

Problems:- Date:- 08-01-2020 27:00

1) A discrete time signal $x(n)$ is defined as $x[n]$

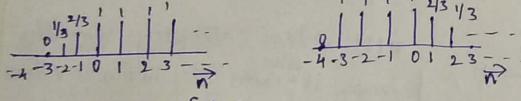
$$= \begin{cases} 1 + \frac{n}{3}, & -3 \leq n \leq 4 \\ 1, & 0 \leq n \leq 3 \\ 0, & \text{elsewhere} \end{cases}$$

- a) Determine its value and sketch the signal $x(n)$.
- $1 + \frac{n}{3}, -3 \leq n \leq -1 \Rightarrow$ at $n = -3, 0$
 $n = -2, \frac{1}{3}$
 $n = -1, \frac{2}{3}$
- from $0 \leq n \leq 3, 1$
 elsewhere, 0

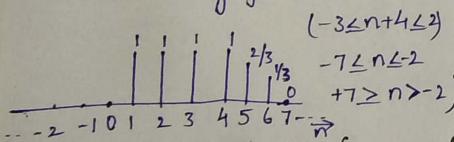
so $x(n) = \begin{cases} \dots, 0, \frac{1}{3}, \frac{2}{3}, 1, 1, 1, 0, \dots \end{cases}$

- b) Sketch its values and the signals that result if we
 i) First fold $x(n)$ and then delay the resulting signal
 by four samples.

$x(-n) = \text{folded signal for } x(n)$



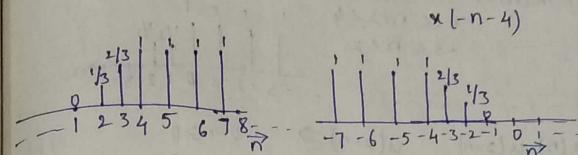
$x(-n+4) \rightarrow$ after
delaying



- 2) First delay $x(n)$ by four samples & then fold the resulting signal.
 $x(n) \rightarrow$ delay by 4 samples.

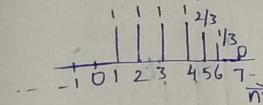
$\text{folding } x(n-4) \rightarrow x(-n-4)$

$x(n-4)$



$x(-n-4)$

c) Sketch the signal $x(-n+4)$



d) Compare the results in parts (b) and (c) and derive a rule for obtaining the signal $x(-n+k)$ from $x(n)$.

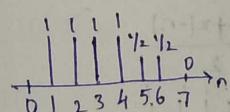
By comparing results in parts (b) & (c) we can say that to get $x(-n+k)$ from $x(n)$ -first we need to fold $x(n)$ which results in $x(-n)$ & then we need to shift by k samples to right if $k > 0$ (or) to left if $k < 0$ results in $x(-n+k)$.

e) Can you express the signal $x(n)$ in terms of signals $\delta(n)$ & $u(n)$?

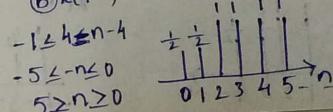
Yes. $x(n) = \frac{1}{3} \delta(n-2) + \frac{2}{3} \delta(n-1) + u(n) - u(n-4)$

2) A discrete time signal $x(n)$ is shown in the figure. Sketch and label the each of following signals:

a) $x(n-2)$



(b) $x(n+4)$



-1 ≤ n ≤ -4

-5 ≤ n ≤ 0

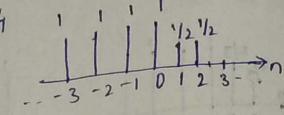
5 ≥ n ≥ 0

1 ≤ n ≤ 4

5 ≥ n ≥ 1

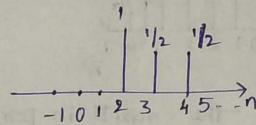
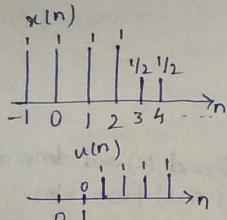
$$c) x(n+2)$$

$$\begin{aligned} -1 &\leq n+2 \leq 4 \\ -3 &\leq n \leq 2 \end{aligned}$$



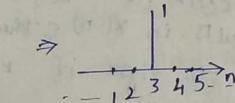
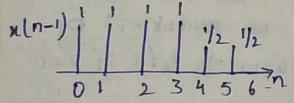
$$d) x(n) u(2-n) \quad u(2-n) \Rightarrow$$

$$\begin{aligned} 2-n &\geq 0 \\ n &\leq 2 \end{aligned}$$

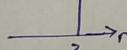


$$e) x(n-1) \delta(n-3)$$

$$x(n-1)\delta(n-3)$$



$$\delta(n-3)$$

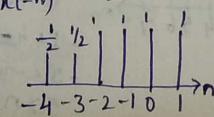
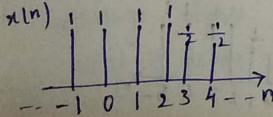


$$f) x(n^2) \Rightarrow x(n) = \{x(-2), x(-1), x(0), x(1), x(2), x(3), x(4), \dots\}$$

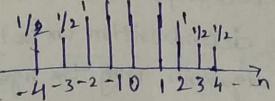
$$x(n^2) = \{x(4), x(1), x(10), x(11), x(14), x(19), \dots\}$$

$$= \left\{ \frac{1}{2}, 1, 1, 1, \frac{1}{2}, 0, 0, \dots \right\}$$

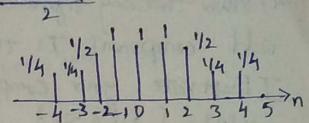
$$g) \text{ even part of } x(n) = \frac{x(n)+x(-n)}{2}$$



$$x(n)+x(-n)$$

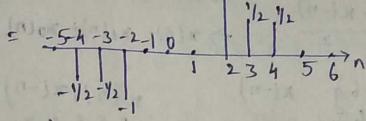


$$x(n)+x(-n)$$

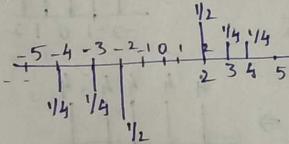


$$\text{odd part } x_0(n) = \frac{x(n)-x(-n)}{2}$$

$$x(n)-x(-n)$$



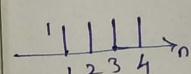
$$\frac{x(n)-x(-n)}{2}$$



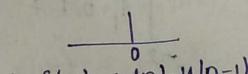
$$3) a) \text{ show that } \delta(n) = u(n) - u(n-1)$$

$$\text{We know } \delta(n) = \frac{1}{0} \quad \frac{1}{1} \quad \frac{1}{1} \quad \frac{1}{1} \quad \frac{1}{0} \quad \frac{1}{1} \quad \frac{1}{1} \quad \frac{1}{1} \quad \frac{1}{0} \quad \dots$$

$$u(n)$$



$$u(n) - u(n-1)$$



$$\therefore \delta(n) = u(n) - u(n-1)$$

$$b) u(n) = \sum_{k=-\infty}^{\infty} \delta(k) = \sum_{k=0}^{\infty} \delta(n-k)$$

$$u(n) = \frac{1}{0} \quad \frac{1}{1} \quad \frac{1}{1} \quad \frac{1}{1} \quad \frac{1}{0} \quad \dots \quad \Rightarrow \sum_{n=-\infty}^{\infty} \delta(k) = u(n) = \begin{cases} 0, & n < 0 \\ 1, & n \geq 0 \end{cases}$$

$$\sum_{k=0}^{\infty} \delta(n-k) = \begin{cases} 0, & n < 0 \\ 1, & n \geq 0 \end{cases}$$

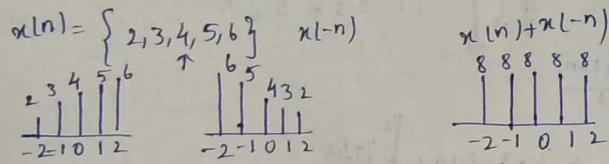
4) Show that any signal can be decomposed into even & odd component. Is the decomposition unique?

Illustrate your proofs using the signal.

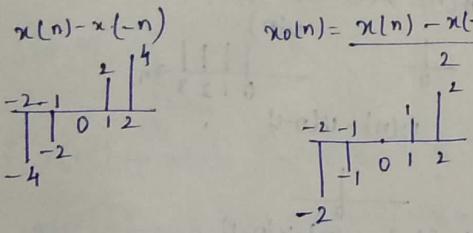
$$x[n] = \{ 2, 3, 4, 5, 6 \}$$

$$x_e(n) = \frac{x(n) + x(-n)}{2} \quad x_e(n) = x(n)$$

$$x_o(n) = \frac{x(n) - x(-n)}{2} \quad \Rightarrow x(n) = x_e(n) + x_o(n)$$



$$x_e(n) = \frac{x(n) + x(-n)}{2} \Rightarrow \begin{array}{c|ccccc} & 4 & 4 & 4 & 4 \\ \hline -2 & & & & & \\ -1 & & & & & \\ 0 & & & & & \\ 1 & & & & & \\ 2 & & & & & \end{array}$$



5) show that the energy (power) of a real-valued energy (power) signal is equal to the sum of the energies (powers) of its even and odd components.

To prove that

$$\sum_{n=-\infty}^{\infty} x_e(n) + x_o(n) = 0 \Rightarrow \sum_{n=-\infty}^{\infty} x_e(n) + x_o(n) = \sum_{m=-\infty}^{\infty} x_e(-m) x_o(-m)$$

$$\Rightarrow \sum_{m=-\infty}^{\infty} x_e(m) x_o(m)$$

$$= - \sum_{m=-\infty}^{\infty} x_e(n) x_o(n) = \sum_{n=-\infty}^{\infty} x_e(n) x_o(n) = 0$$

Energy (powers)

$$\Rightarrow \sum_{n=-\infty}^{\infty} x^2(n) = \sum_{n=-\infty}^{\infty} [x_e(n) + x_o(n)]^2$$

$$= \sum_{n=-\infty}^{\infty} x_e^2(n) + x_o^2(n) + 2x_e(n) \cdot x_o(n)$$

$$= \sum_{n=-\infty}^{\infty} x_e^2(n) + \sum_{n=-\infty}^{\infty} x_o^2(n) + 2 \sum_{n=-\infty}^{\infty} x_e(n) x_o(n)$$

$$= E_e + E_o + 0$$

$$E = E_e + E_o$$

b) Consider the system $y(n) = T[x(n)] = x(n^2)$

a) Determine if the S/I/M is time invariant.

$$\text{given } y(n) = T[x(n)] = x(n^2)$$

$$x(n-k) \Rightarrow y(n) = x[(n-k)^2]$$

$$= x(n^2 + k^2 - 2nk)$$

$$x(n-k) \neq y(n-k)$$

So, the given system is ~~time invariant~~ time variant.

b) To clarify the result in part (a) assume that

$$\text{the signal } x(n) = \begin{cases} 1, & 0 \leq n \leq 3 \\ 0, & \text{otherwise.} \end{cases}$$

is applied into the S/I/M.

1) Sketch the signal $x(n)$

$$x(n) = \left\{ \dots, 0, 1, 1, 1, 0, \dots \right.$$

$\uparrow \cdot \cdot \cdot$

n

1 1 1
0 1 2 3

2) Determine & sketch the signal $y(n) = T[x(n)]$.

$$y(n) = T[x(n)] = x(n^2) \Rightarrow \{x(0), x(1), x(2^2), x(3^2), x(4^2), \dots\}$$

$$\Rightarrow \{x(0), x(1), x(4), x(9), \dots\}$$

$$y(n) = x(n^2) = \left\{ \dots, 0, 1, 1, 0, 0, 0, \dots \right.$$

\uparrow
0 1 2 3 4

3) Sketch the signal $y_2(n) = y(n-2)$

$$y(n-2) = \left\{ \dots, 0, 0, 1, 1, 0, 0, 0, \dots \right.$$

\uparrow
0

4) Sketch the signal $x_2(n) = x(n-2)$

$$x(n-2) = \left\{ \dots, 0, 0, 1, 1, 1, 1, 0, \dots \right.$$

\uparrow
0 1 2 3 4 5 6

5) Determine and sketch the signal $y_2(n) = T[x_2(n-2)]$

$$y_2(n) = T[x_2(n-2)] = \{ \dots, x(0), x(1), x(2), x(3), x(4), \dots \}$$

$$= \{ \dots, 0, 1, 0, 0, 0, 1, 0, \dots \}$$

\uparrow

6) Compare the signals $y_1(n)$ and $y_2(n-2)$. What is your conclusion?

