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### Signaling of demodulation reference signal configuration for uplink short TTI transmissions

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#### Abstract

A method, wireless device and network node for multiplexing demodulation reference signals, DMRS, during short transmission time intervals, sTTIs. According to one aspect, a method includes generating an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS transmission. The method further includes transmitting to the wireless device the indication of IFDMA subcarrier configuration.

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**Field of Classification Search**

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**References Cited**

**U.S. PATENT DOCUMENTS**

Patent No.	Issued Date	Patentee Name	U.S. Cl.	CPC
10897385	12/2020	Gao	N/A	H04L 27/2636
11343837	12/2021	Lee	N/A	H04W 72/0446
11689331	12/2022	Gao	370/329	H04L 27/2085
11929793	12/2023	Da Silva	N/A	H04B 7/0408
2007/0060161	12/2006	Chindapol et al.	N/A	N/A
2012/0281656	12/2011	Hooli	370/328	H04J 11/0023
2012/0320839	12/2011	Noh	370/329	H04L 1/1861
2015/0065153	12/2014	Nissila	455/450	H04L 25/0226
2016/0029331	12/2015	Seo et al.	N/A	N/A
2017/0164352	12/2016	Yang	N/A	H04W 72/21
2018/0288787	12/2017	Hooli	N/A	H04L 5/0053
2019/0253300	12/2018	Munier	N/A	H04L 27/2666
2019/0268934	12/2018	Korhonen	N/A	H04L 27/2613
2020/0021336	12/2019	Da Silva	N/A	H04W 36/0085
2020/0028723	12/2019	Gao	N/A	H04L 27/2636
2021/0111937	12/2020	Gao	N/A	H04L 27/2607
2021/0184739	12/2020	Sang	N/A	H04L 5/0048
2021/0345304	12/2020	Munier	N/A	H04W 72/23
2022/0217747	12/2021	Lee	N/A	H04L 27/2607
2023/0224811	12/2022	Xu	370/311	H04W 56/0035
2023/0239193	12/2022	Munier	370/329	H04L 1/0071
2024/0195459	12/2023	Da Silva	N/A	H04B 7/0695

**FOREIGN PATENT DOCUMENTS**

Patent No.	Application Date	Country	CPC
2747320	12/2013	EP	N/A
2445745	12/2011	RU	N/A
2006134949	12/2005	WO	N/A
2011047462	12/2010	WO	N/A
2011074807	12/2010	WO	N/A
2015115991	12/2014	WO	N/A
2016148789	12/2015	WO	N/A

**OTHER PUBLICATIONS**

International Search Report and Written Opinion dated Feb. 7, 2018 for International Application No. PCT/ EP2017/078172 filed on Nov. 3, 2017, consisting of 9-pages. cited by applicant

3GPP TSG RAN WG1 #86b R1-1610008; Title: UL Design for Shortened TTI; Agenda Item: 7.2.10.2.1; Source: Qualcomm Incorporated; Document for: Discussion and Decision; Location and Date: Lisbon, Portugal Oct. 10-14, 2016, consisting of 7-pages. cited by applicant

3GPP TSG-RAN WG1 #86bis R1-1609849; Title: On RPF, Control Signaling, and Power Boosting for UL DMRS; Agenda Item: 7.2.2.3; Source: Ericsson; Document for: Discussion and Decision; Location and Date: Lisbon, Portugal Oct. 10-14, 2016, consisting of 3-pages. cited by applicant

3GPP TSG RAN WG1 Meeting #84bis R1-163173; Title: PUSCH design for shortened TTI; Agenda Item: 7.3.10.2; Source: NTT Docomo, Inc; Document for: Discussion and Decision; Location and Date: Busan, Korea Apr. 11-15, 2016, consisting of 6-pages. cited by applicant

3GPP TSG-RAN WG1 #87 R1-1611522; Title: Design aspects of sPUSCH; Agenda Item: 6.2.10.2.5; Source: Ericsson; Document for: Discussion, Decision; Location and Date: Reno, USA Nov. 14-18, 2016, consisting of 4-pages. cited by applicant

Release 12; 3GPP TS 36.213 V12.8.0; 3rd Generation Partnership Project; Technical Specification Group Radio Access Network; Evolved Universal Terrestrial Radio Access (E-UTRA); Physical layer procedures (Release 12); Dec. 2015; Consisting of 241 pages. cited by applicant

Release 14; 3GPP TS 36.213 V14.4.0; 3rd Generation Partnership Project; Technical Specification Group Radio Access Network; Evolved Universal Terrestrial Radio Access (E-UTRA); Physical layer procedures (Release 14); Sep. 2017; Consisting of 462 pages. cited by applicant

RP-161299; 3GPP TSG RAN Meeting #72; Busan, Korea; Jun. 13-16, 2016; Consisting of 9 pages. cited by applicant

3rd Generation Partnership Project; Technical Specification Group Radio Access Network; Evolved Universal Terrestrial Radio Access (E-UTRA); Physical layer procedures (Release 12) 3GPP TS 36.213 V12.12.0; Sep. 2017; Consisting of 241 pages. cited by applicant

R1-162777; Ericsson; NB-IoT—UL Reference signals; Discussion and Decision; 3GPP TSG-RAN WG1 Meeting #84bis; Busan, Korea, Apr. 11-15, 2016; Consisting of 8 pages. cited by applicant

3GPP TR 36.881 V1.0.0; 3rd Generation Partnership Project; Technical Specification Group Radio Access Network; Evolved Universal Terrestrial Radio Access (E-UTRA); Study on latency reduction techniques for LTE; Release 14; May 2016; Consisting of 248 pages. cited by applicant

R1-060385; Motorola; Cubic Metric in 3GPP-LTE; 3GPP TSG RAN WG1 #44; Denver, USA; Feb. 13-17, 2006; Consisting of 7 pages. cited by applicant

Release 13; 3rd Generation Partnership Project; Technical Specification Group Radio Access Network; Evolved Universal Terrestrial Radio Access (E-UTRA); Physical channels and modulation; 3GPP TS 36.211 V13.0.0; Dec. 2015; Consisting of 141 pages. cited by applicant

R1-167087; Nokia, Alcatel-Lucent Shanghai Bell; UL DMRS Base Sequences with IFDMA; Discussion and Decision; 3GPP TSG RAN WG1 Meeting #86; Gothenburg, Sweden; Aug. 22-26, 2016; Consisting of 4 pages. cited by applicant

Evolved Universal Terrestrial Radio Access (E-UTRA); 3rd Generation Partnership Project; Technical Specification Group Radio Access Network; Physical layer procedures; Release 13; 3GPP TS 36.213 V13.0.0; Dec. 2015; Consisting of 326 pages. cited by applicant

Official Action dated Nov. 18, 2019 and English language translation for Russian Patent Application No. 2019117040, consisting of 12-pages. cited by applicant

Indian Examination Report dated Nov. 5, 2020 for Application No. 201917016726, consisting of 6-pages. cited by applicant

EPO Communication dated Aug. 6, 2021 for Patent Application No. 17793943.6, consisting of 7-pages. cited by applicant

3GPP TSG RAN WG1 Meeting #86 R1-166341; Title: Control Signalling for UL DMRS with IFDMA; Agenda item: 7.2.4.1.3; Source: Nokia, Alcatel-Lucent Shanghai Bell; Document for:

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## **Background/Summary**

CROSS-REFERENCE TO RELATED APPLICATIONS (1) This application is a Continuation of U.S. application Ser. No. 16/333,490 filed Mar. 14, 2019, entitled "SIGNALING OF DEMODULATION REFERENCE SIGNAL CONFIGURATION FOR UPLINK SHORT TTI TRANSMISSIONS", which is a National Stage application of International Application No. PCT/EP2017/078172, filed Nov. 3, 2017, entitled "SIGNALING OF DEMODULATION REFERENCE SIGNAL CONFIGURATION FOR UPLINK SHORT TTI TRANSMISSIONS", which claims priority to U.S. Provisional Application No. 62/418,031 filed Nov. 4, 2016, entitled "SIGNALING OF DEMODULATION REFERENCE SIGNAL CONFIGURATION FOR UPLINK SHORT TTI TRANSMISSIONS", the entireties of all of which are incorporated herein by reference.

### **TECHNICAL FIELD**

(1) This disclosure relates to wireless communications, and in particular, signaling of demodulation reference signals (DMRS) configurations for uplink short transmission time interval (sTTI) transmissions.

### **BACKGROUND**

(2) In third generation partnership long term evolution (3GPP LTE) systems, data transmissions in both the downlink (i.e., from a network node or base station such as an eNodeB (eNB)) to a user's wireless device (user equipment (UE)) and the uplink (from a wireless device to a network node or eNB) are organized into radio frames of 10 ms, each radio frame consisting of ten equally-sized subframes of length  $T_{\text{subframe}}=1$  ms, as shown in FIG. 1

(3) LTE uses Orthogonal Frequency Division Multiplexing (OFDM) in the downlink and Single Carrier OFDM (SC-OFDM) in the uplink. The basic LTE downlink physical resource can thus be seen as a time-frequency grid as illustrated in FIG. 2, where each resource element corresponds to one OFDM subcarrier during one OFDM symbol interval.

(4) Furthermore, the resource allocation in LTE is typically described in terms of resource blocks (RBs), where a resource block corresponds to one slot (0.5 ms) in the time domain and 12 contiguous subcarriers in the frequency domain. Resource blocks are numbered in the frequency domain, starting with 0 from one end of the system bandwidth.

(5) Similarly, the LTE uplink resource grid is illustrated in FIG. 3, where  $N_{\text{sub.RB.sup.UL}}$  is the number of resource blocks (RBs) contained in the uplink system bandwidth,  $N_{\text{sub.sc.sup.RB}}$  is the number of subcarriers in each RB, typically  $N_{\text{sub.sc.sup.RB}}=12$ ,  $N_{\text{sub.symb.sup.UL}}$  is the number of SC-OFDM symbols in each slot.  $N_{\text{sub.symb.sup.UL}}=7$  for normal cyclic prefix (CP) and  $N_{\text{sub.symb.sup.UL}}=6$  for extended CP. A subcarrier and a SC-OFDM symbol forms an uplink resource element (RE).

(6) Downlink data transmissions from an eNB to a wireless device are dynamically scheduled, i.e., in each subframe the base station transmits control information concerning which terminal's data is transmitted and upon which resource blocks the data is transmitted, in the current downlink subframe. This control signaling is typically transmitted in the first 1, 2, 3 or 4 OFDM symbols in each subframe. A downlink system with 3 OFDM symbols as control is illustrated in FIG. 4.

