

Ch-4 PULSE MODULATION & DEMODULATION :-

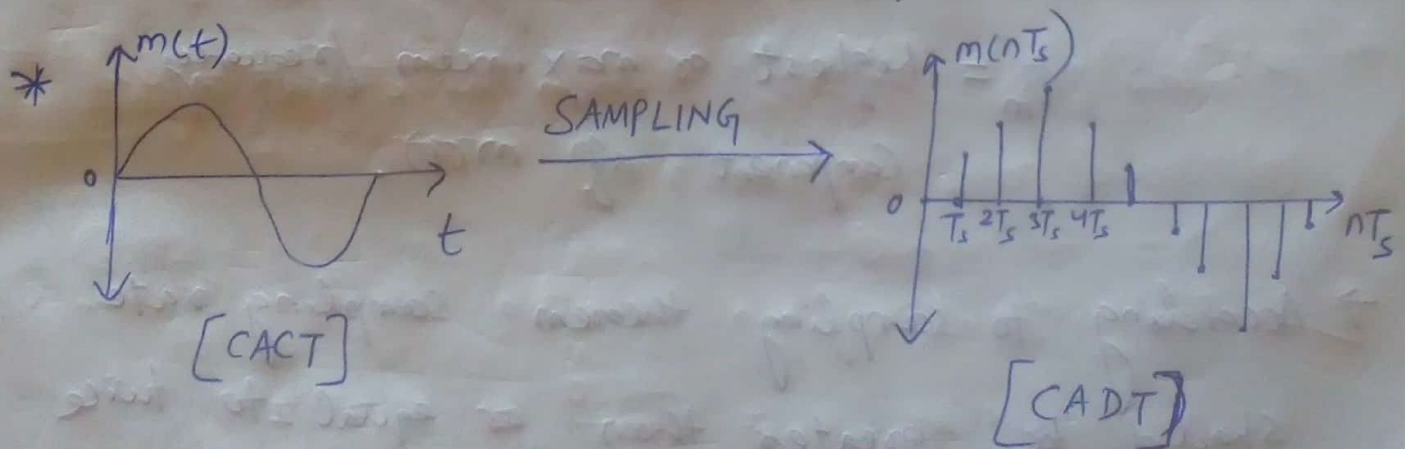
Chapter Coverage

Sampling theorem (Low pass & band pass signal),
natural sampling, flat top sampling, signal recovery
through holding, PAM, channel bandwidth for PAM,
PDM, PWM, PPM.

SAMPLING : It is the process which converts
a continuous amplitude continuous time (CACT)
signal to continuous amplitude discrete time (CADT)
signal.

→ CACT is analog signal $\rightarrow m(t)$

CADT is discrete time signal $\rightarrow m(nT_s)$



→ But SAMPLING process is done in
accordance with a theorem called SAMPLING Theorem.

→ Sampling thm can be studied in two categories

(i) sampling theorem for Low pass signal

(ii) " " " Bandpass signal

(i) Sampling theorem for Lowpass signal :

" If the highest frequency contained in an ~~analog~~ analog signal $m(t)$ is f_m and the signal is sampled at a rate $f_s \geq 2f_m$, then $m(t)$ can be exactly recovered from its sample values."

→ Here f_s = sampling rate = sampling frequency

$\frac{1}{f_s} = T_s$ = sampling time interval

f_m = highest or maximum frequency component of $m(t)$.

→ According to sampling theorem sampling rate should be greater than or equal to twice the maximum frequency.

i.e. $f_s \geq 2f_m$

$$\Rightarrow \frac{1}{T_s} \geq 2f_m$$

$$\Rightarrow \boxed{T_s \leq \frac{1}{2f_m}}$$

* Sampling are of 3 types

① oversampling if $f_s > 2f_m$

② critical sampling if $f_s = 2f_m$

③ Under sampling if $f_s < 2f_m$


* The sampling rate $f_s = 2f_{\max} = 2f_m$ is called
Nyquist rate or Nyquist frequency.

* Nyquist Interval $= T_s = \frac{1}{2f_{\max}} = \frac{1}{2f_m}$

* Undersampling ($f_s < 2f_{\max}$) is totally avoided
in communication.

(II) Sampling theorem for bandpass signal:

"A bandpass signal with highest frequency f_H and bandwidth B , can be recovered from its samples through bandpass filtering by sampling it with frequency $f_s = \frac{2f_H}{K}$, where K is the largest integer not exceeding $\frac{f_H}{B}$."

 \rightarrow All frequencies higher than f_s but below $2f_H$ may or may not be useful for band pass ~~filtering~~ sampling depending on overlap of shifted spectrum.

(Q) A signal $m(t) = 2\cos 6000\pi t + 4\cos 8000\pi t + 6\cos 10000\pi t$ is to be truthfully represented by its samples. What is the minimum sampling rate from

- (a) low pass sampling theorem consideration and
- (b) band pass consideration?

Solⁿ: The signal given has two frequency components.

$$\text{highest } f_H = \frac{10000}{2} = 5000 \text{ Hz}$$

$$\text{lowest } f_L = \frac{6000}{2} = 3000 \text{ Hz}$$

(a) Minimum sampling frequency from low pass consideration = $2f_H = 2 \times 5000 = 10000$ sps

~~(b) $BW = f_H - f_L = 5000 - 3000 = 2000$ Hz~~

~~$K = f_{\text{low}} \left(\frac{f_H}{B} \right) = f_{\text{low}} \left(\frac{5000}{2000} \right) = f_{\text{low}} (2.5) = 2$~~

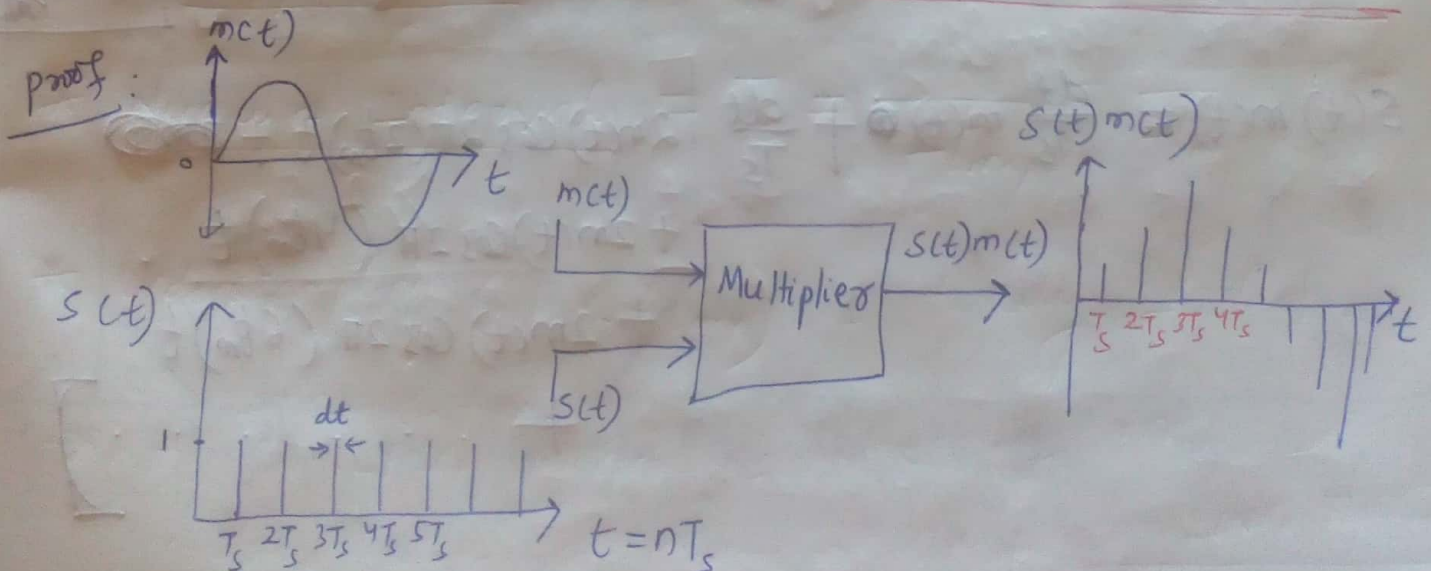
~~→ Then required sampling frequency~~

~~$= \frac{2f_H}{K} = 2 \times \frac{5000}{2} = 5000$ sps~~

~~Sps = samples per second.~~

* proof of Low pass sampling Theorem :

→ Lowpass sampling thm suggests to set sampling frequency $f_s \geq 2f_{\max}$. Then why not $f_s < 2f_{\max}$?