$y_1(n) \neq y_2(n-2) \Rightarrow$ system is time variant

c) Repeat part (b) for the system $y(n) = x(n) - x(n-1)$. Can you see this statement about the time invariance of T ?

$$1) x(n) = \frac{\overline{1111}}{0123} \Rightarrow \{1, 1, 1, 1\}$$

$$2) y(n) = x(n) - x(n-1)$$

$$\begin{array}{c} x(n) \\ \hline 1 & 1 & 1 & 1 \\ 0 & 1 & 2 & 3 \\ \hline n \end{array}$$

$$\begin{array}{c} x(n-1) \\ \hline 1 & 1 & 1 & 1 \\ 0 & 1 & 2 & 3 \\ \hline 4 \end{array}$$

$$y(n) = x(n) - x(n-1) = \frac{1}{0123} = \{0, 1, 0, 0, -1\}$$

$$3) y(n-2) \Rightarrow$$

$$\frac{1}{012345} \Rightarrow \{0, 0, 1, 0, 0, -1\}$$

$$4) x(n-2)$$

$$\frac{1 & 1 & 1}{2345} \Rightarrow \{0, 0, 1, 1, 1\}$$

$$5) y_2(n) = \frac{0, 0, 1, 0, 0, -1}{\uparrow}$$

b) $y_2(n) = y(n-2) \Rightarrow$ system is time invariant.

d) Repeat parts (b) & (c) for the s/m $y(n) = T[x(n)] = nx(n)$

$$1) y(n) = nx(n)$$

$$x(n) = \{ \dots, 0, 1, 1, 1, 1, 0, \dots \}$$

\uparrow

$$2) y(n) = \{ \dots, 0, 1, 2, 3, 4, \dots \}$$

$$3) y(n-2) = \{ \dots, 0, 0, 0, 1, 2, 3, 4, \dots \}$$

$$4) x(n-2) = \{ \dots, 0, 0, 0, 1, 1, 1, \dots \}$$

\uparrow

$$5) y_2(n) = \gamma[x(n-2)] = \{-\dots, -5, -4, -3, -2, -1, 0, 1, 2, 3, 4, 5, \dots\}$$

6) $y_2(n) \neq y_1(n-2) \Rightarrow$ system is time variant.

- 7) ① static or dynamic
 2) linear or non-linear.
 3) Time invariant or varying.
 4) Causal or non-causal.
 5) Stable or unstable.

Examine the following S/I's w.r.t. the properties.

$$a) y(n) = \cos[x(n)]$$

i) static (only present i/p) iii) $y_1(n) = \cos[x_1(n)]$

$$(iii) y(n) = \cos[x(n-n_0)]$$

$$y'(n) = \cos[x(n-n_0)]$$

\Rightarrow Time Variant.

(iv) Only Present i/p \Rightarrow causal \Rightarrow Non-linear.

v) stable

$$b) y(n) = \sum_{k=-\infty}^{n+1} x(k)$$

Dynamic depends on future values. Linear, time invariant, non-causal (also depends on future value, unstable).

$$c) y(n) = x(n) \cos(\omega_0 n)$$

Static, linear, Time variant, Causal, stable.

$$y(n) = x(n-n_0) \cos(\omega_0 n + n_0)$$

$$y'(n) = x(n-n_0) \cos(\omega_0 n)$$

$$d) y(n) = x(-n+2)$$

Dynamic

$$\text{at } n=0 \Rightarrow y(0) = x(2)$$

future value

$$y_1(n) = x_1(-n+2) + x_2(-n+2)$$

$$y_2(n) = [x_1(-n+2) + x_2(-n+2)]$$

Linear

Non Causal, stable, Time invariant.

$$e) y(n) = \text{Trun}[x(n)]$$

Static, non-linear, time invariant, Causal, stable

$$f) y(n) = \text{Round}[x(n)]$$

Static, non-linear, time invariant, Causal, stable.

$$g) y(n) = |x(n)|$$

Static, non-linear, time invariant, Causal, stable

$$h) y(n) = x(n).u(n)$$

Static, linear, time invariant, causal, stable

$$i) y(n) = x(n) + n \cdot x(n+1)$$

Dynamic, linear, time variant, Non-causal, unstable.

$$j) y(n) = x(2n)$$

Dynamic, linear, time variant, Non-causal, stable.

$$k) y(n) = \begin{cases} x(n); & \text{if } x(n) \geq 0 \\ 0; & \text{if } x(n) < 0 \end{cases}$$

Static, linear, time invariant, Non-causal, stable.

$$l) y(n) = x(-n)$$

Dynamic, linear, Time invariant, Causal, stable.

$$m) y(n) = \text{sign}[x(n)]$$

Non linear, Time invariant, Causal, stable

- 1) The ideal sampling of a continuous signal.
- $x(n) = x_0(nT), -\infty < n < \infty$
- $x(n) = x_0(nT)$
is static, linear, time invariant, non-causal, stable.
- q) Let T be an LTI, relaxed and BIBO stable system with i/p $x(n)$ and o/p $y(n)$. Show that.
- If $x(n)$ is periodic with period N . [ie., $x(n) = x(n+N)$ for all $n \geq 0$], the output $y(n)$ tends to a periodic signal with the same period.
- $x(n) = x(n+N) \quad \forall n \geq 0$
- $y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$
- $y(n+N) = \sum_{k=-\infty}^{n+N} h(k) x(n+N-k)$
- $y(n+N) = \sum_{k=n+1}^{n+N} h(k) x(n-k) + \sum_{k=-\infty}^{n-1} h(k) x(n-k)$
- $y(n+N) = y(n) + \sum_{k=n+1}^{n+N} h(k) x(n-k)$
- For BIBO System $\lim_{n \rightarrow \infty} |h(n)| = 0$
- $\lim_{n \rightarrow \infty} \sum_{k=n+1}^{n+N} h(k) x(n-k) = 0$
- $\lim_{n \rightarrow \infty} y(n+N) = y(n)$
- $\therefore y(n) = y(n+N)$
- If $x(n)$ is bounded and tends to a constant, the output will also tend to a constant.

$x(n) = x_0(n) + a u(n)$

$x_0(n) \rightarrow$ bounded with $\lim_{n \rightarrow \infty}$

$\Rightarrow y(n) = a \sum_{k=0}^{\infty} h(k) u(n-k) + \sum_{k=0}^{\infty} h(k) x_0(n-k) = a \sum_{k=0}^{\infty} h(k) x_0(n-k) = a \sum_{k=0}^{\infty} h(k)$

$\Rightarrow \sum_n x_0^2(n) < \infty \Rightarrow \sum_n y_0^2(n) < \infty$

hence $\lim_{n \rightarrow \infty} |y_0(n)| = \infty = a \sum_{k=0}^{\infty} h(k) = \text{const.}$

- If $x(n)$ is an energy signal, the output $y(n)$ will also be an energy signal.

$y(n) = \sum_k h(k) x(n-k)$

$\sum_{-\infty}^{\infty} y^2(n) = \sum_{-\infty}^{\infty} \left[\sum_k h(k) x(n-k) \right]^2 = \sum_k \sum_{n_1}^{\infty} h(k) h(k) \sum_n x(n-k) x(n)$

but $\sum_n x(n-k) x(n-1) \leq \sum_n x^2(n) / h(1)$

for BIBO stable system $\sum_k |h(k)| < M$

Hence $E_y \leq M^2 E_x$, so that $E_y < \infty$ if $E_x < \infty$

- The following i/P-DP pairs have been observed during the operation of a time invariant system:

$x_1(n) = \begin{cases} 1, 0, 2 \\ \uparrow \end{cases} \xrightarrow{T} y_1(n) = \begin{cases} 0, 1, 2 \\ \uparrow \end{cases}$

$x_2(n) = \begin{cases} 0, 0, 3 \\ \uparrow \end{cases} \xrightarrow{T} y_2(n) = \begin{cases} 0, 1, 0, 2 \\ \uparrow \end{cases}$

$x_3(n) = \begin{cases} 0, 0, 0, 1 \\ \uparrow \end{cases} \xrightarrow{T} y_3(n) = \begin{cases} 1, 2, 1 \\ \uparrow \end{cases}$

Can you draw any conclusion regarding the linearity of the S/m. What is the impulse response of the system?

As this is a time-invariant system.
 $y_2(n)$ should have only 3 elements and
 $y_3(n)$ should have 4 elements.
 So, it is non-linear.

- 1) The following I/P-O/P pairs have been observed during the operation of a linear system?

$$x_1(n) = \{1, 2, 1\} \xrightarrow{T} y_1(n) = \{1, 2, -1, 0, 1\}$$

$$x_2(n) = \{1, -1, 1\} \xrightarrow{T} y_2(n) = \{-1, 1, 0, 2\}$$

$$x_3(n) = \{0, 1, 1\} \xrightarrow{T} y_3(n) = \{1, 2, 1\}$$

Can you draw any conclusions about the time invariance of this system?

Since $x_1(n) + x_2(n) = \delta(n)$, system is linear, the impulse response of the S/m is

$$y_1(n) + y_2(n) = \{0, 3, -1, 2, 1\}$$

If S/m were time invariant the response of $x_3(n)$ would be $\{3, 2, 1, 3, 1\}$

- 2) The only available information about the S/m consists of N I/P-O/P pairs of signals $y(n) = \{y_k(n)\}_{k=1}^N$

- a) What is the class of I/P signals for which we can determine the O/P, using the information

above, if the S/m is known to linear?
 Any linear combination of signal in the form of

$$x_i(n), i=1, 2, \dots, N$$

because if we take $i=1, 3$

$$y_1(n) = x_1(n) \Rightarrow y(n) = y_1(n) + y_3(n) = x_1(n) + x_3(n)$$

$$y_3(n) = x_3(n)$$

$$y(n) = x_1(n) + x_3(n)$$

linear

- b) Same repeat for the S/m is invariant.

Any $x_i(n-k)$ where k is any integer; $i=1, 2, \dots, N$

1st replace $n=n-n_0 \Rightarrow n_i(n-n_0-k)$

$x(n)$ by $x(n-n_0) \Rightarrow x_i(n-k-n_0)$

[Time invariant]

- 13) Show that the necessary and sufficient condition for a relaxed LTI S/m to be BIBO stable is

$$\sum_{n=-\infty}^{\infty} |h(n)| \leq m_n < \infty \text{ for constant } m_n$$

A system to be BIBO stable only when bounded O/P produce bounded input

$$y(n) = \sum_k h(k) x(n-k)$$

$$|y(n)| = \sum_k |h(k)| |x(n-k)|$$

$$= \sum_k |x(n-k)| \leq m_n \text{ [some constant]}$$

$$\text{so } |y(n)| = M_n \sum_{k=0}^n |h(k)|$$

$|y(n)| < \infty$ for all n , if and only if

$$\sum_{k=0}^{\infty} |h(k)| < \infty \text{ so } \sum_{n=0}^{\infty} |y(n)|$$

\Rightarrow A s/m to be BIBO stable only when bounded if and bounded output.

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k); n \leq n-k$$

$$|y(n)| = \sum_{k=-\infty}^{\infty} |h(k)| |x(n-k)|$$

as $\sum_{k=-\infty}^{\infty} |x(n-k)| \leq M_n$ for some const.

$$|y(n)| = M_n \sum_{k=-\infty}^{\infty} |h(k)|; n \leq n-k$$

$|y(n)|$ is $< \infty$ if and only if $\sum_{k=-\infty}^{\infty} |h(k)| < \infty$

$$\text{so } \sum_{n=-\infty}^{\infty} |h(n)| < \infty.$$

Q4) Show that

a) A related line s/m is causal if and only if for any $x(n)$ such that $x(n)=0$ for $n < n_0 \Rightarrow y(n)=0$ for $n < n_0$.