(7) Similar to the downlink, uplink transmissions from a wireless device to an eNB are also dynamically scheduled through the downlink control channel. When a wireless device receives an uplink grant in subframe  $n$ , the wireless device transmits data in the uplink at subframe  $n+k$ , where  $k=4$  for a frequency division duplex (FDD) system and  $k$  varies for time division duplex (TDD) systems.

(8) In LTE, a number of physical channels are supported for data transmissions. A downlink or an uplink physical channel corresponds to a set of resource elements carrying information originating from higher layers, while a downlink or an uplink physical signal is used by the physical layer but does not carry information originating from higher layers. Some of the downlink physical channels and signals supported in LTE are: Physical Downlink Shared Channel, PDSCH Physical Downlink Control Channel, PDCCH Enhanced Physical Downlink Control Channel, EPDCCH Reference signals: Cell Specific Reference Signals (CRS) DeModulation Reference Signal (DMRS) for PDSCH Channel State Information Reference Signals (CSI-RS)

(9) The PDSCH is used mainly for carrying user traffic data and higher layer messages in the downlink and is transmitted in a downlink (DL) subframe outside of the control region as shown in FIG. 4. Both the PDCCH and the EPDCCH are used to carry Downlink Control Information (DCI) such as physical resource block (PRB) allocation, modulation level and coding scheme (MCS), precoder used at the transmitter, and etc. The PDCCH is transmitted in the first one to four OFDM symbols in a DL subframe, i.e., the control region, while the EPDCCH is transmitted in the same region as the PDSCH.

(10) Some of the uplink (UL) physical channels and signals supported in LTE are: Physical Uplink Shared Channel, PUSCH Physical Uplink Control Channel, PUCCH Reference Signals DeModulation Reference Signal (DMRS) for PUSCH DeModulation Reference Signal (DMRS) for PUCCH

(11) The PUSCH is used to carry uplink data or/and uplink control information from the wireless device to the eNodeB. The PUCCH is used to carry uplink control information (UCI) from the wireless device to the eNodeB.

(12) Demodulation reference symbols (DMRS) for the PUSCH is used for PUSCH demodulations. More specifically, the DMRS is used by the network node or eNB for uplink channel estimation in the RBs scheduled for the associated PUSCH. DMRS is time multiplexed with the associated PUSCH and occupies the same RBs as the PUSCH. The DMRS is transmitted on the resource elements (Res) of the 3<sup>rd</sup> SC-OFDM symbol of each slot of a subframe as shown in FIG. 5, where only one RB is shown. It can be seen that the DMRS occupies all the subcarriers in the 3<sup>rd</sup> symbols of each slot.

(13) The followings are the main design goals for uplink DMRS: Constant amplitude over transmitted subcarriers for uniform channel excitation and estimation; Low peak to average power ratio (PAPR) or cubic metric (CM) in time domain for efficient Power Amplifier (PA) utilization; Low cross correlation between different DMRS sequences for low inter-cell interference where different sequences are used in different cells.

(14) The above goals were achieved in LTE by using a combination of computer generated (CG) highly optimized base sequences for 1RB and 2RBs and cyclically extended Zadoff-Chu sequences for 3RBs or larger.

(15) Let  $r_{\text{sub.PUSCH.sup.}(\lambda)}$  be the DMRS sequence associated with an uplink multi-input-multi-output (MIMO) layer  $\lambda$ , then the DMRS sequence in LTE is defined as

$$r_{\text{sub.PUSCH.sup.}(\lambda)}(m_{\text{sub.s.Math.M.sub.sc.sup.RS}}+n)=w_{\text{sup.}(\lambda)}(m_{\text{sub.s}})r_{\text{sub.u,v.sup.}(\alpha_{\text{sup.}\lambda.\text{sup.}})(n)} \quad (1)$$

(16) where  $m_{\text{sub.s}}=0,1$  corresponds to slot 0 and slot 1, respectively, as shown in FIG. 5,  $n=0, \dots, M_{\text{sub.sc.sup.RS}}-1$  and  $M_{\text{sub.sc.sup.RS}}=M_{\text{sub.sc.sup.PUSCH}}$  is the number of subcarriers of the RBs scheduled for the associated PUSCH. In legacy LTE,  $w_{\text{sup.}(\lambda)}$  can be configured with  $[w_{\text{sup.}(\lambda)}(0) w_{\text{sup.}(\lambda)}(1)]=[1 \ 1]$ , or  $[1 \ -1]$  according to Table 5.5.2.1.1-1 of TS36.211, which is copied in

Table 3.  $\alpha.\text{sub}.\lambda$  is a cyclic shift configured for a MIMO layer  $\lambda$ . The cyclic shift  $\alpha.\text{sub}.\lambda$  in a slot  $n.\text{sub}.\text{s}$  is given as  $\alpha.\text{sub}.\lambda=2\pi n.\text{sub}.\text{cs},\lambda/12$  with  $n.\text{sub}.\text{cs},\lambda=(n.\text{sub}.\text{DMRS}.\text{sup}.\text{(1)}+n.\text{sub}.\text{DMRS},\lambda.\text{sup}.\text{(2)}+n.\text{sub}.\text{PN}(n.\text{sub}.\text{s}))\bmod 12$  (2)

(17) where the values of  $n.\text{sub}.\text{DMRS}.\text{sup}.\text{(1)}$  are configured by higher layers,  $n.\text{sub}.\text{DMRS},\lambda.\text{sup}.\text{(2)}$  is given by the cyclic shift for the DMRS field in most recent uplink-related DCI for the transport block associated with the corresponding PUSCH transmission: where the value of  $n.\text{sub}.\text{DMRS},\lambda.\text{sup}.\text{(2)}$  is given in Table 5.5.2.1.1-1 of technical standard TS36.211, which is copied in Table 3.  $n.\text{sub}.\text{PN}(n.\text{sub}.\text{s})$  is a cell specific number generated pseudo-randomly in a slot by slot basis.  $n.\text{sub}.\text{s}\in\{0, 1, \dots, 9\}$  is the slot index within a subframe.  $r.\text{sub}.\text{u},\text{v}.\text{sup}.\text{(}\alpha\text{)}(n)$  is a reference signal sequence and is defined by a cyclic shift  $\alpha$  of a base sequence  $r.\text{sub}.\text{u},\text{v}(n)$  according to  $r.\text{sub}.\text{u},\text{v}.\text{sup}.\text{(}\alpha\text{)}(n)=e.\text{sup}.\text{j}\alpha n r.\text{sub}.\text{u},\text{v}(n), 0\leq n<M.\text{sub}.\text{sc}.\text{sup}.\text{RS}$  (3)

(18) where  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}=mN.\text{sub}.\text{sc}.\text{sup}.\text{RB}$  is the length of the reference signal sequence and  $m$  is the number of RBs scheduled for the PUSCH. Multiple reference signal sequences are defined from a single base sequence  $r.\text{sub}.\text{u},\text{v}(n)$  through different values of  $\alpha$ .

(19) Base sequences  $r.\text{sub}.\text{u},\text{v}(n)$  are divided into groups, where  $u\in\{0, 1, \dots, 29\}$  is the group number and  $v$  is the base sequence number within the group. So there are 30 groups of base sequences for each sequence length. For  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}=mN.\text{sub}.\text{sc}.\text{sup}.\text{RB}$ ,  $1\leq m\leq 5$ , each group contains one base sequence ( $v=0$ ). For  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}=mN.\text{sub}.\text{sc}.\text{sup}.\text{RB}$ ,  $m\geq 6$ , there are two base sequences ( $v=0,1$ ) in each group.

(20) The definition of the base sequence  $r.\text{sub}.\text{u},\text{v}(0), \dots, r.\text{sub}.\text{u},\text{v}(M.\text{sub}.\text{sc}.\text{sup}.\text{RS}-1)$  depends on the sequence length  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}$ . For base sequences of length  $3N.\text{sub}.\text{sc}.\text{sup}.\text{RB}$  or larger,  $r.\text{sub}.\text{u},\text{v}(n)$  is generated through cyclic extension of a Zadoff-Chu (ZC) sequence  $x.\text{sub}.\text{q}(m)$  as follows

$r.\text{sub}.\text{u},\text{v}(n)=x.\text{sub}.\text{q}(n \bmod N.\text{sub}.\text{ZC}.\text{sup}.\text{RS}), 0\leq n\leq M.\text{sub}.\text{sc}.\text{sup}.\text{RS}$  (4)

(21) where the  $q.\text{sup}.\text{th}$  root Zadoff-Chu sequence is defined by

(22)  $x_q(m) = e^{-j\frac{qm(m+1)}{N_{\text{ZC}}^{\text{RS}}}}, 0\leq m\leq N_{\text{ZC}}^{\text{RS}}-1$  (5)

(23) with  $q$  given by

$q=\lfloor q+1/2\rfloor+v.\text{Math}.\text{(-1)}.\text{sup}.\lfloor 2q\rfloor$

$q=N.\text{sub}.\text{ZC}.\text{sup}.\text{RS}.\text{Math}.\text{(}u+1\text{)}/31$

The length  $N.\text{sub}.\text{ZC}.\text{sup}.\text{RS}$  of the Zadoff-Chu sequence is given by the largest prime number such that  $N.\text{sub}.\text{ZC}.\text{sup}.\text{RS}<M.\text{sub}.\text{sc}.\text{sup}.\text{RS}$ .

(24) By using cyclic extension of the Zadoff-Chu sequences, the base sequences have a constant amplitude over frequency and also maintain the zero auto-correlation cyclic shift orthogonality property of the Zadoff-Chu sequences, which allows generating multiple orthogonal sequences by using different cyclic shifts on a single base sequence. Using extension, not truncation, in general provides better CM for 3 and more RBs. In addition, at least 30 base sequences can be generated this way.

