

# Summary of ELEC 402

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## 1 Fabrication of ICs

### 1.1 Digital Circuits

#### 1.1.1 Packaging

- Packaged or covered with epoxy resin (plastic) or ceramic
- Wire bonding: connect the chip to the package
- PADS: connect the package to the PCB

IO buffer: connect the chip to the package

- change the voltage level
- remove noise
- improve rise/fall time
- protect the chip from ESD
- provide a constant current source

Driving PADS:

- connect the package to the PCB
- provide a constant current source
- remove noise
- protect the chip from ESD

## **1.2 Wafer Fabrication**

Built in a clean room.

Wafer: a thin slice of semiconductor material.

- Silicon
- Gallium Arsenide
- Silicon Carbide

Growing a Silicon Crystal:

- Czochralski process
- Float zone process
- Epitaxial growth

## **1.3 MOSFET**

MOSFET (Metal Oxide Semiconductor Field Effect Transistor):

- NMOS: n-channel MOSFET
- PMOS: p-channel MOSFET
- CMOS: complementary MOSFET

## **1.4 Lithography**

- Photolithography: use light to transfer a geometric pattern from a photomask to a light-sensitive chemical (photoresist) on the substrate.
- Etching: remove the unwanted material
- Ion implantation: change the electrical properties of the material

### **1.4.1 N/P-Well CMOS Process**

## 2 RLC Circuits

### 2.1 Circuits Review

#### 2.1.1 Resistor

- $V = IR$
- $P = IV = I^2R = \frac{V^2}{R}$

#### 2.1.2 Capacitor

- $I = C \frac{dV}{dt}$
- $V = \frac{1}{C} \int I dt$
- $P = IV = CV \frac{dV}{dt} = \frac{V^2}{R}$

#### 2.1.3 Inductor

- $V = L \frac{dI}{dt}$
- $I = \frac{1}{L} \int V dt$
- $P = IV = LI \frac{dI}{dt} = \frac{I^2}{R}$

## 2.2 RLC Circuits

### 2.2.1 Series RLC Circuit

Wires have resistance, inductance and capacitance. Thus the circuit is not ideal.

For a change of voltage to propagate through the wire, it takes time.

### 2.2.2 Rise/Fall Time

Rise time is the time it takes for the voltage to rise from 10% to 90% of the final value.

Fall time is the time it takes for the voltage to fall from 90% to 10% of the final value.

If there is no inductance, then the time of fall  $t_f = RC \times \ln(9) = 2.2RC$ .

Let  $\tau = RC$ , then  $t_f = 2.2\tau$ . If  $\tau$  is too large, the signal will converge given a fluctuating input since it never reaches the final value before the next fluctuation.

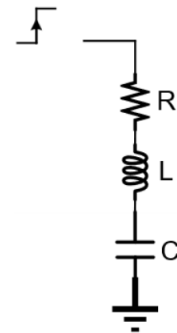


Figure 1: RLC Circuit

## 3 MOSFET

### 3.1 Structure

The given image shows a MOSFET with P-type body and N-type source and drain. This is a NMOS transistor. Voltage applied to gate will allow current to flow from source to drain.

- **Polysilicon:** Silicon formed from many small silicon crystals.
- **Gate Oxide:** Insulating layer between the gate and the channel.
- Source and Drain are doped with N-type impurities. PMOS has P-type impurities with body as N-type.
- **Bulk:** Body of the transistor.

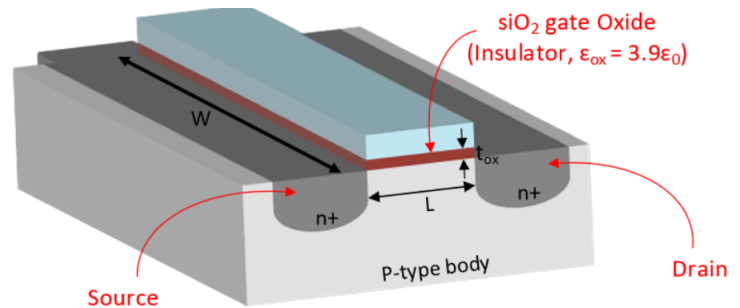


Figure 2: MOSFET

#### 3.1.1 Gate-body Structure

1. Accumulation mode: Gate voltage is negative or zero. Holes are attracted to the gate thus forming a channel. Low resistance.
2. Depletion mode: Gate voltage is positive but less than threshold voltage. Holes are repelled from the gate. High resistance.
3. Inversion mode: Gate voltage is greater than threshold voltage. Electrons are attracted to the gate thus forming holes at the body. Low resistance.

