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United States Patent Application Publication

20250267033

Kind Code

A1

Publication Date

August 21, 2025

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INTERFACE MODULE FOR A SUBSCRIBER STATION OF A SERIAL BUS SYSTEM AND METHOD FOR TRANSMITTING A MESSAGE IN A SERIAL BUS SYSTEM

Abstract

An interface module for a subscriber station of a serial bus system. A transmitting module transmits a digital transmission signal as an analog differential signal onto a bus of the bus system to transmit a message to at least one other subscriber station of the bus system. Bits in the digital transmission signal have a longer bit duration in a first communication phase than in a second. The interface module has a time measuring block for measuring, with a predetermined timing clock, a predetermined time in the digital transmission signal, which is designed to stop its measurement if the measured value of the time measuring block has reached a predetermined limit value, and an evaluation block for evaluating the measured value of the time measuring block to set the transmitting module for operation in the first communication phase or in the second communication phase based on the evaluation result.

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Family ID: 1000008629570

Appl. No.: 18/857682

Filed (or PCT Filed): June 26, 2023

PCT No.: PCT/EP2023/067277

Foreign Application Priority Data

DE 10 2022 208 457.8

Aug. 15, 2022

Publication Classification

Int. Cl.: H04L12/40 (20060101); H04L25/02 (20060101); H04L25/49 (20060101)

U.S. Cl.:

CPC H04L12/40013 (20130101); H04L12/4013 (20130101); H04L25/0272 (20130101); H04L25/4902 (20130101); H04L2012/40215 (20130101)

Background/Summary

FIELD

[0001] The present invention relates to an interface module for a subscriber station of a serial bus system and to a method for transmitting a message in a serial bus system, which is in particular a CAN XL bus system.

BACKGROUND INFORMATION

[0002] Serial bus systems, in particular a CAN bus system, are used for message or data transmission in technical systems. For example, a serial bus system can make communication possible between sensors and control units in a vehicle or a technical production plant, etc.

[0003] In a CAN bus system, messages are transmitted using the CAN and/or CAN FD protocol, as described in the standard ISO-11898-1:2015 as CAN protocol specification with CAN FD. With CAN FD, transmission on the bus switches back and forth between a slow operating mode in a first communication phase (arbitration phase) and a fast operating mode in a second communication phase (data phase). Most manufacturers use CAN FD with a 500 kbit/s arbitration bit rate and a 2 Mbit/s data bit rate in the vehicle.

[0004] Successor bus systems for CAN FD, such as CAN SiC and CAN XL, are compatible with CAN FD and are designed for even higher data rates in the second communication phase. With CAN SiC according to the CiA601-4 standard of the CAN in Automation (CiA) organization, a data rate of about 5 to 8 Mbit/s can be achieved in the second communication phase. In the case of CAN XL according to the CiA601-3 standard, a data rate of >10 Mbit/s is required in the second communication phase. In addition to pure data transport via the CAN bus, CAN XL is also intended to support other functions, such as functional safety, data security and quality of service (QOS). These are elementary properties which are required, for example, in an autonomously driving vehicle.

[0005] In the CAN XL, CAN FD and CAN SiC bus systems, in the second communication phase the data are thus sent to the bus at a higher data rate than in the first communication phase. In CAN XL, a bit of a transmission signal in the second communication phase not only has a shorter bit duration or bit time or temporal length than in the first communication phase, but is usually also sent with a different physical layer to the bus and received with a different receiving threshold than in the first communication phase. Consequently, with CAN XL, the bus levels of the bus signals CAN_H, CAN_L for the first communication phase can differ from the bus levels of the second communication phase. In CAN XL, the type of communication in the second communication phase is also called FAST MODE. The physical layer corresponds to the bit transmission layer or layer 1 of the conventional OSI model (Open Systems Interconnection Model).

[0006] Communication on the bus that is as trouble-free as possible can therefore only take place when CAN XL subscriber stations of the bus system use a transmitting/receiving device that detects and carries out the switching between the two communication phases in a message to be sent to or received from the bus as error-free as possible and converts the level of a transmission

signal correctly for the bus. The transmitting/receiving device can also be referred to as a CAN transceiver or CAN FD transceiver, etc.

[0007] For this reason, a transmitting/receiving device (transceiver) is required for CAN XL, which ensures communication that is as error-free as possible for all operating phases of communication on the bus.

SUMMARY

[0008] It is an object of the present invention to provide an interface module for a subscriber station for a serial bus system and a method for transmitting a message in a serial bus system which solve the aforementioned problems. In particular, an interface module for a subscriber station of a serial bus system and a method for transmitting a message in a serial bus system should make possible a reliable and uncomplicated creation/generation of bus signals, even when the physical layer is switched between two communication phases during communication on the bus.

[0009] The object may be achieved by an interface module for a subscriber station of a serial bus system having certain features of the present invention. In the bus system, a transmitting module is designed to transmit a digital transmission signal as an analog differential signal to a bus of the bus system in order to transmit a message to at least one other subscriber station of the bus system, wherein bits in the digital transmission signal have a longer bit duration in a first communication phase than in a second communication phase of the transmission signal. According to an example embodiment of the present invention, the interface module has at least one time measuring block for measuring, with a predetermined timing, a predetermined time in the digital transmission signal, wherein the at least one time measuring block is designed to stop its measurement when the measured value of the at least one time measuring block has reached a predetermined limit value, and an evaluation block for evaluating the measured value of the at least one time measuring block at the end of the predetermined time in order to set the transmitting module for operation in the first communication phase or for operation in the second communication phase on the basis of the evaluation result.

[0010] The described interface module of the present invention is designed in such a way that a reliable, precise and nevertheless uncomplicated creation of a transmission signal for bus signals takes place during operation of the bus system. The transmission signal of the subscriber station can indicate how the operating mode for the transmitting module is to be preselected or set in order to correctly switch within a message between two communication phases for communication on the bus.

[0011] In addition, the described interface module of the present invention is designed in such a way that the bus signals are reliably and readily detected during operation of the bus system. All this applies in particular to such communication in which the physical layer is switched between two communication phases for communication on the bus.

[0012] The described interface module of the present invention may ensure very low-error and therefore largely interference-free communication, in particular in accordance with the CiA610-3 standard of CAN XL, between subscriber stations of the bus system. Additional advantages are explained in more detail in the description of the figures.

[0013] Furthermore, the described interface module of the present invention also enables the functionality for using different receiving thresholds for arbitration and data phase. Not only is communication in the bus system between other subscriber stations realized at higher bit rates but also the transmittable bit rate is not reduced by errors in the communication.

[0014] Advantageous further embodiments of the interface module are disclosed herein.

[0015] According to an example embodiment of the present invention, the evaluation block may be designed to switch the transmitting module to a first operating mode for the first communication phase for transmitting on the basis of the evaluation result or to switch the transmitting module to a second operating mode or third operating mode for the second communication phase for transmitting.

[0016] In one embodiment variant of the present invention, the at least one time measuring block can have at least four time measuring blocks, wherein the evaluation block is designed to evaluate the measured value of the at least four time measuring blocks at the end of the predetermined time in order to decode a pulse-width-modulated symbol of the transmission signal by evaluating the measured values of the at least four time measuring blocks in order to output a pulse-width-demodulated transmission signal to the transmitting module.

