



Chapter 4:

# Gate Level Minimization

# Introduction

- Complexity of digital circuit & complexity of algebraic expression.
  - A function's truth-table representation is unique; its algebraic expression is not.
  - Simplification by algebraic means is awkward (from algorithmic point of view);
  - Algebraic manipulations is hard since there is not a uniform way of doing it
- The map method called **Karnaugh map (or K-map)** [M. Karnaugh, 1953] is straightforward and commonly used.

# K-map

- A diagram made up of **squares** each representing one **minterm**.
  - A pictorial form of the truth table;
  - A visual diagram of all possible ways a function may be expressed—the simplest one can easily be identified.
- It produces a circuit diagram with a **minimum** number of **gates** and the **minimum** number of **inputs** to the **gate**.
- It is sometimes possible to find **two or more** expressions that satisfy the minimization criteria.

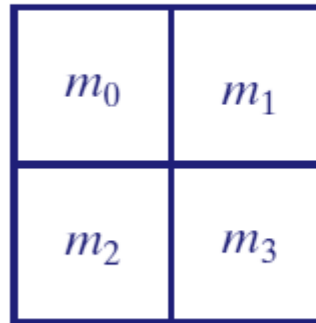
Logic circuit  $\Leftrightarrow$  Boolean function  $\Leftrightarrow$  Truth table  $\Leftrightarrow$  **K-map**  
 $\Leftrightarrow$  Canonical form (sum of minterms, product of maxterms)  
 $\Leftrightarrow$  **(Simplifier) standard form** (sum of products, product of sums)

# Two-Variable K-Map

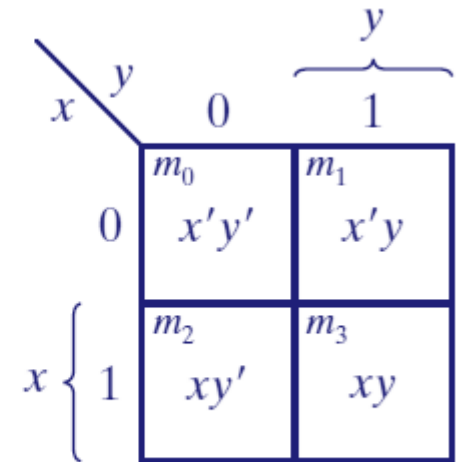
- Two-variable has **four minterms**, and consists of **four squares**.

$$m_1 + m_2 + m_3 = x'y + xy' + xy = x + y$$

- ✓ Four Minterms
- ✓ Two variables
- ✓ Four squares for four minterms



(a)



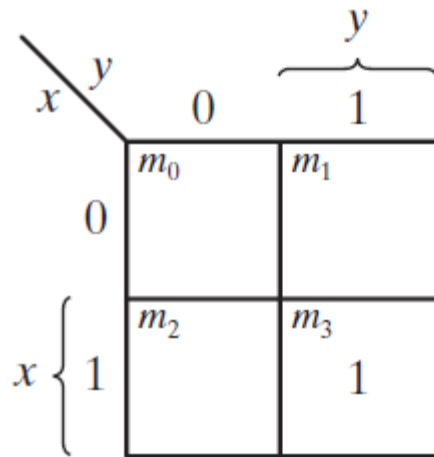
(b)

- ✓ Useful for representation of 16 Boolean functions.

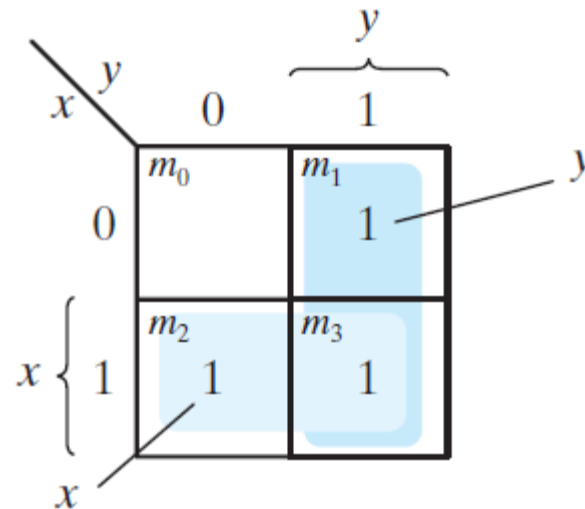
## Example:

- i. If  $xy$  is equal to  $m_3$ , a 1 is placed inside the square that belongs to  $m_3$ .
- ii. If  $m_1 = m_2 = m_3 = 1$  then

$$m_1 + m_2 + m_3 = x'y + xy' + xy = x + y \text{ (OR function)}$$



(a)  $xy$



(b)  $x + y$

# Three-Variable K-Map

$m_0$	$m_1$	$m_3$	$m_2$
$m_4$	$m_5$	$m_7$	$m_6$

(a)

		$y$				
		$yz$	00	01	11	10
$x$	0	$m_0$ $x'y'z'$	$m_1$ $x'y'z$	$m_3$ $x'yz$	$m_2$ $x'yz'$	
	1	$m_4$ $xy'z'$	$m_5$ $xy'z$	$m_7$ $xyz$	$m_6$ $xyz'$	
			$z$			

(b)

- ✓ There are 8 minterms for 3 variables.
- ✓ So, there are 8 squares.
- ✓ Minterms are arranged not in a binary sequence, but in sequence similar to the Gray code.

## Cont...

- Two adjacent squares differs by one variable (one primed other is not).
  - So, they can be minimized easily

### Example:

$$m_5 + m_7 = xy'z + xyz = xz(y' + y) = xz$$

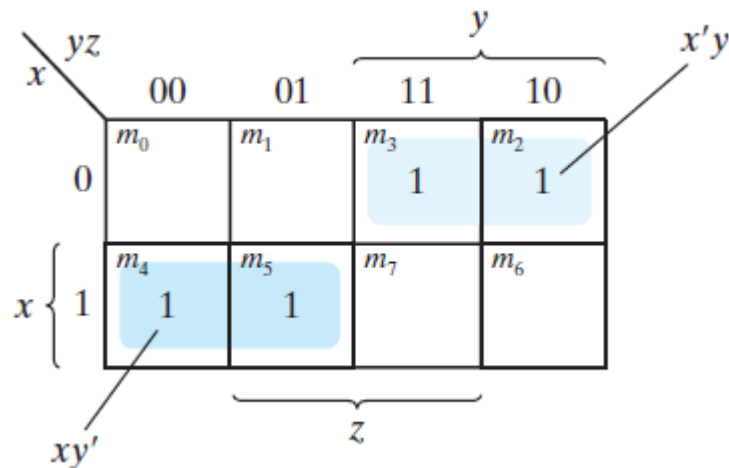
Any two minterms in adjacent squares (vertically or horizontally, but not diagonally, adjacent) that are ORed together will cause a removal of the dissimilar variable.



# Example

- Simplify the Boolean function

$$F(x, y, z) = \sum m(2, 3, 4, 5)$$



$$F(x, y, z) = x'y + xy'$$

## Steps

- a 1 is marked in each minterm square that represents the function.
- Find possible adjacent squares (Two shaded rectangles, each enclosing two 1's)
- The sum of four minterms can be replaced by a sum of only two product terms.

## Cont...

- In some cases, some adjacent squares don't touch each other.

- E.g

- *$m_0$  is adjacent to  $m_2$  and  $m_4$  is adjacent to  $m_6$ .*

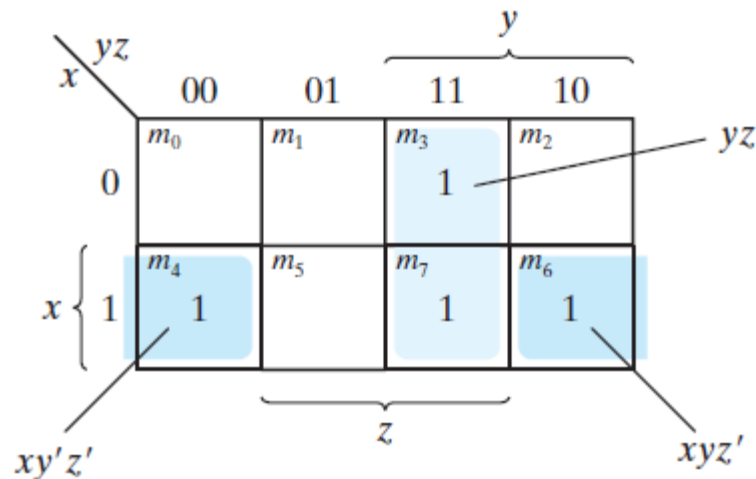
- i.  $m_0 + m_2 = x'y'z' + x'yz' = x'z'(y' + y) = x'z'$

- ii.  $m_4 + m_6 = xy'z' + xyz' = xz'(y' + y) = xz'$

# Example

- Simplify the Boolean function

$$F(x,y,z) = \sum m(3,4,6,7)$$



$$F(x,y,z) = yz + xz'$$

## Example

- A combination of four adjacent squares in the three-variable map represents the logical sum of four minterms and results in an expression with only one literal.

$$\begin{aligned}m_0 + m_2 + m_4 + m_6 &= x'y'z' + x'yz' + xy'z' + xyz' \\&= x'z'(y' + y) + xz'(y' + y) \\&= x'z' + xz' \\&= z'(x' + x) \\&= z'\end{aligned}$$

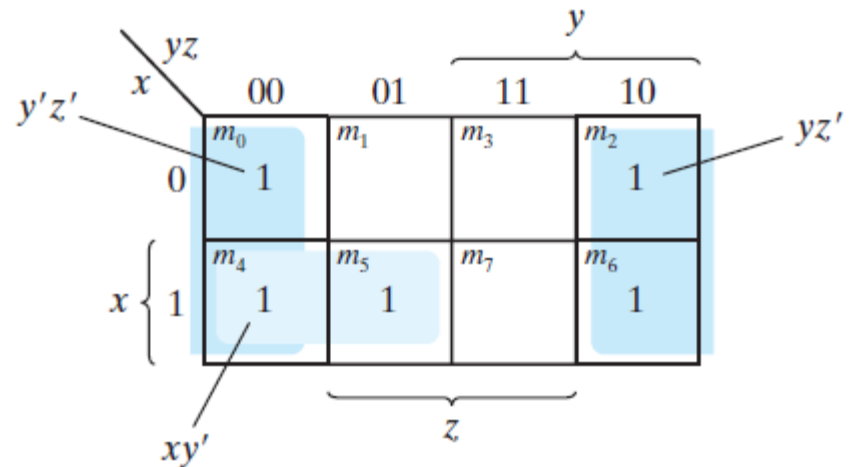
# Example

- Simplify the Boolean function

$$F(x,y,z) = \sum m(0,2,4,5,6)$$

*Sol*

$$F(x,y,z) = z' + xy'$$



# Summary

- ▶ On a 3-variable K-Map:
  - One square represents a minterm with three variables
  - Two adjacent squares represent a product term with two variables
  - Four “adjacent” squares represent a product term with one variable
  - Eight “adjacent” squares is the function of all ones (logic 1).

