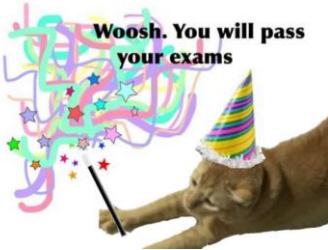


Fourier formula	Geometric Series	Causal: $n \leq 0$ Anti-Causal: $n > 0$ Non-Causal: $n > 0 \& n \leq 0$ Abs Sum: $\sum_n x[n] < \infty$ Square Sum: $\sum_n x[n] ^2 < \infty$ Energy: $\sum_{n=-\infty}^{\infty} x[n] ^2$ Power: $\frac{1}{2L+1} \sum_{n=-L}^{n=L} x[n] ^2$ P.Power: $\frac{1}{N} \sum_{n=0}^{N-1} x[n] ^2$ Power(Freq): $\sum_{n=0}^{N-1} C_k ^2$ Stable System: 1) BIBO ($ x[n] < B_x < \infty$) 2) $h[n]$ abs Summable
	Euler's Formula $\cos \theta = (e^{j\theta} + e^{-j\theta})/2 = \sin(\theta + (\pi/2))$ $\sin \theta = (e^{j\theta} - e^{-j\theta})/2j = \cos(\theta - (\pi/2))$ $X = \sin(at) + \cos(bt)$ $T_x = \text{lcm}(1/a, 1/b)$	
		
	<u>Parseval Identity</u> DTFS: $\frac{1}{N} \sum_{n=0}^{N-1} x[n] ^2 = \sum_{n=0}^{N-1} C_k ^2$ DTFT: $\sum_{n=-\infty}^{\infty} x[n] ^2 = \int_0^{2\pi} X(e^{j\omega}) ^2 d\omega$	
Evolution Core	Z-transform	Trigo Identities
		PFE

Down sampling

Up sampling



Filter Specifications

Cut-off Freq : $\omega_c = 0.5A$ (A = amplitude)
 Cut-off Attenuation: $20\log_{10}|H(e^{j\omega_c})| = -3\text{dB}$
 Passband Edge : $\omega_p = \text{Freq which } (1 + \delta_p) > A > (1 - \delta_p)$
 Passband Attenuation: $20\log_{10}(1 + \delta_p)/(1 - \delta_p)$
 Stopband Edge : $\omega_s = \text{Freq where } A < \delta_s$
 Stopband Attenuation: $20\log_{10}\delta_s$

Non-Int Conversion Ratio

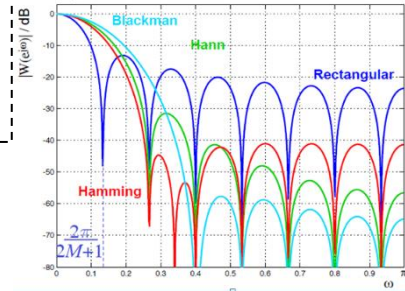
Upsample (Interpolator) then Downsample (Decimator) (i.e. L = 100, M = 101, T' = 1.01T)
 LPF Cut-off = $\min(\frac{\pi}{M}, \frac{\pi}{L})$, Gain = L

Increasing Window Length

- 1) Main Lobe Reduced
- 2) Height of first Side Lobe same
- 3) Narrower transition band

$$|H(\omega)| = \frac{1}{N} \left| \frac{\sin(\frac{N\omega}{2})}{\sin(\frac{\omega}{2})} \right|$$

$$\angle H(\omega) = -\frac{(N-1)\omega}{2}$$



Finite Impulse Response (FIR) Filter Design (Zero Terms)

$$h_i(n \neq 0) = \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} H(\omega) e^{j\omega n} d\omega \quad h_i(n = 0) = \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} H(\omega) e^{j\omega 0} d\omega$$

$$h_i(n \neq 0) = \frac{\sin(n\omega_c)}{\pi n} \quad h_i(n = 0) = \frac{\omega_c}{\pi}$$

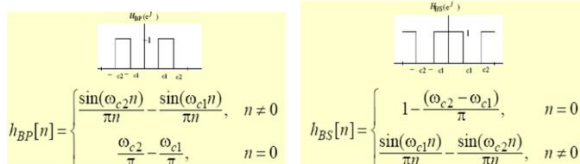
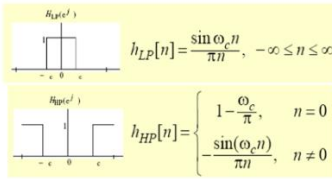
Hamming = $0.54 + 0.46 \cos(n\pi/N) - N$ is order, = L-1

Hanning = $0.5(1 - \cos(2\pi n/N))$

Filter Length Estimation:

$$\text{Kaiser} : N \cong \frac{-10\log_{10}(\delta_p\delta_s) - 13}{14.6 \frac{(\omega_s - \omega_p)}{2\pi}}$$

$$\text{Bellanger} : N \cong \frac{-2\log_{10}(10\delta_p\delta_s)}{3(f_s - f_p)} - 1$$



$$H(e^{j\omega}) = e^{-j\alpha\omega} \{H_r(e^{j\omega})\} = e^{-j\theta(\omega)} \{H_r(e^{j\omega})\}$$

Phase shift Real part

Since $H_r(e^{j\omega})$ is real, its inverse DTFT is even and symmetric
 Since $\theta(\omega) = \alpha\omega$ is linear phase, so it merely shifts the corresponding even and symmetric magnitude in the time domain, so the resulting impulse response is symmetric.

$H_r(e^{j\omega})$ is imaginary then IDTFT is odd symmetric.

The frequency responses of the four types of FIR filters are summarized below:

$$H(e^{j\omega}) = e^{-j[(N/2)\omega]} \left\{ h\left(\frac{N}{2}\right) + 2 \sum_{n=1}^{N/2} h\left[\frac{N}{2} - n\right] \cos(n\omega) \right\}$$

for type I

$$H(e^{j\omega}) = e^{-j[(N/2)\omega]} \left\{ 2 \sum_{n=1}^{(N+1)/2} h\left[\frac{N+1}{2} - n\right] \cos\left(\left(n - \frac{1}{2}\right)\omega\right) \right\}$$

for type II

$$H(e^{j\omega}) = e^{-j[(N\omega - \pi)/2]} \left\{ 2 \sum_{n=1}^{N/2} h\left[\frac{N}{2} - n\right] \sin(n\omega) \right\}$$

for type III

$$H(e^{j\omega}) = e^{-j[(N\omega - \pi)/2]} \left\{ 2 \sum_{n=1}^{(N+1)/2} h\left[\frac{N+1}{2} - n\right] \sin\left(\left(n - \frac{1}{2}\right)\omega\right) \right\}$$

for type IV