[0017] The interface module described above can also have a switching block for outputting an oscillating signal for specifying the predetermined timing clock with which all time measuring blocks intermittently carry out their measurement.

[0018] In one embodiment variant of the present invention, a first time measuring block is designed to restart the measurement with each edge of the transmission signal and to stop the measurement when the measured value of the first time measuring block has reached a predetermined limit value, wherein the evaluation block is designed to preselect the second operating mode for the transmitting module for the second communication phase when the measured value of the first time measuring block corresponds to a first predetermined time.

[0019] In this case, according to an example embodiment of the present invention, a second time measuring block can be designed to start the measurement with each rising edge of the transmission signal and to measure until the next rising edge of the transmission signal until the predetermined limit value is exceeded, wherein a third time measuring block is designed to start the measurement with each falling edge of the transmission signal and to measure until the next falling edge of the transmission signal until the predetermined limit value is exceeded, and wherein the evaluation block is designed to switch the transmitting module for the second communication phase to the second operating mode if the evaluation of the measurement result of the second and third time measuring blocks shows that in the transmission signal the distance between a rising edge and a falling edge or the distance between a falling edge and a rising edge is equal to a second predetermined time, which is the bit duration of bits of the second communication phase, and which is shorter than the first predetermined time.

[0020] In addition, according to an example embodiment of the present invention, a fourth time measuring block can be designed to start the measurement with each rising edge of the transmission signal and to measure until the next falling edge of the transmission signal until the predetermined limit value is exceeded, wherein a fifth time measuring block is designed to start the measurement with each falling edge of the transmission signal and to measure until the next rising edge of the transmission signal until the predetermined limit value is exceeded, and wherein the evaluation block is designed to compare the measurement of the second and fourth time measuring blocks with each other and to compare the measurement of the third and fifth time measuring blocks with each other, and to output a pulse-width-demodulated transmission signal to the transmitting module on the basis of the comparison results.

[0021] According to one exemplary embodiment of the present invention, the at least one time measuring block is a counter. In this case, at least two time measuring blocks of the at least one time measuring block can be designed to stop counting at a count value which corresponds to the symbol time period for a PWM symbol of a pulse-width-modulated signal, wherein the evaluation block is designed to evaluate the value of the PWM symbol of the pulse-width-modulated signal at the end of a PWM symbol using the at least one time measuring block and to output a level for the pulse-width-demodulated transmission signal to the transmitting module using the evaluation result.

[0022] According to one exemplary embodiment of the present invention, the at least one time measuring block is designed as a capacitor with a voltage measuring device for measuring the voltage at the capacitor, wherein at least one switchable constant-current source is provided as a switching block which is designed to switch the at least one constant-current source in order to charge the capacitor over time, and wherein the at least one time measuring block is designed to

stop the charging of the at least one capacitor at a predetermined limit voltage value which corresponds to the predetermined time. The evaluation block can be designed to evaluate the voltage value of at least one time measuring block.

[0023] The interface module of the present invention described above can be part of a transmitting/receiving device which also has a transmitting module for transmitting a digital transmission signal as an analog differential signal to a bus of the bus system in order to transmit a message to at least one other subscriber station of the bus system, and also has a receiving module for receiving signals from the bus and for generating a digital reception signal from the analog differential signal.

[0024] The transmitting module of the present invention can be designed to generate the analog differential signals in the first communication phase of the message with a different physical layer than in the second communication phase.

[0025] The transmitting/receiving device described above can be part of a subscriber station for a serial bus system. The subscriber station can also be a communication control device for controlling communication in the bus system and for generating a digital transmission signal as the basis for the digital pulse-width-modulated transmission signal for the interface module.

[0026] Optionally, the subscriber station is designed for communication in a bus system in which an exclusive, collision-free access of a subscriber station to the bus of the bus system is guaranteed at least temporarily.

[0027] The aforementioned object may further be achieved by a method for transmitting a message in a serial bus system having features of the present invention. According to an example embodiment of the present invention, the method is carried out using an interface module for a subscriber station of a serial bus system, wherein the subscriber station has a transmitting module and wherein the interface module has at least one time measuring block and an evaluation block, and wherein the method comprises the steps of measuring, using the at least one time measuring block and using a predetermined timing clock, a predetermined time in the digital transmission signal, wherein the at least one time measuring block stops its measurement when the at least one time measuring block has reached a predetermined limit value; evaluating, using the evaluation block, the measurement result of the at least one time measuring block at the end of the predetermined time; setting the transmitting module for operation in the first communication phase or for operation in the second communication phase on the basis of the evaluation result, and transmitting, using the transmitting module, a digital, pulse-width-demodulated transmission signal as an analog differential signal onto the bus of the bus system in order to transmit a message to at least one other subscriber station of the bus system, wherein bits in the digital, pulse-width-demodulated transmission signal have a longer bit duration in a first communication phase than in a second communication phase of the transmission signal.

[0028] The method of the present invention offers the same advantages as those mentioned above in relation to the interface module of the present invention.

[0029] Further possible implementations of the present invention also include combinations, even those not explicitly mentioned, of features or embodiments described above or below with respect to the exemplary embodiments. In this case, a person skilled in the art will also add individual aspects as improvements or additions to the relevant basic form of the present invention.

Description

BRIEF DESCRIPTION OF THE DRAWINGS

[0030] The present invention is described in more detail below with reference to the figures and on the basis of exemplary embodiments.

[0031] FIG. 1 shows a simplified block diagram of a bus system according to a first exemplary

embodiment of the present invention.

[0032] FIG. 2 shows a diagram for illustrating the structure of a message that can be transmitted by a subscriber station of the bus system according to the first exemplary embodiment of the present invention.

[0033] FIG. 3 shows an example of the ideal time sequence of bus signals CAN_H, CAN_L, which are sent to a bus of the bus system by subscriber stations of the bus system for the message in FIG. 2.

[0034] FIG. 4 shows the time sequence of a differential voltage VDIFF which forms on the bus of the bus system as a result of the bus signals in FIG. 3.

[0035] FIG. 5 shows a block diagram of transmitting/receiving device for a subscriber station of the bus system according to the first exemplary embodiment of the present invention.

[0036] FIG. 6 shows an example of a time sequence of a digital transmission signal which in the data phase of a message indicates to the transmitting/receiving device for a transmitting node the switching to an operating mode for transmitting.

[0037] FIG. 7 shows the temporal course of a digital transmission signal which the transmitting/receiving device generates from the signal in FIG. 6 in order to transmit the message to the bus.

[0038] FIG. 8 shows an example of a temporal course of a digital transmission signal which in the data phase of a message indicates to the transmitting/receiving device for a receiving node the switching to an operating mode for receiving.

[0039] FIG. 9 shows the temporal course of a digital transmission signal which the transmitting/receiving device generates from the signal in FIG. 8.

[0040] FIG. 10 to FIG. 15 show signal courses to explain the function of an interface module of the transmitting/receiving device for the case in which the associated subscriber station is the transmitter of the message to the bus in the data phase (transmitting node) and for which reason the transmitting/receiving device is switched in the data phase to an operating mode for transmitting the transmission signal to the bus.

