

(12) **United States Patent**
Berger et al.

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(54) **RESONANT ELEMENTS AND OSCILLATORS**

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(60) Continuation of application No. 17/384,535, filed on Jul. 23, 2021, now Pat. No. 11,897,757, which is a (Continued)

(51) **Int. Cl.**
B81B 3/00 (2006.01)
H03H 9/02 (2006.01)
H03H 9/24 (2006.01)

(52) **U.S. Cl.**
CPC **B81B 3/0024** (2013.01); **H03H 9/02448** (2013.01); **H03H 9/2457** (2013.01); **H03H 9/2405** (2013.01)

(58) **Field of Classification Search**

CPC B81B 7/0083-0096; B81B 7/009-0096; B81B 3/0024; H03H 9/2405; (Continued)

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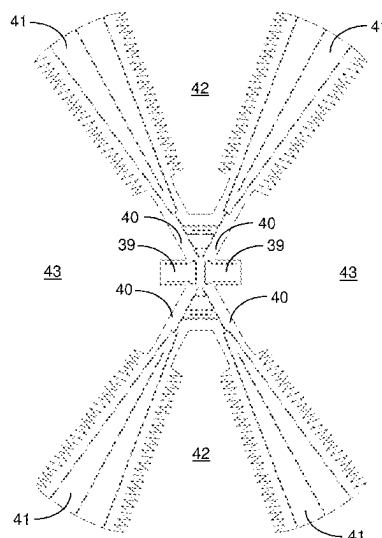
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(57) **ABSTRACT**

The present inventions, in one aspect, are directed to micro-machined resonator comprising: a first resonant structure extending along a first axis, wherein the first axis is different from a crystal axis of silicon, a second resonant structure extending along a second axis, wherein the second axis is different from the first axis and the crystal axis of silicon and wherein the first resonant structure is coupled to the second resonant structure, and wherein the first and second resonant structures are comprised of silicon (for example, substantially monocrystalline) and include an impurity dopant (for example, phosphorus) having a concentrations which is greater than 10^{19} cm^{-3} , and preferably between 10^{19} cm^{-3} and 10^{21} cm^{-3} .

22 Claims, 19 Drawing Sheets



Related U.S. Application Data

division of application No. 15/627,049, filed on Jun. 19, 2017, now Pat. No. 11,807,518, which is a division of application No. 13/759,013, filed on Feb. 4, 2013, now Pat. No. 9,695,036.

- (60) Provisional application No. 61/617,230, filed on Mar. 29, 2012, provisional application No. 61/594,357, filed on Feb. 2, 2012.

(58) **Field of Classification Search**

CPC H03H 9/02448; H03B 5/30-42; H01L 2224/32145; H01L 2924/1461

See application file for complete search history.

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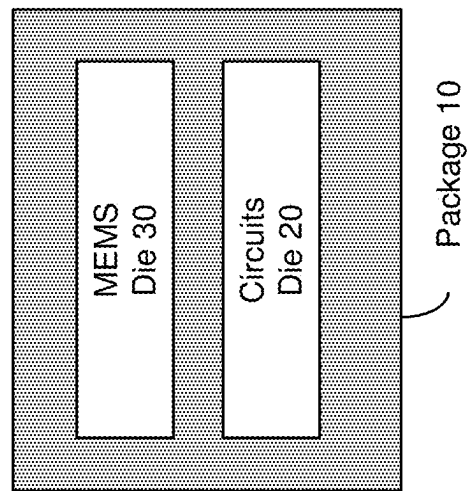


