Control Systems

Subject Code: EC380

Transient & Steady State Response Analysis

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Introduction

The time response of a control system consists of two parts:





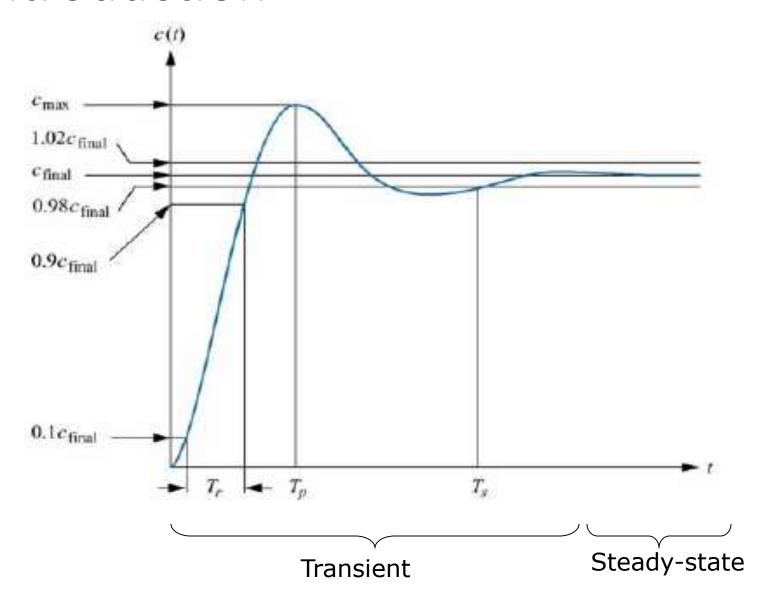
1. Transient response

 from initial state to the final state – purpose of control systems is to provide a desired response.

2. Steady-state response

- the manner in which the system output behaves as *t* approaches infinity – the error after the transient response has decayed, leaving only the continuous response.

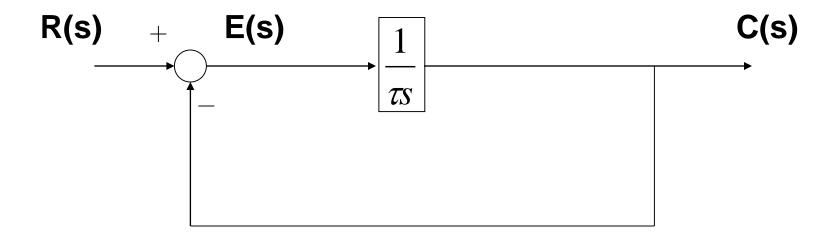
Introduction



Performances of Control Systems

- Specifications (time domain)
 - Max OS, settling time, rise time, peak time,
- Standard input signals used in design
 - actual signals unknown
 - standard test signals:
 - step, ramp, parabola, impulse, etc. sinusoid
 - (study freq. response later)
- Transient response
- Steady-state response
- Relate to locations of poles and zeros

First Order System



Test signal is step function, R(s)=1/s

A first-order system without zeros can be represented by the following transfer function

$$\frac{C(s)}{R(s)} = \frac{1}{\tau s + 1}$$

 Given a step input, i.e., R(s) = 1/s, then the system output (called step response in this case) is

$$C(s) = \frac{1}{\tau s + 1} R(s) = \frac{1}{s(\tau s + 1)} = \frac{1}{s} - \frac{1}{s + \frac{1}{\tau}}$$

Taking inverse Laplace transform, we have the step response

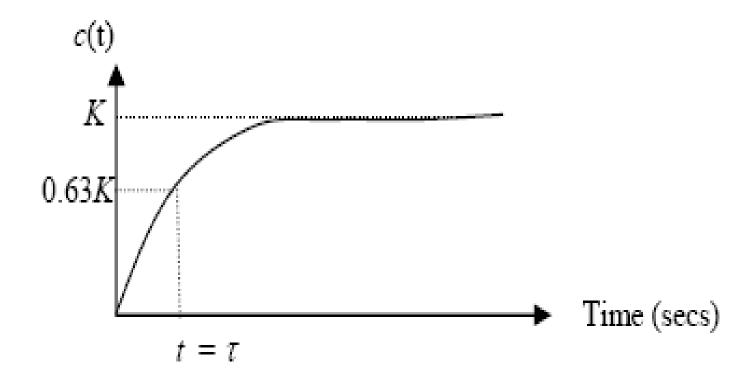
$$c(t) = 1 - e^{-\frac{t}{\tau}}$$

Time Constant: If $t = \tau$, So the step response is

$$C(T) = (1 - 0.37) = 0.63$$

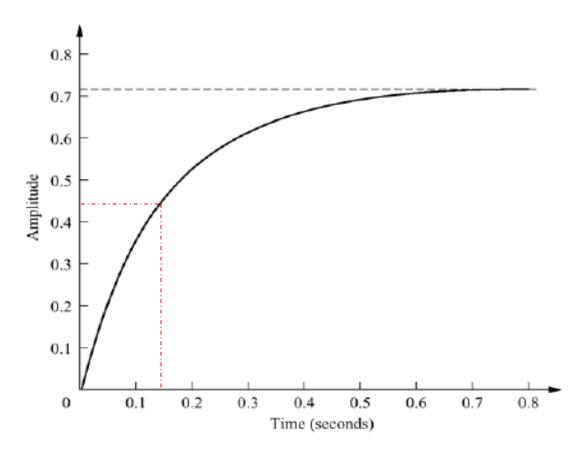
 τ is referred to as the **time constant** of the response. In other words, the time constant is the time it takes for the step response to rise to 63% of its final value. Because of this, the time constant is used to measure how fast a system can respond. The time constant has a unit of seconds.