[0041] FIG. 16 to FIG. 21 show signal waveforms to explain the function of the interface module of the transmitting/receiving device for the case in which the associated subscriber station in the data phase is only a receiver, but not a transmitter, of the message on the bus (receiving node) and for which reason the transmitting/receiving device is switched in the data phase to an operating mode for receiving the message from the bus.

[0042] In the figures, identical or functionally identical elements are given the same reference signs unless otherwise indicated.

DETAILED DESCRIPTION OF EXAMPLE EMBODIMENTS

[0043] FIG. 1 shows a bus system 1, which can, for example, at least in sections, be a CAN bus system, a CAN-FD bus system, etc. The bus system 1 can be used in a vehicle, in particular a motor vehicle, an aircraft, etc., or in a hospital, etc.

[0044] In FIG. 1, the bus system 1 has a plurality of subscriber stations 10, 20, 30, which are each connected to a bus 40 with a first bus wire 41 and a second bus wire 42. The bus wires 41, 42 can also be referred to as CAN_H and CAN_L for the signals on the bus 40. Messages 45, 46, 47 in the form of signals can be transmitted between the individual subscriber stations 10, 20, 30 via the bus 40. The subscriber stations 10, 20, 30 are, for example, control devices or display devices of a motor vehicle. As shown in FIG. 1, the subscriber stations 10, 30 each have a communication control device 11 and a transmitting/receiving device 12. The transmitting/receiving device 12 has a transmitting module 121 and a receiving module 122.

[0045] The subscriber station 20 has a communication control device 21 and a transmitting/receiving device 22. The transmitting/receiving device 22 has a transmitting module 221 and a receiving module 222.

[0046] The transmitting/receiving devices 12 of the subscriber stations 10, 30 and the

transmitting/receiving device **22** of the subscriber station **20** are in each case directly connected to the bus **40**, even if this is not shown in FIG. **1**.

[0047] The communication control devices **11**, **21** are each used for controlling a communication of the relevant subscriber station **10**, **20**, **30** via the bus **40** with at least one other subscriber station of the subscriber stations **10**, **20**, **30** which are connected to the bus **40**.

[0048] The communication control device **11** creates and reads first messages **45**, **47**, which are, for example, modified CAN messages **45**, **47**. Here the modified CAN messages **45**, **47** are based on the CAN XL format, for example. The transmitting/receiving device **12** serves for transmitting and receiving the messages **45**, **47** from the bus **40**. The transmitting module **121** receives a digital transmission signal TxD created by the communication control device **11** for one of the messages **45**, **47** and converts this into signals on the bus **40**. The digital transmission signal TxD can be a pulse-width-modulated signal, at least in sections. The receiving module **122** receives signals transmitted on the bus **40** corresponding to the messages **45** to **47** and from these generates a digital reception signal RxD. The receiving module **122** sends the reception signal RxD to the communication control device **11**. In addition, the communication control device **11** can be designed to create and read second messages **46**, which are, for example, CAN SiC messages **46**. The transmitting/receiving device **12** can be designed accordingly.

[0049] The communication control device **21** can be designed as a conventional CAN controller according to ISO 11898-1:2015, i.e. as a CAN FD-tolerant Classical CAN controller or as a CAN FD controller. The communication control device **21** creates and reads second messages **46**, for example CAN SiC messages. The transmitting/receiving device **22** serves for transmitting and receiving the messages **46** from the bus **40**. The transmitting module **221** receives a digital transmission signal TxD created by the communication control device **21** and converts this into signals for a message **46** on the bus **40**. The receiving module **222** receives signals transmitted on the bus **40** corresponding to the messages **45** to **47** and from these generates a digital reception signal RxD. The transmitting/receiving device **22** may be designed like a conventional CAN SiC transceiver.

[0050] To transmit messages **45**, **46**, **47** with CAN SiC or CAN XL, proven properties are adopted that are responsible for the robustness and user-friendliness of CAN and CAN FD, in particular the frame structure with identifier and arbitration according to the well-conventional CSMA/CR method. The CSMA/CR method has the consequence that so-called recessive states are mandatory on the bus **40**, which can be overwritten on the bus **40** with dominant levels or dominant states by other subscriber stations **10**, **20**, **30**.

[0051] With the two subscriber stations **10**, **30**, a formation and then a transmission of messages **45**, **47** with different CAN formats, in particular the CAN FD format or the CAN SiC format or the CAN XL format, as well as the reception of such messages **45**, **47**, can be realized. This is described in more detail below for a message **45**.

[0052] FIG. **2** shows for the message **45** a frame **450**, which is in particular a CAN XL frame, such as is provided by the communication control device **11** for the transmitting/receiving device **12** for transmission to the bus **40**. Here, the communication control device **11** creates the frame **450** in the present exemplary embodiment as compatible with CAN FD. Alternatively, the frame **450** is compatible with any successor standard for CAN FD.

[0053] According to FIG. **2**, for CAN communication on the bus **40** the frame **450** is divided into different communication phases **451**, **452**, namely an arbitration phase **451** (first communication phase) and a data phase **452** (second communication phase). The frame **450** has, following a start bit SOF, an arbitration field **453**, a control field **454**, a data field **455**, a checksum field **456** and a frame termination field **457**. The checksum field **456** and the frame termination field **457** form a frame end phase **456**, **457** of the frame **450**.

[0054] In the arbitration phase **451**, with the aid of an identifier (ID) in the arbitration field **453**, negotiation takes place bitwise between the subscriber stations **10**, **20**, **30** as to which subscriber

station **10**, **20**, **30** wishes to transmit the message **45**, **46** with the highest priority and will therefore receive exclusive access to the bus **40** of the bus system **1** for the next time for transmitting in the subsequent data phase **452**. A physical layer such as in CAN and CAN FD is used in the arbitration phase **451**. The physical layer corresponds to the bit transmission layer or layer 1 of the convention OSI model (Open Systems Interconnection Model).

[0055] An important point during the phase **451** is that the conventional CSMA/CR method is used, which allows simultaneous access of the subscriber stations **10**, **20**, **30** to the bus **40** without the higher-priority message **45**, **46** being destroyed. As a result, further bus subscriber stations **10**, **20**, **30** can be added relatively easily to the bus system **1**, which is very advantageous.

[0056] The CSMA/CR method has the consequence that so-called recessive states are mandatory on the bus **40**, which can be overwritten on the bus **40** with dominant levels or dominant states by other subscriber stations **10**, **20**, **30**. In the recessive state, high-impedance conditions prevail at the individual subscriber station **10**, **20**, **30**, which in combination with the parasites on the bus circuit results in longer time constants. This leads to a limitation of the maximum bit rate of the present-day CAN-FD physical layer at currently about 2 megabits per second in real vehicle use.

[0057] In the data phase **452**, in addition to a portion of the control field **454**, the payload data of the CAN XL frame **450** or of the message **45** are transmitted from the data field **455** as well as the checksum field **456**. The checksum field **456** may contain a checksum of the data of the data phase **452** including the stuff bits, which are inserted by the transmitter of the message **45** as an inverse bit after each predetermined number of identical bits, in particular **10** identical bits. At the end of the data phase **452**, the system switches back to the arbitration phase **451**.

[0058] An end field in the frame termination field **457** may contain at least one acknowledge bit. This can be followed by a sequence of **11** identical bits that indicate the end of the CAN XL frame **450**. The at least one acknowledge bit can be used to indicate whether or not a receiver has detected an error in the received CAN XL frame **450** or in the message **45**.

