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COEN6521 Design for Testability

Project Report Design Testing of 8-Bit Multiplier Using JTAG

Submitted to Dr. Mahmoud Masadeh

Submitted by Group 2

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DATE: 06-08-2022

Abstract

Design for testability (DFT), is a design strategy that, by including extra circuitry on the chip, makes testing a chip feasible and affordable. It also enhances the controllability and observability of internal nodes to allow the testing of embedded functionalities. This project aims to test an 8-bit array multiplier which is a combinational circuit by incorporating JTAG IEEE std 1149.1 (boundary-scan) within it.

The following tools are used in this project,

- Mentor Graphics VHDL Modelsim tool
- Mentor Graphics Precision RTL Logic Synthesis tool

This project report provides brief description about,

- i. Designed VLSI system (8-bit multiplier)
- ii. Testing the VLSI system (structural vs behavioral)
- iii. JTAG inserted into the VLSI system.
- iv. RTL synthesis of various built combinational components.
- v. Overhead and benefits of the added JTAG.
- vi. Testing the VLSI using the JTAG.

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1. Introduction

Complex chips are being planned, developed, and manufactured as advances in integrated circuit (IC) manufacturing technology continue to lower defect density and minimum feature size. It can be difficult to ensure that a VLSI circuit is fault-free, and the time and effort required to test for issues can dramatically raise the cost of IC fabrication. The significant time and labor costs associated with creating test vector sequences for VLSI circuits are attempted to be reduced by design-fortestability (DFT) methodologies. Board-level test issues were first resolved by the Joint Test Action Group (JTAG). This project aims to create JTAG for testing 8-bit array multiplier. The JTAG design includes Test Access Ports (TAP), Data Registers (Boundary Scan Register, Bypass Register), Instruction Register, Instruction Decoder, TAP Controller.

2. Designed VLSI system (8-bit array multiplier)

The 8-bit combinational array multiplier has two 8-bit values, multiplies them, and gives 16-bit output. The multiplier has 64 AND gates and 56 full adders in its construction. These are arranged in an array fashion, with the input of the subsequent row of full adders being the nth output from the first row's full adders. An output joins the product when it appears at the start of a row.

3. Testing the VLSI system

Between 60 and 80 percent of the design process is spent testing. To achieve high yield and accurate detection of defective chips after production, a well-structured testing procedure must be used.

3.1. Structural Testing

Structural test is a testing procedure that finds all chip structure faults. Stuck-at faults, coupling faults, short-circuits, crosstalk, etc. are a few examples. The structural test offers a way to check for even unmodeled manufacturing flaws. All possible directions of datapath can be tested during structural level testing. The process variation-aware test values aid in the detection of flaws even when process variations are present.

Without considering the circuit's overall functionality, structural testing aims to find manufacturing flaws by validating the correct operation of a circuit's underlying

gates and connections. A structured DFT is being used to analyze the circuit, by evaluating the internal system and accepting control input. There are several ways to carry this out. Scan design is a popular and helpful approach that may be used to change the internal sequential circuitry architecture.

3.2. Functional Testing

Even if a circuit passes a structural test, it does not guarantee that it will operate as expected. A functional test is a testing procedure that confirms the tested circuit performs as intended. Only the functional behavior might be understood by the users in VLSI devices; the precise circuit implementation is always unknown. Based on the functional behavior of the circuit being tested, a systematic method may be utilized to identify and locate stuck-at and bridging faults on the primary input and output lines.

4. JTAG

JTAG is an industry standard for verifying designs and testing printed circuit boards after production. It was called after the Joint Test Action Group, which defined it. As an additional tool to digital simulation in electronic design automation (EDA), JTAG provides standards for on-chip instrumentation.

JTAG components,

- Test Access Port (TAP)
- TAP controller
- Registers
 - > Instruction Register,
 - Boundary Scan Register,
 - Bypass Register
- Instruction Decoder

4.1. TAP

JTAG offers four mandatory pins in the configuration as follows,

- TDI (Test Data Input),
- TDO (Test Data Output),
- TCK (Test Clock),
- TMS (Test Mode Select)

These pins are collectively referred to as the Test Access Port (TAP).