If a system is causal o/p depends only on present and past inputs as $x(n)=0$ for $n < n_0$ the $y(n)$ also became zero for $n < n_0$.

b) A related LTI s/m is causal if and only if

$$h(n)=0 \text{ for } n < 0$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

For finite impulse response.

$$h(n)=0, n < 0 \text{ and } n \geq m.$$

$$\text{so } y(n) \text{ reduces to } y(n) = \sum_{k=0}^{n-1} h(k) x(n-k)$$

$$y(n) = \sum_{k=0}^{\infty} h(k) x(n-k)$$

15) Show that for any real or complex constants, and any finite integer numbers m and N , we have:

$$a) \sum_{n=M}^N a^n = \begin{cases} \frac{a^m - a^{N+1}}{1-a}, & \text{if } a \neq 1 \\ N-m+1, & \text{if } a=1 \end{cases}$$

$$\text{for } a=1, \sum_{n=M}^N a^n = N-m+1$$

$$\text{for } a \neq 1, \sum_{n=M}^N a^n = \frac{a^m - a^{N+1}}{1-a}$$

$$(1-a)^N \sum_{n=M}^N a^n = a^m + a^{m+1} - a^{m+2} + \dots - a^{N-1} - a^N + a^{N+1}$$

$$= a^M - a^{N+1}$$

b) Show that if $|a| < 1$, then

$$\sum_{n=0}^{\infty} a^n = \frac{1}{1-a}$$

for $M=0, |a| < 1$ and $N \rightarrow \infty$

$$\sum_{n=0}^{\infty} a^n = \frac{1}{1-a}; |a| < 1$$

16) a) If $y(n) = x(n) * h(n)$, show that $E_y = \sum_x \sum_h$ where

$$\sum_x = \sum_{n=-\infty}^{\infty} x(n)$$

$$y(n) = \sum_k h(k) x(n-k)$$

$$\sum_n y(n) = \sum_n \sum_k h(k) x(n-k)$$

$$\sum_n y(n) = \sum_k h(k) \sum_{n=-\infty}^{\infty} x(n-k)$$

$$\sum_n y(n) = (\sum_k h(k)) (\sum_n x(n))$$

b) Compute the convolution $x(n) * h(n)$ of the following signals and check the correctness of the results by using the test in (a).

$$1) x(n) = \{1, 2, 4\}, h(n) = \{1, 1, 1, 1\}$$

$$y(n) = \{1, 3, 7, 7, 7, 6, 4\}$$

$$\sum_n y(n) = 35; \sum_n x(n) = 7, \sum_n h(n) = 5$$

$$\sum_n y(n) = \sum_n x(n) \cdot \sum_n h(n)$$

$$35 = 7 \times 5$$

$$35 = 35$$

$$\begin{array}{c} h(n) \\ \diagdown \quad \diagup \\ x(n) \end{array} \begin{array}{r} 1 \quad 2 \quad 4 \\ \hline 1 \quad 1 \quad 2 \quad 4 \\ 1 \quad 1 \quad 2 \quad 4 \\ 1 \quad 1 \quad 2 \quad 4 \\ 1 \quad 1 \quad 2 \quad 4 \end{array}$$

$$2) x(n) = \{1, 2, -1\}, h(n) = x(n)$$

$$x(n) = \{1, 2, -1\}, h(n) = \{1, 2, -1\}$$

$$y(n) = x(n) * h(n)$$

$$y(n) = \{1, 4, 2, -4, 1\}$$

$$\begin{array}{c} h(n) \\ \diagdown \quad \diagup \\ x(n) \end{array} \begin{array}{r} 1 \quad 2 \quad -1 \\ \hline 1 \quad 1 \quad 2 \quad -1 \\ 2 \quad 2 \quad 4 \quad 6 \\ -1 \quad -1 \quad -2 \quad -1 \end{array}$$

$$3) x(n) = \{0, 1, -2, 3, -4\}, h(n) = \{\frac{1}{2}, \frac{1}{2}, 1, \frac{1}{2}\}$$

$$y(n) = \{0, \frac{1}{2}, -\frac{1}{2}, \frac{3}{2}, -2, 0, -\frac{5}{2}, 2\}$$

$$\sum_n y(n) = -5, \sum_n x(n) = -2, \sum_n h(n) = \frac{5}{2}$$

$$\sum_n y(n) = \sum_n x(n) h(n)$$

$$-5 = -5$$

$$\begin{array}{c} h(n) \\ \diagdown \quad \diagup \\ x(n) \end{array} \begin{array}{r} 0 \quad 1 \quad -2 \quad +3 \quad -4 \\ \hline \frac{1}{2} \quad 0 \quad \frac{1}{2} \quad -1 \quad \frac{3}{2} \quad -2 \\ \frac{1}{2} \quad 0 \quad \frac{1}{2} \quad -1 \quad \frac{3}{2} \quad -2 \\ \frac{1}{2} \quad 0 \quad \frac{1}{2} \quad -1 \quad \frac{3}{2} \quad -2 \\ \frac{1}{2} \quad 0 \quad \frac{1}{2} \quad -1 \quad \frac{3}{2} \quad -2 \end{array}$$

$$4) x(n) = \{1, 2, 3, 4, 5\}, h(n) = \{1\}$$

$$y(n) = \{1, 2, 3, 4, 5\}$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

$$15 = 15(1)$$

$$15 = 15$$

$$5) x(n) = \{1, 2, 3\}, h(n) = \{0, 0, 1, 1, 1\}$$

$$y(n) = \{0, 0, 1, -1, 2, 2, 1, 3\}$$

$$\sum_n y(n) = 8; \sum_n x(n) = 2; \sum_n h(n) = 4$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

$$8 = 8(1)$$

$$\begin{array}{c} h(n) \\ \diagdown \quad \diagup \\ x(n) \end{array} \begin{array}{r} 1 \quad -2 \quad 3 \\ \hline 0 \quad 0 \quad 0 \\ 0 \quad 0 \quad 0 \\ 1 \quad -2 \quad 3 \end{array}$$

$$6) x(n) = \{0, 0, 1, 1, 1\}, h(n) = \{1, -2, 0\}$$

$$y(n) = \{0, 0, 1, -1, 2, 2, 1, 3\}$$

$\sum_n y(n) = 8, \sum_n x(n) = 4, \sum_n h(n) = 2$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

x(n)	0	0	1	1	1
h(n)	1	0	0	1	1
y(n)	-2	0	0	-2	-2
	3	0	0	3	3

$$7) x(n) = \{0, 1, 4, -5\}, h(n) = \{1, 0, -1\}$$

$$y(n) = \{0, 1, 4, -4, -5, -1, 3\}$$

$\sum_n y(n) = -2, \sum_n x(n) = -2, \sum_n h(n) = 1$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

x(n)	0	1	4	-3
h(n)	1	0	1	-1
y(n)	0	0	0	0
	-1	0	-1	-4
	-1	0	-1	-4

$$8) x(n) = \{1, 1, 2\}, h(n) = u(n)$$

$$y(n) = \{1, 2, 4, 3, 2\}$$

$$\sum_n y(n) = 12, \sum_n x(n) = 4, \sum_n h(n) = 3$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

x(n)	1	1	2
h(n)	1	1	1
y(n)	1	1	1
	1	1	2

$$12 = 4 \times 3$$

$$12 = 12$$

$$9) x(n) = \{1, 1, 0, 1, 1\}, h(n) = \{1, -2, -3, 4\}$$

$$y(n) = \{1, -1, -5, 2, 3, -5, 1, 4\}$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

x(n)	1	1	0	1	1
h(n)	1	1	1	0	1
y(n)	-2	-2	-2	0	-2
	-3	-3	-3	0	-3
	-4	-4	-4	0	-4

$$10) x(n) = \{1, 2, 0, 2, 1\}, h(n) = x(n)$$

$$y(n) = \{1, 4, 4, 4, 10, 4, 4, 4, 1\}$$

x(n)	1	2	0	2	1
h(n)	1	1	2	0	2
y(n)	2	2	4	0	4
	0	0	0	0	0
	2	2	4	0	4
	1	1	2	0	2

$$11) x(n) = \left(\frac{1}{2}\right)^n u(n), h(n) = \left(\frac{1}{4}\right)^n u(n)$$

$$y(n) = \left[2\left(\frac{1}{2}\right)^n - \left(-\frac{1}{4}\right)^n u(n)\right]$$

$$\sum_n y(n) = \frac{8}{3}, \sum_n h(n) = \frac{4}{3}, \sum_n x(n) = 2$$

17) Compute and plot convolutions $x(n) * h(n)$ and $h(n) * x(n)$ for the pairs of signal shown below.

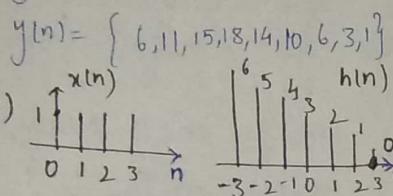
$$a) x(n)$$

$$x(n) = \{1, 1, 1, 1\}$$

x(n)	1	1	1	1
h(n)	6	5	4	3
y(n)	6	6	6	6
	5	5	5	5
	4	4	4	4
	3	3	3	3
	2	2	2	2
	1	1	1	1

$$x(n) = \{1, 1, 1, 1\}, h(n) = \{6, 5, 4, 3, 2, 1, 0, 0\}$$

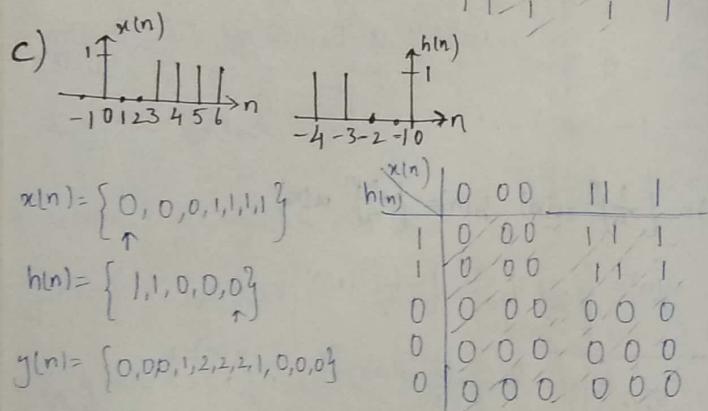
$$y(n) = x(n) * h(n)$$



$$x(n) = \begin{cases} 1, & n=0 \\ 0, & \text{elsewhere} \end{cases}$$

$$h(n) = \begin{cases} 1, & -3 \leq n \leq 3 \\ 0, & \text{elsewhere} \end{cases}$$

$$y(n) = \sum_{k=-3}^{3} x(k)h(n-k) = \begin{cases} 6, & n=-3, -2, -1, 0, 1, 2, 3 \\ 0, & \text{elsewhere} \end{cases}$$



$$x(n) = \begin{cases} 1, & n=0 \\ 0, & \text{elsewhere} \end{cases}$$

$$h(n) = \begin{cases} 1, & -4 \leq n \leq 0 \\ 0, & \text{elsewhere} \end{cases}$$

$$y(n) = \sum_{k=-4}^{0} x(k)h(n-k) = \begin{cases} 1, & n=-4, -3, -2, -1, 0 \\ 0, & \text{elsewhere} \end{cases}$$

18) Determine and sketch the convolution $y(n)$ of the signals.

$$x(n) = \begin{cases} \frac{1}{3}n, & 0 \leq n \leq 6 \\ 0, & \text{elsewhere} \end{cases}$$

$$h(n) = \begin{cases} 1, & -2 \leq n \leq 2 \\ 0, & \text{elsewhere} \end{cases}$$

a) graphically:

$$x(n) = \begin{cases} 0, & n < 0 \\ \frac{1}{3}, & n=0 \\ \frac{2}{3}, & n=1 \\ \frac{4}{3}, & n=2 \\ \frac{5}{3}, & n=3 \\ \frac{6}{3}, & n=4 \\ \frac{7}{3}, & n=5 \\ \frac{8}{3}, & n=6 \end{cases}$$

$$h(n) = \begin{cases} 1, & -2 \leq n \leq 2 \\ 0, & \text{elsewhere} \end{cases}$$

$$y(n) = \begin{cases} 0, & n < -2 \\ \frac{1}{3}, & n=-1 \\ \frac{10}{3}, & n=0 \\ \frac{20}{3}, & n=1 \\ \frac{11}{3}, & n=2 \\ \frac{12}{3}, & n=3 \end{cases}$$

b) Analytically:

$$x(n) = \frac{1}{3}n[u(n) - u(n-7)]$$

$$h(n) = u(n+2) - u(n-3)$$

$$\begin{aligned} y(n) &= \frac{1}{3}n[u(n) - u(n-7)] * u(n+2) - u(n-3) \\ &= \frac{1}{3}n[u(n) * u(n+2) - u(n-7) * u(n+2) - u(n) * u(n-3) + u(n-7) * u(n-3)] \\ &\quad * \frac{1}{3}n[u(n+2) + u(n-1) * u(n-3)] \end{aligned}$$

20) Consider the following three operations.

a) multiply the integer numbers: 131 and 122.

$$131 \times 122 = 15982$$

b) Compute the convolution of signals: $\{1, 3, 1\} * \{1, 2, 2\}$

$$y(n) = \{1, 5, 9, 8, 2\}$$

$$y(n) = \begin{cases} 1, & n=-2 \\ 5, & n=-1 \\ 9, & n=0 \\ 8, & n=1 \\ 2, & n=2 \\ 0, & \text{elsewhere} \end{cases}$$

c) multiply the polynomials:

$$1+3z+z^2 \text{ and } 1+2z+2z^2.$$

$$(z^2+3z+1) * (2z^2+2z+1) = z^4+8z^3+9z^2+5z+1$$

c) Repeat part (a) for the numbers 1.31 and 12.2.
 $1.31 \times 12.2 = 15.982$

e) Comment on your results.