Plot c(t) versus time:



Example 1

The following figure gives the measurements of the step response of a first-order system, find the transfer function of the system.



First – order system Transient Response Analysis

Rise Time T_r :

The rise-time (symbol T_r units s) is defined as the time taken for the step response to go from **10% to 90%** of the final value.

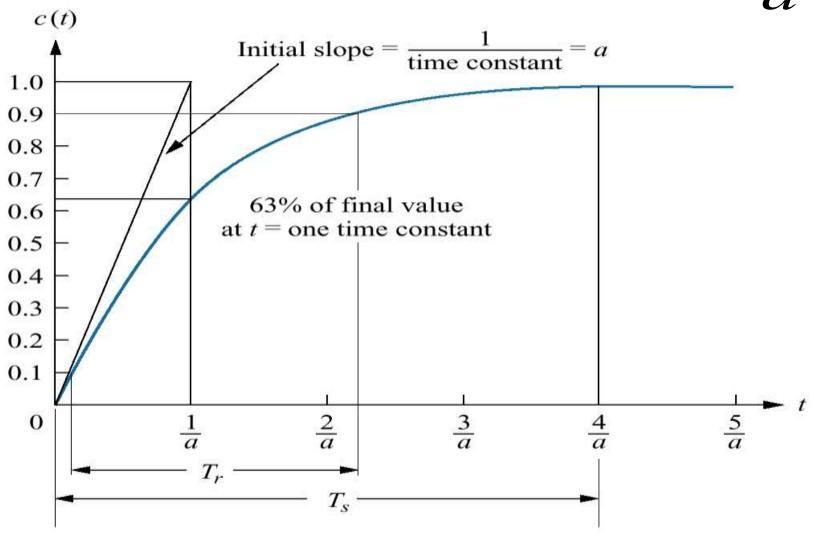
$$T_r = 2.31\tau - 0.11\tau = 2.2\tau$$

Settling Time *Ts***:**

Defined the settling-time (symbol T_s units s) to be the time taken for the step response to come to within **2% of the final value** of the step response.

$$T_s = 4\tau$$

$$\tau = \frac{1}{\alpha}$$



- Second-order systems exhibit a wide range of responses which must be analyzed and described.
 - Whereas for a *first-order system*, varying a single parameter changes the speed of response, changes in the parameters of a *second order* system can change the form of the response.
 - For example: a second-order system can display characteristics much like a first-order system or, depending on component values, display damped or pure oscillations for its transient response.

- A general second-order system is characterized by the following transfer function:

$$G(s) = \frac{b}{s^2 + as + b}$$

- We can re-write the above transfer function in the following form (closed loop transfer function):

$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$\omega_n \quad (\omega_n = \sqrt{b})$$

- referred to as the un-damped natural frequency of the second order system, which is the frequency of oscillation of the system without damping.

$$\zeta \left(\zeta = \frac{a}{2\sqrt{b}} \right)$$

 ζ ($\zeta = \frac{a}{2\sqrt{b}}$) second order system, which is a measure of the degree of residual the degree of resistance to change in the system output.

Poles;
$$-\omega_{n}\zeta + \omega_{n}\sqrt{\zeta^{2} - 1}$$
$$-\omega_{n}\zeta - \omega_{n}\sqrt{\zeta^{2} - 1}$$

Poles are complex if ζ < 1!

- According the value of ζ , a second-order system can be set into one of the four categories:
 - 1. Overdamped when the system has two real distinct poles ($\zeta > 1$).
 - 2. *Underdamped* when the system has two complex conjugate poles $(0 < \zeta < 1)$
 - 3. *Undamped* when the system has two imaginary poles ($\zeta = 0$).
 - 4. *Critically damped* when the system has two real but equal poles ($\zeta = 1$).

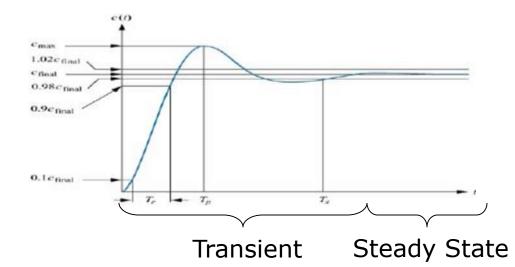
Time-Domain Specification

Given that the closed loop TF

$$T(s) = \frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\varsigma\omega_n s + \omega_n^2}$$

The system (2nd order system) is parameterized by ς and ω_n

For $0 < \varsigma < 1$ and $\omega_n > 0$, we like to investigate its response due to a unit step input



Two types of responses that are of interest:

- (A)Transient response
- (B)Steady state response

(A) For transient response, we have 4 specifications:

(a)
$$T_r$$
 - rise time = $\frac{\pi}{\omega_n \sqrt{1-\varsigma^2}}$

(b)
$$T_p$$
 - peak time = $\frac{\pi}{\omega_n \sqrt{1-\varsigma^2}}$

(c) %MP – percentage maximum overshoot =
$$e^{-\frac{\pi\varsigma}{\sqrt{1-\varsigma^2}}}$$
 $x100\%$

(d)
$$T_s$$
 – settling time (2% error) = $\frac{4}{\varsigma \omega_n}$

(B) Steady State Response

(a) Steady State error

Question : How are the performance related to ς and ω_n ?

