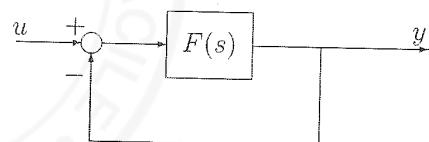


Domanda Scritta di Controlli Automatici (9CFU) - 19/11/2012

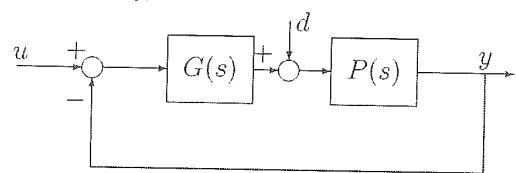
Esercizio 1 È dato il sistema di controllo:



in cui: $F(s) = \frac{K}{s(s-p)}$. Utilizzando il criterio di Nyquist, studiare la stabilità del sistema a ciclo chiuso, per $K \in \mathbb{R}$, $K \neq 0$, $p \in \mathbb{R}$, $p \neq 0$.

Esercizio 2

È dato il sistema di controllo:



in cui $P(s) = \frac{(s+3)}{s(s+4)}$; $d(t) = \delta_{-1}(t)$.

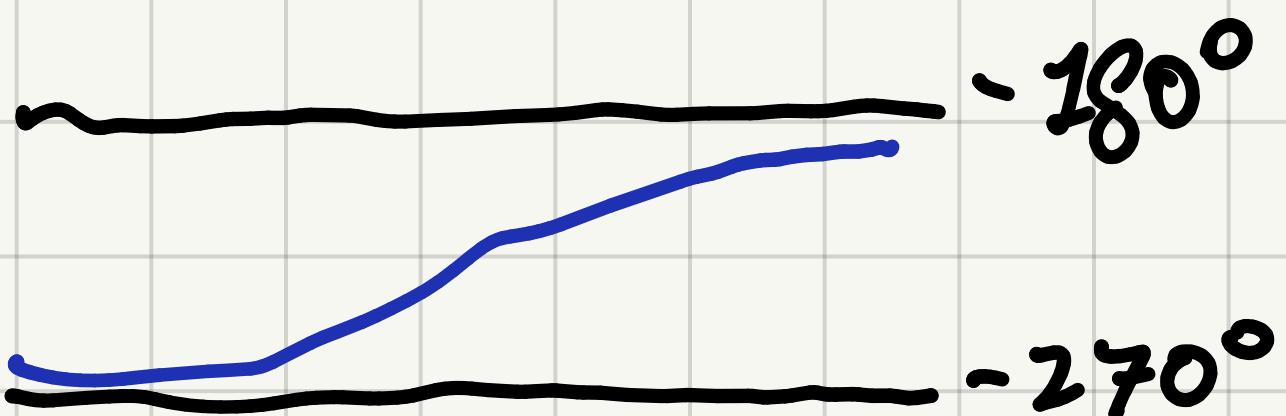
Utilizzando la sintesi con il luogo delle radici, progettare $G(s)$ in modo che:

- il sistema sia astatico rispetto al disturbo $d(t)$.
- tutti i poli della funzione di trasferimento in catena chiusa abbiano parte reale minore di -3 .

Calcolare infine la risposta a regime permanente all'ingresso $u(t) = (3t + 2)\delta_{-1}(t)$.

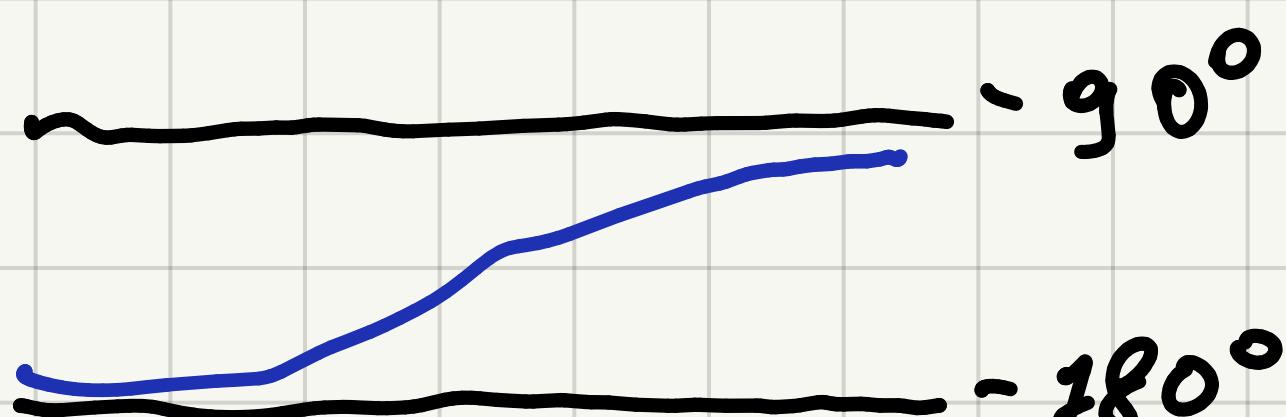
$$F(s) = -\frac{K}{P} \cdot \frac{1}{s(1-\frac{s}{P})}$$

CASO 1: $K > 0, P > 0$
 $M(0^+) = \infty, \varphi(0^+) = -270^\circ$



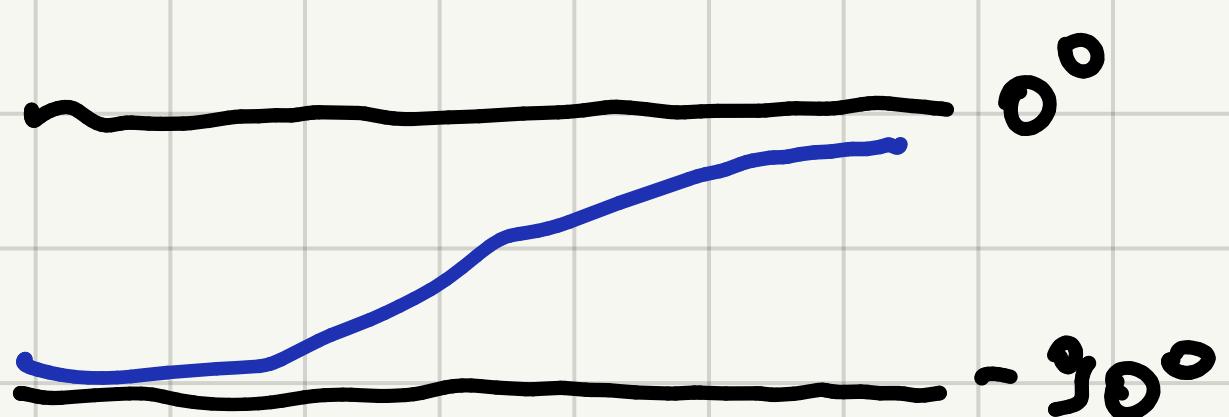
$\tilde{N} \neq -P_f \Rightarrow$ SISTEMA INSTABILE

CASO 1: $K > 0, P < 0$
 $M(0^+) = \infty, \varphi(0^+) = -180^\circ$



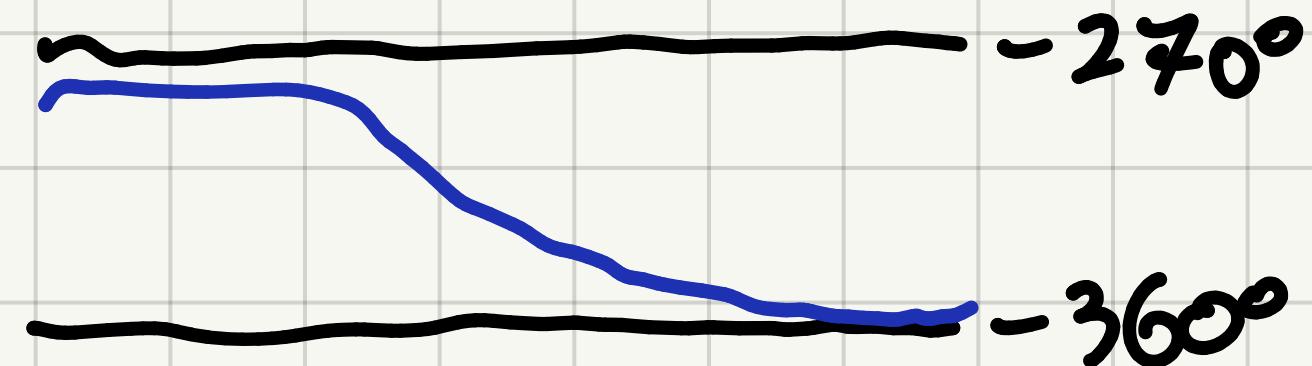
$\tilde{N} = -P_f \Rightarrow$ SISTEMA STABILE