4.2.TAP Controller

A 16-state machine called a TAP controller is programmed by the inputs Test Mode Select (TMS) and Test Clock (TCK), and it regulates the flow of data bits to the Instruction Register (IR) and the Data Registers (DR).

On the positive edge of TCK, bits are shifted in, and on the negative edge, they are moved out. The register into which the bits are moved is controlled by the TMS signal (instruction register, bypass register or boundary scan register). A register is captured, a new value is shifted in from TDI while the old value is simultaneously shifted out on TDO, and then the register is updated with the new value. The instruction register or one of the other registers can be used by the TAP controller to shift values.

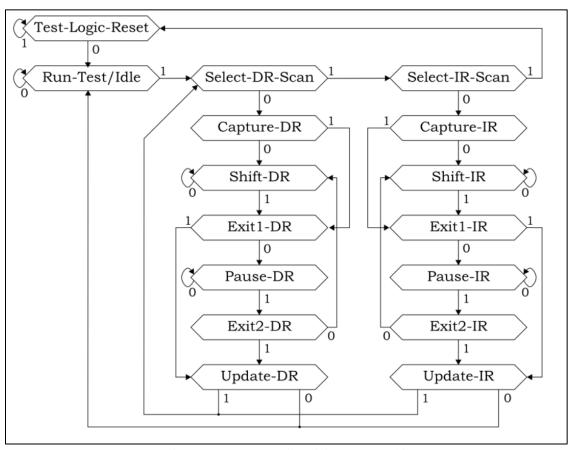


Figure 1: TAP Controller Finite State Machine

On the rising edge of TCK, TMS is sampled and used to advance the state. The following steps were taken in each state:

Test-Logic-Reset

All test modes, including extest mode, are reset in this condition, disabling them from operating and enabling the chip to continue operating normally. The external

logic will force TMS high on startup for at least five TCK cycles. This ensures that the Test-Logic-Reset condition is reached and maintained.

Run-Test/Idle

When everything is running normally, this is the resting state.

Select-DR-Scan, Select-IR-Scan

These are the corresponding beginning states for accessing the instruction register or one of the data registers.

Capture-DR, Capture-IR

These capture the current value of the instruction register or one of the data registers into the scan cells, respectively. This is a little misnomer for the instruction register since, when using Capture-IR, status information is often captured rather than the actual instruction.

Shift-DR, Shift-IR

Shift a bit from the presently chosen data or instruction register into TDI (on the rising edge of TCK) and out onto TDO (on the falling edge of TCK), accordingly.

Exit1-DR, Exit1-IR

These are the shift states' respective exit states. From this point, the state machine has two options: update state or pause state.

Pause-DR, Pause-IR

Data shifting into the instruction or data register is paused. This enables, for instance, the reloading of buffers on test equipment that supplies TDO.

Exit2-DR, Exit2-IR

These are the equivalent pause state's departure states. The state machine can then either start shifting again or go into the update stage.

Update-DR, **Update-IR**

The chip's inputs or the connection are used to drive the value that was moved into the scan cells during the previous states (for outputs). With the ability to pause during the shifting, our basic state machine enables either data registers or the instruction register to complete its capture-shift-update cycle.

4.3. Register

A minimum of two data registers, the boundary scan register, and the bypass register, must be included in a boundary scan logic design.

4.3.1. Boundary Scan Data Register

The Boundary Register, which has a boundary-scan cell next to each input and output pin, is the most significant. The device's input and output pins can be controlled and monitored using this register. IEEE 1149.1 requires the Boundary register as a feature. By entering 0000 into the instruction register, the boundary scan data register is chosen. According to the requirements of the IEEE 1149.1 standard, the boundary scan register is operated by the TAP controller's Shift-DR, Update-DR, and Capture-DR states.