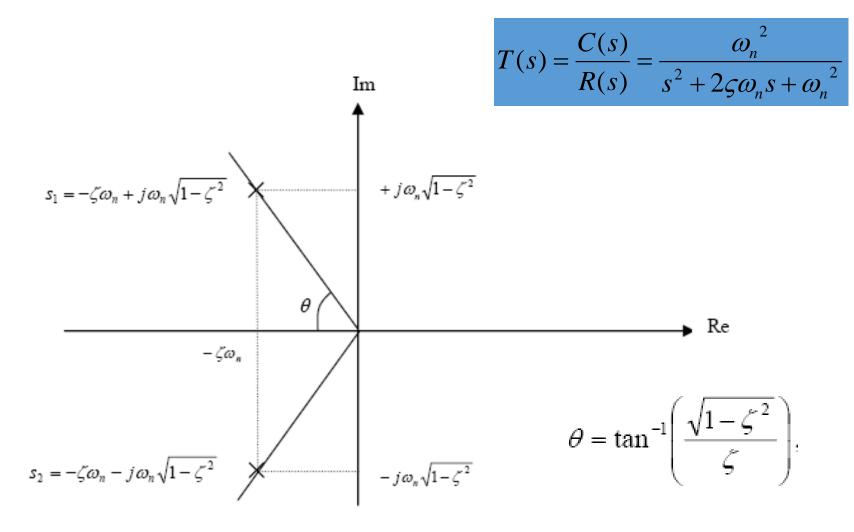
- Given a step input, i.e., R(s) = 1/s, then the system output (or step response) is;

$$C(s) = R(s)G(s) = \frac{\omega_n^2}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}$$

- Taking inverse Laplace transform, we have the step response;

$$c(t) = 1 - \frac{1}{\sqrt{1 - \zeta^2}} e^{-\zeta \omega_n t} \sin(\omega_n \sqrt{1 - \zeta^2} t + \theta)$$

Where;
$$\theta = \tan^{-1} \left(\frac{\sqrt{1-\zeta^2}}{\zeta} \right)$$
 or $\theta = \cos^{-1}(\xi)$



Mapping the poles into s-plane

Lets re-write the equation for c(t):

Let:
$$\beta = \sqrt{1-\xi^2}$$
 and
$$\omega_d = \omega_n \sqrt{1-\xi^2}$$
 Damped natural frequency
$$\omega_n > \omega_d$$

Thus:
$$c(t) = 1 - \frac{1}{\beta} e^{-\xi \omega_n t} \sin(\omega_d t + \theta)$$
 where $\theta = \cos^{-1}(\xi)$

Transient Response Analysis

1) Rise time, Tr. Time the response takes to rise from 0 to 100%

$$c(t)\big|_{t=T_r} = 1 - \frac{1}{\beta} e^{-\xi \omega_n t} \sin(\omega_d t + \theta) = 1$$

$$\neq 0 \qquad = 0$$

$$\sin(\omega_d T_r + \theta) = 0$$

$$\omega_d T_r + \theta = \sin^{-1}(0) = \pi$$

$$T_r = \frac{\pi - \theta}{\omega_n \sqrt{1 - \xi^2}}$$

Transient Response Analysis

2) Peak time, T_p - The peak time is the time required for the response to reach the first peak, which is given by;

$$\begin{aligned} y(t) \Big|_{t=T_p} &= 0 \\ y(t) \Big|_{t=T_p} &= -\frac{1}{\beta} (-\varsigma \omega_n) e^{-\varsigma \omega_n t} \sin(\omega_d t + \theta) - \frac{1}{\beta} e^{-\varsigma \omega_n t} \cos(\omega_d t + \theta) \Big[\omega_n \sqrt{1 - \varsigma^2} \Big] = 0 \\ &\frac{\varsigma \omega_n}{\beta} e^{-\varsigma \omega_n T_p} \sin(\omega_d T_p + \theta) = \frac{\left[\omega_n \sqrt{1 - \varsigma^2} \right]}{\beta} e^{-\varsigma \omega_n T_p} \cos(\omega_d T_p + \theta) \\ &\frac{1}{\beta} e^{-\varsigma \omega_n T_p} \sin(\omega_d T_p + \theta) = \frac{\left[\omega_n \sqrt{1 - \varsigma^2} \right]}{\beta} e^{-\varsigma \omega_n T_p} \cos(\omega_d T_p + \theta) \end{aligned}$$

$$\tan(\omega_d T_p + \theta) = \frac{\sqrt{1-\zeta^2}}{\zeta}$$

$$s_1 = -\zeta \omega_n + j\omega_n \sqrt{1-\zeta^2}$$

$$-\zeta \omega_n$$

$$s_2 = -\zeta \omega_n - j\omega_n \sqrt{1-\zeta^2}$$

$$-j\omega_n \sqrt{1-\zeta^2}$$

We know that $\tan(\theta) = \tan(\pi + \theta)$ So, $\tan(\omega_d T_p + \theta) = \tan(\pi + \theta)$

From this expression:

$$\omega_d T_p + \theta = \pi + \theta$$
$$\omega_d T_p = \pi$$

$$T_{p} = \frac{\pi}{\omega_{d}} = \frac{\pi}{\omega_{n} \sqrt{1 - \varsigma^{2}}}$$

Transient Response Analysis

3) Percent overshoot, %OS - The percent overshoot is defined as the amount that the waveform at the peak time overshoots the steady-state value, which is expressed as a percentage of the steady-state value.

$$\%MP = \frac{C(T_p) - C(\infty)}{C(\infty)} x 100\%$$

OR

$$\%OS = \frac{C \max - Cfinal}{Cfinal} \times 100$$

$$\frac{C(T_p)-1}{1}x100\% = -\frac{1}{\beta}e^{-\xi\omega_n t}\sin(\omega_d t + \theta)x100\%$$