CASO 1: $K < 0, P > 0$
 $M(0^+) = \infty, \varphi(0^+) = -90^\circ$



$\tilde{N} \neq -P_f \Rightarrow$ SISTEMA INSTABILE

CASO 1: $K < 0, P < 0$
 $M(0^+) = \infty, \varphi(0^+) = -270^\circ$



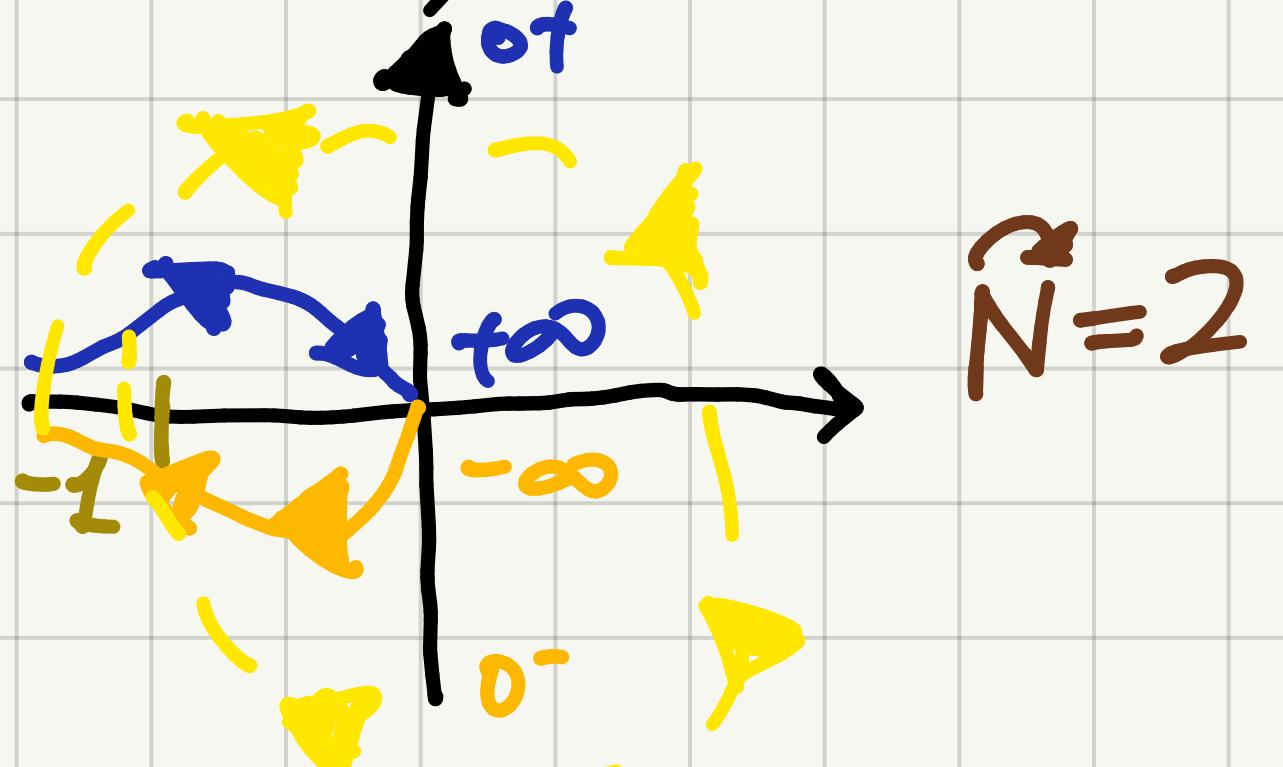
$\tilde{N} \neq -P_f \Rightarrow$ SISTEMA INSTABILE

