

# Semester Project

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# Contents

<b>1</b>	<b>Introduction</b>	<b>3</b>
<b>2</b>	<b>Theory</b>	<b>4</b>
2.1	Randomness and Entropy . . . . .	4
2.2	True vs. Pseudo Randomness . . . . .	4
2.3	Hypothesis . . . . .	4
2.4	Theoretical Modeling of Analog Noise . . . . .	5
<b>3</b>	<b>Implementation</b>	<b>7</b>
3.1	Hardware Setup . . . . .	7
3.2	Data Transmission . . . . .	7
<b>4</b>	<b>Results and Analysis</b>	<b>8</b>
4.1	Results from the Raw Signal . . . . .	8
4.2	Processing . . . . .	9
4.3	Network Design . . . . .	11
4.4	Subnetting and Addressing . . . . .	11
4.5	Data Security and Transmission Integrity . . . . .	11
<b>5</b>	<b>Conclusion</b>	<b>13</b>
5.1	Future Work . . . . .	13

# 1 Introduction

In this project we use the STM32 microcontroller along with analog sensors to gather data that we analyze for statistical and informational randomness, with the goal of evaluating how feasible such a setup is as a hardware-based true random number generator (TRNG). The data will be sent via serial connection to a computer to analyze the quality of the data.

We will combine theory from information theory, probability, signal processing, and networking to assess both the quality of randomness and how such a system could integrate into a distributed architecture.

We will discuss:

- Theoretical background on randomness and entropy.
- Implementation and data transfer.
- Statistical analysis and randomness testing.
- Viability discussion and possible improvements.
- Network design for scaling such systems.

## 2 Theory

### 2.1 Randomness and Entropy

Randomness can be defined both in a statistical and algorithmic sense. A sequence is considered random if it is unpredictable and lacks compressible structure.

#### Definition 2.1: Shannon Entropy

Given a discrete random variable  $X$  that takes values in  $\chi$  with probability distribution  $p : \chi \rightarrow [0, 1]$ , its entropy is

$$H(X) = - \sum_{x \in \chi} p(x) \log_2 p(x).$$

(as introduced by Shannon [1] and discussed in [2, 4]). Entropy is a central measure in information theory pertaining to the randomness of data and it will be referenced extensively throughout this text.

### 2.2 True vs. Pseudo Randomness

A pseudo-random number generator (PRNG) produces deterministic sequences from an initial seed, while a true random number generator (TRNG) relies on physical entropy [3]. For hardware-based RNGs, ensuring unbiased and unpredictable output requires:

- High-quality analog entropy sources.
- Statistical post-processing.

### 2.3 Hypothesis

We believe that due to the inherent noise of electrical circuits that dominate at low frequencies, the least significant bits (LSBs) of our signal should be a source of high entropy. Then, with further post-processing we should get a high-quality source of random numbers which can be applied in processes which require randomness.

## 2.4 Theoretical Modeling of Analog Noise

To justify that the least significant bits of our ADC contains randomness, we model the analog sensor output as a combination of deterministic signal and noise:

$$V_{\text{ADC}}(t) = V_{\text{Signal}}(t) + V_{\text{Noise}}(t)$$

- $V_{\text{Signal}}$  represents the slowly varying, predictable component (e.g. ambient light level in the case of photoresistor).
- $V_{\text{Noise}}$  represents the unpredictable physical noise (thermal-, shot noise, quantization error, sensor-specific noise).

The noise, which is the sum of thermal-, shot-, and other electronic noise is typically modeled as a Gaussian random variable [5]:

$$V_{\text{Noise}}(t) \sim \mathcal{N}(0, \sigma_{\text{Noise}}^2)$$

The **ADC quantization** transforms the continuous voltage into discrete levels:

$$X = \text{ADC}(V_{\text{ADC}}) \in \{0, 1, \dots, 2^{12} - 1\}$$

The voltage corresponding to one LSB (least significant bit) step is

$$\Delta V = \frac{V_{\text{ref}}}{2^{12}}$$

for a 12-bit ADC with reference voltage  $V_{\text{ref}}$ .

