^2 g \end{cases}$$

• Case $y(k) \leq 0$ (ground level): $v(k+1) = -(1-\alpha)v(k)$ $y(k+1) = y(k-1) = y(k) - T_s v(k)$

$$\begin{cases} v(k+1) &= -(1-\alpha)v(k) \\ y(k+1) &= y(k) - T_s v(k) \end{cases}$$

- We need a binary variable $[\delta(k)=1] \leftrightarrow [y(k) \leq 0]$

BOUNCING BALL - HYSDEL MODEL

```
SYSTEM bouncing ball {
INTERFACE {
/* Description of variables and constants */
        STATE { REAL height [-10.10];
                REAL velocity [-100,100]; }
        PARAMETER (
                REAL q;
                REAL alpha: /* 0=elastic. 1=completely anelastic */
                REAL Ts: }
IMPLEMENTATION (
        AUX { BOOL negative;
                REAL hnext:
                REAL vnext:
        AD I
              negative = height <= 0; }
        DA (
                hnext = { IF negative THEN height-Ts*velocity
                        ELSE height+Ts*velocity-Ts*Ts*q};
                vnext = { IF negative THEN -(1-alpha)*velocity
                        ELSE velocity-Ts*q}; }
        CONTINUOUS (
                height = hnext:
                velocity = vnext;}
}}
```

go to demo demos/hybrid/bball.m

BOUNCING BALL - SIMULATION

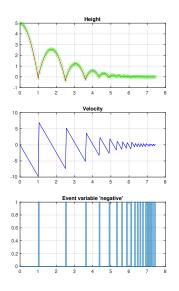
```
>> Ts=0.05;
>> g=9.8;
>> alpha=0.3;

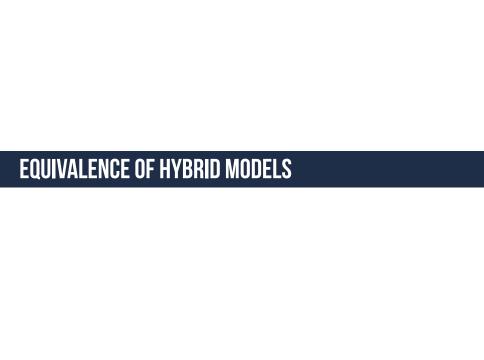
>> S=mld('bouncing_ball',Ts);

>> N=150;
>> U=zeros(N,0);
>> x0=[5 0]';

>> [X,T,D]=sim(S,x0,U);
```

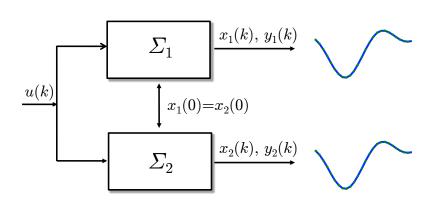
Note: no Zeno effect in discrete time!





EQUIVALENCE OF HYBRID MODELS

• Two hybrid models Σ_1, Σ_2 are **equivalent** if for all initial states $x_1(0) = x_2(0)$ and input excitations $u_1(k) \equiv u_2(k)$, the corresponding trajectories $x_1(k) \equiv x_2(k)$ and $y_1(k) \equiv y_2(k), \forall k = 0, 1, \dots$



EQUIVALENCE OF HYBRID MODELS

• MLD and PWA systems are equivalent (Bemporad, Ferrari-Trecate, Morari, 2000)

<u>Proof</u>: For a given combination (x_ℓ, u_ℓ, δ) of an MLD model, the state and output equation are linear and valid in a polyhedron.

Conversely, a PWA system can be modeled as MLD system (see next slide)

 Efficient conversion algorithms from MLD to PWA form exist (Bemporad, 2004) (Geyer, Torrisi, Morari, 2003)

 Further equivalences exist with other classes of hybrid dynamical systems, such as Linear Complementarity (LC) systems (Heemels, De Schutter, Bemporad, 2001)

