

US Patent & Trademark Office

Patent Public Search | Text View

United States Patent	12395204
Kind Code	B2
Date of Patent	August 19, 2025
Inventor(s)	Howard; John

Random, sequential, or simultaneous multi-beam circular antenna array and beam forming networks with up to 360° coverage

Abstract

A beam forming network system includes a first beam forming network having first and second ports, in which each of the first ports is operatively coupled to an antenna element; and a second beam forming network including third and fourth ports, in which each of the third ports is operatively coupled to one of the second ports using at least one of a phase shifter, attenuator, power divider, and/or hybrid coupler. A method of beam forming includes coupling each of the first ports associated with a first beam forming network operatively to one antenna element, and coupling each of the third ports associated with a second beam forming network operatively to one of the second ports associated with the first beam forming network using at least one of a phase shifter, attenuator, power divider, and/or hybrid coupler.

Inventors:	Howard; John (Upper Mount Bethel, PA)
Applicant:	Howard; John (Upper Mount Bethel, PA)
Family ID:	1000008763462
Assignee:	Howard; John (Upper Mount Bethel, PA)
Appl. No.:	18/386626
Filed:	November 03, 2023

Prior Publication Data

Document Identifier	Publication Date
US 20240137065 A1	Apr. 25, 2024
US 20240235603 A9	Jul. 11, 2024

Related U.S. Application Data

continuation parent-doc US 16891244 20200603 US 11855680 child-doc US 18386626
continuation-in-part parent-doc US 14227634 20140327 US 10734733 20200804 child-doc US
16891244
us-provisional-application US 61874407 20130906

Publication Classification

Int. Cl.: **H01Q3/40** (20060101); **H04B1/04** (20060101); **H04B1/58** (20060101); **H04B7/0426**
(20170101)

U.S. Cl.:

CPC **H04B1/583** (20130101); **H04B1/0458** (20130101); **H04B7/043** (20130101);

Field of Classification Search

CPC: H04B (1/583); H04B (1/0458); H04B (7/043); H04B (7/0617); H01Q (3/40); H01Q
(21/205); H01Q (25/00)

References Cited

U.S. PATENT DOCUMENTS

Patent No.	Issued Date	Patentee Name	U.S. Cl.	CPC
3736592	12/1972	Coleman	N/A	N/A
3906502	12/1974	Connolly	N/A	N/A
4408205	12/1982	Hockham	342/368	H01Q 3/2617
4652879	12/1986	Rudish et al.	N/A	N/A
4980692	12/1989	Rudish et al.	N/A	N/A
5093668	12/1991	Sreenivas	342/373	H01Q 25/00
5151706	12/1991	Roederer et al.	N/A	N/A
5179386	12/1992	Rudish et al.	N/A	N/A
5325101	12/1993	Rudish et al.	N/A	N/A
5430453	12/1994	Rudish et al.	N/A	N/A
5610617	12/1996	Gans	342/372	H01Q 3/24
5805575	12/1997	Kamin, Jr.	370/335	H01Q 3/22
5929804	12/1998	Jones et al.	N/A	N/A
5987037	12/1998	Gans	370/480	H01Q 25/00
6633551	12/2002	Kent	370/316	H04B 7/02
6791507	12/2003	Johansson et al.	N/A	N/A
6864853	12/2004	Judd et al.	N/A	N/A
6992621	12/2005	Casas et al.	N/A	N/A
6992622	12/2005	Chiang et al.	N/A	N/A
6995730	12/2005	Pleva et al.	N/A	N/A
7183995	12/2006	Pleva	342/374	G01S 7/352
7245938	12/2006	Sobczak	343/890	H01Q 3/40
7248215	12/2006	Pleva et al.	N/A	N/A
7567213	12/2008	Liu	N/A	N/A
8604989	12/2012	Olsen	N/A	N/A

2003/0098814	12/2002	Keller, III et al.	N/A	N/A
2004/0160374	12/2003	Johansson et al.	N/A	N/A
2005/0259005	12/2004	Chiang	342/373	H01Q 25/00
2006/0145919	12/2005	Pleva	342/368	H01Q 1/3283
2010/0066590	12/2009	Brown et al.	N/A	N/A
2011/0032849	12/2010	Yeung	370/328	H04B 7/0617
2011/0102263	12/2010	Angeletti	342/373	H01Q 3/40
2014/0375518	12/2013	Powell	343/816	H01Q 3/40
2015/0070241	12/2014	Howard et al.	N/A	N/A
2015/0341098	12/2014	Angeletti	375/267	H01Q 3/40
2015/0349421	12/2014	Sharawi	342/373	H01Q 21/0006

FOREIGN PATENT DOCUMENTS

Patent No.	Application Date	Country	CPC
2846401	12/2014	EP	N/A
2005/122328	12/2004	WO	N/A

OTHER PUBLICATIONS

European Search Report for corresponding European Application No. EP14172161, Jan. 23, 2015. cited by applicant

Non-Final Office Action received for U.S. Appl. No. 14/227,634 dated Oct. 9, 2015, 32 pages. cited by applicant

Ireland et al. Dielectric Embedded Multi-Beam Adaptive Array Antenna, Journal of the Japan Society of Applied Electromagnetics and Mechanics, vol. 15, Supplement (2007). cited by applicant

Panduro et al. Simplifying the Feeding Network for Multi-Beam Circular Antenna Arrays by Using Corps, Progress nElectromagnetics Research Letters, vol. 21, 119-128, 2011. cited by applicant

Final Office Action received for U.S. Appl. No. 14/227,634 dated Mar. 23, 2016, 11 pages. cited by applicant

Non-Final Office Action received for U.S. Appl. No. 14/227,634 dated Jan. 4, 2017, 13 pages. cited by applicant

Final Office Action received for U.S. Appl. No. 14/227,634 dated Jun. 14, 2017, 18 pages. cited by applicant

Bhowmik et al. "Optimum Design of a 4×4 Planar Butler Matrix Array for WLAN Application, Journal of Telecommunications, vol. 2, Issue 1, Apr. 2010". cited by applicant

Non-Final Office Action received for U.S. Appl. No. 14/227,634 dated May 8, 2018, 24 pages. cited by applicant

Cetinoneri et al. "An 8 × 8 Butler Matrix in 0.13-um CMOS for 5-6-GHz Multibeam Applications, IEEE Transactions on Microwave Theory and Techniques, vol. 59, No. 2, Feb. 2011". cited by applicant

Final Office Action received for U.S. Appl. No. 14/227,634 dated Nov. 29, 2018, 16 pages. cited by applicant

Non-Final Office Action received for U.S. Appl. No. 14/227,634 dated Jun. 10, 2019, 15 pages. cited by applicant

Communication pursuant to Article 94(3) EPC received for EP Patent Application Serial No. 14172161.3 dated Feb. 24, 2016, 7 page. cited by applicant

Communication pursuant to Article 94(3) EPC received for EP Patent Application Serial No. 14172161.3 dated Oct. 6, 2016, 8 page. cited by applicant

Final Office Action received for U.S. Appl. No. 14/227,634 dated Dec. 9, 2019, 18 pages. cited by applicant

Notice of Allowance received for U.S. Appl. No. 14/227,634 dated Apr. 10, 2020, 17 pages. cited

by applicant

Office Action for corresponding Application No. 14172161.3, Aug. 21, 2017, 11 pages,. cited by applicant

Primary Examiner: Lopez Cruz; Dimary S

Assistant Examiner: Immanuel; Bamidele A

Attorney, Agent or Firm: FisherBroyles, LLP

Background/Summary

CROSS-REFERENCE TO RELATED APPLICATION (1) This application is a continuation application of U.S. application Ser. No. 16/891,244, filed Jun. 3, 2020, which is a continuation-in-part application of U.S. application Ser. No. 14/227,634, filed Mar. 27, 2014, which claims the benefit of and priority to U.S. Provisional Application No. 61/874,407, filed Sep. 6, 2013, the disclosures of which are incorporated by reference herein in their entireties.