FIGURE 1A

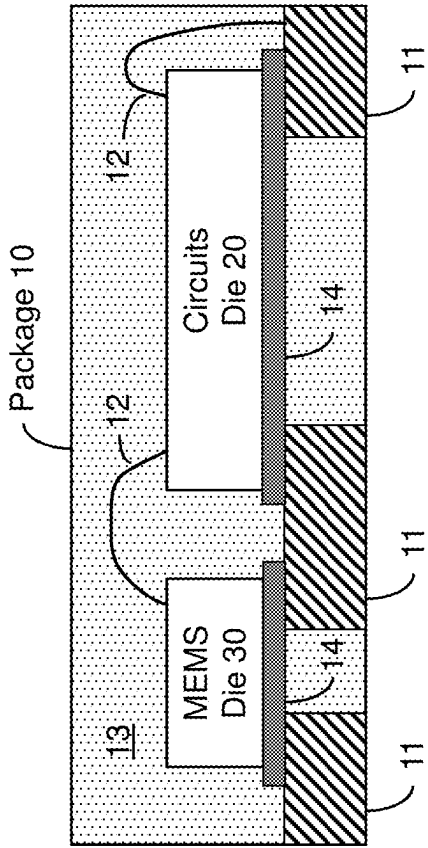


FIGURE 1B

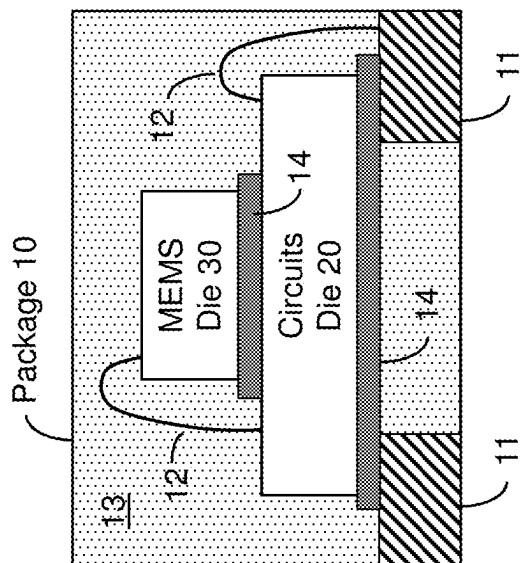


FIGURE 1C

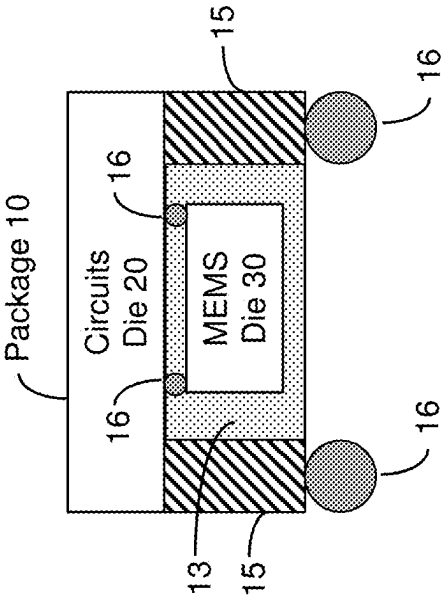


FIGURE 1D

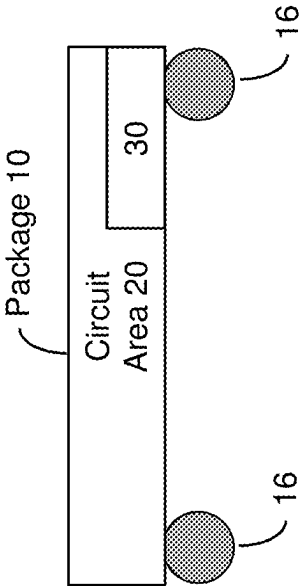


FIGURE 1E

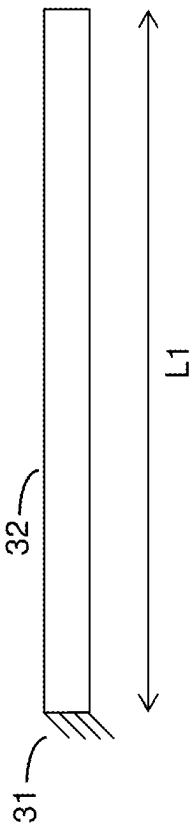


FIGURE 2A

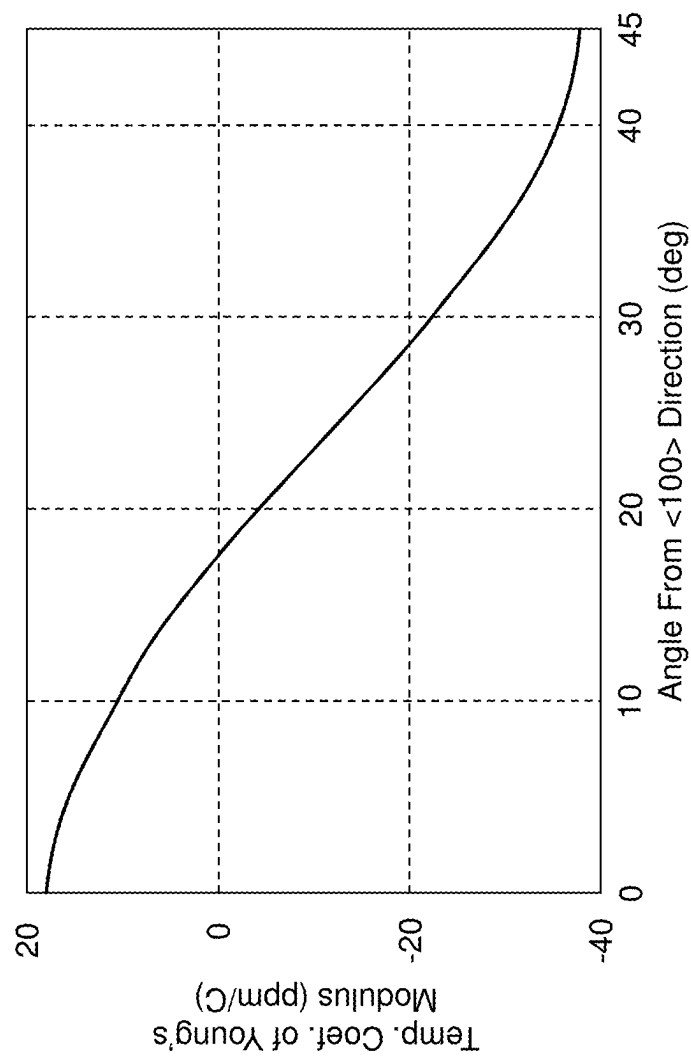


FIGURE 2B

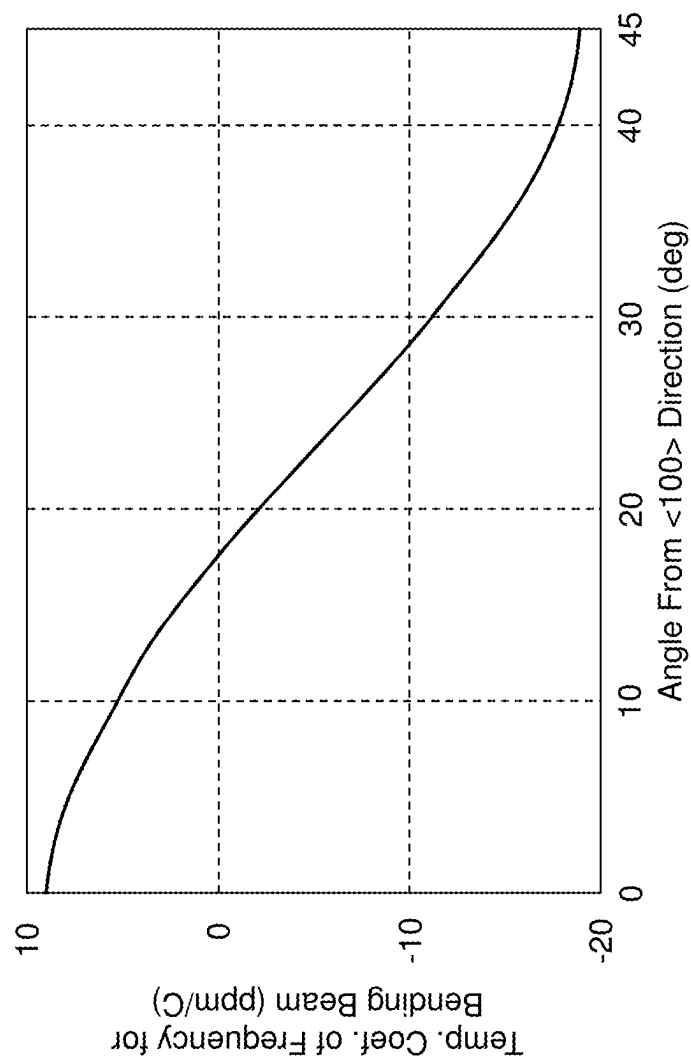


FIGURE 2C

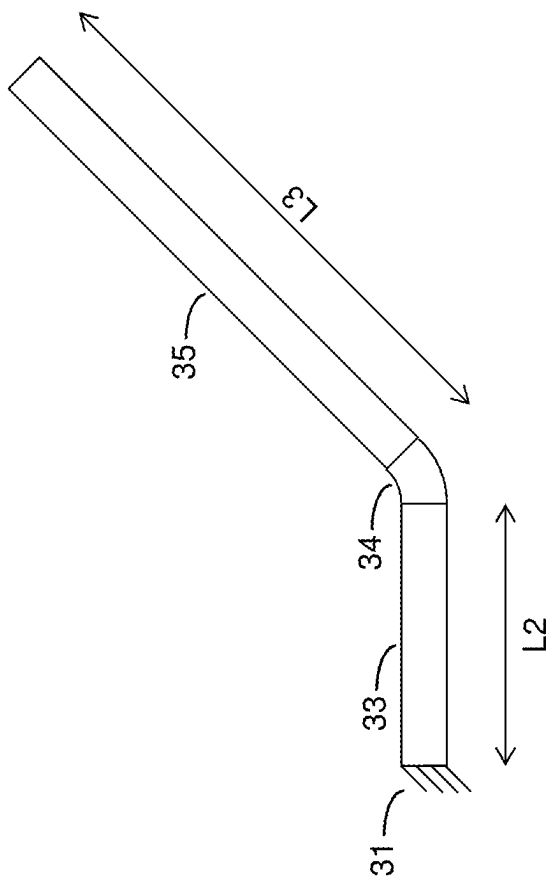


FIGURE 2D

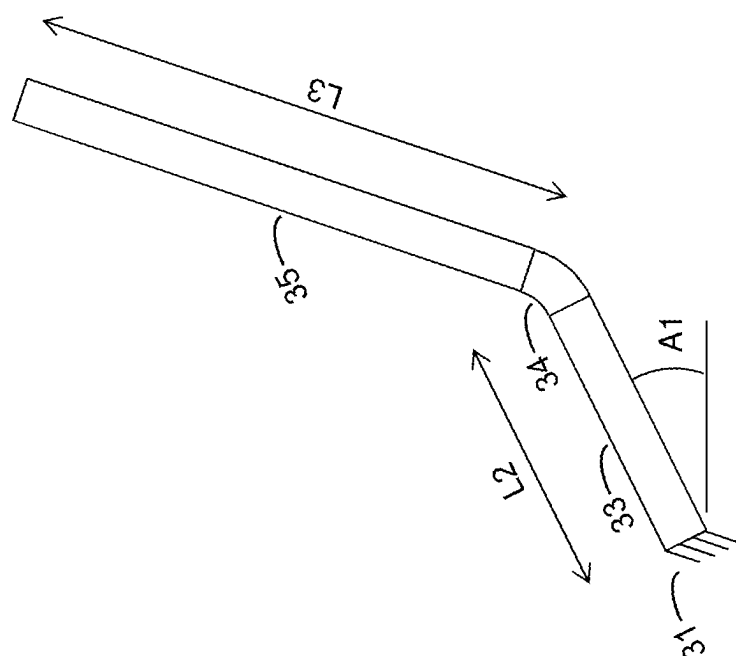


FIGURE 2E

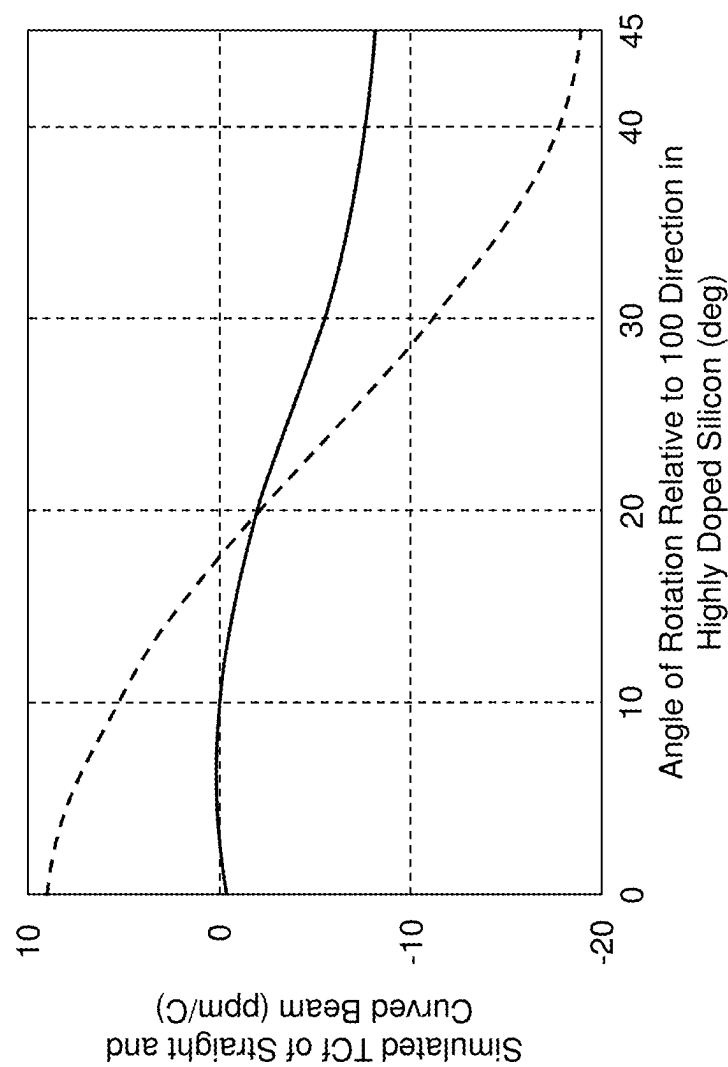


FIGURE 2F

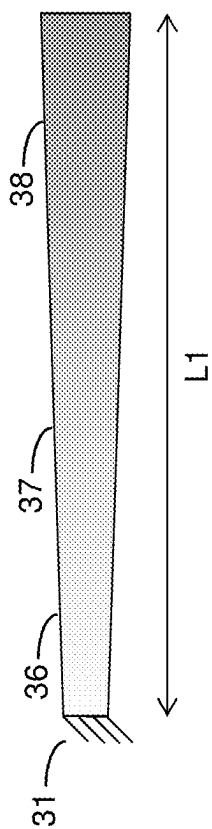


FIGURE 3

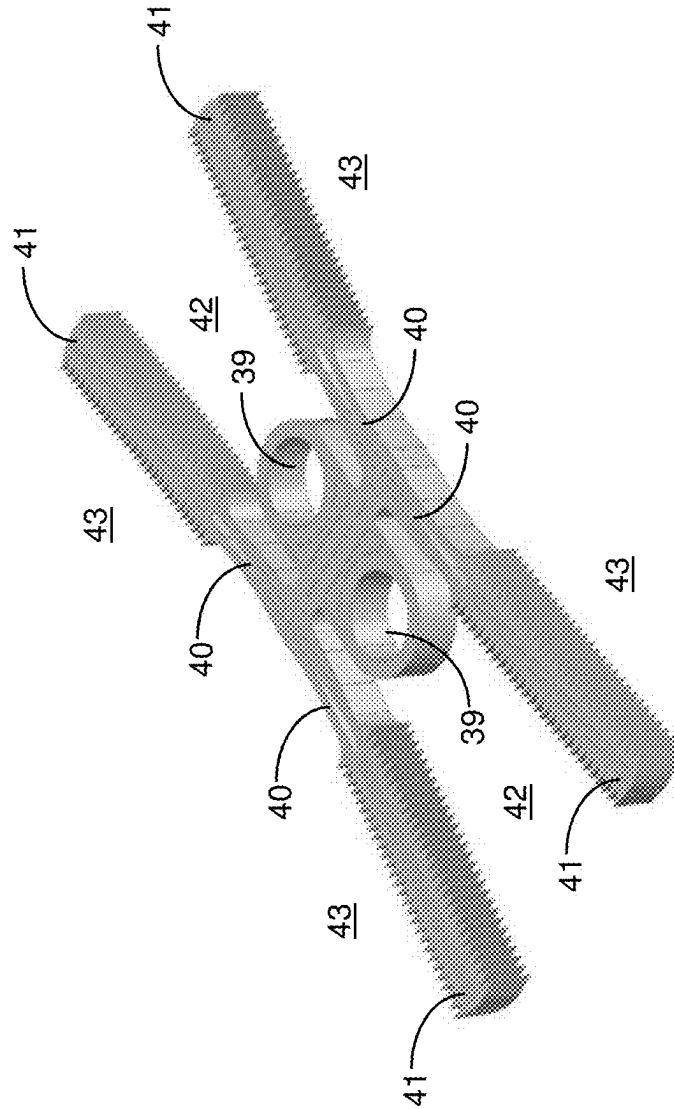


FIGURE 4A

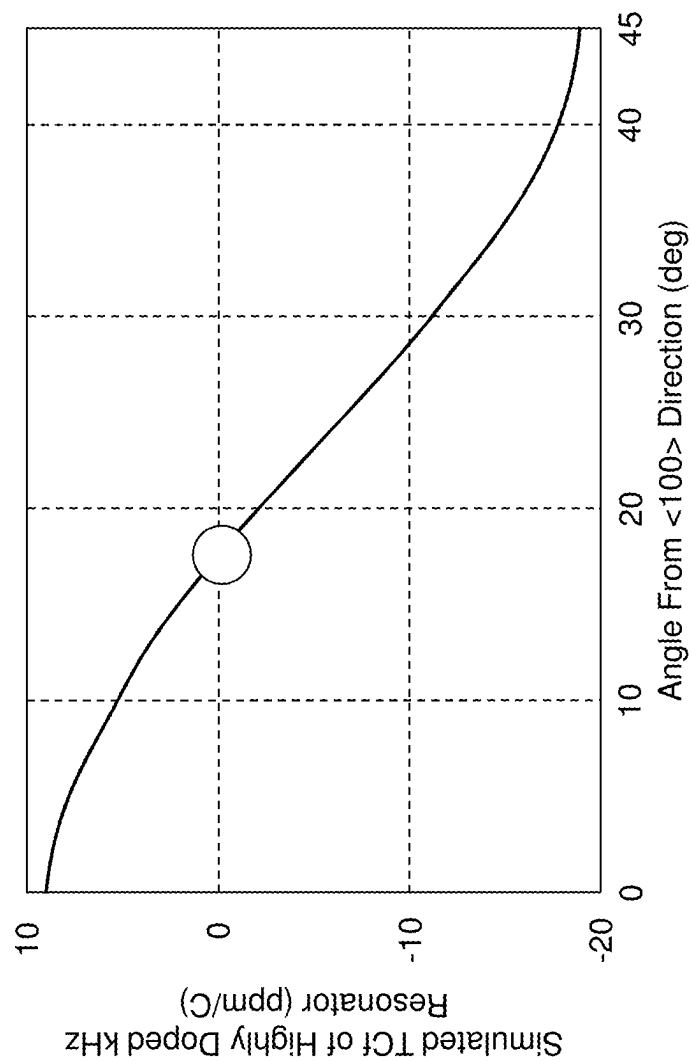


FIGURE 4B

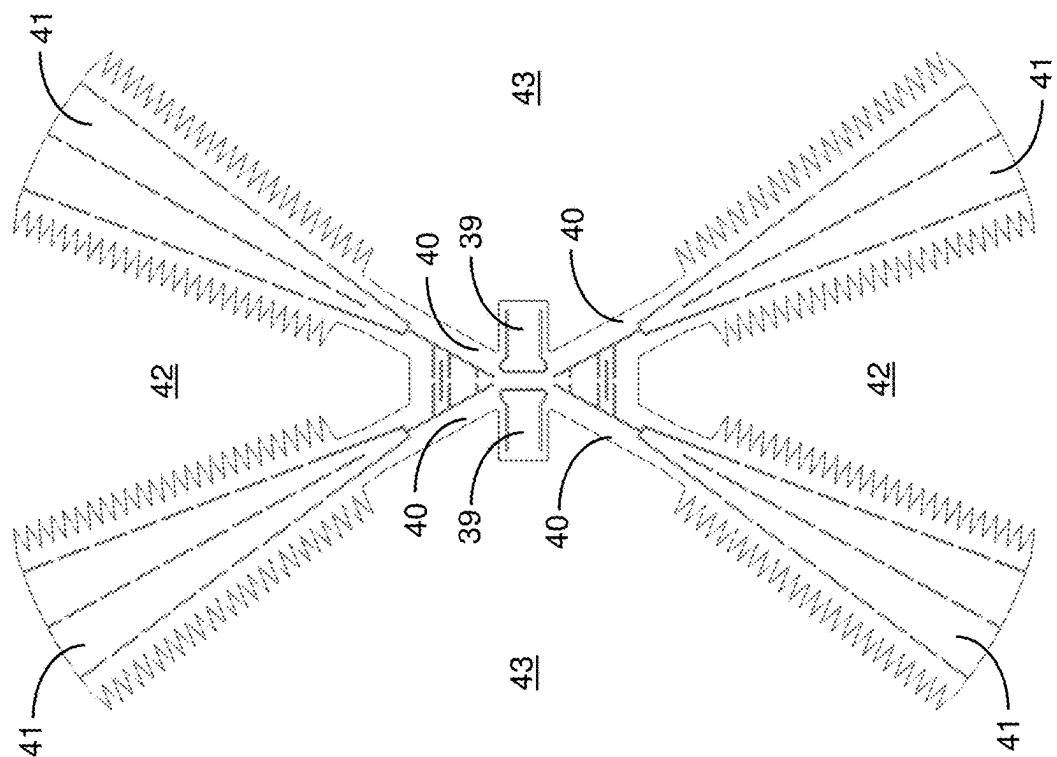


FIGURE 5

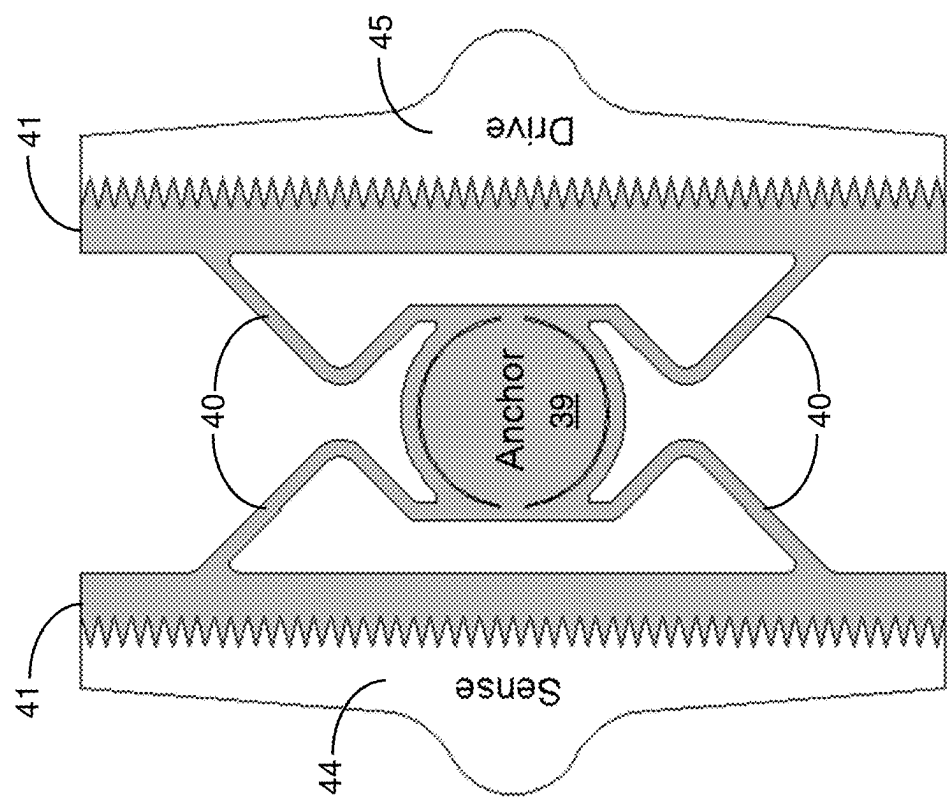


FIGURE 6

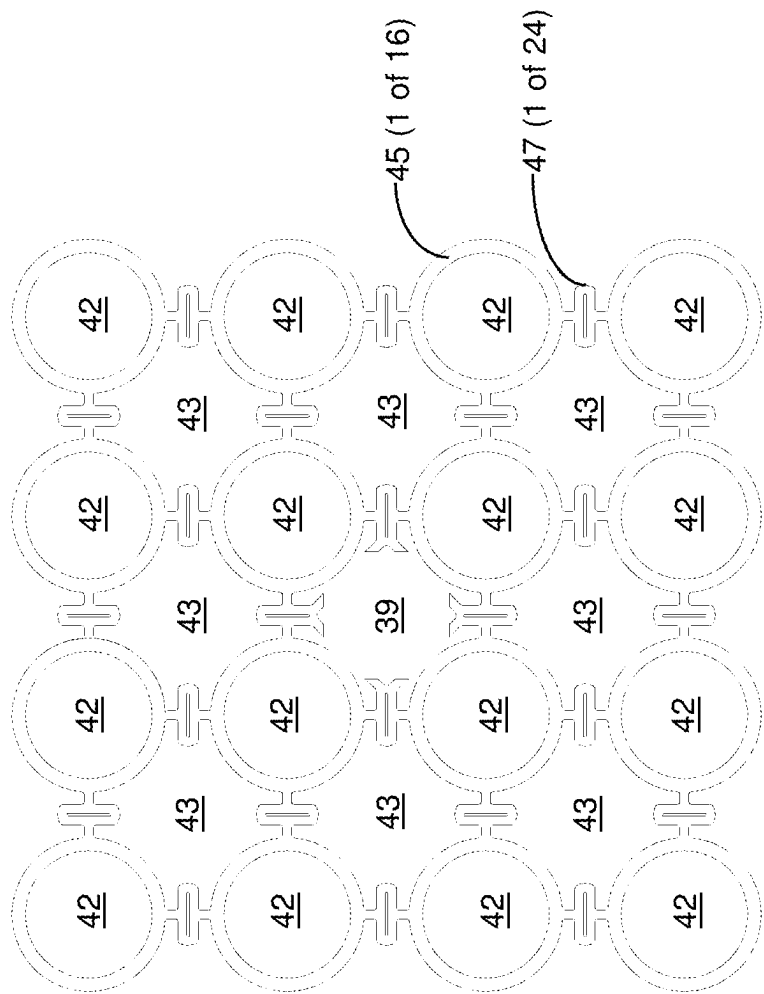


FIGURE 7A

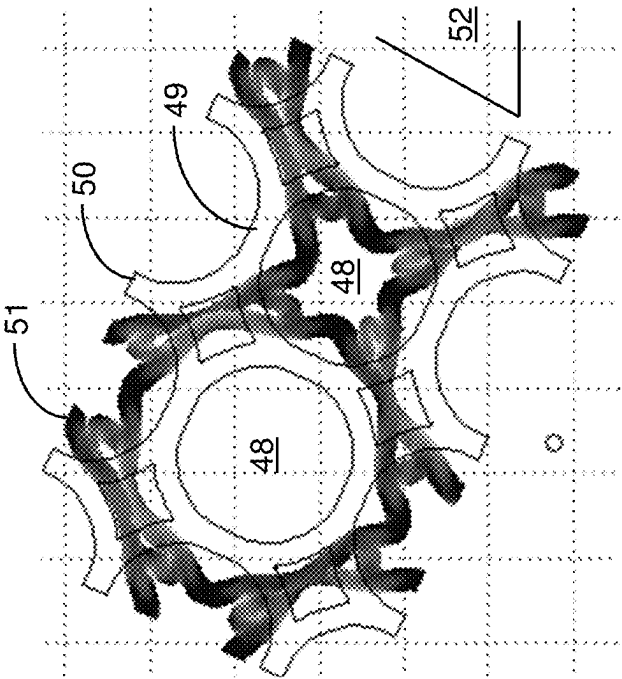


FIGURE 7B

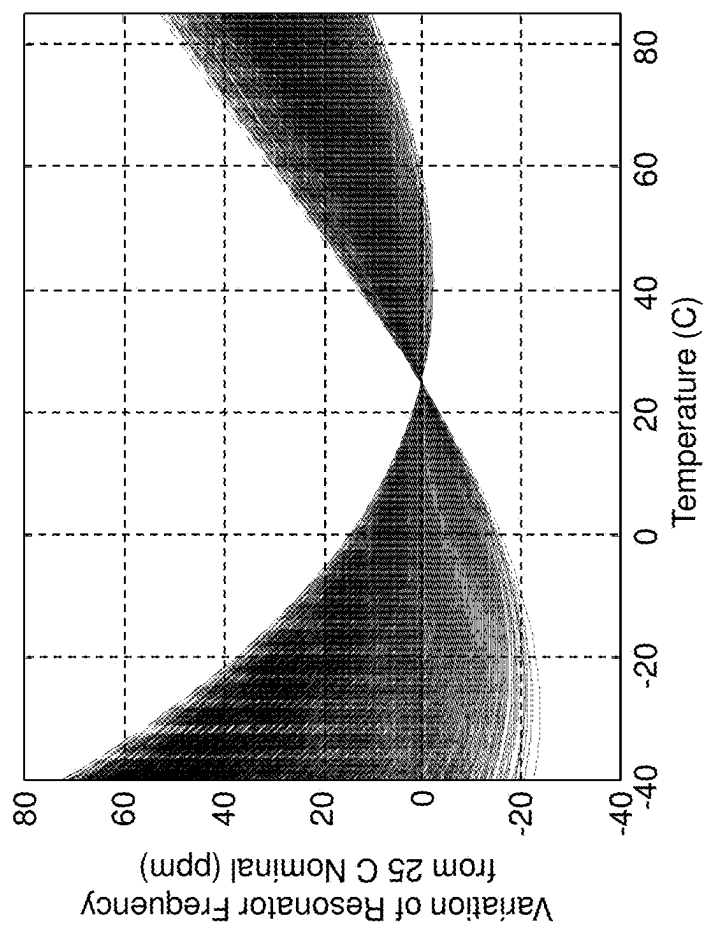


FIGURE 8

RESONANT ELEMENTS AND OSCILLATORS

CROSS REFERENCE TO RELATED APPLICATIONS

This application is a continuation of U.S. application Ser. No. 17/384,535, filed Jul. 23, 2021, which is a divisional of U.S. application Ser. No. 15/627,049 filed Jun. 19, 2017 (now U.S. patent Ser. No. 11/807,518), which is a divisional of U.S. application Ser. No. 13/759,013 filed Feb. 4, 2013 (now U.S. Pat. No. 9,695,036), which claims priority to (i) U.S. Provisional Application No. 61/594,357, filed Feb. 2, 2012, and (ii) U.S. Provisional Application No. 61/617,230, filed Mar. 29, 2012. Each of the above-identified patent applications is hereby incorporated by reference herein.

INTRODUCTION

In one aspect, the present inventions relate to methods of manufacturing microelectromechanical and/or nanoelectromechanical (collectively hereinafter “MEMS”) structures and devices/systems including same; and more particularly, in various aspects, to methods of manufacturing MEMS structures comprising one or more moveable, resonating or resonant members (hereinafter “resonant elements”) that exhibit desirable resonant frequency stability or response across temperature (for example, a predetermined temperature range such as a desired or predetermined operating range of the resonating or resonant members). In further aspects to designing oscillators that include MEMS resonant elements and electronics to produce output signals with stable frequency across temperature.

