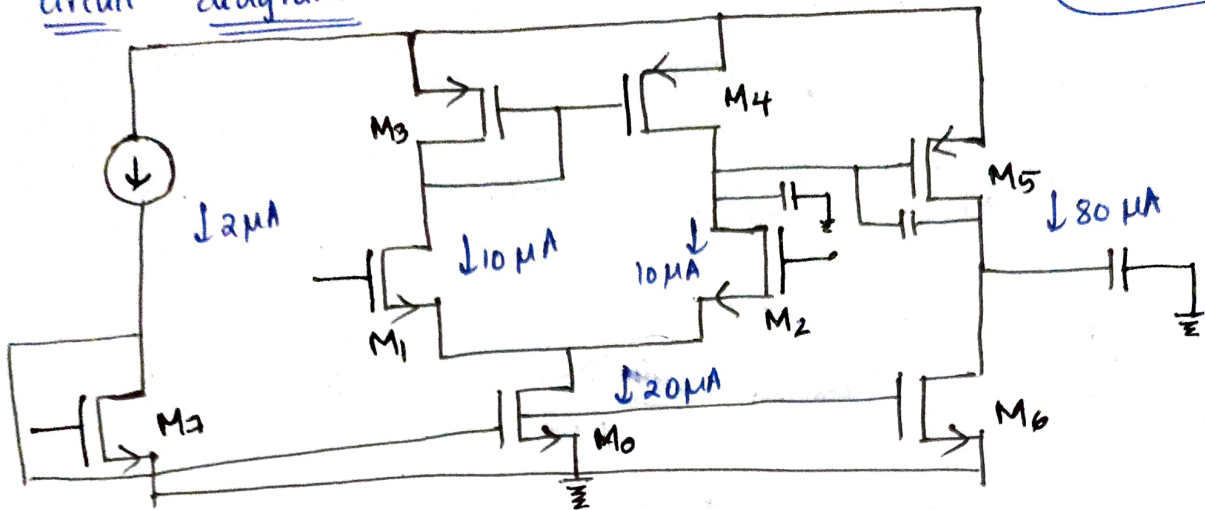


2 stage OTA

Circuit diagram



$$I_{ref} = 20\mu A$$

$$\text{Tail current} = 2 \times 10 = 20\mu A$$

Given

$$(V_{th})_n = 0.37V$$

$$(V_{th})_p = 0.39V$$

$$\mu_{nlox} = 230 \mu A/V^2$$

$$\mu_{plox} = 100 \mu A/V^2$$

$$l_{min} = 0.18 \mu m$$

$$W_{min} = 0.27 \mu m$$

$$\lambda = 0.1 \text{ for } 180nm \text{ technology}$$

$$L_{min} = 0.18 \mu m$$

$$\text{considering } V_{gs} - V_{th} = 200mV \text{ initially}$$

$$L = 5 \times 0.18 \mu m = 0.9 \mu m \approx 1 \mu m \text{ (for all)}$$

$$\frac{(I)_0}{(W/L)_0} = \frac{I_{ref}}{(W/L)_{ref}}$$

$$\frac{20}{(W/L)_0} = \frac{2}{2}$$

$$\boxed{(W/L)_0 = 20}$$

For M_1 and M_2

For M_1 to be in saturation

$$V_{gs} > V_{th}$$

$$0.9 - V_s > 0.37$$

$$\boxed{V_s < 0.53}$$

$$A_v \quad g_m = \frac{2I_D}{V_{gs} - V_{th}}$$

For max gain, high $g_m \Rightarrow$ low $V_{gs} - V_{th}$

$$\text{let } V_s = 0.2V$$

$$V_s = 0.33V$$

Then,

$$10\mu = \frac{1}{2} (230\mu) \left(\frac{W}{L}\right)_{M1} (0.9 - 0.33 - 0.37)^2$$