- In general,
  - The more the squares combined, the fewer the literals of a product term
  - Overlap is allowed.

## Example

$$F = AB'C' + AB'C + ABC + ABC' + A'B'C + A'BC'$$

		BC			
		00	01	11	10
A	0	0	1	0	1
	1	1	1	1	1

$$F = A + B'C + BC'$$

## Example

■  $F(x, y, z) = \sum m(2, 3, 6, 7)$

		Y			
		00	01	11	10
$\frac{yz}{x}$	0			1	1
	1			1	1

Z

- Applying the Minimization Theorem three times:

$$\begin{aligned} F(x, y, z) &= \bar{x} y z + x y z + \bar{x} y \bar{z} + x y \bar{z} \\ &= yz + y\bar{z} \\ &= y \end{aligned}$$

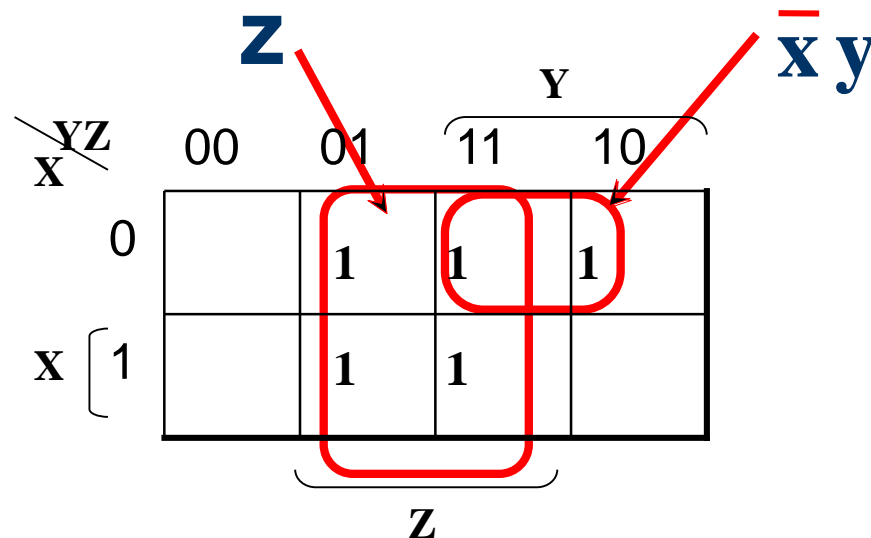


## Example:

- Simplify

$$F(X, Y, Z) = X'Z + X'Y + XY'Z + YZ$$

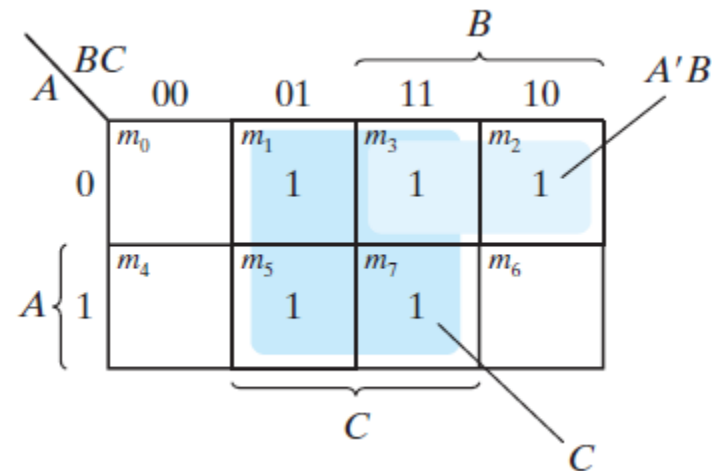
$$F(X, Y, Z) = \Sigma_m (1, 2, 3, 5, 7)$$



# Example

■  $F = A'C + A'B + AB'C + BC$  (Boolean function)

- Express  $F$  as a sum of minterms.
- Find the minimal sum-of-products expression



■ *Sol*

- $F(A,B,C) = \sum m(1,2,3,5,7)$
- $F = C + A'B$

# Four-Variable Map

- 16 minterms (and squares) for 4 variables.

$m_0$	$m_1$	$m_3$	$m_2$
$m_4$	$m_5$	$m_7$	$m_6$
$m_{12}$	$m_{13}$	$m_{15}$	$m_{14}$
$m_8$	$m_9$	$m_{11}$	$m_{10}$

(a)

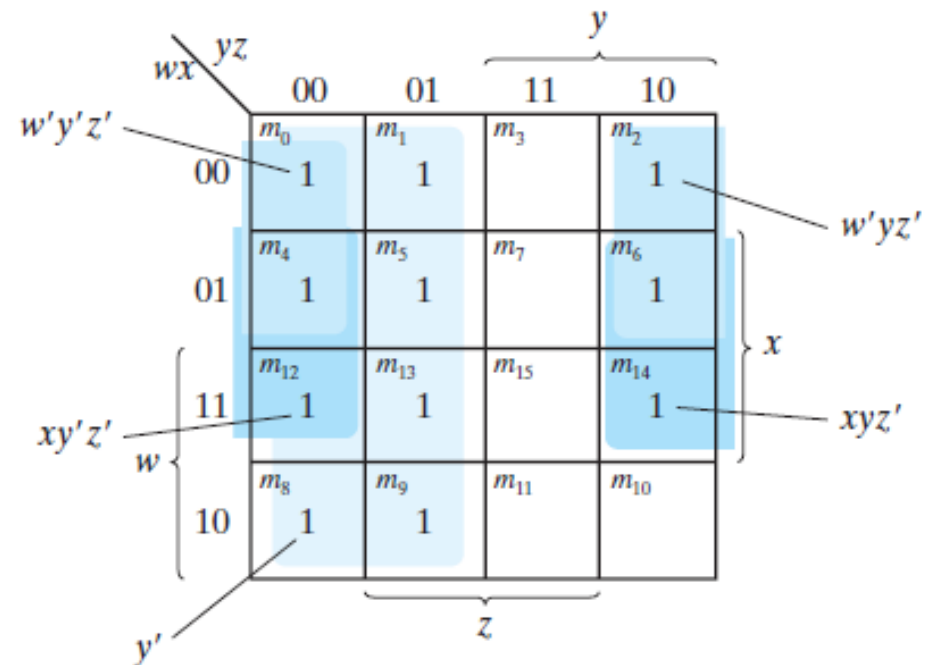
		$y$			
		$yz$	00	01	11
$w$	00	$m_0$ $w'x'y'z'$	$m_1$ $w'x'y'z$	$m_3$ $w'x'yz$	$m_2$ $w'x'yz'$
	01	$m_4$ $w'xy'z'$	$m_5$ $w'xy'z$	$m_7$ $w'xyz$	$m_6$ $w'xyz'$
	11	$m_{12}$ $wxy'z'$	$m_{13}$ $wxy'z$	$m_{15}$ $wxyz$	$m_{14}$ $wxyz'$
	10	$m_8$ $wx'y'z'$	$m_9$ $wx'y'z$	$m_{11}$ $wx'yz$	$m_{10}$ $wx'yz'$
		$z$			

(b)

- ✓ One square represents one minterm, giving a term with four literals.
- ✓ Two adjacent squares represent a term with three literals.
- ✓ Four adjacent squares represent a term with two literals.
- ✓ Eight adjacent squares represent a term with one literal.
- ✓ Sixteen adjacent squares produce a function that is always equal to 1.

# Example

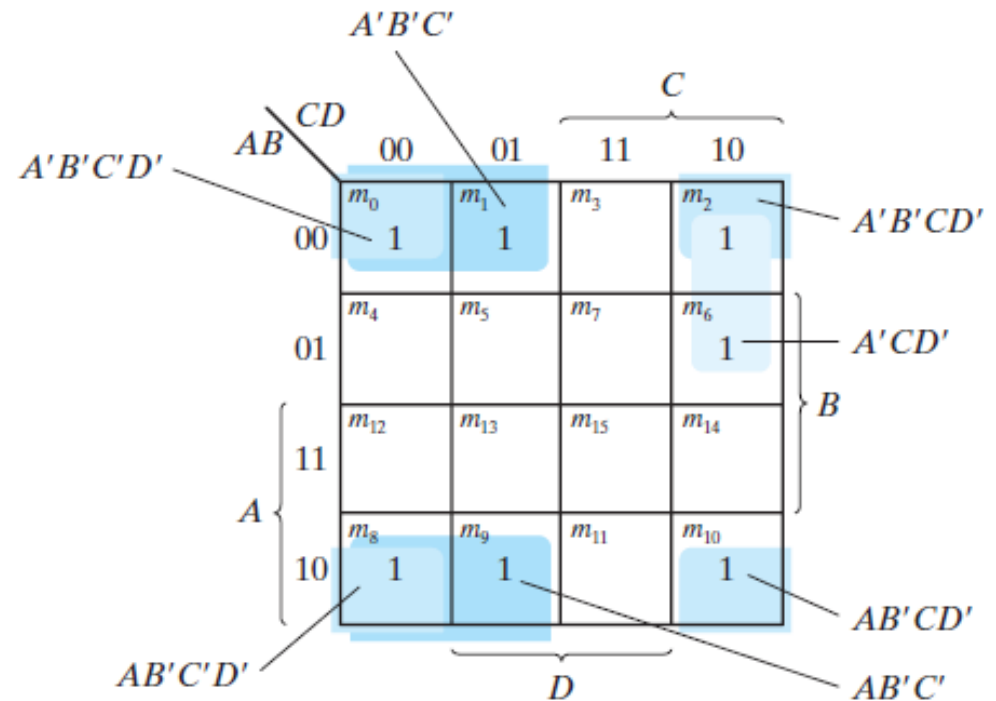
- Simplify  $F(w,x,y,z) = \sum m(0,1,2,4,5,6,8,9,12,13,14)$



$$F(w,x,y,z) = y' + w'z' + xz'$$

# Example

- Simplify  $F = A'B'C' + B'CD' + A'BCD' + AB'C'$



$$F = B'D' + B'C' + A'CD'$$

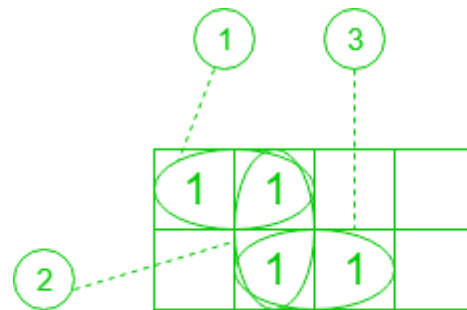
Note:  $A'B'C'D' + A'B'CD' = A'B'D'$   
 $AB'C'D' + AB'CD' = AB'D'$   
 $A'B'D' + AB'D' = B'D'$   
 $A'B'C' + AB'C' = B'C'$

# Prime Implicants

- A **prime implicant** is a product term obtained by combining the maximum possible number of adjacent squares in the map.
- If a minterm is covered by only one prime implicant, that prime implicant is said to be *essential*.