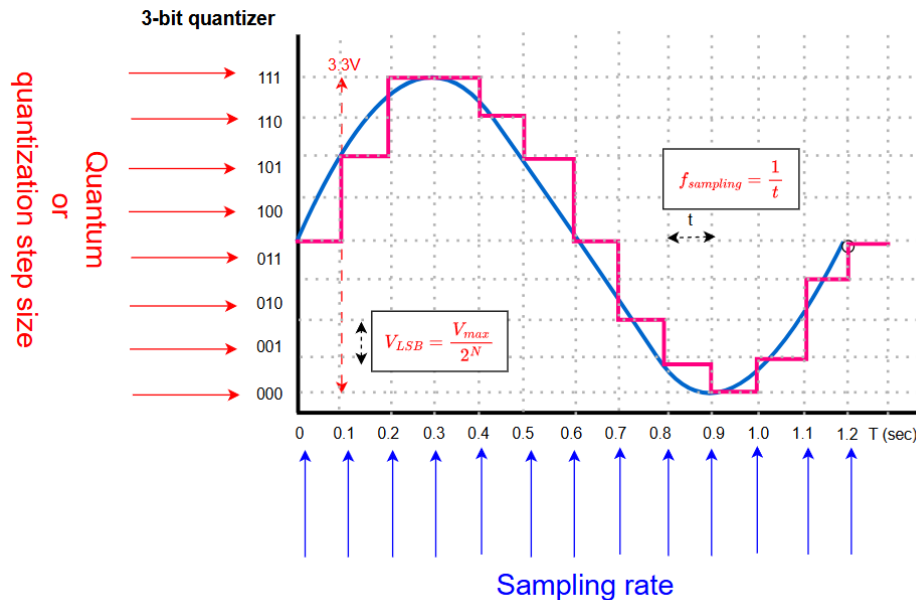


Figure 1: Quantization of analog signal.

The randomness of the extracted bits is measured by the Shannon Entropy. For a discrete random variable  $X$  with  $k$  possible outcomes, the maximum possible entropy  $H_{\text{MAX}} = \log_2(k)$  bits.

We extract the 4 LSBs, which gives  $2^4 = 16$  possible outcomes from  $0000_b \rightarrow 1111_b$ . The maximum entropy then is

$$H_{\text{MAX}} = 4 \text{ bits}$$

Maximum entropy is achieved if the the probability of observing any of the 16 outcomes is uniform [2, 4]:

$$P(D_{\text{LSB}} = i) = \frac{1}{16} \text{ for } 0 \leq i \leq 15$$

To achieve a near-uniform distribution across the 16 LSB states, the standard deviation of the noise  $\sigma_{\text{Noise}}$  must be large enough to span multiple quantization steps.

Let  $\Delta V_{4\text{-bit}}$  be the voltage corresponding to the 4 LSBs:

$$\Delta V_{4\text{-bit}} = 16\Delta V$$

Consider an input voltage  $V_{\text{ADC}}$  that falls within a  $\Delta V_{4\text{-bit}}$  window. The probability that the resulting digital word  $D$  falls into a specific 4-bit LSB state  $i$  depends on the area under the Gaussian Probability Density function of the noise that falls into the corresponding voltage interval  $I_i$ .

The critical condition for near-maximum entropy is then:

$$\sigma_{\text{Noise}} \gg \Delta V$$

If the noise standard deviation is significantly larger than the LSB step size then the Gaussian noise distribution becomes "smeared" across multiple LSB intervals. Since the noise is zero-mean and  $\sigma_{\text{Noise}}$  is large, the probability of the total input  $V_{\text{ADC}}$  falling into any one LSB interval  $I_i$  is nearly equal to the probability of it falling into an adjacent  $I_{i\pm 1}$ .

### 3 Implementation

#### 3.1 Hardware Setup

We will use the STM32f767zi microcontroller, connected to a photoresistor.