MODELING A PWA SYSTEM IN MLD FORM

 $\bullet \;\; \mathsf{PWA}$ system with bounded states and inputs and s regions

$$\begin{array}{rcl} x(k+1) & = & A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)} \\ y(k) & = & C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)} \\ i(k) & = & \operatorname{such} \operatorname{that} \begin{bmatrix} x(k) \\ u(k) \end{bmatrix} \in \mathcal{C}_{i(k)} \end{array}$$

with
$$C_i = \{ \begin{bmatrix} x \\ u \end{bmatrix} : H_i x + J_i u \leq K_i \}$$
, and $\overset{\circ}{C_i} \cap \overset{\circ}{C_j} = \emptyset$, $\forall i \neq j, i, j = 1, \dots, s$ ($\{C_i\}$ is a polyhedral partition of the set $C \triangleq \cup_{i=1}^s C_i$)

• Introduce s binary variables $\delta_i, i=1,\ldots,s$ and the logic constraints

$$[\delta_i = 1] \to [H_i x + J_i u \le K_i]$$

$$\bigoplus_{i=1}^s [\delta_i = 1] = \texttt{true}$$

$$\sum_{i=1}^s \delta_i = 1$$

$$H_i x + J_i u \le K_i + M_i (1 - \delta_i)$$

were the vector M_i of upper-bounds can be computed, e.g., via LP

MODELING A PWA SYSTEM IN MLD FORM

• Introduce auxiliary real vectors z_i , w_i defined by if-then-else rules

$$z_i = \left\{ \begin{array}{ll} A_i x + B_i u + f_i & \text{if } \delta_i = 1 \\ 0 & \text{otherwise} \end{array} \right. \quad w_i = \left\{ \begin{array}{ll} C_i x + D_i u + g_i & \text{if } \delta_i = 1 \\ 0 & \text{otherwise} \end{array} \right.$$

and convert the relations above into mixed-integer inequalities

• Finally, write the state update and output equations

$$\begin{cases} x(k+1) &= \sum_{i=1}^{s} z_i(k) \\ y(k) &= \sum_{i=1}^{s} w_i(k) \end{cases}$$

MODELING PWA SYSTEMS BY DISJUNCTIVE PROGRAMMING

(Balas, 1985)

A PWA system with bounded states and inputs is equivalent to the disjunction

$$\bigvee_{i=1}^{s} \begin{bmatrix} H_i x(k) + J_i u(k) \le K_i \\ x(k+1) = A_i x(k) + B_i u(k) + f_i \end{bmatrix} \qquad x_{\ell b} \le x(k) \le x_{ub}$$
$$u_{\ell b} \le u(k) \le u_{ub}$$

- Introduce s binary variables $\delta_1(k),\dots,\delta_s(k)$ subject to $\sum_i \delta_i(k)=1$
- Introduce the convex hull relaxation of the disjunction

$$x(k) = \sum_{i=1}^{s} v_i(k), \quad x_{\ell b} \delta_i(k) \le v_i(k) \le x_{ub} \delta_i(k)$$
$$u(k) = \sum_{i=1}^{s} w_i(k), \quad u_{\ell b} \delta_i(k) \le w_i(k) \le u_{ub} \delta_i(k)$$

and impose

$$x(k+1) = \sum_{i=1}^{s} A_i v_i(k) + B_i w_i(k) + f_i \delta_i(k), \quad H_i v_i(k) + J_i w_i(k) \le K_i \delta_i(k)$$

MODELING PWA SYSTEMS WITHOUT AUX CONTINUOUS VARIABLES

(Bemporad, 2022)

• Only introduce s binary variables δ_i , $i=1,\ldots,s$ and set:

$$\underbrace{m_i^x(1 - \delta_i(k)) \le x(k+1) - A_i x(k) - B_i u(k) - f_i \le M_i^x(1 - \delta_i(k))}_{[\delta_i(k) = 1] \to [x(k+1) = A_i x(k) + B_i u(k) + f_i]} \\
m_i^y(1 - \delta_i(k)) \le y(k) - C_i x(k) - D_i u(k) - g_i \le M_i^y(1 - \delta_i(k))$$

$$H_i x(k) + J_i u(k) \le K_i + M_i (1 - \delta_i(k))$$
 $[\delta_i(k) = 1] \to [H_i x(k) + J_i u(k) \le K_i]$