Resonant elements exhibit variation in resonant frequency from, among other things, their temperature. This may be described as a frequency function of temperature, a change in frequency with temperature, temperature sensitivity, temperature drift, or by other names (collectively hereinafter “temperature sensitivity”). Controlling, setting, adjusting, or minimizing temperature sensitivity is important to many applications. In this regard, in most implementations, the resonance frequency of the MEMS resonant element should remain substantially stable over a range of operating temperatures. This, however, will not typically be the case as thermally induced changes to the Young’s modulus (or other variables) tend to change the mechanical stiffness of the structure. That is, thermally-induced changes to the Young’s modulus tend to cause considerable variation or change in the frequency of the output of the resonant element.

The temperature sensitivity of the resonant element may be compensated via circuitry with measured temperature to produce a more stable frequency over temperature. (See, for example, U.S. Pat. No. 6,995,622).

Temperature sensitivity may be described with measured or expected resonant frequencies at measured or specified temperatures (hereinafter “frequency/temperature pairs”), where there may be few or many frequency/temperature pairs. Additionally, temperature sensitivity may be described by mathematical abstractions that define the frequency as a function of temperature, or the temperature as a function of frequency, or other relationships. One common abstraction is to use frequency/temperature pairs and interpolate and/or extrapolate them, where the number of pairs is less than the number of temperature at which the frequency may be determined. The interpolation or extrapolation may be of a simple linear nature or more complex, for instance a cubic spline. Another common abstraction is a polynomial fit. In

this approach an offset, line, or curve represents the frequency with a polynomial of the temperature, or the temperature as a polynomial of frequency, or other polynomial relationships. It is important to note that frequency/temperature pairs need not be measured or identified. Further, the abstraction need not describe the physics behind the temperature sensitivity, for instance the cause of change in Young’s modulus, linear expansion, stress, etc. The abstraction may describe the temperature sensitivity with accuracy suitable for the application. When describing the temperature sensitivity in this way, e.g. as a polynomial, one assumes that the abstraction is a sufficiently accurate description or model of the system.