These are some of the ways to perform convolution.

2) Compute the convolution $y(n) * h(n)$ of the following pairs of the signals.

a) $x(n) = a^n u(n)$, $h(n) = b^n u(n)$, when $a \neq b$, and when $a = b$.

$$\begin{aligned} y(n) &= x(n) * h(n) \\ &= a^n u(n) * b^n u(n) \\ &= [a^n * b^n] u(n) \\ g(n) &= \sum_{k=0}^n a^k u(k) b^{n-k} u(n-k) \\ &= b^n \sum_{k=0}^n a^k u(k) b^{-k} \\ &= b^n \sum_{k=0}^n (ab)^{-k} \end{aligned}$$

$$\text{if } a \neq b, \text{ then } y(n) = \frac{b^{n+1} - a^{n+1}}{b-a} u(n)$$

$$\text{if } a=b \Rightarrow b^n (n+1) u(n)$$

$$b) x(n) = \begin{cases} 1; & n = -2, 1 \\ 2; & n = -1 \\ 0; & \text{elsewhere} \end{cases} \quad h(n) = \delta(n) - \delta(n-1) + \delta(n-4) + \delta(n-5)$$

$$x(n) = \{1, 2, 1, 1\} \quad h(n) = \{1, -1, 0, 0, 1, 1\}$$

$$y(n) = \{1, 1, -1, 0, 0, 3, 3, 2, 1\}$$

$$\begin{aligned} c) \quad x(n) &= u(n+1) - u(n-2) - \delta(n-5) \\ h(n) &= [u(n+2) - u(n-3)](3 - 1)n! \\ x(n) &= \{1, 1, 1, 1, 0, -1\} \\ h(n) &= \{1, 2, 3, 2, 1\} \\ y(n) &= \{1, 3, 6, 8, 9, 8, 5, 1, -2, -2, 1\} \end{aligned}$$

$$\begin{array}{c|cccccc} x(n) & 1 & 1 & 1 & 1 & 0 & -1 \\ \hline h(n) & 1 & 1 & 1 & 1 & 1 & 0 \\ 2 & 2 & 2 & 2 & 2 & 2 & -2 \\ 3 & 3 & 3 & 3 & 3 & 3 & -3 \\ 2 & 2 & 2 & 2 & 2 & 2 & -2 \\ 1 & 1 & 1 & 1 & 1 & 1 & 0 \end{array}$$

$$\begin{aligned} d) \quad x(n) &= u(n) - u(n-5); \quad h(n) = u(n-2) - u(n+8) + u(n-11) - u(n-17) \\ x(n) &= \{1, 1, 1, 1, 1\} \\ h(n) &= \{0, 0, 1, 1, 1, 1\} \\ h(n) &= h'(n) + h(n-9) \\ y(n) &= y'(n) + y(n-9); \text{ where} \\ y'(n) &= \{0, 0, 1, 2, 3, 4, 5, 5, 4, 3, 2, 1\} \end{aligned}$$

22) Let $x(n)$ be the IIP signal to a discrete time filter with impulse response $h(n)$ and let $y_i(n)$ be the corresponding output. Compute and sketch $x(n)$ and $y_i(n)$ in the following

case using the small scale in all the figures.

$$m(n) = \{1, 4, 2, 3, 5, 3, 4, 5, 7, 6, 9\}$$

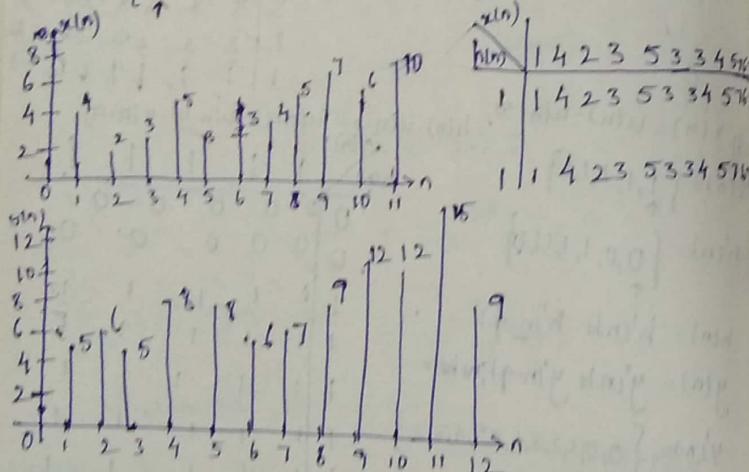
$$h_1(n) = \{1, 1\}, h_2(n) = \{1, 2, 1\}, h_3(n) = \{1/2, 1/2\}, h_4(n) = \{1/4, 1/2\}$$

$$h_5(n) = \{1/4, -1/2, 1/4\}$$

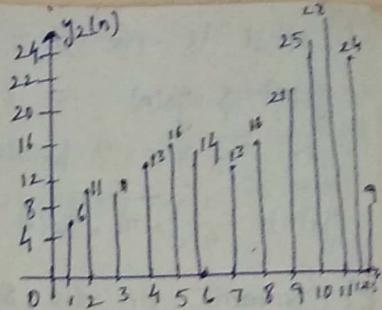
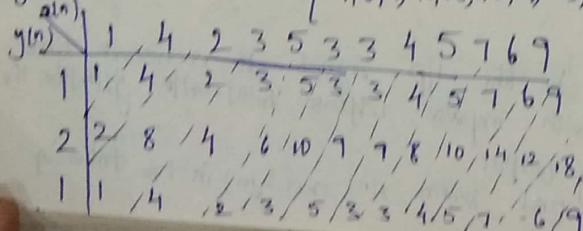
Sketch $x(n), y_1(n), y_2(n)$ on one graph and $x(n), y_3(n), y_4(n), y_5(n)$ on another graph.

$$y_1(n) = x(n) * h_1(n)$$

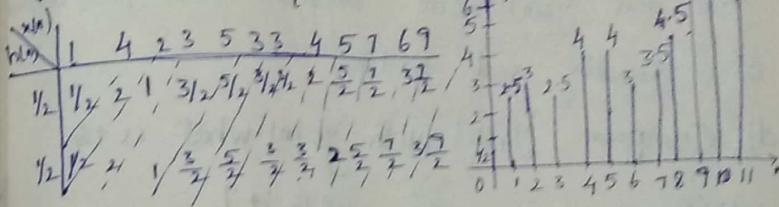
$$y_1(n) = \{1, 5, 6, 5, 8, 8, 6, 7, 9, 12, 12, 15, 9\}$$



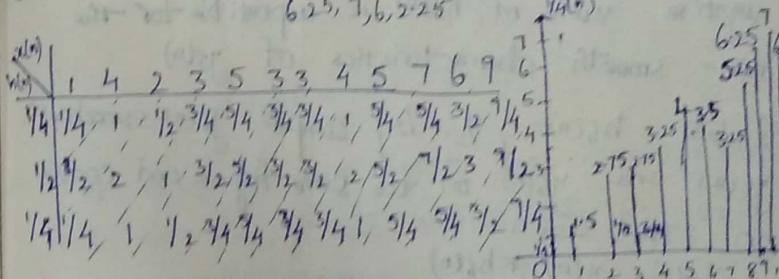
$$y_2(n) = x(n) * h_2(n) = \{1, 6, 11, 11, 13, 16, 14, 13, 16, 21, 25, 23, 24, 7\}$$



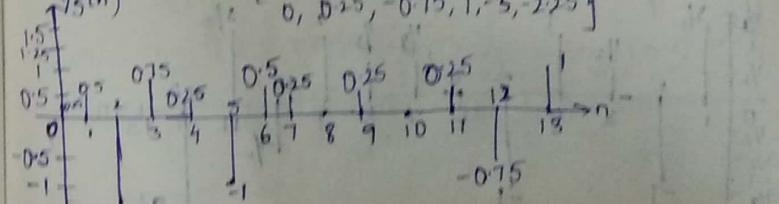
$$\Rightarrow y_3(n) = x(n) * h_3(n) = \{1/2, 2, 5, 3, 2, 5, 4, 4, 3, 3, 5, 4, 5, 6, 6, 7\}$$



$$y_4(n) = x(n) * h_4(n) = \{0.25, 1.5, 2.75, 2.75, 3.25, 4, 3.5, 3.25, 5.25, 6.25, 7, 6, 2.25\}$$



$$y_5(n) = x(n) * h_5(n) = \{0.25, 0.5, -1.25, 0.75, 0.25, -1.05, 0.25, 0.25, 0.25, 0, 0.25, -0.75, 1, -3, -2.25\}$$



b) What is the diff. b/w $y_2(n)$ and $y_4(n)$ and b/w

$y_3(n)$ & $y_4(n)$.

$$y_3(n) = \frac{1}{2} y_1(n); h_3(n) = \frac{1}{2} h_1(n)$$

$$y_4(n) = \frac{1}{4} y_2(n); h_4(n) = \frac{1}{4} h_2(n)$$

c) Comment on the smoothness of $y_2(n)$ & $y_4(n)$. Which factors affect the smoothness?

$y_2(n)$ and $y_4(n)$ are smoother than $y_1(n)$ because of smaller scale factor.

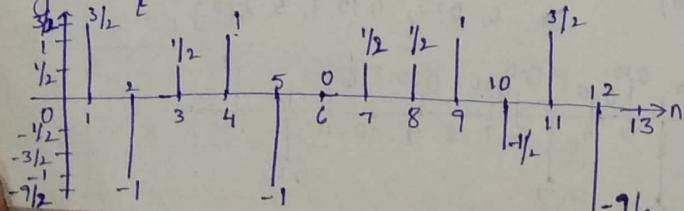
d) Compare $y_4(n)$ with $y_5(n)$. What is the difference can you explain it?

$y_4(n)$ results in smoother output than $y_5(n)$. The negative value of $h_5(0)$ is responsible for the non-smooth characteristics of $y_5(n)$.

e) Let $h_6(n) = \left\{ \frac{1}{2}, -\frac{1}{2} \right\}$ Compute $y_6(n)$. Sketch $x(n)$, $y_2(n)$ and $y_6(n)$ on the same figure and comment on the results.

$$y_6(n) = x(n) * h_6(n)$$

$$y_6(n) = \left\{ \frac{1}{2}, \frac{3}{2}, -1, \frac{1}{2}, 1, -1, 0, \frac{1}{2}, \frac{1}{2}, 1, -\frac{1}{2}, \frac{3}{2}, -\frac{9}{2} \right\}$$



$y_2(n)$ is more smaller than $y_6(n)$.

23) Express the output $y(n)$ of a linear time-invariant system with impulse response $h(n)$ in terms of its step response $\delta(n) = h(n) * u(n)$ and the input $x(n)$.