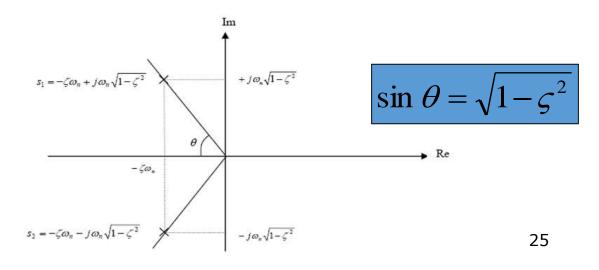
$$= -\frac{1}{\beta}e^{-\xi\omega_n \left[\frac{\pi}{\omega_n\sqrt{1-\varsigma^2}}\right]}\sin(\omega_d \left(\frac{\pi}{\omega_d}\right) + \theta)x100\%$$

$$= -\frac{1}{\beta}e^{-\frac{\pi\varsigma}{\sqrt{1-\varsigma^2}}}\sin(\pi + \theta)x100\%$$

$$= \frac{\sin(\theta)}{\beta}e^{-\frac{\pi\varsigma}{\sqrt{1-\varsigma^2}}}x100\% = e^{-\frac{\pi\varsigma}{\sqrt{1-\varsigma^2}}}x100\%$$

From slide 24

$$\beta = \sqrt{1 - \xi^2}$$



Therefore,

$$\% MP = e^{-\frac{\pi\varsigma}{\sqrt{1-\varsigma^2}}} x100\%$$

- For given %OS, the damping ratio can be solved from the above equation;

$$\varsigma = \frac{-\ln(\%MP/100)}{\sqrt{\pi^2 + \ln^2(\%MP/100)}}$$

Transient Response Analysis

4) Setting time, T_s - The settling time is the time required for the amplitude of the sinusoid to decay to 2% of the steady-state value.

To find T_s , we must find the time for which c(t) reaches & stays within $\pm 2\%$ of the steady state value, $c_{final.}$ The settling time is the time it takes for the amplitude of the decaying sinusoid in c(t) to reach 0.02, or

$$e^{-\varsigma\omega_n T_s} \frac{1}{\sqrt{1-\varsigma^2}} = 0.02$$

Thus,

$$T_{s} = \frac{4}{\varsigma \omega_{n}}$$

UNDERDAMPED

Example 2: Find the natural frequency and damping ratio for the system with transfer function

$$G(s) = \frac{36}{s^2 + 4.2s + 36}$$

Solution:

Compare with general TF_

$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$
•\omega n = 6
•\xi = 0.35

UNDERDAMPED

Example 3: Given the transfer function

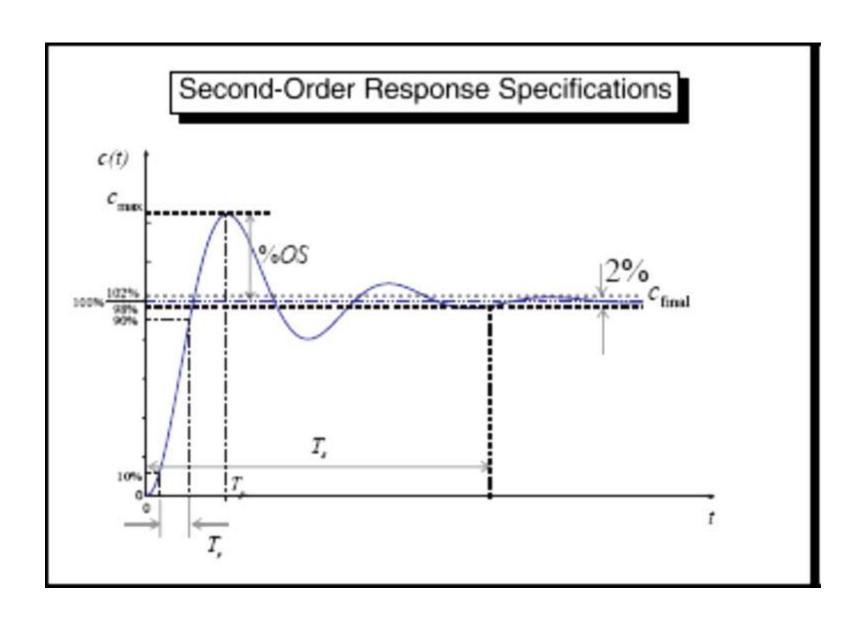
$$G(s) = \frac{100}{s^2 + 15s + 100}$$

find T_s , %OS, T_p

 $T_s = 0.533 s$, %OS = 2.838%, $T_p = 0.475 s$ Solution:

