

## Parcial 1: Señales y Sistemas

Presentado por:

Wilmer Sebastián Pérez Cuastumal

Profesor:

Andrés Marino Álvarez Meza

Jueves 21 de Marzo

Departamento de Ingeniería Eléctrica, Electrónica y computación

Universidad Nacional de Colombia – Sede Manizales

2024 – 1

## Componentes teóricas de solución a mano

### A. Distancia media de dos señales periódicas $d^2(x_1 - x_2)^2$ con potencia

$$a. d^2(x_1, x_2) = \overline{P_{x_1 - x_2}} = \lim_{T \rightarrow \infty} \frac{1}{T} \int_T |x_1(t) - x_2(t)|^2 dt \quad \begin{aligned} &: x_1(t) = A e^{j\omega_0 t} \\ &x_2(t) = B e^{j5\omega_0 t} \end{aligned}$$

$A, B, T \in \mathbb{R}^+$        $\omega_0 = \frac{2\pi}{T}$

$$= \lim_{T \rightarrow \infty} \frac{1}{T} \int_T |A e^{j\omega_0 t} - B e^{j5\omega_0 t}|^2 dt$$

$$= \lim_{T \rightarrow \infty} \frac{1}{T} \left[ \underbrace{\int_T (A e^{j\omega_0 t})^2 dt}_{(1)} - \underbrace{2AB \int_T e^{j\omega_0 t} e^{j5\omega_0 t} dt}_{(2)} + \underbrace{\int_T (B e^{j5\omega_0 t})^2 dt}_{(3)} \right]$$

$$|x(t)|^2 = x(t) \cdot x(t)^* \rightarrow \text{conjugado.}$$

$$\begin{aligned} \therefore (1) \int_T |A e^{j\omega_0 t}|^2 dt &= A^2 \int_T e^{j\omega_0 t} \cdot e^{-j\omega_0 t} dt \\ &= A^2 \int_T e^0 dt \\ &= A^2 e \Big|_0^T \\ &= A^2 T - A^2 \cdot 0 \end{aligned}$$

$$\boxed{\int_T |A e^{j\omega_0 t}|^2 dt = A^2 T}$$

$$\begin{aligned} (2) \int_T (B e^{j5\omega_0 t})^2 dt &= B^2 \int_T e^{j5\omega_0 t} \cdot e^{-j5\omega_0 t} dt \\ &= B^2 \int_T e^0 dt \\ &= B^2 t \Big|_0^T \\ &= B^2 T - B^2 \cdot 0 \end{aligned}$$

$$\boxed{\int_T (B e^{j5\omega_0 t})^2 dt = B^2 T}$$

$$\begin{aligned} (3) \int_T e^{j\omega_0 t} \cdot e^{j5\omega_0 t} dt &= \int_T e^{j6\omega_0 t} dt \\ &= \int_T \cos(6\omega_0 t) + j \sin(6\omega_0 t) dt \\ &= \int_T \cos\left(\frac{12\pi}{T} t\right) + j \sin\left(\frac{12\pi}{T} t\right) dt \\ &= \frac{T}{12\pi} \left[ \cancel{\sin\left(\frac{12\pi}{T} t\right)} \Big|_0^T - \cancel{\cos\left(\frac{12\pi}{T} t\right)} \Big|_0^T \right] + \frac{T}{12} \left[ -\underbrace{\cos\left(\frac{12\pi}{T} t\right)}_1 \Big|_0^T - \underbrace{\cos\left(\frac{12\pi}{T} t\right)}_1 \Big|_0^T \right] \\ &= \frac{T}{12\pi} [0 - 0] - \frac{T}{12\pi} [1 - 1] \end{aligned}$$

$$\boxed{\int_T e^{j\omega_0 t} \cdot e^{j5\omega_0 t} dt = 0}$$

$$\begin{aligned} \therefore d^2(x_1, x_2) &= \lim_{T \rightarrow \infty} \frac{1}{T} (A^2 T - 2AB \cdot 0 + B^2 T) \\ &= \lim_{T \rightarrow \infty} A^2 + B^2 \end{aligned}$$

$$\boxed{d^2(x_1, x_2) = A^2 + B^2} \rightarrow \text{distancia de las dos señales.}$$

media.

B. Señal obtenida en el tiempo discreto utilizando un conversor análogo digital a una frecuencia de muestreo  $f_s = 5\text{kHz}$

$$b. \quad x(t) = 3 \cos(\underbrace{1000\pi t}_{\omega_1}) + 5 \sin(\underbrace{2000\pi t}_{\omega_2}) + 10 \cos(\underbrace{11000\pi t}_{\omega_3}) \quad : f_s = 5\text{kHz}$$

→ buscamos el periodo para cada señal

$$\therefore T_1 = \frac{2\pi}{\omega_1} = \frac{2\pi}{1000\pi} = \frac{1}{500} \text{ [s]}$$

$$T_2 = \frac{2\pi}{\omega_2} = \frac{2\pi}{2000\pi} = \frac{1}{1000} \text{ [s]}$$

$$T_3 = \frac{2\pi}{\omega_3} = \frac{2\pi}{11000\pi} = \frac{1}{5500} \text{ [s]}$$

Nota: se puede observar que  $x(t)$  es cuasiperiódica  $\therefore$  la división de sus frecuencias es  $\mathbb{Q}$