①

$$F(iw) = -\frac{K}{P} \cdot \frac{1}{iw(1-\frac{iw}{P})}$$

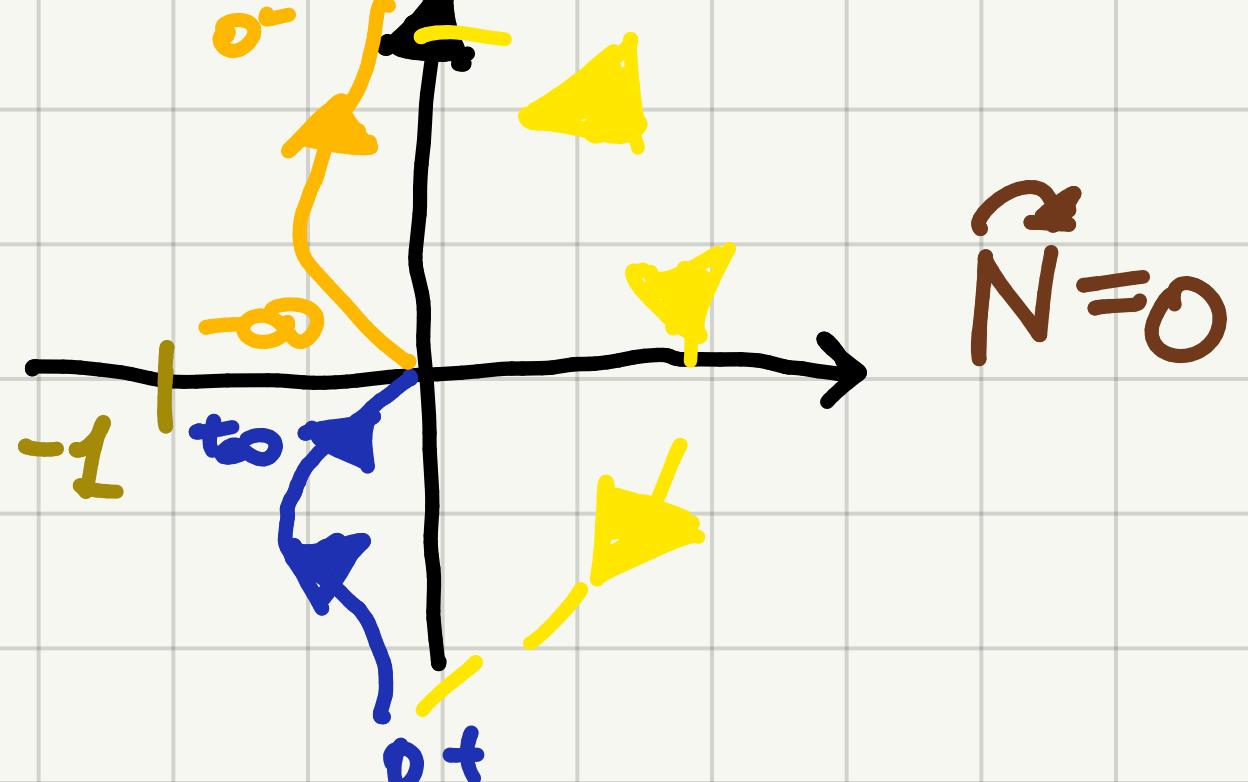
$P_f = 1$

$M(+\infty) = 0, \varphi(+\infty) = -180^\circ$



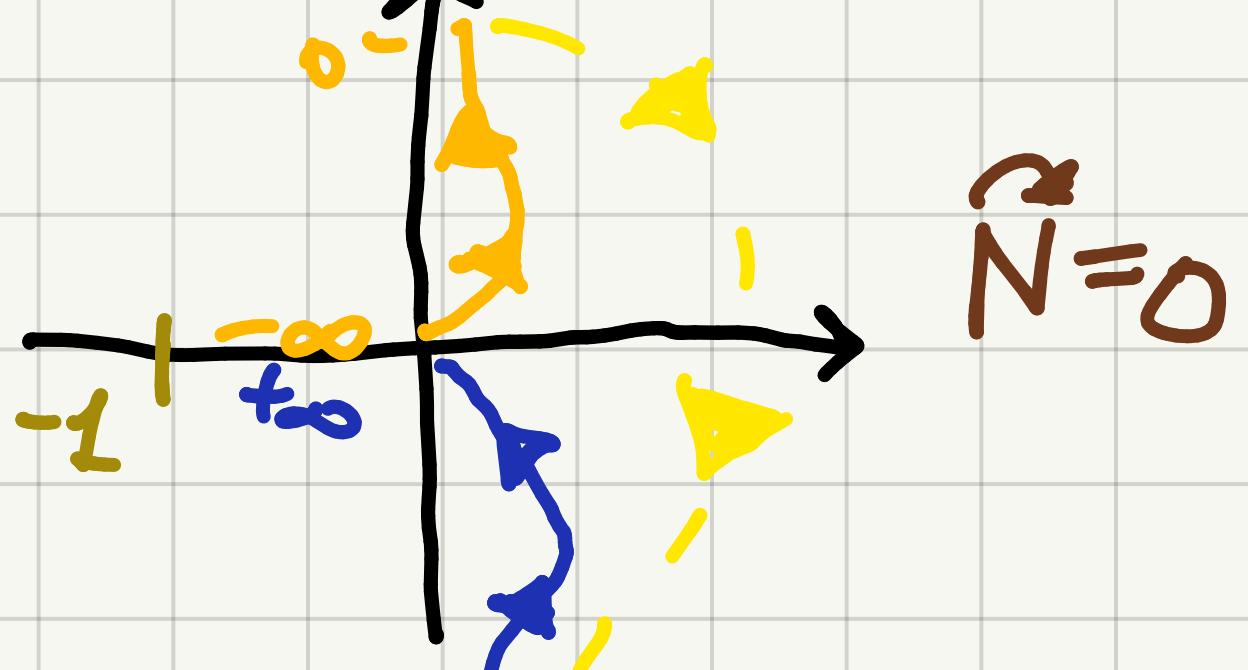
$P_f = 0$

$M(+\infty) = 0, \varphi(+\infty) = -90^\circ$



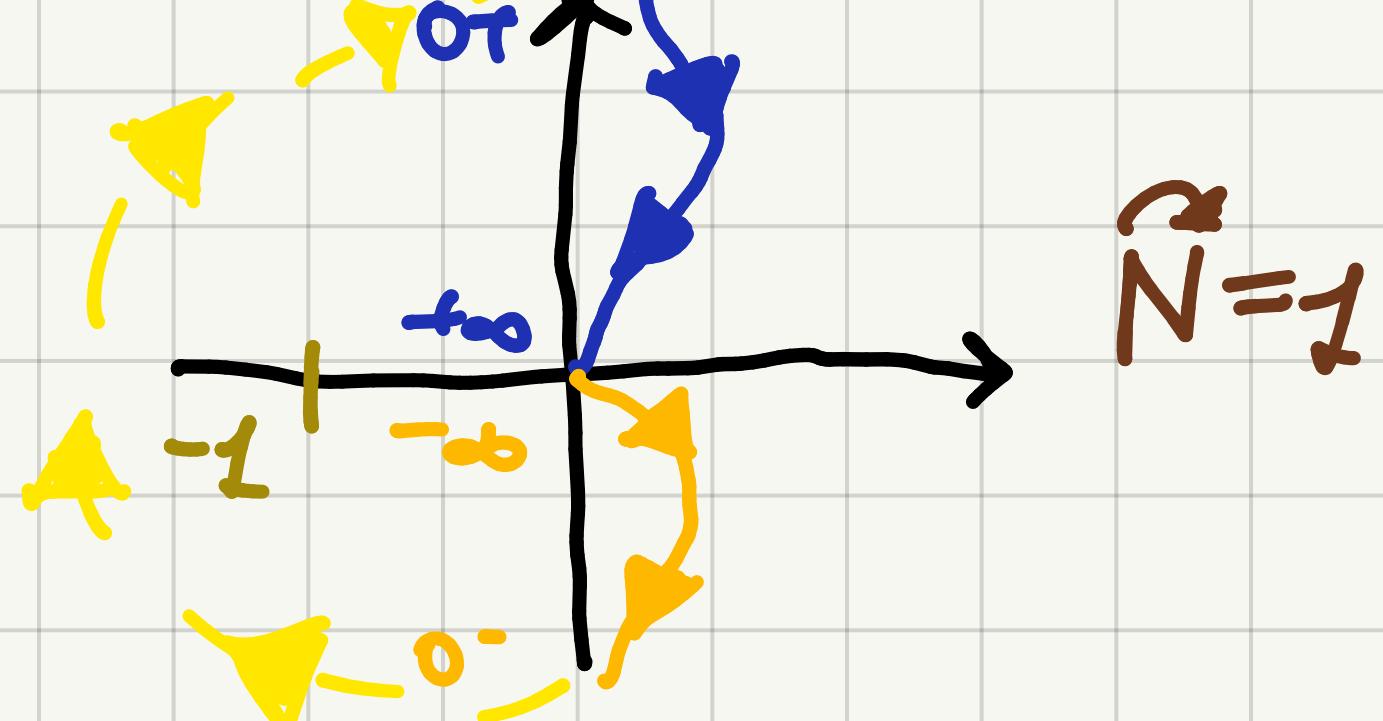
$P_f = 1$

$M(+\infty) = 0, \varphi(+\infty) = 0$



$P_f = 0$

$M(+\infty) = 0, \varphi(+\infty) = -360^\circ$



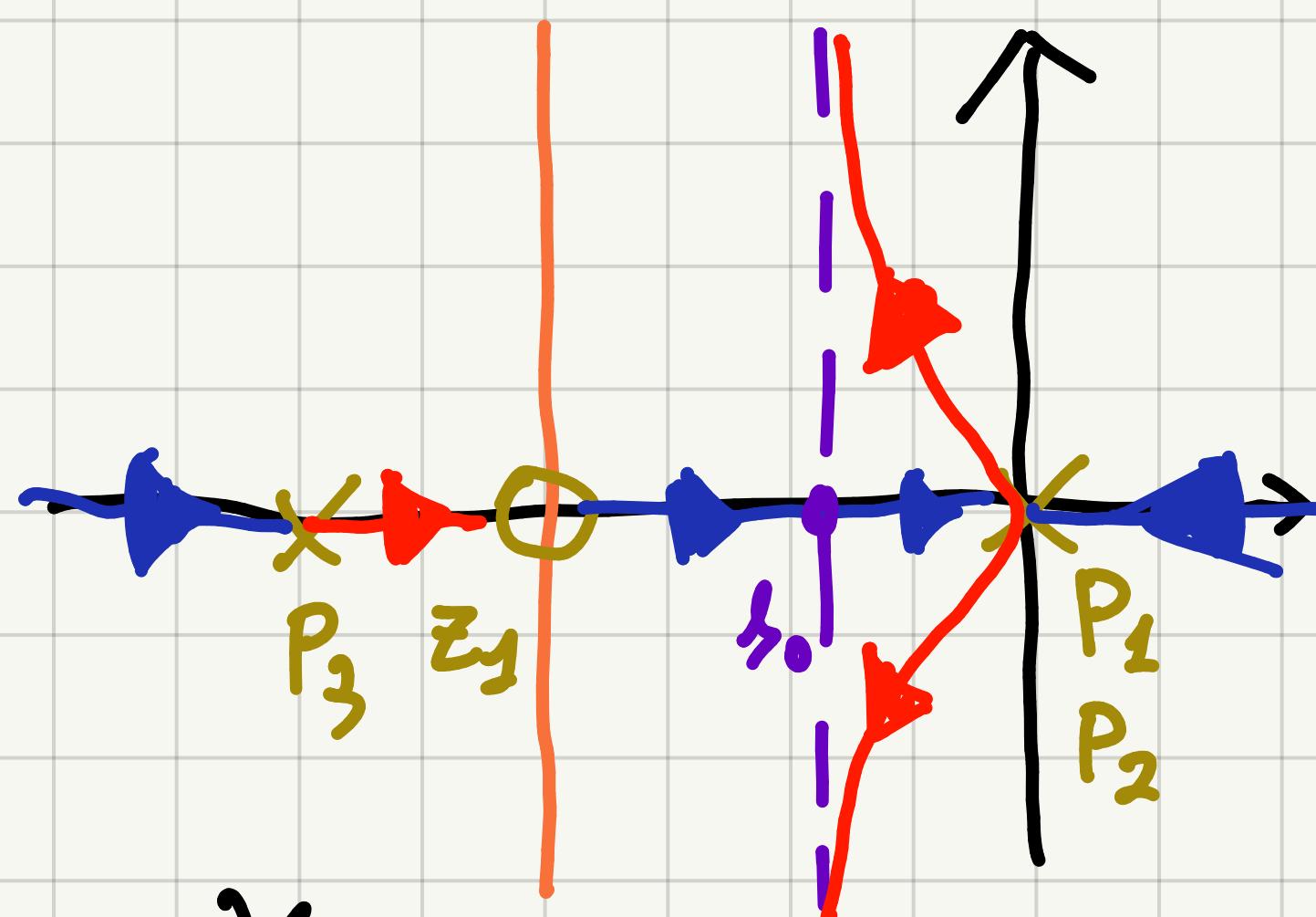
(2)