(25) For one and two RBs, however, only a small number of low CM extended Zadoff-Chu sequences are available. To achieve similar inter-cell interference randomization as in the case of 3 or more PRBs, 30 base sequences are desirable. Thus, base sequences for one and two RBs were obtained by computer searches. Only quadrature phase shift keying (QPSK) based sequences were selected to reduce memory size for storage and computational complexity. The base sequences for one and two RBs (i.e.  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}=N.\text{sub}.\text{sc}.\text{sup}.\text{RB}$  and  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}=2N.\text{sub}.\text{sc}.\text{sup}.\text{RB}$ ) are defined as

$r.\text{sub}.\text{u},\text{v}(n)=e.\text{sup}.\text{j}\varphi(n)\pi/4, 0\leq n\leq M.\text{sub}.\text{sc}.\text{sup}.\text{RS}-1$  (6)

(26) where the value of  $\varphi(n)$  is given by Table 1 for  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}=N.\text{sub}.\text{sc}.\text{sup}.\text{RB}$  and by Table 2 for  $M.\text{sub}.\text{sc}.\text{sup}.\text{RS}=2N.\text{sub}.\text{sc}.\text{sup}.\text{RB}$ .

(27) TABLE-US-00001 TABLE 1  $u \varphi(0), \dots, \varphi(11)$  0 -1 1 3 -3 3 3 1 1 3 1 -3 3 1 1 1 3 3 3 -1 1 -3 -3 1 -3 3 2 1 1 -3 -3 -3 -1 -3 -3 1 -3 1 -1 3 -1 1 1 1 1 -1 -3 -3 1 -3 3 -1 4 -1 3 1 -1 1 -1

-3 -1 1 -1 1 3 5 1 -3 3 -1 -1 1 1 -1 -1 3 -3 1 6 -1 3 -3 -3 -3 3 1 -1 3 3 -3 1 7 -3 -1 -1 1  
-3 3 -1 1 -3 3 1 8 1 -3 3 1 -1 -1 -1 1 1 3 -1 1 9 1 -3 -1 3 3 -1 -3 1 1 1 1 1 10 -1 3 -1 1 1 -3  
-3 -1 -3 -3 3 -1 1 1 3 1 -1 -1 3 3 -3 1 3 1 3 3 12 1 -3 1 1 -3 1 1 1 -3 -3 -3 1 13 3 3 -3 3 -3 1 1  
3 -1 -3 3 3 14 -3 1 -1 -3 -1 3 1 3 3 3 -1 1 15 3 -1 1 -3 -1 -1 1 1 3 1 -1 -3 16 1 3 1 -1 1 3 3 3  
-1 -1 3 -1 17 -3 1 1 3 -3 3 -3 -3 3 1 3 -1 18 -3 3 1 1 -3 1 -3 -3 -1 -1 1 -3 19 -1 3 1 3 1 -1 -1  
3 -3 -1 -3 -1 20 -1 -3 1 1 1 1 3 1 -1 1 -3 -1 21 -1 3 -1 1 -3 -3 -3 -3 -3 1 -1 -3 22 1 1 -3 -3  
-3 -3 -1 3 -3 1 -3 3 23 1 1 -1 -3 -1 -3 1 -1 1 3 -1 1 24 1 1 3 1 3 3 -1 1 -1 -3 -3 1 25 1 -3 3 3  
1 3 3 1 -3 -1 -1 3 26 1 3 -3 -3 3 -3 1 -1 -1 3 -1 -3 27 -3 -1 -3 -1 -3 3 1 -1 1 3 -3 -3 28 -1 3  
-3 3 -1 3 3 -3 3 3 -1 -1 29 3 -3 -3 -1 -1 -3 -1 3 -3 3 1 -1

(28) TABLE-US-00002 TABLE 2  $u \varphi(0), \dots, \varphi(23)$  0 -1 3 1 -3 3 -1 1 3 -3 3 1 3 -3 3 1 1 -1 1 3  
-3 3 -3 -1 -3 1 -3 3 -3 -3 -3 1 -3 -3 3 -1 1 1 1 3 1 -1 3 -3 -3 1 3 1 1 -3 2 3 -1 3 3 1 1 -3 3 3  
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-3 3 1 -1 3 3 24 1 -1 3 3 -1 -3 3 -3 -1 -1 3 -1 3 -1 -1 1 1 1 1 -1 -1 -3 -1 3 25 1 -1 1 -1 3 -1  
3 1 1 -1 -1 -3 1 1 -3 1 3 -3 1 1 -3 -3 -1 -1 26 -3 -1 1 3 1 1 -3 -1 -1 -3 3 -3 3 1 -3 3 -3 1 -1  
1 -3 1 1 1 27 -1 -3 3 3 1 1 3 -1 -3 -1 -1 -1 3 1 -3 -3 -1 3 -3 -1 -3 -1 -3 -1 28 -1 -3 -1 -1 1  
-3 -1 -1 1 -1 -3 1 1 -3 1 -3 -3 3 1 1 -1 3 -1 -1 29 1 1 -1 -1 -3 -1 3 -1 3 -1 1 3 1 -1 3 1 3 -3  
-3 1 -1 -1 1 3

(29) It should be noted that a phase shift of a reference signal sequence does not change its peak to average power ration (PAPR) or CM. Also, the magnitude of a reference signal sequence's autocorrelation or cross correlation with other reference signal sequences does not change if the reference signal is phase shifted. Therefore, a reference signal  $r'_{\text{sub},u,v}(n) = e^{j\zeta} r_{\text{sub},u,v}(n)$  is equivalent to  $r_{\text{sub},u,v}(n)$ , where  $\zeta$  is a real number.

(30) A given reference sequence  $r_{\text{sub},u,v}(n)$  with sequence number  $u$  (for example, corresponding to a row in Tables 1 and 2 will have a given value of PAPR or CM. Also, a sequence  $r_{\text{sub},u,\text{sub},1,\text{sub},v}(n)$  with sequence number  $u_{\text{sub},1}$  and a sequence  $r_{\text{sub},u,\text{sub},2,\text{sub},v}(n)$  with sequence number  $u_{\text{sub},2}$  will have some cross-correlation  $c_{\text{sub},u,\text{sub},1,\text{sub},u_{\text{sub},2}}(l_{\text{sub},1}, l_{\text{sub},2})$ , where  $l_{\text{sub},1}$  and  $l_{\text{sub},2}$  are correlation lags. Good reference signal sequences should have low PAPR or CM and low cross-correlation.

(31) The CM for a signal,  $v(t)$ , with 3.84 MHz nominal bandwidth is defined according to

$$(32) \quad \text{CM} = \frac{20 \log_{10} \{ \text{rms}[v_{\text{norm}}^3(t)] \} - 1.52}{1.56} \text{dB} \quad (7)$$

where  $20 \log_{10} \{ \text{rms}[v_{\text{sub},\text{norm},\text{sup},3}(t)] \}$  is called raw cubic metric (in dB) of the signal, and

$$(33) \quad \text{rms}(x) = \frac{\sqrt{(x \cdot x)}}{N}, \quad v_{\text{norm}}(t) = \frac{\text{Math. } v(t) \cdot \text{Math.}}{\text{rms}[v(t)]}.$$

This definition is used in the CM calculations in the following sections.

(34) For a set of DMRS base sequences  $\{r_{\text{sub},u,v}(n), n=0, 1, \dots, M_{\text{sub},\text{sc}}-1\}$ , the cross correlation between two sequences  $r_{\text{sub},u1,v1}(n)$  and  $r_{\text{sub},u2,v2}(n)$  is defined as





either the first or the last entry of the sequence. A set of 30 base sequences have been proposed for length 18 sequences through computer search.

(43) For length 6 sequences, it has been proposed to reuse a set of 14 length-6 sequences that was agreed to be introduced for Narrow Band Internet of Things (NB-IOT) in LTE Release 14. The set of 14 length-6 sequences is a subset of the 16 sequences and is shown in Table 4 below.

(44) TABLE-US-00004 TABLE 4  $\mu$   $\varphi(0)$ ,  $\varphi(1)$ ,  $\varphi(2)$ ,  $\varphi(3)$ ,  $\varphi(4)$ ,  $\varphi(5)$ , 0 1 1 1 1 3 -3 1 1 1 3 1 -3 3 2 1 -1 -1 -1 1 -3 3 1 -1 3 -3 -1 -1 4 1 3 1 -1 -1 3 5 1 -3 -3 1 3 1 6 -1 -1 1 -3 -3 -1 7 -1 -1 -1 3 -3 -1 8 3 -1 1 -3 -3 3 9 3 -1 3 -3 -1 1 10 3 -3 3 -1 3 3 11 -3 1 3 1 -3 -1 12 -3 1 -3 3 -3 -1 13 -3 3 -3 1 1 -3

(45) In some cases, optimizing for minimum CM or peak to average power ratio (PAPR) or for low correlation magnitudes may not be essential. This design flexibility can be used to improve compatibility with “RPF=1” Rel-13 wireless devices that transmit DMRS without repetition or IFDMA.

(46) First, we recall that IFDMA wireless devices using RPF=2 transmit their sequence on every other subcarrier, as described above. More generally, IFDMA with D repetitions (‘RPF=D’), can be written as follows. Note that D=2 for the case with RPF=2.

$re(Dk, l) = r'.sub.u, v, sup.(\alpha)(n)$  (9)  $re(Dk, l)$  is a resource element with subcarrier index Dk in uplink SC-OFDM symbol l.  $r'.sub.u, v, sup.(\alpha)(n)$  is the n.sup.th element of a new DMRS reference sequence to be used for RPF=D with group index u and sequence index v with cyclic shift  $\alpha$ .

(47) If the RPF=D wireless devices use the same reference sequence element values on their occupied resource elements as the Rel-13 wireless devices do, then cyclic shift orthogonality can be maintained for all provided that the RPF=D sequence length is at least length 12. Then the new RPF=D sequence  $r'.sub.u, v(n)$  can be defined as a decimated version of the Rel-13 DMRS sequence, which can be expressed:

$$r'.sub.u, v(n) = r.sub.u, v(Dn + \Delta.sub.r) \quad (10)$$

(48) Where

(49)  $r.sub.u, v(n)$  is the n.sup.th element of the Rel-13 DMRS base sequence with group index u, group sequence index v.

(50)  $\Delta.sub.r \in \{0, 1, \dots, D\}$  is an offset used to select which portion of the Rel-13 DMRS base sequence is used.

(51) An example is shown in FIG. 7, where wireless device 1 is a new wireless device with IFDMA and wireless device 2 is a legacy wireless device.