We can express $\delta(n) = u(n) - u(n-1)$

$$h(n) = h(n) * \delta(n)$$

$$= h(n) * [u(n) - u(n-1)]$$

$$= h(n) * u(n) - h(n) * u(n-1)$$

$$= \delta(n) - \delta(n-1)$$

then $y(n) = h(n) * x(n)$

$$= [\delta(n) - \delta(n-1)] * x(n)$$

$$= \delta(n) * x(n) - \delta(n-1) * x(n)$$

24) The discrete time system $y(n) = ny(n-1) + x(n)$, $n \geq 0$ is at rest [i.e., $y(-1) = 0$], check if the s/m is LTI and BIBO stable.

$$y(n) = ny_1(n-1) + x_1(n), \quad n \geq 0$$

$$y_1(n) = ny_1(n-1) + x_1(n) \quad \text{if } n > 0 \\ \Rightarrow y(n) = ny_1(n-1) + x_1(n)$$

$$y_2(n) = ny_2(n-1) + x_2(n) \quad \text{if } n > 0 \\ \Rightarrow y(n) = ny_2(n-1) + x_2(n)$$

$$y(n) = ny(n-1) + x(n)$$

$$x(n) = ny(n-1) + f(n)$$

$$x(n) = ax(n) + bx(n)$$

$$y(n) = ay(n) + by(n)$$

Hence the system is linear.

$$\Rightarrow y(n-1) = (n-1)y(n-2) + x(n-1)$$

$$\text{delayed} \Rightarrow y(n-1) = ny(n-2) + x(n-1)$$

so the system is time variant.

\Rightarrow If $x(n) = u(n)$, then $|x(n)| \leq 1$, for this bounded input, output is $y(0)=0, y(1)=2, y(2)=5 \dots$ unbounded

So system is unstable.

25) Consider the signal $s(n) = a^n u(n)$; $a > 1$

a) show that any sequence $x(n)$ can be decomposed as $x(n) = \sum_{n=-\infty}^{\infty} c_k s(n-k)$ and express c_k in terms of $s(n)$

$$s(n) = f(n) - af(n-1)$$

$$s(n-k) = f(n-k) - af(n-k-1)$$

$$x(n) = \sum_{k=-\infty}^{\infty} x(k) s(n-k)$$

$$= \sum_{k=-\infty}^{\infty} x(k) [f(n-k) - af(n-k-1)]$$

$$x(n) = \sum_{k=-\infty}^{\infty} x(k) f(n-k) - a \sum_{k=-\infty}^{\infty} x(k) f(n-k-1)$$

$$x(n) = \sum_{k=-\infty}^{\infty} x(k) f(n-k) - a \sum_{k=-\infty}^{\infty} x(k-1) f(n-k)$$

$$= \sum_{k=-\infty}^{\infty} [x(k) - ax(k-1)] f(n-k)$$

$$\text{Thus } c_k = x(k) - ax(k-1)$$

b) Use the property of linearity and time invariance to express the output $y(n) = T[x(n)]$ in terms of the input $x(n)$ and the signal $g(n) = T[f(n)]$, where $T[\cdot]$ is an LTI sm

$$y(n) = T[x(n)]$$

$$= T\left[\sum_{k=-\infty}^{\infty} c_k f(n-k)\right]$$

$$= \sum_{k=-\infty}^{\infty} c_k T[f(n-k)]$$

$$= \sum_{k=-\infty}^{\infty} c_k g(n-k)$$

c) Express the impulse response $h(n) = T[s(n)]$ in terms of $g(n)$.

$$h(n) = T[s(n)]$$

$$h(n) = T[f(n) - af(n-1)]$$

$$= g(n) - ag(n-1)$$

26) Determine the zero i/p-response of the sm described by the second order differential equation

$$(x(n) - 3y(n-1) - 4y(n-2)) = 0$$

With $x(n) = 0$

$$-3y(n-1) - 4y(n-2) = 0 \quad (\div (-3))$$

$$y(n-1) + \frac{4}{3}y(n-2) = 0$$

at $n=0$

$$y(-1) = -\frac{4}{3}y(-2)$$

$$\text{at } n=1 \quad y(0) = -\frac{4}{3}y(-1) = \left(-\frac{4}{3}\right)^2 y(-2)$$

$$y(1) = \left(-\frac{4}{3}\right)^3 y(-2)$$

$$y(k) = \left(-\frac{4}{3}\right)^{k+2} y(-2)$$

\rightarrow zero if p response.

27) Determine the particular solution of the difference equation $y(n) = \frac{5}{6}y(n-1) - \frac{1}{6}y(n-2) + x(n)$ when the forcing function is $x(n) = 2^n u(n)$

$$y(n) = \frac{5}{6}y(n-1) - \frac{1}{6}y(n-2) + x(n)$$

$$x(n) = y(n) - \frac{5}{6}y(n-1) + \frac{1}{6}y(n-2)$$

Characteristic equation is

$$\lambda^2 - \frac{5}{6}\lambda + \frac{1}{6} = 0; \quad \lambda = \frac{1}{2}, \frac{1}{3}$$

$$\text{So } y_h(n) = C_1 \left(\frac{1}{2}\right)^n + C_2 \left(\frac{1}{3}\right)^n$$

$$x(n) = 2^n u(n) \quad y_{pl}(n) = k[2^n] u(n)$$

$$\text{so } k[2^n] u(n) - k\left(\frac{5}{6}\right)(2^{n-1}) u(n-1) + k\left(\frac{1}{6}\right)(2^{n-2}) u(n-2) \\ = 2^n u(n)$$

for $n=2$

$$4k - \frac{5k}{3} + \frac{k}{6} = 4$$

$$k = \frac{8}{5}$$

Total solution is

$$y_{pl}(n) + y_h(n) = y(n)$$

$$y(n) = \frac{8}{5}(2^n) u(n) + C_1 \left(\frac{1}{2}\right)^n u(n) + C_2 \left(\frac{1}{3}\right)^n u(n)$$

$$\text{Assume } y(-2) = y(-1) = 0 \quad \text{so } y(0) = 1$$

$$\text{then } y(1) = \frac{5}{6}y(0) + 2 = 17/6$$

$$\text{so } \frac{8}{5} + C_1 + C_2 = 1 \quad C_1 + C_2 = 3/5 \rightarrow 0$$

$$\frac{16}{5} + \frac{1}{2}C_1 + \frac{1}{3}C_2 = 17/6$$

$$3C_1 + 2C_2 = -11/5 \rightarrow 2$$

By solving ① & ②

$$C_1 = -1, C_2 = 2/5$$

so, the total solution is

$$y(n) = \left[\frac{8}{5}(2)^n - \left(\frac{1}{2}\right)^n + \frac{2}{5}\left(\frac{1}{3}\right)^n\right] u(n)$$

$$28) \text{ In the given equation } y(n) = (-a_1)^{n+1} y(-1) + \frac{(1-a_1)^{n+1}}{1+a_1}$$

for $n \geq 1$ separate old sequence $y(n)$ into the transient response and the steady state response. Plot these responses for $a_1 = 0.9$.

at $y(-1) = 1$

$$\text{The given equation } y(n) = \frac{(-a)^{n+1} + (1-a)^{n+1}}{1+a}$$

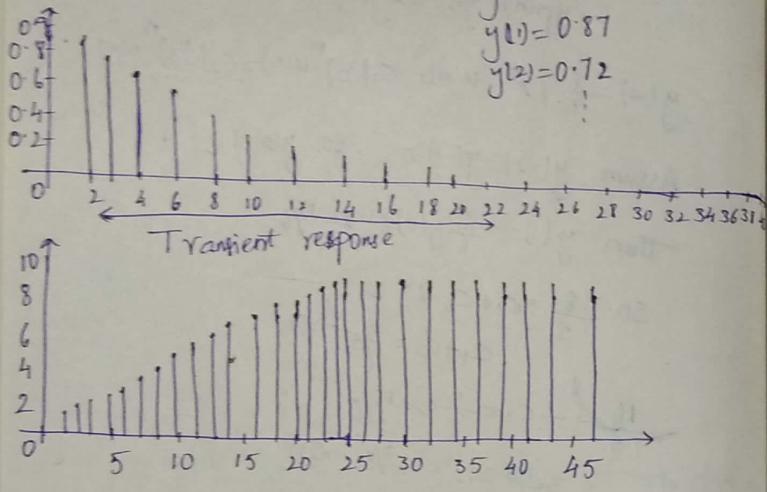
$$y(n) = y_{zi}(n) + y_{zs}(n)$$

= Transient + Steady state

$$y(0) = 0.9$$

$$y(1) = 0.87$$

$$y(2) = 0.72$$



← Steady state response →

29) Determine the impulse response for the cascade of two linear time invariant systems having impulse responses.

$$h_1(n) = a^n [u(n) - u(n-N)] \text{ and } h_2(n) = [u(n) - u(n-m)]$$

$$h(n) = h_1(n) * h_2(n)$$

$$= \sum_{k=-\infty}^{\infty} a^k [u(k)u(k-N)] [u(n-k) - u(n-k-m)]$$

$$= \sum_{k=-\infty}^{\infty} a^k u(k) u(n-k) - \sum_{k=-\infty}^{\infty} a^k u(k) u(n-k-m) -$$

$$\sum_{k=-\infty}^{\infty} a^k u(k) u(n-k) + \sum_{k=-\infty}^{\infty} a^k u(k-N) u(n-k-m)$$

$$= \left(\sum_{k=0}^N a^k - \sum_{k=0}^{n-M} a^k \right) - \left(\sum_{k=N}^n a^k - \sum_{k=N}^{n-M} a^k \right)$$

$$h(n) = 0$$

30) Determine the response $y(n)$, $n \geq 0$ of the system described by the 2nd order difference equation:

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

to the input $x(n) = 4^n u(n)$.

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

characteristic equation is

$$\lambda^2 - 3\lambda - 4 = 0$$

$$\lambda = 4, -1$$

$$so \quad \lambda_h(n) = C_1 4^n + C_2 (-1)^n$$

$$x(n) = 4^n u(n)$$

$$y_p(n) = k n 4^n u(n)$$

$$k n 4^n u(n) - 3k(n-1) 4^{n-1} u(n-1) - 4k(n-2) 4^{n-2} u(n-2)$$

$$= 4^n u(n) + 2(-1)^{n-1} u(n-1)$$

$$\text{for } n=2 \quad k(32-12) = 4^2 + 8 = 24 \rightarrow k = 6/5$$

The total solution is

$$y(n) = y_p(n) + y_h(n)$$

$$= \left[\frac{6}{5} n_4^n + c_1 4^n + c_2 (-1)^n \right] u(n)$$

To find c_1 and c_2 , let $y(-2) = y(-1) = 0$ then

$$y(0) = 1$$

$$y(1) = 3 y(0) + 4 + 2 = 9$$

$$c_2 + c_1 = 1 \rightarrow ①$$

$$\frac{24}{5} + 4c_1 - c_2 = 9 \Rightarrow 4c_1 - c_2 = \frac{21}{5} \rightarrow ②$$

From ① & ②

$$c_1 = \frac{26}{25} \text{ & } c_2 = \frac{-1}{25}$$

$$\text{so } y(n) = \left[\frac{6}{5} n_4^n + \frac{26}{25} 4^n - \frac{1}{25} (-1)^n \right] u(n)$$

31) Determine the impulse response of the following causal system $y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$

Solve characteristic equation $\lambda^2 - 3\lambda - 4 = 0$

$$\lambda = -1, 4$$

$$y_h(n) = c_1 4^n + c_2 (-1)^n$$

$$x(n) = \delta(n)$$

$$y(0) = 1 \text{ and } y(1) - 3y(0) = 2$$

$$\therefore c_1 - c_2 = 5$$

$$\text{so } c_1 + c_2 = 1 \rightarrow ①$$

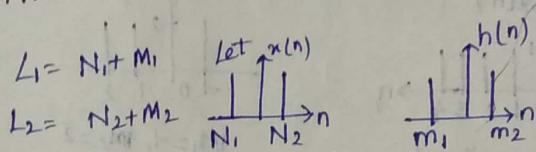
$$4c_1 - c_2 = 5 \rightarrow ②$$

From ① & ② $c_1 = 6/5$ & $c_2 = -1/5$

$$\therefore h(n) = \left[\frac{6}{5} 4^n - \frac{1}{5} (-1)^n \right] u(n)$$

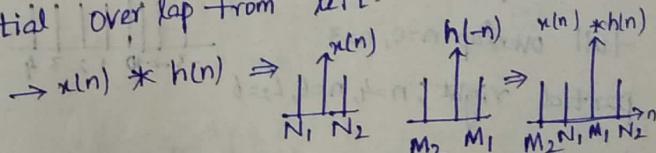
32) Let $x(n)$, $N_1 \leq n \leq N_2$ and $h(n)$, $M_1 \leq n \leq M_2$ be two finite duration signals.

a) Determine the range $L_1 \leq n \leq L_2$ of their convolution in terms of N_1, N_2, M_1 and M_2 .