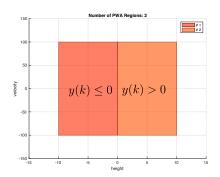
$$\sum_{i=1}^s \delta_i(k) = 1$$
 $\bigoplus_{i=1}^s [\delta_i(k) = 1] = \mathsf{true}$

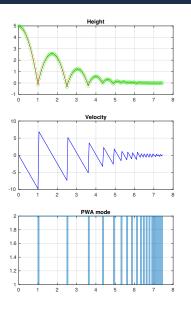
 $[\delta_i(k) = 1] \rightarrow [u(k) = C_i x(k) + D_i u(k) + q_i]$

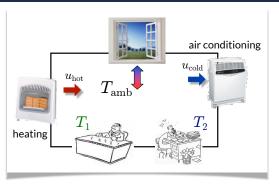
where $m_i^x, M_i^x, m_i^y, M_i^y$ are suitably defined upper and lower bounds

BOUNCING BALL - PWA EQUIVALENT

```
>> P=pwa(S);
>> plot(P)
>> [X,T,I]=sim(P,x0,U);
(Bemporad, 2004)
```







discrete dynamics

- #1 = cold → heater = on
- #2 = cold \rightarrow heater = on unless #1 hot
- A/C activation has similar rules

continuous dynamics

$$\frac{dT_i}{dt} = -\alpha_i(T_i - T_{\text{amb}}) + k_i(u_{\text{hot}} - u_{\text{cold}})$$

$$i = 1, 2$$

go to demo demos/hybrid/heatcool.m

```
SYSTEM heatcool {
INTERFACE {
   STATE { REAL T1 [-10.501:
           REAL T2 [-10,501: }
   INPUT { REAL Tamb [-50,501; }
   PARAMETER (
      REAL Ts, alphal, alpha2, k1, k2;
      REAL Thot1, Tcold1, Thot2, Tcold2, Uc, Uh; }
   }
IMPLEMENTATION {
   AUX { REAL uhot, ucold:
        BOOL hot1, hot2, cold1, cold2; }
   AD { hot1 = T1>=Thot1:
        hot2 = T2>=Thot2:
        cold1 = T1<=Tcold1:
        cold2 = T2<=Tcold2; }
   DA { uhot = { IF cold1 | (cold2 & ~hot1) THEN Uh ELSE 0 }:
        ucold = { IF hot1 | (hot2 & ~cold1) THEN Uc ELSE 0 }; }
   CONTINUOUS { T1 = T1+Ts*(-alpha1*(T1-Tamb)+k1*(uhot-ucold)):
                T2 = T2+Ts*(-alpha2*(T2-Tamb)+k2*(uhot-ucold)); }
```

>> S=mld('heatcoolmodel',Ts);

get the MLD model in MATLAB

>> [XX,TT]=sim(S,x0,U);

simulate the MLD model

MLD model of the room temperature system

$$\begin{cases} x(k+1) &= Ax(k) + B_1 u(k) + B_2 \delta(k) + B_3 z(k) + B_5 \\ y(k) &= Cx(k) + D_1 u(k) + D_2 \delta(k) + D_3 z(k) + D_5 \\ E_2 \delta(k) &+ E_3 z(k) \le E_4 x(k) + E_1 u(k) + E_5 \end{cases}$$

- 2 continuous states

(temperature T_1, T_2)

- 1 continuous input

(room temperature $T_{
m amb}$)

- 2 auxiliary continuous vars

(power flows $u_{\rm hot}$, $u_{\rm cold}$)

- 6 auxiliary binary vars

(4 threshold events + 2 for the OR condition)

20 mixed-integer inequalities

• In principle, we have $2^6=64$ possible combinations of binary variables

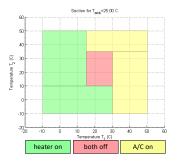
• PWA model of the room temperature system

$$\begin{array}{rcl} x(k+1) & = & A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)} \\ y(k) & = & C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)} \end{array}$$

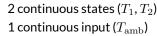
>> P=pwa(S);