(52) The new reference symbol sequence for RPF=D with cyclic shift is then determined somewhat differently than for Rel-13 using the following equation. Note that a factor of D is used in the exponent so that the Rel-13 and new reference signal have the same values when mapped to the same subcarriers.

$$r'.sub.u, v, sup.(\alpha)(n) = e.sup.jD\alpha n r'.sub.u, v(n), 0 \leq n < M'.sub.sc, sup.RS \quad (11)$$

(53) Where  $M'.sub.sc, sup.RS = \lfloor M.sub.sc, sup.RS / D \rfloor$  is the length of the new reference signal sequence and  $M.sub.sc, sup.RS$  is the length of the Rel-13 sequence from which it is decimated.

(54) The RPF=D sequence may alternatively be constructed by setting elements of the Rel-13 sequence to zero, and transmitting the modified sequence in the same REs as the Rel-13 sequence. This RE mapping can be expressed:

$$(55) \quad re(k, l) = \begin{cases} r_{u, v}^{(\alpha)}(n); & k = \text{Math. } n / D \text{ Math. } + r \\ 0; & \text{otherwise} \end{cases} \quad (12)$$

(56) Where

(57)  $r.sub.u, v, sup.(\alpha)(n)$  is the n.sup.th element of a Rel-13 DMRS reference sequence with group index u and group sequence index v with cyclic shift  $\alpha$ .

(58)  $\Delta.sub.r \in \{0, 1, \dots, D\}$  is an offset used to select which portion of the Rel-13 DMRS subcarriers are nonzero and contain values of the Rel-13 sequence.

(59) The approach can be extended to scenarios where a new wireless device configured with IFDMA may be paired with more than one legacy wireless devices, each occupying a different part of the bandwidth scheduled for the new wireless device. An example is shown in FIG. 8. In this case, the base sequence for the new wireless device may be a decimated version of the two base sequences associated with the two legacy wireless devices.

(60) Packet data latency is one of the performance metrics that vendors, operators and also end-users (via speed test applications) regularly measures. Latency measurements are done in all phases of a radio access network system lifetime, when verifying a new software release or system component, when deploying a system and when the system is in commercial operation.

(61) Shorter latency than previous generations of 3GPP RATs was one performance metric that guided the design of Long Term Evolution (LTE). LTE is also now recognized by the end-users to be a system that provides faster access to internet and lower data latencies than previous generations of mobile radio technologies.

(62) Packet data latency is important not only for the perceived responsiveness of the system, it is also a parameter that indirectly influences the throughput of the system. Hypertext transfer protocol/transmission control protocol (HTTP/TCP) is the dominating application and transport layer protocol suite used on the Internet today. According to HTTP Archive (<http://httparchive.org/trends.php>) the typical size of HTTP based transactions over the Internet are in the range of a few 10's of Kbytes up to 1 Mbyte. In this size range, the TCP slow start period is a significant part of the total transport period of the packet stream. During TCP slow start the performance is latency limited. Hence, improved latency can rather easily be shown to improve the average throughput, for this type of TCP based data transactions.

(63) Radio resource efficiency can be positively impacted by latency reductions. Lower packet data latency could increase the number of transmissions possible within a certain delay bound; hence, higher Block Error Rate (BLER) targets could be used for data transmissions, freeing up radio resources and potentially improving the capacity of the system.

(64) One approach to latency reduction is the reduction of transport time of data and control signaling, by addressing the length of a transmission time interval (TTI). By reducing the length of a TTI and maintaining the bandwidth, the processing time at the transmitter and the receiver nodes is also expected to be reduced, due to less data to process within the TTI. As described above, in LTE release 8, a TTI corresponds to one subframe (SF) of length 1 millisecond. One such 1 ms TTI is constructed by using 14 OFDM or SC-FDMA symbols in the case of normal cyclic prefix and 12 OFDM or SC-FDMA symbols in the case of extended cyclic prefix. In LTE release 14, a study item on latency reduction has been conducted, with the goal of specifying transmissions with shorter TTIs, such as a slot or a few symbols. A work item with the goal of specifying short TTI (sTTI) started in August 2016.

(65) An sTTI can be decided to have any duration in time and comprises resources on a number of OFDM or SC-FDMA symbols within a 1 ms SF. As one example shown in FIG. 9, the duration of the uplink short TTI is 0.5 ms, i.e. seven SC-FDMA symbols for the case with normal cyclic prefix. As another example shown in FIG. 10, the durations of the uplink short TTIs within a subframe are of 2 or 3 symbols. Here, the “R” in the figures indicate the DMRS symbols, and “S” indicate the sounding reference signal (SRS) symbols.

(66) Throughout this written description, short PDSCH (sPDSCH) and short PUSCH (sPUSCH) are used to denote the downlink and uplink physical shared channels with sTTIs, respectively.

(67) In order to reduce the DMRS overhead, in LTE Release 14, it has been agreed that DMRS multiplexing/sharing should be supported for uplink short TTI transmissions. More specifically, based on the outcome of a study item on latency reduction, the following are recommended to be supported for the design of DMRS for sPUSCH: For the case of 1-slot TTI length, reuse the current DMRS For the case of less than 1-slot TTI length, support DMRS sharing/multiplexing of consecutive TTIs from one or multiple wireless devices. The DMRS symbols may be

shared/multiplexed between sTTIs, e.g. DMRS symbols as shown in FIGS. 9 to 11. At least 2 contiguous TTIs can be shared/multiplexed.

(68) When the same wireless device is scheduled on consecutive sTTIs, an effective way to reduce the overhead of reference signals for UL data transmission is DMRS sharing. This means that the DMRS is not transmitted in each sTTI. Instead, a certain periodicity in terms of the number of sTTIs is assumed for transmitting DMRS. FIG. 11 illustrates an example of DMRS sharing for the case of a 2/3-symbol sTTI configuration in an uplink subframe. In this example, DMRS transmitted in sTTI 0 and sTTI 3 are used for the channel estimation for sTTI 1 and sTTI 4, respectively.

(69) In case different wireless devices are scheduled in consecutive short TTIs, DMRS sharing cannot be used. Instead, to reduce the DMRS overhead for UL data transmissions, DMRS multiplexing can be considered. DMRS multiplexing means multiple wireless devices share the same SC-FDMA symbol but have separate SC-FDMA symbols for the data. FIG. 12 illustrates examples of DMRS multiplexing for two different 2/3-symbol sTTI configurations within an uplink subframe. The DMRS from different wireless devices are multiplexed in the same SC-FDMA symbol marked with “R”. The orthogonality between the multiplexed DMRS from different wireless devices needs to be ensured in order to guarantee good channel estimation and successful data decoding.

(70) For uplink sTTI transmissions, DMRS multiplexing/sharing can be used to reduce the DMRS overhead. With the same frequency allocation, the DMRS of different wireless devices can be multiplexed on the same SC-FDMA symbol by using different cyclic shifts.

(71) In order to keep the scheduling flexibility, different wireless devices can be allocated with different frequency resources, where part of their frequency allocation is overlapped, as shown in FIG. 13, where reference symbol 2 is for a first wireless device 1 (WD1), reference symbol 4 is for a second wireless device (WD2), data 6 is for WD1 and data 8 is for WD2. DMRS multiplexing should also be supported in this partially overlapped frequency allocation case to reduce the DMRS overhead, and at the same time, keep the scheduling flexibility.

(72) The IFDMA-based DMRS multiplexing method discussed above for MU-MIMO transmission on PUSCH can also be used for supporting DMRS multiplexing on sPUSCH with partially overlapped frequency allocations.

(73) Different from DMRS multiplexing for MU-MIMO transmissions on PUSCH, in most cases, there is at most one DMRS symbol per sPUSCH transmission. This implies that OCC cannot be used for DMRS multiplexing when considering sPUSCH transmissions. A new signaling method needs to be designed in order to support IFDMA-based DMRS multiplexing for sPUSCH transmissions with partially overlapped frequency allocation.

## SUMMARY

(74) Some embodiments advantageously provide a method and network node for signaling demodulation reference symbols, DMRS, configurations of uplink short transmission time interval, sTTI, transmissions, the signaling supporting multiplexing of DMRS of different wireless devices for uplink sTTIs.

(75) According to one aspect, a method in a network node for configuring a wireless device for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The method includes generating an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The further includes transmitting to the wireless device the indication of IFDMA subcarrier configuration.

(76) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the indication of the IFDMA subcarriers is contained in downlink control information (DCI). In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift. In some embodiments, the method further includes indicating whether a DMRS configuration is an IFDMA-based DMRS configuration. In some embodiments, a

sTTI is one of 2 and 3 symbols duration. In some embodiments, the IFDMA has a repetition factor of 2. In some embodiments, a sTTI transmission is a short physical uplink shared channel, sPUSCH, transmission. In some embodiments, the method further includes determining one of whether only IFDMA-based DMRS multiplexing is used for sTTIs and whether both cyclic shift-based DMRS multiplexing and IFDMA-based DMRS multiplexing are used for sTTIs.

(77) According to another aspect, a network node for configuring a wireless device for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The network node includes processing circuitry configured to generate an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The network node further includes a transceiver configured to transmit to the wireless device the indication of IFDMA, subcarrier configuration.

(78) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the indication of the IFDMA subcarriers is contained in downlink control information, DCI. In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift. In some embodiments, the processing circuitry is further configured to indicate whether a DMRS configuration is an IFDMA-based DMRS configuration. In some embodiments, a sTTI is one of 2 and 3 symbols duration. In some embodiments, the IFDMA has a repetition factor of 2. In some embodiments, a sTTI transmission is a short physical uplink shared channel, sPUSCH, transmission. In some embodiments, the processing circuitry is further configured to determine one of whether only IFDMA-based DMRS multiplexing is used for sTTIs and whether both cyclic shift-based DMRS multiplexing and IFDMA-based DMRS multiplexing are used for sTTIs.

(79) According to yet another aspect, a network node for configuring a wireless device for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The network node includes an IFDMA determination module configured to generate an indication of an interleaved frequency division multiple access (IFDMA), subcarrier configuration for DMRS transmission. The network node further includes a transceiver module configured to transmit to the wireless device the indication of IFDMA, subcarrier configuration.

(80) According to another aspect, a method in a wireless device for configuring demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The method includes receiving from a network node an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The method further include configuring DMRS transmissions according to the indication.

(81) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift.

(82) According to yet another aspect, a wireless device for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The wireless device includes a transceiver configured to receive from a network node an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The wireless device further includes processing circuitry configured to configure DMRS transmissions according to the indication.

(83) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift.

(84) According to another aspect, a wireless device for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The wireless device includes a transceiver module configured to receive from a network node an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS

transmission. The wireless device further includes a DMRS configuration module configured to configure DMRS transmissions according to the indication.