BACKGROUND

Field

(1) Embodiments of the invention generally relate to antennas and, more particularly, relate to random, sequential or simultaneous multi-beam antenna arrays with up to 360° antenna coverage.

SUMMARY

(2) In accordance with one embodiment, a beam forming network system is disclosed, which includes a first beam forming network including a plurality of first ports and a plurality of second ports, in which each of the plurality of first ports is configured to be operatively coupled to one of a plurality of antenna elements; a second beam forming network including a plurality of third ports and a plurality of fourth ports, in which each of the plurality of third ports is operatively coupled to one of the plurality of second ports; and a switch sequentially coupling each of the plurality of fourth ports to a signal by sweeping the switch through a plurality of positions, thereby enabling the plurality of antenna elements to provide sequential 360° coverage.

(3) The first beam forming network may be a $K \times N$ beam forming network, in which K is greater than or equal to N , and the second beam forming network may be an $N \times M$ beam forming network, in which M is less than or equal to N . At least one of the first beam forming network, second beam forming network may include at least one of a Butler matrix, Blass matrix, Nolen matrix, Shelton matrix, McFarland matrix, Davis matrix.

(4) In accordance with another embodiment, a method of beam forming, is disclosed, which includes coupling each of a plurality of first ports associated with a first beam forming network operatively to one of a plurality of antenna elements; coupling each of a plurality of third ports associated with a second beam forming network operatively to one of a plurality of second ports associated with the first beam forming network; and coupling each of a plurality of fourth ports associated with the second beam forming network sequentially to a signal by sweeping a switch through a plurality of positions, thereby enabling the antenna elements to provide sequential 360° coverage.

(5) The first beam forming network may be a $K \times N$ beam forming network, in which K is greater than or equal to N , and the second beam forming network may be an $N \times M$ beam forming network, in which M is less than or equal to N . At least one of the first beam forming network, second beam forming network may include at least one of a Butler matrix, Blass matrix, Nolen matrix, Shelton matrix, McFarland matrix, Davis matrix.

(6) In accordance with another embodiment, a beam forming network system is disclosed, which includes at least one first beam forming network including a plurality of first ports and a plurality of second ports, in which each of the plurality of first ports is configured to be operatively coupled to one of a plurality of antenna elements; and at least one second beam forming network including a plurality of third ports and a plurality of fourth ports, in which each of the plurality of third ports being operatively coupled to one of the plurality of second ports using at least one of a first variable phase shifter, first fixed phase shifter, first attenuator, first power divider, first hybrid coupler.

(7) The first beam forming network may be an $MN \times MN$ beam forming network, in which N is an integer greater than or equal to one (1) and M is an integer greater than or equal to one (1); the second beam forming network may be an $N \times N$ beam forming network, in which N is an integer greater than or equal to one (1); and the first beam forming network may be an $N \times (N+M)$ beam forming network, in which N is an integer greater than or equal to one (1) and M is an integer greater than or equal to one (1). At least one of the first beam forming network, second beam forming network may include at least one of a Butler matrix, Blass matrix, Nolen matrix, Shelton matrix, McFarland matrix, Davis matrix. The first hybrid coupler may include at least one of a 90 degree hybrid coupler, 180 degree hybrid coupler. At least one of amplitude, phase may be controlled for sidelobe reduction in at least one of azimuth, elevation using at least one of the first variable phase shifter, first fixed phase shifter, first attenuator, first power divider, first hybrid coupler. The first beam forming network may be an $N \times N$ beam forming network, in which N is an integer greater than or equal to one (1). Each of the plurality of fourth ports may be configured to be operatively coupled to a switch operatively coupling each of the plurality of fourth ports to a signal by sweeping the switch through a plurality of positions. Each of the plurality of fourth ports may be configured to be operatively coupled to one of a plurality of transceivers operatively coupling one of the plurality of fourth ports to a signal. The plurality of antenna elements may be configured in at least one of a circle, cylinder, semi-circle, arc, line, sphere, conformal shape, curvilinear shape. The beam forming network system may include at least one third beam forming network including a plurality of fifth ports and a plurality of sixth ports, in which the plurality of fifth ports is configured to be operatively coupled to a one of the plurality of fourth ports. The beam forming network system may include at least one fourth beam forming network including a plurality of seventh ports and a plurality of eighth ports, in which each of the plurality of seventh ports is operatively coupled to one of the plurality of sixth ports using at least one of a second variable phase shifter, second fixed phase shifter, second attenuator, second power divider, second hybrid coupler. The second hybrid coupler may include at least one of a 90 degree hybrid coupler, 180 degree hybrid coupler. At least one of amplitude, phase may be controlled for sidelobe reduction in at least one of azimuth, elevation using at least one of the second variable phase shifter, second fixed phase shifter, second attenuator, second power divider, second hybrid coupler. Each of the plurality of eighth ports may be configured to be operatively coupled to a switch selectively coupling each of the plurality of eighth ports to a signal by sweeping the switch through a plurality of positions. Each of the plurality of eighth ports may be configured to be operatively coupled to one of a plurality of transceivers operatively coupling one of the plurality of eighth ports to a signal. The second beam forming network may include a power divider.

(8) In accordance with another embodiment, a method of beam forming is disclosed, which includes coupling each of a plurality of first ports associated with at least one first beam forming network operatively to one of a plurality of antenna elements and coupling each of a plurality of third ports associated with at least one second beam forming network operatively to one of a plurality of second ports associated with the first beam forming network using at least one of a first variable phase shifter, first fixed phase shifter, first attenuator, first power divider, first hybrid coupler.

(9) The first beam forming network may be an $MN \times MN$ beam forming network, in which N is an integer greater than or equal to one (1) and M is an integer greater than or equal to one (1); the

second beam forming network may be an $N \times N$ beam forming network, in which N is an integer greater than or equal to one (1); and the first beam forming network may be an $N \times (N+M)$ beam forming network, in which N is an integer greater than or equal to one (1) and M is an integer greater than or equal to one (1). At least one of the first beam forming network, second beam forming network may include at least one of a Butler matrix, Blass matrix, Nolen matrix, Shelton matrix, McFarland matrix, Davis matrix. The first hybrid coupler may include at least one of a 90 degree hybrid coupler, 180 degree hybrid coupler. The method may include controlling at least one of amplitude, phase for sidelobe reduction in at least one of azimuth, elevation using at least one of the first variable phase shifter, first fixed phase shifter, first attenuator, first power divider, first hybrid coupler. The first beam forming network may be an $N \times N$ beam forming network, in which N is an integer greater than or equal to one (1). The method may include coupling each of a plurality of fourth ports associated with the second beam forming network operatively to a switch operatively coupling each of the plurality of fourth ports to a signal by sweeping the switch through a plurality of positions. The method may include coupling each of a plurality of fourth ports associated with the second beam forming network operatively to one of a plurality of transceivers operatively coupling one of the plurality of fourth ports to a signal. The plurality of antenna elements may be configured in at least one of a circle, cylinder, semi-circle, arc, line, sphere, conformal shape, curvilinear shape. The method may include coupling a plurality of fifth ports associated with at least one third beam forming network operatively to one of a plurality of fourth ports associated with the second beam forming network. The method may include coupling each of a plurality of seventh ports associated with at least one fourth beam forming network operatively to one of a plurality of sixth ports associated with the at least one third beam forming network using at least one of a second variable phase shifter, second fixed phase shifter, second attenuator, second power divider, second hybrid coupler. The second hybrid coupler may include at least one of a 90 degree hybrid coupler, 180 degree hybrid coupler. The method may include controlling at least one of amplitude, phase for sidelobe reduction in at least one of azimuth, elevation using at least one of the second variable phase shifter, second fixed phase shifter, second attenuator, second power divider, second hybrid coupler. The method may include coupling a plurality of eighth ports associated with the at least one fourth beam forming network operatively to a switch selectively coupling each of the plurality of eighth ports to a signal by sweeping the switch through a plurality of positions. The method may include coupling a plurality of eighth ports associated with the at least one fourth beam forming network operatively to one of a plurality of transceivers operatively coupling one of the plurality of eighth ports to a signal. The second beam forming network may include a power divider.