Here $m(t)$ = baseband message signal (CACT)

$s(t)$ = periodic pulse train of period T_s
and width dt .

Here T_s = sampling time interval

$s(t)m(t)$ = Sampled signal (CADD) =

In mathematics

$$s(t) = \frac{dt}{T_s} + \frac{2dt}{T_s} \left(\cos 2\pi \frac{t}{T_s} + \cos 2 \times 2\pi \frac{t}{T_s} + \dots \right)$$

~~sampled~~

sampled signal

$$s(t)m(t) = \frac{dt}{T_s} m(t) + \frac{2dt}{T_s} \left(2m(t) \cos 2\pi \frac{t}{T_s} + 2m(t) \cos 2 \times 2\pi \frac{t}{T_s} + \dots \right)$$
$$= dt b_s m(t) + dt b_s \left[2m(t) \cos 2\pi b_s t + 2m(t) \cos 2\pi (2b_s) t + \dots \right]$$

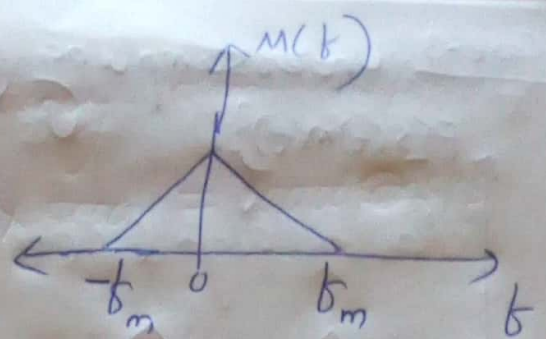
The important is to set the value of T_s +

Case-1: critical sampling $b_s = 2b_m \Rightarrow \frac{1}{T_s} = 2b_m \Rightarrow \boxed{T_s = \frac{1}{2b_m}}$

$$s(t)m(t) = \frac{dt}{T_s} m(t) + \frac{dt}{T_s} \left[2m(t) \cos 2\pi (2b_m) t + 2m(t) \cos 2\pi (4b_m) t + 2m(t) \cos 2\pi (6b_m) t + \dots \right]$$

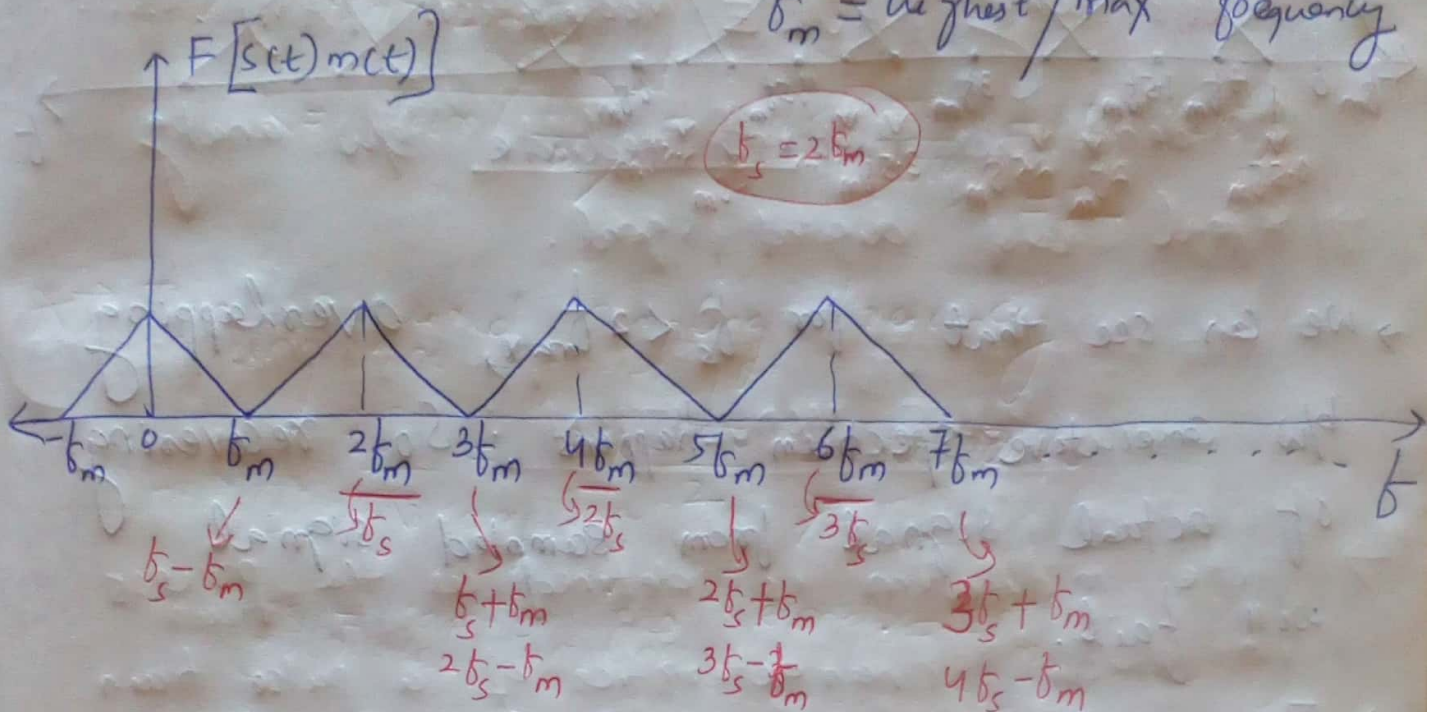
Spectral diagram :