[0059] A transmitter of the message **45** begins to transmit bits of the data phase **452** to the bus **40** only when the subscriber station **10** as the transmitter has won the arbitration and the subscriber station **10** as transmitter thus has exclusive access to the bus **40** of the bus system **1** for transmitting.

[0060] In the arbitration phase **451** as the first communication phase the subscriber stations **10**, **30** thus use in part, in particular up to the FDF bit (inclusive), a format from CAN/CAN FD according to ISO11898-1:2015. However, compared to CAN or CAN FD, an increase in the net data transfer rate, in particular to over 10 megabits per second, is possible in the data phase **452** as the second communication phase. Increasing the size of the payload data per frame, in particular to about 2 kbyte or any other value, is also possible.

[0061] As shown in FIG. **3**, in the arbitration phase **451**, the transmitting/receiving devices **12** use a physical layer **451_P** to transmit to the bus **40** a transmission signal TxD (FIG. **1**) over time t as signals CAN_H, CAN_L. The same applies to the transmitting/receiving device **22**. In contrast, in the data phase **452** the transmitting/receiving device **12** may use a physical layer **452_P** that is different from the physical layer **451_P** in order to transmit to the bus **40** the transmission signal TxD (FIG. **1**) as signals CAN_H, CAN_L, as described above. There are two operating modes for the physical layer **452_P**, namely FAST_TX and FAST_RX, as described in more detail below. The course of the corresponding transmission signals is explained in more detail below with the aid of FIG. **6** to FIG. **21**.

[0062] FIG. **3** shows on the left side that in the arbitration phase **451** the subscriber stations **10**, **20**, **30** each transmit signals CAN_H, CAN_L over time t to the bus **40**, which have a first bit duration t_{bt1} . The signals CAN_H, CAN_L are serial signals and have alternately at least one dominant state **401** or at least one recessive state **402**. After arbitration in the arbitration phase **451**, one of the subscriber stations **10**, **20**, **30** emerges the winner.

[0063] Assuming that the first subscriber station **10** has won the arbitration, at the end of the

arbitration phase **451**, the transmitting/receiving device **12** of the subscriber station **10** switches its physical layer **451_P** from a first operating mode (SLOW), which can also be implemented as an SiC operating mode, to a second operating mode (FAST_TX), since in the data phase **452** the subscriber station **10** is the transmitter of the message **45**. As shown in FIG. 3, in the data phase **452** or in the second operating mode (FAST_TX) the transmitting module **121** then, depending on a transmission signal TxD, generates the states L0 or L1 with the physical layer **452_P** for the signals CAN_H, CAN_L on the bus **40**, one after the other and thus serially. The frequency of the signals CAN_H, CAN_L can be increased in the data phase **452**. In the example in FIG. 3, the bit time or bit duration t_{bt2} in the data phase **452** is shorter or less than the bit time or bit duration t_{bt1} in the arbitration phase **451**. In the example in FIG. 3, the net data transfer rate in the data phase **452** is thus increased compared to the arbitration phase **451**. In contrast, at the end of the arbitration phase **451** the transmitting/receiving device **12** of the subscriber station **30** switches its physical layer **451_P** from the first operating mode (SLOW or SiC) to a third operating mode (FAST_RX), since in the data phase **452** the subscriber station **30** is only a receiver, i.e. not a transmitter, of the frame **450**.

[0064] If the transmitting/receiving device **12** detects that a switchover from the data phase **452** back to the arbitration phase **451** is to be made, the transmitting/receiving device **12** will be switched from transmitting (operating mode FAST_TX) and/or receiving (operating mode FAST_RX) signals with the physical layer **452_P** to transmitting and/or receiving in the physical layer **451_P**. Accordingly, after the end of the arbitration phase **451** all transmitting/receiving devices **12** of the subscriber stations **10**, **30** switch their operating mode to the first operating mode (SLOW or SiC). All transmitting/receiving devices **12** can thus not only switch between the bit durations t_{bt1} , t_{bt2} but also switch their physical layer, as described above.

[0065] According to FIG. 4, in the arbitration phase **451** over time t in the ideal case a difference signal $V_{DIFF} = CAN_H - CAN_L$ with values of $V_{DIFF} = 2V$ for dominant states **401** and $V_{DIFF} = 0V$ for recessive states **402** is formed on the bus **40**. This is shown on the left-hand side in FIG. 4. In contrast, in the data phase **452**, a difference signal $V_{DIFF} = CAN_H - CAN_L$ corresponding to the states L0, L1 in FIG. 4 is formed over time t on the bus **40**, as shown on the right-hand side in FIG. 4. The state L0 has a value $V_{DIFF} = 1V$. The state L1 has a value $V_{DIFF} = -1V$.

[0066] The receiving module **122** can distinguish the states **401**, **402** with in each case two of the receiving thresholds T1, T2, T3, which lie in the ranges TH_T1, TH_T2, TH_T3. For this purpose, the receiving module **122** samples the signals in FIG. 3 or FIG. 4 at times t_A , as shown in FIG. 4. To evaluate the sampling result, in the arbitration phase **451** the receiving module **122** uses the receiving threshold T1 of, for example, 0.7 V and the receiving threshold T2 of, for example, -0.35 V. In contrast, in the data phase **452** the receiving module **122** only uses signals that were evaluated with the receiving threshold T3. When switching between the first to third operating modes (SLOW or SiC, FAST_TX, FAST_RX) described above with reference to FIG. 3, the receiving module **122** switches in each case the receiving thresholds T2, T3.

[0067] The receiving threshold T2 is used to detect whether the bus **40** is free when the subscriber station **12** is newly connected to communication on the bus **40** and attempts to integrate itself into communication on the bus **40**. Each subscriber station **10**, **30** switches the operating mode of the transmitting/receiving device **12** to the operating mode of the arbitration phase **451** when the subscriber station **12** is (newly) connected to communication on the bus **40** or attempts to reintegrate into communication on the bus **40** following an error in bus communication. Only when it is detected that the bus **40** is free, may the subscriber station **10** in the cases mentioned itself transmit data, in particular messages **45**, **47**, to the bus **40**.

[0068] Upon receiving the corresponding signals from the bus **40**, each transmitting/receiving device **12** generates the associated reception signal RxD, as illustrated in FIG. 1 and FIG. 5. Ideally, the reception signal RxD has no time offset from the transmission signal TxD.

[0069] FIG. 5 shows the basic structure of the subscriber station **10** with its communication control device **11** and its transmitting/receiving device **12** which comprises the transmitting module **121** and the receiving module **122**. The subscriber station **10** also has a microcontroller **13**, a system ASIC (ASIC=application-specific integrated circuit) **14**, a first interface module **15** and a second interface module **16**. The second interface module **16** has time measuring blocks **161**, **1621**, **1622**, **1631**, **1632**, a switching block **164**, and an evaluation block **165**.

[0070] The first interface module **15** is assigned to the microcontroller **13** and to the device **11**. The first interface module **15** is arranged between the device **11** and a terminal TXD of the microcontroller **13**. The first interface module **15** is connected to the terminal TXD of the microcontroller **13**. The first interface module **15** outputs a signal TxD to the second interface module **16**.