(10) Other embodiments of the invention will become apparent from the following detailed description considered in conjunction with the accompanying drawings. It is to be understood, however, that the drawings are designed as an illustration only and not as a definition of the limits of any embodiments of the invention.

Description

BRIEF DESCRIPTION OF THE DRAWINGS

- (1) The following drawings are provided by way of example only and without limitation, wherein like reference numerals (when used) indicate corresponding elements throughout the several views, and wherein:
- (2) FIG. 1 shows a matrix fed circular array for continuous scanning;
- (3) FIG. 2 shows an embodiment of a circular antenna array, in which variable and fixed phase shifters shown in FIG. 1 have been replaced with a Butler matrix;
- (4) FIG. 3 shows another embodiment of a circular antenna array, in which variable and fixed phase

shifters shown in FIG. 1 have been replaced with a Butler matrix;

(5) FIG. 4 shows an antenna beam pattern providing 360° coverage;

(6) FIG. 5 shows another embodiment of a circular antenna array, in which the Butler matrices shown in FIG. 3 are implemented as $K \times N$ and $N \times M$ beam forming networks;

(7) FIG. 6 shows another embodiment, in which an $N \times N$ beam forming network is connected to a $3N \times 3N$ beam forming network using variable phase shifters, fixed phase shifters, and three-way power dividers, which include an even or uneven power split;

(8) FIG. 7 shows another embodiment, in which an $N \times N$ beam forming network is connected to a $2N \times 2N$ beam forming network using variable phase shifters, fixed phase shifters, and 180 degree hybrid couplers;

(9) FIG. 8 shows another embodiment, in which an $N \times N$ beam forming network is connected to a $2N \times 2N$ beam forming network using variable phase shifters, fixed phase shifters, and 90 degree hybrid couplers;

(10) FIG. 9 shows another embodiment, in which an $N \times N$ beam forming network is connected to a $N \times (N+M)$ beam forming network using one or more variable phase shifters, fixed phase shifters, and 90 degree hybrid couplers;

(11) FIG. 10 shows another embodiment including a circular antenna array, $N \times N$ beam forming network, variable phase shifters, attenuators, power divider, and signal transceiver;

(12) FIG. 11 shows another embodiment including a circular antenna array, first $M \times M$ beam forming network, attenuators, second $M \times M$ beam forming network, operational switch, and signal transceiver;

(13) FIG. 12 shows another embodiment including a circular antenna array, first $M \times M$ beam forming network, attenuators, second $M \times M$ beam forming network, and independent signal transceivers;

(14) FIG. 13 shows another embodiment including a circular antenna array, $N \times N$ beam forming network, attenuators, hybrid couplers, power dividers, and/or phase shifters, $M \times M$ beam forming network, operational switch, and signal transceiver;

(15) FIG. 14 shows another embodiment including a circular antenna array, $N \times N$ beam forming network, attenuators, hybrid couplers, power dividers, and/or phase shifters, $M \times M$ beam forming network, and signal transceivers;

(16) FIG. 15 shows another embodiment including a circular antenna array, $N \times N$ beam forming network, attenuators, phase shifters, and/or power dividers, $M \times M$ beam forming network, and signal transceivers;

(17) FIG. 16 shows another embodiment including a cylindrical antenna array, $N \times N$ beam forming network, attenuators, hybrid couplers, power dividers, and/or phase shifters, $M \times M$ beam forming network, and signal transceivers;

(18) FIG. 17 shows another embodiment including a conformal array in a single portion of a curvilinear array or in a single portion of a surface conformal array, $N \times N$ beam forming network, attenuators, hybrid couplers, power dividers, and/or phase shifters, $M \times M$ beam forming network, and signal transceivers;

(19) FIG. 18 shows another embodiment including a plurality of cylindrical arrays, beam forming networks, attenuators, hybrid couplers, power dividers, and/or phase shifters, and signal transceivers; and

(20) FIG. 19 shows another embodiment including a plurality of conformal arrays, beam forming networks, attenuators, hybrid couplers, power dividers, and/or phase shifters, and signal transceivers.

(21) It is to be appreciated that elements in the figures are illustrated for simplicity and clarity. Common but well-understood elements, which are useful or necessary in a commercially feasible embodiment, are not shown in order to facilitate a less hindered view of the illustrated embodiments.

DETAILED DESCRIPTION

(22) Embodiments disclosed herein replace variable phase shifters and fixed phase shifters with a Butler matrix beam forming network. Phase and/or amplitude tapering is used to generate narrow beams with reduced sidelobes in azimuth and/or elevation. The elements of the array may be omni and/or directional radiators in broad and/or narrow band configurations.

(23) FIG. 1 shows a matrix fed circular array **10** configured for continuous scanning. The matrix fed circular antenna array **10** includes a circular antenna array **12**, a plurality of antenna elements **14**, a Butler matrix **16**, variable phase shifters **18**, fixed phase shifters **20**, and a power divider **22**. The circular antenna array **12** is coupled to output ports of the Butler matrix **16** by lines **26** of equal length. Each input port of the Butler matrix **16** is coupled to an output port of the power divider **22** through a variable phase shifter **18** and a fixed phase shifter **20**. The power divider **22** is coupled to a transceiver **24**.

(24) FIG. 2 shows a first embodiment **28**, which includes a circular array **42**, a plurality of antenna elements **44**, a first Butler matrix **34**, a second Butler matrix **30**, and an optional switch **32**. The switch **32** can be an analog or a digital switch that selectively directs one or more signals to produce a beam in a certain location of 360° depending on which input of the Butler matrix is chosen. By sweeping through the positions of the switch **32**, the beam can be swept to cover a 360° footprint.

(25) Each of the antenna elements **44** in the circular array **42** is coupled to an output port of the first Butler matrix **34** by lines **36** of equal length. Each input port of the first Butler matrix **34** is coupled to an output port of the second Butler matrix **30**. The second Butler matrix **30** effectively replaces the variable phase shifters **18** and fixed phase shifters **20** shown in FIG. 1. The optional switch **32** selectively couples input ports of the second Butler matrix **30** to a transceiver **38**, and allows a user to switch through each beam to achieve simultaneous or sequential 360° coverage. For example, if the switch **32** applies the signal from the transceiver **38** to each of the inputs of the second Butler matrix, simultaneous 360° coverage is achieved. In addition, if the switch **32** sequentially applies the signal from the transceiver **38** to each of the inputs of the second Butler matrix, sequential 360° coverage is achieved. Further, if the switch **32** applies the signal from the transceiver **38** to less than all of the inputs of the second Butler matrix, partial coverage is achieved. The use of two Butler matrices **30**, **34** enables antenna transmissions to cover 360° simultaneously, which cannot be performed using conventional antenna systems.

(26) FIG. 3 shows a second embodiment having ten (10) input ports to the second Butler matrix **30**. If the Butler matrix **30** is configured correctly, an antenna beam is provided every 36° , that is, at 0° , 36° , 72° , etc. If each of the input ports of the second Butler matrix **30** is connected to a transceiver **48**, as shown in FIG. 3, transmissions can occur simultaneously or sequentially at 360° . In contrast, conventional approaches, such as that shown in FIG. 1, include variable phase shifters **18** and fixed phase shifters **20** that can only sweep through an arc of a predetermined number of degrees in a manner that is similar to a clock's second hand that moves slowly around a central axis. However, this conventional approach provides discontinuous and non-simultaneous coverage over the predetermined arc. Since the variable phase shifters **18** and fixed phase shifters **20** require a certain amount of time to sweep through the predetermined arc, a potential target may be missed or may be allowed to pass through the predetermined arc without being detected due to latency in the phase shifters **18**, **20**. The second embodiment **46** shown in FIG. 3 enables connection of a multi-output transceiver **48** to couple each of the outputs of the second Butler matrix **30** to one or more transceivers **48** to provide 360° coverage.