$$\omega_n = 10$$
 $\xi = 0.75$

UNDERDAMPED



$$G(s) = \frac{b}{s^2 + as + b}$$

Overdamped Response

a = 9

Overdamped system

$$R(s) = \frac{1}{s}$$

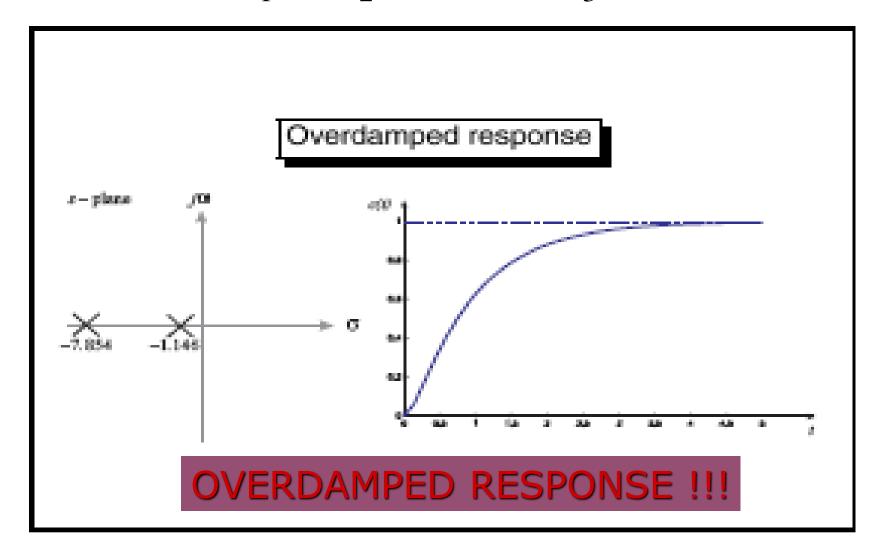
$$s^{2} + 9s + 9$$

$$2 \text{ poles. No zeros.}$$

$$C(s) = \frac{9}{s(s^2 + 9s + 9)} = \frac{9}{s(s + 7.854)(s + 1.146)}$$

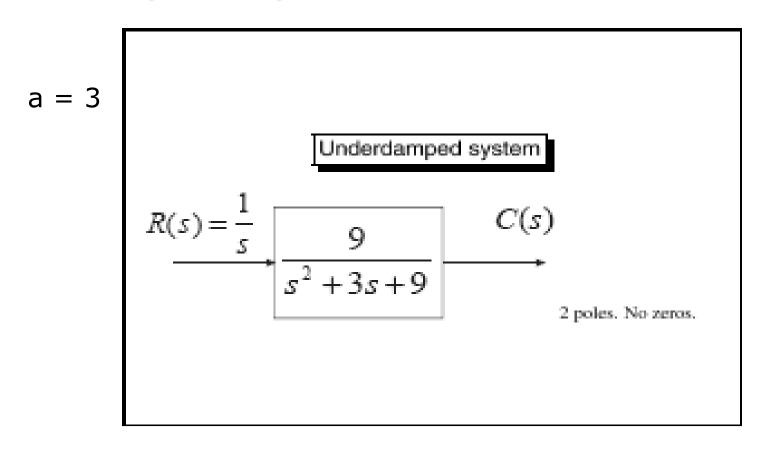
s = 0; s = -7.854; s = -1.146 (two real poles)

$$c(t) = K_1 + K_2 e^{-7.854t} + K_3 e^{-1.146t}$$



$$G(s) = \frac{b}{s^2 + as + b}$$

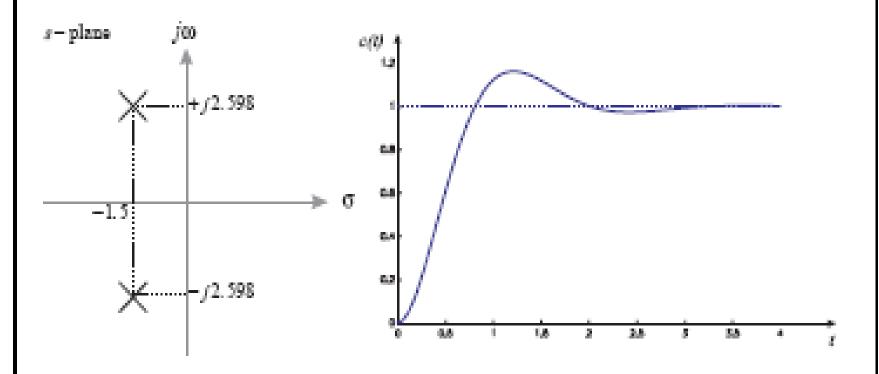
<u>Underdamped Response</u>



$$c(t) = K_1 + e^{-1.5t} (K_2 \cos 2.598t + K_3 \sin 2.598t)$$

$$s = 0$$
; $s = -1.5 \pm j2.598$ (two complex poles)

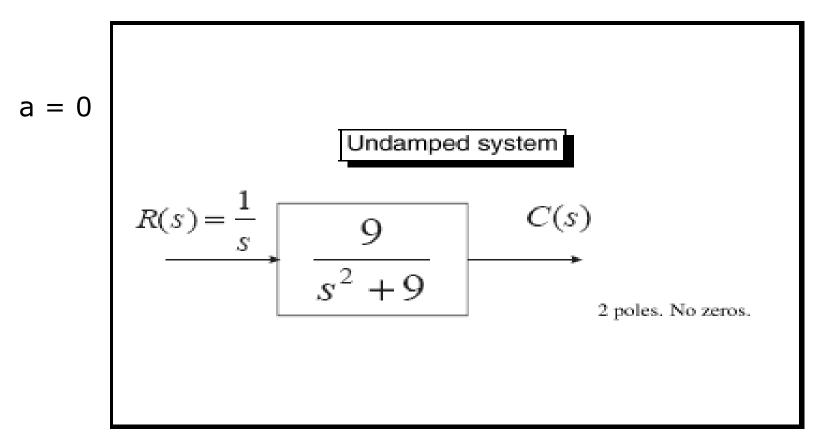
Underdamped response



UNDERDAMPED RESPONSE !!!