Each of the processor interface pins may be accessed serially using the boundary scan register. Therefore, it is possible to load and monitor logic values on the processor pins using the boundary scan register. Board-level connection testing is the primary use of the boundary scan register.

4.3.2. Bypass Register

Bypass register provides a direct link between TDI and TDO, bypassing the device. A single bit makes up the bypass register. Data is sent from TDI to TDO with a one TCK cycle delay when the BYPASS instruction is the current instruction in the IR: in the Shift-DR mode. The Bypass Register shortens the shift route between the test data input (TDI) and test data output (TDO) of a boundary scan device with a fixed binary "0" output when it is activated by the Bypass Instruction.

4.3.3. Instruction Register and Instruction Decoder

The instruction register must be at least 2 bits long. During the Update-IR state, the instruction is given to an instruction decoder. The TAP controller state machine defines the conditions under which the instruction decoder interprets and executes the instructions. IEEE 1149.1 mandates a minimum of 4 instructions:

BYPASS

Utilize the bypass entry to capture, move, and update data. This enables the chip to carry on with its routine operations. This instruction must include just 1s according to IEEE 1149.1.

SAMPLE

Data entering and leaving the chip through its inputs and outputs can be sampled by capturing and shifting data through the boundary scan register. The update phase, however, does not force data onto inputs or outputs.

PRELOAD

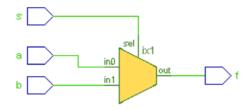
Shift information across the boundary scan register to create a value in the scan cells that may be used in the future. The previous value is not entered into the cell for this instruction during the capture phase, and neither is data loaded into the inputs or outputs during the update phase. This instruction was paired with SAMPLE in early versions of the standard.

EXTEST

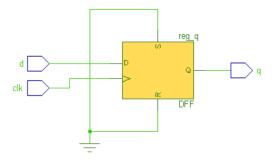
Before data is recorded, shifted, and updated through the boundary scan register, the semiconductor is put into EXTEST mode. This is employed to check the connection of several semiconductors. The chip does not attempt to drive outputs or take inputs when in EXTEST mode. Prior to EXTEST, it is typical to use PRELOAD to configure the boundary scan register.

5. RTL synthesis of various built combinational components

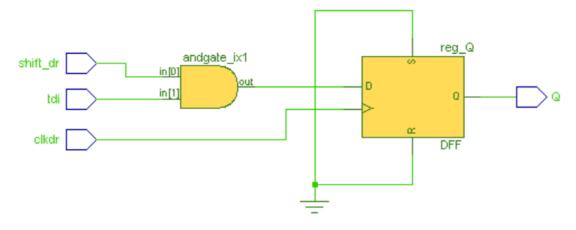
5.1.RTL synthesis of Multiplexer



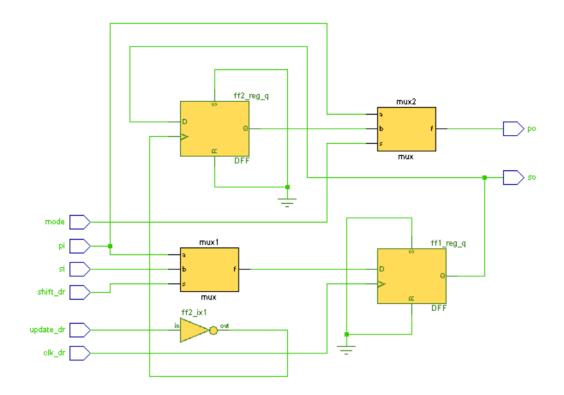
5.2.RTL synthesis of D-Flipflop for raising and falling edge