(27) Further, variable, fixed, and/or digital phase shifters are not as reliable as Butler matrices because the phase shifters are active and not passive. However, Butler matrices are passive and thus more robust and less likely to fail. In addition, Butler matrices can be made to cover a very broad band, which is larger than that of variable, fixed, and/or digital phase shifters.

(28) Thus, the embodiments disclosed herein provide for random, simultaneous and/or sequential

360° antenna coverage without the necessity of scanning. Although 10 (input)×10 (output) Butler matrices are shown and described herein, it is to be understood that any configuration of Butler matrix, such as 8×8, 16×16, and the like may be used while remaining within the intended scope of the disclosure.

(29) FIG. 4 shows an antenna beam pattern **50** with lobes **52** that shows an example of simultaneous 360° antenna coverage provided by the embodiment disclosed herein. In contrast, conventional approaches can only provide for an antenna pattern including fewer than each of the lobes **52**, which are swept through a predetermined arc as function of time and cannot provide for 360° coverage at any given moment in time as shown in FIG. 4. Any combination of beams can be used to provide the 360° coverage, such as without limitation 2, 4, 6, 8, 24, and the like beams. The combination of beams depends on the construction and phase of the Butler matrices. The crossing and/or overlap between beams can also vary depending on the design of the Butler matrices.

(30) FIG. 5 shows a third embodiment **54**, which includes a circular antenna array **56**, a plurality of antenna elements **58**, a first non-square ($K \times N$) beam forming network **60**, a second non-square ($N \times M$) beam forming network **62**, and a plurality of transceivers **64**. In the depicted embodiment, the first and second beam forming networks are $K \times N$ and $N \times M$ beam forming networks, respectively. The circular antenna array **56** includes the plurality of antenna elements **58** and is operatively coupled to the first non-square ($K \times N$) beam forming network **60**. The first non-square ($K \times N$) beam forming network **60** is operatively coupled to the second non-square ($N \times M$) beam forming network **62**, which is operatively coupled to the plurality of transceivers **64**. It should be understood that each of the beam forming networks **60**, **62** is not limited to a Butler matrix. Other beam forming networks such as at least one of a Butler matrix, Blass matrix, Nolen matrix, Shelton matrix, McFarland matrix, and/or Davis matrix may be used as well.

(31) Each of the antenna elements **58** in the circular array **56** is coupled to an output port of the $K \times N$ beam forming network **60** using K lines **66** of substantially equal length. Each input port of the $K \times N$ beam forming network **60** is coupled to an output port of the $N \times M$ beam forming network **62** by N lines **68** of substantially equal length. A combination of the $K \times N$ beam forming network **60** and the $N \times M$ beam forming network **62** effectively replaces the variable phase shifters **18** and fixed phase shifters **20** shown in FIG. 1. Each of the input ports of the $N \times M$ beam forming network **62** is connected to each of the plurality of transceivers **64** using M lines **70** of substantially equal length such that transmission and/or reception can occur simultaneously or sequentially at 360°. In another embodiment, each of the input ports of the $N \times M$ beam forming network **62** may be connected to a switch, such as the switch **32** shown in FIG. 2, such that transmission and/or reception can occur simultaneously or sequentially at 360°. These embodiments provide unique radiation patterns that are different and distinct from antenna array systems using traditional square ($N \times N$) Butler matrices. The quantity of K lines **66** is greater than or equal to the quantity of N lines **68**, and the quantity of M lines **70** is less than or equal to the quantity of N lines **68**.

(32) FIGS. 6-8 illustrate examples of multibeam antenna systems implemented using unequal or equal beamforming networks. For example, FIG. 6 shows another embodiment, in which an $N \times N$ beam forming network **72** is operatively coupled to a $MN \times MN$ beamforming network **80**, in which $M=3$, using one or more variable phase shifters **76**, fixed phase shifters **78**, and/or three-way power dividers **74**, which include an even or uneven power split. M can be any integer greater than or equal to two (2). M can also be equal to one (1), which results in a retro-directive antenna.

(33) FIG. 7 shows another embodiment, in which an $N \times N$ beam forming network **82** is operatively coupled to a $MN \times MN$ beam forming network **90**, in which $M=2$, using one or more variable phase shifters **86**, fixed phase shifters **88**, and/or 180 degree hybrid couplers **84**. M can be any integer greater than or equal to two (2). M can also be equal to one (1), which results in a retro-directive antenna.

(34) FIG. 8 shows another embodiment of the disclosed subject matter, in which an $N \times N$ beam forming network **92** is operatively coupled to a $MN \times MN$ beam forming network **100**, in which

M=2, using one or more variable phase shifters **96**, fixed phase shifters **98**, and/or 90 degree hybrid couplers **94**. M can be any integer greater than or equal to two (2). M can also be equal to one (1), which results in a retro-directive antenna.

(35) FIG. **9** shows another embodiment of the disclosed subject matter, in which an $N \times N$ beam forming network **102** is operatively coupled to a $N \times (N+M)$ beam forming network **110**, using one or more variable phase shifters **106**, fixed phase shifters **108**, and 90 degree hybrid couplers **104**. M can be any integer greater than or equal to two (2). M can also be equal to one (1), which results in a retro-directive antenna.

(36) In accordance with one or more of the disclosed embodiments, signals are able to be received from one direction and transmitted in a different direction. In addition, one or more techniques described in U.S. Pat. No. 8,170,634 may be implemented between beam forming networks in accordance with the disclosed subject matter to reduce sidelobe levels and provide power dividers with power division having various phase variations. Further, embodiments in accordance with the disclosed subject matter utilize amplitude modes as well as phase modes, wherein amplitude and/or phase is controlled and/or tapered for sidelobe reduction in azimuth and/or elevation. Yet further, it is to be noted that utilizing power division, unequal power, and/or various phases of M outputs connected to an $MN \times MN$ beam forming network enables the antenna pattern of an antenna array to yield increased gain, reduced beam width, reduced sidelobe levels, and the selection of beam crossing width to increase the signal-to-noise ratio. In accordance with one or more embodiments disclosed herein, it is to be noted that there is no constraint regarding multiple beam forming networks being required to be of the same order, type, and/or quantity of input and/or output connections, such as a requirement that both beam forming networks be 8×8 or 16×16 , as shown in, for example, in FIG. **2**.

(37) FIG. **9** shows another embodiment of the disclosed subject matter, in which an $N \times N$ beam forming network **102** is operatively coupled to a $N \times (N+M)$ beam forming network **110**, in which $M=7$, using one or more variable phase shifters **106**, fixed phase shifters **108**, and/or 90 degree hybrid couplers **104**. M can be any integer greater than or equal to one (1). This embodiment provides beam crossovers at different values and reduces sidelobe levels, which results in a substantially improved signal-to-noise ratio that is advantageous in 5G communication applications.

(38) FIG. **10** shows a circular antenna array **120**, an $N \times N$ beam forming network **118**, variable phase shifters **116**, attenuators **114**, a power divider **112**, and a signal transceiver **122**. The circular antenna array **120** is operatively coupled to output ports of the $N \times N$ beam forming network **118**. Each input port of the $N \times N$ beam forming network **118** is operatively coupled to an output port of the power divider **112** through one or more variable phase shifters **116** and/or attenuators **114**. The power divider **112** is operatively coupled to the signal transceiver **122**. In this embodiment, amplitude tapering is performed using the variable phase shifters **116** and/or attenuators **114** for sidelobe reduction in azimuth. It is to be noted that if the circular antenna array is configured vertically rather than horizontally with similar amplitude tapering, then sidelobe reduction occurs in elevation. Crossings of beams are in the region of -3 dB.