Undamped Response

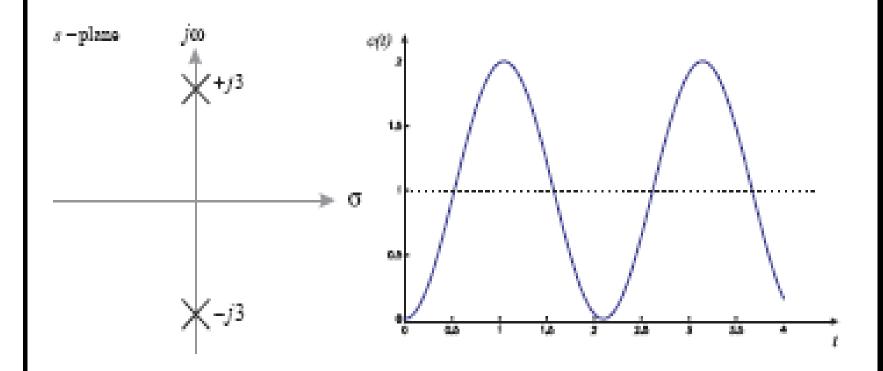
$$G(s) = \frac{b}{s^2 + as + b}$$



$$c(t) = K_1 + K_2 \cos 3t$$

s = 0; $s = \pm j3$ (two imaginary poles)

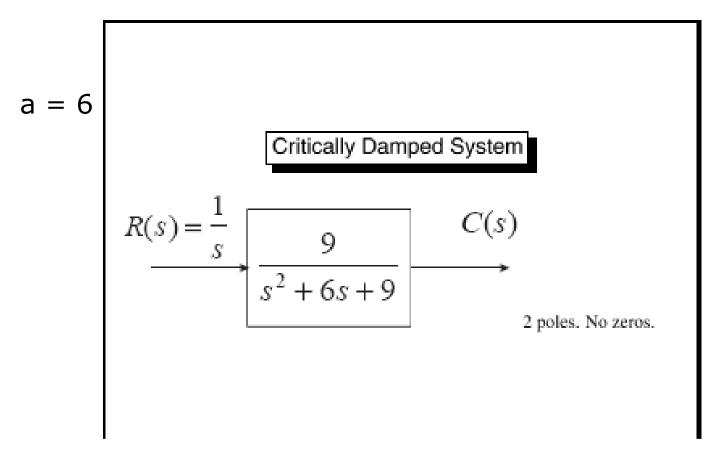
Undamped response



UNDAMPED RESPONSE !!!

Critically Damped System

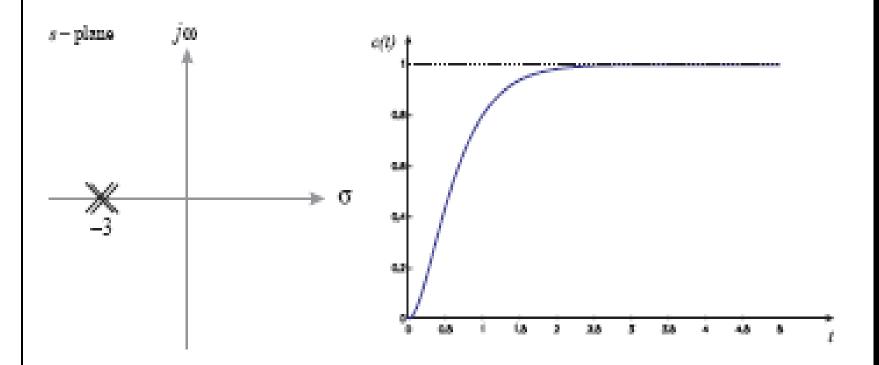
$$G(s) = \frac{b}{s^2 + as + b}$$



$$c(t) = K_1 + K_2 e^{-3t} + K_3 t e^{-3t}$$

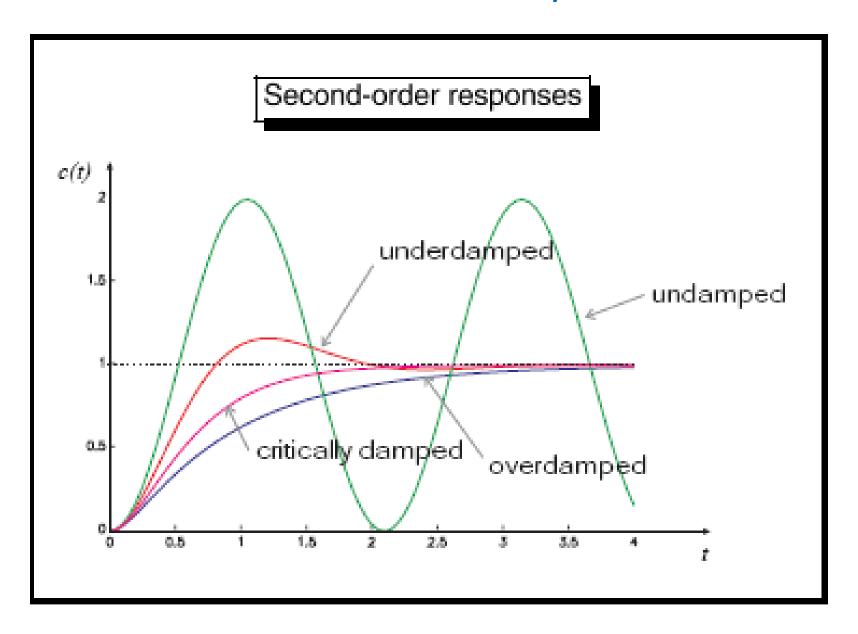
S = 0; s = -3,-3 (two real and equal poles)

