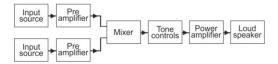
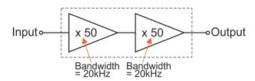
Audio Systems*

Revision sheet

1 Introduction

different There are lots components which The go into audio system. diagram shows mono audio system.



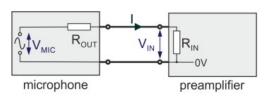


The gain and bandwidth can be calculated using the GBP equation $GainBandwidthProduct = Gain \times Bandwith$

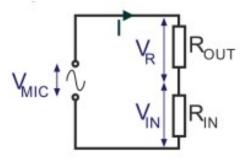
2 Imput & Pre-amp

A typical input would be a microphone. This produces a very small output so before the signal can be processed, it has to be amplified.

2.1 Signal transfer



As there are two resisitors drawn in series, this could also be shown as a potential divider.



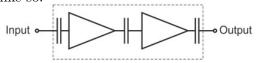
For maximum voltage teansfer, we want Rin >> Rout, this should be about 10X greater.

2.2 Pre-amplifier

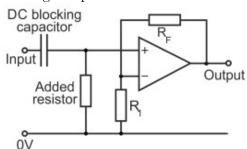
A multi-stage amplifier is usually used here. To maximise bandwidth, the stages need equal gain.

2.2.1 AC coupling

The op-amps used in the circuit amplify all frequencies, right down to 0Hz. We need to block these signals as they are DC signals and will break things further down the circuit. To do this, we use DC blocking coupling capacitors placed like so.



The amplifier design is based off of a non-inverting amplifier with one added resistor to give the DC signals a path to ground. Below is one stage of a two-stage amp, you'd need to connect the output of one into the input of the next to get a two-stage amp.

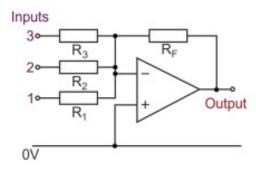


3 Audio Mixer

The audio mixer is based off of a Summing Amplifier.

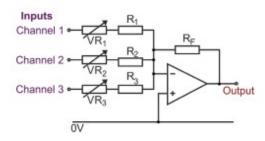
^{*}Images from Wjec E-book

3.1Summing amplifier



 $Vout = -R_f(\frac{V_1}{R_1} + \frac{V_2}{R_2} + \frac{V_3}{R_3})$ This basic design isn't great as you can't contol the volume of each channel.

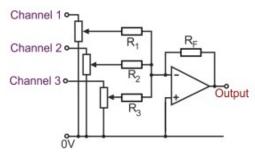
Adding gain control 3.2



Channel gain: $Ch.gain = \frac{-R_F}{F_1}$

Minimum channel gain: $Min.ch.gain = \frac{-R_F}{VR+R}$ Maximum channel gain: $Max.ch.gain = \frac{-R_F}{-R_F}$ There is a flaw with this design. The gain is always greater than 1 therefore you can't completely mute a channel (as this would require Gain=0). There a better design which fixes this.

3.3 Potentiometer design



As this design has potentiometers, we can change the ratio of resistors so 'all the voltage flows through the one which we aren't taking the output of', allowing us to set gain=0.

Tone Controls 4

Tone controls are designed using active filters. The main difference between active and passive Gain at high frequencies: $G = -\frac{r}{R}$

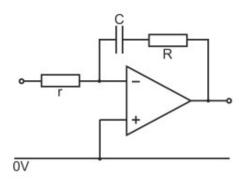
filters is that active filters are built using an opamp therefore their gain can exceed 1, allowing us to have boost filter.

Break Frequency 4.1

It is defined as the frequency at which the reactance of the capacitor is equal to the resistance of the resistor that is in the RC network. This can be calculated using the formula; $f_b = \frac{1}{2\Pi RC}$ The formula is the same for all the types of active filter. It is important to choose the right resistor, which is the one 'touching' the capacitor.

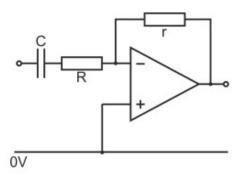


Bass Boost 4.1.1

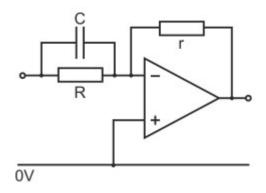


Gain at high frequencies: $G = -\frac{R}{r}$

Bass Cut 4.1.2

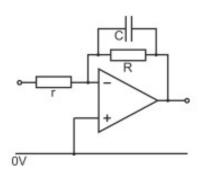


4.1.3 Treble Boost



Gain at low frequencies: $G = -\frac{r}{R}$

4.1.4 Treble cut

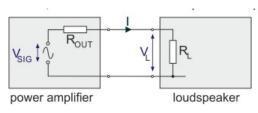


Gain at low frequencies: $G = -\frac{R}{r}$

5 Power Amplifiers

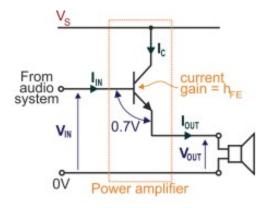
This section also covers transfer to loudspeakers. Power Amplifiers are used to amplify the current of the output of the tone controls before the signal enters the loudspeaker.