### 3.2 NMOS and PMOS

Drain-Source current is controlled by the gate-source voltage.

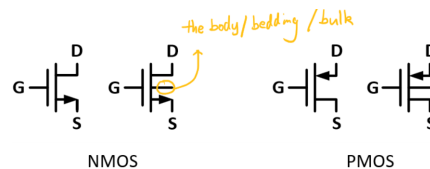


Figure 3: MOSFET

**NMOS:** When  $V_{GS} > V_{th}$ , the transistor is on. When  $V_{GS} < V_{th}$ , the transistor is off.

**PMOS:** When  $V_{GS} < V_{th}$ , the transistor is on. When  $V_{GS} > V_{th}$ , the transistor is off.

### 3.2.1 nMOS modes

Source and drain are symmetric diffusion terminals. For nMOS, source is the terminal at lower voltage ( $V_{ds} > 0$ )

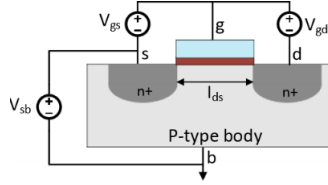


Figure 4: nMOS modes

- Cutoff:  $V_{GS} < V_{th}$ ,  $I_{DS} = 0$ . It is the depletion mode.
- Triode/Linear:  $V_{GS} > V_{th}$ ,  $V_{DS} < V_{GS} - V_{th}$ . It is the linear mode.
  - Conductivity is proportional to  $V_{GS} - V_{th}$ . The larger the  $V_{GS} - V_{th}$ , the larger the conductivity, the smaller the resistance.
  - Current flow from D to S. Electrons flow from S to D.
- Saturation:  $V_{GS} > V_{th}$ ,  $V_{DS} > V_{GS} - V_{th}$ . It is the saturation mode.
  - Channel is pinched off at the drain end. The current is independent of  $V_{DS}$ .

### 3.3 MOSFET Characteristics

Carrier velocity is proportional to the electric field.  $v_d = \mu_n E$ .  $\mu_n$  is the carrier mobility of the electrons.

Electric field is proportional to the voltage gradient.  $E = \frac{dV}{dx}$ . Can approximate as  $E = \frac{V}{L}$  where  $L$  is the length of the channel.

Thus time it takes for the electron to travel from source to drain is  $t = \frac{L}{v_d} = \frac{L}{\mu_n E} = \frac{L^2}{\mu_n V}$ .

$C_{ox}$  is the capacitance of the oxide layer.  $C_{ox} = \frac{\epsilon_{ox}}{t_{ox}}$ . With  $W$  as the width of the channel,  $L$  as the length of the channel,  $C_{ox} = \frac{\epsilon_{ox} W L}{t_{ox}}$ .

Let  $\beta = \mu_n C_{ox} \frac{W}{L}$  be the gain factor.

- Cut-off:  $I_{DS} = 0$ .
- Triode:  $I_{DS} = \beta(V_{GS} - V_{th} - \frac{V_{DS}}{2})V_{DS}$ .
- Saturation:  $I_{DS} = \frac{\beta}{2}(V_{GS} - V_{th})^2$ . independent of  $V_{DS}$ .

For a fixed  $V_{ds}$  and  $V_{GS} > V_{th}$ , the current is proportional to:

- $L$ : The length of the channel. The longer the channel, the longer the time it takes for the electron to travel from source to drain.
- $W$ : The width of the channel. The wider the channel, the more electrons can flow through.
- $V_{th}$ : The threshold voltage. The larger the threshold voltage, the larger the current.
- $\epsilon_{ox}$ : The permittivity of the oxide layer. The larger the permittivity, the larger the current.
- $t_{ox}$ : The thickness of the oxide layer. The thinner the oxide layer, the larger the current.
- $\mu_n$ : The carrier mobility of the electrons. The larger the carrier mobility, the larger the current.

$$- \mu_n = 500 \text{ cm}^2/\text{V} \cdot \text{sec} \approx 2.5 \mu_p. \mu_p = 180 \text{ cm}^2/\text{V} \cdot \text{sec}.$$

- Thus need to increase the size of pmos to match the current of nmos.