Temperature T, (C)

$$i(k)$$
 s.t. $H_{i(k)}x(k) + J_{i(k)}u(k) \le K_{i(k)}$



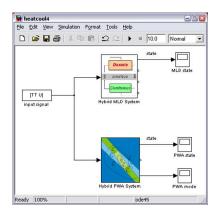
1

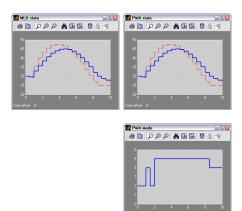


-20

Temperature T₂ (C)

5 polyhedral regions (partition does not depend on input)





• MLD and PWA models are equivalent, hence simulated states are the same

USING PWA EQUIVALENCE FOR MODEL ANALYSIS

- Assume plant + controller can be modeled as DHA:
 - plant = approximated as PWA system (e.g.: nonlinear switched model)
 - controller = switched linear controller (e.g: combination of threshold conditions, logic, linear feedback laws, ...)
- Convert DHA to MLD form, then to PWA form
- The resulting closed-loop PWA model reveals how the closed-loop system behaves in different regions of the state-space
- Can analyze closed-loop stability analysis using piecewise quadratic Lyapunov functions (Johansson, Rantzer, 1998) (Mignone, Ferrari-Trecate, Morari, 2000)

OTHER EXISTING HYBRID MODELS

(Heemels, De Schutter, Bemporad, 2001)

Linear complementarity (LC) systems (Heemels, 1999)

$$x(k+1) = Ax(k) + B_1u(k) + B_2w(k)$$

$$y(k) = Cx(k) + D_1u(k) + D_2w(k)$$

$$v(k) = E_1x(k) + E_2u(k) + E_3w(k) + E_4$$

$$0 \le v(k) \perp w(k) \ge 0$$

Examples: mechanical systems, electrical circuits



• Min-max-plus-scaling (MMPS) systems (De Schutter, Van den Boom, 2000)

$$x(k+1) = M_x(x(k), u(k), w(k))$$

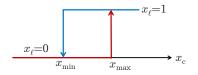
 $y(k) = M_y(x(k), u(k), w(k))$
 $0 \ge M_c(x(k), u(k), w(k))$

Example: discrete-event system k = event counter

where $M_{()}$ are MMPS functions defined by the grammar $M:=x_i|\alpha|\max(M_1,M_2)|\min(M_1,M_2)|M_1+M_2|\beta M_1$

Example:
$$x(k+1) = 2\max(x(k),0) + \min(\frac{1}{2}u(k),1)$$

MODELING HYSTERESIS



- Hysteresis between $x_{\min} \le x_c(k) \le x_{\max}$
- Introduce two binary variables

$$[\delta_{\min}(k) = 1] \quad \leftrightarrow \quad [x_c(k) \le x_{\min}]$$

$$[\delta_{\max}(k) = 1] \quad \leftrightarrow \quad [x_c(k) \ge x_{\max}]$$

• Introduce logic state $x_{\ell} \in \{0,1\}$ with dynamics

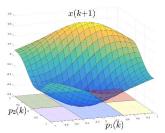
$$x_{\ell}(k+1) = (x_{\ell}(k) \land \neg \delta_{\min}(k)) \lor (\neg x_{\ell}(k) \land \delta_{\max}(k))$$



HYBRID SYSTEM IDENTIFICATION

- A hybrid model of the process may not be available from physical principles
- Therefore, a model must be either
 - estimated from data (model is unknown)
 - or hybridized (model is known but nonlinear)
- If one linear model is enough: easy problem (SYS-ID TBX) (Ljung, 1999)
- If switching sequence known: easy, just identify one linear model per mode
- If modes & dynamics must be identified simultaneously, we need hybrid system identification (or piecewise affine regression)

In industrial MPC most effort is spent in identifying (multiple) linear prediction models from data

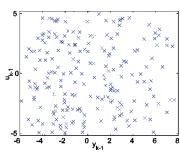


LEARNING PWA MODELS FROM DATA

Estimate from data **both** the **parameters** of the affine submodels and the **partition** of the PWA map