Let $m(t)$ is such that



$b_m = \text{highest / max}^m \text{ frequency}$

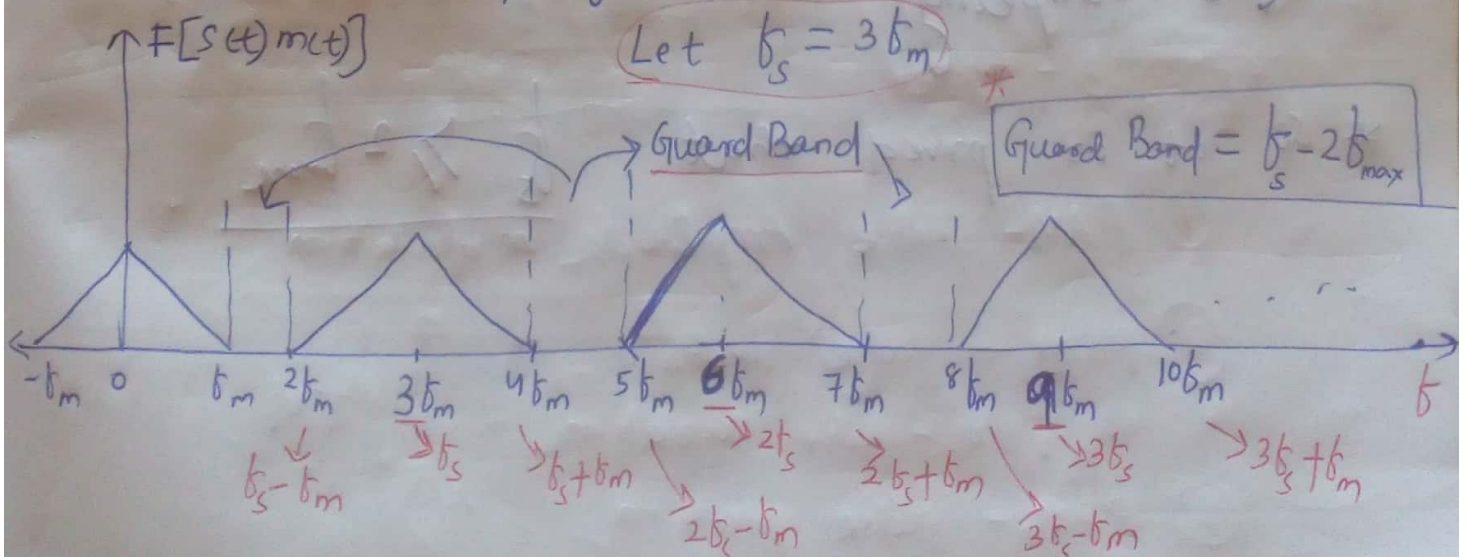
$$b_s = 2b_m$$



* From Case-1 Spectral diagram we can draw Spectral diagram of Case-2 ($b_s > 2b_m$) and Case-3 ($b_s < 2b_m$).

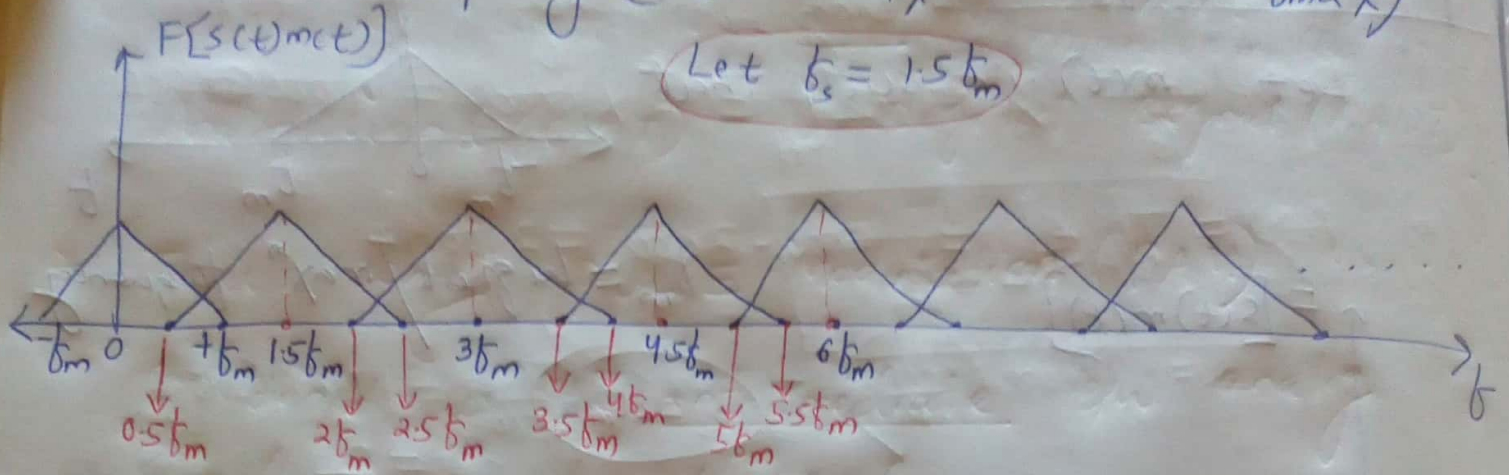
* Case-2: oversampling ($b_s > 2b_m$ or $T_s < \frac{1}{2b_m}$)

$$\text{Let } b_s = 3b_m$$



Case-3: Undersampling ($f_s < 2f_{max}$ or $T_s > \frac{1}{2f_{max}}$)

Let $f_s = 1.5f_m$



→ We can see that for $f_s < 2f_{max}$, overlapping b/w consecutive spectrum happens. So recovering of actual signal from sampled signal will fail.

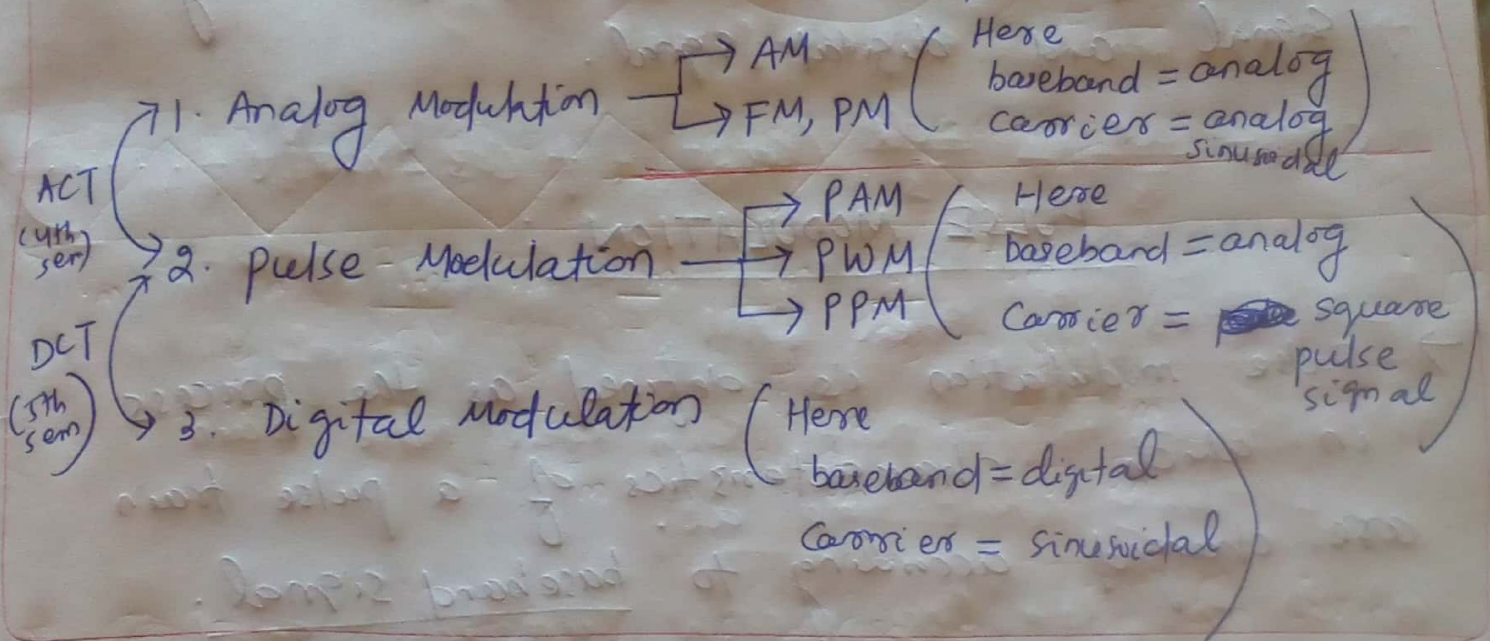
→ ~~The effect~~ The signal loss due to overlapping of consecutive spectrum in case of $f_s < 2f_{max}$ is known as ALISING.