The microcontroller samples the analog voltage via the ADC and stores or streams the digital values over a serial or network interface. We will be sampling at 12-bits, keeping the 4 least significant bits as we believe these will be most susceptible to noise.

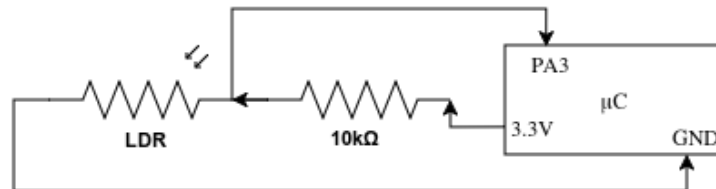


Figure 2: Diagram

#### 3.2 Data Transmission

Data is transmitted from the STM32 to a computer for analysis. In our experimentation we use US-ART/Serial Connection using a micro-usb connection from the microcontroller to a computer.

## 4 Results and Analysis

### 4.1 Results from the Raw Signal

We begin by analyzing the apparent randomness of the raw signal.

In our experimentation we first use Python with the serial library to read the incoming data and matplotlib to plot the signal over time.

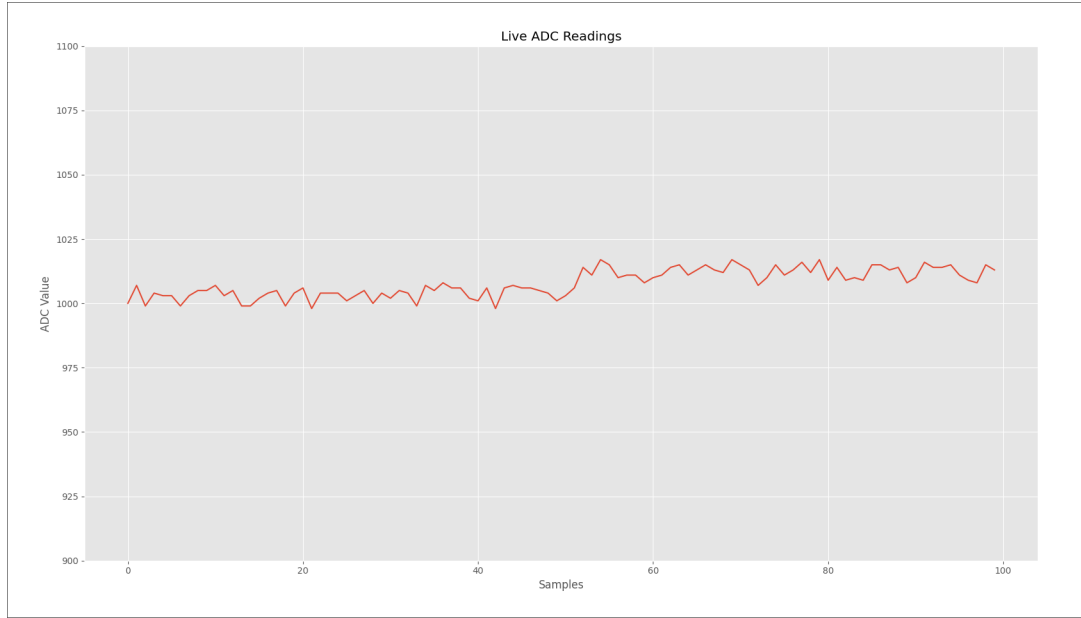


Figure 3: Plot of the raw signal over time

As we can see in this plot, while the signal remains relatively stable, reflecting the stable light level we were testing against, there are definite micro-jitters present. This is a good sign and is almost certainly due to the noise which we predicted would dominate at small levels.

We do have to take into account the fact that the micro-jitters might be due to slight fluctuations of the ambient light-level in our tests. To mitigate this possibility we will from this time forward be testing with the photoresistor covered with electrical tape.



Now we look at the observed probabilities pertaining to the 4 LSBs. We want a roughly equal probability that a given bit in the 4 LSBs is 0 or 1.