Example: Let the data be generated by the PWARX system

$$y_k = \left\{ \begin{array}{ll} \left[\begin{array}{ccc} -0.4 & 1 & 1.5 \end{array} \right] \phi_k + \epsilon_k \\ \text{if} \left[\begin{array}{ccc} 4 & -1 & 10 \end{array} \right] \phi_k < 0 \\ \left[\begin{array}{ccc} 0.5 & -1 & -0.5 \end{array} \right] \phi_k + \epsilon_k \\ \text{if} \left[\begin{array}{ccc} -4 & 1 & 10 \\ 5 & 1 & -6 \end{array} \right] \phi_k \leq 0 \\ \left[\begin{array}{ccc} -0.3 & 0.5 & -1.7 \end{array} \right] \phi_k + \epsilon_k \\ \text{if} \left[\begin{array}{ccc} -5 & -1 & 6 \end{array} \right] \phi_k < 0 \end{array} \right.$$



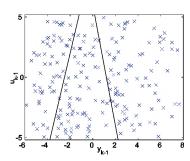
with
$$\phi_k = [y_{k-1} \, u_{k-1} \, 1]'$$
, $|u_k| \le 5$, and $|\epsilon_k| \le 0.1$

PWA IDENTIFICATION PROBLEM

Estimate from data **both** the **parameters** of the affine submodels and the **partition** of the PWA map

Example: Let the data be generated by the PWARX system

$$y_k = \left\{ \begin{array}{l} \left[-0.4 \quad 1 \quad 1.5 \right] \phi_k + \epsilon_k \\ \text{if} \left[4 \quad -1 \quad 10 \right] \phi_k < 0 \\ \left[0.5 \quad -1 \quad -0.5 \right] \phi_k + \epsilon_k \\ \text{if} \left[-4 \quad 1 \quad 10 \right] \phi_k \leq 0 \\ \left[-4 \quad 1 \quad 10 \right] \phi_k \leq 0 \\ \left[-4 \quad 1 \quad 10 \right] \phi_k \leq 0 \\ \left[-4 \quad 1 \quad 10 \right] \phi_k \leq 0 \end{array} \right.$$

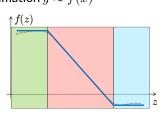


with
$$\phi_k = [y_{k-1} \ u_{k-1} \ 1]'$$
, $|u_k| \le 5$, and $|\epsilon_k| \le 0.1$

PWA REGRESSION PROBLEM

• **Problem**: Given input/output pairs $\{x(k),y(k)\}$, $k=1,\ldots,N$ and number s of models, compute a **piecewise affine** (PWA) approximation $y\approx f(x)$

$$v(k) = \begin{cases} F_1 z(k) + g_1 & \text{if } H_1 z(k) \leq K_1 \\ \vdots \\ F_s z(k) + g_s & \text{if } H_s z(k) \leq K_s \end{cases}$$
$$v(k) = \begin{bmatrix} x(k+1) \\ y(k) \end{bmatrix}, \quad z(k) = \begin{bmatrix} x(k) \\ u(k) \end{bmatrix}$$



- Need to learn **both** the parameters $\{F_i,\,g_i\}$ of the affine submodels **and** the partition $\{H_i,\,K_i\}$ of the PWA map from data (offline learning)
- Possibly update model+partition as new data are available (online learning)
- Any ML technique can be applied that leads to PWA models, such as (leaky)ReLU-NNs, decision trees, softmax regression, KNN, ...

APPROACHES TO PWA IDENTIFICATION

- Mixed-integer linear or quadratic programming (Roll, Bemporad, Ljung, 2004)
- Partition of infeasible set of inequalities (Bemporad, Garulli, Paoletti, Vicino, 2005)
- K-means clustering in a feature space (Ferrari-Trecate, Muselli, Liberati, Morari, 2003)
- Bayesian approach (Juloski, Wieland, Heemels, 2004)
- Kernel-based approaches (Pillonetto, 2016)
- Hyperplane clustering in data space (Münz, Krebs, 2002)
- Recursive multiple least squares & PWL separation (Breschi, Piga, Bemporad, 2016)
- Piecewise affine regression and classification (PARC) (Bemporad, 2022)

1. Estimate models $\{F_i, g_i\}$ recursively. Let $e_i(k) = y(k) - F_i x(k) - g_i$ and only update model i(k) such that

$$i(k) \leftarrow \arg\min_{i=1,\dots,s} \underbrace{e_i(k)' \Lambda_e^{-1} e_i(k)}_{\text{one-step prediction error}} + \underbrace{(x(k) - c_i)' R_i^{-1} (x(k) - c_i)}_{\text{proximity to centroid}}$$
of cluster #i

using recursive LS and inverse QR decomposition (Alexander, Ghirnikar, 1993)