(39) FIG. **11** shows a circular antenna array **124**, a first $M \times M$ beam forming network **126**, attenuators **128**, a second $M \times M$ beam forming network **130**, an operational switch **132**, and a signal transceiver **134**. The circular antenna array **124** is operatively coupled to output ports of the first $M \times M$ beam forming network **126**. Each input port of the first $M \times M$ beam forming network **126** is operatively coupled to an output port of the second $M \times M$ beam forming network **130** through one or more attenuators **128**. Each input port of the second $M \times M$ beam forming network **130** is selectively coupled to the operational switch **132**. The operational switch **132** can be an analog or a digital switch that selectively directs one or more signals to produce a beam in a certain location of 360° , depending on which input of the second $M \times M$ beam forming network **130** the signal **134** is directed to by the operational switch **134**. By sweeping through the positions of the

switch **132**, the beam can be swept to cover 360° . In this embodiment, amplitude tapering for sidelobe reduction is performed in either azimuth or elevation using the attenuators **128**, depending upon whether the antenna array is configured horizontally or vertically. Crossings of beams are in the region of -3 dB.

(40) FIG. **12** shows a circular antenna array **136**, a first $M \times M$ beam forming network **138**, attenuators **140**, a second $M \times M$ beam forming network **142**, and independent signal transceivers **144**. The circular antenna array **136** is operatively coupled to the first $M \times M$ beam forming network **138**, which is operatively coupled to a second $M \times M$ beam forming network **142** using one or more of the attenuators **140**. The attenuators **40** are operatively coupled to the independent signal transceivers **144**. The second beam forming network **142** is configurable to provide an antenna beam every 36° , that is, at 0° , 36° , 72° , and the like. If each of the input ports of the second beam forming network **142** is connected to a transceiver **144**, as shown in FIG. **12**, transmission can occur simultaneously or sequentially over a 360° area. In this embodiment, amplitude tapering can be performed for sidelobe reduction in either azimuth and/or elevation using the attenuators **140**, depending upon whether the circular array is configured horizontally or vertically. Crossings of beams are in the region of -3 dB.

(41) FIG. **13** shows a circular antenna array **146**, an $N \times N$ beam forming network **148**, attenuators **150**, hybrid couplers, power dividers, and/or phase shifters **151**, an $M \times M$ beam forming network **152**, an operational switch **154**, and a signal transceiver **156**. The circular antenna array **146** is operatively coupled to the $N \times N$ beam forming network **148**, which is operatively coupled to the $M \times M$ beam forming network **152** using one or more attenuators **150**, hybrid couplers, power dividers, and/or phase shifters **151**. The $M \times M$ beam forming network **152** is selectively coupled to the operational switch **154**, which is operatively coupled to the signal transceiver **156**. In this embodiment, amplitude tapering for sidelobe reduction is performed in either azimuth or elevation using the attenuators **150**, hybrid couplers, power dividers, and/or phase shifters **151**, depending upon whether the antenna array is configured horizontally or vertically. Crossings of beams may be configured using the attenuators **150**, hybrid couplers, power dividers, and/or phase shifters **151**, which also control the beam widths, at any level, such as for example -10 dB.

(42) FIG. **14** shows a circular antenna array **158**, an $N \times N$ beam forming network **160**, attenuators **162**, hybrid couplers, power dividers, and/or phase shifters **163**, an $M \times M$ beam forming network **164**, and signal transceivers **166**. The circular antenna array **158** is operatively coupled to the $N \times N$ beam forming network **160**, which is operatively coupled to the $M \times M$ beam forming network **164** using one or more attenuators **162**, hybrid couplers, power dividers, and/or phase shifters **163**. The $M \times M$ beam forming network **164** is operatively coupled to the one or more of the signal transceivers **166**. In this embodiment, amplitude tapering for sidelobe reduction is performed in either azimuth or elevation using the attenuators **162**, hybrid couplers, power dividers, and/or phase shifters **163**, depending upon whether the antenna array is configured horizontally or vertically. Crossings of beams may be configured using the, attenuators **162**, hybrid couplers, power dividers, and/or phase shifters **163**, which also control the beam widths, at any level, such as for example -10 dB.

(43) FIG. **15** shows a circular antenna array **168**, an $N \times N$ beam forming network **170**, attenuators **172**, phase shifters **174**, and/or power dividers **175**, an $M \times M$ beam forming network **176**, and signal transceivers **178**. The circular antenna array **168** is operatively coupled to the $N \times N$ beam forming network **170**, which is operatively coupled to the $M \times M$ beam forming network **176** using one or more attenuators **172**, phase shifters **174**, and/or power dividers **175**. The $M \times M$ beam forming network **176** is operatively coupled to one or more of the signal transceivers **178**. In this embodiment, amplitude tapering for sidelobe reduction is performed in either azimuth or elevation using the attenuators **172**, phase shifters **174**, and/or power dividers **175**, depending upon whether the antenna array is configured horizontally or vertically. Crossings of beams may be configured using the attenuators **172**, phase shifters **174**, and/or power dividers **175**, which also control the

beam widths, at any level, such as for example -10 dB.

(44) FIG. 16 shows a cylindrical antenna array **180**, an $N \times N$ beam forming network **182**, attenuators **184**, hybrid couplers, power dividers, and/or phase shifters **185**, an $M \times M$ beam forming network **186**, and signal transceivers **188**. The cylindrical antenna array **180** is operatively coupled to the $N \times N$ beam forming network **182**, which is operatively coupled to the $M \times M$ beam forming network **186** using one or more of the attenuators **184**, hybrid couplers, power dividers, and/or phase shifters **185**. The $M \times M$ beam forming network **186** is operatively coupled to the one or more of the signal transceivers **188**. In this embodiment, amplitude tapering for sidelobe reduction is performed in either azimuth or elevation using the attenuators **184**, hybrid couplers, power dividers, and/or phase shifters **185**, depending upon whether the antenna array is configured horizontally or vertically. Crossings of beams may be configured using the attenuators **184**, hybrid couplers, power dividers, and/or phase shifters **185**, which also control the beam widths, at any level, such as for example -10 dB.

(45) FIG. 17 shows a conformal array as a single portion of a curvilinear array or as a single portion of a surface conformal array **190**, an $N \times N$ beam forming network **192**, attenuators **194**, hybrid couplers, power dividers, and/or phase shifters **195**, an $M \times M$ beam forming network **196**, and signal transceivers **198**. The conformal array **190** is operatively coupled to the $N \times N$ beam forming network **192**, which is operatively coupled to the $M \times M$ beam forming network **196** using one or more of the attenuators **194**, hybrid couplers, power dividers, and/or phase shifters **195**. The $M \times M$ beam forming network **196** is operatively coupled to one or more of the signal transceivers **198**. In this embodiment, amplitude tapering for sidelobe reduction is performed in either azimuth or elevation using the attenuators **194**, hybrid couplers, power dividers, and/or phase shifters **195**, depending upon whether the antenna array is configured horizontally or vertically. Crossings of beams may be configured using the attenuators **194**, hybrid couplers, power dividers, and/or phase shifters **195**, which also control the beam widths, at any level, such as for example -10 dB.