$$10 = \frac{1}{2} (230) \left(\frac{W}{L}\right)_{M1} (0.2)^2$$

$$\left(\frac{W}{L}\right)_{M1} = \frac{10}{11.5(0.2)^2} = 2.1739$$

$$\left(\frac{W}{L}\right)_{M1} = \left(\frac{W}{L}\right)_{M2} = 2.1739$$

To get max ~~swt~~ $(g_m)_{M1} = \frac{2(10\mu)}{0.2} = 100\mu S$

$$r_o = \frac{1}{\lambda I_D} = \frac{1}{0.1(10\mu)} = 1000k\Omega$$

$$(r_o)_{M1} = (r_o)_{M2} = 1000k\Omega$$

③ For M3 and M4

M3 should be on and M1 should be in saturation

$$(V_{sg})_3 > (V_{th})_p$$

$$1.8 - (V_g)_3 > 0.39$$

$$(V_g)_3 < 1.41$$

$$(V_d)_1 < 1.41V$$

$$(V_g)_3 = (V_d)_1$$

and M1 saturation

$$(V_{ds})_1 > (V_{gs})_1 - (V_{th})_n$$

$$(V_d)_1 > (V_g)_1 - (V_{th})_n$$

$$V_d > 0.9 - 0.37$$

$$(V_d)_1 > 0.53$$

$$(V_d)_1 \in (0.53, 1.41)$$

$$\Rightarrow \boxed{(V_d)_1 = 1.02V}$$

Then,

$$10\mu = \frac{1}{2} (\mu_p \times) \left(\frac{W}{L}\right)_{M3} (V_{sg} - V_{tp})^2$$

$$10\mu = \frac{1}{2} (100\mu) \left(\frac{W}{L}\right)_{M3} (1.8 - 1.02 - 0.39)^2$$

$$\left(\frac{W}{L}\right)_{M3} = \frac{2}{10(0.39)^2} = 1.314$$

$$g_m = \frac{2I_D}{V_{SG} - V_{tp}} = \frac{2(10\mu)}{0.39} = 51.3 \mu S$$

$$(r_o)_{M3} = (r_o)_{M4} = \frac{1}{\lambda I_D} = 1000 k\Omega$$

④ For M5 (consider $80\mu A$ in stage 2)

M5 on

$$(V_{SG})_5 > (V_{th})_p$$

$$1.8 - (V_G)_5 > 0.39$$

$$(V_G)_5 < 1.41$$

$$\text{and } (V_G)_5 > 0.53$$

$$(V_G)_5 \in (0.53, 1.41)$$

$$(V_G)_5 = 1.02V$$

$$80\mu = \frac{1}{2} (100\mu) \left(\frac{W}{L} \right)_5 (1.8 - 1.02 - 0.39)^2$$

$$\left(\frac{W}{L} \right)_5 = 1.314 \times 8 = 10.512$$

$$(g_m)_5 = \frac{2(80\mu)}{0.39} = 410.25 \mu S$$

$$(r_o)_5 = (r_o)_6 = \frac{1}{\lambda I_D} = \frac{1}{0.1 \times 80\mu} = 125 k\Omega$$

⑤ For M6

$$\frac{(I)_6}{(W/L)_6} = \frac{(I)_{nub}}{(W/L)_{nub}}$$

$$\frac{80}{(W/L)_6} = \frac{2}{2}$$

$$(W/L)_6 = 80$$

$$r_o = \frac{1}{\lambda I_D}$$

$$(g_m)_6 = 2(80\mu)$$

Tabular form considering $\lambda = 0.1$

Transistor	w/L	gm	v _{drive}	r _o	i _D
NMOS { M ₁ M ₂	2.1739	100 μ S	0.2V	1000K Ω	20 μ A
	2.1739	100 μ S	0.2V	1000K Ω	
PMOS { M ₃ M ₄	1.314	51.3 μ S	0.39V	1000K Ω	20 μ A
	1.314	51.3 μ S	0.39V	1000K Ω	
PMOS M ₅	10.512	410.25 μ S	0.39V	125K Ω	80 μ A
NMOS M ₆	80	11823 μS 1777 μ S	0.002V 0.09V	125K Ω	80 μ A
tail M ₀	20	444.4 μ S	0.09V	500K Ω	40 μ A
reference M ₇	20	44.44 μ S	0.09V	5000K Ω	2 μ A