This also splits the data points x(k) in clusters $C_i = \{x(k) : i(k) = i\}$

2. Compute a polyhedral partition $\{H_i, K_i\}$ of the regressor space via multi-category linear separation

$$\phi(x) = \max_{i=1,\dots,s} \{w_i' x - \gamma_i\}$$



PWA REGRESSION EXAMPLES

(Breschi, Piga, Bemporad, 2016)

Identification of piecewise-affine ARX model

$$\begin{split} \begin{bmatrix} y_1(k) \\ y_2(k) \end{bmatrix} &= \begin{bmatrix} -0.83 & 0.20 \\ 0.30 & -0.52 \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix} + \begin{bmatrix} -0.34 & 0.45 \\ -0.30 & 0.24 \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix} \\ &+ \begin{bmatrix} 0.20 \\ 0.15 \end{bmatrix} + \max \left\{ \begin{bmatrix} 0.20 & -0.90 \\ 0.10 & -0.42 \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix} \right. \\ &+ \begin{bmatrix} 0.42 & 0.20 \\ 0.50 & 0.64 \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix} + \begin{bmatrix} 0.40 \\ 0.30 \end{bmatrix}, \begin{bmatrix} 0 \\ 0 \end{bmatrix} \right\} + e_0(k), \end{split}$$

• Quality of fit: best fit rate (BFR) = $\max\Big\{1-\frac{\|y_{\mathrm{o},i}-\hat{y}_i\|_2}{\|y_{\mathrm{o},i}-\bar{y}_{\mathrm{o},i}\|_2},0\Big\}$, i=1,2 y_{o} = measured, \hat{y} = open-loop simulated, \bar{y} = sample mean of y_{o}

			N = 4000	N = 20000	N = 100000
	y_1	(offline) RLP	96.0 %	96.5 %	99.0 %
		(Offline) RPSN	96.2 %	96.4 %	98.9 %
		(Online) ASGD	86.7 %	95.0 %	96.7 %
ĺ	y_2	(offline) RLP	96.2 %	96.9 %	99.0 %
ı		(offline) RPSN	96.3 %	96.8 %	99.0 %
		(online) ASGD	87.4 %	95.2 %	96.4 %

BFR on validation data, open-loop validation

• CPU time for computing the partition: (i7 2.40-GHz Intel core)

RLP = Robust linear programming

(Bennett, Mangasarian, 1994)

RPSN = Piecewise-smooth Newton method (Bemporad, Bernardini, Patrinos, 2015)

ASGD = Averaged stochastic gradient descent (Bottou, 2012)

	N = 4000	N = 20000	N = 100000
(Offline) RLP	0.308 s	3.227 s	112.435 s
(Offline) RPSN	0.016 s	0.086 s	0.365 s
(Online) ASGD	0.013 s	0.023 s	0.067 s

Identification of linear parameter varying ARX model

$$\begin{bmatrix} y_1(k) \\ y_2(k) \end{bmatrix} = \begin{bmatrix} \bar{a}_{1,1}(p(k)) \ \bar{a}_{1,2}(p(k)) \\ \bar{a}_{2,1}(p(k)) \ \bar{a}_{2,2}(p(k)) \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix}$$

$$+ \begin{bmatrix} \bar{b}_{1,1}(p(k)) \ \bar{b}_{1,2}(p(k)) \\ \bar{b}_{2,1}(p(k)) \ \bar{b}_{2,2}(p(k)) \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix} + e_0(k)$$

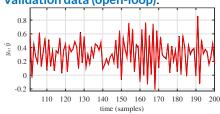
 $\bar{a}(p)$ = PWA function of p $\bar{b}(p)$ has quadratic and sin terms

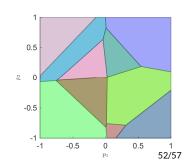
Quality of fit (BFR):

	y_1	y_2
PWA regression	87 %	84 %
parametric LPV*	80 %	70 %

(Bamieh, Giarré, 2002)

Validation data (open-loop):



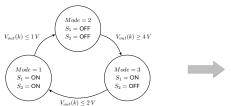


"Model Predictive Control" - © 2023 A. Bemporad. All rights reserved.