# Prime Implicants

- This is a group of square or rectangle made up of bunch of adjacent minterms which is allowed by definition of K-Map. i.e. all possible groups formed in K-Map.

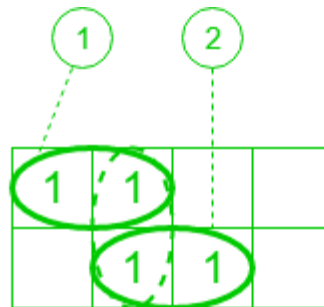


No. of Prime Implicants = 3



# Essential Prime Implicants

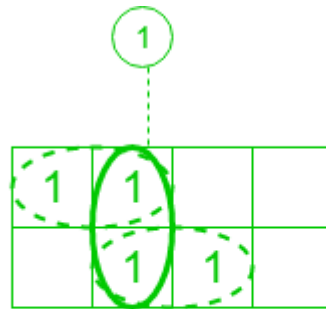
- These are those subcubes (groups) which cover at least one minterm that can't be covered by any other prime implicant.
- **Essential prime implicants(EPI)** are those prime implicants which always appear in final solution.



No. of Essential Prime Implicants = 2

# Redundant Prime Implicants

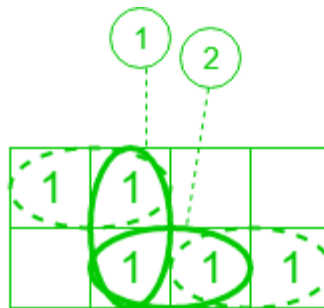
- The prime implicants for which each of its minterm is covered by some essential prime implicant are **redundant prime implicants(RPI)**.
- This prime implicant never appears in final solution.



No. of Redundant Prime Implicants = 1

# Selective Prime Implicants (SPI)

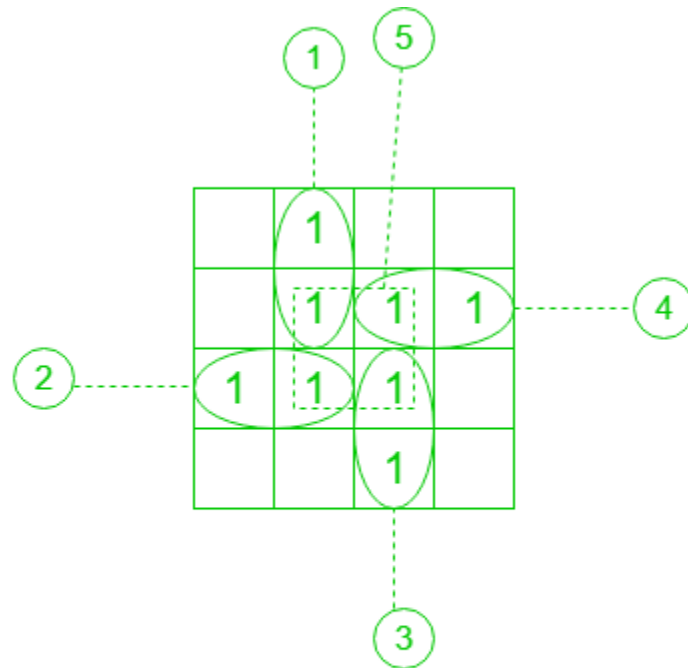
- These are the prime implicants for which are neither essential nor redundant prime implicants.
- These are also known as non-essential prime implicants.
- They may appear in some solution or may not appear in some solution.



No. of Selective Prime Implicants = 2

# Example

- Given  $F = \sum(1, 5, 6, 7, 11, 12, 13, 15)$ , find number of implicant, PI, EPI, RPI and SPI.



No. of Implicants = 8

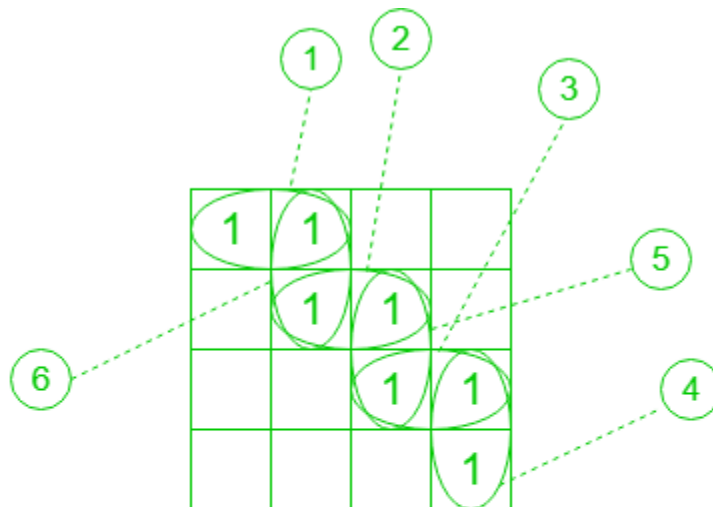
PI = (1,2,3,4,5)

EPI = (1,2,3,4)

RPI = (5)

$$F = \textcircled{1} + \textcircled{2} + \textcircled{3} + \textcircled{4}$$

- Given  $F = \sum(0, 1, 5, 7, 15, 14, 10)$ , find number of implicant, PI, EPI, RPI and SPI.



No. of Implicants = 7

PI = (1,2,3,4,5,6)

EPI = (1,4)

SPI = (2,3,5,6)

$$F = (1) + (2) + (3) + (4)$$

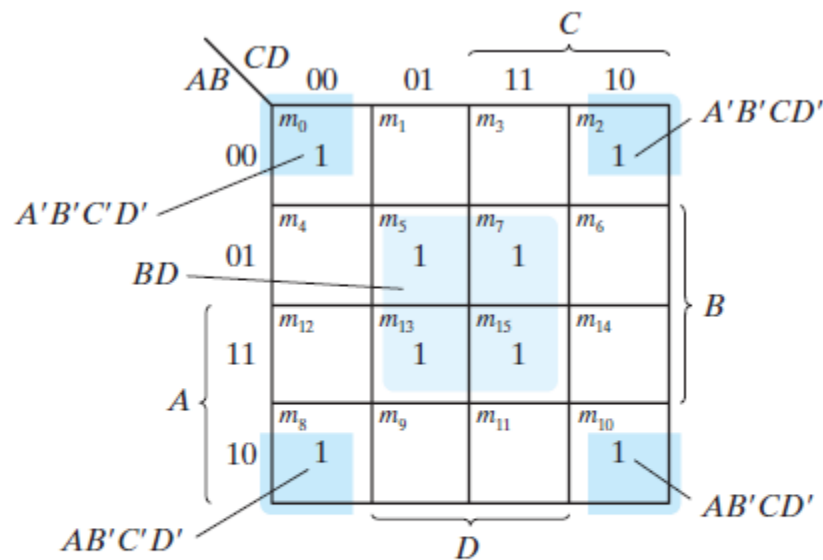
OR

$$F = (1) + (5) + (6) + (4)$$

# Example

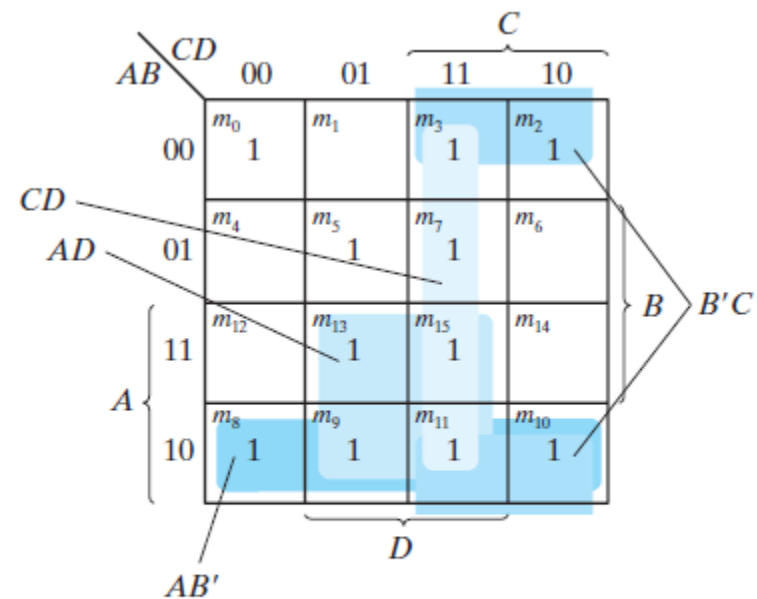
■  $F(A,B,C,D) = \sum m(0,2,3,5,7,8,9,10,11,13,15)$

$$\begin{aligned} F &= BD + B'D' + CD + AD \\ &= BD + B'D' + CD + AB' \\ &= BD + B'D' + B'C + AD \\ &= BD + B'D' + B'C + AB' \end{aligned}$$



Note:  $A'B'C'D' + A'B'CD' = A'B'D'$   
 $AB'C'D' + AB'CD' = AB'D'$   
 $A'B'D' + AB'D' = B'D'$

(a) Essential prime implicants  
 $BD$  and  $B'D'$



(b) Prime implicants  $CD$ ,  $B'C$ ,  
 $AD$ , and  $AB'$

# Five-Variable Map

- Not simple, is not usually used.
  - 5 variables = 32 squares.
  - 6 variables = 64 squares.