Offset, linear, parabolic, and cubic functions are examples of polynomials, the order may be increased as needed, usually providing better agreement between the abstraction and the actual resonant element temperature sensitivity. It is generally understood that the higher order polynomials may also contain lower order terms, so for instance a cubic polynomial may include parabolic, linear, and offset terms. The order is often described with ordinals, for instance third order for cubic, second order for parabolic, first order for linear, and zero order for offset. For polynomials beyond cubic it is common to refer to the ordinal, for instance fourth or fifth order.

The temperature sensitivity may include a polarity; positive (increasing in frequency with increasing temperature), negative (decreasing in frequency with increasing temperature), or zero (unchanging in frequency with temperature). A numerical value may be assigned to the temperature sensitivity, for instance -31 ppm/C. When a value is assigned it may be used to describe a single value across temperature or a value at a specific temperature. When a value is described as applying over a range of temperatures that implies the temperature sensitivity is not a function of temperature, for instance that it is linear. When the temperature sensitivity is a function of temperature, for instance in a second or higher order polynomial, then a specific sensitivity may be applied at one or more temperatures or a sensitivity range may be given over a temperature range.

In one embodiment a resonant element’s frequency across a given temperature range may be made to be within a desired delta from a given frequency. In another embodiment, a resonant element’s frequency measured across a temperature range may be made to be within a delta from that resonant element’s frequency at a given temperature. In yet another embodiment, a resonant element’s frequency across a given temperature range may be made such that the frequency sensitivity has a slope within a given range. In yet another embodiment, the resonant element’s temperature sensitivity at some temperature, within a given temperature range, may be made to be zero.