Gain of 1st stage

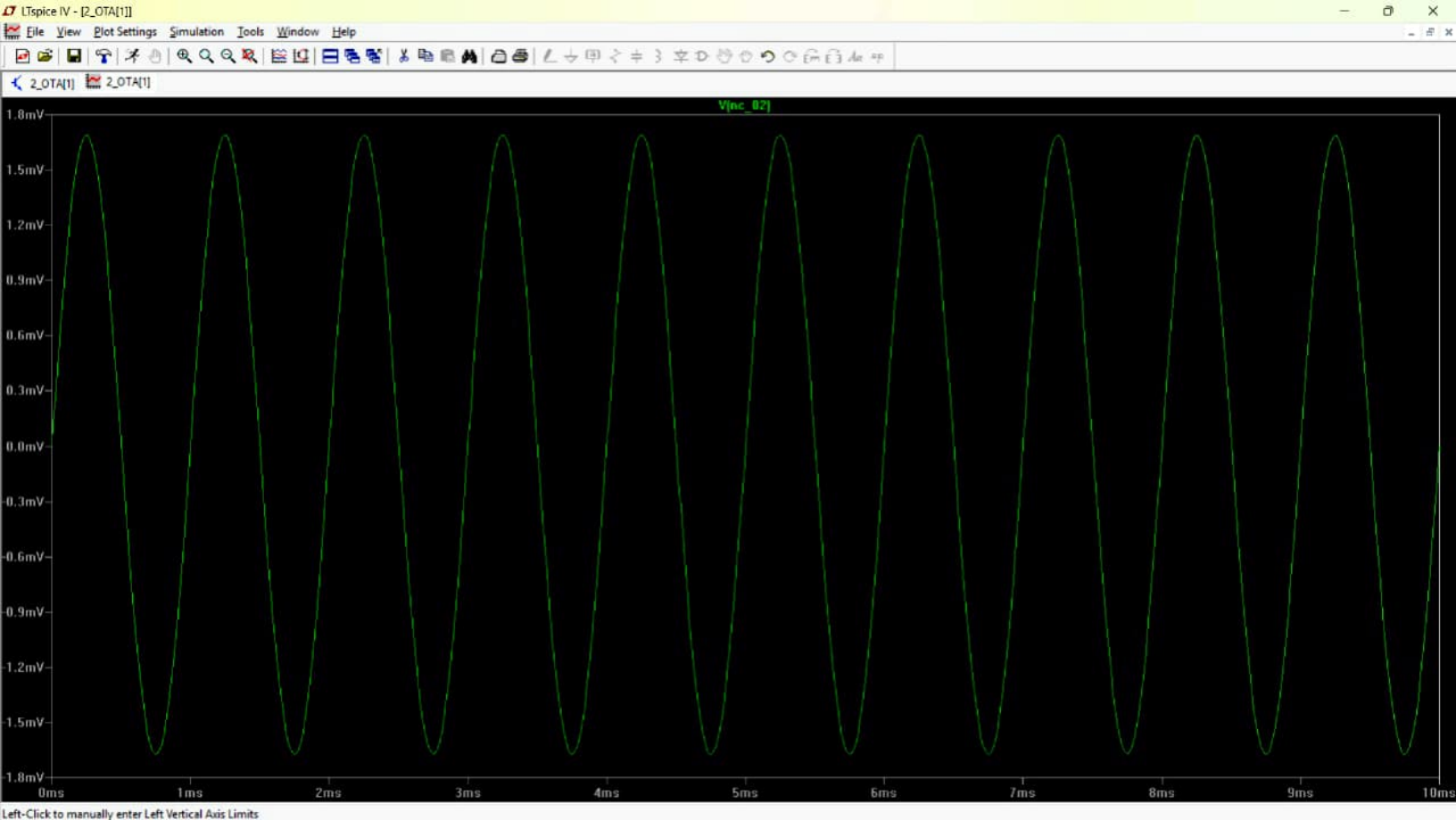
$$(A_v)_1 = \frac{1.7\text{mV} - (-1.7\text{mV})}{20\mu} = \frac{3.4\text{mV}}{20\mu} = 170 = 44.6\text{dB}$$