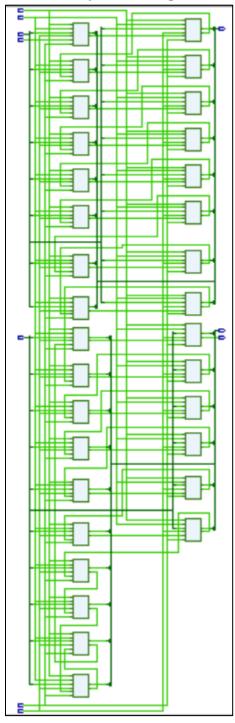
5.3.RTL synthesis of Bypass Register



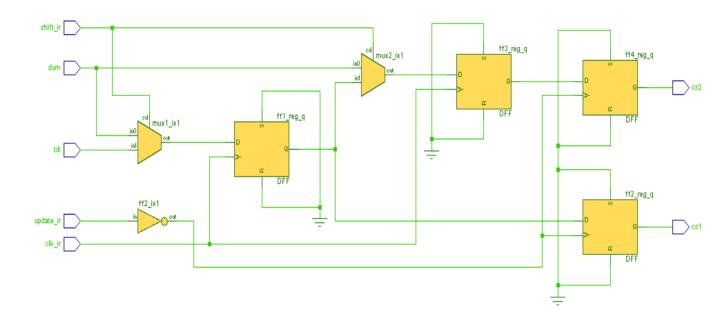
5.4.RTL synthesis of Boundary Scan Cell



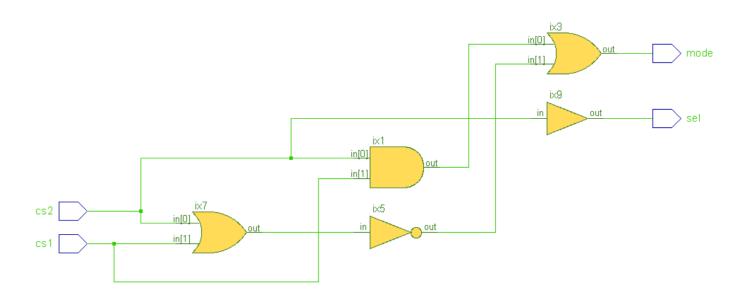
5.5.RTL synthesis of Boundary Scan Register



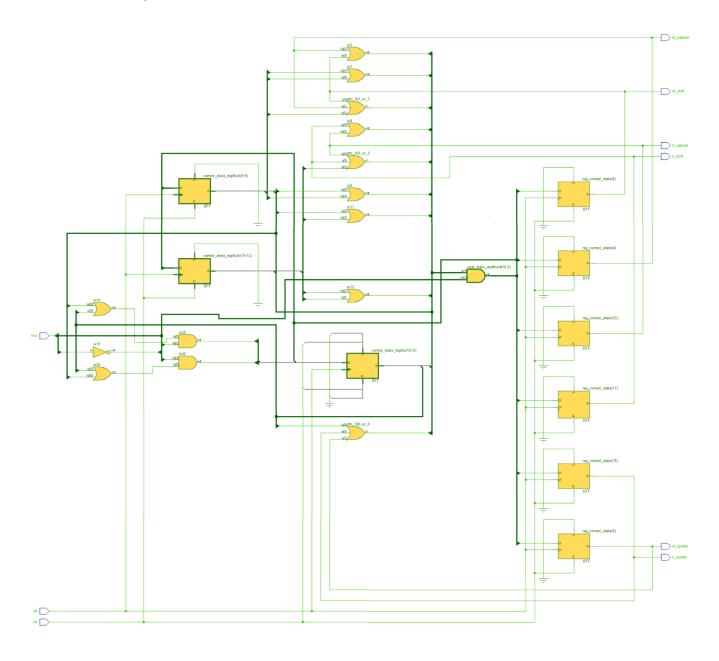
5.6.RTL synthesis of Instruction Register