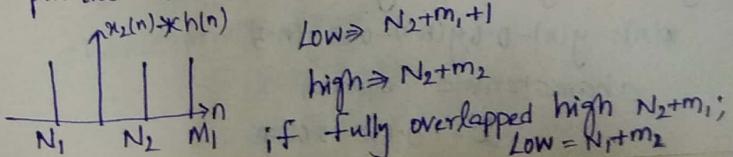
b) Determine the limits of the cases of partial overlap from the left, full overlap and partial overlap from right. For convenience, assume that $h(n)$ has shorter duration than $x(n)$.

partial overlap from left



LOW $N_1 + M_1$ & high $m_2 + N_1 - 1$
It fully overlaps then $N_1 + m_2$ (LOW) & high $N_2 + m_1$

partial overlap from right



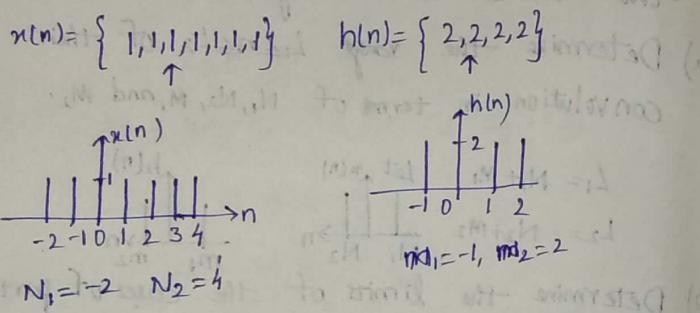
$$\text{LOW} \Rightarrow N_2 + m_1 + 1$$

$$\text{high} \Rightarrow N_2 + m_2$$

if fully overlapped high $N_2 + m_1$
 $\text{LOW} = N_1 + m_2$

c) Illustrate the validity of your result by computing the convolution of the signal $x(n)$

$$= \begin{cases} 1, & -2 \leq n \leq 4 \\ 0, & \text{elsewhere} \end{cases} \quad h(n) = \begin{cases} 2, & -1 \leq n \leq 2 \\ 0, & \text{elsewhere} \end{cases}$$



Partial overlap from left

$$\text{low } N_1 + m_1 = -3 \\ \text{high } m_2 + N_1 - 1 = 2 - 2 - 1 = -1$$

full overlap, $n=0, n=3$

partial right, $n=4, n=6, l_2=6$

33) Determine the impulse response and the unit step response the systems described by the difference equation:

$$a) y(n) = 0.6y(n-1) - 0.08y(n-2) + x(n)$$

$$x(n) = y(n) - 0.6y(n-1) - 0.08y(n-2)$$

Characteristic equation

$$\lambda^2 - 0.6\lambda + 0.08 = 0$$

$$\lambda = 1/2, 1/5$$

$$y(n) = c_1(1/5)^n + c_2(2/5)^n$$

Impulse response $x(n) = \delta(n)$ with $y(0)=1$

$$y(1) - 0.6y(0) = 0 \Rightarrow y(1) = 0.6$$

$$\text{so } c_1 + c_2 = 1 \rightarrow ①$$

$$\frac{1}{5}c_1 + \frac{2}{5}c_2 = 0.6 \rightarrow ②$$

$$\text{from } ① \& ② \quad c_1 = -1, c_2 = 3$$

$$\therefore h(n) = [-(\frac{1}{5})^n + 2(\frac{2}{5})^n]u(n)$$

step response $x(n) = u(n)$

$$\delta(n) = \sum_{k=0}^n y_h(n-k), n \geq 0$$

$$= \sum_{k=0}^n [2(\frac{2}{5})^{n-k} - (\frac{1}{5})^{n-k}]$$

$$= 2(\frac{2}{5})^n \sum_{k=0}^n (\frac{2}{5})^k - (\frac{1}{5})^n \sum_{k=0}^n (\frac{1}{5})^k$$

$$= \left\{ 2\left(\frac{2}{5}\right)^n \left[\left(\frac{2}{5}\right)^{n+1} - 1 \right] - \left(\frac{1}{5}\right)^n \left[5^{n+1} - 1 \right] \right\} u(n)$$

$$b) y(n) = 0.7y(n-1) - 0.1y(n-2) + 2x(n) - x(n-2)$$

$$2x(n) - x(n-2) = y(n) - 0.7y(n-1) + 0.1y(n-2)$$

characteristic equation,

$$\lambda^2 - 0.7\lambda + 0.1 = 0$$

$$\lambda = 1/2, 1/5$$

$$y[n] = C_1 \left(\frac{1}{2}\right)^n + C_2 \left(\frac{1}{5}\right)^n$$

Impulse response $x[n] = \delta[n]$ $y[0] = 2$

$$y[1] - 0.7y[0] = 0 \Rightarrow y[1] = 1.4$$

$$C_1 + C_2 = 2$$

$$\frac{1}{2}C_1 + \frac{1}{5}C_2 = \frac{7}{5} \rightarrow ①$$

$$C_1 + \frac{2}{5}C_2 = \frac{14}{5} \rightarrow ②$$

Solving ① & ②

$$C_1 = \frac{10}{3}, C_2 = -\frac{4}{3}$$

$$\text{So, } h[n] = \left[\frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u[n]$$

$$\text{Step response } s[n] = \sum_{k=0}^n h[n-k]$$

$$\begin{aligned} &= \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k} \\ &= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k \\ &= \frac{10}{3} \left[\frac{1}{2}^n (2^{n+1} - 1) u(n) \right] - \frac{4}{3} \left[\frac{1}{5}^n (5^{n+1} - 1) u(n) \right] \end{aligned}$$

34) Consider a S/I with impulse response

$$h[n] = \begin{cases} \left(\frac{1}{2}\right)^n, & 0 \leq n \leq 4 \\ 0, & \text{elsewhere} \end{cases}$$

Determine the i/p $x[n]$ for $0 \leq n \leq \infty$ that will generate the o/p sequence

$$y[n] = \{1, 2, 2.5, 3, 3, 2, 1, 0\}$$

$$h[n] = \{1, \frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{16}, \dots\}$$

$$y[n] = \{1, 2, 2.5, 3, 3, 2, 1, 0, \dots\}$$

$$y[0] = x[0] h[0]$$

$$y[0] = x[0] \cdot 1 \Rightarrow x[0] = 1$$

$$y[1] = x[1] + h[1] x[0]$$

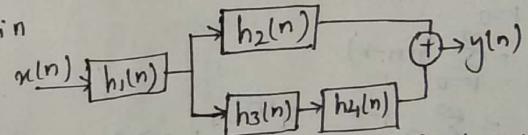
$$2 = x[1] + \frac{1}{2}(1) \Rightarrow x[1] = \frac{3}{2}$$

$$y[2] = x[2] + h[2] x[1] + h[1] x[0]$$

$$2.5 = x[2] + \frac{1}{4}(\frac{3}{2}) + \frac{1}{2}(1)$$

$$\text{So } x[n] = \{\frac{1}{2}, \frac{3}{2}, \frac{3}{2}, \frac{7}{4}, \frac{3}{2}, \dots\}$$

35) Consider the interconnection of LTI S/I's as shown in



a) Express the overall impulse response in terms of $h_1[n]$, $h_2[n]$, $h_3[n]$, $h_4[n]$

$$h[n] = h_1[n] * [h_2[n] - \{h_3[n] * h_4[n]\}]$$

b) Determine $h[n]$ when $h_1[n] = \{\frac{1}{2}, \frac{1}{4}, \frac{1}{2}\}$

$$h_2[n] = h_3[n] = (n+1) u[n]$$

$$h_4[n] = \delta[n-2]$$

$$h_3[n] * h_4[n] = (n+1) u(n) * \delta(n-2)$$

$$= (n+1) u(n-2) = (n+1) u(n-2)$$

$$h_2[n] - [h_3[n] * h_4[n]] = (n+1) u(n) - (n+1) u(n-2)$$

$$= 2u[n] - \delta(n)$$

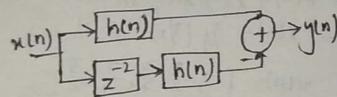
$$h_1[n] = \frac{1}{2} \delta(n) + \frac{1}{4} \delta(n-1) + \frac{1}{2} \delta(n-2)$$

$$h[n] = \left[\frac{1}{2} \delta(n) + \frac{1}{4} \delta(n-1) + \frac{1}{2} \delta(n-2) \right] * [2u[n] - \delta(n)]$$

$$= \frac{1}{2} \delta(n) + \frac{1}{2} \delta(n-1) + 2 \delta(n-2) + \frac{5}{2} \delta(n-3)$$

- c) Determine the response of the S/m in part (b) if
 $x(n) = \delta(n+2) + 3 \delta(n-1) - 4 \delta(n-3)$
 $\star(n) = \begin{cases} 1, 0, 0, 3, 0, -4 \end{cases}$

- 36) Consider the S/m fig with $h(n) = a^n u(n), -1 < a < 1$.
Determine the response $y(n)$ of the S/m to the excitation. $x(n) = u(n+5) - u(n-10)$



$$s(n) = u(n) * h(n)$$

$$s(n) = \sum_{k=0}^{\infty} u(k) h(n-k)$$

$k=0$

$$= \sum_{k=0}^{\infty} h(n-k)$$

$$= \sum_{k=0}^{\infty} a^{n-k} = \frac{a^{n+1}}{a-1}; n \geq 0$$

For $x(n) = u(n+5) - u(n-10)$ then

$$s(n+5) - s(n-10) = \frac{a^{n+6}-1}{a-1} u(n+5) - \frac{a^{-9}-1}{a-1} u(n-10)$$

from given figure $y(n) = x(n) * h(n) - x(n) * h(n-2)$

$$y(n) = \frac{a^{n+6}-1}{a-1} u(n+5) - \frac{a^{-9}-1}{a-1} u(n-10) -$$

$$\frac{a^{n+4}-1}{a-1} u(n+3) + \frac{a^{n-11}-1}{a-1} u(n-12)$$

- 37) Compute and sketch step response of the S/m.

$$y(n) = \frac{1}{M} \sum_{k=0}^{M-1} x(n-k)$$

$$h(n) = \left[\frac{u(n) - u(n-M)}{M} \right]$$

$$s(n) = \sum_{k=-\infty}^{\infty} u(k) h(n-k)$$

$$= \sum_{k=0}^n h(n-k) = \begin{cases} \frac{n+1}{M}, & n \leq M \\ 1, & n > M \end{cases}$$

- 38) Determine the range of values of the parameter a for which the linear time-invariant system with impulse response

$$h(n) = \begin{cases} a^n; n \geq 0, n \text{ even} \\ 0, \text{ otherwise} \end{cases}$$

$$\sum_{k=-\infty}^{\infty} |h(n)| = \sum_{n=0}^{\infty} |a|^n$$

$$= \sum_{n=0}^{\infty} |a|^n = \frac{1}{1-|a|^2}$$

stable if $|a| < 1$

- 39) Determine the response of the system with impulse response $h(n) = a^n u(n)$, to i/p signal $x(n) = u(n) - u(n-10)$.

$$h(n) = a^n u(n)$$

$$y_1(n) = \sum_{k=0}^{\infty} u(k) h(n-k)$$

$$= \sum_{k=0}^n a^{n-k}$$

$$= a^m \sum_{k=0}^n a^{-k} = \frac{1-a^{n+1}}{1-a} u(n)$$

$$y(n) = y_1(n) - y_1(n-10)$$

$$= \frac{1}{1-a} \left[(1-a^{n+1}) u(n) - (1-a^{n-9}) u(n-10) \right]$$

41) Determine the response of S/I/M characterized by the impulse response $h(n) = \left(\frac{1}{2}\right)^n u(n)$ to the input signals.

a) $x(n) = 2^n u(n)$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= \sum_{k=0}^n \left(\frac{1}{2}\right)^k 2^{n-k}$$

$$= 2^n \sum_{k=0}^n \left(\frac{1}{4}\right)^k$$

$$= 2^n \left[1 - \left(\frac{1}{4}\right)^{n+1} \right] \left(\frac{4}{3}\right)$$

$$= \frac{2}{3} \left[2^{n+1} - \left(\frac{1}{2}\right)^{n+1} \right] u(n)$$

b) $x(n) = u(-n)$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= \sum_{k=0}^{\infty} h(k) = \sum_{k=0}^{\infty} \left(\frac{1}{2}\right)^k = 2, n \geq 0$$

$$y(n) = \sum_{k=n}^{\infty} h(k)$$

$$= \sum_{k=n}^{\infty} \left(\frac{1}{2}\right)^k$$

$$= \sum_{k=0}^{\infty} \left(\frac{1}{2}\right)^k - \sum_{k=0}^{n-1} \left(\frac{1}{2}\right)^k$$

$$= 2 - \left(\frac{1 - \left(\frac{1}{2}\right)^n}{1 - \frac{1}{2}} \right) = 2 \left(\frac{1}{2}\right)^n, n \geq 0$$

42) Three S/I/Ms with impulse responses $h_1(n) = \delta(n) - \delta(n-1)$, $h_2(n) = h(n)$, and $h_3(n) = u(n)$, are connected in cascade.

a) What is the impulse response, $h_c(n)$ of the overall system.

$$h_c(n) = h_1(n) * h_2(n) * h_3(n)$$

$$= [\delta(n) - \delta(n-1)] * u(n) * h(n)$$

$$= [u(n) - u(n-1)] * h(n)$$

$$= \delta(n) * h(n) = h(n)$$

b) Does the order of interconnection affect the overall S/I/M?