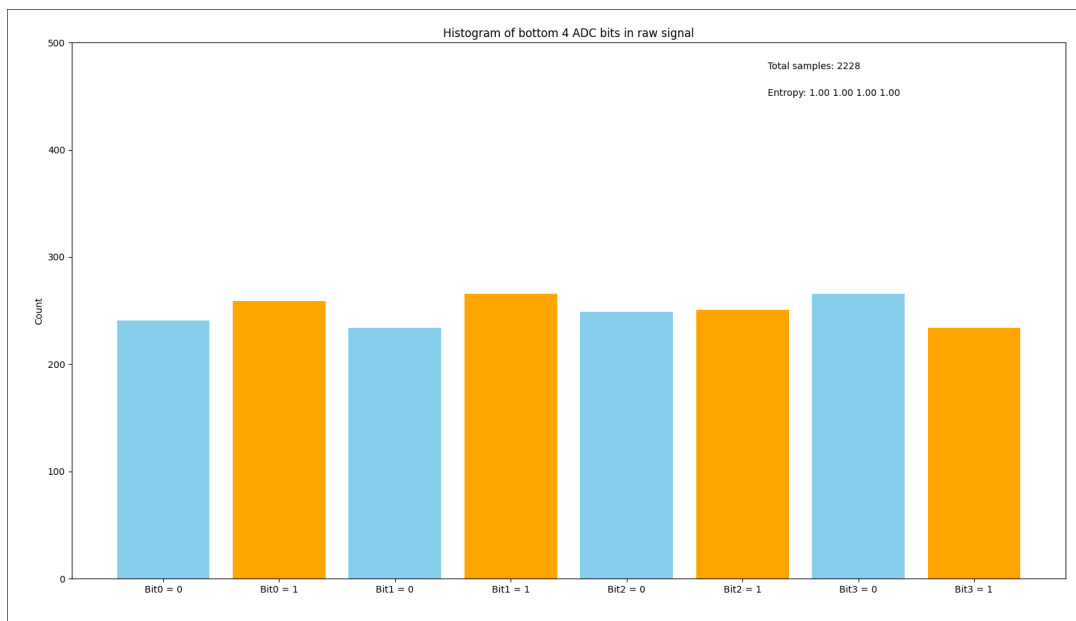


Figure 4: Histogram of 4 LSBs with photoresistor covered

As we can see there appears to be a roughly equal chance that a given bit in the 4 LSBs is 0 or 1, pointing towards high entropy in the LSBs.

Furthermore, we can see that each bit has an entropy of 1 and the sum of the 4 LSBs entropies is 4. As we determined earlier the maximum entropy  $H_{MAX} = 4$ , meaning that the 4 LSBs exhibit maximum entropy.

## 4.2 Processing

Our goal is to transform the raw data stream (which is **biased** and **predictable**) into a shorter, statistically perfect random stream. So far we've already done some processing in discarding all, but the 4 LSBs.

As discussed previously, the 4 LSBs are a good source of entropy, but using 4 bits at a time is not sensible as it is relatively easy to predict given the small number of combinations at only 16.

Now, since we have observed that each individual bit of the LSBs exhibit a maximum entropy (1), then taking 64 samples and appending each by bit-shifting 4 bits to the left should give us a 256 bit number with maximum entropy.

In this section we will examine 4 central metrics in determining the quality of the 256-bit number.

1. **Entropy.** Entropy is still important and we want an entropy as close to 1 as possible in this case.
2. **Bit frequency.** We examine how many 0s vs. 1s there are in a given bit-string. A 50/50 distribution is obviously desirable.
3. **Autocorrelation.** The correlation between consecutive bits is important and we want to be as close to 0 as possible, indicating no correlation/pattern in consecutive bits.
4. **Runs.** Runs are the number of "blocks" of like bits. For example  $0111110_b$  contains 3 blocks, the first 0, the consecutive 1s and the final 0. A run of 256 is bad because it indicates alternating

0s and 1s. 1 is also very bad as it indicates that the string only has 0s or 1s. Hence we want a run-count somewhere in the middle of these extremes.