This is called a turnover temperature, and describes a behavior where a resonant element has a temperature sensitivity that changes from one sign to the opposite sign, crossing through zero slope at a temperature. In one embodiment a resonant element has a parabolic temperature sensitivity where the maximum or minimum of the parabolic function is at or near a specific desired turnover temperature. In another embodiment the resonant element has a cubic temperature sensitivity where the function may have two turnover points, one at higher temperature than the other. Higher polynomial orders may lead to further turnover temperatures. It should be recognized that an exact turnover temperature is may not be possible, but rather a turnover temperature within a certain temperature range may suffice. Further, the frequency at the turnover temperature may be a

specific value or may be within a range of frequencies. A key characteristic of a turnover temperature is that the change in frequency as a function of temperature may be zero or nearly zero.

A resonant element with a turnover temperature has the advantage that it may be maintained near that temperature and in so doing provide a highly stable reference frequency. The resonant element may be designed such that the turnover temperature is at a higher value than the highest ambient operating temperature. The resonant element may then be heated to near the turnover temperature and maintained at that point as the ambient temperature varies. The result is that the resonant element may have a more stable frequency than a system in which the resonant element's temperature is allowed to vary with the ambient temperature.

It is an embodiment of these inventions that a MEMS resonant element may have a frequency sensitivity that gives a predetermined and stable frequency across a range of temperatures. It is a further embodiment that a MEMS resonant element may have a frequency sensitivity that gives a frequency across a range of temperatures, where the absolute frequency is not tightly determined but the variation is minimized. It is a further embodiment that a MEMS resonant element may have a temperature sensitivity that may have minimal slope across a range of temperatures. It is a further embodiment that a MEMS resonant element may have one or more turnover temperatures within one or more temperature ranges.

A resonant element may also have various frequency functions of temperature depending upon its construction or upon which resonant mode is excited. Two resonant elements may be thermally coupled and operated simultaneously so that their temperatures are nearly equal. In this way their relative frequencies may be used to determine the temperature. In such an approach one resonant element may be made to have minimum temperature sensitivity, a stable frequency, while the other may be made to have maximum temperature sensitivity. The relative frequencies then provide both a stable reference frequency and a measurement of the temperature, or relative temperature, which may be used to compensate the stable frequency to provide a more stable frequency. This may also be done with the more stable resonant element having a turnover temperature and where the relative frequency is used to measure and maintain the temperature at or near the turnover temperature.

The two resonant elements need not be distinct but may be one structure that is resonated (or resonates) in multiple modes (for example, two modes). In this way the temperature of the two resonant elements may be more closely matched because the elements may be physically co-located.

It should be appreciated that the temperature also may be measured by non-resonant means, such as with an electronic sensor. Examples of electronic sensors include thermistors and bipolar transistor circuits. There are a vast number of ways to measure temperature and this discussion is meant to include all arrangements, and is not limited to those described here.

The measured temperature may be used to correct or adjust the frequency of the resonant element. This may be done by pulling the resonant frequency, for instance by varying a bias voltage or load capacitance(s), or this may be done by electronically synthesizing a frequency based on the reference frequency. The adjusted or synthesized frequency often may be more stable over temperature than the reference frequency, but this is not required, in some cases it may be desirable for the adjusted or synthesized frequency to have a specific or predetermined function of temperature.

One example of this is where the frequency is a linear function of temperature and thus may be used to transmit that temperature value to external devices.

The circuit that adjusts or synthesizes the frequency may be on an integrated circuit placed far from or near the resonant element(s). When mounted near the resonant element(s) it may be mounted in thermal contact with the resonant element(s) in order that the temperature of the integrated circuits may be coupled to the resonant element(s) so that the temperature measured at the circuit reflects the temperature of the resonant element(s).

Additional advantages may be found where the integrated circuit is in close proximity to the resonant element(s), for instance decreased stray capacitances and thus improved performance, decreased noise, decreased power consumption, and decreased susceptibility to external interference. Thus even when two or more resonant elements are used to both provide a reference and measure temperature, it may be advantageous to have the integrated circuit that drives the resonant elements mounted in close proximity to those elements.

In one embodiment a circuit is mounted proximal to the resonant element(s) to drive the resonant element(s) into motion. The circuit may include functions to measure temperature, and may have functions to adjust the frequency, possibly with a fractional phase locked loop (frac-N PLL), or possibly with a programmable divider. The circuit may have other functions, including non-volatile memory, digital control, test elements, and I/O blocks.

In another embodiment the circuit may be mounted in thermal contact with the resonant element(s), where the circuit die is mounted in contact with the MEMS die. The mounting may include electrical and/or thermal contact, for instance coupled with electrical and/or thermally conductive epoxy or metallic interconnects.

In an embodiment, the MEMS resonant element and circuit may be built into a package. The package may be an epoxy or chip scale package similar to those used for integrated circuits.

The resonant element(s) may be built into a die or lidded such that they are encapsulated. In this way they may be more readily handled, e.g. tested, diced, mounted, and packaged.

In an embodiment, the resonant element(s) may be built on a wafer and lidded with a cover wafer to protect it/them during wafer dicing and packaging. The cover wafer may be attached with gold compression bonding, eutectic bonding, frit-glass bonding or other bonding techniques.

In another embodiment, the resonant element(s) may be built into an encapsulation structure or structures in or on a carrier wafer. The encapsulation structure or structures may be formed over or around the resonant element(s) in wafer fabrication, and may include silicon on insulator (SOI) structures. All wafer covering and encapsulation techniques, presently known or unknown, are included in the scope of these inventions.

In one embodiment, a MEMS resonant element may be built from a semiconductor, for instance silicon, germanium or carbon. The semiconductor may also be compound materials, for example, silicon-germanium, silicon-carbide, gallium-arsenide, or other.

Specifically, when a resonant element is fabricated in highly doped material, it may provide both positive and negative temperature coefficients of frequency, or turnover temperatures, depending on its structure and the dopant concentration.

In one embodiment, the resonant element may be fabricated from silicon having doping concentrations greater than 10^{15} cm^{-3} , and preferably, between 10^{15} cm^{-3} and 10^{21} cm^{-3} . In another embodiment, the resonant element may be fabricated from silicon having doping concentrations greater than 10^{19} cm^{-3} , and preferably, between 10^{19} cm^{-3} and 10^{21} cm^{-3} .

The doping of the material from which a resonant element is fabricated may be n-type or p-type or may be a combination of n-type and p-type. For instance in silicon the dopant(s) may include but are not limited to phosphorus, arsenic, antimony, boron, and aluminum. The dopants may be used individually or in combination. For instance in one embodiment, silicon may be dominantly doped with phosphorus, but there may be arsenic and/or boron present, or the dominant dopant may be boron but there may be phosphorus and/or arsenic present. Other elements, not normally considered dopants, may be present as well, for instance carbon and/or oxygen. Oxygen in particular may be included in concentration levels that contribute to the mechanical behavior of the structures.

In some embodiments the doping may not be uniform but rather may vary in concentration across the resonant elements. The type as well as the concentration may vary.

In one embodiment the resonant elements are built from silicon with a phosphorus dopant concentration between 10^{19} cm^{-3} and 10^{21} cm^{-3} . The dopant is not necessarily uniform, and other elements, such as oxygen, may be present in the material.

In an additional embodiment the material from which the resonant elements may be fabricated may be monocrystalline, polycrystalline, amorphous, or a combination of these. The polycrystalline material may be of random order or ordered, where ordered polycrystalline material may have a preferred or dominant axis of alignment. The material may vary as a function of location in a resonant element, for instance with one or more parts being monocrystalline and others polycrystalline.

The monocrystalline material may be aligned in a controlled way such that a resonant element built in or on it has a desired or predetermined orientation with respect to the multiple crystal axis, planes, normal vectors, crystallographic orientation, or principle axis (hereinafter the "orientation"). This may be the case with, for instance, silicon wafers or SOI wafers that may be procured in a preferred orientation with specified alignment accuracy, for instance a $\langle 100 \rangle$ wafer. Hereafter, the term "angle" refers to the orientation of a resonant element or a member of a resonant element with respect to the orientation of the single crystal material or the dominant orientation of the polycrystalline material from which it is made.

The material, crystalline type, doping type, doping concentration, resonant element shape, resonant mode shape, and orientation contribute to a resonant element's temperature sensitivity and to the variation of that temperature sensitivity over process variations.

These and other parameters may be selected to minimize the sensitivity of the temperature sensitivity to fabrication process variations. For example, in one embodiment, the material, crystalline type, doping type, doping concentration, resonant element shape, resonant mode shape, and/or orientation, may be selected to produce resonant members that are maximally insensitive to process variation. In this embodiment, although the parameters may not provide a minimum temperature sensitivity characteristic of a particular resonant element, the parameters may provide a mini-

mum temperature sensitivity of the locus of resonant elements over manufacturing variations or tolerances.

BRIEF DESCRIPTION OF THE DRAWINGS

In the course of the detailed description to follow, reference will be made to the attached drawings. These drawings show different aspects of the present inventions and, where appropriate, reference numerals illustrating like structures, components, materials and/or devices in different figures are labeled similarly. It is understood that various combinations of the structures, components, and/or devices, other than those specifically shown, are contemplated and are within the scope of the present inventions.

Moreover, there are many inventions described and illustrated herein. The present inventions are neither limited to any single aspect nor embodiment thereof, nor to any combinations and/or permutations of such aspects and/or embodiments. Moreover, each of the aspects of the present inventions, and/or embodiments thereof, may be employed alone or in combination with one or more of the other aspects of the present inventions and/or embodiments thereof. For the sake of brevity, certain permutations and combinations are not discussed and/or illustrated separately herein.