(46) FIG. 18 shows a plurality of cylindrical arrays **200**, a plurality of beam forming network stages **202**, **204**, **206**, **208**, which are connected using attenuators **210**, **211**, **224**, hybrid couplers, power dividers, and/or phase shifters **212**, **213**, and signal transceivers **214**. The cylindrical arrays **200** are operatively coupled to a first stage of beam forming networks **202**, which are operatively coupled to a second stage of beam forming networks **204** using one or more attenuators **210**, hybrid couplers, power dividers, and/or phase shifters **212**. The second stage of beam forming networks **204** are operatively coupled to one or more third stage beam forming network **206**, which is operatively coupled to one or more fourth stage beam forming network **208** using one or more of the attenuators **211**, hybrid couplers, power dividers, and/or phase shifters **213**. One or more fourth stage beam forming network **208** is operatively coupled to one or more of the signal transceivers **214**. In this embodiment, amplitude tapering for sidelobe reduction is performed using the attenuators **210**, **211**, hybrid couplers, power dividers, and/or phase shifters **212**, **213** in azimuth and/or elevation. Crossings of beams in azimuth and/or elevation may also be configured using the attenuators **210**, **211**, hybrid couplers, power dividers, and/or phase shifters **212**, **213**, which also control the beam widths, at any level, such as for example -10 dB.

(47) FIG. 19 shows a plurality of conformal arrays **214**, a plurality of beam forming network stages **216**, **218**, **220**, **222**, which are connected using attenuators **224**, **225**, hybrid couplers, power dividers, and/or phase shifters **226**, **227**, and a signal transceivers **228**. The conformal arrays **214** are operatively coupled to a first stage of beam forming networks **216**, which are operatively coupled to a second stage of beam forming networks **218** using one or more attenuators **224**, hybrid couplers, power dividers, and/or phase shifters **226**. The second stage of beam forming networks **218** are operatively coupled to one or more third stage beam forming network **220**, which is operatively coupled to one or more fourth stage beam forming network **222** using one or more of the attenuators **225**, hybrid couplers, power dividers, and/or phase shifters **227**. One or more fourth stage beam forming network **222** is operatively coupled to one or more of the signal transceivers

228. In this embodiment, amplitude tapering for sidelobe reduction using the attenuators 224, 235, hybrid couplers, power dividers, and/or phase shifters 226, 227 is performed in both azimuth and/or elevation. Crossings of beams in azimuth and/or elevation may also be configured using the attenuators 224, 225, hybrid couplers, power dividers, and/or phase shifters 226, 227, which also control the beam widths, at any level, such as for example -10 dB.

(48) In accordance with one embodiment, an antenna array system that provides simultaneous transmission and/or reception with up to 360° coverage is disclosed, which includes Butler matrix beam forming networks connected together to an antenna array, which includes narrow and/or broadband elements, and multiple transmitters, receivers, or transceivers to allow for 360° transmission and/or reception. The antenna array system provides multiple beams, such as without limitation 8 or 16 beams, which can vary in beam crossing and/or overlap to provide simultaneous coverage of up to 360°.

(49) In accordance with another embodiment, an antenna array system is provided, which includes a plurality of antenna elements configured in an array, a first Butler matrix operatively coupled to the plurality of antenna elements, and a second Butler matrix operatively coupled to the first Butler matrix.

(50) The first Butler matrix may include a plurality of output ports and a plurality of input ports. Each of the plurality of output ports associated with the first Butler matrix may be operatively coupled to each of the plurality of antenna elements, and each of the plurality of input ports associated with the first Butler matrix may be coupled to each of a plurality of output ports associated with the second Butler matrix. The second Butler matrix may include a plurality of output ports and a plurality of input ports. Each of the plurality of output ports associated with the second Butler matrix may be operatively coupled to each of a plurality of input ports associated with the first Butler matrix, and each of the plurality of input ports associated with the second Butler matrix may be coupled to a transceiver. The antenna array system may include a switch, which can have one or multiple outputs and inputs. The second Butler matrix may include a plurality of output ports and a plurality of input ports. Each of the plurality of output ports associated with the second Butler matrix may be operatively coupled to each of a plurality of input ports associated with the first Butler matrix, each of the plurality of input ports associated with the second Butler matrix may be coupled to the output of the switch, and the input of switch may be coupled to a transceiver. The plurality of antenna elements may be configured to provide 360° coverage in response to the switch being swept through a plurality of positions. At least one of the plurality of antenna elements may include at least one of a bow tie antenna, log periodic antenna, and Vivaldi antenna. The plurality of antenna elements may be configured in at least one of a circle, cylinder semi-circle, arc, line, sphere, and/or any conformal shaped array.

(51) In accordance with another embodiment, a method of providing simultaneous 360° coverage is provided, which includes configuring a plurality of antenna elements in an array, coupling a first Butler matrix operatively to the plurality of antenna elements, and coupling a second Butler matrix operatively to the first Butler matrix.

(52) The method may also include coupling each of a plurality of output ports associated with the first Butler matrix operatively to each of the plurality of antenna elements, and coupling each of a plurality of input ports associated with the first Butler matrix to each of a plurality of output ports associated with the second Butler matrix. The method may include coupling each of a plurality of output ports associated with the second Butler matrix operatively to each of a plurality of input ports associated with the first Butler matrix, and coupling each of a plurality of input ports associated with the second Butler matrix to a transceiver. The method may include coupling each of a plurality of output ports associated with the second Butler matrix operatively to each of a plurality of input ports associated with the first Butler matrix, coupling each of a plurality of input ports associated with the second Butler matrix to the output of a switch, and coupling the input of switch operatively to a transceiver. The method may include configuring the plurality of antenna

elements to provide 360° coverage in response to the switch being swept through a plurality of positions. At least one of the plurality of antenna elements may include at least one of a bow tie antenna, log periodic antenna, and Vivaldi antenna. The method, configuring the plurality of antenna elements as at least one of a circle, semi-circle, arc, line, sphere, and/or any conformal shape.

(53) In accordance with another embodiment, an antenna array system is provided, which includes a plurality of antenna elements configured in an array, a first beam forming network operatively coupled to the plurality of antenna elements, a second beam forming network operatively coupled to the first beam forming network, and a switch. The switch includes an output and an input, and the second beam forming network includes a plurality of output ports and a plurality of input ports. Each of the plurality of output ports associated with the second beam forming network is operatively coupled to one of a plurality of input ports associated with the first beam forming network. The switch sequentially couples each of the plurality of input ports associated with the second beam forming network to a signal from a transceiver by sweeping the switch through a plurality of positions, thereby enabling the antenna to provide sequential 360° coverage.

(54) The first beam forming network may be a $K \times N$ beam forming network, in which K is greater than or equal to N . The second beam forming network may be an $N \times M$ beam forming network, in which M is less than or equal to N . At least one of the first and second beam forming networks may include at least one of a Butler matrix, Blass matrix, Nolen matrix, Shelton matrix, and/or Davis matrix.

(55) In accordance with another embodiment, a method of providing simultaneous 360° coverage using a multi-beam antenna array is provided, which includes configuring a plurality of antenna elements in an array, coupling a first beam forming network operatively to the plurality of antenna elements, coupling a second beam forming network operatively to the first beam forming network, coupling each of a plurality of output ports associated with the second beam forming network operatively to one of a plurality of input ports associated with the first beam forming network, coupling sequentially each of a plurality of input ports associated with the second beam forming network to a signal from a transceiver by sweeping a switch through a plurality of positions, thereby enabling the antenna to provide sequential 360° coverage.

(56) The first beam forming network may be a $K \times N$ beam forming network, in which K is greater than or equal to N . The second beam forming network may be a $N \times M$ beam forming network, in which M is less than or equal to N . At least one of the first and second beam forming networks may include at least one of a Butler matrix, Blass matrix, Nolen matrix, Shelton matrix, McFarland matrix, and/or Davis matrix.

(57) Although the specification describes components and functions implemented in the embodiments with reference to particular standards and protocols, the embodiment are not limited to such standards and protocols. It is to be understood that the various references throughout this disclosure made to input and output ports are not intended as a limitation on the direction of energy passing through these ports since, by the Reciprocity Theorem, energy is able to pass in either direction. Rather these references are merely intended as a convenient method of referring to various portions of the disclosed embodiments.