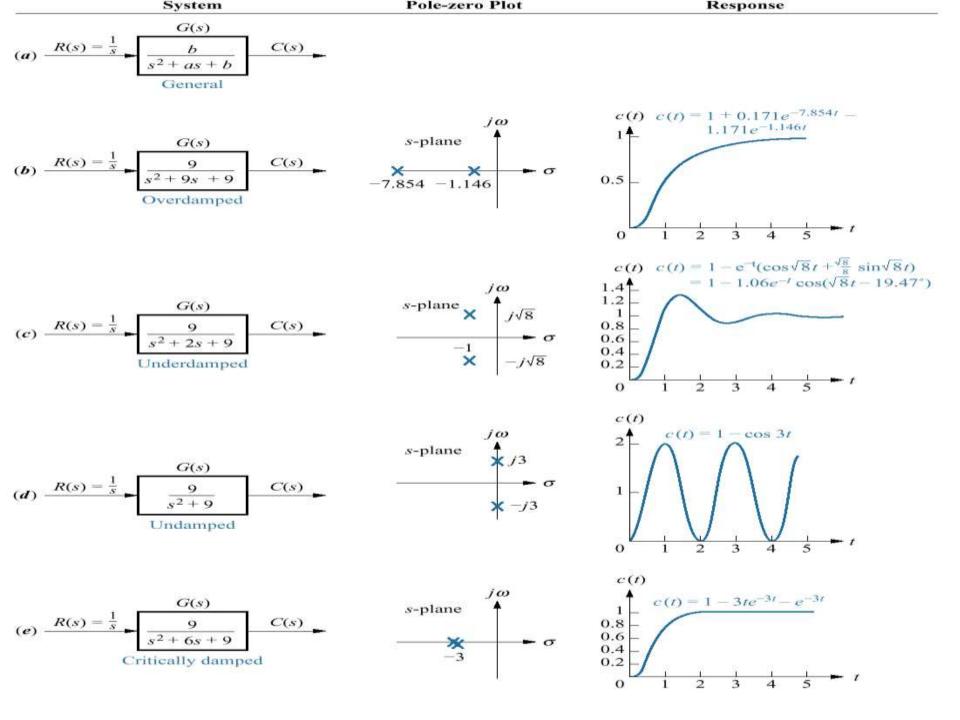
Critically Damped Response



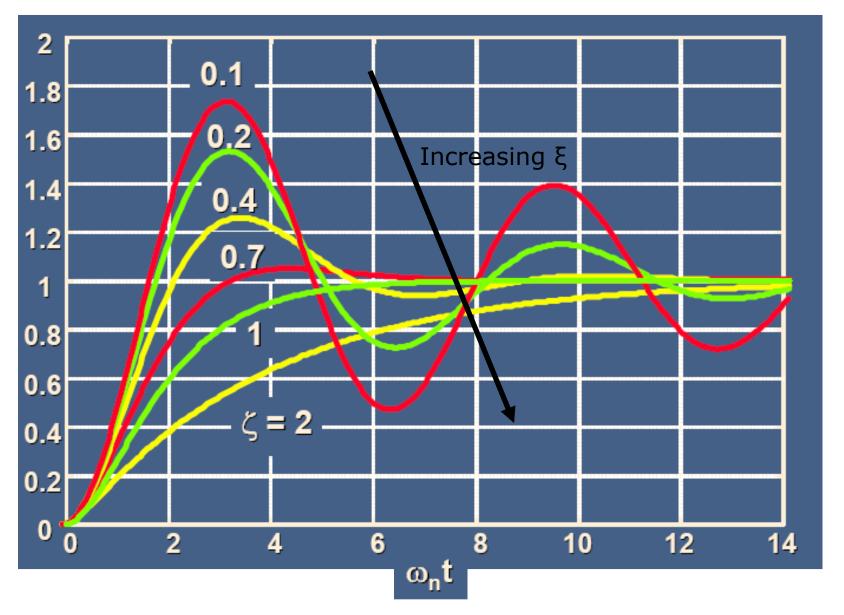
CRITICALLY DAMPED RESPONSE !!!

Second – Order System





Effect of different damping ratio, ξ



Second – Order System

<u>Example 4</u>: Describe the nature of the second-order system response via the value of the damping ratio for the systems with transfer function