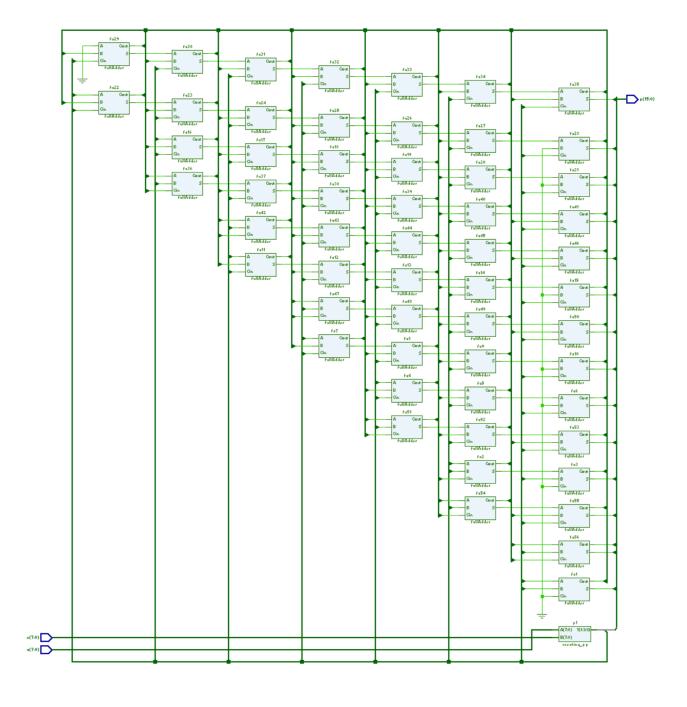
5.7.RTL synthesis of Instruction Decoder



5.8.RTL synthesis of TAP Controller



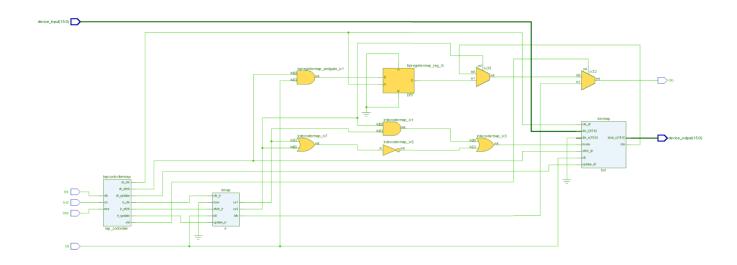
5.9.RTL synthesis of Multiplier



Area Report

Resource				Used	Αı	/ail	Utilization
IOS				32	14	40	22.86%
Global Buffers							0.00%
LUTS				158	28	316	5.61%
CLB Slices				79	14	408	5.61%
Dffs or Latche	S			0	32	236	0.00%
Block RAMs				0	12	2	0.00%
Block Multipli	ers			0	1.2	l.	0.00%
Block Multipli	er Dffs						0.00%
GT_CUSTOM				0	4		0.00%
******	*****	****	****	******	****	*****	***
Library: work	Cell:	mult	i_8b	it Vi	.ew: I	Behavi	oral
*****	*****	****	****	******	****	*****	***
Cell	Library	Re	fere	nces	Tota	al Are	a
IBUF	xcv2p	16	х				
LUT2 LUT3 LUT4 OBUF creating_pp	xcv2p	1	х	1	1	LUTS	
LUT3	xcv2p	14	Х	1	14	LUTS	
LUT4	xcv2p	106	Х	1	106	LUTS	
OBUF	xcv2p	16	Х				
creating_pp	work	1	х	64	64	gates	
				37	37	LUTS	
Number of por	ts:					32	
Number of net						206	
Number of ins	tances :					154	
Number of ref	erences t	o th	is v	iew :		0	
Total accumula	ted area	:					
Number of LUT	s:					158	
Number of gat	es:					185	
Number of acc						190	

5.10. RTL synthesis of JTAG



Area Report

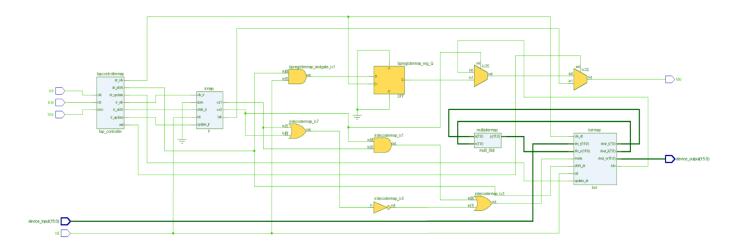
******	*******	****	****	****	*****	*****	******		
Device Utilization for 2VP2fg256									