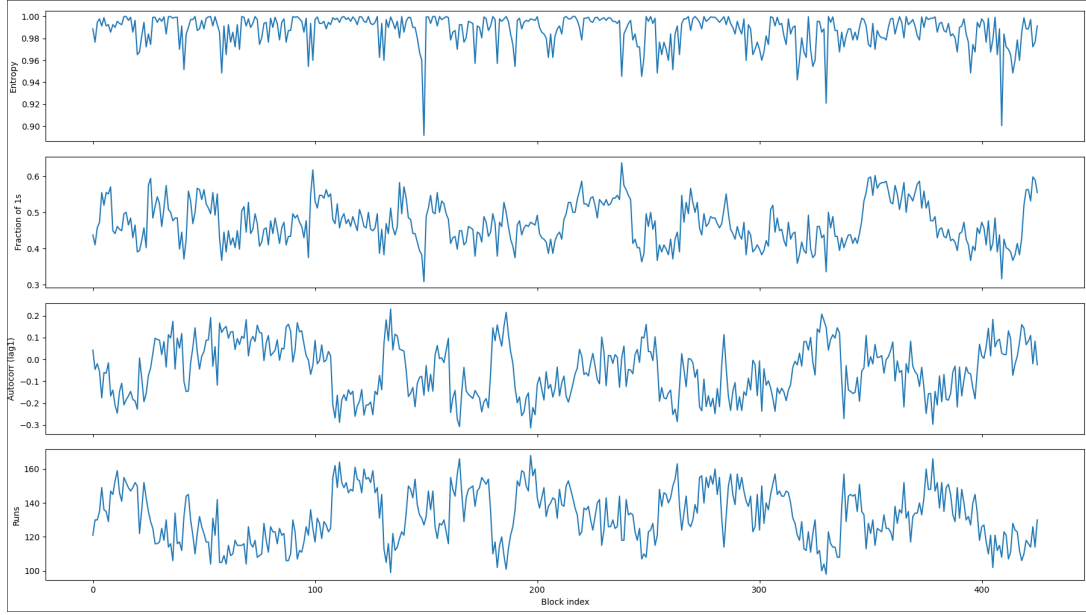


Figure 5: Readings of various metrics in 256-bit number.

Our readings show that our 256-bit numbers tend to hover around the desired values, never reaching any disconcerting values during the test that produced this plot. During this test, 434 numbers were analyzed with the following averages:

1. Average Entropy: 0.987
2. Average Bit Frequency: 0.472
3. Average Autocorrelation:  $-0.045$
4. Average Runs: 131.990

These results are promising, but we want to employ a more stringent test to analyze the **RNG**-quality of our 256-bit blocks. The employed test suite enforces a rigorous hypothesis testing framework, where the **null hypothesis**  $H_0$  posits the bitstream as being perfectly random, and results are accepted only if the calculated P-value is greater than the significance level  $\alpha = 0.01$ . Specifically, the **Monobit Frequency Test** assesses the uniformity of 0s and 1s by calculating the Z-score,

$$Z = |S_n - n/2| / \sqrt{n/4}$$

, where  $S_n$  is the sum of ones in the  $n = 256$  bit sequence, and deriving the P-value via the complementary error function,  $P_1 = \text{erfc}(Z/\sqrt{2})$ . The **Runs Test** evaluates the occurrence of contiguous identical bits by calculating the observed number of runs  $V_n$  and comparing it against the expected mean  $\mu$  and variance  $\sigma^2$ , again yielding a Z-score

$$Z = |V_n - \mu| / \sqrt{\sigma^2}$$

for the P-value  $P_2$ . Finally, the **Serial Overlapping Blocks Test** serves as a strong proxy for Approximate Entropy by using a  $\chi^2$  goodness-of-fit statistic

$$\chi^2 = \sum_{i=1}^{16} \frac{(C_i - E)^2}{E}$$

across all  $2^m = 16$  possible  $m = 4$  bit blocks, with the P-value  $P_3$  determined by the Chi-squared survival function  $\chi^2.sf(\chi^2, 2^m - 1)$ . The block is certified as *random* if and only if  $P_1 \geq \alpha \wedge P_2 \geq \alpha \wedge P_3 \geq \alpha$ , providing a composite measure highly resistant to statistical bias.