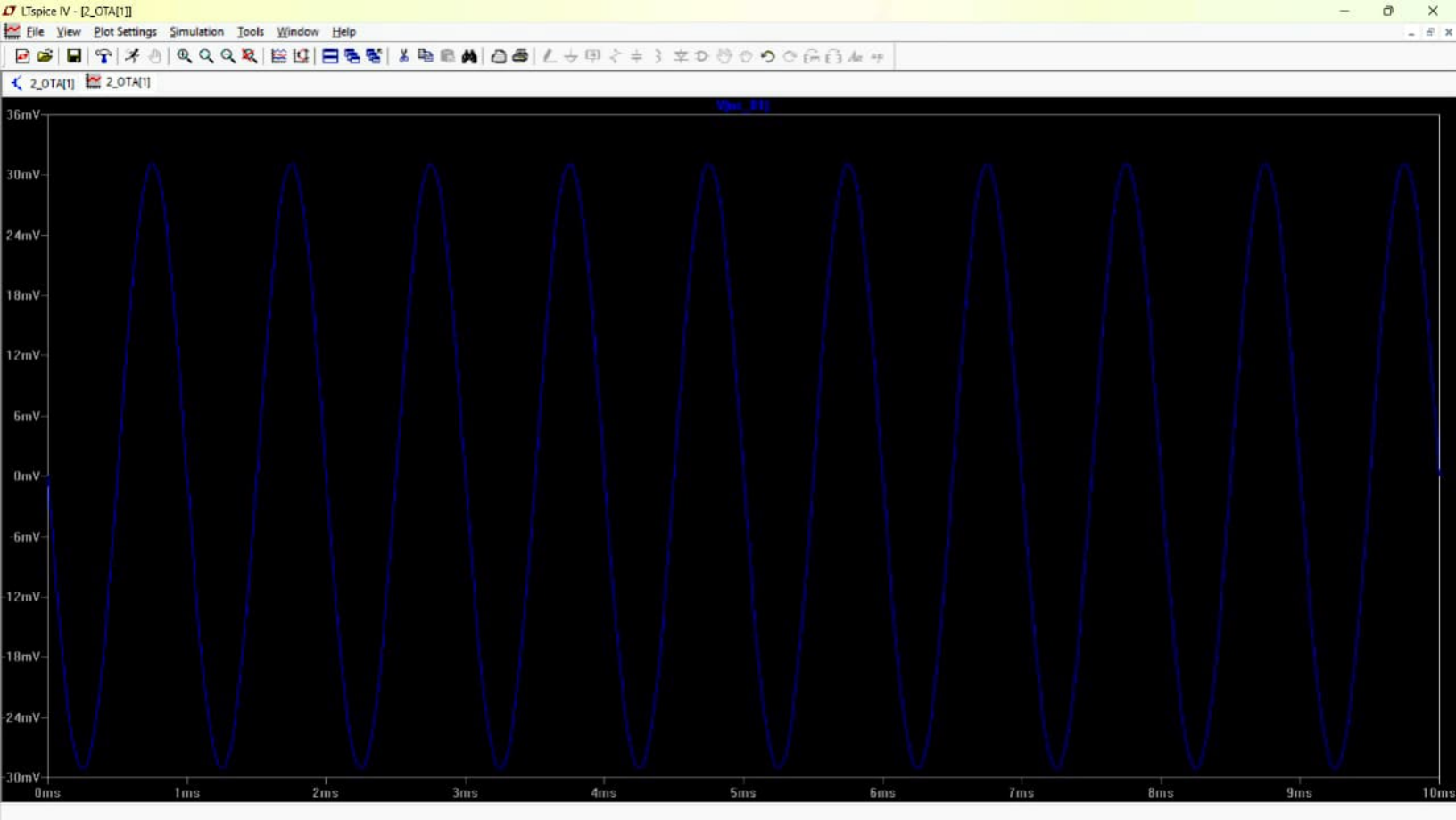
$$\text{2nd stage } (A_v)_2 = \frac{30\text{mV} - (-30\text{mV})}{20\mu} = \frac{60\text{mV}}{20\mu} = 3000 = 69.54\text{dB}$$

Overall gain \rightarrow 114.14dB

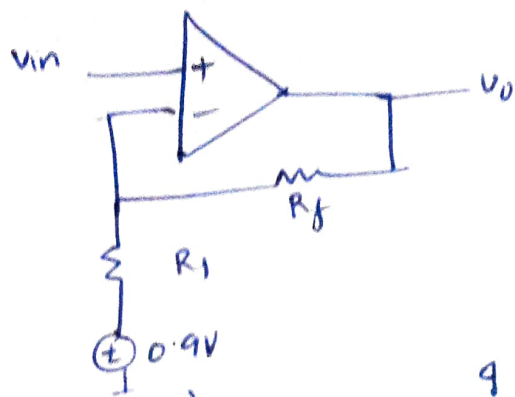
Power consumed \rightarrow 20 μ A + 80 μ A = 102 μ A
V = 1.8V

$$\text{Power} = 183\mu\text{W}$$





⑤ Non-inverting amplifier of gain 2



$$1 + \frac{R_f}{R_1} = 2$$

$$\frac{R_f}{R_1} = 1$$

$$R_f = R_1$$

I considered $R_f = R_1 = 5k\Omega$

$$\text{Gain} = 20 \log(2) = 6 \text{ dB}$$

(I got gain around 5.98 dB)