Resource							lization		
Resource					MAGIL		TIZACION		
IOS			37		140	26	.43%		
Global Buffers			1		140 16		.25%		
LUTS			55				.95%		
CLB Slices			35		2816 1408	2	499		
Dffs or Latches			69		3236	2	.13%		
Block RAMS			0			0	.00%		
Block Multipliers			0	0 12		0	.00%		
Block Multiplier I			0		432	0	.00%		
GT CUSTOM			0		4		.00%		
	-								
******	*******	****	****	****	****				
Library: work	cell: top	V	iew:	behav	rioral				
*******	*******	****	****	****	****				
Cell	Library	Ref	erenc	es	Tota	al Area	ı		
BUFGP	xcv2p	1	x				or Latches or Latches		
FDR	xcv2p	3	x	1	3	Dffs o	r Latches		
FD 1	xcv2p	2	x	1	2	Dffs o	r Latches		
IBUF	xcv2p	19	x						
LUT1	xcv2p	2	x	1	2	LUTS			
LUT3	xcv2p	1	x	1	1	LUTS			
LUT4	xcv2p			1	1	LUTS			
OBUF	xcv2p xcv2p	17	**						
bsr	work		x	49	49	gates			
				48	48	gates LUTs			
1				48	48		or Latches		
tap controller	work	1	×	48 21	21	gates	or Latches		
				21	21	LUTS			
				16	16	Dffs o	or Latches		
1				_					
Number of ports	:				3	7			
Number of nets: 90									
Number of instances : 48									
Number of referen		nis v	riew :		(
Total accumulated	area :								
Number of Dffs or Latches : 69									
Number of LUTs: 55									
Number of gates	:				74	5			
Number of accumu		ance	es :		179	9			

Timing Report/Delay

```
NAME
                                             GATE
                                                        DELAY
                                                                  ARRIVAL DIR FANOUT
tapcontrollermap/reg_current_state(5)/C FDC tapcontrollermap/reg_current_state(5)/Q FDC
                                                                  0.000
                                                       0.370
                                                                  0.370
tapcontrollermap/dr_shift
                                                       0.640
                                                                 1.010
tapcontrollermap/ix22095z1329/I1
                                          LUT2
                                                                          up
                                                       0.264
tapcontrollermap/ix22095z1329/0
                                          LUT2
                                                                  1.274
                                                                          up
tapcontrollermap/dr_clk
                                          (net)
                                                       0.640
                                                                               34
tapcontrollermap/ix22095z1322/I1
                                                                 1.914
                                          LUT2
                                                                          up
tapcontrollermap/ix22095z1322/0
                                          LUT2
                                                       0.264
                                                                  2.178
                                                                          up
tapcontrollermap/next_state(6)
                                                       0.280
                                                                                1
                                          (net)
tapcontrollermap/reg_current_state(6)/D FDC
                                                                  2.458
                 Initial edge separation:
                                               10.000
                 Source clock delay:
                                               1.359
                 Dest clock delay:
                                                1.359
                 Edge separation:
                                               10.000
                 Setup constraint:
                                                0.174
                 Data required time:
                                                9.826
                                                         ( 36.53% cell delay, 63.47% net delay )
                 Data arrival time:
                                               2.458
                                                7.368
                 slack:
End CTE Analysis ..... CPU Time Used: 0 sec.
```

5.11. RTL synthesis of JTAG and Multiplier



Area Report

Resource						Utilization	
IOS							
Global Buffers				1	16	26.43% 6.25%	
LUTS				258	2816	9.16%	
CLB Slices				258 129	1408	9.16%	
Dffs or Latches				85	3236	2.63%	
Block RAMs				0	12	0.00%	
Block Multipliers				0	12	0.00%	
Block Multiplier	Dffs			0	432	0.00%	
GT_CUSTOM				85 0 0 0	4	0.00%	