[0071] The second interface module **16** is assigned to the device **12** and to the system ASIC **14**. The second interface module **16** is arranged between a terminal TXD of the system ASIC **14** and the transmitting module **121**. The second interface module **16** is connected to the terminal TXD of the system ASIC **14** and/or a corresponding terminal TXD of the device **12**.

[0072] The communication control device **11** can be designed as a protocol controller for transmitting and/or receiving CAN XL messages **45**, **47**. Optionally, the device **11** is designed for transmitting and/or receiving CAN FD or CAN SiC messages **46**.

[0073] The microcontroller **13** generates or processes data which are to be converted by the device **11** into the corresponding frame format for transmitting messages **45**, **47** via the bus **40**. To transmit the data on the bus **40**, the communication control device **11** generates a transmission signal TxD_PC according to the corresponding standard for CAN FD, CAN SiC or CAN XL messages **45**, **47** and outputs the transmission signal TxD_PC to the first interface module **15**.

[0074] The first interface module **15** is designed for pulse-width modulation (PWM) of the transmission signal TxD_PC. Accordingly, the transmission signal TxD at the terminal TXD of the microcontroller **13** is a transmission signal that is pulse-width-modulated at least in parts.

[0075] The second interface module **16** is designed to process the transmission signal TxD. In particular, the second interface module **16** is designed for pulse-width demodulation of the transmission signal TxD. Consequently, the transmission signal TxD_TC output by the module **16** is a decoded or pulse-width- demodulated transmission signal. These functions are explained in more detail with the aid of FIG. 6 to FIG. 9.

[0076] The system ASIC **14** of the subscriber station **10** in FIG. 5 can alternatively be a system basis chip (SBC) on which several functions necessary for an electronic module of the subscriber station **10** are combined. In the ASIC **14** system, a power supply device, not shown, which supplies the transmitting/receiving device **12** with electrical energy can be installed in addition to the transmitting/receiving device **12**. The power supply device usually supplies a voltage CAN_supply of 5 V. Depending on requirements, however, the power supply device **19** can provide a different voltage with a different value. Additionally or alternatively, the power supply device can be designed as a current source.

[0077] FIG. 5 shows the transmitting module **121** only in a simplified representation. The transmitting module **121** is connected directly to the bus **40** via terminals CANH, CANL in order to be able to transmit a transmission signal TxD TC of the second interface module **16** to the bus **40**.

[0078] According to FIG. 5, the receiving module **122** is also directly connected to the bus **40** via the terminals CANH, CANL. The receiving module **122** is designed to generate the digital reception signal RxD. The receiving module **122** transmits or outputs the reception signal RxD to the terminal RXD of the microcontroller **13** or the communication control device **11** via the terminal RXD of the system ASIC **14** or of the transmitting/receiving device **12**.

[0079] In the present exemplary embodiment, the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632** are in each case designed as counters. The switching block **164** operates or clocks the measuring, in particular the counting, of the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632**.

The time measuring blocks **161**, **1621**, **1622**, **1631**, **1632** do not carry out their measurements permanently but rather at predetermined time intervals or intermittently.

[0080] The switching block **164** may be designed as an oscillator block that outputs an oscillating signal FO (FIG. **12**) with a clock TK or a frequency f that corresponds to the period of the signal in FIG. **12**. The clock TK, and thus also the frequency f , is tuned to a time t_d in FIG. **6** or t-decode. In the time period or time t_d , during sampling of a bit of the transmission signal TxD, the level or bit value of the bit is to be determined, namely high (H=high) or low (L=low). The time t_d in FIG. **6** begins or ends with a bit of the transmission signal TxD. In addition, the value of the bit in the middle of a bit can be sampled with the time t_d (FIG. **6**), which starts or ends with the middle of the bit time period t_{per} . Since the time t_d is set to be not less than or shorter than 5 ns, the frequency f of the switching block 164 is chosen to be $f > 200$ MHz, in particular about 300 MHz. This means that the requirement for time t_d can be met. The bit time t_{per} of bits of the data phase **452** can have a value < 205 ns in the example given. As described in more detail below, the time period t_{per} is divided into the time periods t_1 , t_2 . Each of the time periods t_1 , t_2 must not be shorter or less than 5 ns.

[0081] The time measuring block **161** is provided for preselecting the operating mode of the data phase **452**. The time measuring blocks **1621**, **1622** are provided for detecting bits of the transmission signal TxD which have the bit time t_{bt2} . The time measuring blocks **1621**, **1631** are used to detect whether the data or bits in the transmission signal TxD are transmitted pulse-modulated in the FAST_TX operating mode. The time measuring blocks **1622**, **1632** are used to detect whether the data or bits in the transmission signal TxD are transmitted pulse-modulated in the FAST_RX operating mode.

[0082] FIG. **6** shows an example of a transmission signal TxD for a transmitting node, which the second interface module **16** receives from the first interface module **15** over time t , and thus serially. FIG. **7** shows the transmission signal TxD_TC, which the interface module **16** generates from the signal in FIG. **6** over time t for the transmitting module **121**. FIG. **8** and FIG. **9** show for a receiving node other examples of the transmission signal TxD and the transmission signal TxD_TC generated from it.

[0083] FIG. **6** and FIG. **8** each show the part of the transmission signal TxD of a message **45** at the end of the arbitration phase **451**. In this phase, the transmitting module **121** is switched to the SLOW or SiC operating mode, as shown in FIG. **6** and FIG. **8** in each case by the state S_TX.

[0084] If an edge occurs in the TxD signal, as shown in FIG. **6** and FIG. **8**, the time period t_{SL} will begin running and the time measuring block **161** starts a time measurement. The time period t_{SL} corresponds to the time period t_{select} , whereby according to CiA610-3 $t_{select} = 500 \dots 980$ ns. Measurement with the time measuring block **161** begins again with each edge of the transmission signal TxD. After the time period t_{SL} has elapsed, the time measuring block **161** stops its measurement and the evaluation block **165** detects the level of the transmission signal TxD. If this level is L (low), the evaluation block **165** specifies the operating mode FAST_TX, as indicated in FIG. **6** with the state F_TX_SL. The FAST_TX operating mode is thus preselected for the transmitting/receiving device **12** towards the end of the arbitration phase **451**. If, on the other hand, the level H (high) is evaluated at the end of the time period t_{SL} , the evaluation block **165** specifies the operating mode FAST_RX, as indicated in FIG. **8** with the state F_RX_SL. The FAST_RX operating mode is thus preselected for the transmitting/receiving device **12** towards the end of the arbitration phase **451**.

[0085] In addition, if the FAST_TX operating mode is preselected, as indicated in FIG. **6** with the state F_TX_SL, the time measuring block **1621** begins to measure whether two rising edges of the transmission signal TxD are transmitted in a time window having a temporal length equal to the time period or symbol duration t_{per} . The time period t_{per} is in particular equal to the time period t_{bt2} . The time period t_{bt2} corresponds to the time period t_{fast} , whereby according to CiA610-3 $t_{fast} = 205 \dots 245$ ns. Starting with a rising edge of the transmission signal TxD, the time

measuring block **1621** measures whether two consecutive rising edges occur in the symbol duration t_{per} of the transmission signal TxD. The time measuring block **1621** stops measurement once the time period t_{bt2} is reached. If two consecutive rising edges occur in the symbol duration t_{per} of the transmission signal TxD, the evaluation block **165** recognizes that the transmitting/receiving device **12** is to be switched to the FAST_TX operating mode, as indicated in FIG. **6** with the state F_TX for the levels F_TX)L1 and F_TX)L0. The FAST_TX operating mode is deactivated again when the evaluation block **165** detects that a longer or shorter time period than the time period t_{bt2} elapses between two consecutive rising edges in the transmission signal TxD. In addition, the FAST_TX operating mode is exited or deactivated again if the evaluation block **165** detects that the transmission signal TxD has the level H or L for a predetermined long time, in particular permanently.