→ Ahora buscamos  $T$  con  $T_1, T_2$  y  $T_3$

$$T = kT_1 = lT_2 = sT_3$$

$$T = k \frac{1}{500} = l \frac{1}{1000} = s \frac{1}{5500}$$

$$11000T = 22k = 11l = 2s$$

$$(M.C.M.) \Rightarrow \begin{array}{ccc} 22 & 11 & 2 \\ 11 & 11 & 1 \\ 1 & 1 & \end{array}$$

$$\left. \begin{array}{ccc} 22 & 11 & 2 \\ 11 & 11 & 1 \\ 1 & 1 & \end{array} \right\} \begin{array}{l} 2 \times 11 = 22 \\ 2 \times 11 = 22 \end{array}$$

Igualemos a  $T$

$$\left. \begin{array}{cccc|c} 500 & 1000 & 5500 & 10 & \\ 50 & 100 & 550 & 10 & \\ 5 & 10 & 55 & 5 & \\ 1 & 2 & 11 & 2 & \\ & 1 & 11 & 11 & \end{array} \right\} 10 \times 11 \times 5 \times 2 = 11000$$

$$11000T = 22$$

$$T = \frac{22}{11000} \rightarrow T = \frac{1}{500}$$

$$\therefore \begin{array}{ccc} T = kT_1 & ; & T = lT_2 & ; & T = sT_3 \\ k = \frac{1}{500} & ; & l = \frac{1}{500} & ; & s = \frac{1}{5500} \\ \frac{1}{500} & & \frac{1}{1000} & & \frac{1}{5500} \\ \boxed{k=1} & & \boxed{l=2} & & \boxed{s=11} \end{array}$$

→ Miremos si  $f_s$  de  $5\text{kHz}$  cumple con Nyquist

$$f_s \geq 2f_{\text{max}} \quad ; \quad \omega = \frac{2\pi}{T} = 2\pi f \quad \therefore f = \frac{\omega}{2\pi}$$

$$f_s \geq 2f_1(f_1, f_2, f_3) \quad f_1 = \frac{1000\pi}{2\pi} = 500 \text{ [Hz]}$$

$$f_s \geq 2(5500)$$

$$f_s \geq 11000 \text{ Hz}$$

$$f_2 = \frac{2000\pi}{2\pi} = 1000 \text{ [Hz]}$$

$$f_3 = \frac{11000\pi}{2\pi} = 5500 \text{ [Hz]}$$

$\therefore 5\text{kHz}$  no cumple con Nyquist  $\rightarrow$  Posible frecuencia a utilizar  $f_s = 20f_3$

$$\boxed{f_s = 110\text{kHz}}$$

cumple Nyquist

→ Ahora discretizamos la señal para 5KHz

$$x(t) = 3 \cos(1000\pi t) + 5 \sin(2000\pi t) + 10 \cos(11000\pi t)$$

$$x(t \rightarrow \frac{n}{f_s}) = 3 \cos\left(1000\pi \frac{n}{f_s}\right) + 5 \sin\left(2000\pi \frac{n}{f_s}\right) + 10 \cos\left(11000\pi \frac{n}{f_s}\right)$$

$$x[n] = 3 \cos\left(\frac{1000\pi n}{5000}\right) + 5 \sin\left(\frac{2000\pi n}{5000}\right) + 10 \cos\left(\frac{11000\pi n}{5000}\right)$$

$$x[n] = 3 \underbrace{\cos\left(\frac{\pi n}{5}\right)}_{\Omega_1} + 5 \underbrace{\sin\left(\frac{2\pi n}{5}\right)}_{\Omega_2} + 10 \underbrace{\cos\left(\frac{11\pi n}{5}\right)}_{\Omega_3}$$

buscamos copias

$$\underbrace{-\pi < \Omega_1 < \pi}_{\text{cumple}} ; \underbrace{-\pi < \Omega_2 < \pi}_{\text{cumple}} ; \underbrace{-\pi < \frac{11\pi}{5} < \pi}_{\text{No cumple}} \therefore$$

$$\hat{\Omega}_3 = \frac{11\pi}{5} \overset{\text{quitamos una vuelta}}{-2\pi}$$

$$\therefore x[n] = 3 \cos\left(\frac{\pi n}{5}\right) + 5 \sin\left(\frac{2\pi n}{5}\right) + 10 \cos\left(\frac{\pi n}{5}\right)$$

$$\hat{\Omega}_3 = \frac{\pi}{5} //$$

$$\boxed{x[n] = 13 \cos\left(\frac{\pi n}{5}\right) + 5 \sin\left(\frac{2\pi n}{5}\right)}$$

C. Encontrar T para discretisar

C.  $x(t) = 20 \cos(\underbrace{\frac{t}{3}}_{\omega_1}) + \cos(\underbrace{\frac{t}{4}}_{\omega_2})$

→ comprobamos es cuasiperiódica

$$\frac{\omega_1}{\omega_2} = \frac{1/3}{1/4} = \frac{4}{3} \in \mathbb{Q} \rightarrow \therefore \text{es cuasiperiódica}$$

→ buscamos  $T$

$$\left. \begin{aligned} T_1 &= \frac{2\pi}{\omega_1} = \frac{2\pi}{1/3} = 6\pi \\ T_2 &= \frac{2\pi}{\omega_2} = \frac{2\pi}{1/4} = 8\pi \end{aligned} \right\} \therefore \begin{aligned} T &= l T_1 = k T_2 \\ T &= l 6\pi = k 8\pi \\ \frac{T}{\pi} &= l 6 = k 8 \end{aligned}$$

$\rightarrow \text{M.C.M.}(6, 8) = 24$

$$\frac{T}{\pi} = 24 \rightarrow \boxed{T = 24\pi}$$

$$\therefore l = \frac{T}{T_1} ; k = \frac{T}{T_2}$$

$$l = \frac{24\pi}{1/3\pi} \quad k = \frac{24\pi}{1/4\pi}$$

$$\boxed{l = 72}$$

$$\boxed{k = 96}$$