⑥ For phase margin (55°-70°), tune the following.
→ Miller capacitor used for compensation (C_c) = 4pF

$$\text{Load capacitor } (C_L) = 18.18 \text{ pF} = \frac{C_c}{0.22}$$

$$R_2 \text{ (for zero)} = 10k$$

This ensures that pole 1 (dominant pole) due to C_c and pole due to C_L are far apart from each other.

$$R_2 = \frac{1}{4 \text{ ms}} = \frac{1}{410.25} \mu\text{s} = 3k\Omega$$

→ To get around 60°, I have tuned the C_c, C_L, R_2 .

Final values (To get 62°)

$$C_c = 4 \text{ pF}$$

$$C_L = 18.18 \text{ pF}$$

$$R_2 = 10k\Omega$$

→ From simulation

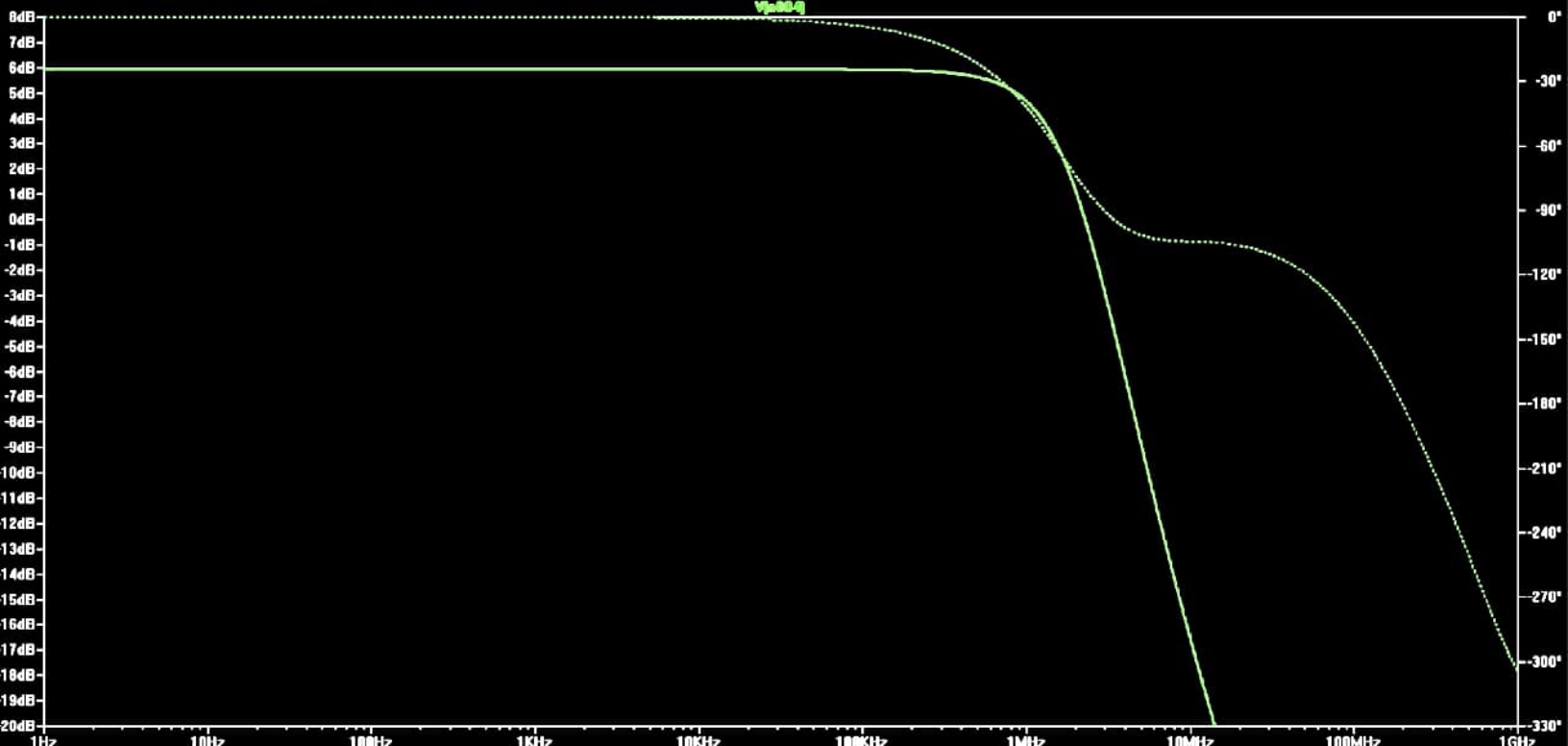
$$\text{Freq at } 0 \text{ dB } (\omega_{gc} = 3.9 \text{ MHz})$$