FIG. 1A-1E illustrates packaging options for the resonant element die and integrated circuit die according to certain aspects and embodiments of the present inventions, including lateral die positioning, die stacking, flip chip construction, and integrated construction;

FIGS. 2A-2F illustrates simplified schematic of idealized resonant structure elements and plots of temperature sensitivity of those elements, where the structures are idealized to make the relationships clear;

FIG. 3 illustrates a simplified schematic of an idealized resonant element with non-uniform doping;

FIGS. 4A and 4B illustrates an embodiment of a kHz resonant element that may exhibit the desired temperature insensitivity as a function of its rotation angle, and a plot of approximately that sensitivity as a function of angle;

FIG. 5 illustrates another embodiment of a kHz resonant element that may show minimal variation in frequency over both temperature and process variations;

FIG. 6 illustrates another embodiment of a kHz resonant element that may show minimal variation in frequency over both temperature and process variations;

FIGS. 7A and 7B illustrates an embodiment of an arrayed MHz resonant element that may show minimal variation in frequency variation over temperature and a detail of two array cells showing both the design shape and the mode shape; and

FIG. 8 illustrates an example of frequency variation over temperature of an embodiment kHz resonant element across various fabrication tolerances.

Again, there are many inventions described and illustrated herein. The present inventions are neither limited to any single aspect nor embodiment thereof, nor to any combinations and/or permutations of such aspects and/or embodiments. Each of the aspects of the present inventions, and/or embodiments thereof, may be employed alone or in combination with one or more of the other aspects of the present inventions and/or embodiments thereof. For the sake of brevity, many of those combinations and permutations are not discussed separately herein.

DETAILED DESCRIPTION

At the outset, it should be noted that there are many inventions described and illustrated herein. The present

inventions are neither limited to any single aspect nor embodiment thereof, nor to any combinations and/or permutations of such aspects and/or embodiments. Moreover, each of the aspects of the present inventions, and/or embodiments thereof, may be employed alone or in combination with one or more of the other aspects of the present inventions and/or embodiments thereof. For the sake of brevity, many of those permutations and combinations will not be discussed separately herein.

Further, in the course of describing and illustrating the present inventions, various architectures, structures, components, functions and/or elements, as well as combinations and/or permutations thereof, are set forth. It should be understood that architectures, structures, components, functions and/or elements other than those specifically described and illustrated, are contemplated and are within the scope of the present inventions, as well as combinations and/or permutations thereof.

With that in mind, in one aspect, the present inventions are directed to a method of manufacturing a microelectromechanical structure having one or more moveable or resonating elements. The method of manufacture fabricates moveable or resonating elements from a substrate (for example, monocrystalline silicon) having a dopant concentration, using, for example, conventional lithographic and etching techniques. A desired dopant type, concentration, and pattern may be developed before or during the wafer fabrication.

MEMS resonator elements, or resonators, may be built into systems, and as such may be parts of larger systems. Resonator elements may be used with drive circuits, particularly CMOS circuits, and may be packaged. With reference to FIG. 1A, a package 10, a circuit die 20 and a MEMS die 30 are shown schematically. The MEMS resonant element, or resonator, is part of the MEMS die. It should be understood that the package 10 and circuits 20 contribute to performance of the system, and in particular to the temperature stability of the resonant element. Throughout this description the resonant frequency of the resonant element, or resonator, is described as belonging to the MEMS element, but it should be understood that the circuit and the package may have influence on it and may need to be considered.

The MEMS die and the circuit die will often be physically packaged together. With reference to FIGS. 1B, 1C, and 1D, one can package them side-by-side, stacked, for instance with the MEMS on the CMOS, with bondwire electrical interconnects, or stacked, for instance with the CMOS on the MEMS, with ball or bump interconnects.

With reference to FIG. 1B, the contacts 11 can be metal pads that are exposed to the outside of the package, the bondwires 12 can connect the pads to the circuitry and the circuitry to the MEMS die, the plastic 13 encloses the die and parts of the contacts, but leaves at least some of the external faces of the contacts 11 exposed, and the epoxy 14 attaches the die to the contacts.

There are many possible variants to the construction shown in FIG. 1B and it should be understood that this description is not limiting. There may be elements not shown that can be used in a package, as well there may be elements that may be omitted from a package.

With reference to FIG. 1C, the MEMS die may be stacked on the circuit die. This may have the advantage that it has a smaller footprint, and may have the disadvantage that it has an overall greater thickness. In some applications and system designs it has the advantage that the MEMS die 30 and the circuit die 20 are in close or intimate electrical

and/or thermal contact. In this way the temperature of the MEMS die 30 may be measured at the circuit die 20 with circuit elements. The epoxy 14, encapsulation plastic 13, bondwires 12, and contacts 11 continue to function in the stacked configuration as they do in the side-by-side configuration except that the epoxy between the die may have different properties than that below the die. For example the inter-die epoxy may be electrically conductive or thermally conductive while the under-die epoxy may be electrically insulating.

In some applications it may be desirable to contact the external environment through balls, or bumps, rather than contact pads. It may also be advantageous to build the package with the minimum footprint. In these cases it may be advantageous to use a chip-scale package. With reference to FIG. 1D, the MEMS and circuits die can be stacked without bondwires where the electrical contacts are made through bumps or balls 16. This type of package is known in circuit applications but can be extended for these resonant element system packages. These packages may use small balls or bumps 16 between the MEMS and circuit die and larger balls or bumps 16 between the external contact and the circuit die. It may be useful, although not always necessary, to build pillars 15 between the external balls or bumps and the circuit die. These pillars can serve to extend the height of the balls to supply room to mount the MEMS die 30 below the circuit die 20. Plastic 13 may be used to enclose the MEMS die but in some applications it may be omitted.

It should be understood that this is an example of flip chip packaging and should not be understood to be limiting. Other components, for instance underfill material between the MEMS die 30 and the circuit die 20 may be used even though it is not shown in this figure.

A fully integrated system may also be built. With respect to FIG. 1E, the MEMS functions and the circuit functions may be integrated onto one die with functional areas 30 and 20. That die may be packaged as a chip with balls or bumps 16 to form a chip-scale package 10. It should be clear that an integrated system is described only schematically here and that there are important details not described. This description must be understood to be non-limiting. Further, an integrated system may be packaged similarly to that shown in FIG. 1B with bondwires but only a single die.

The system may also be packaged in a lead frame (not shown). It should be understood that the full diversity of packages used for circuits can potentially be used for the MEMS and circuit combination. With reference to FIG. 1A, the package 10 can take many forms, some common and some uncommon in the electronics industry. All such forms are contemplated in this work. Indeed, new package types developed in the future are contemplated in this work and all packages now known or yet to be developed are explicitly included in this description.

The resonant element can be of simple or complex shapes, but looking first at a simple structure can show important characteristics that can carry to the more complex structures. With reference to FIG. 2A, a resonant element 32 is attached at an anchor 31 and has length L1. This resonant element may move in-plane or out of plane, and that distinction is not made in this figure. The first bending mode is a swaying of the end. There are other modes as well, including higher order bending modes and extensional modes.

With reference to FIG. 2B, the resonant beams may have positive or negative temperature coefficients of Young's modulus. FIG. 2B shows an example coefficient of a highly doped crystalline semiconductor as a function of the angle of the resonant beam with respect to the crystal orientation of

that semiconductor. Note that there are many possible curves or relationships depending on the semiconductor, dopant type, and the dopant density. FIG. 2B represents one that may be found for highly phosphorus doped silicon.

It should be understood that the example of phosphorus in silicon in FIG. 2B should not be taken as limiting. As previously discussed, there are many options for materials and dopants, including but not limited to silicon, silicon-germanium, boron, aluminum, phosphorus, arsenic, and oxygen.

The Young's modulus, shown in FIG. 2B contributes to the resonant frequency function of temperature, the temperature sensitivity, but is not the only contributor. The expansion of the material over temperature contributes, as well as potential stress from the package and electrical interactions with the drive circuit, for instance drive phase and load capacitance.

With reference to FIG. 2C, the variation in frequency, or the temperature sensitivity, can be plotted as a function of angle from the crystal lattice. As described above, this plot is exemplary of phosphorus doped silicon, but many other dopants and base materials are possible.

An important characteristic of the curve shown in FIG. 2C is that the temperature sensitivity can be positive or negative and that it can cross through zero. This suggests that a resonant element may be configured for zero temperature sensitivity by building it at a specific angle to the crystal lattice of the semiconductor. It should be remembered that this is just one of many potential curves, and that it is a function of doping concentration, dopant type, semiconductor type, and other variables.

More complex resonant elements may be constructed from simple elements. With reference to FIG. 2D one can construct a beam that includes two (or more) sections at different angles. These beams, 33 and 35 can be connected with section 34 and anchored with 31. In this example they have length L2 and L3. It should be understood that this example shows only two beams while systems with three or more beams are clearly possible.

TABLE 1 (immediately below) tabulates the effective temperature coefficient of frequency, or temperature sensitivity, TCf, as a function of the ratio of the length of the two beams L3 to L2, where they are at an angle of nominally 45 degrees. When they are the same length (L3/L2=1) they show a net temperature sensitivity of +6 ppm/C. When they have a length ratio of 4, that is (L3/L2=4), the two beams system shows a temperature sensitivity of 0 ppm/C.

TABLE 1

Ratio of L3/L2	TCf of 1 st Mode
1	+6 ppm/C.
2	+4 ppm/C.
4	0 ppm/C.
6.5	-4 ppm/C.

It should be well understood that this is a typical case for the silicon and phosphorus materials where the phosphorus concentration is about 10^{20} cm^{-3} . For other semiconductors, other dopants, or other dopant densities there will be other values for similar tables. One important point that is shown in this table is that the base resonant element beam seen as 33 in FIG. 2D may be aligned with the crystal axis and the net temperature sensitivity may be zero.

Another important point is that the net temperature sensitivity can be interpreted as the weighted sum of the contributing elements. The weighting is partially based on

the stress levels seen in each element for the particular resonant mode. The results tabulated in TABLE 1 therefore are for a particular resonant mode, in this case the first bending mode. Other modes may show other temperature sensitivities.