(58) The illustrations of embodiments described herein are intended to provide a general understanding of the structure of various embodiments, and are not intended to serve as a complete description of all the elements and features of apparatus and systems that might make use of the structures described herein. Many other embodiments will be apparent to those of skill in the art upon reviewing the above description. Other embodiments are utilized and derived therefrom, such that structural and logical substitutions and changes are made without departing from the scope of this disclosure. Figures are also merely representational and are not drawn to scale. Certain proportions thereof are exaggerated, while others are decreased. Accordingly, the specification and drawings are to be regarded in an illustrative rather than a restrictive sense.

(59) Such embodiments of the inventive subject matter are referred to herein, individually and/or collectively, by the term “embodiment” merely for convenience and without intending to limit the scope of this application to any single embodiment or inventive concept. Thus, although specific embodiments have been illustrated and described herein, it should be appreciated that any arrangement calculated to achieve the same purpose may be substituted for the specific embodiments shown. This disclosure is intended to cover any and all adaptations or variations of various embodiments. Combinations of the above embodiments, and other embodiments not specifically described herein, will be apparent to those of skill in the art upon reviewing the above description.

(60) In the foregoing description of the embodiments, various features are grouped together in a single embodiment for the purpose of streamlining the disclosure. This method of disclosure is not to be interpreted as reflecting that the claimed embodiments have more features than are expressly recited in each claim. Rather, as the following claims reflect, inventive subject matter lies in less than all features of a single embodiment. Thus, the following claims are hereby incorporated into the detailed description, with each claim standing on its own as a separate example embodiment.

(61) The abstract is provided to comply with 37 C.F.R. § 1.72(b), which requires an abstract that will allow the reader to quickly ascertain the nature of the technical disclosure. It is submitted with the understanding that it will not be used to interpret or limit the scope or meaning of the claims. In addition, in the foregoing Detailed Description, it can be seen that various features are grouped together in a single embodiment for the purpose of streamlining the disclosure. This method of disclosure is not to be interpreted as reflecting an intention that the claimed embodiments require more features than are expressly recited in each claim. Rather, as the following claims reflect, inventive subject matter lies in less than all features of a single embodiment. Thus, the following claims are hereby incorporated into the Detailed Description, with each claim standing on its own as separately claimed subject matter.

(62) Although specific example embodiments have been described, it will be evident that various modifications and changes are made to these embodiments without departing from the broader scope of the inventive subject matter described herein. Accordingly, the specification and drawings are to be regarded in an illustrative rather than a restrictive sense. The accompanying drawings that form a part hereof, show by way of illustration, and without limitation, specific embodiments in which the subject matter are practiced. The embodiments illustrated are described in sufficient detail to enable those skilled in the art to practice the teachings herein. Other embodiments are utilized and derived therefrom, such that structural and logical substitutions and changes are made without departing from the scope of this disclosure. This Detailed Description, therefore, is not to be taken in a limiting sense, and the scope of various embodiments is defined only by the appended claims, along with the full range of equivalents to which such claims are entitled.

(63) Given the teachings of the invention provided herein, one of ordinary skill in the art will be able to contemplate other implementations and applications of the techniques of the invention. Although illustrative embodiments of the invention have been described herein with reference to the accompanying drawings, it is to be understood that the invention is not limited to those precise embodiments, and that various other changes and modifications are made therein by one skilled in the art without departing from the scope of the appended claims.

Claims

1. A beam forming network system, which comprises: a first beam forming network comprising a plurality of first ports and a plurality of second ports, each of the plurality of first ports configured to be operatively coupled to one of a plurality of antenna elements associated with a circular array antenna; a second beam forming network comprising a plurality of third ports and a plurality of fourth ports, each of the plurality of third ports being operatively coupled to one of the plurality of

second ports using a four-port hybrid coupler; and a switch being configured to perform operations comprising: sequentially coupling the plurality of fourth ports to a signal by sweeping the switch through a plurality of positions, thereby enabling the plurality of antenna elements to provide sequential 360° coverage during a first operation; simultaneously coupling the plurality of fourth ports to the signal to enable the plurality of antenna elements to provide simultaneous 360° coverage during a second operation; and selectively coupling a portion of the plurality of fourth ports to the signal to enable the plurality of antenna elements to provide a partial 360° coverage during a third operation, the beam forming network system configured to perform amplitude tapering to cause sidelobe reduction in at least one of azimuth or elevation using the four-port hybrid coupler, a quantity of the plurality of fourth ports being less than a quantity of the plurality of first ports, the beam forming network system configured to enable selection of beam crossing width causing a signal-to-noise ratio to be increased.

2. The beam forming network system, as defined by claim 1, wherein the first beam forming network is a $K \times N$ beam forming network, K being greater than or equal to N .
3. The beam forming network system, as defined by claim 1, wherein the second beam forming network is an $N \times M$ beam forming network, M being less than or equal to N .
4. The beam forming network system, as defined by claim 1, wherein at least one of the first beam forming network and the second beam forming network comprises at least one of a Butler matrix, a Blass matrix, a Nolen matrix, a Shelton matrix, a McFarland matrix, or a Davis matrix.
5. A method of beam forming, the method comprising: coupling each of a plurality of first ports associated with a first beam forming network operatively to one of a plurality of antenna elements associated with a circular array antenna; coupling each of a plurality of third ports associated with a second beam forming network operatively to one of a plurality of second ports associated with the first beam forming network using a four-port hybrid coupler; coupling a plurality of fourth ports associated with the second beam forming network sequentially to a signal by sweeping a switch through a plurality of positions, thereby enabling the antenna elements to provide sequential 360° coverage during a first operation; coupling the plurality of fourth ports simultaneously to the signal to enable the plurality of antenna elements to provide simultaneous 360° coverage during a second operation; selectively coupling a portion of the plurality of fourth ports to the signal to enable the plurality of antenna elements to provide a partial 360° coverage during a third operation, a quantity of the plurality of fourth ports being less than a quantity of the plurality of first ports; performing amplitude tapering to cause sidelobe reduction in at least one of azimuth or elevation using the four-port hybrid coupler; and enabling selection of beam crossing width causing a signal-to-noise ratio to be increased.
6. The method of beam forming, as defined by claim 5, wherein the first beam forming network is a $K \times N$ beam forming network, K being greater than or equal to N .
7. The method of beam forming, as defined by claim 5, wherein the second beam forming network is an $N \times M$ beam forming network, M being less than or equal to N .
8. The method of beam forming, as defined by claim 5, wherein at least one of the first beam forming network and the second beam forming network comprises at least one of a Butler matrix, a Blass matrix, a Nolen matrix, a Shelton matrix, a McFarland matrix, or Davis matrix.
9. A beam forming network system, which comprises: at least one first beam forming network comprising a plurality of first ports and a plurality of second ports, each of the plurality of first ports configured to be operatively coupled to one of a plurality of antenna elements associated with a circular array antenna; at least one second beam forming network comprising a plurality of third ports and a plurality of fourth ports, each of the plurality of third ports being operatively coupled to one of the plurality of second ports using a first four-port hybrid coupler; a switch being configured to perform operations comprising: sequentially coupling the plurality of fourth ports to a signal by sweeping the switch through a plurality of positions, thereby enabling the plurality of antenna elements to provide sequential 360° coverage during a first operation; simultaneously coupling the

plurality of fourth ports to the signal to provide simultaneous 360° coverage during a second operation; and selectively coupling a portion of the plurality of fourth ports to the signal to provide a partial 360° coverage during a third operation, the beam forming network system configured to perform amplitude tapering to cause sidelobe reduction in at least one of azimuth or elevation using the first four-port hybrid coupler, a quantity of the plurality of fourth ports being less than a quantity of the plurality of first ports, the beam forming network system configured to enable selection of beam crossing width causing a signal-to-noise ratio to be increased; and at least one third beam forming network comprising a plurality of fifth ports and a plurality of sixth ports, the plurality of fifth ports configured to be operatively coupled to a one of the plurality of fourth ports.