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## Description

### BRIEF DESCRIPTION OF THE DRAWINGS

(1) A more complete understanding of the present embodiments, and the attendant advantages and features thereof, will be more readily understood by reference to the following detailed description when considered in conjunction with the accompanying drawings wherein:

- (2) FIG. 1 is a timing diagram of a radio frame;
- (3) FIG. 2 is a diagram of resource elements;
- (4) FIG. 3 is an LTE uplink resource grid;
- (5) FIG. 4 is a diagram of 3 OFDM symbols;
- (6) FIG. 5 is a diagram of SC-OFDM symbol resource elements;
- (7) FIG. 6 is a diagram of SC-OFDM symbols;
- (8) FIG. 7 is a diagram DMRS sequences;
- (9) FIG. 8 is a diagram of DMRS sequences;
- (10) FIG. 9 is diagram of sTTIs;
- (11) FIG. 10 is a diagram of sTTIs;
- (12) FIG. 11 is a diagram of sTTIs;
- (13) FIG. 12 is a diagram of sTTIs;
- (14) FIG. 13 is a diagram of allocation of frequency resources among WDs;
- (15) FIG. 14 is a block diagram of a wireless communication network constructed in accordance with principles set forth herein;
- (16) FIG. 15 is a block diagram of a network node constructed in accordance with principles set forth herein;
- (17) FIG. 16 is a block diagram of an alternative embodiment of a network node;
- (18) FIG. 17 is a block diagram of a wireless device constructed in accordance with principles set forth herein;
- (19) FIG. 18 is a block diagram of an alternative embodiment of a wireless device; and
- (20) FIG. 19 is a flowchart of an exemplary process performed at a wireless device.
- (21) FIG. 20 is a flowchart of an exemplary process for DMRS signaling;
- (22) FIG. 21 is a diagram of allocation of frequency resources among WDs;
- (23) FIG. 22 is a diagram of allocation of frequency resources among WDs;
- (24) FIG. 23 is a diagram of allocation of frequency resources among WDs;
- (25) FIG. 24 is a diagram of allocation of frequency resources among WDs;
- (26) FIG. 25 is a diagram of allocation of frequency resources among WDs; and
- (27) FIG. 26 is a diagram of allocation of frequency resources among WDs.

### DETAILED DESCRIPTION

(28) Before describing in detail exemplary embodiments, it is noted that the embodiments reside primarily in combinations of apparatus components and processing steps related to signaling of DMRS configurations for uplink short transmission time interval (sTTI) transmissions.

Accordingly, components have been represented where appropriate by conventional symbols in the drawings, showing only those specific details that are pertinent to understanding the embodiments so as not to obscure the disclosure with details that will be readily apparent to those of ordinary skill in the art having the benefit of the description herein.

(29) As used herein, relational terms, such as “first” and “second,” “top” and “bottom,” and the like, may be used solely to distinguish one entity or element from another entity or element without necessarily requiring or implying any physical or logical relationship or order between such entities

or elements.

(30) Note that functions described herein as being performed by a wireless device or a network node may be distributed over a plurality of wireless devices and/or network nodes. In other words, it is contemplated that the functions of the network node and wireless device described herein are not limited to performance by a single physical device and, in fact, can be distributed among several physical devices.

(31) Embodiments provide methods of signaling DMRS configurations of uplink short TTI transmissions. In particular, the signaling methods support multiplexing of DMRS of different wireless devices for uplink short TTI transmissions. In some embodiments, a network node configures a wireless device for multiplexing DMRS by generating an indication of interleaved frequency division multiple access (IFDMA) subcarrier configuration and by transmitting the indication of the IFDMA subcarrier configuration to the wireless device. The IFDMA subcarrier configuration indication may include which subcarriers are to be used for DMRS transmission and may further indicate whether or not IFDMA is to be used by the wireless device.

(32) An LTE subframe lasting 1 ms contains 14 OFDM symbols for normal CP. A New Radio (5G), NR, subframe may have a fixed duration of 1 ms and may therefore contain a different number of OFDM symbols for different subcarrier spacings. An LTE slot corresponds to 7 OFDM symbols for normal CP. An NR slot corresponds to 7 or 14 OFDM symbols; at 15 kHz subcarrier spacing, a slot with 7 OFDM symbols occupies 0.5 ms. Concerning NR terminology, reference is made to 3GPP TR 38.802 v14.0.0 and later versions.

(33) Aspects of this disclosure may be applicable to either LTE or NR radio communications. References to a short TTI may alternatively be considered as a mini-slot, according to NR terminology. The mini-slot may have a length of 1 symbol, 2 symbols, 3 or more symbols, or a length of between 1 symbol and a NR slot length minus 1 symbol. The short TTI may have a length of 1 symbol, 2 symbols, 3 or more symbols, an LTE slot length (7 symbols) or a length of between 1 symbol and a LTE subframe length minus 1 symbol. The short TTI, or mini-slot, may be considered as having a length less than 1 ms or less than 0.5 ms.

(34) Returning to the drawing figures, in which like elements are referred to by like reference numerals, there is shown in FIG. 14 a block diagram of a wireless communication system 10 constructed according to principles set forth herein. The wireless communication network 10 includes a cloud 12 which may include the Internet and/or the public switched telephone network (PSTN). Cloud 12 may also serve as a backhaul network of the wireless communication network 10. The wireless communication network 10 includes one or more network nodes 14A and 14B, which may communicate directly via an X2 interface in LTE embodiments, and are referred to collectively as network nodes 14. It is contemplated that other interface types can be used for communication between network nodes 14 for other communication protocols such as New Radio (NR). The network nodes 14 may serve wireless devices 16A and 16B, referred to collectively herein as wireless devices 16. Note that, although only two wireless devices 16 and two network nodes 14 are shown for convenience, the wireless communication network 10 may typically include many more wireless devices (WDs) 16 and network nodes 14. Further, in some embodiments, WDs 16 may communicate directly using what is sometimes referred to as a side link connection.

(35) The term “wireless device” or mobile terminal used herein may refer to any type of wireless device communicating with a network node 14 and/or with another wireless device 16 in a cellular or mobile communication system 10. Examples of a wireless device 16 are user equipment (UE), target device, device to device (D2D) wireless device, machine type wireless device or wireless device capable of machine to machine (M2M) communication, PDA, tablet, smart phone, laptop embedded equipped (LEE), laptop mounted equipment (LME), USB dongle, etc.

(36) The term “network node” used herein may refer to any kind of radio base station in a radio network which may further comprise any base transceiver station (BTS), base station controller

(BSC), radio network controller (RNC), evolved Node B (eNB or eNodeB), NR gNodeB, NR gNB, Node B, multi-standard radio (MSR) radio node such as MSR BS, relay node, donor node controlling relay, radio access point (AP), transmission points, transmission nodes, Remote Radio Unit (RRU) Remote Radio Head (RRH), nodes in distributed antenna system (DAS), etc.

(37) Although embodiments are described herein with reference to certain functions being performed by network node **14**, it is understood that the functions can be performed in other network nodes and elements. It is also understood that the functions of the network node **14** can be distributed across network cloud **12** so that other nodes can perform one or more functions or even parts of functions described herein.

(38) As shown in FIG. **14**, the network node **14** includes a transceiver **18** configured to transmit an indication of an interleaved frequency division multiple access subcarrier configuration for DMRS transmission. Complementarily, the wireless device **16** includes a transceiver to receive the indication of an interleaved frequency division multiple access subcarrier configuration for DMRS transmission.

(39) FIG. **15** is a block diagram of a network node **14** for signaling demodulation reference symbol, DMRS, configurations of uplink short transmission time interval, sTTI, transmissions, the signaling supporting multiplexing of DMRS of different wireless devices for uplink sTTIs. In some examples, the DMRS symbols may be shared/multiplexed between different wireless devices, each using one or more sTTIs, as shown in FIGS. **6** and **9** to **11**. Any of the arrangements or features described herein, including those in the background section, may be combined with the arrangements or features described in any example of the disclosure. For example, the wireless device **16** or network node **14** according to any example may be operating in a multiple input multiple output (MIMO) mode or a multi-user (MU)-MIMO mode.

(40) The network node **14** has processing circuitry **22**. In some embodiments, the processing circuitry may include a memory **24** and processor **26**, the memory **24** containing instructions which, when executed by the processor **26**, configure processor **26** to perform the one or more functions described herein. In addition to a traditional processor and memory, processing circuitry **22** may comprise integrated circuitry for processing and/or control, e.g., one or more processors and/or processor cores and/or FPGAs (Field Programmable Gate Array) and/or ASICs (Application Specific Integrated Circuitry).

(41) Processing circuitry **22** may include and/or be connected to and/or be configured for accessing (e.g., writing to and/or reading from) memory **24**, which may comprise any kind of volatile and/or non-volatile memory, e.g., cache and/or buffer memory and/or RAM (Random Access Memory) and/or ROM (Read-Only Memory) and/or optical memory and/or EPROM (Erasable Programmable Read-Only Memory). Such memory **24** may be configured to store code executable by control circuitry and/or other data, e.g., data pertaining to communication, e.g., configuration and/or address data of nodes, etc. Processing circuitry **22** may be configured to control any of the methods described herein and/or to cause such methods to be performed, e.g., by processor **26**. Corresponding instructions may be stored in the memory **24**, which may be readable and/or readably connected to the processing circuitry **22**. In other words, processing circuitry **22** may include a controller, which may comprise a microprocessor and/or microcontroller and/or FPGA (Field-Programmable Gate Array) device and/or ASIC (Application Specific Integrated Circuit) device. It may be considered that processing circuitry **22** includes or may be connected or connectable to memory, which may be configured to be accessible for reading and/or writing by the controller and/or processing circuitry **22**.

(42) The memory **24** is configured to store an IFDMA subcarrier configuration indicator **30** which may be transmitted to a wireless device **16**. The processor **26** includes an IFDMA determination unit **32** configured to generate an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS transmission. The transceiver **28** is configured to transmit to the wireless device **16** the indication of IFDMA subcarrier configuration.

(43) FIG. 16 is a block diagram of an alternative embodiment of a network node 14 for configuring a wireless device for multiplexing demodulation reference signals, DMRS, during short transmission time intervals, sTTIs. The network node 14 may be implemented at least in part by software executable by a processor to perform functions described herein. An IFDMA determination module 33 generates an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS transmission. The transceiver module 29 transmits to the wireless device the indication of IFDMA subcarrier configuration.

