

## Solutions to Examples from Discussion 6 (ECE 213, Spring 2024)

1. For each of the following LTI systems, determine if the system is (i) causal and/or (ii) BIBO stable.

(a) The system is defined by  $y(t) = \int_t^\infty x(\tau) e^{t-\tau} d\tau$ .

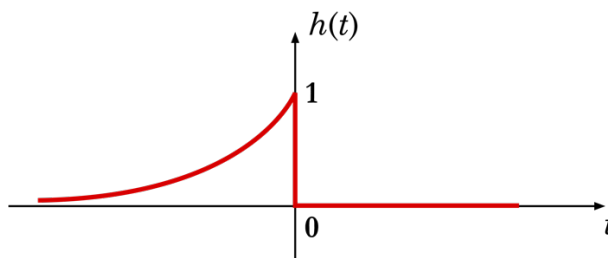
(b) The impulse response is  $h(t) = \frac{1}{(t+1)^2} u(t)$ .

### ANSWERS:

- (a) The impulse response  $h(t)$  is equal to...

$$h(t) = \int_t^\infty \delta(\tau) e^{t-\tau} d\tau = \int_{-\infty}^\infty u(\tau - t) \delta(\tau) e^{t-\tau} d\tau = u(-t)e^t$$

which is shown below...



Since  $h(t)$  is non-zero for  $t < 0$ , this system is **not causal**.

To test for BIBO stability...

$$\int_{-\infty}^\infty |h(t)| dt = \int_{-\infty}^0 e^t dt = 1 = \text{finite}$$

Therefore, this system is **BIBO stable**.

- (b) The impulse response is equal to 0 for  $t < 0$ , so this system is **causal**.

To test for BIBO stability...

$$\int_{-\infty}^\infty |h(t)| dt = \int_0^\infty \frac{dt}{(t+1)^2} = 1 = \text{finite}$$

Therefore, this system is also **BIBO stable**.

## Solutions to Examples from Discussion 6 (ECE 213, Spring 2024)

2. For each of the following non-LTI systems, determine if the system is (i) causal and/or (ii) BIBO stable.

(a)  $y(t) = \int_t^{2t} x(\tau) d\tau$

(b)  $y(t) = x(|t|)$

### ANSWERS:

Because these systems are not LTI, causality and stability cannot be found as above. Instead, we will use the definitions.

- (a) The output at time  $t$  depends on the input between time  $t$  and time  $2t$ . For  $t > 0$ , this time frame is in the future, so this system is **not causal**.

To test for BIBO stability, let's choose a simple bounded input  $x(t) = C$ , where  $C$  is non-zero. The corresponding output is...

$$y(t) = \int_t^{2t} C d\tau = Ct$$

The magnitude of  $y(t)$  goes infinite as  $t$  goes infinite, so this system is **not BIBO stable**.

- (b) For  $t < 0$ , the output depends on the input at a positive time, so this system is **not causal**.

To test for BIBO stability, assume that the input is bounded, i.e.,

$$|x(t)| < C$$

where  $C > 0$ . Therefore...

$$|y(t)| = |x(|t|)| < C$$

Since the output is bounded for a bounded input, this system is **BIBO stable**.

## Solutions to Examples from Discussion 6 (ECE 213, Spring 2024)

3. Consider the following system defined by an LCCDE:

$$\frac{d^2 y}{dt^2} + 5 \frac{dy}{dt} + 4 y(t) = \frac{d^2 x}{dt^2} + 3 \frac{dx}{dt} + 6 x(t)$$

(a) Find the frequency response function  $H(\omega)$ .

(b) Find  $y(t)$  for  $x(t) = 10 \sin(2t) u(t)$ .

### ANSWERS:

As shown in class, using an input  $x(t) = e^{j\omega t}$ , the output is  $y(t) = H(\omega) e^{j\omega t}$ , where  $H(\omega)$  is found by replacing the  $n$ th derivative of  $x$  with  $(j\omega)^n x(t)$  and replacing the  $n$ th derivative of  $y$  with  $(j\omega)^n y(t)$ .