* To remove alising, we use Anti ALising Filter before sampling.

* So it is proved that $f_s \geq 2f_{max}$

* Sampling technique is a pulse Modulation Technique but with a rule $f_s \gg 2f_{max}$.

In Communication we will study 3-types of Conventional modulation techniques.



* Sampling process is a PAM-type of pulse modulation.

* Sampling is of 3-types / PAM is of 3-types

① Ideal / instantaneous Sampling / PAM

② Natural Sampling / PAM

③ Flat-Top Sampling / PAM

* Sampling process is used to convert Analog signal (CACT) to Digital signal (DADT)

* That's why PULSE Modulation (PAM) is considered as the intermediate b/w analog signal and digital signal.

PULSE MODULATION :

→ pulse modulation is defined as the process in which characteristics of a pulse train are changed according to baseband signal.

→ Here carrier = ~~signal~~ pulse train
message = analog signal

→ Like analog modulation pulse modulation is of 3 - types.

- ① PAM - pulse Amplitude modulation
- ② PWM - pulse width " } PTM
or PDM - pulse Duration "
- ③ PPM - pulse position "

* PWM or PDM & PPM togetherly called as
pulse Time Modulation (PTM).

PAM

Defⁿ: In PAM, Amplitude of pulse train will
be changed according to baseband signal.

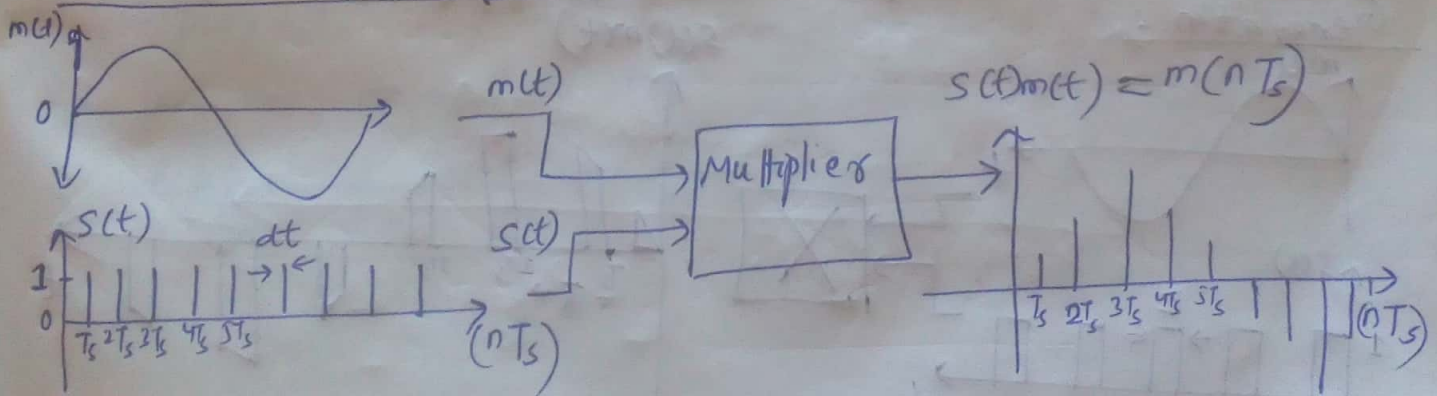
→ PAM signals are known as Sampled
signal if $t_s \geq 2 t_{max}$.

→ 3-types of PAM signal.

- (i) Ideal / instantaneous PAM
- (ii) Natural PAM
- (iii) Flat Top PAM

(i) Ideal PAM / Ideal Sampling:

Generation of Ideal PAM:



→ It is called instantaneous PAM or sampling because width of pulse train is infinitesimally small i.e. dt .

Here sampling interval $= T_s$

→ We can write $s(t)m(t) = m(nT_s)$

→ $s(t)$ can be expressed as

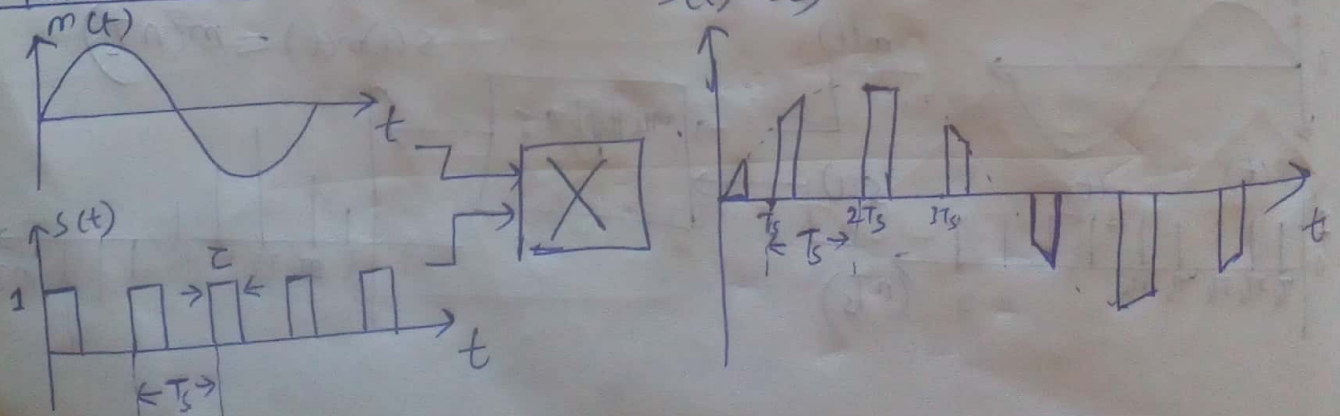
$$s(t) = \frac{dt}{T_s} + \frac{2dt}{T_s} \left(\cos 2\pi f_s t + \cos 2 \times 2\pi f_s t + \dots \right)$$

Limitations of Ideal PAM:

- (i) Difficult to construct such ideal pulse train of very short time dt .
- (ii) In this sampling, each sampled signal contains infinitesimal energy so when this signal will be communicated it will be lost by external noise.

(ii) Natural PAM / SAMPLING:

Generation:



Here

$$s(t) = \frac{T}{T_s} + \frac{2T}{T_s} \left(C_1 \cos 2\pi f_s t + C_2 \cos 2 \cdot 2\pi f_s t + \dots \right)$$

where

$$C_n = \frac{\sin\left(\frac{n\pi T}{T_s}\right)}{\left(n\pi T/T_s\right)}$$

→ Natural Sampled Signal

$$s(t)m(t)$$

$$= \frac{T}{T_s} m(t) + \frac{2T}{T_s} \left[m(t) C_1 \cos 2\pi f_s t + m(t) C_2 \cos 4\pi f_s t + \dots \right]$$

→ To demodulate message signal $m(t)$ from $s(t)m(t)$ we have to use a Lowpass Filter.