For small  $V_{DS}$ , the transistor is viewed as a linear resistor.

$$R_{on} = \frac{1}{\beta(V_{GS} - V_{th})}.$$

### 3.3.1 Region of Operation

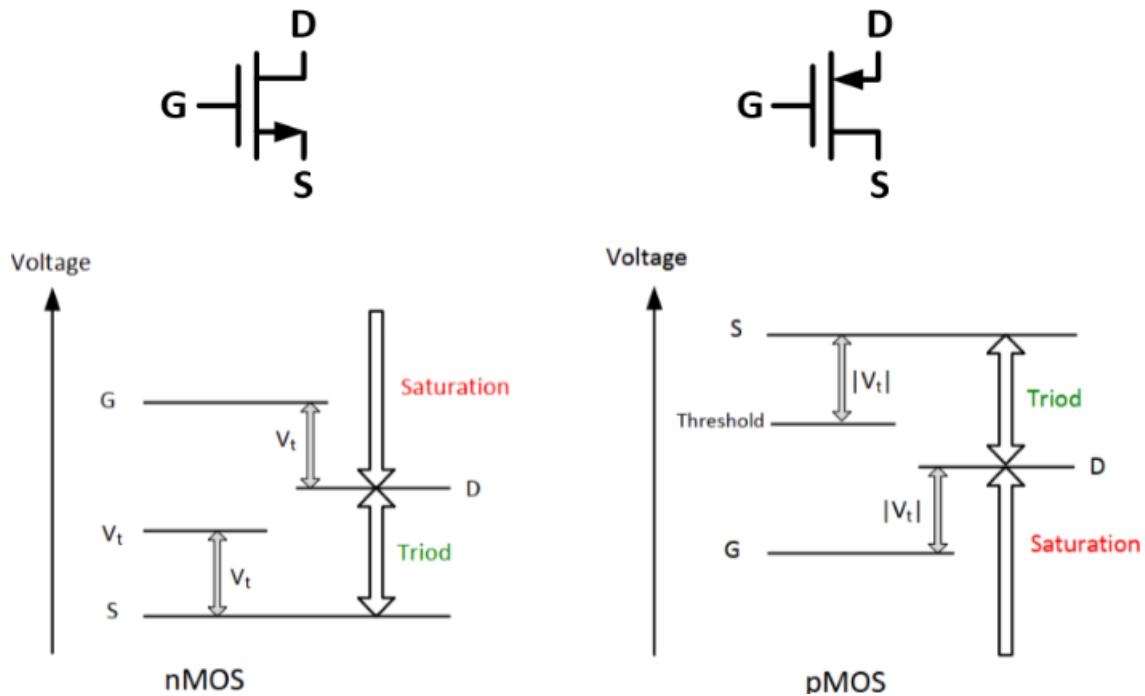


Figure 5: Region of Operation

## 4 Combinational Logic

### 4.1 Static complementary CMOS

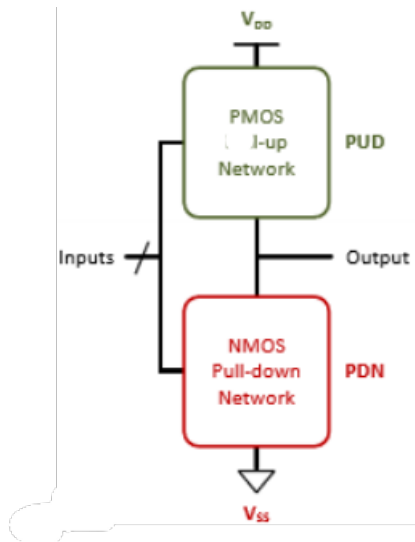
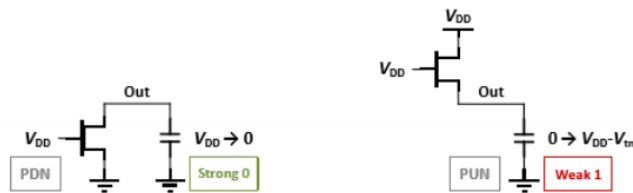


Figure 6: Static CMOS

- Pull-up Network (PUN): Constructed using PMOS transistors
- Pull-down Network (PDN): Constructed using NMOS transistors

Why is PDN made of NMOS and PUN made of PMOS?

- NMOS produces "strong zeros" and "weak ones."



- PMOS produces "strong ones" and "weak zeros."

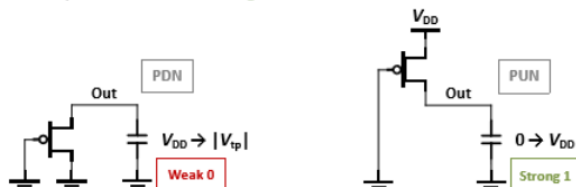


Figure 7: PDN and PUN

PUN and PDN are dual logic networks. One and only one of the networks conduct at DC.