- The left-hand four-variable map represents the 16 squares where  $A=0$ , and the other four-variable map represents the squares where  $A=1$ .
- In addition, each square in the  $A=0$  map is adjacent to the corresponding square in the  $A=1$  map.

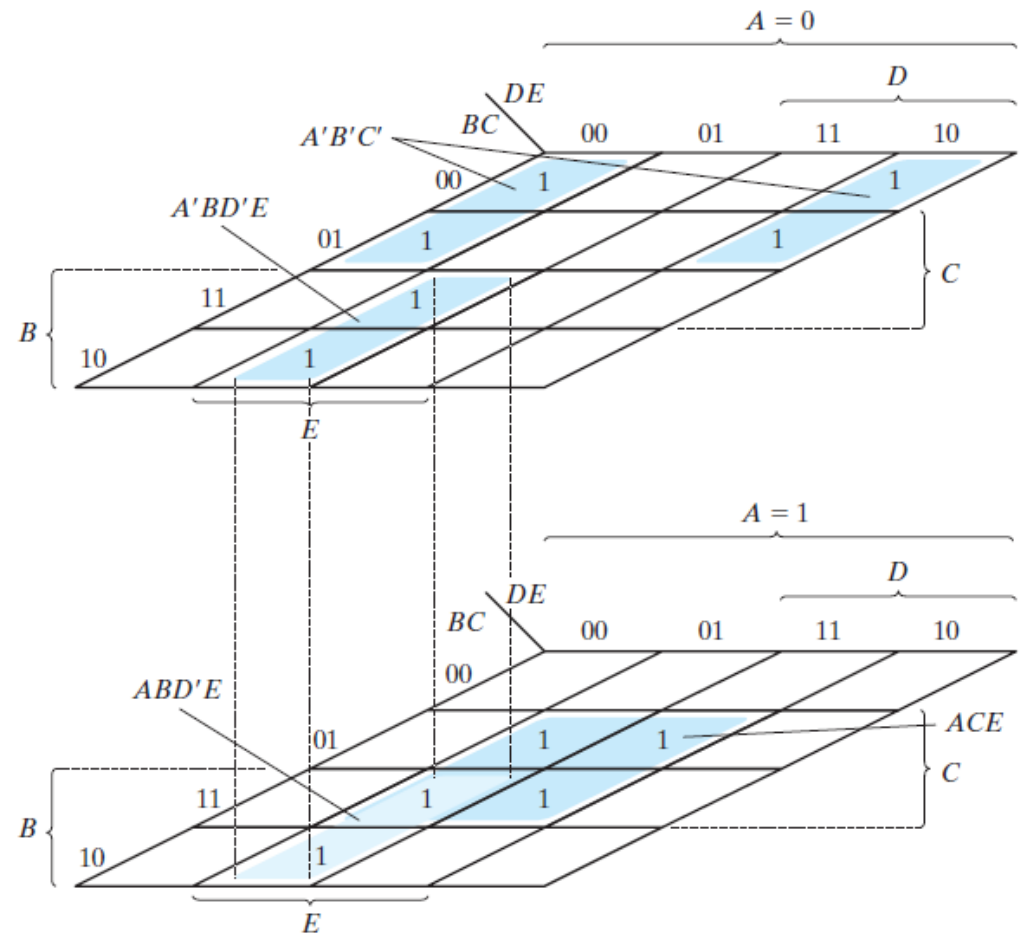
$A = 0$					
		$DE$		$D$	
$BC$		00	01	11	10
$B$	00	0	1	3	2
	01	4	5	7	6
	11	12	13	15	14
	10	8	9	11	10
		$E$			

$A = 1$					
		$DE$		$D$	
$BC$		00	01	11	10
$B$	00	16	17	19	18
	01	20	21	23	22
	11	28	29	31	30
	10	24	25	27	26
		$E$			



- Simplify  $F(A,B,C,D,E) = \sum m(0,2,4,6,9,13,21,23,25,29,31)$

$$F = A'B'E' + BD'E' + ACE$$



***The Relationship Between the Number of Adjacent Squares  
and the Number of Literals In the Term***

<i>K</i>	<b>Number of Adjacent Squares</b> $2^k$	<b>Number of Literals in a Term in an <i>n</i>-variable Map</b>			
		<i>n</i> = 2	<i>n</i> = 3	<i>n</i> = 4	<i>n</i> = 5
0	1	2	3	4	5
1	2	1	2	3	4
2	4	0	1	2	3
3	8		0	1	2
4	16			0	1
5	32				0

# Product-of-Sums Simplification

- Minterms represented by 1's placed in the squares of the map.
- The minterms not included is denote by its complement 0's.
- The complement of SOP of the function  $F$  gives us back the function  $F$  in POS form (a consequence of DeMorgan's theorem).

# Example

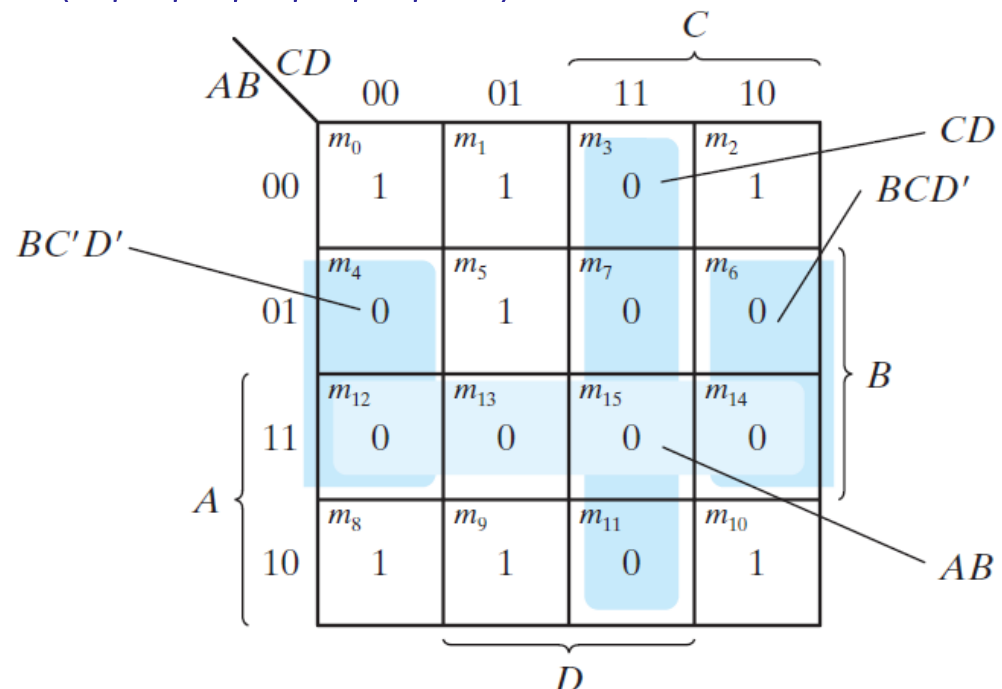
## ■ Simplify the following Boolean function into

1. Sum-of-products form and
2. Product-of-sums form:

$$F(A, B, C, D) = (0, 1, 2, 5, 8, 9, 10)$$

## ■ Solution

- The 1's in the map represent all the minterms of the function.
- The squares marked with 0's represent the minterms **not included** in  $F$  and therefore denote the complement of  $F$



## 1. SOP

- Combine the squares with 1's in the map

$$F = B'D' + B'C' + A'C'D$$

## 2. POS

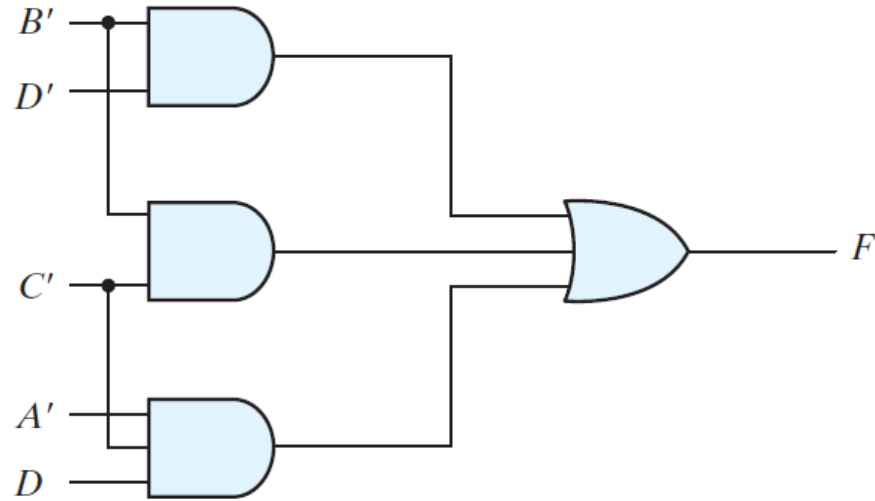
- Combining squares marked with 0's, we obtain the simplified complemented function:

$$F' = AB + CD + BD'$$

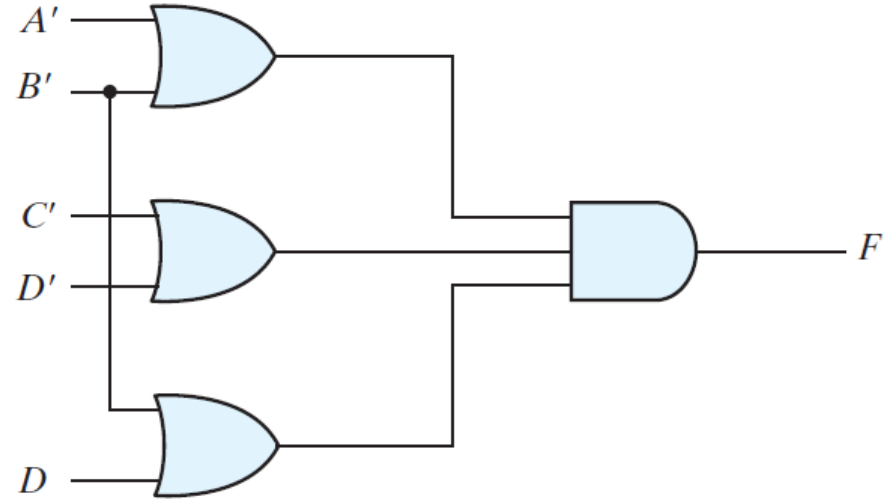
- Applying DeMorgan's theorem (to above function)

$$F = (A' + B') (C' + D') (B' + D)$$

$$\begin{aligned}
 F(A, B, C, D) &= (0, 1, 2, 5, 8, 9, 10) \\
 &= B'D' + B'C' + A'C'D \\
 &= (A' + B')(C' + D')(B' + D)
 \end{aligned}$$