[0086] At the end of and after the arbitration phase **451**, the transmission signal TxD in FIG. **6** is a sectionally pulse-width-modulated signal (PWM signal). In the pulse-width-modulated signal, PWM symbols have the symbol duration t_{per} , which in turn is divided into the time periods t_1 , t_2 . The transmission signal TxD changes after the time period t_1 from a first level, for example L (low), to a second level, for example H (high), as shown in FIG. **6**. The evaluation block **165** evaluates, with the aid of the levels L, H of the transmission signal TxD and the time periods t_1 , t_2 , which PWM symbol is encoded in the transmission signal TxD, and thus which level is to be created for bits in the transmission signal TxD_TC in FIG. **7** in order to generate with the transmitting module **121** the associated state (L1, L0 in FIG. **4**) on the bus **40**. For this purpose, the evaluation block **165** evaluates the measurements of at least two of the time measuring blocks **1621**, **1622**, **1631**, **1632** as described in more detail below.

[0087] In addition, if the FAST_RX operating mode is preselected, as indicated in FIG. **8** with the state F_RX_SL, the time measuring block **1622** begins to measure whether two falling edges of the transmission signal TxD are transmitted in a time window having a time length equal to the time period of a PWM symbol or symbol duration t_{per} . The time period t_{per} is in particular equal to the time period t_{bt2} , as mentioned above. The time measuring block **1622** measures, starting or beginning with a falling edge of the transmission signal TxD, whether two consecutive falling edges occur in the symbol duration t_{per} of the transmission signal TxD. The time measuring block **1622** stops measurement once the time period t_{bt2} is reached. If two consecutive falling edges occur in the symbol duration t_{per} of PWM symbols of the transmission signal TxD, the evaluation block **165** recognizes that the transmitting/receiving device **12** is to be switched to the FAST_RX operating mode, as indicated in FIG. **8** with the state F_RX. The FAST_RX operating mode is deactivated again when the evaluation block **165** detects that a longer or shorter time period than the time period t_{bt2} elapses between two consecutive falling edges in the transmission signal TxD. In addition, the FAST_RX operating mode is quit or deactivated again if the evaluation block **165** detects that the transmission signal TxD has the level H or L for a predetermined long time, in particular permanently.

[0088] To decode the pulse-width-modulated signal TxD in FIG. **6** into the signal TxD_TC in FIG. **7**, the following procedure can be followed. The same applies to decoding the pulse-width-modulated signal TxD in FIG. **8** into the signal TxD_TC in FIG. **9**.

[0089] FIG. **10** shows for the data phase **452** an example of a transmission signal TxD_PC, which the device **11** in FIG. **3** generates over time t with the values H, L for bits with the symbol duration t_{per} or bit time or bit_duration t_{bt2} and outputs serially to the first interface module **15** in FIG. **3**. FIG. **11** shows the transmission signal TxD, which the first interface module **15** generates from the signal in FIG. **10** by means of pulse modulation and outputs serially to the terminal TXD of the device **12** in FIG. **3**. The transmission signal TxD encodes with PWM symbols the states L0, L1 for the bus signal in FIG. **4**, as mentioned above. The time t_2 in the signal TxD in FIG. **11** corresponds to the on time of the transmission signal TxD. The on time is the time for which the transmission signal TxD has the level H or is switched on. For the remaining time at the start or

from the beginning of the symbol time t_{per} , the transmission signal TxD has the level L or is switched off. If communication takes place on bus 40 at 10 Mbit/s, the symbol duration t_{per} will have a length of 100 ns, for example. In this case, the time t_1 for a PWM symbol encoding the state L0 can be chosen to be, for example, 25 ns. In addition, a length of 75 ns can be chosen as the time t_1 for a PWM symbol that encodes the state L1. The time t_1 is therefore either, for example, 25% of the symbol duration t_{per} or, for example, 75% of the symbol duration t_{per} , in order to encode the states L0, L1. Of course, other time lengths for the time t_1 are possible, in particular those permitted by the CiA610-3 standard for CAN XL.

[0090] The transmission signal TxD is demodulated or decoded in the second interface module **16**. Here, the oscillator signal FO in FIG. **12** is used to trigger the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632**. With the aid of the corresponding signals in FIG. **13** and FIG. **14**, the second interface module **16** generates the transmission signal TxD_TC, as already mentioned above. The second interface module **16** generates the transmission signal TxD_TC with a predetermined time offset to the transmission signal TxD, as shown by the arrows between FIG. **11** and FIG. **15**. According to FIG. **11** and FIG. **15**, the time offset is slightly more than one bit time or bit_duration t_{bt2} .

[0091] If the FAST_TX operating mode is preselected, as indicated in FIG. **6** with the state F_TX_SL, the time measuring block **1621** begins measuring from one rising edge to the next rising edge of the transmission signal TxD, as shown in FIG. **13**. The count value C2_FT of the time measuring block **1621** increases maximally up to a limit value GW, which corresponds to the time period t_{per} or t_{bt2} . The time measuring block **1621** stops when the limit value GW (maximum count value) or the time period t_{per} is reached. In addition, as shown in FIG. **14**, the time measuring block **1631** begins measuring from one rising edge to the next falling edge of the transmission signal TxD. The time measuring block **1631** stops when the time period t_{per} is reached. At the end of the PWM symbol, i.e. the second rising edge of the transmission signal TxD, the PWM symbol is decoded. The digital circuit_design allows an easy halving of the count value C2_FT of the time measuring block **1631** by means of a digital shift operation by one place. The comparison then takes place. The evaluation block **165** evaluates the state L1 as a value for the measured PWM symbol of the transmission signal TxD if the count value $C3_{\text{FT}} < \text{count value } C2_{\text{FT}}/2$, so that the value L is to be generated in the transmission signal TxD_TC in FIG. **15**. The evaluation block **165** evaluates L0 as the value for the measured PWM symbol of the transmission signal TxD if count value $C3_{\text{FT}} > \text{count value } C2_{\text{FT}}/2$, so that the value L is to be generated in the transmission signal TxD_TC in FIG. **15**. It is not defined which value or level is to be generated in the transmission signal TxD_TC if count value $C3_{\text{FT}} = \text{count value } C2_{\text{FT}}/2$.

[0092] The respective result of the evaluation by the evaluation block **165** is passed on to the transmitting module **121** as the decoded transmission signal TxD_TC in FIG. **15**. The transmitting module **121** converts an L level of the transmission signal TxD_TC into an L0 signal or an L0 state on the bus (VDIFF=1V, FIG. **4**). The transmitting module **121** converts an H level of the transmission signal TxD_TC into an L1 signal or an L1 state on the bus **40** (VDIFF=-1V, FIG. **4**).