(a) Therefore, the LCCDE becomes...

$$(j\omega)^2 y(t) + 5 j\omega y(t) + 4 y(t) = (j\omega)^2 x(t) + 3 j\omega x(t) + 6 x(t)$$

or...

$$y(t) = \frac{(j\omega)^2 + 3j\omega + 6}{(j\omega)^2 + 5j\omega + 4} x(t)$$

So, the frequency response function is...

$$H(\omega) = \frac{(j\omega)^2 + 3j\omega + 6}{(j\omega)^2 + 5j\omega + 4}$$

(b) Rewriting the input as a linear combination of complex exponentials...

$$x(t) = 10 \sin(2t) u(t) = \frac{10}{j2} (e^{j2t} - e^{-j2t}) u(t)$$

... the response becomes...

$$\begin{aligned} y(t) &= \frac{10}{j2} (H(2) e^{j2t} - H(-2) e^{-j2t}) u(t) \\ &= \frac{10}{j2} \left( \frac{(j2)^2 + 3(j2) + 6}{(j2)^2 + 5(j2) + 4} e^{j2t} - \frac{(-j2)^2 - 3(-j2) + 6}{(-j2)^2 - 5(-j2) + 4} e^{-j2t} \right) u(t) \end{aligned}$$

Note that this is a purely real-valued function! But it needs to be simplified (a lot) to see it. The result is...

$$y(t) = 6 \sin(2t) u(t) - 2 \cos(2t) u(t)$$

## Solutions to Examples from Discussion 6 (ECE 213, Spring 2024)

4. Consider an LTI system with impulse response  $h(t) = 6 e^{-2t} u(t)$ .

Find the output  $y(t)$  when the input is (a)  $x(t) = 8 u(t)$  and (b)  $x(t) = 10 \sin 4t u(t)$ .

### ANSWERS:

First, find the frequency response function  $H(\omega)$  using the following integral...

$$H(\omega) = \int_0^{\infty} h(t) e^{-j\omega t} dt = \int_0^{\infty} 6 e^{-2t} e^{-j\omega t} dt = 6 \int_0^{\infty} e^{(-2-j\omega)t} dt = \frac{6}{2+j\omega}$$

For (a), recognize that 8 can be written as  $8 e^{j0t}$ , i.e., use  $H(0)$  to find the output. The result is...

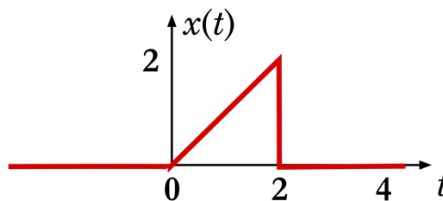
$$y(t) = 8 H(0) e^{j0t} u(t) = 8 \frac{6}{2+j0} e^{j0t} u(t) = 24 u(t)$$

For (b), write  $10 \sin 4t$  in terms of complex exponentials, then find the output for each separately and combine them. The result is...

$$\begin{aligned} y(t) &= \frac{10}{j2} (H(4) e^{j4t} - H(-4) e^{-j4t}) u(t) = \frac{10}{j2} \left( \frac{6}{2+j4} e^{j4t} - \frac{6}{2-j4} e^{-j4t} \right) u(t) \\ &= \frac{10}{j2} \frac{1}{20} (6(2-j4) e^{j4t} - 6(2+j4) e^{-j4t}) u(t) = 6 \sin 4t u(t) - 12 \cos 4t u(t) \end{aligned}$$

Note that this result is purely real, because the input is purely real.

5. A signal  $x(t)$  is represented below. What is its Laplace Transform  $X(s)$ ?



### ANSWER:

First, rewrite  $x(t)$  using unit step functions...

$$x(t) = r(t) - r(t-2) - 2u(t-2) = t u(t) - (t-2) u(t-2) - 2u(t-2)$$

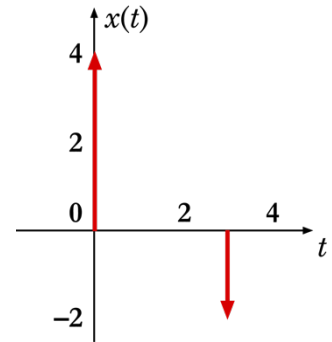
Next, transform each  $u(t)$  and  $t u(t)$ , then use the time-shift property (of the Laplace Transform) to get...

$$X(s) = \frac{1}{s^2} - \frac{1}{s^2} e^{-2s} - \frac{1}{s} e^{-2s}$$

## Solutions to Examples from Discussion 6 (ECE 213, Spring 2024)

6. A signal is represented to the right. The arrows in the figure are impulse functions, with each value representing the area “below” it.

What is its Laplace Transform  $X(s)$ ?



**ANSWER:**

First, write  $x(t)$  using impulse functions...

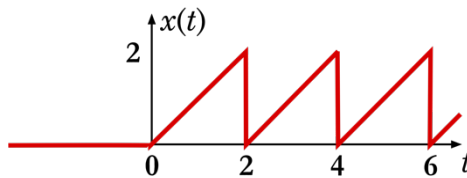
$$x(t) = 4\delta(t) - 2\delta(t - 3)$$

Next, transform each  $\delta(t)$ , then use the time-shift property to get...

$$X(s) = 4 - 2e^{-3s}$$

7. For each of the following signals  $x(t)$ , find its Laplace Transform  $X(s)$ .

(a) The causal periodic signal shown below.