$$G(s) = \frac{K_0}{s} \quad F(s) = G(s) \cdot P(s) = K_0 \cdot \frac{(s+3)}{s^2(s+4)}$$

$$n=3, m=1 \Rightarrow n-m=2$$

$$P_1 = P_2 = 0, P_3 = -4; Z_1 = -3$$

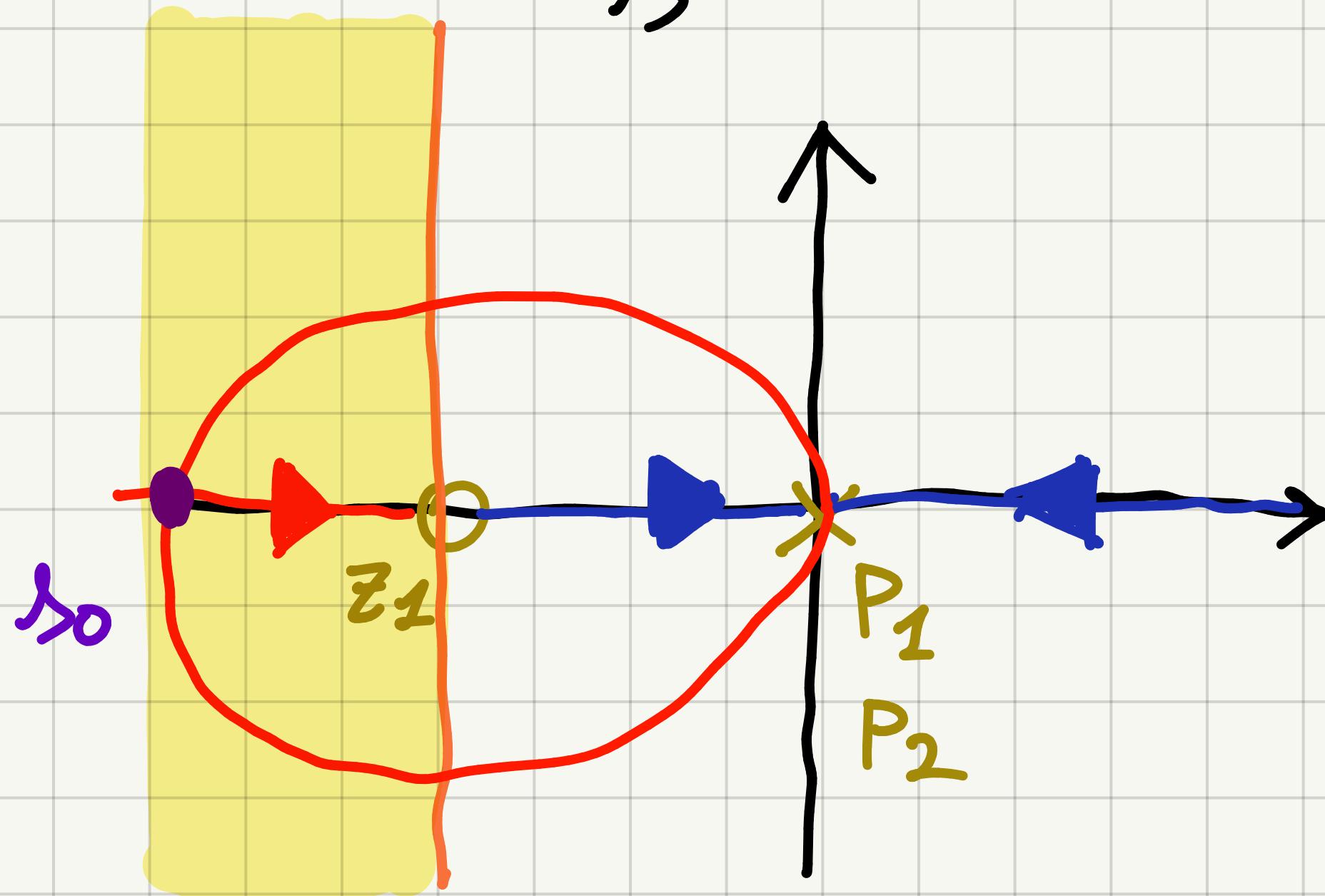
$$D_0 = \frac{\sum P - \sum Z}{n-m} = -\frac{1}{2}$$



SPECIFICA
NON SODDISFATTA

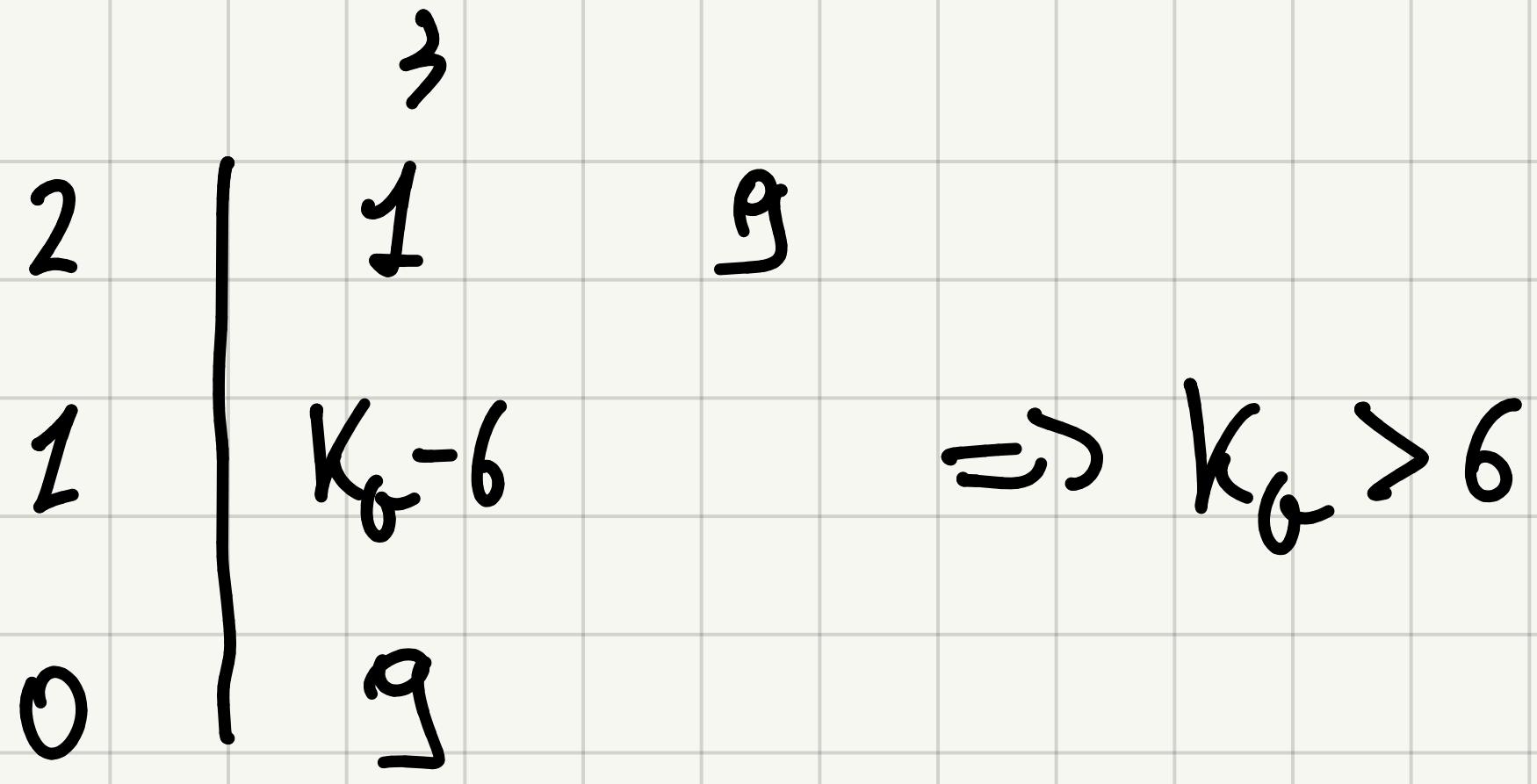
$$G(s) = \frac{K_0}{s} \cdot (s - Z_1) \Rightarrow n-m=1 \quad Z_1 = -4$$