A further important point is that while FIG. 2A was not described as being in-plane or out-of-plane, one can see that as the structures become more intricately shaped they can be built more readily in-plane. FIG. 2D can be built in plane by lithographically defining and etching the shape. The curved section 34 can have a large or small radius, and indeed the choice will impact the results tabulated in TABLE 1 because the stresses through the radius can have various directions with respect to their crystallographic angle.

The temperature sensitivity can be further reduced by rotating the structure as shown in FIG. 2E. In this case both the beams 33 and 35 are rotated from the crystal axis, with 33 rotated by an angle A1. Other elements, radius or curved section 34 and lengths of the beams L2 and L3, are maintained.

FIG. 2F plots the temperature sensitivity, or TCf, of the rotated assembly. The solid line shows data for the rotated assembly, the dashed line shows data for the simple beam plotted in FIG. 2C for comparison. The rotated assembly of FIG. 2F can be seen to exhibit at least two important characteristics. First the amplitude of the temperature sensitivity over angle is reduced, and second at the points where the temperature sensitivities are zero their slopes are reduced. These observations mean that the rotated assembly has decreased sensitivity to fabrication errors that may result in angle variations. Indeed, FIG. 2F shows an angle near 8 degrees in which the sensitivity to angle errors approaches zero.

It should be born in mind once again that the data plotted in FIG. 2F is exemplary of many potential data sets. This data is again for a highly doped phosphorus and silicon system but other systems may show these behaviors.

Extending the exposition, one need not assume that a resonant structure has a uniform doping. FIG. 3 shows a tapered beam with a dopant density that decreases along its length from locations 36, to 37, to 38, where again it is anchored at 31 and has length L1. The temperature sensitivity is again a weighed sum but in this case it is both a weight of the mode shape and the dopant density at each location. As such it is a function of the resonant mode.

The dopant density may also have a gradient across the width of the beam. In particular it may have higher density at the outer parts of the beam and lower in the center parts. This can be the case with diffusive in-doping from the trench sidewalls.

An analysis of temperature sensitivity can be extended to include variation of doping density and of rotated compound shapes. Those analyses may find low slope and low amplitude zero temperature sensitivity angles for particular modes. However, as the resonant elements become more structurally complex it becomes more effective to think in terms of aggregate temperature sensitivity of resonant Eigen modes rather than of weighted sums of contributors. In this perspective there are only Eigen modes of the structure, not sub-structures, and those Eigen modes have temperature sensitivities. Experimental data is usually also of the form that specific modes have specific temperature sensitivities, rather than containing data on individual sub-components of the resonating elements.

It is also important to understand that simulations in general and the finite element simulations for these resonant structures in particular, are usually of limited accuracy. One

must sometimes iterate over design and testing to refine the structure. This design process, including finite element design, and experimental iteration of resonant structures to find low temperature sensitivity structures is part of the inventions described here.

FIG. 4A shows a 524 kHz resonator with anchors **39**, flexures **40**, and proof masses **41**. Drive and sense electrodes (not shown) engage with the resonant element at the serrated edges of **41** and occupy areas **42** and **43**.

FIG. 4B shows the temperature sensitivity, or TCf, of the resonator shown in FIG. 4A as a function of rotation to the crystallographic axis. The angle where the temperature sensitivity is zero, about 18 degrees in this case, is marked with a circle. Therefore, this structure may be able to be made insensitive to temperature by rotating it about 18 degrees with respect to the crystallographic axis.

Not that the data in 4B is again for a particular dopant density and gradient, and for a particular dopant in a particular semiconductor. In this case it is for a phosphorus concentration of about 10^{20} cm^{-3} in <100> aligned silicon. It should be understood that the resonator shown in FIG. 4A and data in FIG. 4B are only one example of many possible instantiations of the inventions described herein.

FIG. 5 shows an alternative 524 kHz resonator with beams that are angled outward from the semiconductor axis. The beams **40** are angled at or near the direction that will produce zero temperature sensitivity, but they are rotated on opposite sides of the direction. In this way if the resonator or crystal are rotated, two of the four beams will tend to increase in angle and two beams will tend to decrease in angle. This will have to first order a canceling effect and the structure will be to first order insensitive to small changes in crystallographic orientation. In this way the manufacturing sensitivity to angle may be reduced. The structure includes beams **40**, proof masses **41** and electrodes **42** and **43**.

FIG. 6 shows a resonator that includes beams **40** built near zero temperature sensitivity angles. Like the resonator in FIG. 5, these can be designed off the zero angles such that changes in crystallographic alignment have a canceling effect, for instance with parts of the beams increasing and parts decreasing in temperature sensitivity. This resonator may have additional virtues, for instance the teeth engaging the sense and drive electrodes **44** and **45** to the proof masses **41** move in a parallel fashion, which can improve electrical efficiency.

It should be understood that the modification of the resonators from what is shown in FIG. 4A to FIGS. 5 and 6 can enhance manufacturing tolerance. However, it should also be understood that there are other factors that go into the choice of resonant structure shape and there are additional criteria that may be considered for the best design. Indeed may additional criteria are considered and the discussion above should not be considered limiting.

FIG. 7A shows a 70 MHz resonator built from an array of sixteen rings **45** with couplers **47** outside electrodes **39** and inside electrodes **42**. These rings oscillate in a tensile breathing mode, where a single mode is coupled across the rings **45** by the flexures **47**. While this figure shows sixteen rings, these resonators can be built with more or less, for instance 64 rings or 9 rings. The anchor **39** shown in the middle can be arrayed across more locations to support larger resonator structures.

FIG. 7B shows two example elements of a different array. In this figure we see inner electrodes **48** but no outer electrodes. The resonator shape is shown in **50** and a finite element deformation of the desired mode in **51**. The angle with respect to the crystallographic axis is shown as **52**.

The beam and mass configurations shown in these figures should be understood to be examples of the many possible and contemplated configurations. Various other structures are contemplated and the descriptions here should not be considered limiting. The techniques described in the above figures and the text can be applied widely. For instance rotation of the beam with respect to the crystallographic axis, construction of compound beams, construction of tapered beams, construction of doping gradients, inclusion of multiple materials, semiconductors, and dopants can be combined in many ways to form resonators not shown in these figures or directly describe in the text. However, these combinations, when used to form resonators are within the scope of this invention.

In the description of resonator angles up to this point the temperature sensitivity was described as a single number that can be minimized. However, in a more nuanced view it can be understood to be a non-linear function that can be described as a polynomial. The first order term represents the value we have been discussing to this point. The higher order terms are also functions of the structure, rotation, material, dopant etc., and they may or may not be able to be simultaneously minimized. Therefore there is often a residual curvature in the temperature sensitivity when the first order is removed or nearly removed.

Over fabrication processes the geometries, dopants, and other parameters may shift. These shifts can change the temperature sensitivity of the resonant elements. It can be important to quantify the total range of temperature sensitivities that a resonant element may show across process variations. FIG. 8 shows the locus of temperature sensitivities that a kHz resonator may show over process. Careful analysis in simulation and through measurement may be helpful for proper resonant element construction. Analyzing these across corners and defining structures that are minimally sensitive to process corners is one of the express objects of this invention.

FIG. 8 also shows a residual higher order curvature in the temperature sensitivity as described above. In this case it is predominantly second order upward, but it can take other shapes as well, particularly it can be cubic in some cases.

It should be further noted that various structures (for example, the structures of the MEMS device), circuits and/or circuitry may be described using computer aided design tools and expressed (or represented), as data and/or instructions embodied in various computer-readable media, in terms of their behavioral, register transfer, logic component, transistor, layout geometries, and/or other characteristics. Formats of files and other objects in which such structure and/or circuit expressions may be implemented include, but are not limited to, formats supporting behavioral languages such as C, Verilog, HDL, and Matlab, formats supporting register level description languages like RTL, and formats supporting geometry description languages such as GDSII, GDSIII, GDSIV, CIF, MEBES and any other suitable formats and languages. Computer-readable media in which such formatted data and/or instructions may be embodied include, but are not limited to, non-volatile storage media in various forms (e.g., optical, magnetic or semiconductor storage media) and carrier waves that may be used to transfer such formatted data and/or instructions through wireless, optical, or wired signaling media or any combination thereof. Examples of transfers of such formatted data and/or instructions by carrier waves include, but are not limited to, transfers (uploads, downloads, e-mail, etc.) over the Internet and/or other computer networks via one or more data transfer protocols (e.g., HTTP, FTP, SMTP, etc.).

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Indeed, when received within a computer system via one or more computer-readable media, such data and/or instruction-based expressions of the above described structures, circuits and/or circuitry may be processed by a processing entity (e.g., one or more processors) within the computer system in conjunction with execution of one or more other computer programs including, without limitation, net-list generation programs, place and route programs and the like, to generate a representation or image of a physical manifestation of such structures, circuits and/or circuitry. Such representation or image may thereafter be used in device fabrication, for example, by enabling generation of one or more masks that are used to form various components of the circuits in a device fabrication process.

Moreover, the various structures (for example, the structures of the MEMS device), circuits and/or circuitry disclosed herein may be represented via simulations using computer aided design and/or testing tools. The simulation of the various structures and/or characteristics or operations thereof may be implemented by a computer system wherein characteristics and operations of such structures and/or circuitry, and techniques implemented thereby, are imitated, replicated and/or predicted via a computer system. The present inventions are also directed to such simulations of the inventive structures and circuitry, and/or techniques implemented thereby, and, as such, are intended to fall within the scope of the present inventions. The computer-readable media corresponding to such simulations and/or testing tools are also intended to fall within the scope of the present inventions.