Library: work	-						
						_	
Cell	Library	Re	fer	ences	Tota	al Area	
BUFGP	xcv2p						
FDR	xcv2p	3	Х	1	3	Dffs or Latches Dffs or Latches	
FD_1	xcv2p	2	Х	1	2	Dffs or Latches	
IBUF	xcv2p	19	Х				
LUT1	xcv2p	2	Х	1	2	LUTS	
LUT2	xcv2p	1	х	1	1	LUTS	
LUT3	xcv2p	1	Х	1	1	LUTS	
LUT4	xcv2p xcv2p xcv2p xcv2p xcv2p	1	Х	1	1	LUTS	
OBUF	xcv2p	17	Х				
OBUF bsr multi_8bit	work	1	Х	65	65	gates	
				56	56	LUTS	
				64	64	Dffs or Latches	
muiti_8bit	work	1	Х	178	178	LUTS	
t				185	185	gates	
tap_controller	WOLK	1	Х	21	21	gates	
				16		Dffs or Latches	
Number of ports					37		
Number of nets :					130		
Number of instan					50		
Number of refere	nces to th	nis '	vie	w :	()	
Total accumulated Number of Dffs o					85		
Number of time .			258				
Number of LUTs :							
Number of LUTs : Number of gates Number of accumu	:				27	7	

Timing Report/Delay

NAME	GATE	3	DELAY	ARRIVAL	DIR	FANOUT
tapcontrollermap/reg current state(5)/C	FDC			0.000	up	
tapcontrollermap/reg_current_state(5)/Q	FDC		0.370	0.370	up	
tapcontrollermap/dr_shift	(net)		0.640			35
tapcontrollermap/ix22095z1329/I1	LUT2			1.010	up	
	LUT2		0.264	1.274	up	
	(net)		0.640			34
	LUT2			1.914	up	
			0.264	2.178	up	
	(net)		0.280			1
tapcontrollermap/reg_current_state(6)/D	FDC			2.458	up	
Initial edge separation:						
Source clock delay:						
Dest clock delay:	+ 1	359				
Edge separation:	10	000				
Setup constraint:).174				
Scrap constraint.						
Data required time:	g	.826				
Data arrival time:	- 2		(36.53	8% cell d	elav.	63.47% net delay)
			, 50101		,	,
Slack:	7	7.368				
End CTE Analysis CPU Time Used: 0	sec.					

6. Overhead and benefits of the added JTAG

The memory takes up a large portion of most design. Circuitry that is added to test the memory creates area overhead. After each DFT pass, area grows along with the test logic and is extremely challenging to control. To facilitate design testing on ATE (Automatic Test Equipment), DFT logic circuitry is introduced throughout the design process (post-manufacturing). The addition of these logic circuits doesn't have an impact on the operation of any design. The DFT logic circuit causes some design space to be added after each iteration of the DFT. The region for a chip is over headed by this DFT circuitry. A DFT engineer must be highly adept at minimizing and controlling the area by maintaining its track. The overhead problem can be reduced by various techniques, such as test microprogram, scan protection scheme, etc., Figure 2 shows the comparison result of area report of multiplier with and without JTAG.

The following are the benefits of JTAG,

- JTAG offers the ability to evaluate PC-board interconnects without the need of physical test probes or test equipment.
- Doesn't need the board to be in a bootable condition to perform fault diagnostics.
- Device level diagnostics and automated test development for DSP initialization, memory, and flash are made possible by JTAG.