$$\Rightarrow G(s) = \frac{K_0 \cdot (s+4)}{s} \Rightarrow F(s) = K_0 \cdot \frac{(s+3)}{s^2}$$



$$f(s, k) = s^2 + K_0(s+3) \rightarrow (\bar{s}-3)^2 + K_0 \bar{s} = 0$$

$$\bar{s}^2 + (K_0 - 6)\bar{s} + 9 = 0$$



$$\Rightarrow u(s) = 8 \cdot \frac{(s+4)}{s} \quad F(s) = 8 \cdot \frac{(s+3)}{s^2}$$

$$U(t) = (3t+2)\delta_{-1}(t) = 3(t)\delta_{-1}(t) + (2)\delta_{-1}(t) \\ = 3U_1(t) + 2U_2(t)$$

• $U_1(t)$

$$\tilde{e}_{U_1}(t) = K_d U_1(t) - \tilde{y}_{U_1}(t)$$

$$\tilde{e}_{U_1}(t) = \frac{1}{K_d \cdot K_p} = \frac{1}{K_F} = \frac{1}{24}$$

$$\tilde{y}_{U_1}(t) = K_d U_1(t) - \tilde{e}_{U_1}(t) = \left(t - \frac{1}{24}\right) \delta_{-1}(t)$$

• $U_2(t)$

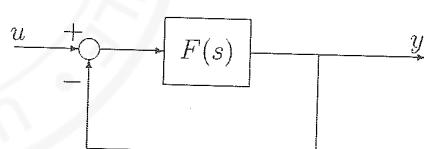
$$(R4DO DI \quad U_2(t) \perp \text{RPO DI } F(s)) \Rightarrow \tilde{y}_{U_2}(t) = \delta_{-1}(t)$$

$$\Rightarrow \tilde{y}(t) = 3\left(t - \frac{1}{24}\right) \delta_{-1}(t) + 2\delta_{-1}(t)$$

Domanda Scritta di Controlli Automatici (9CFU) - 17/12/2012

Esercizio 1

È dato il sistema in controreazione:

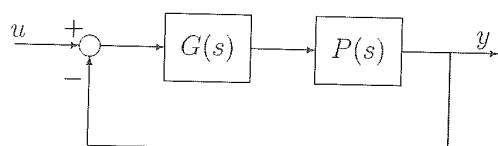


$$\text{in cui } F(s) = \frac{K(s+1)^2}{(s-1)(s+3)(s+5)(s+8)}, K \in \mathbb{R}.$$

- Tracciare il luogo positivo delle radici;
- tracciare il luogo negativo delle radici;
- determinare per quali valori di K il sistema a ciclo chiuso è asintoticamente stabile;
- se $K = 10$, esiste la risposta a regime permanente a ciclo chiuso per un ingresso a gradino? Motivare la risposta.

Esercizio 2

È dato il sistema di controllo:



in cui:

$$P(s) = \frac{50(s+1)}{(s+4)^2}.$$

Progettare $G(s)$ con la sintesi per tentativi in ω in modo che:

- $|\tilde{\epsilon}_1(t)| \leq 0.04$, essendo $\tilde{\epsilon}_1(t)$ l'errore a regime permanente per un ingresso di riferimento a rampa unitaria;
- $M_r \leq 2 \text{ dB}$;
- $B_3 \simeq 1.5 \text{ Hz}$.

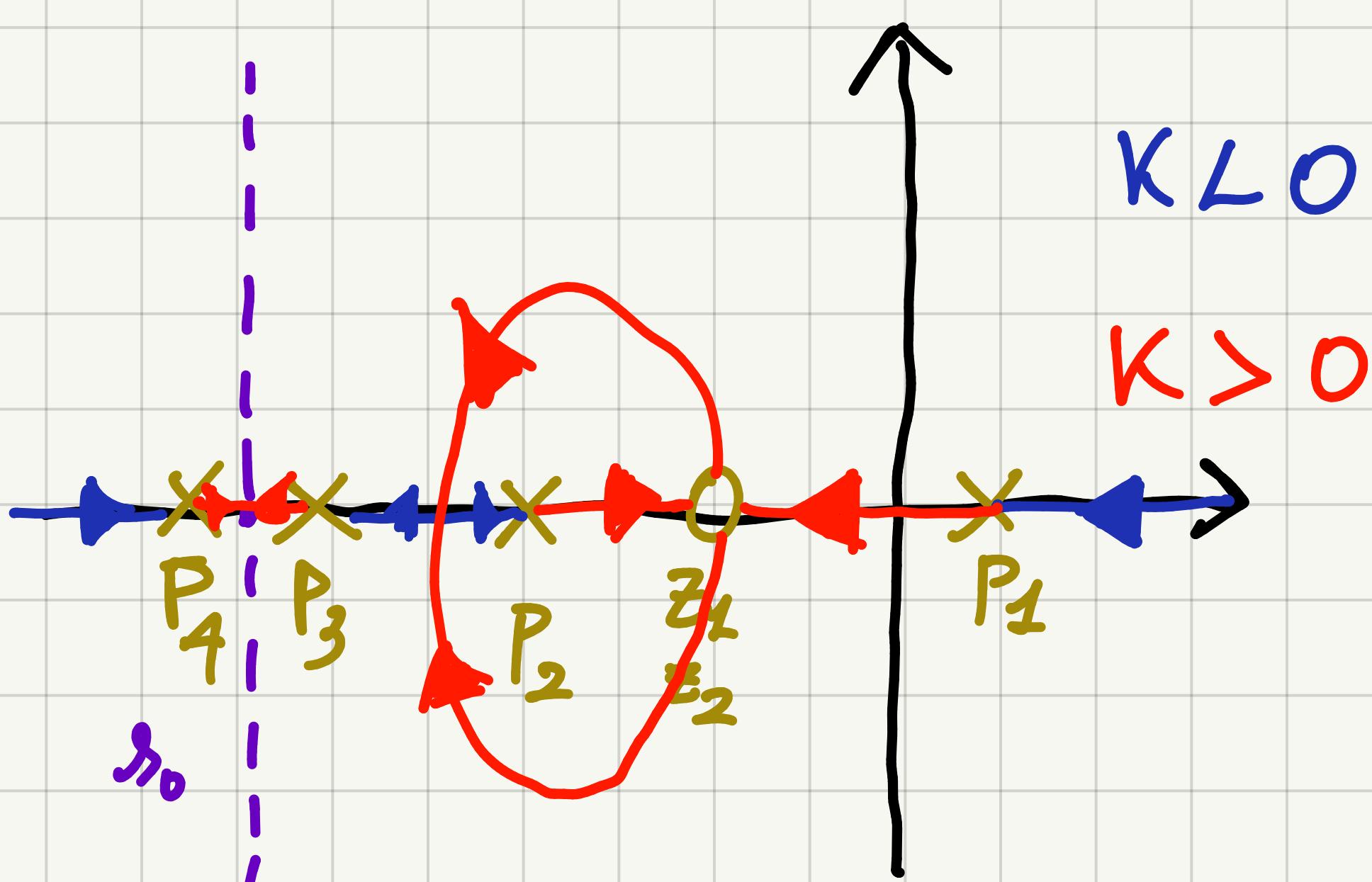
Calcolare infine la risposta a regime permanente all'ingresso: $u(t) = (3t+5) \cdot \delta_{-1}(t)$.