No.

43) a) Prove and explain graphically the diff b/w the relations $x(n) \delta(n-n_0) = x(n_0)$, $\delta(n-n_0) = x(n-n_0)$ and $x(n) * \delta(n-n_0) = x(n-n_0)$.

$x(n) \delta(n-n_0) = x(n_0)$. Thus only the value of $x(n)$ at $n=n_0$ is of interest.

$x(n) * \delta(n-n_0) = x(n-n_0)$. Thus, we obtain shifted version of $x(n)$ sequence.

b) Show that a discrete-time system, which is described by a convolution summation, is LTI & relaxed.

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= h(n) * x(n)$$

$$\text{linearity: } x_1(n) \rightarrow y_1(n) = h(n) * x_1(n)$$

$$\begin{aligned} x_2(n) \rightarrow y_2(n) &= h(n) * x_2(n) \\ &= \alpha h(n) * x_1(n) + \beta h(n) * x_2(n) \\ &= \alpha y_1(n) + \beta y_2(n) \end{aligned}$$

Time invariance:

$$\begin{aligned} x(n) \rightarrow y_1(n) &= h(n) * x(n) \\ x(n-n_0) \rightarrow y_1(n) &= h(n) * x(n-n_0) \\ &= \sum_k h(k) x(n-n_0-k) \\ &= y(n-n_0) \end{aligned}$$

e) What is the impulse response of the system described by $y(n) = x(n-n_0)$

Ans:- $h(n) = \delta(n-n_0)$.

45) Compute the zero state response of the system described by the difference equation $y(n) + \frac{1}{2}y(n-1) = x(n) + 2x(n-2)$ to the input $x(n) = \{1, 2, 3, 4, 2, 1\}$ by solving the difference equation

recursively $y(n) = -\frac{1}{2}y(n-1) + x(n) + 2x(n-2)$

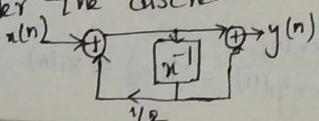
at $y(-2) = -\frac{1}{2}y(-3) + x(-2) + 2x(-4) = 1$

$y(-1) = -\frac{1}{2}y(-2) + x(-1) + 2x(-3) = \frac{3}{2}$

$y(0) = -\frac{1}{2}y(-1) + 2x(-2) + x(0) = \frac{17}{4}$

$y(1) = -\frac{1}{2}y(0) + x(1) + 2x(-1) = \frac{47}{8}$

47) Consider the discrete time shown in fig



a) Compute the 10 first samples of its impulse response.

b) Find the i/p-o/p relation.

c) Apply the i/p $x(n) = \{1, 1, 1, \dots\}$ and compute the first 10 samples of the output.

d) Compute the first 10 samples of the output for the i/p given in part (c) by using convolution.

e) Is the system causal? Is it stable?

a) $x(n) = \{1, 0, 0, \dots\}$

$y(n) = -\frac{1}{2}y(n-1) + x(n) + 2x(n-2)$

$y(0) = x(0) = 1$

$y(1) = -\frac{1}{2}y(0) + x(1) + 2x(0) = \frac{3}{2}$

$y(2) = -\frac{1}{2}y(1) + x(2) + 2x(1) = \frac{3}{4}$. Thus we obtain

$y(n) = \left\{ 1, \frac{3}{2}, \frac{3}{4}, \frac{3}{8}, \frac{3}{16}, \dots \right\}$

b) $y(n) = -\frac{1}{2}y(n-1) + x(n) + 2x(n-2)$

c) as in part (a) we obtain

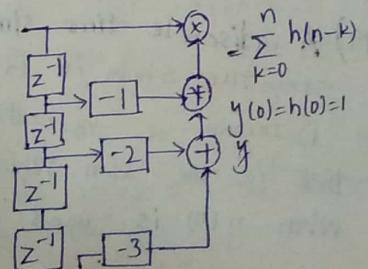
$y(n) = \left\{ 1, \frac{5}{2}, \frac{13}{4}, \frac{27}{8}, \frac{61}{16}, \dots \right\}$

d) $y(n) = u(n) * h(n)$

$= \sum_k u(k) h(n-k)$

$y(1) = h(0) + h(1) = \frac{5}{2}$

$y(2) = h(0) + h(1) + h(2) = \frac{13}{4}$.



e) from part (a), $h(n)=0$ for $n \leq 0 \Rightarrow$ The s/m is causal. $\sum_{n=0}^{\infty} |h(n)| = 1 + \frac{3}{2}(1 + \frac{1}{2} + \frac{1}{4} + \dots) = 4 \Rightarrow$ s/m is stable.

48) Consider the s/m described by the diff. eqn. $y(n) = ay(n-1) + bx(n)$.

a) Determine b in terms of a so that

$$\sum_{n=-\infty}^{\infty} h(n) = 1$$

b) Compute the zero-state step response $s(n)$ of the s/m and choose b so that $s(\infty) = 1$.

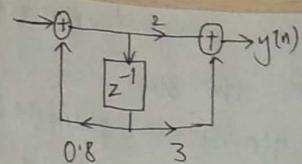
c) Compare the values of b obtained in parts (a) and (b). When did you notice?

$y(n) = ay(n-1) + bx(n)$ $h(n) = ba^n u(n)$ $\sum_{n=0}^{\infty} h(n) = \frac{b}{1-a}$ $b = 1-a$	$b) s(n) = \sum_{k=0}^n h(n-k)$ $= b \left[\frac{1-a^{n+1}}{a+1} \right] u(n)$ $s(\infty) = \frac{b}{1-a}$ $b = 1-a$
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c) $b = 1-a$ in both the cases.

49) A discrete-time s/m is realised by the structure shown in fig P2.49

a) Determine a realization for its inverse s/m, that is, the s/m which produces $x(n)$ as an op. when $y(n)$ is used as an ip.



$$a) y(n) = 0.8y(n-1) + 2x(n) + 3x(n-1)$$

$$y(n) - 0.8y(n-1) = 2x(n) + 3x(n-1)$$

The characteristic equation is

$$\lambda - 0.8 = 0$$

$$\lambda = 0.8$$

$$y(n) = C(0.8)^n$$

Let us first consider the response of the s/m.

$$y(n) - 0.8y(n-1) = x(n)$$

To $x(n) = \delta(n)$, since $y(0) = 1$, it follows that $C = 1$. Then, the impulse response of the original s/m is,

$$h(n) = 2(0.8)^n u(n) + 3(0.8)^{n-1} u(n-1)$$

$$= 2\delta(n) + 4 \cdot 6(0.8)^{n-1} u(n-1)$$

b) The inverse s/m is characterized by the difference equation.

$$x(n) = -1.5x(n-1) + \frac{1}{2}y(n) - 0.4y(n-1)$$

50) Consider the discrete-time s/m shown below

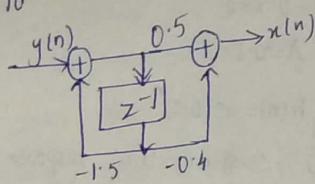
- Compute the first six values of the impulse response of a s/m.
- Compute the first six values of the zero-state step response of the s/m.

c) Determine an analytical expression for the impulse response of the SIR.

$$y(n) = 0.9 y(n-1) + x(n) + 2x(n-1) + 3x(n-2)$$

$$y(n) = 0.9 y(n-1) = x(n) + 2x(n-1) + 3x(n-2)$$

a) for $x(n) = \delta(n)$, we have
 $y(0) = 1, y(1) = 2.9, y(2) = 5.61, y(3) = 5.049, y(4) = 4.544$
 $y(5) = 4.090$



b) $s(0) = y(0) = 1$

$$s(1) = y(0) + y(1) = 3.9$$

$$s(2) = y(0) + y(1) + y(2) = 9.5$$

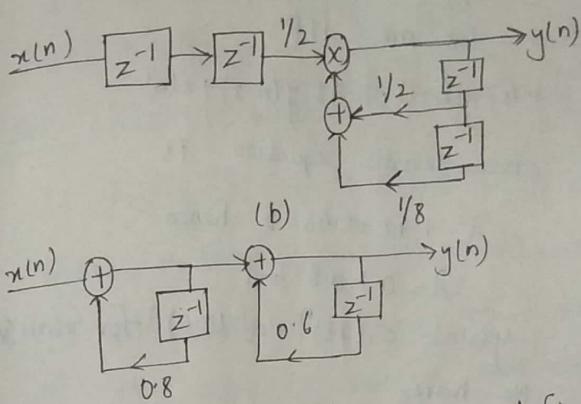
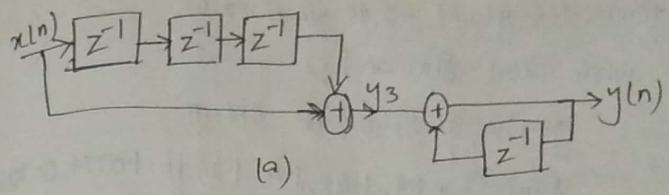
$$s(3) = y(0) + y(1) + y(2) + y(3) = 14.56$$

$$s(4) = \sum_0^4 y(n) = 19.10$$

$$s(5) = \sum_0^5 y(n) = 23.19$$

c) $h(n) = (0.9)^n u(n) + 2(0.9)^{n-1} u(n-1) + 3(0.9)^{n-2} u(n-2)$
 $= \delta(n) + 2.9 \delta(n-1) + 5.61 (0.9)^{n-2} u(n-2)$

5) Determine & sketch the impulse response of the following systems for $n=0, 1, \dots, 9$



a) fig P2.51(a) b) fig P2.51(b) c) fig P2.51(c)

d) classify the SIRs above as FIR or IIR.

e) find an explicit expression for the impulse response of the SIR in part(c).

a) $y(n) = \frac{1}{3} x(n) + \frac{1}{3} x(n-1) + y(n-1)$

for $x(n) = \delta(n)$, we have

$$h(n) = \left\{ \frac{1}{3}, \frac{1}{3}, \frac{1}{3}, \frac{2}{3}, \frac{2}{3}, \frac{2}{3}, \dots \right\}$$

b) $y(n) = \frac{1}{2} y(n-1) + \frac{1}{8} y(n-2) + \frac{1}{2} x(n-2)$

with $x(n) = \delta(n)$ and

$$y(-1) = y(-2) = 0, \text{ we obtain}$$

$$h(n) = \left\{ 0, 0, \frac{1}{2}, \frac{1}{4}, \frac{3}{16}, \frac{1}{8}, \frac{1}{128}, \frac{15}{256}, \frac{41}{1024}, \dots \right\}$$

c) $y(n) = 1.4y(n-1) - 0.48y(n-2) + x(n)$

with $x(n) = \delta(n)$ and

$$y(-1) = y(-2) = 0 \text{ we obtain}$$

$$h(n) = \{1, 1.4, 1.48, 1.4, 1.2496, 1.0774, 0.9086\}$$

d) All three s/m's are IIR.