then LPF will result

$$s_0(t) = \frac{T}{T_s} m(t)$$

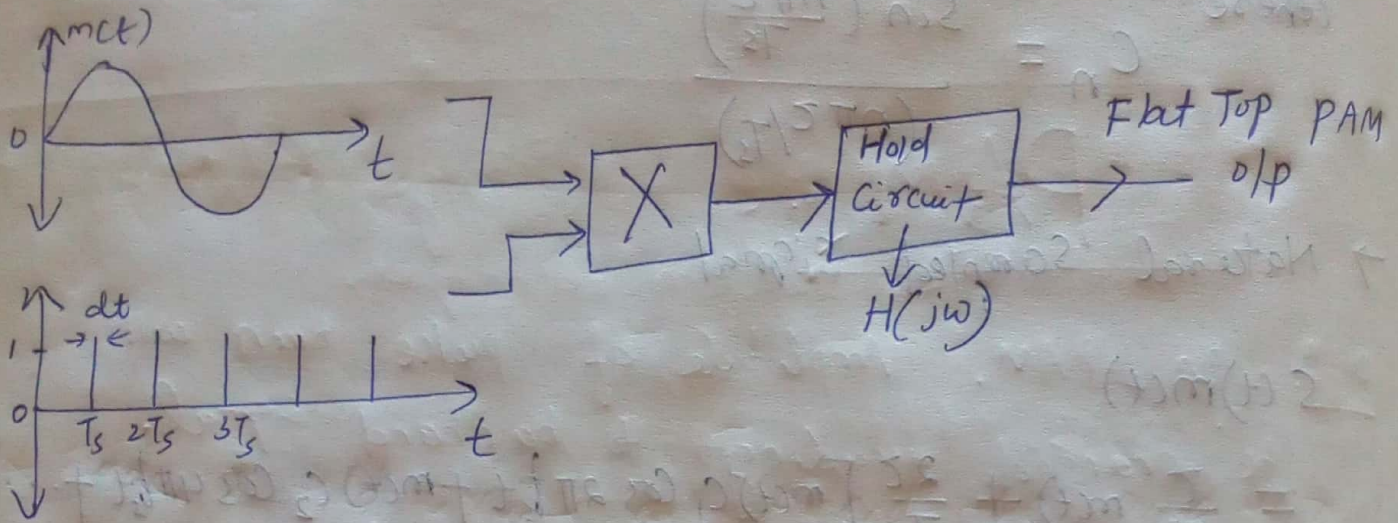
Limitation of Natural PAM :

→ The sampled signal of Natural PAM has top part not constant rather it follows the waveform of the signal which is not frequently employed.

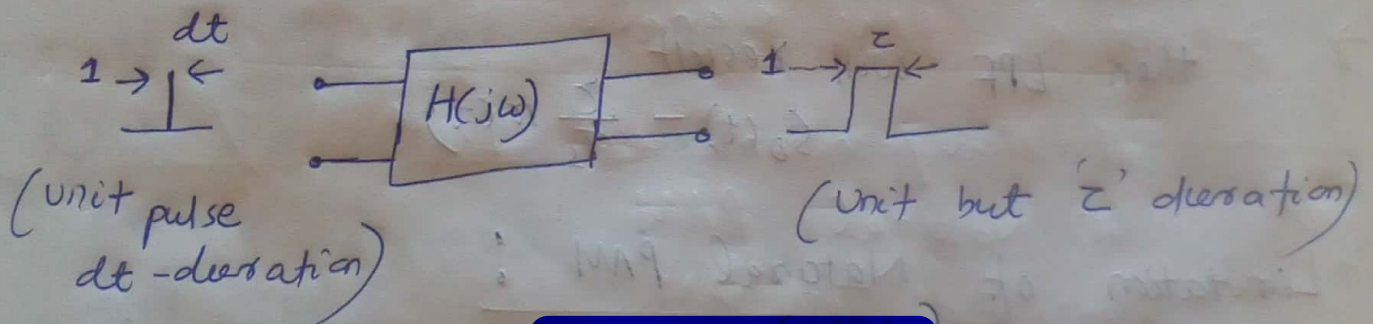
→ Due to nonuniform top, electronic circuits face difficulty to process it.

(iii) Flat Top PAM / Flat-Top Sampling :-

Generation :

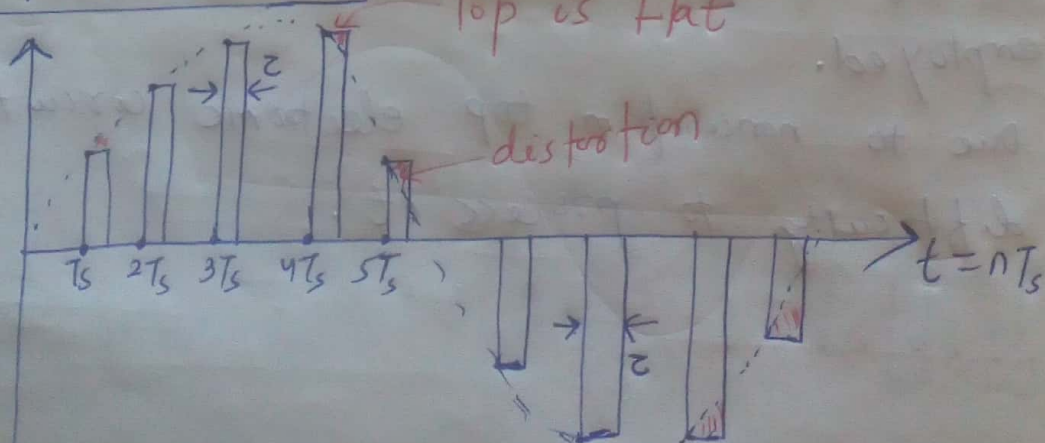


* $H(jw)$ = Transfer function of Hold circuit. It helps to extend the time duration of pulse.



$$H(jw) = \frac{z}{dt} \frac{\sin(wz/2)}{wz/2}$$

→ Flat Top PAM signal :



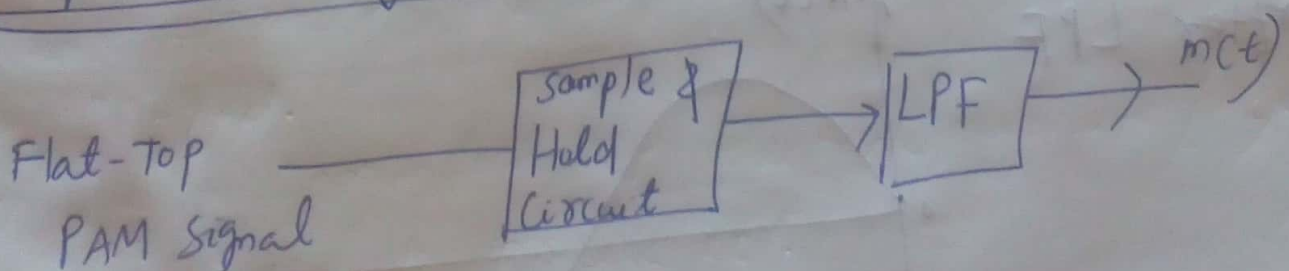
→ The gap b/w original signal and flat-top signal is appeared as a distortion. It is known as Aperture Effect Distortion.

→ More the width of the top (τ) more will be distortion. If in order to reduce distortion ' τ ' will be reduced then signal strength will also be decreased which is a challenge to maintain desired SNR.