$$F(s) = K \cdot \frac{(s+1)^2}{(s-1)(s+3)(s+5)(s+8)} \quad K \in \mathbb{R}$$

$$n=4, m=2 \Rightarrow n-m=2$$

$$\begin{array}{ll} P_1 = 1 & P_2 = -3 \\ P_3 = -5 & P_4 = -8 \\ Z_1 = Z_2 = -1 & \end{array} \quad \sum P - \sum Z = -13$$

$$J_0 = \frac{-13}{n-m} = \frac{-13}{2} = -6,5$$



$$\rho(s, K) = (s-1)(s+3)(s+5)(s+8) + K(s+1)^2 \Big|_{s=0} = 0$$

$$-120 + K = 0 \quad K = 120$$

SISTEMA STABILE $\forall K > 120$

$\Rightarrow \exists$ RISPOSTA A REGIME PERMANENTE PER $K = 10$

$$G(s) = -\frac{K}{s} \quad | \tilde{e}_1 | = \left| \frac{\frac{K}{s}^2}{K_G \cdot K_P} \right| \leq 0,04 \quad K_P = \frac{50}{16} \Rightarrow K_G \geq 8$$

②

$$M_r \leq 2 \text{ dB} \Rightarrow M_\varphi \geq 47^\circ$$

$$B_3 \approx \frac{1}{2} \text{ Hz} \Rightarrow W_C = 3 \div 5 B_3 = 4 B_3 \approx 6 \frac{\text{rad}}{\text{s}}$$

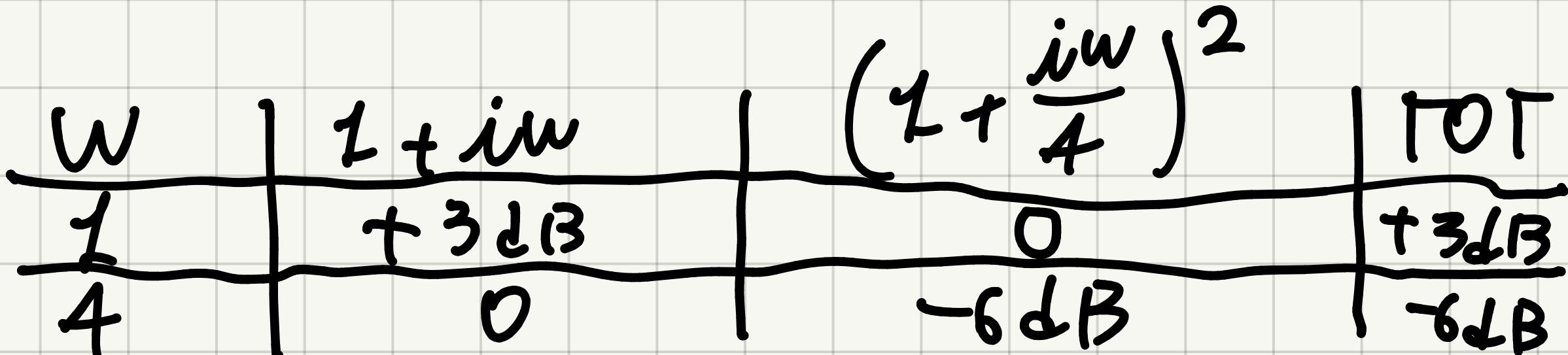
$$F(s) = G(s) \cdot P(s) = 25 \cdot \frac{(1+s)}{s(1+\frac{s}{4})^2}$$

$$F(iw) = 25 \cdot \frac{(1+iw)}{iw(1+\frac{iw}{4})^2} \quad 25 \rightarrow 20 \log_{10}(25) \approx 28 \text{ dB}$$

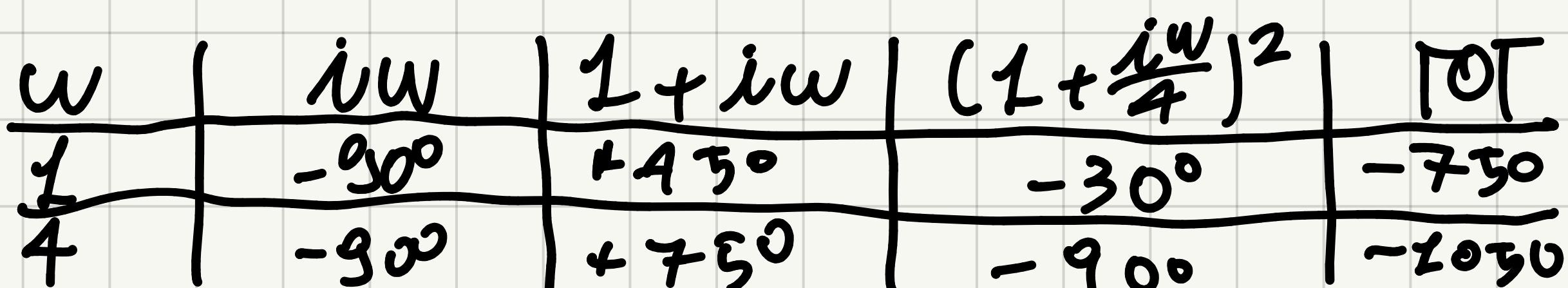
PUNTI DI ROTURA

• $w=0$	●	-20dB	-90°	-20dB	-90°
• $w=1$	●	+20dB	+90°	0	0
• $w=4$	●	-40dB	-180°	-40dB	-180°