5.1 Maximum Power Transfer theorem



The maximum power is transferred when the resistance of the load is equal to the output resistance of the amplifier.

5.2 Emmitter Follower

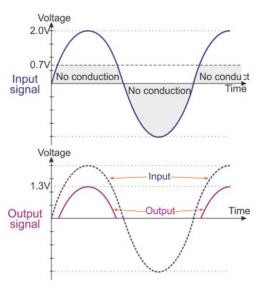


Power in: $P_{in} = V_{in} \times I_B$ Power out: $P_{out} = h_{FE} \times P_{in}$ Imput Impedance: $R_{in} = \frac{V_{in}}{h_{FE}}$ Output Impedance: $R_{out} = \frac{R_{in}}{h_{FE}}$ The output impedance is much smaller than the

The output impedance is much smaller than the input impedance therefore this an impedance buffer as it transforms the impedance. This matches the high output impedance of the tone controls to the low load impedance of a loud-speaker.

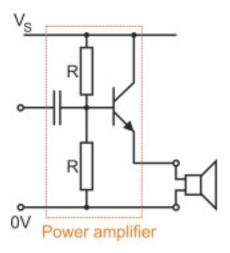
There are a number of problems with this design which make it completely impractical to use.

- Not efficient
- Waste lots of power
- They have a region where nothing is amplified.
- Generally, they're terrible



As the graph shows, everything below 0.7V is not amplified, and what is above it is reduced by 0.7V. To sovle this, we need to add a DC voltage to Vin, this is done by biasing the amp.

5.3 Biased amplifier



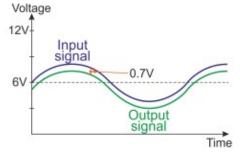
As the resistors are equal, Vs will be split into two, giving a bias voltage.

$$V_{bias} = \frac{V_s}{2}$$

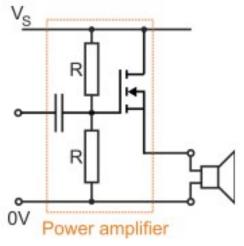
 $V_{bias} = \frac{V_s}{2}$ $V_x = V_{in} + V_{bias}$ (V_x is the voltage after the

$$V_{out} = V_x - 0.7$$

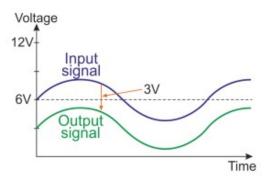
This gives much better output



Source Follower **5.4**



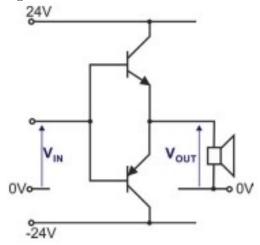
This type of amplifer has an advantage over the emitter follower due to its infinite input resitance. The MOSFET doesn't conduct until the input signal reaches 3V, hence the amplitude of the output is 3V smaller than the amplitude of the input.

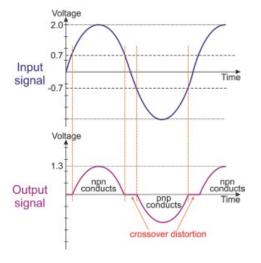


A disadvantage of both the Source follower and the emitter follower is that the transistor is aloways conducting, even when there is no input signal - so it is always dissipating energy as heat.

Push-pull amplifiers 5.5

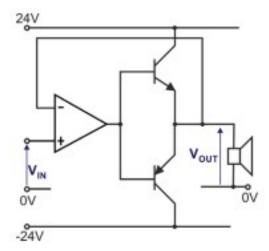
If we use a NPN transistor (the type we've been using already) and a PNP (the same as a NPN but reversed polarity) together, we can get an amplifier which will amplifier both positive and negative signals.





As seen on the graph, there is a region (between 0.7 and -0.7) where neither of the amplifiers conducts. This is called crossover distortion and it can be fixed by adding an op-amp to the circuit.

5.6 Improved Push-Pull amplifer



This solves the crossover distortion problem. This works by the op-amp trying to keep the inverting and non-inverting inputs at the same voltage, unless the ouptu saturates. The inverting input is directly connected to the push-pull output and so the voltage at the output is always equal to the voltage at the non-inverting input, hence eliminating crossover distortion.

5.7 Maximising power output

 $Max.P_{out}=P_{RMS_{MAX}}=\frac{V_s^2}{8R_L}$ where V_s is the voltage between the rails.