(a)  $F = B'D' + B'C' + A'C'D$



(b)  $F = (A' + B')(C' + D')(B' + D)$

# Example

Express the following sum-of-minterms as POS

$$F(x, y, z) = \sum m(1, 3, 4, 6)$$

Sol

- In product-of-maxterms form

$$F(x, y, z) = \prod M(0, 2, 5, 7)$$

SOP

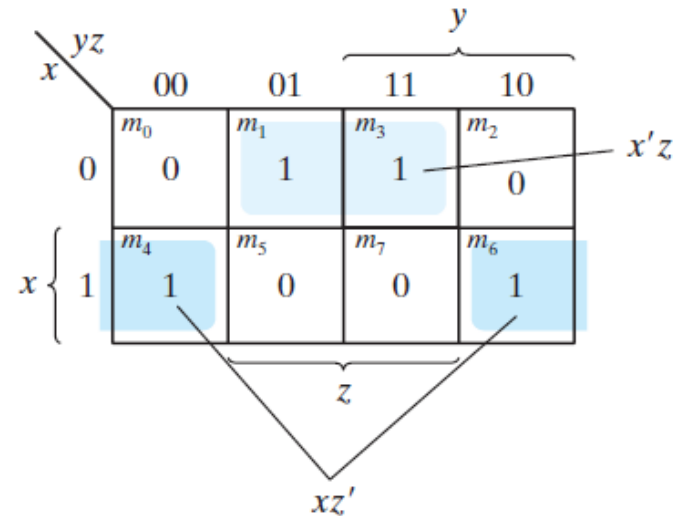
$$F = x'z + xz'$$

POS

- Combine the 0's and take the complement

$$F' = xz + x'z'$$

$$F = (x' + z')(x + z)$$



# Don't-Care Conditions

- A combination of variables whose logical value is not specified.
  - Cannot be marked with a 1 or 0 in the map
  - Marked with an X
- An X inside a square in the map indicates that we don't care whether the value of 0 or 1 is assigned to  $F$  for the particular minterm.



## Example

Simplify the Boolean function

$$F(w, x, y, z) = (1, 3, 7, 11, 15)$$

which has the don't-care conditions

$$d(w, x, y, z) = (0, 2, 5)$$

Sol

		y			
		00	01	11	10
wx	yz				
		$m_0$	$m_1$	$m_3$	$m_2$
w'x'	00	X	1	1	X
	01	$m_4$	$m_5$	$m_7$	$m_6$
w	01	0	X	1	0
	11	$m_{12}$	$m_{13}$	$m_{15}$	$m_{14}$
	10	0	0	1	0
		z			

(a)  $F = yz + w'x'$

		y			
		00	01	11	10
wx	yz				
		$m_0$	$m_1$	$m_3$	$m_2$
w'z	00	X	1	1	X
	01	$m_4$	$m_5$	$m_7$	$m_6$
w	01	0	X	1	0
	11	$m_{12}$	$m_{13}$	$m_{15}$	$m_{14}$
	10	0	0	1	0
		z			

(b)  $F = yz + w'z$

- In (a),

$$F = yz + w'x'$$

$$F(w, x, y, z) = yz + w'x' = (0, 1, 2, 3, 7, 11, 15)$$

- In (b)

$$F = yz + w'z$$

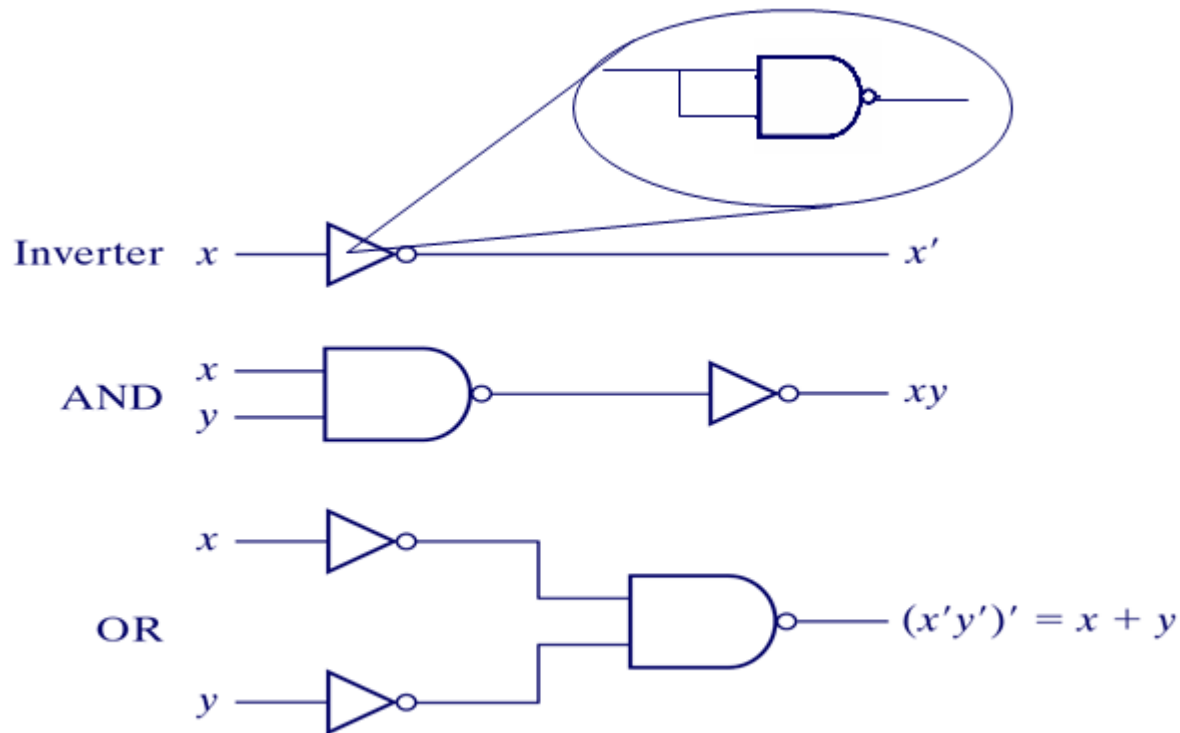
$$F(w, x, y, z) = yz + w'z = (1, 3, 5, 7, 11, 15)$$

# NAND AND NOR Implementation

- Digital circuits are constructed with NAND or NOR
  - Easier to fabricate with electronic components
  - The basic gates used in all IC digital logic families.

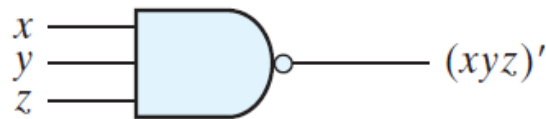
# NAND Circuits

- Universal gate – any logic circuit can be implemented with it.

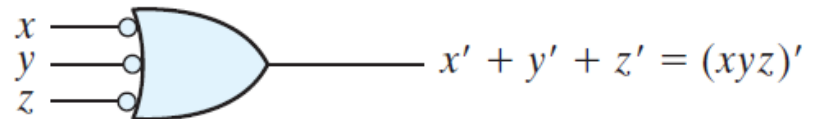


Logic Operations with NAND Gates

- Alternative graphic symbol for the NAND gate



(a) AND-invert



(b) Invert-OR

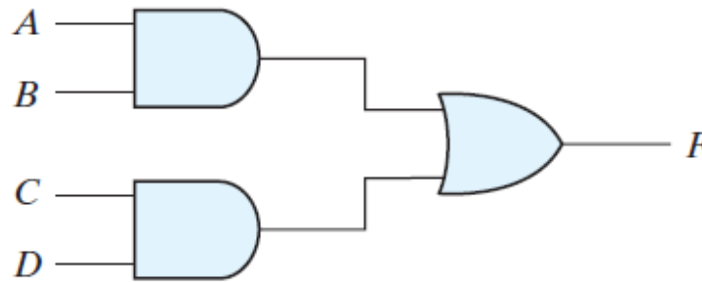
- Can be implemented in

- ☐ 2 Level
- ☐ Multilevel

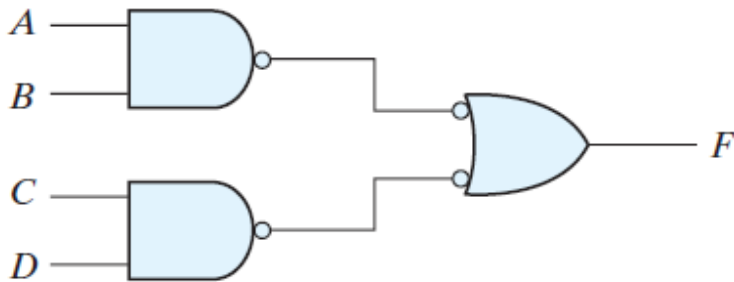
# Two-Level Implementation

- Expressed in sum-of-products form.