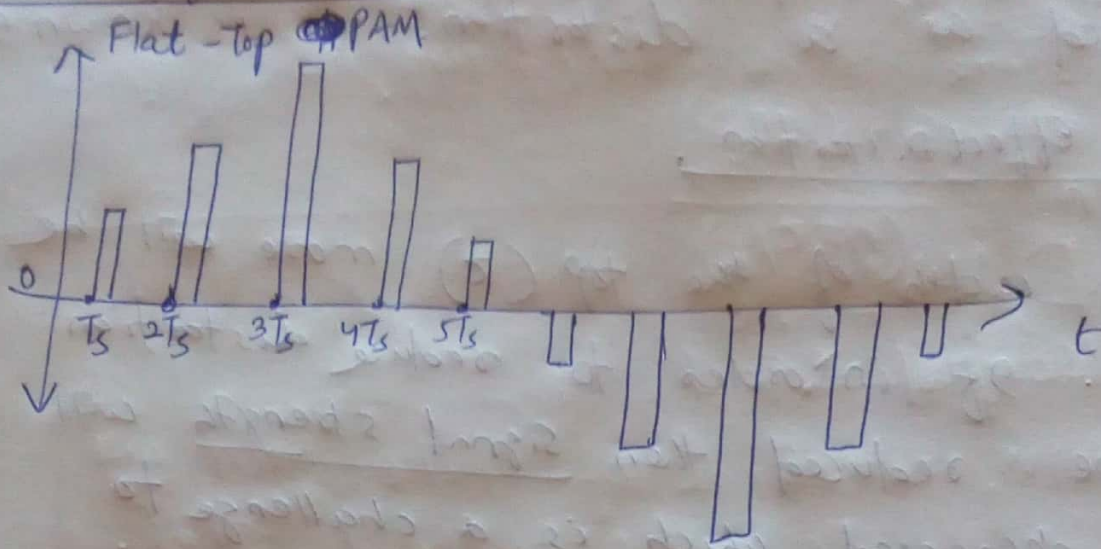
→ So not ~~compromising~~ compromising ' τ ' value and to reduced distortion, we will use an Equalizer at the receiver side.

* To recover $m(t)$ from Flat-Top Sampled signal, only Lowpass Filter is not sufficient

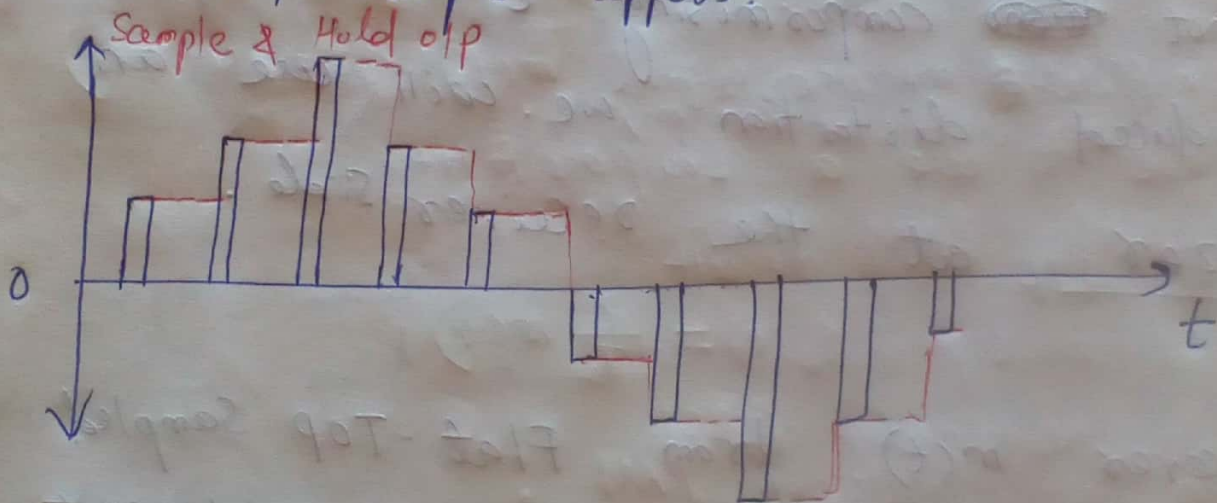
* Signal Recovery through Holding:



operation of Recovery.

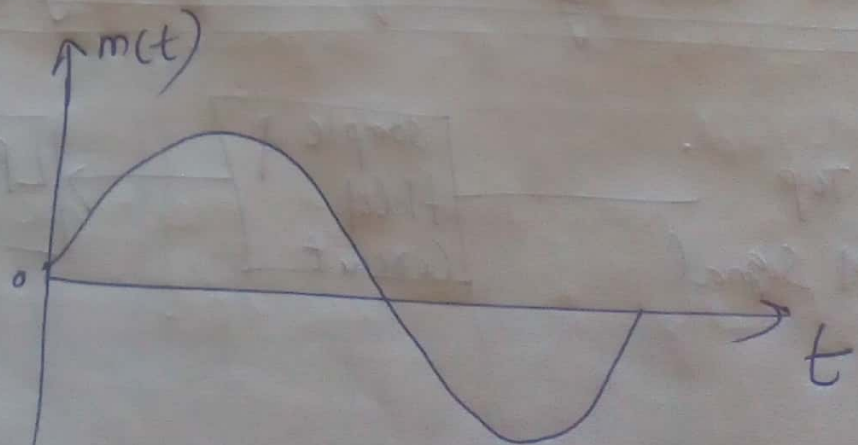


* Sample & Hold circuit will extend the pulse upto the next sample appears.

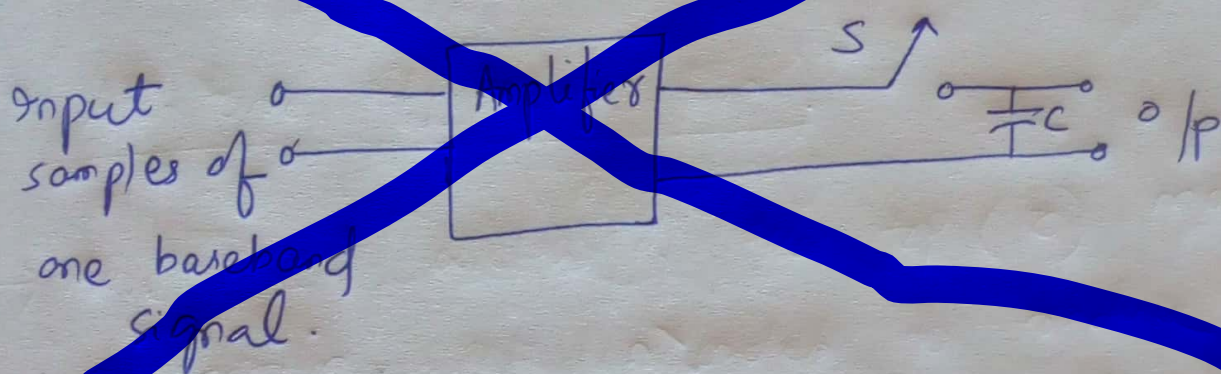


* The Red lined is the output of Sample & Hold ckt.

* After LPF



* Hold ckt uses a Capacitor as below



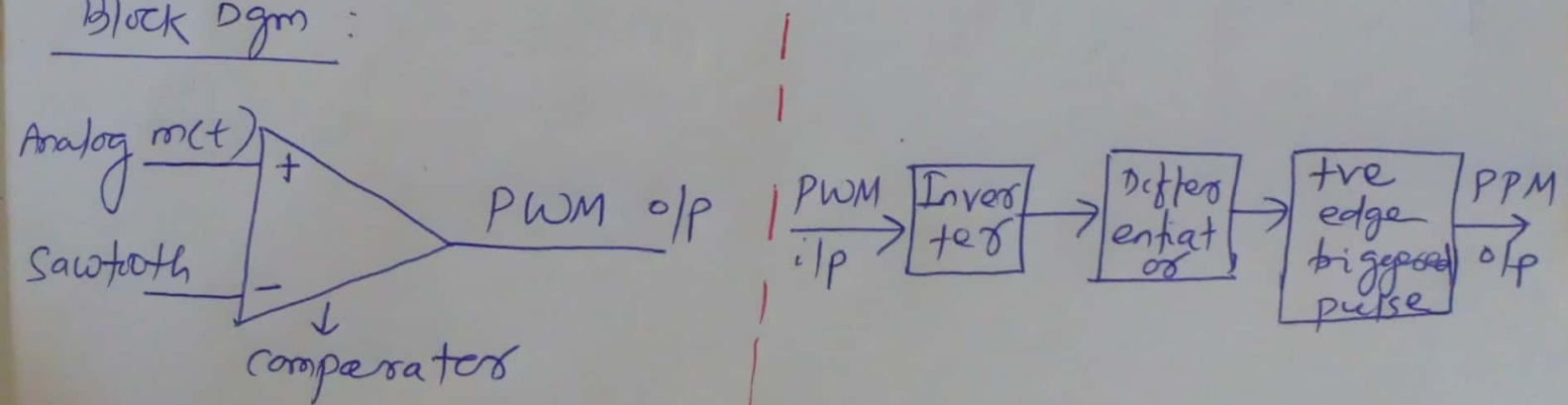
PWM & PPM

PWM \rightarrow In pulse width modulation, width of pulse will be changed according to message ~~signal~~ signal.