e) $y(n) = 1.4y(n-1) - 0.48y(n-2) + x(n)$

The characteristic equation is

$$\lambda^2 - 1.4\lambda + 0.48 = 0 \text{ hence}$$

$$\lambda = 0.8, 0.6 \text{ and}$$

$$y(n) = C_1(0.8)^n + C_2(0.6)^n \text{ for } x(n) = \delta(n),$$

we have,

$$C_1 + C_2 = 1 \text{ and}$$

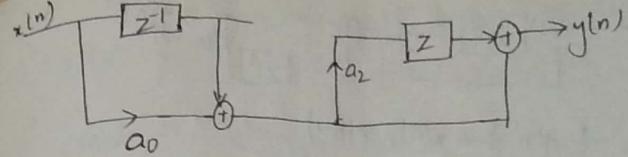
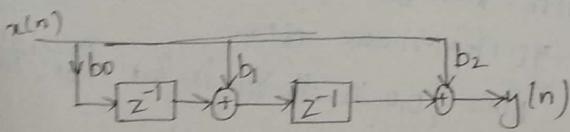
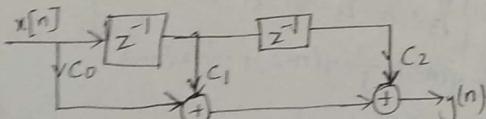
$$0.8C_1 + 0.6C_2 = 1.4$$

$$C_1 = 4$$

$$C_2 = -3$$

$$h(n) = [4(0.8)^n - 3(0.6)^n] u(n)$$

Q52) Consider s/m's shown below.



a) determine & sketch their impulse response $h_1(n)$, $h_2(n)$ & $h_3(n)$.

$$\text{Sol: } h_1(n) = c_0\delta[n] + c_1\delta[n-1] + c_2\delta[n-2]$$

$$h_2(n) = h_2\delta[n] + b_1\delta[n-1] + b_0\delta[n-2]$$

$$h_3(n) = a_0\delta[n] + (a_1 + a_0a_2)\delta[n-1] + a_0a_2\delta[n-2]$$

b) Is it possible to choose the coefficients of these s/m's in such a way that-

$$h_1(n) = h_2(n) = h_3(n).$$

Sol: The only question is whether

$$h_3(n) = h_2(n) = h_1(n)$$

$$a_0 = c_0$$

$$a_1 + a_0a_2 = c_1$$

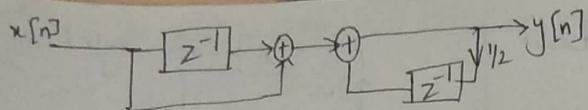
$$a_0a_2 = c_2$$

$$\text{Hence } \frac{c_2}{a_2} + a_2c_0 - c_1 = 0$$

$$c_0a_2^2 - c_1a_2 + c_2 = 0$$

For $c_0 \neq 0$ the quadratic eq. has real solution if and only if $c_1^2 - 4c_0c_2 \geq 0$

Q53) Consider the s/m shown in figure.
a) determine the impulse response $h[n]$.



$$y(n) = \frac{1}{2}y(n-1) + x(n) + x(n-1)$$

for $y(n) = \frac{1}{2}y(n-1) = \delta(n)$ the solution is

$$h(n) = \left(\frac{1}{2}\right)^n u[n] + \left(\frac{1}{2}\right)^{n-1} u[n-1]$$

b) Show that $h(n)$ is equal to the convolution of the following signals.

$$h_1(n) = \delta(n) + \delta(n-1)$$

$$h_2(n) = \left(\frac{1}{2}\right)^n u[n]$$

$$\text{sol: } h_1(n) * [\delta(n) + \delta(n-1)] = \left(\frac{1}{2}\right)^n u(n) + \left(\frac{1}{2}\right)^{n-1} u[n-1]$$

54) Compare and sketch the convolution $y_1[n]$ and correlation $r_1[n]$ sequences for the following pair of signals & comment on the results obtained.

$$a) x_1(n) = \{1, 2, 4\} \quad h_1(n) = \{1, 1, 1, 1\}$$

sol:-

$$\begin{array}{|c|c|c|c|} \hline & 1 & 2 & 4 \\ \hline 1 & | & 1 & 2 & 4 \\ 1 & | & 1 & 2 & 4 \\ 1 & | & 1 & 2 & 4 \\ 1 & | & 1 & 2 & 4 \\ \hline \end{array}$$

Convolution

$$y(n) = \{1, 3, 7, 7, 6, 4\}$$

Correlation

$$r_1(n) = \{1, 3, 7, 7, 6, 4\}$$

$$b) x_2(n) = \{0, 1, -2, 3, -4\} \quad h_2(n) = \{\frac{1}{2}, 1, 2, 1, \frac{1}{2}\}$$

$$\text{Convolution } y_2(n) = \{\frac{1}{2}, 0, \frac{3}{2}, -2, \frac{1}{2}, -6, -5, 2, 2\}$$

$$\text{Correlation: } r_2(n) = \{\frac{1}{2}, 0, \frac{3}{2}, -2, \frac{1}{2}, -6, -5, 2, 2\}$$

Note that $y_2(n) = r_2(n)$, because $h_2(n) = b_2[n]$

$$c) h_3(n) = \{1, 2, 3, 4\} \quad h_3(n) = \{4, 3, 2, 1\}$$

convolution, $y_3(n) = \{4, 11, 20, 30, 20, 11, 4\}$

correlation, $r_3(n) = \{1, 4, 10, 20, 25, 24, 16\}$

$$d) h_4(n) = \{1, 2, 3, 4\} \quad h_4(n) = \{1, 2, 3, 4\}$$

sol: Convolution $y_4(n) = \{1, 4, 10, 20, 25, 24, 16\}$

correlation $r_4(n) = \{4, 11, 20, 30, 20, 11, 4\}$

Note that $h_3[-n] = h_4[n+3]$

hence $r_3(n) = y_4(n+3)$

and $h_4(-n) = h_3(n+3)$

$$\Rightarrow r_4(n) = y_3(n+3)$$

55) The zero-state response of a causal LTI system to the IIP:

$$x[n] = \{1, 3, 3, 1\} \text{ is } y[n] = \{1, 4, 6, 4, 1\}$$

Determine its impulse response.

sol: the length of $h[n]$ is,

$$h[n] = \{h_0, h_1\}$$

$$h_0 = 1$$

$$3h_0 + h_1 = 4$$

$$\Rightarrow h_0 = 1, h_1 = 1$$

56) prove by direct substitution of equivalence of eqn $w[n] = -\sum_{k=1}^n a_{1k} w[n-k] + x[n]$ and $y(n) = \sum_{k=1}^m b_k w[n-k]$ which describe the direct form II structure to the

to the relation $y[n] = -\sum_{k=1}^M a_k y[n-k] + \sum_{k=0}^M b_k n[n-k]$, which describes the direct form I structure.

$$\text{Sol: } y[n] = -\sum_{k=1}^M a_k y[n-k] + \sum_{k=0}^M b_k x[n-k] \rightarrow 0$$

$$w[n] = -\sum_{k=1}^N a_k w[n-k] + x[n] \rightarrow ②$$

$$y[n] = \sum_{k=0}^M b_k w[n-k] \rightarrow ③$$

from eq ② we obtain,

$$x[n] = w[n] + \sum_{k=1}^N a_k w[n-k] \rightarrow ④$$

By substituting eq ③ for $y[n]$ and eq ④ into ①, we get LHS = RHS.

57) Determine the response $y[n]$, $n \geq 0$ of the S/I M described by the second order diff. equation

$$y[n] - 4y[n-1] + 4y[n-2] = x(n) - x(n-1)$$

where imp is $x[n] = (-1)^n u(n)$ & the initial condition are $y(-1) = y(-2) = 0$.

$$y[n] - 4y[n-1] + 4y[n-2] = x(n) - x(n-1)$$

The characteristic equation is

$$\lambda^2 - 4\lambda + 4 = 0$$

$$\lambda = 2e^{j\omega}$$

$$\text{Hence } y_h[n] = C_1 2^n + C_2 n 2^n$$

The particular solution is

$$y_p[n] = k(-1)^n u[n].$$

Substituting this solution into the diff. equation we

$$\begin{aligned} \text{obtain } k(-1)^n u[n] - 4k(-1)^{n-1} u[n-1] + 4k(-1)^{n-2} u[n-2] \\ = (-1)^n u(n) - (-1)^{n-1} u(n-1). \end{aligned}$$

for $A=2$, $K[1+4+4]=2 \Rightarrow k=2/9$. The total solution is

$$y[n] = [C_1 2^n + C_2 n 2^n + 2/9 (-1)^n] u(n)$$

from the initial conditions we obtain.

$$y[0] = 1, y[1] = 2.$$

$$\text{Then } C_1 + 2C_2 = 1$$

$$\Rightarrow C_1 = 7/9$$

$$2C_1 + 2C_2 - 2C_2 = 2$$

$$\frac{14}{9} - \frac{7}{9} + 3C_2 = 2 \Rightarrow 3C_2 = 2 - \frac{12}{9} = \frac{6}{9}$$

$$C_2 = \frac{3}{9} = \frac{1}{3}$$

58) Determine the impulse response $h[n]$ after the S/I M described by the second order diff. equation

$$y[n] - 4y[n-1] + 4y[n-2] = x[n] - x[n-1].$$

Sol: from prev. problem

$$h[n] = [C_1 2^n + C_2 n 2^n] u(n)$$

with $y[0] = 1, y[1] = 3$, we have

$$C_1 = 1, 2C_1 + 2C_2 = 3.$$

$$C_2 = 1/2$$

$$\text{This, } h[n] = [2^n + \frac{1}{2} n 2^n] u(n)$$

with $y[0] = 1, y[1] = 3$, we have $C_1 = 1$

$$2c_1 + 2c_2 = 3$$

$$\Rightarrow c_2 = \frac{3}{2}$$

$$\text{Thus, } h[n] = \left[2^n + \frac{1}{2} 2^n \right] u(n)$$

59) Show that any discrete-time signal $x(n)$ can be expressed by $x[n] = \sum_{k=-\infty}^{\infty} [x(k) - x(k-1)] u[n-k]$ where $u(n-k)$ is a unit step delayed by k units in time, that is $u(n-k) = \begin{cases} 1, & n \geq k \\ 0, & \text{otherwise.} \end{cases}$

$$x(n) = x(n) * \delta(n)$$

$$= x(n) * [u(n) - u(n-1)]$$

$$= [x(n) - x(n-1)] * u(n)$$

$$x(n) = \sum_{k=-\infty}^{\infty} [x(k) - x(k-1)] u(n-k)$$

60) Show that the output of an LTI system can be expressed in terms of its unit step response

$\delta(n)$ as follows.

$$y(n) = \sum_{k=-\infty}^{\infty} [\delta(k) - \delta(k-1)] x(n-k)$$

$$= \sum_{k=-\infty}^{\infty} [x(k) - x(k-1)] \delta(n-k)$$

Let $h(n)$ be the impulse response of the system.

$$h(k) = \sum_{m=-\infty}^k h(m)$$

$$h(k) = \delta(k) - \delta(k-1)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= \sum_{k=-\infty}^{\infty} [\delta(k) - \delta(k-1)] x(n-k)$$

61) Compute the correlation seq. $r_{xx}(l)$ and $r_{xy}(l)$ for the following signal sequences.

$$x(n) = \begin{cases} 1, & n_0 - N \leq n \leq n_0 + N \\ 0, & \text{otherwise.} \end{cases}$$

$$y(n) = \begin{cases} 0, & -N \leq n \leq N \\ 1, & \text{otherwise.} \end{cases}$$

$$x(n) = \begin{cases} 1, & n_0 - N \leq n \leq n_0 + N \\ 0, & \text{otherwise} \end{cases} \quad y(n) = \begin{cases} 1, & -N \leq n \leq N \\ 0, & \text{otherwise} \end{cases}$$

$$r_{xx}(l) = \sum_{n=-\infty}^{\infty} x(n) x(n-l)$$

The range of non-zero values of $r_{xx}(l)$ is determined by $n_0 - N \leq n \leq n_0 + N$

$$n_0 - N \leq n - l \leq n_0 + N$$

which implies

$$-2N \leq l \leq 2N$$

for a given shift l , the no. of terms in the summation for which both $x(n)$ and non-zero is $2N+1 - |l|$ and the value of each term is 1. Hence

$$r_{xx}(l) = \begin{cases} 2N+1 - |l|, & -2N \leq l \leq 2N \\ 0, & \text{otherwise.} \end{cases}$$

for $r_{xy}(l)$ we have

$$r_{xy}(l) = \begin{cases} 2N+1 - |l|, & n_0 - 2N \leq l \leq n_0 + 2N \\ 0, & \text{otherwise.} \end{cases}$$