What is claimed is:

1. A micromachined resonator comprising:

first, second, third and fourth beams, each of the beams comprising at least one layer of crystal silicon having an impurity dopant having concentration which is greater than or equal to 10^{19} cm^{-3} , each of the beams an anchored end and a swinging end;

at least one connecting structure that mechanically couples together the anchored ends of the beams;

at least one electrode to sense a frequency of resonant motion of the micromachined resonator; and wherein

the micromachined resonator has a first axis, the swinging end of the first beam and the swinging end of the second beam are arranged symmetrically about the first axis, and are configured to simultaneously deflect under the application of electrical stimuli, in a common direction along a second axis perpendicular to the first axis,

the swinging end of the third beam and the swinging end of the fourth beam are also arranged symmetrically about the first axis, and are also configured to simultaneously deflect under the application of the electrical stimuli, in a common direction along the second axis,

the swinging end of the first beam and the swinging end of the third beam are configured to simultaneously deflect under the application of the electrical stimuli in opposite directions along the second axis,

the swinging end of the second beam and the swinging end of the fourth beam are configured to simultaneously deflect under the application of the electrical stimuli in opposite directions along the second axis,

the at least one layer of crystal silicon has a silicon crystal axis;

the micromachined resonator is oriented, relative to the silicon crystal axis, such that a first two of the first

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beam, the second beam, the third beam and the fourth beam are oriented at respective, reciprocal angles about the silicon crystal axis and such that a second two of the first beam, the second beam, the third beam and the fourth beam are oriented at respective, reciprocal angles about the silicon crystal axis;

the first two and the second two are oriented at one or more angles relative to each other, such that a contribution of a temperature coefficient of frequency (TCF) of the micromachined resonator from the first two is substantially negated by a contribution to the TCF from the second two; and

the sensed frequency of resonant motion is dependent on motion of the four beams.

2. The micromachined resonator of claim 1 wherein the impurity dopant is one of phosphorus, arsenic, bismuth or antimony.

3. The micromachined resonator of claim 1 wherein the concentration of impurity dopant is non-uniform within each of the four beams along at least one dimensional axis.

4. The micromachined resonator of claim 3 wherein the at least one dimensional axis comprises a length dimension of the respective beam.

5. The micromachined resonator of claim 1 wherein the impurity dopant is phosphorus.

6. The micromachined resonator of claim 1, embodied as a multi-die package, the multi-die package comprising a circuits die and a MEMS die, the micromachined resonator being carried by the MEMS die, the MEMS die being flip-chip mounted to the circuits die.

7. A device comprising:

a first die, having a micromachined resonator and at least one electrode to sense a frequency of resonant motion of the micromachined resonator, wherein the micromachined resonator comprises first, second, third and fourth beams, each of the beams comprising at least one layer of crystal silicon having an impurity dopant having concentration which is greater than or equal to 10^{19} cm^{-3} , each of the beams an anchored end and a swinging end, and at least one connecting structure that mechanically couples together the anchored ends of the beams; and

a second die having electronic circuitry that is structured so as to, during operation of the integrated circuit package, excite the micromachined resonator to vibrating motion and to generate, from a signal provided by the at least one electrode, a timing signal;

wherein

the micromachined resonator has a first axis, the swinging end of the first beam and the swinging end of the second beam are arranged symmetrically about the first axis, and are configured to simultaneously deflect under the application of electrical stimuli, in a common direction along a second axis perpendicular to the first axis,

the swinging end of the third beam and the swinging end of the fourth beam are also arranged symmetrically about the first axis, and are also configured to simultaneously deflect under the application of the electrical stimuli, in a common direction along the second axis,

the swinging end of the first beam and the swinging end of the third beam are configured to simultaneously deflect under the application of the electrical stimuli in opposite directions along the second axis,

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the swinging end of the second beam and the swinging end of the fourth beam are configured to simultaneously deflect under the application of the electrical stimuli in opposite directions along the second axis, the at least one layer of crystal silicon has a silicon crystal axis,

the micromachined resonator is oriented, relative to the silicon crystal axis, such that a first two of the first beam, the second beam, the third beam and the fourth beam are oriented at respective, reciprocal angles about the silicon crystal axis and such that a second two of the first beam, the second beam, the third beam and the fourth beam are oriented at respective, reciprocal angles about the silicon crystal axis,

the first two and the second two are oriented at one or more angles relative to each other, such that a contribution of a temperature coefficient of frequency (TCF) of the micromachined resonator from the first two is substantially negated by a contribution to the TCF from the second two, and the sensed frequency of resonant motion is dependent on motion of the four beams.

8. The device of claim 7 wherein the impurity dopant is one of phosphorus, arsenic, bismuth or antimony.

9. The device of claim 7 further comprising at least one encapsulant, wherein the at least one encapsulant is configured such that the first die is completely encapsulated relative to an ambient environment external to the device.

10. The device of claim 7 wherein the micromachined resonator is formed from a silicon wafer in a manner such that the micromachined resonator is anchored at an anchor point and wherein two beams, of the first, second, third and fourth beams, are mirrored versions of one another about at least one axis which passes through the anchor point.

11. The device of claim 10 wherein each beam of the two beams is a mirrored version of a first other one of the first, second, third and fourth beams about a first axis of the at least one axis, and is also a mirrored version of a second other one of the first, second, third and fourth beams about a second axis of the at least one axis.

12. The device of claim 7 wherein:

a constituent material of the micromachined resonator has a temperature-dependent vibration characteristic; and the electronic circuitry comprises active compensation circuitry that is to generate the timing signal dependent on the vibration of the micromachined resonator in a manner such that the timing signal has a temperature-invariant frequency.

13. The device of claim 12 wherein the electronic circuitry is to generate a bias voltage in a manner that varies with temperature, wherein the first die is coupled to the second die to receive the bias voltage and wherein the first die is configured to apply the bias voltage to the micromachined resonator to correct the temperature-dependent vibration characteristic.

14. The device of claim 12 wherein the first die is coupled to the second die to provide to the second die a resonator signal, from the at least one electrode, the resonator signal having a temperature-dependent frequency that varies in dependence on the temperature-dependent vibration characteristic, and wherein the electronic circuitry comprises frequency synthesis circuitry which is configured to electronically generate the timing signal from the resonator signal, in a manner such that the timing signal has a temperature-invariant frequency.

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15. The device of claim 14 wherein the device is further characterized by external electrical contacts and wherein the device is configured so as to provide the timing signal as an output signal via at least one of the external electrical contacts.

16. The device of claim 14 wherein the electronic circuitry comprises a temperature sensor and wherein the first die and the second die are coupled together by a thermally conductive epoxy.

17. The device of claim 16 further comprising at least one encapsulant, wherein the at least one encapsulant is configured such that the first die is completely encapsulated relative to an ambient environment external to the device.

18. The device of claim 7 further comprising a heating element, wherein the electronic circuitry is configured to control the heating element so as to heat the micromachined resonator to a predetermined temperature during operation of the device.

19. The device of claim 18 wherein the micromachined resonator is characterized by a first turnover temperature and wherein the first turnover temperature is greater than the predetermined temperature.

20. The device of claim 7, embodied as a multi-die package, the multi-die package comprising a circuits die and a MEMS die, the micromachined resonator being carried by the MEMS die, wherein the MEMS die is mounted to the circuits die at least in part by, and is electrically connected with the circuits die by, a plurality of solder ball connections, and wherein the multi-chip package further comprises a number of stand-off structures, the MEMS die being mounted in a stacked relationship with the circuits die, with the stand-off structures having a greater height than the MEMS die so as to support the MEMS die in a flip-chip configuration between the circuits die and substrate surface, to which the multi-die package is to be mounted.

21. An oscillator integrated circuit device comprising:

a first die, having a micromachined resonator and at least one electrode to sense a frequency of resonant motion of the micromachined resonator, wherein the micromachined resonator comprises first, second, third and fourth beams, each of the beams comprising at least one layer of crystal silicon having an impurity dopant having concentration which is greater than or equal to 10^{19} cm^{-3} , each of the beams an anchored end and a swinging end, and at least one connecting structure that mechanically couples together the anchored ends of the beams;

a second die having electronic circuitry that is structured so as to, during operation of the integrated circuit package, excite the micromachined resonator to vibrating motion and to generate, from a signal provided by the at least one electrode, a timing signal;

an encapsulant that in combination with one or more structures, insulates the first die from an ambient environment external to the oscillator integrated circuit device; and

one or more electrical contacts on an exterior surface of the oscillator integrated circuit device, the one or more electrical contacts to provide the timing signal as an output of the oscillator integrated circuit device;

wherein

the micromachined resonator has a first axis,

the swinging end of the first beam and the swinging end of the second beam are arranged symmetrically about the first axis, and are configured to simultaneously deflect under the application of electrical

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stimuli, in a common direction along a second axis perpendicular to the first axis,

the swinging end of the third beam and the swinging end of the fourth beam are also arranged symmetrically about the first axis, and are also configured to simultaneously deflect under the application of the electrical stimuli, in a common direction along the second axis,

the swinging end of the first beam and the swinging end of the third beam are configured to simultaneously deflect under the application of the electrical stimuli in opposite directions along the second axis,

the swinging end of the second beam and the swinging end of the fourth beam are configured to simultaneously deflect under the application of the electrical stimuli in opposite directions along the second axis,

the at least one layer of crystal silicon has a silicon crystal axis,

the micromachined resonator is oriented, relative to the silicon crystal axis, such that a first two of the first beam, the second beam, the third beam and the fourth beam are oriented at respective, reciprocal angles about the silicon crystal axis and such that a second two of the first beam, the second beam, the

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third beam and the fourth beam are oriented at respective, reciprocal angles about the silicon crystal axis,

the first two and the second two are oriented at one or more angles relative to each other, such that a contribution of a temperature coefficient of frequency (TCF) of the micromachined resonator from the first two is substantially negated by a contribution to the TCF from the second two, and

the sensed frequency of resonant motion is dependent on motion of the four beams.

22. The oscillator integrated circuit device of claim **21**, embodied as a multi-die package, the multi-die package comprising a circuits die and a MEMS die, the micromachined resonator being carried by the MEMS die, wherein the MEMS die is mounted to the circuits die at least in part by, and is electrically connected with the circuits die by, a plurality of solder ball connections, and wherein the multi-chip package further comprises a number of stand-off structures, the MEMS die being mounted in a stacked relationship with the circuits die, with the stand-off structures having a greater height than the MEMS die so as to support the MEMS die in a flip-chip configuration between the circuits die and substrate surface, to which the multi-die package is to be mounted.

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