CORREZIONE MODULO

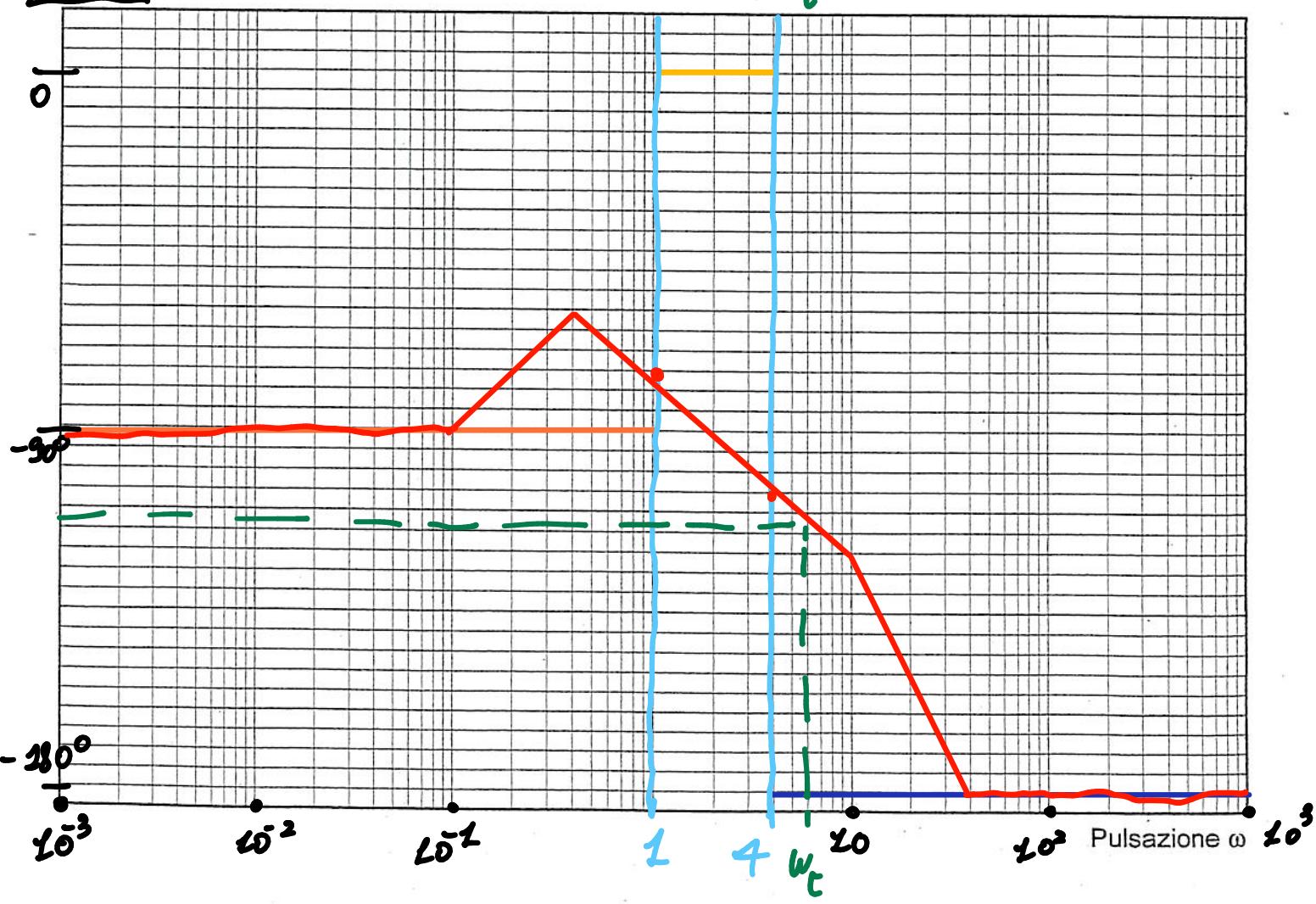
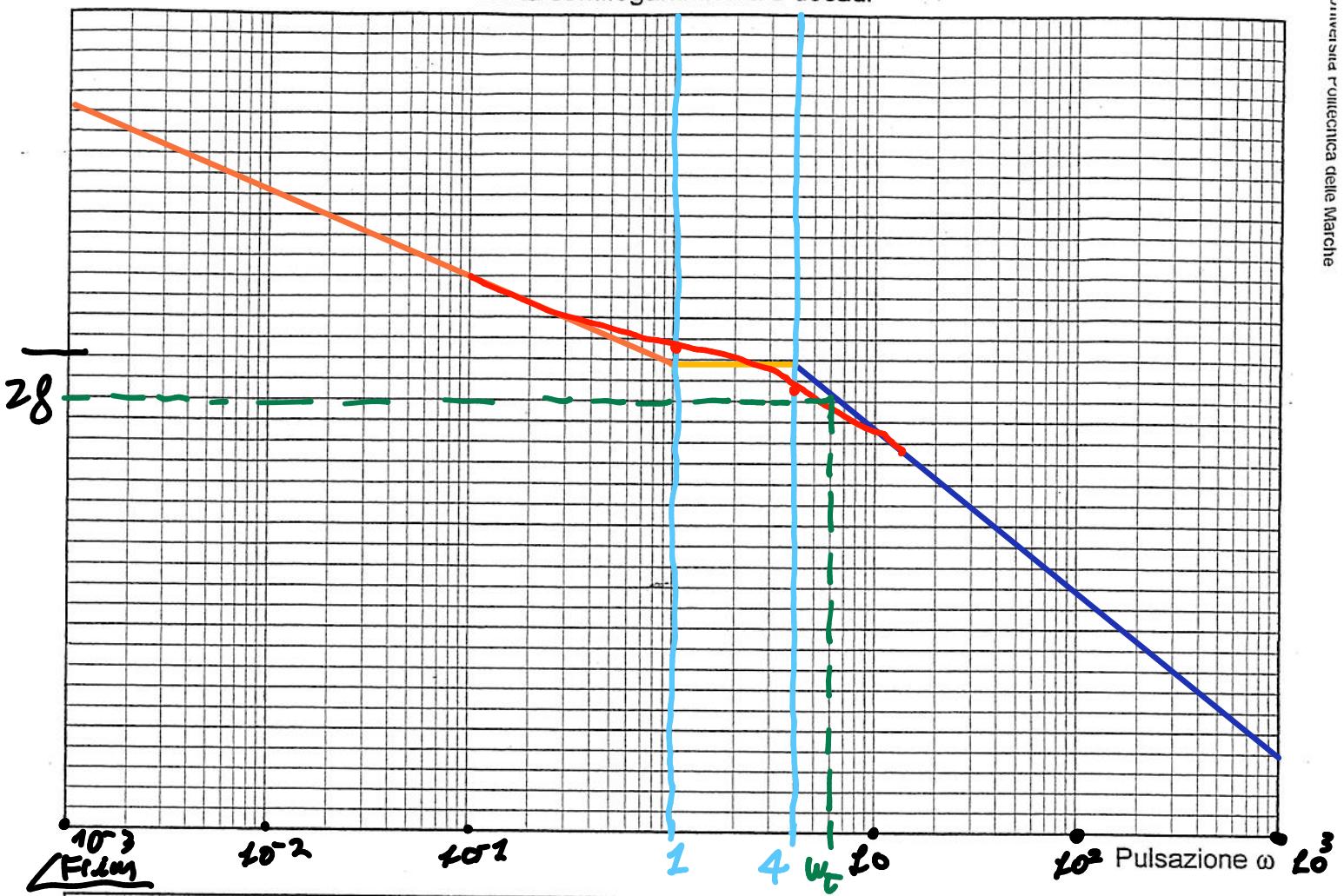


CORREZIONE FASE



$|F(\omega)|$

Carta semilogaritmica a 6 decadri



$$|F(i\omega_c)| = 18 \text{ dB}$$

$$\angle F(i\omega_c) = -115^\circ \Rightarrow M_\varphi = 65^\circ$$

OBIETTIVO:

$$|F(i\omega_c)| = 0 \quad \text{X}$$

$$M_\varphi \geq 47^\circ \quad \checkmark \Rightarrow \text{FUNZIONE ATTENUATRICE}$$