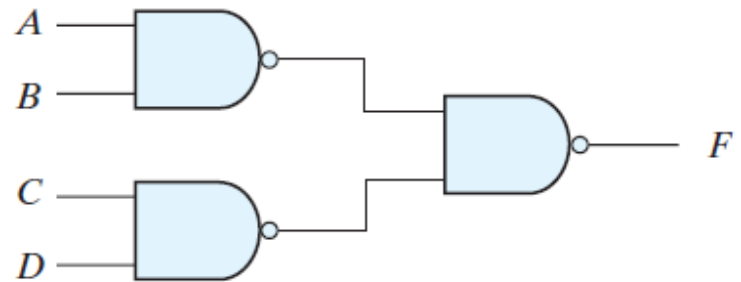
$$F = AB + CD = ((AB)'(CD)')'$$



(a)



(b)



(c)

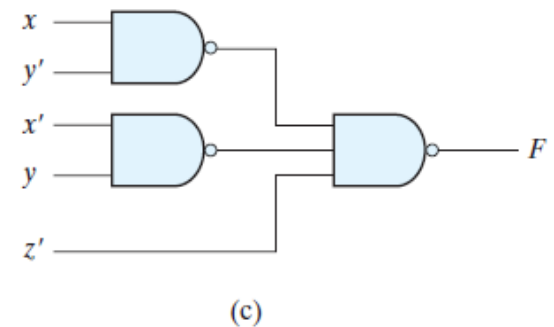
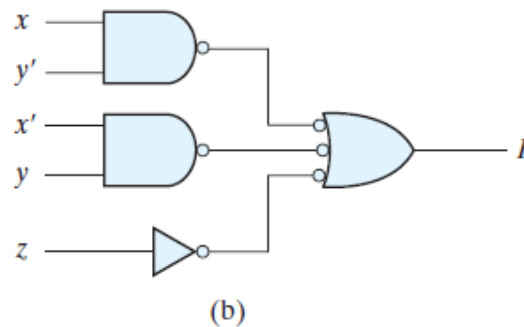
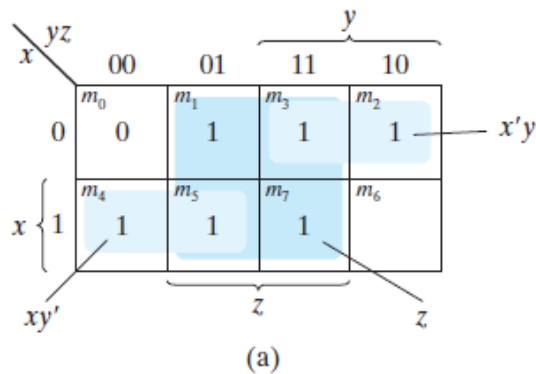
# Example

- Implement the following Boolean function with NAND gates:

$$F(x, y, z) = (1, 2, 3, 4, 5, 7)$$

- Sol

$$F = xy' + x'y + z$$



## Steps

1. Simplify the function and express it in SOP
2. Draw a NAND gate for each product term of the expression that has at least two literals. The inputs to each NAND gate are the literals of the term.
3. Draw a single gate using the AND-invert or the invert-OR graphic symbol in the second level, with inputs coming from outputs of first-level gates.
4. A term with a single literal requires an inverter in the first level. However, if the single literal is complemented, it can be connected directly to an input of the second level NAND gate.



# Multilevel NAND Circuits

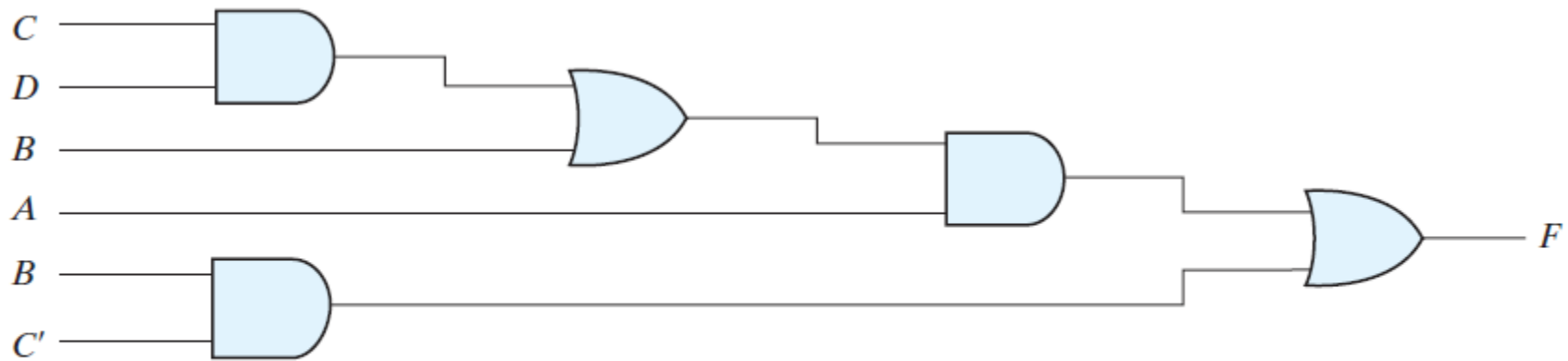
- Consider the Boolean function

$$F = A(CD + B) + BC'$$

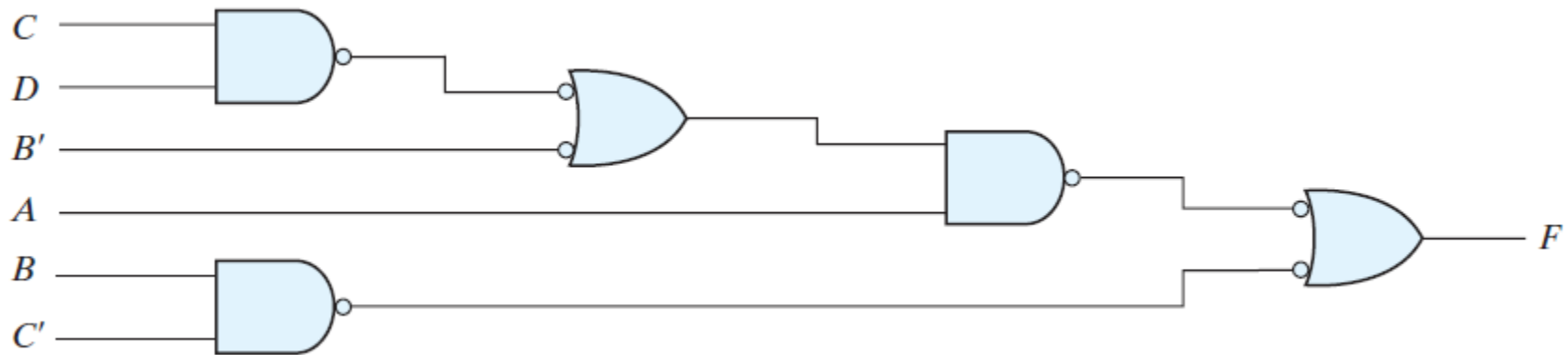
*Sol*

- **Steps**

1. Convert all AND gates to NAND gates with AND-invert graphic symbols.
2. Convert all OR gates to NAND gates with invert-OR graphic symbols.
3. Check all the bubbles in the diagram. For every bubble that is not compensated by another small circle along the same line, insert an inverter (a one-input NAND gate) or complement the input literal.



(a) AND-OR gates



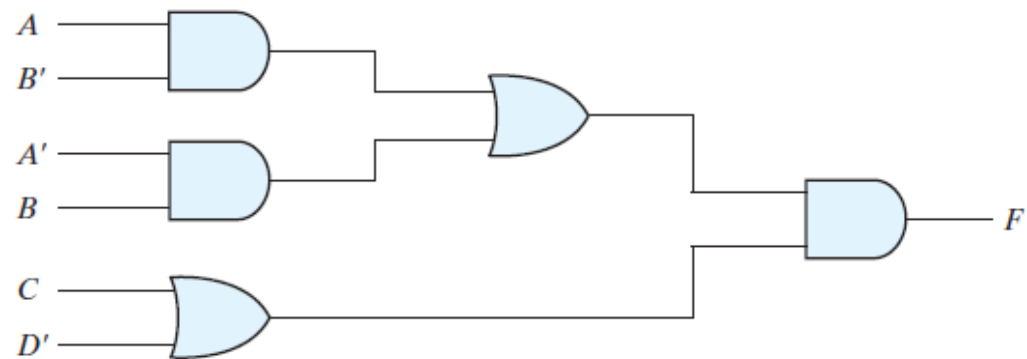
(b) NAND gates

# Example

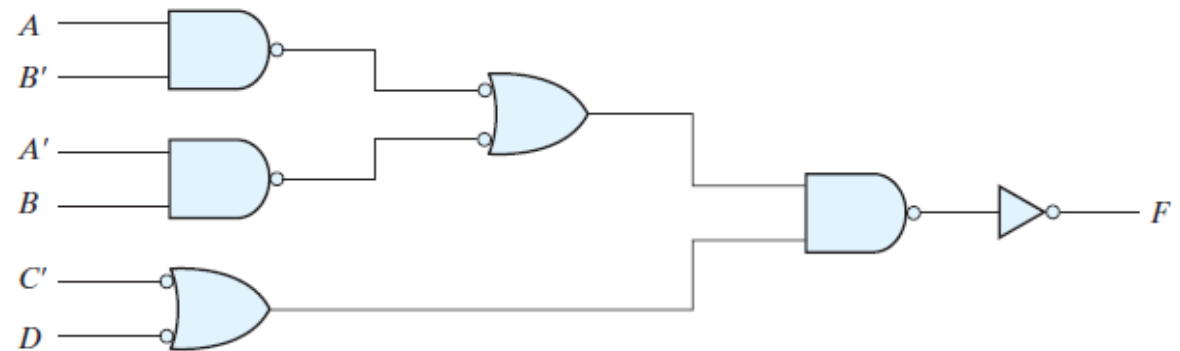
- Consider the multilevel Boolean function

$$F = (AB' + A'B)(C + D')$$

- Sol



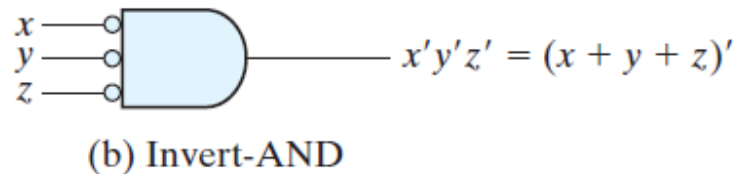
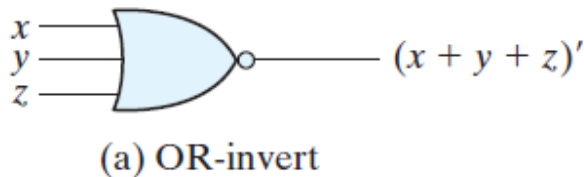
(a) AND-OR gates