[0093] FIG. **16** to FIG. **21** show the corresponding signals for a receiving node for which the FAST_RX operating mode is preselected for the data phase **452**, as indicated in FIG. **8** with the state F_RX_SL. In this case, the transmission signal TxD_PC is transmitted with level H in the data phase **452**, as shown in FIG. **16**. The interface module **15** generates the transmission signal TxD accordingly with PWM symbols L1, as shown in FIG. **17**. In this case, triggered by the signal according to FIG. **18** of the switching block **164**, the time measuring block **1622** begins measuring from a falling edge of the transmission signal TxD to the next falling edge of the transmission signal TxD. The time measuring block **1622** thus measures the duration of the time period t_{per} . The count value C2_FR of the time measuring block **1622** in FIG. **19** increases maximally up to a limit value GW which corresponds to the time period t_{per} . The time measuring block **1622** stops when the limit value GW or the time period t_{per} is reached. In addition, as shown in FIG. **20**, the

time measuring block **1632** begins measuring from a falling edge of the transmission signal TxD to the next rising edge of the transmission signal TxD. The time measuring block **1632** thus measures the duration of the off time of the PWM symbol. The time measuring block **1632** stops when the time period t_{per} or t_{bt2} is reached. At the end of the PWM symbol, i.e. of the second falling edge of the transmission signal TxD in FIG. 17, validation of the PWM symbol is undertaken. The PWM symbol is accepted by the evaluation block **165** if $C3_FR > 1$ and $C3_FR < GW - 1$. If the PWM symbol is accepted, an H level will be evaluated as the value for the measured PWM symbol of the transmission signal TxD. The H level corresponds to a recessive state **402** (FIG. 4) on the bus **40**.

[0094] The above-described design of the second interface module **16** forms a digital circuit concept. This concept makes possible a high degree of accuracy in the execution of the three functions (operating mode preselection, detection of the bit time t_{bt2} of the data phase **452** for switching between the operating modes for the transmitting module **121**, decoding of the PWM symbols of the transmission signal TxD) of the second interface module **16**. This is made possible by using a balanced circuit block **164**, in particular a ring oscillator block which outputs an oscillating signal FO, as shown in FIG. 12. The achievable accuracy is approximately $\pm 7\%$.

[0095] An advantage of the second interface module **16** is in particular the high accuracy for evaluating the pulse-modulated signal TxD, since the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632** are operated with the same clock TK or the same frequency f of the switching block **164**. This in particular also applies to the time measuring blocks **1621**, **1622** for period t_{bt2} and the time measuring blocks **1631**, **1632** for the on time t_1 or off time t_2 , since the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632** also work very well together.

[0096] In addition, the above-described design of the second interface module **16** makes a very low silicon area consumption possible in the hardware. This makes the interface module **16** very inexpensive and space-saving. This also contributes to the fact that a higher-level transmitting/receiving device (transceiver) **12** can be manufactured very inexpensively and in a space-saving manner.

[0097] Yet another advantage of the second interface module **16** lies in a high test coverage and the associated component quality.

[0098] In addition, there is a high degree of flexibility for adaptations in additional development steps of the second interface module **16**.

[0099] There is also the option of using the second interface module **16** as a digital part in a CAN-XL transmitting/receiving device **12**. This allows functional expansions in the device **12**, such as security functions, particularly against manipulation.

[0100] According to a second exemplary embodiment, the evaluation block **165** is designed to additionally evaluate the measurement of the time measuring block **1632** in the FAST_RX operating mode. The FAST_RX operating mode remains active for as long as the value of the time measuring block **1632** is greater than or equal to $1 \cdot t_d$, AND the value of the time measuring block **1632** is less than the value of the time measuring block **1622** minus $1 \cdot t_d$.

[0101] This provides a redundancy which prevents the FAST_RX operating mode from being deactivated too quickly or remaining active for too long.

[0102] According to a third exemplary embodiment, at least one of the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632** is realized according to an analog circuit concept.

[0103] In contrast to the transmitting/receiving device **12** of the above-described exemplary embodiment, at least one of the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632** is designed to carry out the above-described time measurement doing so using at least one switched constant current. At least one of the time measuring blocks **161**, **1621**, **1622**, **1631**, **1632** comprises a capacitor. In addition, the at least one time measuring block **161**, **1621**, **1622**, **1631**, **1632** has at least one voltmeter for measuring the voltage across the capacitor.

[0104] The constant currents are switched in such a way that the capacitor is charged or discharged as required. The evaluation block **165** can be designed to evaluate the voltage measurement. The

evaluation block **165** can in particular carry out a voltage comparison of the measured capacitor voltages.

[0105] Accordingly, the switching block **164** may include at least one switchable constant current source. The switching block **164** is designed to switch the at least one constant current source in order to charge the capacitor over time. The at least one time measuring block **161**, **1621**, **1622**, **1631**, **1632** is designed to stop the switching of the at least one constant current source and thus the charging of the at least one capacitor at a predetermined limit voltage value (GW) which corresponds to the predetermined time t_{per} or t_1 or t_{bt2} or t_{SL} .

[0106] In this way too, the functions described above for the second interface module **16** in the above-described embodiments can also be advantageously implemented.

[0107] However, such an implementation requires a significantly higher silicon area consumption for the second interface module **16** than in the above-described embodiments.

[0108] All above-described embodiments of the interface module **16** of the transmitting/receiving device **12**, of the subscriber stations **10**, **20**, **30**, the bus system **1** and of the method carried out therein according to the exemplary embodiments and their modifications can be used individually or in all possible combinations. Additionally or alternatively, the following modifications are in particular possible.

[0109] The above-described bus system **1** according to at least one of the exemplary embodiments is described on the basis of a bus system based on the CAN protocol. However, the bus system **1** according to at least one of the exemplary embodiments may alternatively be another type of communication network in which the signals are transmitted as differential signals. It is advantageous, but not a necessary condition, for an exclusive, collision-free access of a subscriber station **10**, **20**, **30** to the bus **40** to be ensured in the case of the bus system **1**, at least for certain time periods.

[0110] The bus system **1** according to at least one of the exemplary embodiments and their modifications is in particular a bus system in which communication can take place between at least two of the subscriber stations **10**, **20**, **30** according to two different CAN standards, such as CAN-HS or CAN FD or CAN SiC or CAN XL. However, the bus system **1** may be another communication network in which the signals are transmitted as differential signals and serially via the bus **40**. The functionality of the above-described exemplary embodiments can thus be used, for example, in transmitting/receiving devices **12**, **22** that are to be operated in such a bus system.

[0111] The number and arrangement of the subscriber stations **10**, **20**, **30** in the bus system **1** according to at least one of the exemplary embodiments and their modifications can be selected as desired.

Claims

1-16. (canceled)

17. An interface module for a subscriber station of a serial bus system, wherein in the serial bus system has a transmitting module configured to transmit a digital transmission signal as an analog differential signal onto a bus of the bus system to transmit a message to at least one other subscriber station of the bus system, wherein bits in the digital transmission signal have a longer bit duration in a first communication phase than in a second communication phase of the transmission signal, the interface module comprising: at least one time measuring block configured to measure, with a predetermined timing clock, a predetermined time in a digital transmission signal, wherein the at least one time measuring block is configured to stop its measurement when a measured value of the at least one time measuring block has reached a predetermined limit value; and an evaluation block configured to evaluate the measured value of the at least one time measuring block at an end of the predetermined time to set the transmitting module, based on an evaluation result of the evaluating, for operation in the first communication phase or for operation in the second

communication phase.