(44) FIG. 17 is a block diagram of a wireless device 16. The wireless device 16 has processing circuitry 42. In some embodiments, the processing circuitry may include a memory 44 and processor 46, the memory 44 containing instructions which, when executed by the processor 46, configure processor 46 to perform the one or more functions described herein. In addition to a traditional processor and memory, processing circuitry 42 may comprise integrated circuitry for processing and/or control, e.g., one or more processors and/or processor cores and/or FPGAs (Field Programmable Gate Array) and/or ASICs (Application Specific Integrated Circuitry).

(45) Processing circuitry 42 may include and/or be connected to and/or be configured for accessing (e.g., writing to and/or reading from) memory 44, which may comprise any kind of volatile and/or non-volatile memory, e.g., cache and/or buffer memory and/or RAM (Random Access Memory) and/or ROM (Read-Only Memory) and/or optical memory and/or EPROM (Erasable Programmable Read-Only Memory). Such memory 64 may be configured to store code executable by control circuitry and/or other data, e.g., data pertaining to communication, e.g., configuration and/or address data of nodes, etc. Processing circuitry 42 may be configured to control any of the methods described herein and/or to cause such methods to be performed, e.g., by processor 46. Corresponding instructions may be stored in the memory 44, which may be readable and/or readably connected to the processing circuitry 42. In other words, processing circuitry 42 may include a controller, which may comprise a microprocessor and/or microcontroller and/or FPGA (Field-Programmable Gate Array) device and/or ASIC (Application Specific Integrated Circuit) device. It may be considered that processing circuitry 42 includes or may be connected or connectable to memory, which may be configured to be accessible for reading and/or writing by the controller and/or processing circuitry 42.

(46) The memory 44 is configured to store an IFDMA subcarrier configuration indicator 50. The processor 46 implements a DMRS configuration unit 54 that is configured to configure DMRS transmissions according to the IFDMA subcarrier configuration indicator 50. The transceiver 52 is configured to receive from the network node 14 the indication of IFDMA subcarrier configuration.

(47) FIG. 18 is a block diagram of an alternative embodiment of a wireless device 16 for configuring DMRS transmissions according to an IFDMA subcarrier configuration indicator received from a network node. The wireless device 16 may be implemented at least in part by software executable by a processor to perform functions described herein. The DMRS configuration module 55 is configured to configure DMRS transmissions according to the IFDMA subcarrier configuration indicator 50. The transceiver module 53 is configured to receive from the network node 14 the indication of IFDMA subcarrier configuration.

(48) FIG. 19 is a flowchart of an exemplary process in a network node 14 for configuring a wireless device for multiplexing demodulation reference signals, DMRS, during short transmission time intervals (sTTIs). The process includes generating an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission (block S100). The process further includes transmitting to the wireless device the indication of IFDMA subcarrier configuration (block S102). Alternatively, in some embodiments, a method may constitute the sole step of transmitting to the wireless device 16 an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS transmission. In some examples, a method in a network node may include receiving a DMRS from one or more wireless devices 16, in which the DMRS have an interleaved frequency division multiple access, IFDMA, subcarrier



configuration.

(49) FIG. 20 is a flowchart of an exemplary process in a wireless device 16 for configuring demodulation reference signals, DMRS, during short transmission time intervals (sTTIs). The process includes receiving from a network node an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS transmission (block S104). The process also includes configuring DMRS transmissions according to the indication (block S106). In some examples, the method may further comprise the wireless device 16 transmitting a DMRS according to an interleaved frequency division multiple access, IFDMA, subcarrier configuration. Alternatively, in some embodiments, the method may constitute the sole step of receiving at the wireless device 16 an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS transmission. Alternatively, in some embodiments, the method may constitute the sole step of the wireless device 16 transmitting a DMRS according to an interleaved frequency division multiple access, IFDMA, subcarrier configuration.

(50) In case only an IFDMA-based method is used for DMRS multiplexing for uplink short TTI transmissions, signaling options are proposed: Option 1: The subcarrier configuration for DMRS transmission is signaled by radio resource control (RRC). The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the legacy cyclic shift mapping table is reused for sPUSCH transmissions. Option 2: The subcarrier configuration for DMRS transmission is signaled by one bit in UL DCI. The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the legacy cyclic shift mapping table is reused for sPUSCH transmissions. Option 3: The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the subcarrier configuration is implicitly indicated by the cyclic shift index.

(51) In case both cyclic-shift-based DMRS multiplexing ( $RPF=1$ ) and IFDMA-based DMRS multiplexing ( $RPF \geq 2$ ) method are used for multi-UE sPUSCH transmissions, three signaling options are proposed: Option 1: RRC signaling to indicate whether the DMRS configuration is an IFDMA-based or no-IFDMA based DMRS configuration, and the subcarrier configurations for DMRS transmission. The legacy cyclic shift mapping table is reused. The cyclic shift parameter is indicated by the legacy 3-bit cyclic shift field in UL DCI. Option 2: Introduce a new field with a single bit in UL DCI to indicate whether the DMRS configuration is a IFDMA-based DMRS configuration or not. The legacy cyclic shift mapping table is reused. The cyclic shift parameter and the subcarrier configuration are indicated by the legacy 3-bit cyclic shift field in UL DCI. Option 3: Use the legacy 3-bit cyclic shift field to indicate the DMRS sequence configuration, including both the selected DMRS multiplexing method (including the subcarrier configuration) and the cyclic shift parameter, by adapting the legacy cyclic shift mapping table for sPUSCH transmissions.

(52) With the proposed solutions, it is possible to multiplex the DMRS of different wireless devices on the same SC-FDMA symbol for uplink short TTI transmissions, in order to reduce the DMRS overhead. In particular, DMRS multiplexing can be supported for uplink short TTI transmissions from multiple wireless devices, whose allocated frequency bandwidth are fully overlapped or partially overlapped.

(53) Methods of signaling DMRS configurations of uplink short TTI transmissions are presented. In particular, the signaling methods support multiplexing of DMRS of different wireless devices for uplink short TTI transmissions. As explained above, in one embodiment a short TTI is one less than 1 milli-second and a short PUSCH is an uplink shared channel with sTTIs.

(54) In an embodiment, only the IFDMA-based DMRS multiplexing method is used for uplink short TTI transmissions. The subcarrier configuration for DMRS transmission is signaled by RRC. The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the legacy cyclic shift mapping table is reused for sPUSCH transmissions.

(55) In an embodiment, only the IFDMA-based DMRS multiplexing method is used for uplink short TTI transmissions. The subcarrier configuration for DMRS transmission is signaled by one bit in UL DCI.

(56) In another embodiment, only the IFDMA-based DMRS multiplexing method is used for uplink short TTI transmissions. The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the subcarrier configuration is implicitly indicated by the cyclic shift index.

(57) In an embodiment, both cyclic-shift-based DMRS multiplexing ( $RPF=1$ ) and IFDMA-based DMRS multiplexing method ( $RPF=2$ ) are used for uplink short TTI transmissions. The DMRS configuration for each wireless device is signaled from the eNB based on one of the three options below. Option 1: RRC signaling to indicate whether the DMRS configuration is a IFDMA-based or no-IFDMA based DMRS configuration, and the subcarrier configurations for DMRS transmission. The legacy cyclic shift mapping table is reused. The cyclic shift parameter is indicated by the legacy 3-bit cyclic shift field in UL DCI. Option 2: Introduce a new field with a single bit in UL DCI to indicate whether the DMRS configuration is a IFDMA-based DMRS configuration. The legacy cyclic shift mapping table is reused. The cyclic shift parameter is indicated by the legacy 3-bit cyclic shift field in UL DCI. Option 3: Use the legacy 3-bit cyclic shift field to indicate the DMRS sequence configuration, including both the selected DMRS multiplexing method and the cyclic shift parameter, by adapting the legacy cyclic shift mapping table for sPUSCH transmissions.

(58) In an embodiment, the RRC signaling explicitly indicates the subcarrier configuration for IFDMA-based DMRS transmission of the wireless device. In another embodiment, the subcarrier configuration for IFDMA-based DMRS transmission is signaled by a bit-field in UL DCI. In another embodiment, the subcarrier configuration for IFDMA-based DMRS transmission is implicitly indicated by a predefined mapping between subcarrier configurations and cyclic shift indices.

(59) Some examples of how to multiplex DMRS from different wireless devices **16** for sPUSCH transmissions, considering different frequency allocation cases, are given. The examples of signaling of the DMRS multiplexing are also given.

(60) FIG. **21** illustrates the case where the multiplexed wireless devices **16** are allocated with the same uplink frequency bandwidth. In this case, three different DMRS multiplexing approaches can be used. The legends **62**, **64** and **66** are RS and data for WD1, e.g., WD **16a** (FIG. **14**) and WD2, e.g., WD **16b** (FIG. **14**) as indicated.

(61) Approach 1: DMRS from different wireless devices are multiplexed on the same SC-FDMA symbol with  $RPF=1$  by using different cyclic shifts.

(62) Approach 2: DMRS from different wireless devices are multiplexed on the same SC-FDMA symbol but different subcarriers with  $RPF=2$ , as shown in FIG. **22**, where the legends **68**, **70**, **72** and **74** are RS and data for WD1, e.g., WD **16a** (FIG. **14**) and WD2, e.g., WD **16b** (FIG. **14**) as indicated.

(63) Approach 3: As shown in FIG. **23**, the DMRS of one wireless device (WD2) is configured with  $RPF=1$ , where the legend **78** indicates RS for WD2, e.g., WD **16b** (FIG. **14**) as indicated. The DMRS of the other wireless device is configured with  $RPF=2$ , as shown in FIG. **23**, where the legend **76** indicates RS for WD1, e.g., WD **16a** (FIG. **14**) as indicated.

(64) When different wireless devices are allocated with partially overlapped frequency bandwidth, approaches 2 and 3 can be used.

(65) FIGS. **24**, **25** and **26** illustrate examples of IFDMA-based DMRS multiplexing of different wireless devices, where the sPUSCH transmissions are allocated with partially overlapped frequency bandwidth, where the legends **80-110** are RS and data for WD1, e.g., WD **16a** (FIG. **14**), WD2, e.g., WD **16b** (FIG. **14**) and WD3, e.g., another WD **16** as indicated.