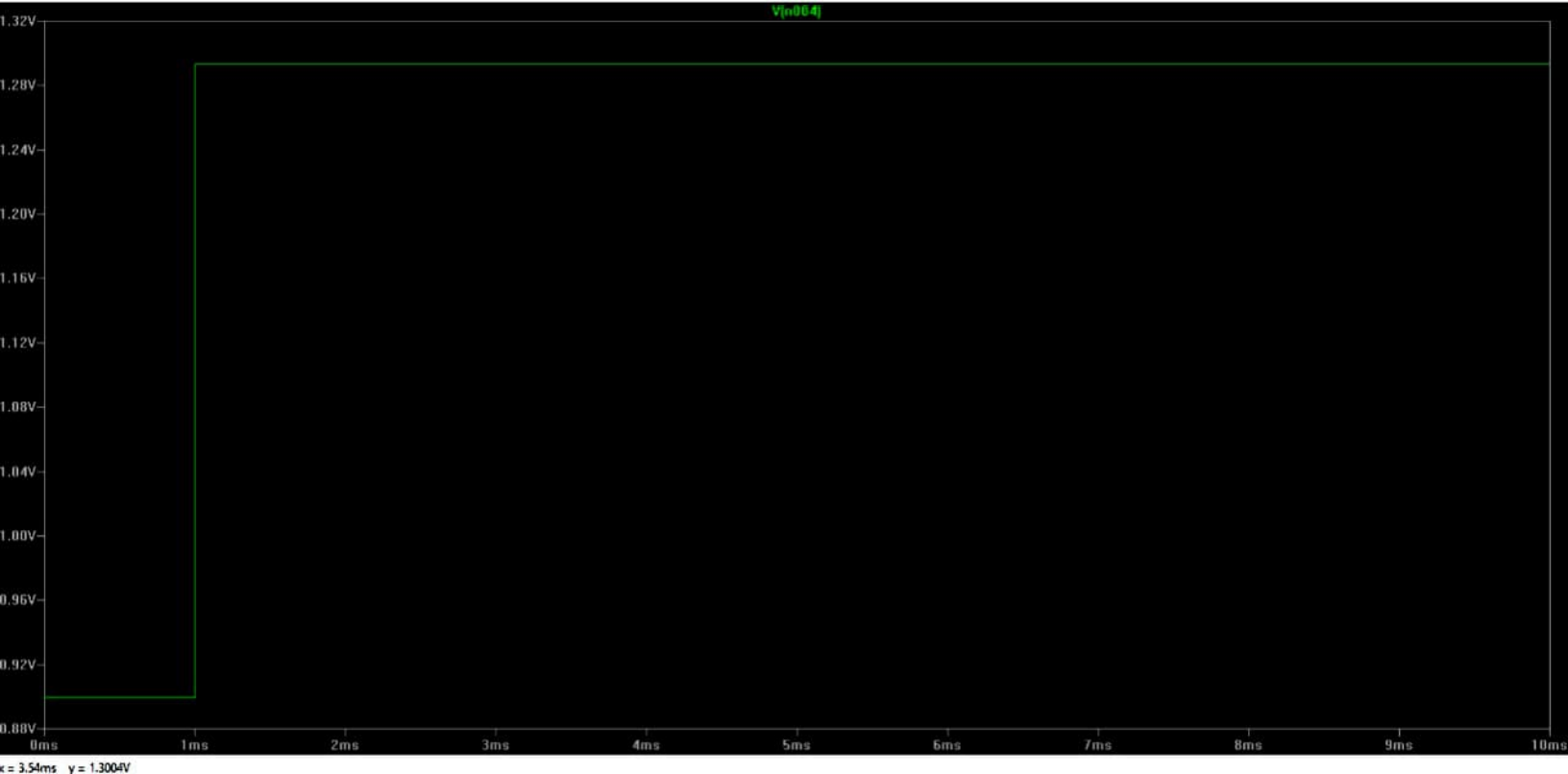
At a that ω_{gc} phase of open loop (GH) = 118°
(in unity gain feedback)