18. The interface module according to claim 17, wherein the evaluation block is configured to, based on the evaluation result: (i) switch the transmitting module to a first operating mode for the first communication phase for transmission, or (ii) switch the transmitting module to a second operating mode or third operating mode for the second communication phase for transmission.

19. The interface module according to claim 18, wherein the at least one time measuring block includes at least four time measuring blocks, and wherein the evaluation block is configured to evaluate a measured value of each of the at least four time measuring blocks at an end of the predetermined time to decode a pulse-width-modulated symbol of the digital transmission signal by evaluating the measured values of the at least four time measuring blocks in order to output a pulse-width-demodulated transmission signal to the transmitting module.

20. The interface module according to claim 19, further comprising: a switching block configured to output an oscillating signal for specifying the predetermined timing clock with which all of the time measuring blocks carry out their measurement intermittently.

21. The interface module according to claim 19, wherein: a first time measuring block of the at least four time measuring blocks is configured to restart the measurement with each edge of the digital transmission signal and to stop the measurement when the measured value of the first time measuring block has reached a predetermined limit value; and the evaluation block is configured to preselect the second operating mode for the transmitting module for the second communication phase when the measured value of the first time measuring block corresponds to a first predetermined time.

22. The interface module according to claim 21, wherein: a second time measuring block of the at least four time measuring blocks is configured to start the measurement with each rising edge of the digital transmission signal and to measure until a next rising edge of the digital transmission signal, until the predetermined limit value is exceeded; a third time measuring block of the at least four time measuring blocks is configured to start the measurement with each falling edge of the digital transmission signal and to measure until the next falling edge of the digital transmission signal, until the predetermined limit value is exceeded; and wherein the evaluation block is configured to switch the transmitting module to the second operating mode for the second communication phase when the evaluation of the measurement result of the second and third time measuring blocks indicates that, in the digital transmission signal, a distance between a rising edge and a falling edge or the distance between a falling edge and a rising edge is equal to a second predetermined time, which is the bit duration of bits of the second communication phase and which is shorter than the first predetermined time.

23. The interface module according to claim 22, wherein: a fourth time measuring block of the at least four time measuring blocks is configured to start the measurement with each rising edge of the digital transmission signal and to measure until the next falling edge of the digital transmission signal, until the predetermined limit value is exceeded; a fifth time measuring block of the at least four time measuring blocks is configured to start the measurement with each falling edge of the digital transmission signal and to measure until the next rising edge of the digital transmission signal, until the predetermined limit value is exceeded; and the evaluation block is configured to compare the measurement of the second and fourth time measuring blocks with each other and to compare the measurement of the third and fifth time measuring blocks with each other, and to output the pulse-width-demodulated transmission signal to the transmitting module based on results of the comparisons.

24. The interface module according to claim 19, wherein the at least one time measuring block includes a counter.

25. The interface module according to claim 24, wherein: at least two time measuring blocks of the at least one time measuring block are configured to stop counting at a count value which corresponds to a symbol time period for a PWM symbol of a pulse-width-modulated (PWM)

signal; and the evaluation block is configured to evaluate a value of the PWM symbol of the pulse-width-modulated signal at an end of a PWM symbol using the at least one time measuring block and, using a result of the evaluation of the value of the PWM symbol, to output a level for the pulse-width-demodulated transmission signal to the transmitting module.

26. The interface module according to claim 17, wherein: the at least one time measuring block includes a capacitor having a voltmeter for measuring a voltage across the capacitor; at least one switchable constant current source is provided as a switching block which is configured to switch the at least one constant current source to charge the capacitor over time; the at least one time measuring block is configured to stop charging of the at least one capacitor at a predetermined limit voltage value which corresponds to the predetermined time.

27. The interface module according to claim 26, wherein the evaluation block is configured to evaluate a voltage value of the at least one time measuring block.

28. A transmitting/receiving device, comprising: a transmitting module configured to transmit a digital transmission signal as an analog differential signal onto a bus of a bus system to transmit a message to at least one other subscriber station of the bus system, wherein bits in the digital transmission signal have a longer bit duration in a first communication phase than in a second communication phase of the transmission signal; a receiving module configured to receive signals from the bus and to generate a digital reception signal from the analog differential signal; and an interface module including: at least one time measuring block configured to measure, with a predetermined timing clock, a predetermined time in a digital transmission signal, wherein the at least one time measuring block is configured to stop its measurement when a measured value of the at least one time measuring block has reached a predetermined limit value, and an evaluation block configured to evaluate the measured value of the at least one time measuring block at an end of the predetermined time to set the transmitting module, based on an evaluation result of the evaluating, for operation in the first communication phase or for operation in the second communication phase.

29. The transmitting/receiving device according to claim 28, wherein the transmitting module is configured to generate the analog differential signals in the first communication phase of the message with a different physical layer than in the second communication phase.

30. A subscriber station for a serial bus system, comprising: a transmitting/receiving device including: a transmitting module configured to transmit a digital transmission signal as an analog differential signal onto a bus of a bus system to transmit a message to at least one other subscriber station of the bus system, wherein bits in the digital transmission signal have a longer bit duration in a first communication phase than in a second communication phase of the transmission signal, a receiving module configured to receive signals from the bus and to generate a digital reception signal from the analog differential signal, and an interface module including: at least one time measuring block configured to measure, with a predetermined timing clock, a predetermined time in a digital transmission signal, wherein the at least one time measuring block is configured to stop its measurement when a measured value of the at least one time measuring block has reached a predetermined limit value, and an evaluation block configured to evaluate the measured value of the at least one time measuring block at an end of the predetermined time to set the transmitting module, based on an evaluation result of the evaluating, for operation in the first communication phase or for operation in the second communication phase; and a communication control device configured to control communication in the bus system and to generate a digital transmission signal as a basis for a digital pulse-width-modulated transmission signal for the interface module.

31. The subscriber station according to claim 30, wherein the subscriber station is configured for communication in a bus system in which an exclusive, collision-free access of a subscriber station to the bus of the bus system is guaranteed at least temporarily.

32. A method for transmitting differential signals in a serial bus system, wherein the method is carried out using an interface module for a subscriber station of a serial bus system, wherein the subscriber station has a transmitting module and wherein the interface module has at least one time

measuring block and an evaluation block, and wherein the method comprises the following steps: measuring, using the at least one time measuring block and using a predetermined timing clock, a predetermined time in a digital transmission signal, wherein the at least one time measuring block stops its measurement when the at least one time measuring block has reached a predetermined limit value; evaluating, using the evaluation block, a measurement result of the measuring of the at least one time measuring block at an end of the predetermined time; setting, based on a result of the evaluating, the transmitting module for operation in the first communication phase or for operation in the second communication phase; and transmitting, using the transmitting module, a digital pulse-width-demodulated transmission signal as an analog differential signal onto the bus of the bus system to transmit a message to at least one other subscriber station of the bus system, wherein bits in the digital pulse-width-demodulated transmission signal have a longer bit duration in a first communication phase than in a second communication phase of the transmission signal.