(66) Signaling for Only IFDMA-Based DMRS Multiplexing for sPUSCH

(67) In one example, only the IFDMA-based DMRS multiplexing method ( $RPF \geq 2$ ) is used for uplink short TTI transmissions.

(68) Solution 1: The RRC signaling explicitly indicates the subcarriers used for DMRS transmission of the wireless device, e.g., the odd or the even subcarriers within the allocated

frequency bandwidth if RPF=2. The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the legacy cyclic shift mapping table is reused.

(69) Solution 2: Introduce a one-bit field in UL DCI to explicitly indicate the subcarriers used for DMRS transmission of the wireless device, e.g., the odd or the even subcarriers within the allocated frequency bandwidth if RPF=2. The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the legacy cyclic shift mapping table is reused.

(70) Solution 3: Implicitly indicate the subcarrier allocation for DMRS by predefining a mapping rule such that a subset of cyclic shifts indicates IFDMA RPF-2 in odd subcarriers, and the complement subset of cyclic shifts indicates IFDMA RPF-2 in even subcarriers. In this case, RRC signaling is not needed.

(71) Table 5 gives an example of the mapping between cyclic shifts and IFDMA subcarrier configurations. The mapping is designed to guarantee the best possible cyclic shift separation in scenarios with high Doppler spread and low delay spread. The cyclic shifts with the highest minimum separation for the first layer are chosen in this example. From the legacy cyclic shift field mapping in Table 3, the cyclic shifts possible for the 1st layer (i.e.,  $\lambda=0$ ) are {0, 2, 3, 4, 6, 8, 9, 10}. The subset {0, 3, 6, 9} provides the best possible minimum cyclic shift difference of 3 for the 1st layer. This corresponds to 000, 001, 010, and 111 in the cyclic shift field in Table 3 above, and thus these cyclic shifts are mapped to the same subcarrier configuration, e.g., the odd subcarriers, for IFDMA-based DMRS multiplexing with RPF=2.

(72) TABLE-US-00005 TABLE 5 Cyclic Shift Field in uplink- related n.sub.DMRS,  $\lambda$ .sup.(2)

IFDMA DCI format	$\lambda = 0$	$\lambda = 1$	$\lambda = 2$	$\lambda = 3$	Configuration
Odd subcarriers	000	0	6	3	9
Odd subcarriers	001	6	0	9	3
Odd subcarriers	010	3	9	6	0
Odd subcarriers	011	4	10	7	1
Even subcarriers	100	2	8	5	11
Even subcarriers	101	8	2	11	5
Even subcarriers	110	10	4	1	7
Even subcarriers	111	9	3	0	6

subcarriers

(73) Consider the IFDMA-based DMRS-multiplexing case shown in FIG. 24, and assume that both wireless device 1 and wireless device 2 are configured with two-layer uplink sTTI transmission. Based on Table 5, one way of signaling the DMRS configuration for wireless device 1 and wireless device 2 is to set the cyclic shift field in UL UCI for wireless device 1 to 000, and set the cyclic shift field for wireless device 2 to 101, to provide the best possible cyclic shift separation.

(74) Signaling for Both IFDMA-Based and No-IFDMA Based DMRS Multiplexing for sPUSCH

(75) In another example, both IFDMA-based (RPF $\geq$ 2) and no-IFDMA based (RPF=1) DMRS multiplexing methods are supported for uplink short TTI transmissions.

(76) Solution 1: The RRC signaling explicitly indicates the subcarrier configuration used for DMRS transmission of the wireless device, i.e., IFDMA RPF-2 in odd subcarriers, IFDMA RPF-2 in even subcarriers, no-IFDMA in all allocated subcarriers. The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the legacy cyclic shift mapping table is reused.

(77) Solution 2: Introduce a new field with 1 bit in UL DCI to indicate the subcarrier configuration for DMRS transmission of the wireless device. The cyclic shift is signaled by the legacy 3-bit cyclic shift field in UL DCI and the legacy cyclic shift mapping table is reused.

(78) Solution 3: Implicitly indicate the subcarrier configuration, including the selected DMRS multiplexing scheme (IFDMA or no-IFDMA), for DMRS transmission by the cyclic shift signaled in 3-bit cyclic shift field in UL DCI. In this case, there is no extra signaling overhead.

(79) Table 6 gives an example of the mapping between cyclic shifts and IFDMA/no-IFDMA configurations. The mapping is designed to guarantee the best possible cyclic shift separation in scenarios with high Doppler spread and low delay spread. The cyclic shifts with the highest minimum separation for the first layer are chosen in this example. From the legacy cyclic shift field mapping in Table 3, the cyclic shifts possible for the 1st layer are 10, 2, 3, 4, 6, 8, 9, 101. The subset {0, 3, 6, 9} provides the best possible minimum cyclic shift difference of 3 for the 1st layer. This corresponds to 000, 001, 010, and 111 in the cyclic shift field in Table 3, and thus these cyclic shifts are reserved for no-IFDMA based DMRS configuration.

(80) TABLE-00006 TABLE 6 Cyclic Shift Field in uplink- related n.sub.DMRS,  $\lambda$ .sup.(2)  
DMRS DCI format  $\lambda = 0$   $\lambda = 1$   $\lambda = 2$   $\lambda = 3$  Configuration 000 0 6 3 9 No IFDMA 001 6 0 9 3 No IFDMA 010 3 9 6 0 No IFDMA 011 4 10 7 1 Odd subcarriers 100 2 8 5 11 Even subcarriers 101 8 2 11 5 Odd subcarriers 110 10 4 1 7 Even subcarriers 111 9 3 0 6 No IFDMA

(81) The remaining 4 cyclic shifts (correspond to 011, 100, 101, 110 in cyclic shift field), two cyclic shifts each should be reserved for IFDMA RPF-2 in odd and even subcarriers. The two code points within odd subcarriers are selected such that they have the highest minimum separation for the first layer, and they do not have overlapped cyclic shifts when supporting multi-layer sPUSCH transmissions. Based on the above rule, the pair (011, 101) can be reserved for IFDMA RPF-2 in odd subcarriers. The remaining code point pair (100, 110) which also satisfy the above rule can be reserved for IFDMA RPF-2 in even subcarriers.

(82) Consider the IFDMA-based DMRS-multiplexing case shown in FIG. 24, and assume that both wireless device 1 and wireless device 2 are configured with two-layer uplink sTTI transmission. Based on Table 6, one way of signaling the DMRS configuration for wireless device 1 and wireless device 2 is to set the cyclic shift filed in UL UCI for wireless device 1 to 011, and set the cyclic shift filed for wireless device 2 to 100, to provide the best possible cyclic separation between different wireless devices and different transmission layers.

(83) Thus, according to one aspect, a method in a network node 14 for configuring a wireless device 16 for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The method includes generating an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The further includes transmitting to the wireless device 16 the indication of IFDMA subcarrier configuration.

(84) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the indication of the IFDMA subcarriers is contained in downlink control information (DCI). In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift. In some embodiments, the method further includes indicating whether a DMRS configuration is an IFDMA-based DMRS configuration. In some embodiments, a sTTI is one of 2 and 3 symbols duration. In some embodiments, the IFDMA has a repetition factor of 2. In some embodiments, a sTTI transmission is a short physical uplink shared channel, sPUSCH, transmission. In some embodiments, the method further includes determining one of whether only IFDMA-based DMRS multiplexing is used for sTTIs and whether both cyclic shift-based DMRS multiplexing and IFDMA-based DMRS multiplexing are used for sTTIs.

(85) According to another aspect, a network node 14 for configuring a wireless device 16 for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The network node 14 includes processing circuitry 22 configured to generate an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The network node 14 further includes a transceiver 28 configured to transmit to the wireless device 16 the indication of IFDMA, subcarrier configuration.

(86) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the indication of the IFDMA subcarriers is contained in downlink control information (DCI). In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift. In some embodiments, the processing circuitry 22 is further configured to indicate whether a DMRS configuration is an IFDMA-based DMRS configuration. In some embodiments, a sTTI is one of 2 and 3 symbols duration. In some embodiments, the IFDMA has a repetition factor of 2. In some embodiments, a sTTI transmission is a short physical uplink shared channel, sPUSCH, transmission. In some embodiments, the processing circuitry 22 is further configured to determine one of whether only IFDMA-based DMRS multiplexing is used for

sTTIs and whether both cyclic shift-based DMRS multiplexing and IFDMA-based DMRS multiplexing are used for sTTIs.

(87) According to yet another aspect, a network node **14** for configuring a wireless device **16** for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The network node **14** includes an IFDMA determination module **33** configured to generate an indication of an interleaved frequency division multiple access (IFDMA), subcarrier configuration for DMRS transmission. The network node **14** further includes a transceiver module **29** configured to transmit to the wireless device **16** the indication of IFDMA, subcarrier configuration.

(88) According to another aspect, a method in a wireless device **16** for configuring demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The method includes receiving from a network node **14** an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The method further include configuring DMRS transmissions according to the indication.

(89) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift.

(90) According to yet another aspect, a wireless device **16** for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The wireless device **16** includes a transceiver **52** configured to receive from a network node **14** an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission. The wireless device **16** further includes processing circuitry **42** configured to configure DMRS transmissions according to the indication.

(91) According to this aspect, in some embodiments, the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission. In some embodiments, the IFDMA subcarrier configuration is indicated by a field indicating a cyclic shift.

(92) According to another aspect, a wireless device **16** for multiplexing demodulation reference signals (DMRS) during short transmission time intervals (sTTIs) is provided. The wireless device **16** includes a transceiver module **53** configured to receive from a network node **14** an indication of an interleaved frequency division multiple access, IFDMA, subcarrier configuration for DMRS transmission. The wireless device **16** further includes a DMRS configuration module **55** configured to configure DMRS transmissions according to the indication.

(93) Some embodiments are as follows:

(94) Embodiment 1. A method in a network node of signaling demodulation reference symbols, DMRS, configurations of uplink short transmission time interval, sTTI, transmissions, the signaling supporting multiplexing of DMRS of different wireless devices for uplink sTTIs, the method including:

(95) signaling a cyclic shift of the DMRS by using a legacy cyclic shift field in the uplink-related downlink control information, DCI, format; and

(96) reusing a legacy cyclic shift mapping table for short physical uplink shared channel transmissions.

(97) Embodiment 2. The method of Embodiment 1, further comprising signaling a subcarrier configuration of DMRS transmission via a radio resource control, RRC, message.