10. The beam forming network system, as defined by claim 9, wherein the first beam forming network is an $MN \times MN$ beam forming network, N being an integer greater than or equal to one (1), M being an integer greater than or equal to one (1).

11. The beam forming network system, as defined by claim 9, wherein the second beam forming network is an $N \times N$ beam forming network, N being an integer greater than or equal to one (1).

12. The beam forming network system, as defined by claim 9, wherein the first beam forming network is an $N \times (N+M)$ beam forming network, N being an integer greater than or equal to one (1), M being an integer greater than or equal to one (1).

13. The beam forming network system, as defined by claim 9, wherein at least one of the first beam forming network and the second beam forming network comprises at least one of a Butler matrix, a Blass matrix, a Nolen matrix, a Shelton matrix, a McFarland matrix, or a Davis matrix.

14. The beam forming network system, as defined by claim 9, wherein the first four-port hybrid coupler comprises at least one of a 90 degree hybrid coupler or a 180 degree hybrid coupler.

15. The beam forming network system, as defined by claim 9, wherein at least one of amplitude or phase is controlled to cause sidelobe reduction in at least one of azimuth or elevation using at least one of a first variable phase shifter, a first fixed phase shifter, a first attenuator, a first power divider, or the first four-port hybrid coupler.

16. The beam forming network system, as defined by claim 9, wherein the first beam forming network is an $N \times N$ beam forming network, N being an integer greater than or equal to one (1).

17. The beam forming network system, as defined by claim 9, wherein each of the plurality of fourth ports is configured to be operatively coupled to one of a plurality of transceivers, each of the plurality of transceivers operatively coupling one of the plurality of fourth ports to the signal.

18. The beam forming network system, as defined by claim 9, wherein the plurality of antenna elements is configured in at least one of a circle, a cylinder, a semi-circle, an arc, a line, a sphere, a conformal shape, or a curvilinear shape.

19. The beam forming network system, as defined by claim 9, further comprising at least one fourth beam forming network comprising a plurality of seventh ports and a plurality of eighth ports, each of the plurality of seventh ports being operatively coupled to one of the plurality of sixth ports using at least one of a second variable phase shifter, a second fixed phase shifter, a second attenuator, a second power divider, or a second four-port hybrid coupler.

20. The beam forming network system, as defined by claim 19, wherein the second four-port hybrid coupler comprises at least one of a 90 degree hybrid coupler or a 180 degree hybrid coupler.

21. The beam forming network system, as defined by claim 19, wherein at least one of amplitude or phase is controlled to cause sidelobe reduction in at least one of azimuth or elevation using at least one of the second variable phase shifter, the second fixed phase shifter, the second attenuator, the second power divider, or the second four-port hybrid coupler.

22. The beam forming network system, as defined by claim 19, wherein each of the plurality of eighth ports is configured to be operatively coupled to the switch, the switch operatively coupling each of the plurality of eighth ports to the signal by sweeping the switch through a plurality of positions.

23. The beam forming network system, as defined by claim 19, wherein each of the plurality of

eight ports is configured to be operatively coupled to one of a plurality of transceivers, each of the plurality of transceivers operatively coupling one of the plurality of eighth ports to the signal.

24. The beam forming network system, as defined by claim 9, wherein the second beam forming network comprises a power divider.

25. A method of beam forming, which comprises: coupling each of a plurality of first ports associated with at least one first beam forming network operatively to one of a plurality of antenna elements associated with a circular array antenna; and coupling each of a plurality of third ports associated with at least one second beam forming network operatively to one of a plurality of second ports associated with the first beam forming network using a first four-port hybrid coupler, the second beam forming network comprising a plurality of fourth ports, a quantity of the fourth ports being less than a quantity of the first ports; sequentially coupling the plurality of fourth ports to a signal to enable the plurality of antenna elements to provide sequential 360° coverage during a first operation; simultaneously coupling the plurality of fourth ports to the signal to enable the plurality of antenna elements to provide simultaneous 360° coverage during a second operation; and selectively coupling a portion of the plurality of fourth ports to the signal to enable the plurality of antenna elements to provide a partial 360° coverage during a third operation; performing amplitude tapering to cause sidelobe reduction in either azimuth or elevation using the first four-port hybrid coupler; and enabling selection of beam crossing width to cause a signal-to-noise ratio to be increased.

26. The method of beam forming, as defined by claim 25, wherein the first beam forming network is an $MN \times MN$ beam forming network, N being an integer greater than or equal to one (1), M being an integer greater than or equal to one (1).

27. The method of beam forming, as defined by claim 25, wherein the second beam forming network is an $N \times N$ beam forming network, N being an integer greater than or equal to one (1).

28. The method of beam forming, as defined by claim 25, wherein the first beam forming network is an $N \times (N+M)$ beam forming network, N being an integer greater than or equal to one (1), M being an integer greater than or equal to one (1).

29. The method of beam forming, as defined by claim 25, wherein at least one of the first beam forming network or the second beam forming network comprises at least one of a Butler matrix, a Blass matrix, a Nolen matrix, a Shelton matrix, a McFarland matrix, or a Davis matrix.

30. The method of beam forming, as defined by claim 25, wherein the first four-port hybrid coupler comprises at least one of a 90 degree hybrid coupler or a 180 degree hybrid coupler.

31. The method of beam forming, as defined by claim 25, further comprising controlling at least one of amplitude or phase to cause sidelobe reduction in at least one of azimuth or elevation using at least one of a first variable phase shifter, a first fixed phase shifter, a first attenuator, a first power divider, or the first four-port hybrid coupler.

32. The method of beam forming, as defined by claim 25, wherein the first beam forming network is an $N \times N$ beam forming network, N being an integer greater than or equal to one (1).

33. The method of beam forming, as defined by claim 25, further comprising coupling each of the plurality of fourth ports associated with the second beam forming network operatively to a switch, the switch selectively coupling each of the plurality of fourth ports to the signal by sweeping the switch through a plurality of positions.

34. The method of beam forming, as defined by claim 25, further comprising coupling each of the plurality of fourth ports associated with the second beam forming network operatively to one of a plurality of transceivers, each of the plurality of transceivers operatively coupling one of the plurality of fourth ports to the signal.

35. The method of beam forming, as defined by claim 25, wherein the plurality of antenna elements is configured in at least one of a circle, cylinder, a semi-circle, an arc, a line, a sphere, a conformal shape, or curvilinear shape.

36. The method of beam forming, as defined by claim 25, further comprising coupling a plurality

of fifth ports associated with at least one third beam forming network operatively to one of the plurality of fourth ports associated with the second beam forming network.

37. The method of beam forming, as defined by claim 36, further comprising coupling each of a plurality of seventh ports associated with at least one fourth beam forming network operatively to one of a plurality of sixth ports associated with the at least one third beam forming network using at least one of a second variable phase shifter, a second fixed phase shifter, a second attenuator, a second power divider, or a second four-port hybrid coupler.

38. The method of beam forming, as defined by claim 37, wherein the second four-port hybrid coupler comprises at least one of a 90 degree hybrid coupler or a 180 degree hybrid coupler.

39. The method of beam forming, as defined by claim 37, further comprising controlling at least one of amplitude or phase to cause sidelobe reduction in at least one of azimuth or elevation using at least one of the second variable phase shifter, the second fixed phase shifter, the second attenuator, the second power divider, or the second four-port hybrid coupler.

40. The method of beam forming, as defined by claim 37, further comprising coupling a plurality of eighth ports associated with the at least one fourth beam forming network operatively to a switch, the switch operatively coupling each of the plurality of eighth ports to the signal by sweeping the switch through a plurality of positions.

41. The method of beam forming, as defined by claim 37, further comprising coupling a plurality of eighth ports associated with the at least one fourth beam forming network operatively to one of a plurality of transceivers, each of the plurality of transceivers operatively coupling one of the plurality of eighth ports to the signal.

42. The method of beam forming, as defined by claim 25, wherein the second beam forming network comprises a power divider.