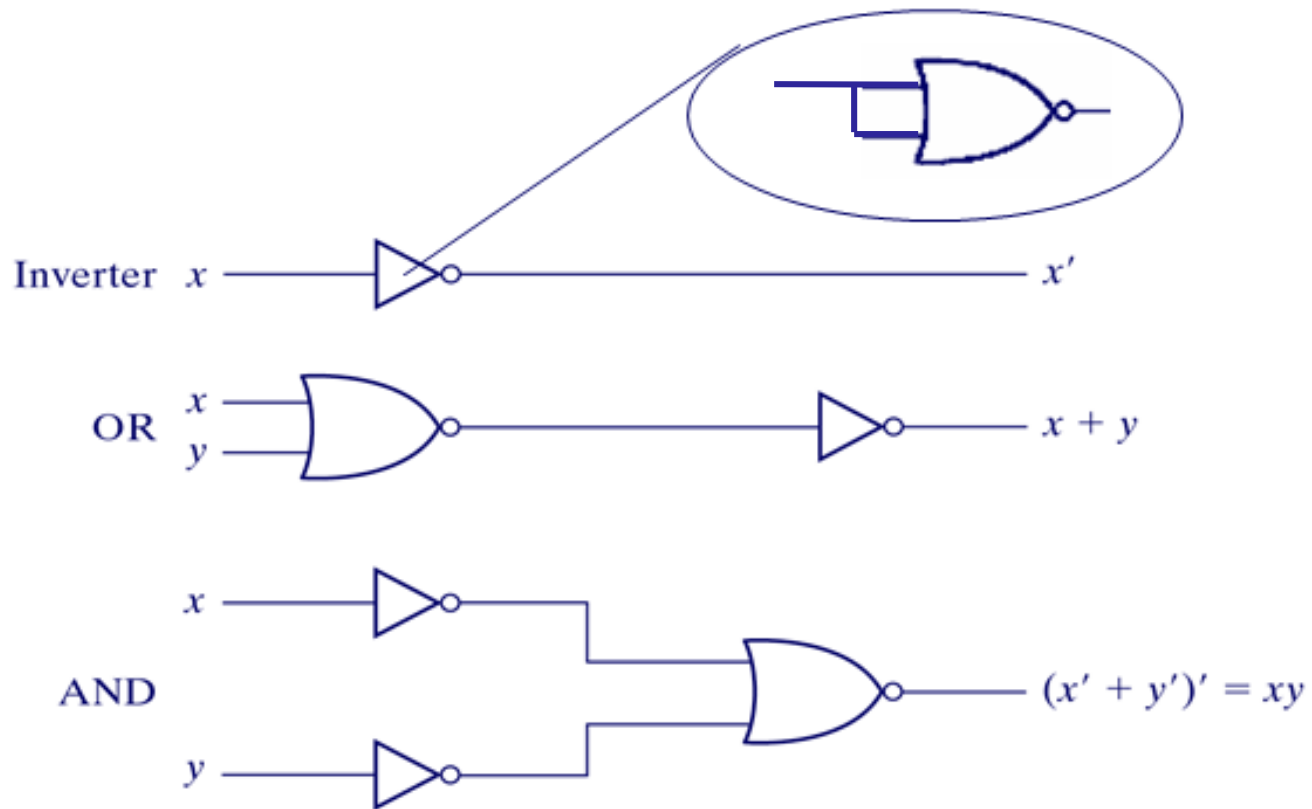
(b) NAND gates

# NOR Implementation

- The NOR operation is the dual of the NAND operation.
  - Procedures and rules for NOR logic are the duals NAND logic.
- **Universal gate** – any logic circuit can be implemented with it.



- Alternative graphic symbol for the NOR gate
  - In part (b), we can place a bubble (NOT) in each input and apply the [DeMorgan's theorem](#), then get a Boolean function in NOR type.



Logic Operations with NOR Gates

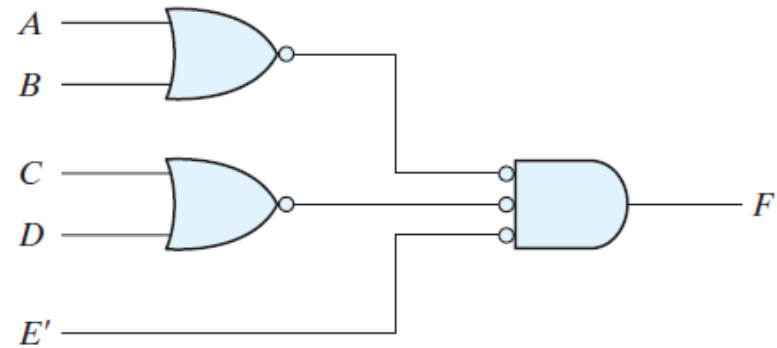
# Implementation

- Simplify the function into POS form.
  - Based on K-map - combining the 0's and complementing.
- Transformation from the OR–AND to a NOR diagram
  - Change the OR gates to NOR gates with OR-invert graphic symbols
  - Change the AND gate to a NOR gate with an invert-AND graphic symbol.
  - A single literal term going into the second-level gate must be complemented.

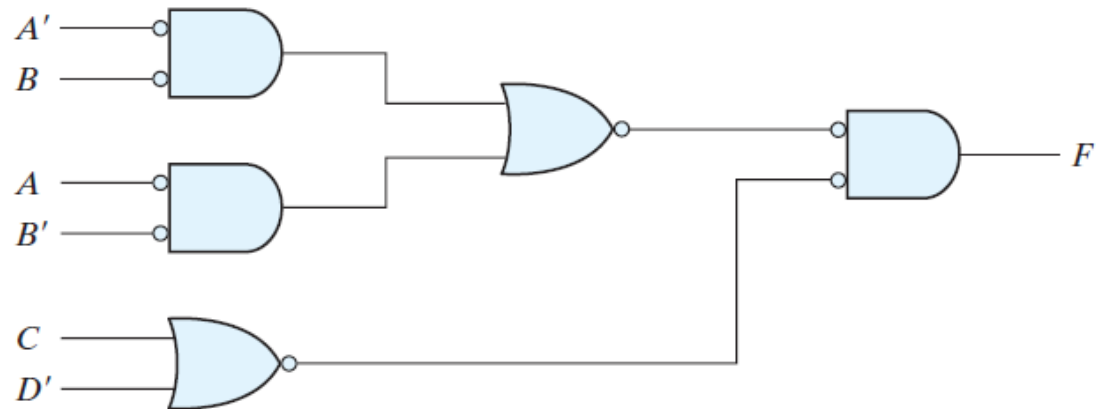
- Express the following function with a NOR diagram

$$F = (A + B)(C + D)E$$

$$F = (AB' + A'B)(C + D)'$$



Implementing  $F = (A + B)(C + D)E$



Implementing  $F = (AB' + A'B)(C + D')$  with NOR gates

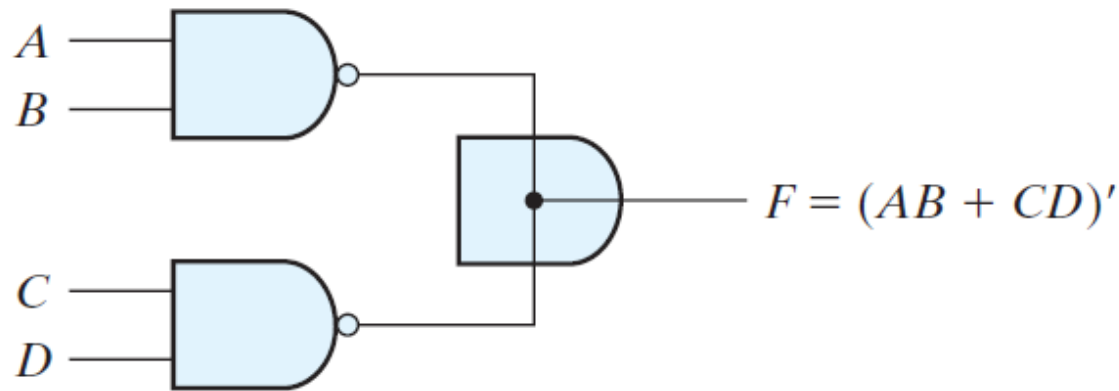


## Other Two-level Implementations

- Some NAND or NOR gates allow the possibility of a wire connection between the outputs of two gates to provide a specific logic function.
- This type of logic is called *wired logic*.

- The logic function called an AND–OR–INVERT can be given as;

$$F = (AB)' \dots (CD)' = (AB + CD)' = (A' + B')(C' + D')$$

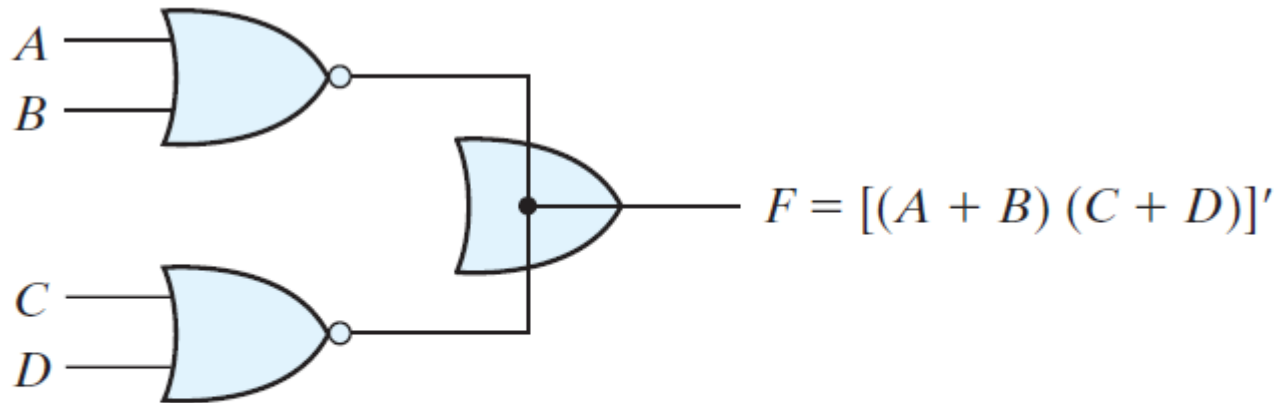


(a) Wired-AND in open-collector  
TTL NAND gates.

(AND–OR–INVERT)

- The NOR outputs of emitter-coupled logic (ECL) gates can be tied together to perform a wired-OR function called an OR–AND–INVERT function.

$$F = (A + B)' + (C + D)' = [(A + B)(C + D)]'$$



(b) Wired-OR in ECL gates

(OR–AND–INVERT)

- Some NAND or NOR gates allow the possibility of a wire connection between the outputs of two gates to provide a **wired logic**.
- Open-collector TTL NAND gates, when **tied together**, perform the **wired-AND logic**
- The wired-AND gate is **not a physical gate**.