(98) Embodiment 3. The method of Embodiment 1, further comprising signaling a subcarrier configuration for DMRS transmission via a bit in the uplink-related DCI.

(99) Embodiment 4. The method of Embodiment 1, wherein the cyclic shift is signaled by using the legacy cyclic shift field in the uplink-related DCI and the subcarrier configuration is implicitly indicated by a cyclic shift index.

(100) Embodiment 5. The method of Embodiment 1, further comprising using radio resource control, RRC, messaging to indicate non-interleaved frequency division multiplexing, non-IFDMA,

DMRS multiplexing.

(101) Embodiment 6. The method of Embodiment 1, further comprising allocating partially overlapping frequencies to different wireless device.

(102) Embodiment 7. The method of Embodiment 1, wherein a field in the uplink-related DCI is used to indicate non-interleaved frequency division multiplexing, non-IFDMA, DMRS multiplexing.

(103) Embodiment 8. The method of Embodiment 1, wherein a legacy cyclic shift field is used to indicate non-interleaved frequency division multiplexing, non-IFDMA, DMRS multiplexing.

(104) Embodiment 9. The method of Embodiment 1, wherein the DMRS configuration specifies at least one of odd subcarriers and even subcarriers.

(105) Embodiment 10. A network node for signaling demodulation reference symbol, DMRS, configurations of uplink short transmission time interval, sTTI, transmissions, the signaling supporting multiplexing of DMRS of different wireless devices for uplink sTTIs, the network node comprising:

(106) processing circuitry including a memory and a processor: the memory configured to store downlink control information, DCI; and the processor configured to: signal a cyclic shift of the DMRS by a legacy cyclic shift field in the uplink-related downlink control information, DCI; and reuse a legacy cyclic shift mapping table for short physical uplink shared channel transmissions.

(107) Embodiment 11. A network node for signaling demodulation reference symbol, DMRS, configurations of uplink short transmission time interval, sTTI, transmissions, the signaling supporting multiplexing of DMRS of different wireless devices for uplink sTTIs, the network node comprising:

(108) a memory module configured to store downlink control information, DCI;

(109) a signaling module configured to signal a cyclic shift of the DMRS by a legacy cyclic shift field in the uplink-related downlink control information, DCI; and

(110) a table reuse module configured to reuse a legacy cyclic shift mapping table for short physical uplink shared channel transmissions.

(111) Embodiment 12. A method in a wireless device, the method comprising:

(112) receiving signaling from a network node, the signaling including a cyclic shift of a demodulation reference symbol, DMRS; and

(113) decoding the signaling to obtain the DMRS.

(114) Embodiment 13. A wireless device, comprising: a transceiver configured to receive signaling from a network node, the signaling including a cyclic shift of a demodulation reference symbol, DMRS; processing circuitry including a memory and a processor: the memory configured to store the DMRS; and the processor configured to decode the signaling to obtain the DMRS.

(115) Embodiment 14. A wireless device, comprising: a transceiver module configured to receive signaling from a network node, the signaling including a cyclic shift of a demodulation reference symbol, DMRS; a memory module configured to store the DMRS; a decoding module configured to decode the signaling to obtain the DMRS.

(116) Embodiment 15. A method performed by a network node of a wireless network, the method comprising:

(117) signaling a demodulation reference symbol, DMRS, configuration for a wireless device in an uplink short TTI, sTTI, transmission, the DMRS configuration being usable by the wireless device and the network node, the DMRS configuration defining at least the cyclic shift and the subcarrier configuration for the DMRS transmission.

(118) Embodiment 16. The method of Embodiment 15, wherein the sTTI has a predetermined duration in time and comprises resources on a number of OFDM or SC-FDMA symbols within a subframe.

(119) Embodiment 17. A network node for a wireless network, the network node comprising:

(120) processing circuitry, the processing circuitry being configured to. signal a demodulation

reference symbol, DMRS, configuration for a wireless device in an uplink short TTI, sTTI, transmission, the DMRS configuration being usable by the wireless device and the network node, the DMRS configuration defining at least the cyclic shift and the subcarrier configuration for the DMRS transmission.

(121) Embodiment 18. The network node of Embodiment 17, wherein the sTTI has a predetermined duration in time and comprises resources on a number of OFDM or SC-FDMA symbols within a subframe.

(122) As will be appreciated by one of skill in the art, the concepts described herein may be embodied as a method, data processing system, and/or computer program product. Accordingly, the concepts described herein may take the form of an entirely hardware embodiment, an entirely software embodiment or an embodiment combining software and hardware aspects all generally referred to herein as a “circuit” or “module.” Furthermore, the disclosure may take the form of a computer program product on a tangible computer usable storage medium having computer program code embodied in the medium that can be executed by a computer. Any suitable tangible computer readable medium may be utilized including hard disks, CD-ROMs, electronic storage devices, optical storage devices, or magnetic storage devices.

(123) Some embodiments are described herein with reference to flowchart illustrations and/or block diagrams of methods, systems and computer program products. It will be understood that each block of the flowchart illustrations and/or block diagrams, and combinations of blocks in the flowchart illustrations and/or block diagrams, can be implemented by computer program instructions. These computer program instructions may be provided to a processor of a general purpose computer, special purpose computer, or other programmable data processing apparatus to produce a machine, such that the instructions, which execute via the processor of the computer or other programmable data processing apparatus, create means for implementing the functions/acts specified in the flowchart and/or block diagram block or blocks.

(124) These computer program instructions may also be stored in a computer readable memory or storage medium that can direct a computer or other programmable data processing apparatus to function in a particular manner, such that the instructions stored in the computer readable memory produce an article of manufacture including instruction means which implement the function/act specified in the flowchart and/or block diagram block or blocks.

(125) The computer program instructions may also be loaded onto a computer or other programmable data processing apparatus to cause a series of operational steps to be performed on the computer or other programmable apparatus to produce a computer implemented process such that the instructions which execute on the computer or other programmable apparatus provide steps for implementing the functions/acts specified in the flowchart and/or block diagram block or blocks.

(126) It is to be understood that the functions/acts noted in the blocks may occur out of the order noted in the operational illustrations. For example, two blocks shown in succession may in fact be executed substantially concurrently or the blocks may sometimes be executed in the reverse order, depending upon the functionality/acts involved. Although some of the diagrams include arrows on communication paths to show a primary direction of communication, it is to be understood that communication may occur in the opposite direction to the depicted arrows.

(127) Computer program code for carrying out operations of the concepts described herein may be written in an object oriented programming language such as Java® or C++. However, the computer program code for carrying out operations of the disclosure may also be written in conventional procedural programming languages, such as the “C” programming language. The program code may execute entirely on the user's computer, partly on the user's computer, as a stand-alone software package, partly on the user's computer and partly on a remote computer or entirely on the remote computer. In the latter scenario, the remote computer may be connected to the user's computer through a local area network (LAN) or a wide area network (WAN), or the connection

may be made to an external computer (for example, through the Internet using an Internet Service Provider).

(128) Many different embodiments have been disclosed herein, in connection with the above description and the drawings. It will be understood that it would be unduly repetitious and obfuscating to literally describe and illustrate every combination and subcombination of these embodiments. Accordingly, all embodiments can be combined in any way and/or combination, and the present specification, including the drawings, shall be construed to constitute a complete written description of all combinations and subcombinations of the embodiments described herein, and of the manner and process of making and using them, and shall support claims to any such combination or subcombination.

(129) It will be appreciated by persons skilled in the art that the embodiments described herein are not limited to what has been particularly shown and described herein above. In addition, unless mention was made above to the contrary, it should be noted that all of the accompanying drawings are not to scale. A variety of modifications and variations are possible in light of the above teachings without departing from the scope of the following claims.

## Claims

1. A method in a network node configured to communicate with a wireless device, the method comprising: transmitting to the wireless device an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for demodulation reference signal (DMRS) transmission, IFDMA subcarrier configuration being indicated by a first field indicating a cyclic shift and a second field separate from the first field indicating whether the IFDMA subcarrier configuration for DMRS transmission is one of a no-IFDMA-based DMRS configuration with a repetition factor (RPF) of 1 and an IFDMA-based DMRS configuration with an RPF of 2.
2. The method of claim 1, wherein the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission.
3. The method of claim 1, wherein the indication of the IFDMA subcarrier configuration is contained in downlink control information.
4. A network node configured to communicate with a wireless device, DMRS, the network node comprising: a transceiver configured to transmit to the wireless device an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for demodulation reference signal (DMRS) transmission, the IFDMA subcarrier configuration being indicated by a first field indicating a cyclic shift and a second field separate from the first field indicating whether the IFDMA subcarrier configuration for DMRS transmission is one of a no-IFDMA-based DMRS configuration with a repetition factor (RPF) of 1 and an IFDMA-based DMRS configuration with an RPF of 2.
5. The network node of claim 4, wherein the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission.
6. The network node of claim 4, wherein the indication of the IFDMA subcarrier configuration is contained in downlink control information, DCI.
7. A method in a wireless device for configuring demodulation reference signals, (DMRS), the method comprising: receiving, from a network node, an indication of an interleaved frequency division multiple access (IFDMA) subcarrier configuration for DMRS transmission; and configuring DMRS transmissions according to the indication, the IFDMA subcarrier configuration being indicated by a first field indicating a cyclic shift and a second field separate from the first field indicating whether the IFDMA subcarrier configuration for DMRS transmission is one of a no-IFDMA-based DMRS configuration with a repetition factor (RPF) of 1 and an IFDMA-based DMRS configuration with an RPF of 2.
8. The method of claim 7, wherein the indication of the IFDMA subcarrier configuration specifies



which subcarriers are to be used for DMRS transmission.

9. A wireless device for multiplexing demodulation reference signals, (DMRS), the wireless device comprising: a transceiver configured to receive, from a network node, an indication of an interleaved frequency division multiple access, (IFDMA) subcarrier configuration for DMRS transmission; and processing circuitry configured to configure DMRS transmissions according to the indication, the IFDMA subcarrier configuration being indicated by a first field indicating a cyclic shift and a second field separate from the first field indicating whether the IFDMA subcarrier configuration for DMRS transmission is one of a no-IFDMA-based DMRS configuration with a repetition factor (RPF) of 1 and an IFDMA-based DMRS configuration with an RPF of 2.

10. The wireless device of claim 9, wherein the indication of the IFDMA subcarrier configuration specifies which subcarriers are to be used for DMRS transmission.

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