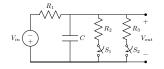
IDENTIFICATION OF HYBRID SYSTEMS WITH LOGIC STATES

(Breschi, Piga, Bemporad, 2016)

Identification of a hybrid model with logic states



true system

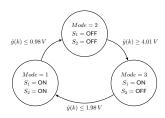




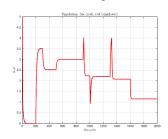
Quality of fit: BFR=96.64 % (validation)

CPU time for identification: 78 ms

(2000 samples, MacBook Pro 2.8 GHz)



identified system



PARC - PIECEWISE AFFINE REGRESSION AND CLASSIFICATION

(Bemporad, 2022)

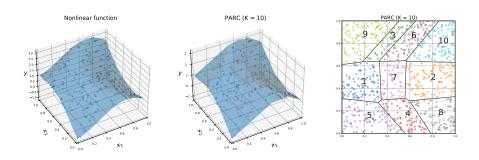
- New Piecewise Affine Regression and Classification (PARC) algorithm
- Training dataset:
 - feature vector $z \in \mathbb{R}^n$ (categorical features one-hot encoded in $\{0,1\}$)
 - target vector $v_c \in \mathbb{R}^{m_c}$ (numeric), $v_{di} \in \{w_{di}^1, \dots, w_{di}^{m_i}\}$ (categorical)
- PARC iteratively clusters training data in K sets and fits linear predictors:
 - 1. fit $v_c = a_j z + b_j$ by ridge regression (= ℓ_2 -regularized least squares)
 - 2. fit $v_{di}=w_{di}^{h_*}$, $h_*=\arg\max\{a_{dih}^hz+b_{di}^h\}$ by softmax regression
 - 3. fit a convex PWL separation function by softmax regression

$$\Phi(z) = \omega^{j(z)} z + \gamma^{j(z)}, \qquad j(z) = \min \left\{ \arg \max_{j=1,\dots,K} \{ \omega^j z + \gamma^j \} \right\}$$

- Data reassigned to clusters based on weighted fit/PWL separation criterion
- PARC is a block-coordinate descent algorithm ⇒ (local) convergence ensured

PARC - PIECEWISE AFFINE REGRESSION AND CLASSIFICATION

- Simple PWA regression example:
 - 1000 samples of $y = \sin(4x_1 5(x_2 0.5)^2) + 2x_2$ (use 80% for training)
 - Look for PWA approximation over K=10 polyhedral regions



Code download:



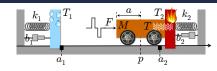
http://cse.lab.imtlucca.it/~bemporad/parc/

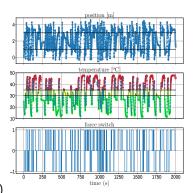
PARC - CART & BUMPERS EXAMPLE

 Example: moving cart and bumpers + heat transfer during bumps.

Spring and viscous forces are nonlinear.

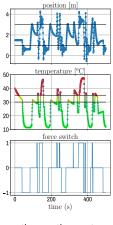
- Categorical input $F \in \{-\bar{F}, 0, \bar{F}\}$ and categorical output $c \in \{green, \underline{yellow}, \underline{red}\}$
- Continuous-time system simulated for 2,000 s, sample time = 0.5 s (=4000 training samples)
- Feature vector $z_k = [y_k, \dot{y}_k, T_k, F_k]$
- $\bullet \ \ \mathsf{Target} \, \mathsf{vector} \, v_k = [y_{k+1}, \dot{y}_{k+1}, T_{k+1}, c_k]$
- Hybrid model learned by PARC (K=5 regions)



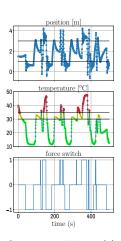


PARC - CART & BUMPERS EXAMPLE

Open-loop simulation on 500 s test data:



continuous-time system



discrete-time PWA model

Model fit is good enough for MPC design purposes (see later ...)