**ANSWER:**

This is the same signal as in problem 5, repeated with a period of  $T = 2$ . For a causal periodic function, as here,  $X(s)$  is the Laplace Transform of the function from 0 to  $T$  divided by  $1 - e^{-Ts}$ . Therefore...

$$X(s) = \frac{\frac{1}{s^2} - \frac{1}{s^2} e^{-2s} - \frac{1}{s} e^{-2s}}{1 - e^{-2s}}$$

(b)  $x(t) = e^{-5t} \sin(2t) u(t)$

**ANSWER:**

Use the known Laplace Transform of  $\sin(2t) u(t)$  and the frequency-shifting property to get...

$$X(s) = \frac{2}{(s + 5)^2 + 2^2}$$

[continued]

## Solutions to Examples from Discussion 6 (ECE 213, Spring 2024)

7. [continued] For each of the following signals  $x(t)$ , find its Laplace Transform  $X(s)$ .

(c)  $x(t) = 6 e^{-2t} u(t - 5)$

**ANSWER:**

First, rewrite the exponential, so that the time shift is more obvious...

$$x(t) = 6 e^{-2(t-5)-10} u(t - 5)$$

Next, simplify the expression...

$$x(t) = 6 e^{-10} e^{-2(t-5)} u(t - 5)$$

In other words, there is an overall factor of  $6 e^{-10}$ , then we have a frequency shift of 2 and a time shift of 5. We do the frequency shift first, i.e., find the Laplace Transform of  $e^{-2t} u(t)$ , then add a factor of  $e^{-Ts}$  for the time shift of  $T = 5$  to get...

$$X(s) = 6 e^{-10} \frac{1}{s + 2} e^{-5s}$$

(d)  $x(t) = t \cos(2t) u(t)$

**ANSWER:**

Let  $x_1(t) = \cos(2t) u(t)$ . The Laplace Transform of  $t x_1(t)$  is found using the frequency-derivative property, that is...

$$X(s) = -\frac{dX_1}{ds} = -\frac{d}{ds} \left( \frac{s}{s^2 + 2^2} \right) = \frac{2s^2}{(s^2 + 4)^2} - \frac{1}{s^2 + 4}$$

(e)  $x(t) = \frac{1-e^{-2t}}{t} u(t)$

**ANSWER:**

Let  $x_1(t) = (1 - e^{-2t}) u(t)$ . The Laplace Transform of  $x_1(t)/t$  is found using the frequency-integral property, that is...

$$X(s) = \int_s^\infty X_1(s') ds' = \int_s^\infty \left( \frac{1}{s'} - \frac{1}{s' + 2} \right) ds' = \ln \left( \frac{s + 2}{s} \right)$$

[continued]

## Solutions to Examples from Discussion 6 (ECE 213, Spring 2024)

7. [continued] For each of the following signals  $x(t)$ , find its Laplace Transform  $X(s)$ .

(f)  $x(t) = \frac{d}{dt}[e^{-5t} \sin(2t) u(t)]$

**ANSWER:**

Let  $x_1(t) = e^{-5t} \sin(2t) u(t)$ . The Laplace Transform of the time-derivative of  $x_1(t)$  is...

$$X(s) = sX_1(s) - x_1(0^-) = s \frac{2}{(s+5)^2 + 2^2} - 0 = \frac{2s}{(s+5)^2 + 2^2}$$

Note that this is **much** easier than taking the derivative of  $e^{-5t} \sin(2t) u(t)$ , then transforming it.

(g)  $x(t) = \text{rect}\left(\frac{1}{2}(t-1)\right) * e^{-5t} u(t)$

**ANSWER:**

The Laplace Transform of  $x_1(t) * x_2(t)$  is simply the product of the two individual Laplace Transforms, i.e.,  $X_1(s) X_2(s)$ . (This is one of the reasons we LOVE the Laplace Transform!) Therefore, turn the time-scaled (signal-expanded), time-shifted rect function into a difference of two unit-step functions...

$$x_1(t) = \text{rect}\left(\frac{1}{2}(t-1)\right) = u(t) - u(t-2)$$

... then convolve them with  $e^{-5t} u(t)$  to find the Laplace Transform...

$$X(s) = \left(\frac{1}{s} - \frac{1}{s} e^{-2s}\right) \cdot \frac{1}{s+5} = \frac{1 - e^{-2s}}{s(s+5)}$$