PPM \rightarrow In pulse position modulation, the time position of occurrence of pulse will be changed according to the message signal.

Generation of PWM & PPM :

Block Dgm :



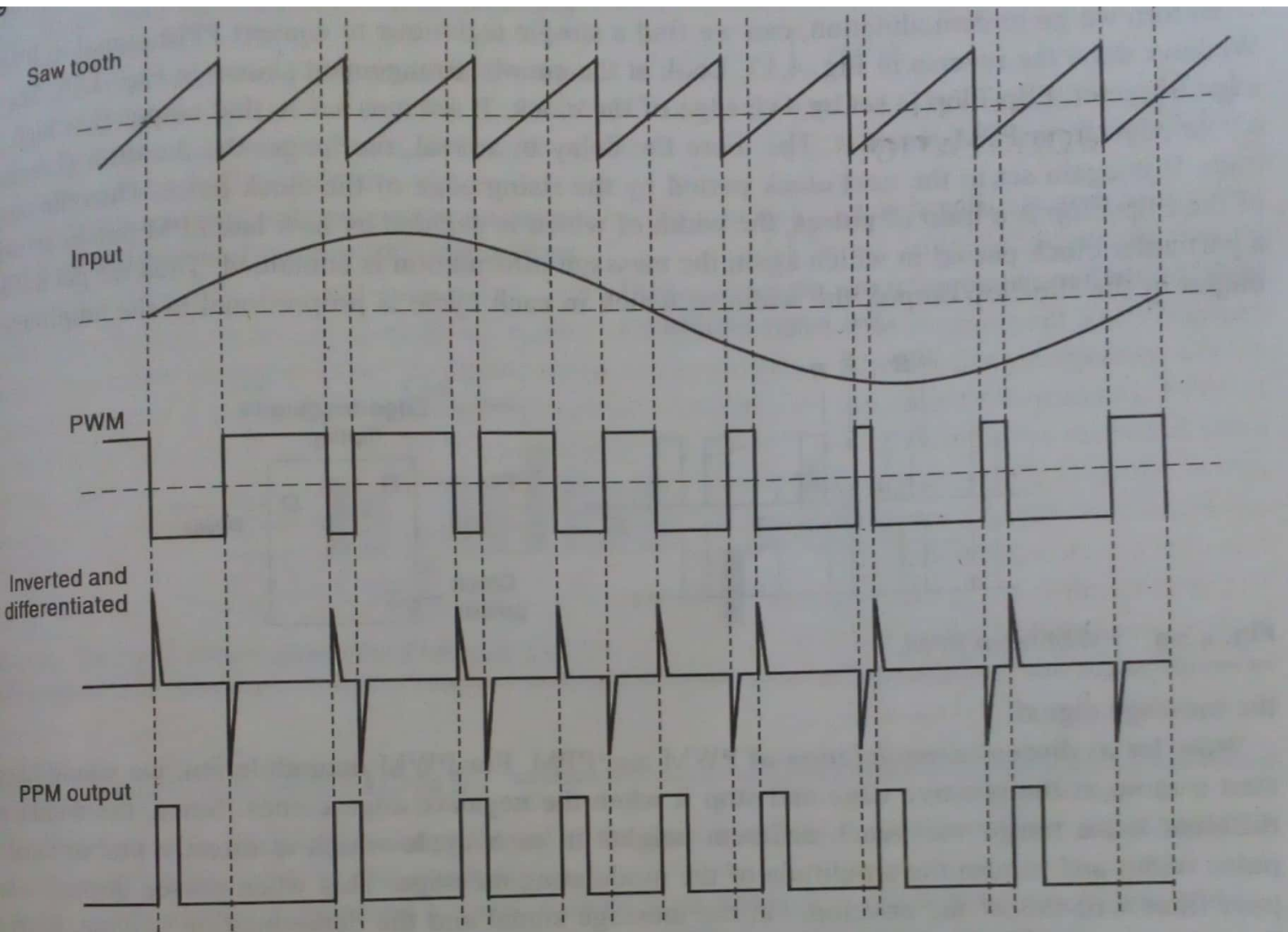
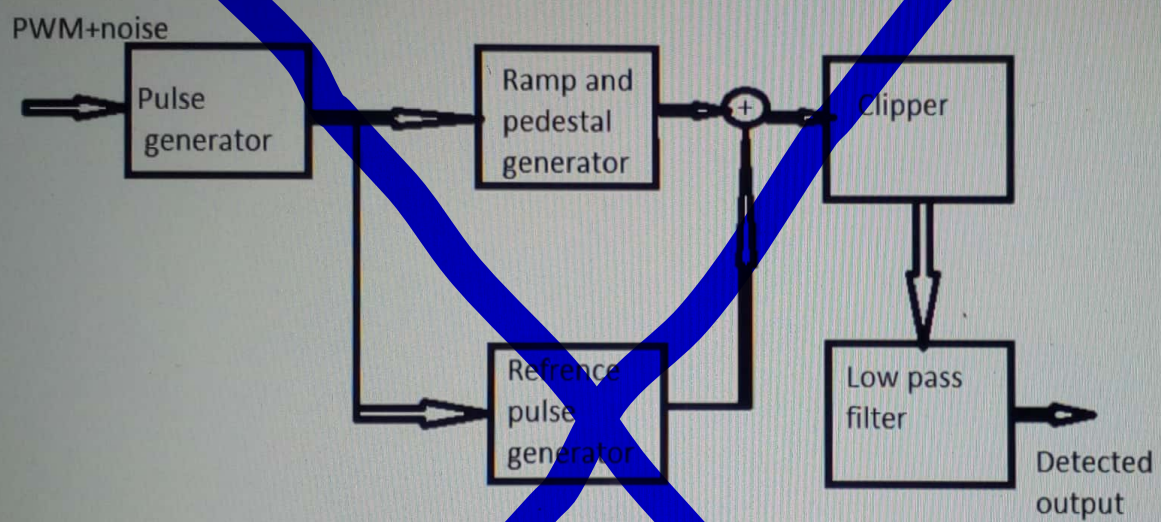
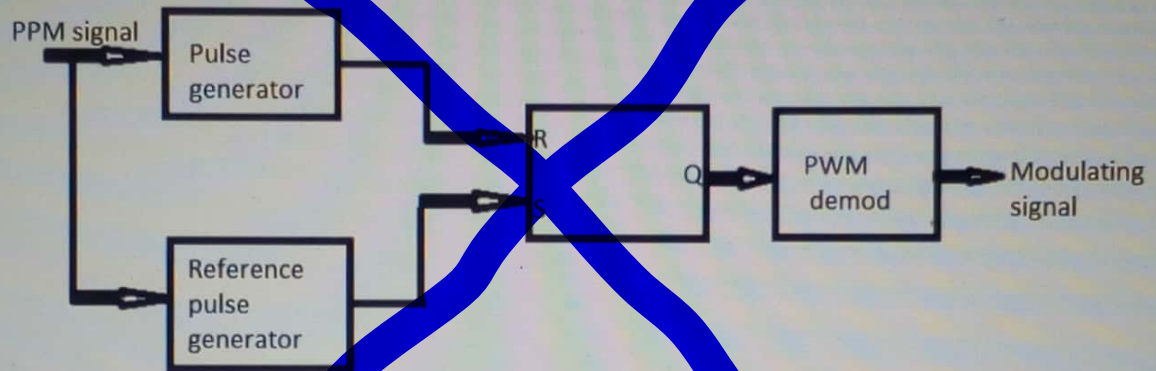


Fig. 4.18 Principle of PWM and PPM generation.