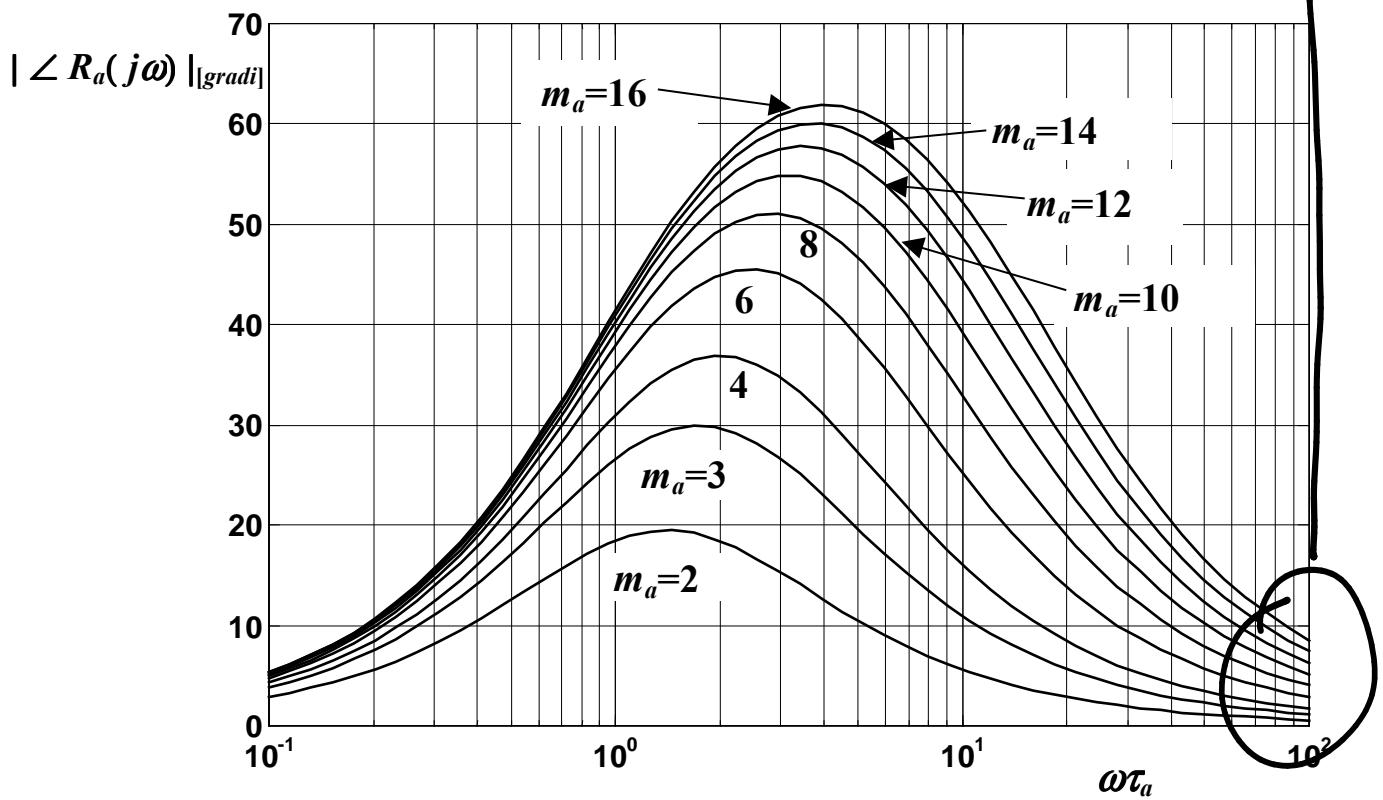
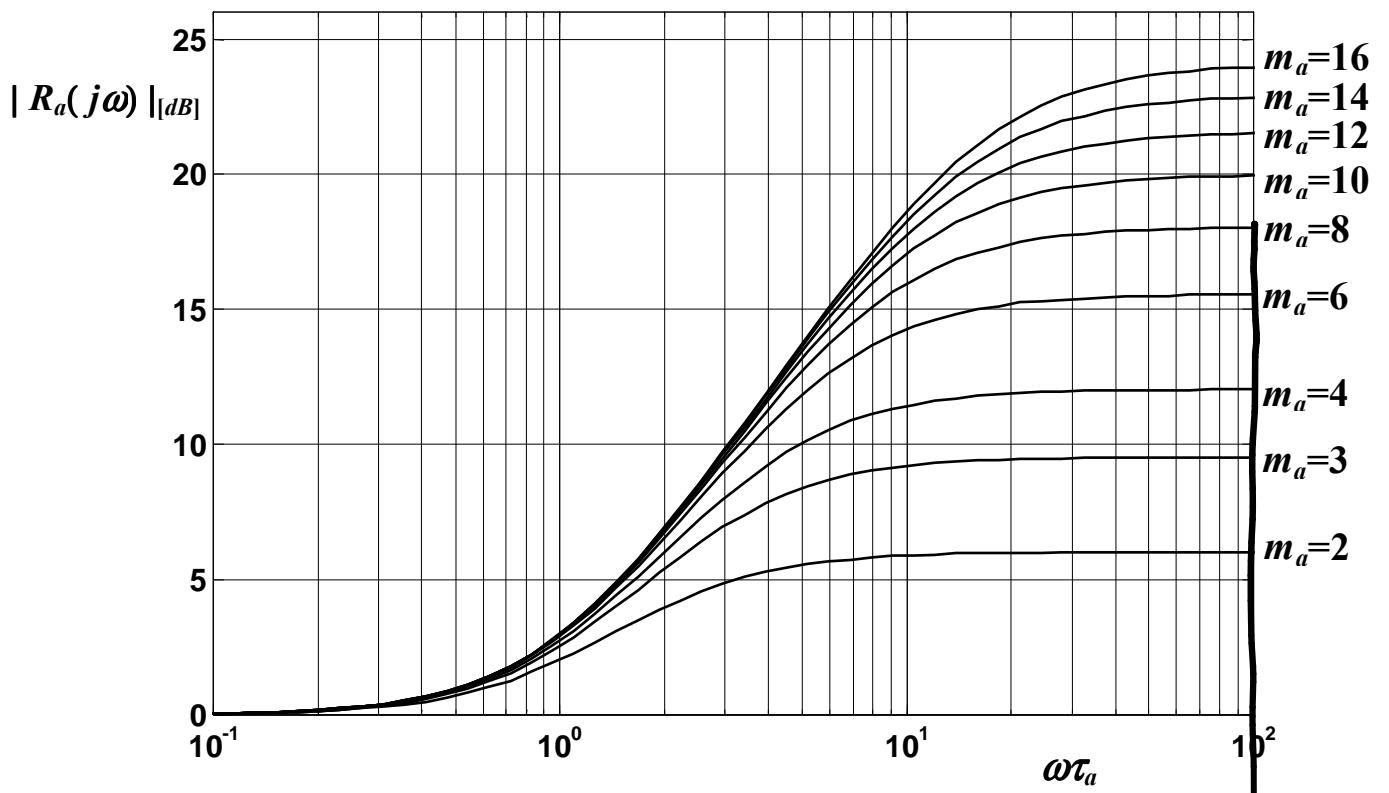
$$W_i \gamma_i = 100, M_i = 8 \quad W_i = 0,06$$

$$\Rightarrow K_{i(\omega)} = \frac{1 + \frac{i\omega}{M_i W_i}}{1 + \frac{i\omega}{W_i}} = \frac{1 + \frac{i\omega}{0,48}}{1 + \frac{i\omega}{0,06}}$$

$$\Rightarrow G(s) = K \cdot \frac{1 + \frac{s}{0,48}}{s(1 + \frac{s}{0,06})}$$

$$F(i\omega) = 25 \cdot \frac{(1+i\omega)}{i\omega(1+\frac{i\omega}{4})^2} \frac{1 + \frac{i\omega}{0,48}}{1 + \frac{i\omega}{0,06}}$$

• $\omega=0$	●	-20dB	-90°	-20dB	-90°
• $\omega=0,06$	●	-20dB	-90°	-40dB	-180°
• $\omega=0,48$	●	+20dB	+90°	-20dB	-90°
• $\omega=1$	●	+20dB	+90°	0	0
• $\omega=4$	●	-40dB	-180°	-40dB	-180°



CORREZIONE MODULO

ω	$1 + \frac{i\omega}{0,06}$	$1 + \frac{i\omega}{0,48}$	$1 + i\omega$	$(1 + \frac{i\omega}{4})^2$	TOT
0,06	-3dB	0	0	0	-3dB
0,48	0	+3dB	+1dB	0	+4dB
1	0	+2dB	+3dB	0	+4dB
4	0	0	0	-6dB	-6dB

CORREZIONE FASE

ω	$i\omega$	$1 + \frac{i\omega}{0,06}$	$1 + \frac{i\omega}{0,48}$	$1 + i\omega$	$(1 + \frac{i\omega}{4})^2$	TOT
0,06	-90°	-45°	+5°	0	0	-130°
0,48	-90°	-80°	+45°	+25°	0	-160°
1	-90°	-90°	+65°	+45°	-20°	-90°
4	-90°	-90°	+80°	+75°	-90°	-115°

$$F(i\omega) = 25 \cdot \frac{(1 + i\omega)}{i\omega (1 + \frac{i\omega}{4})^2} \cdot \frac{1 + \frac{i\omega}{0,48}}{1 + \frac{i\omega}{0,06}}$$