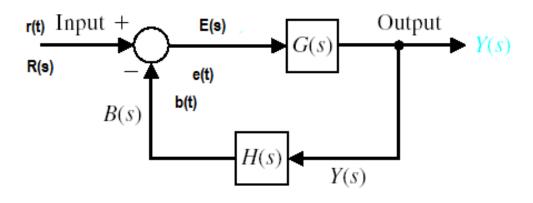
1.
$$G(s) = \frac{12}{s^2 + 8s + 12}$$

2.
$$G(s) = \frac{16}{s^2 + 8s + 16}$$

3.
$$G(s) = \frac{20}{s^2 + 8s + 20}$$

Steady State Error Analysis

Steady State Error Analysis



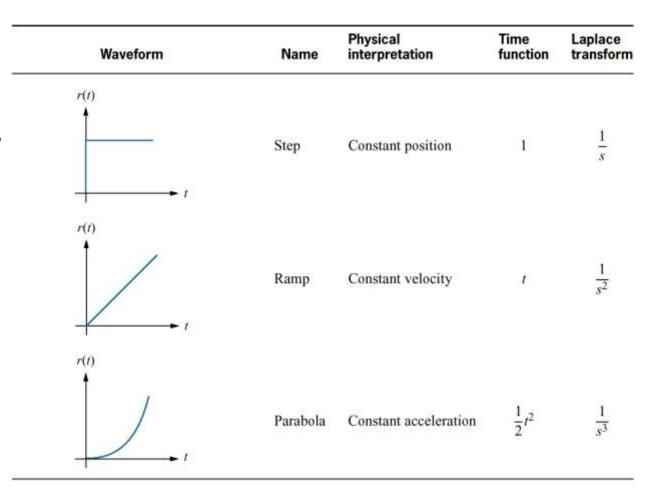
$$Error, e(t) = r(t) - b(t)$$

$$E(s) = R(s) - B(s) = \frac{1}{1 + G(s)H(s)} \times R(s)$$

Steady State Error= $\lim_{t\to\infty} e(t) = \lim_{s\to 0} sE(s)$

Steady state error analysis

Test waveforms for evaluating steady-state errors of position control systems

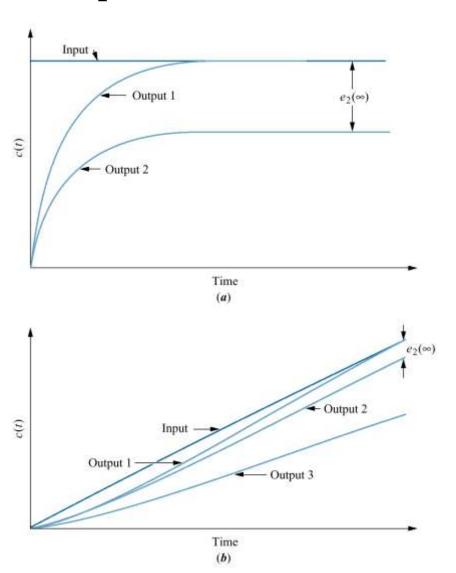


System response to different inputs

Steady-state error:

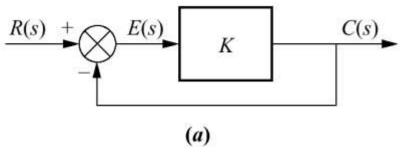
a. step input;

b. ramp input

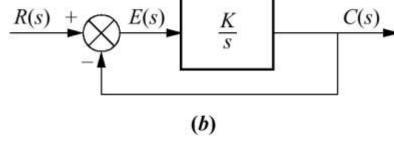


Intuition/motivation for steady state error analysis

System with: a. finite steady-state error for a step input; **b.** zero steady-state error for step input

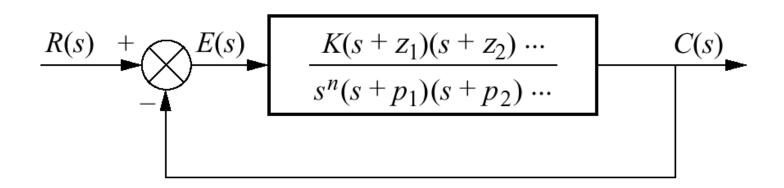


 $E_{steady \, state} = (1/K)C_{steady \, state}$



$$E_{steady\ state} = 0 \ for \ C_{steady\ state} \neq 0$$

Feedback control system for defining system type



Type of the system=n

Relationships between input, system type, static error constants, and steady-state errors

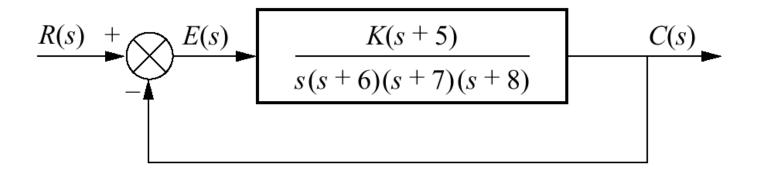
Input	Steady-state error formula	Type 0		Type 1		Type 2	
		Static error constant	Error	Static error constant	Error	Static error constant	Error
Step, $u(t)$	$\frac{1}{1+K_p}$	$K_p =$ Constant	$\frac{1}{1+K_p}$	$K_p = \infty$	0	$K_p = \infty$	0
Ramp, tu(t)	$\frac{1}{K_{\nu}}$	$K_v = 0$	∞	$K_v =$ Constant	$\frac{1}{K_{\nu}}$	$K_v = \infty$	0
Parabola, $\frac{1}{2}t^2u(t)$	$\frac{1}{K_a}$	$K_a = 0$	∞	$K_a = 0$	∞	$K_a =$ Constant	$\frac{1}{K_a}$

Position error constatnt, $K_p = \lim_{s \to 0} G(s)H(s)$

Velocity error constatnt, $K_v = \lim_{s \to 0} sG(s)H(s)$

Accelaration error constatnt, $K_a = \lim_{s \to 0} s^2 G(s) H(s)$

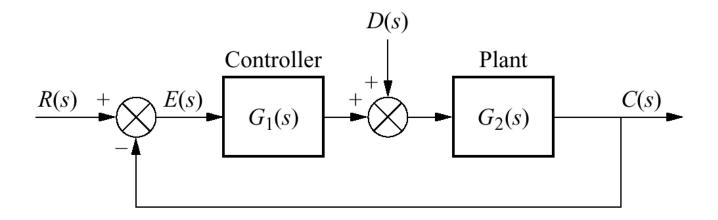
Example 1



Determine position error, velocity error, and acceleration error.

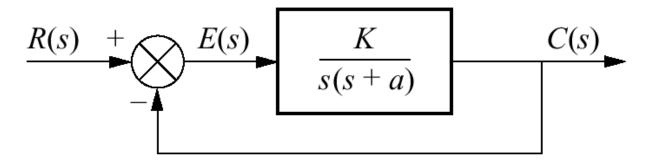
Example-2

Feedback control system showing disturbance



Determine position error, velocity error, and acceleration error.

Example-3



Determine position error, velocity error, and acceleration error.

Thank You