Detection of PWM



Detection of PPM



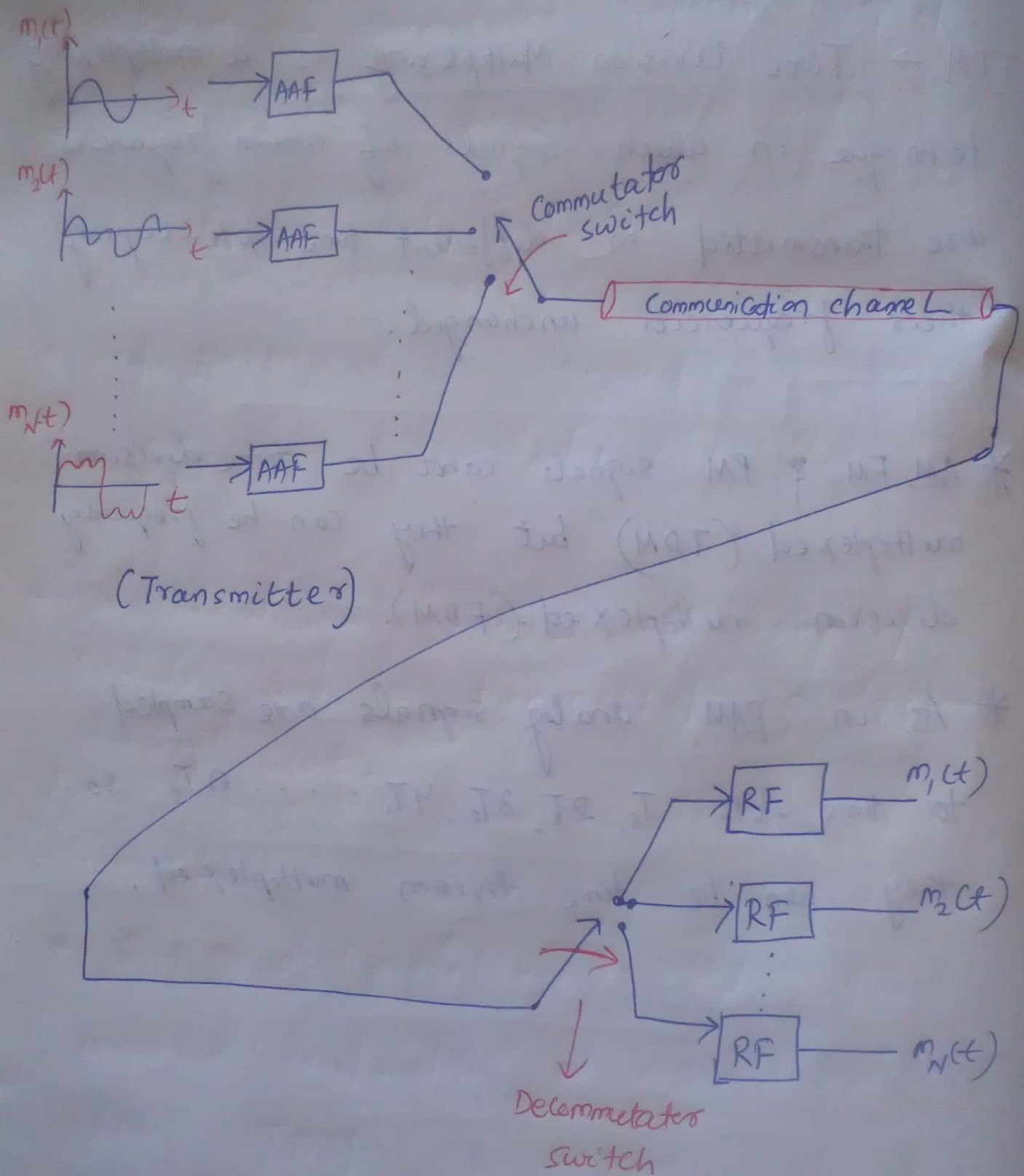
PAM - TDM:

TDM \rightarrow Time Division Multiplexing is a multiplexing technique in which signals of common frequencies are transmitted in different time slots keeping their frequencies unchanged.

* AM, FM & PM signals cannot be time division multiplexed (TDM) but they can be frequency division multiplexed (FDM).

* As in PAM, analog signals are sampled to time slots $T_s, 2T_s, 3T_s, 4T_s, \dots, nT_s$, so they can be time division multiplexed.

Block Diagram of PAM-TDM:



* Here AAF - Anti Aliasing effect

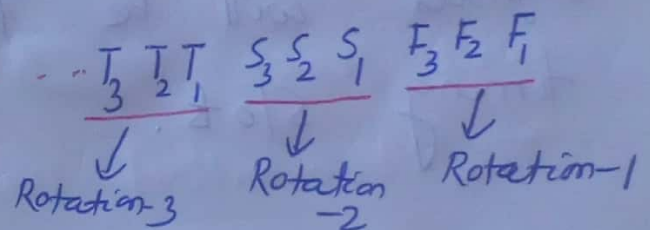
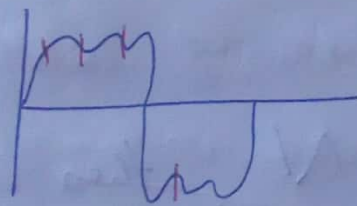
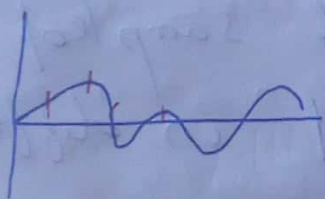
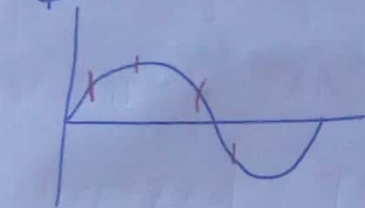
RF - Reconstruction Filter

→ AAF will bandlimit the input signals.

→ Commutator switch will rotate to cut samples from each signal.

Explanation

Let n signals are



* The red spots are samples of each signal.

F_1, F_2, F_3 → First samples of Signal 1, 2, 3 (Let)

S_1, S_2, S_3 → Second samples of "

T_1, T_2, T_3 → Third "

→ If N number of signals, then on first rotation of commutator switch sampled signals will be

$$F_N \dots F_4 F_3 F_2 F_1$$

on 2nd rotation

$$S_N \dots S_4 S_3 S_2 S_1$$

* In communication channel all sampled signals will be transmitted in different timing slots.

Remember:

~~* If N no. of signals will be N , then minimum bandwidth of ~~all~~ all signals after TDM is Nf_m if f_m is the highest frequency of each signal after AAF.~~