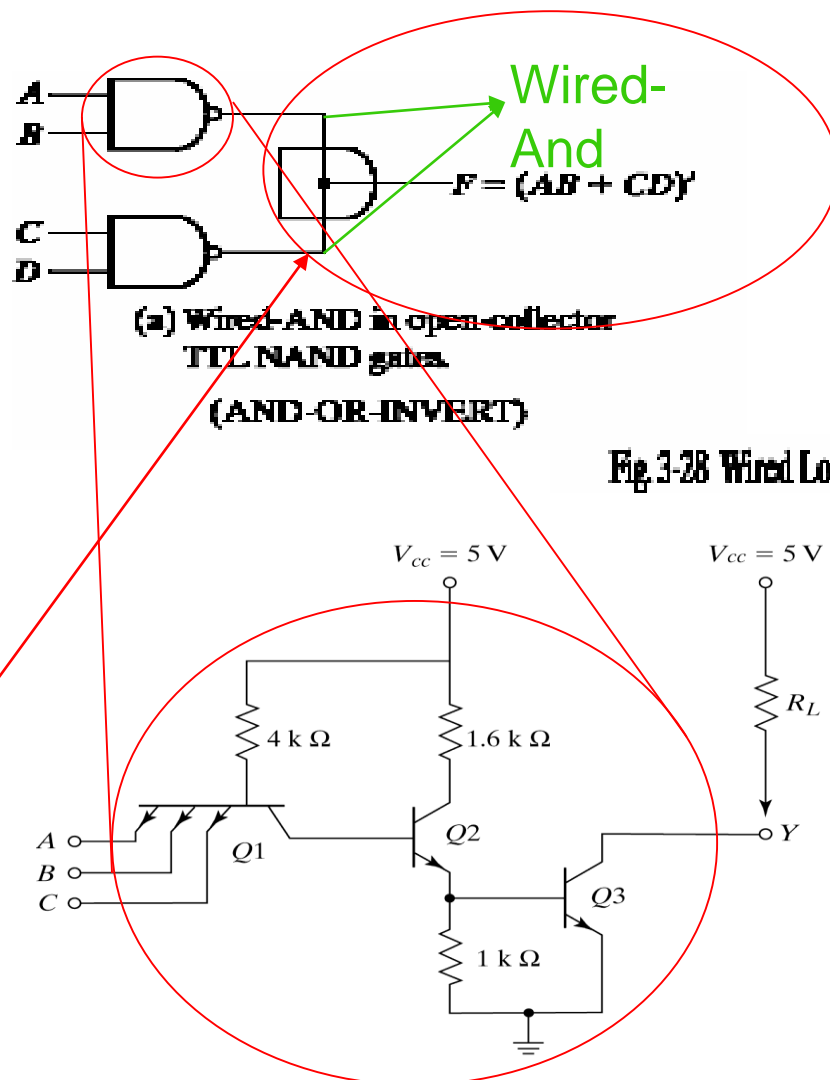


Fig. 10-11 Open-Collector TTL Gate

# Nondegenerate forms

- We consider four types of gates: AND, OR, NAND, and NOR. These will have 16 combinations of two-level forms. (*by assigning one type of gate for the first level and one type for the second level*)
- Eight of these combinations are said to be degenerate forms, because they degenerate to a single operation.
- The other eight nondegenerate forms produce an SOPs or POSs as follows:

AND-OR → 3-4

NAND-NAND → 3-6

NOR-OR

OR-NAND

OR-AND → 3-4

NOR-NOR → 3-6

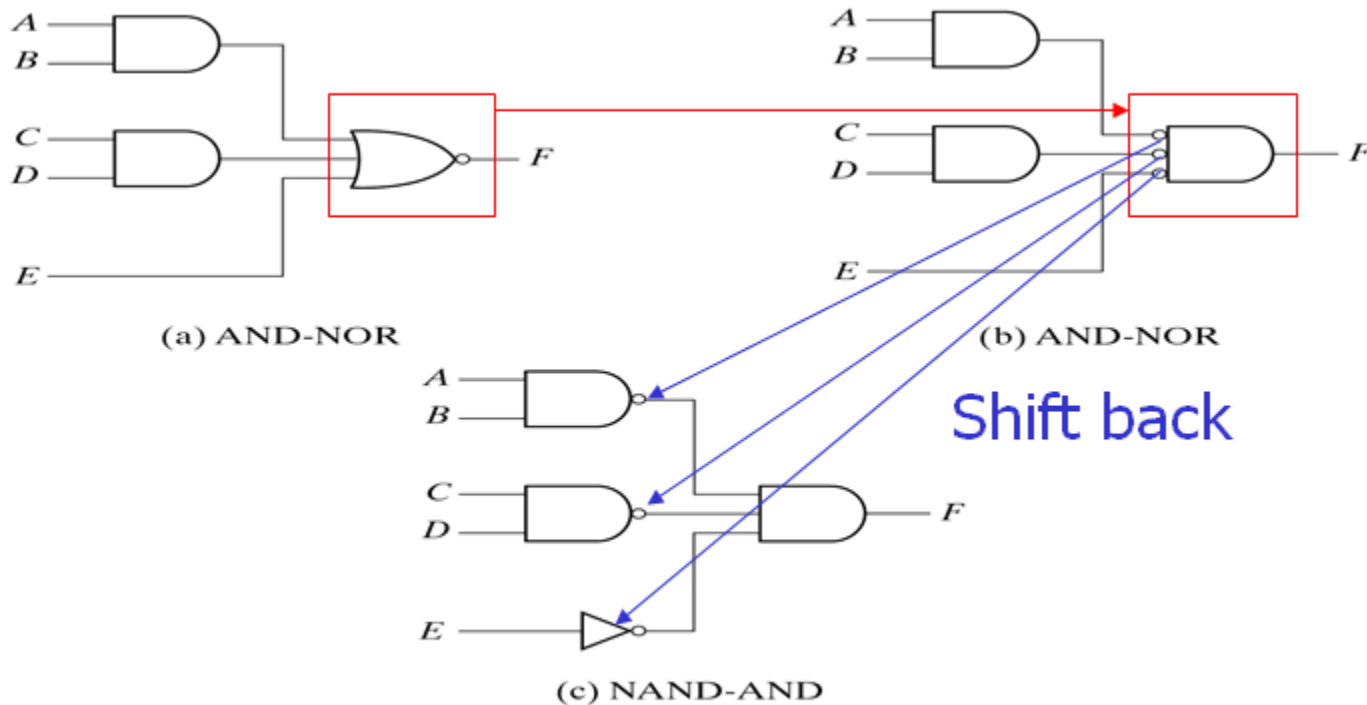
NAND-AND

AND-NOR

# AND-OR-INVERT implementation

- The two forms **NAND-AND** and **AND-NOR** are **equivalent forms** and can be treated together.

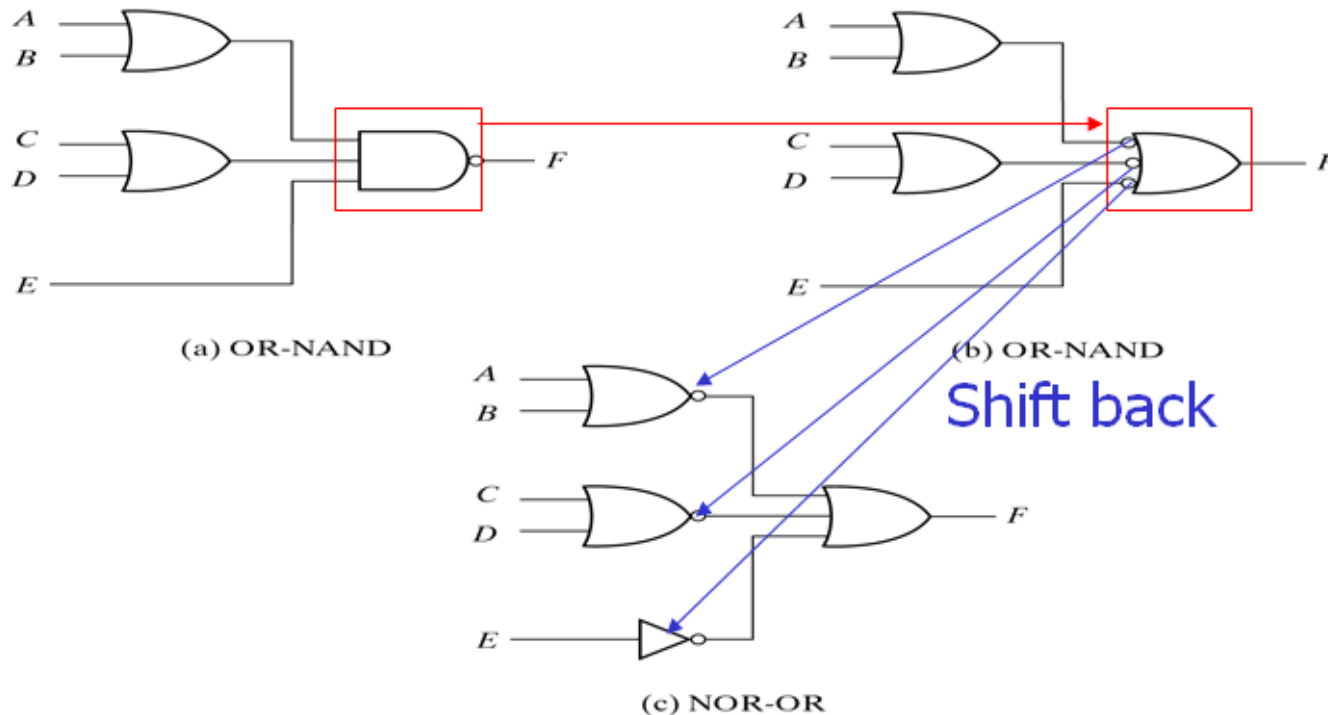
$$F = (AB + CD + E)'$$



AND-OR-INVERT Circuits;  $F = (AB + CD + E)'$

# OR-AND-INVERT implementation

- The OR–NAND and NOR–OR forms perform the OR–AND–INVERT function.
- The **OR-NAND** form resembles the **OR-AND** form, except for the inversion done by the bubble in the NAND gate.



$$\text{OR-AND-INVERT Circuits; } F = [(A + B)(C + D)E]'$$

## Tabular summary

- Because of the **INVERT** part in each case, it is **convenient to use the simplification of  $F'$**  of the function.

Equivalent Nondegenerate Form		Implements the Function	Simplify $F'$ into	To Get an Output of
(a)	(b)*			
AND-NOR	NAND-AND	AND-OR-INVERT	Sum-of-products form by combining 0's in the map.	$F$
OR-NAND	NOR-OR	OR-AND-INVERT	Product-of-sums form by combining 1's in the map and then complementing.	$F$





## ■ Homework

- Exclusive-OR Function
- Odd Function
- Parity Generation and Checking