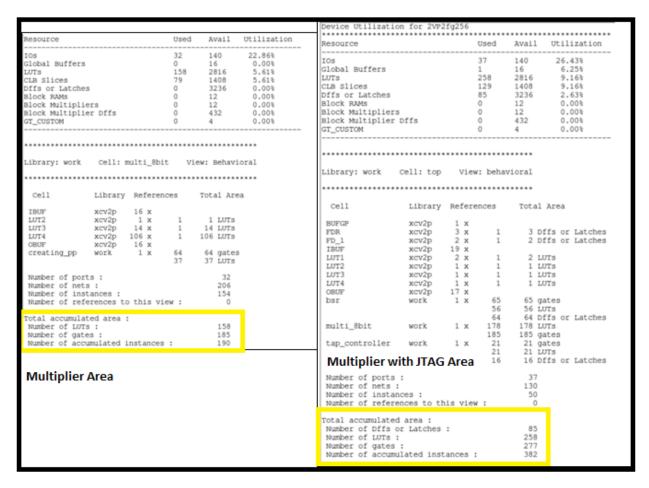
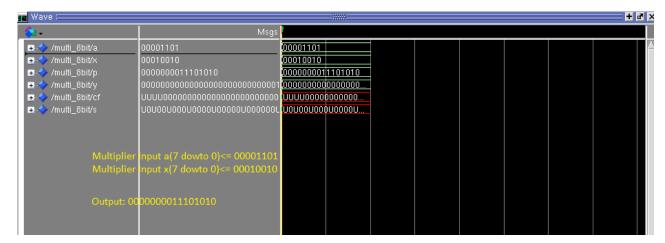
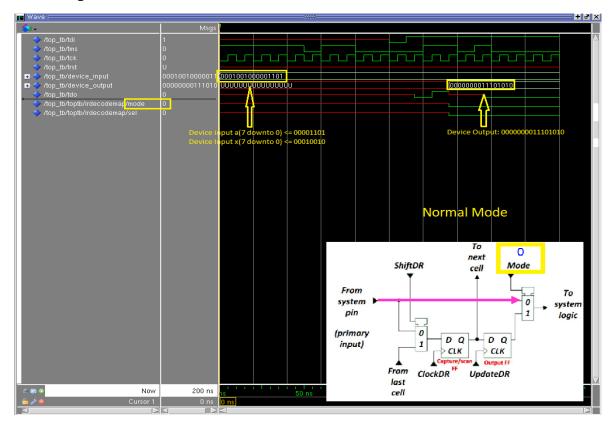


Figure 2: Accumulated Area of multiplier without JTAG and with JTAG

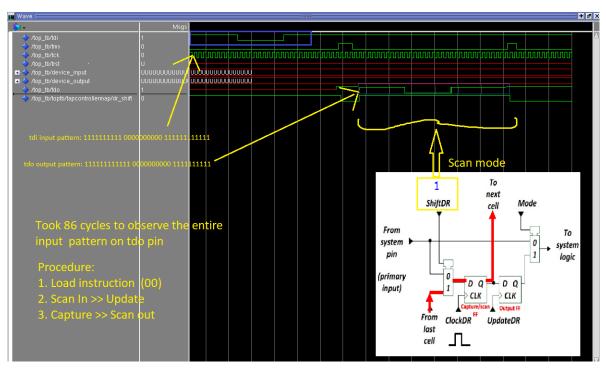
7. Testing 8-bit multiplier using the JTAG Simulated output of giver 8-bit multiplier design:



7.1.Output of multiplier with JTAG JTAG operation - Normal:



JTAG operation – Scan-in, update, capture and scan-out through 32 cells:



8. Conclusion:

Scan-based DFT improves the system's testability by making its internal nodes easier to observe and control. A direct link to the debug logic is made possible by the JTAG path. The advantages of using JTAG methodology extend from chip design through all stages of system development and the device product lifecycle since the JTAG scan channel offers direct core access. Reusing test patterns at different levels of hierarchy, from chip to board to board to system level, is possible with the JTAG approach. As an illustration, the board-level test can employ all or a portion of the chip level test vectors as its base. JTAG technique is also cost-effective because it expedites the testing procedure, ensures greater manufacturing yields, and takes numerous other money-saving steps.

9. Reference:

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