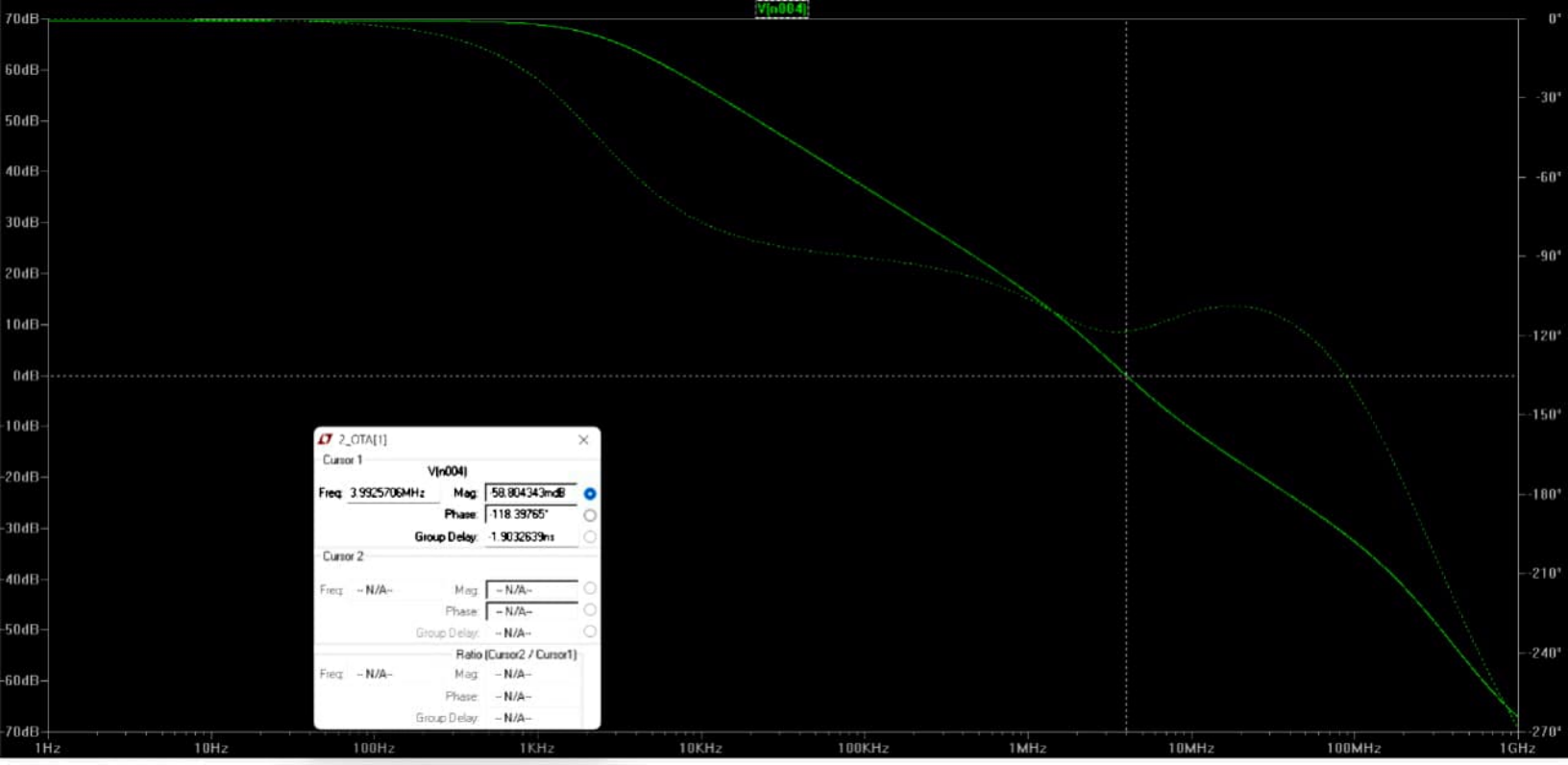
$$\text{Phase margin} = 180^\circ + \phi$$

$$= 180^\circ - 118^\circ$$

$$\boxed{\text{PM} = 62^\circ} \quad \checkmark$$







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