With this approach we quickly discover the limitations of using only the raw signal of our LDR to create our numbers.

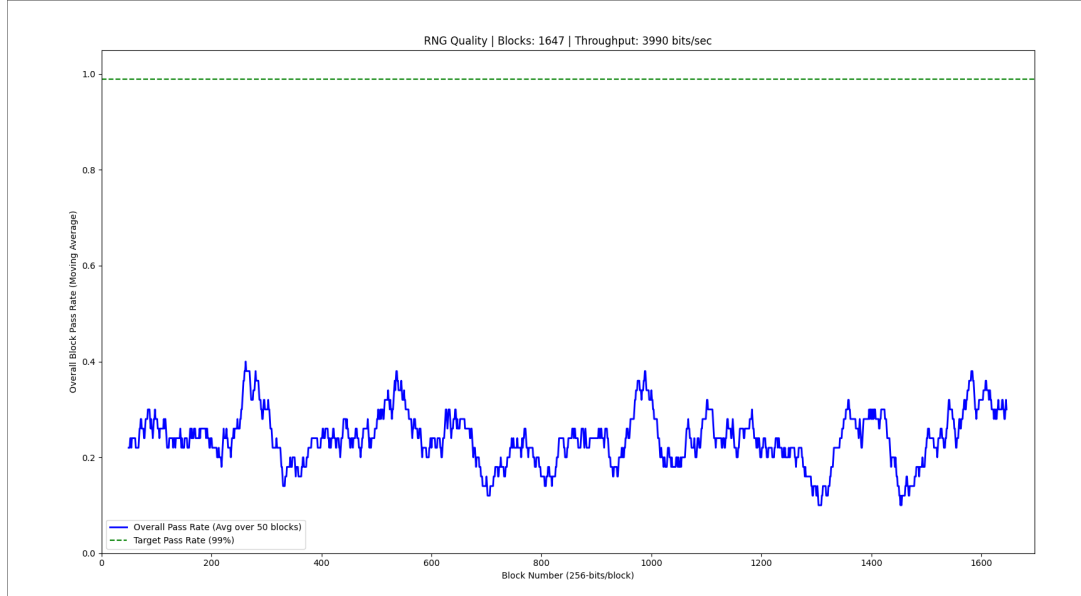


Figure 6: Results from our test suite

It is clear, from the fact that less than half of our 256-bit numbers pass all our tests, that some more involved processing is required to get high-enough-quality random numbers.

### 4.3 Network Design

A possible setup:

- Each microcontroller is connected to a local router.
- A dedicated subnet for sensor nodes (e.g., 192.168.10.0/24).
- Central analysis server on a separate subnet.

### 4.4 Subnetting and Addressing

We show how to subnet the network efficiently:

- Example: dividing a /24 into four /26 subnets.
- Assigning IP ranges for sensors, analysis nodes, and administration.

### 4.5 Data Security and Transmission Integrity

We briefly discuss:

- Packet integrity verification (CRC or checksum).
- Optional encryption for data in transit.
- Synchronization and time-stamping for accurate sampling.

## 5 Conclusion

We summarize: Fundamentals of Precision ADC Noise Analysis

- Theoretical feasibility of analog-sensor-based TRNGs.
- Experimental results and limitations.
- Potential for distributed entropy networks.

### 5.1 Future Work

- Hardware whitening circuits and amplification.
- FPGA or ASIC implementations.
- Scaling to larger networks and entropy pooling.

## References

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