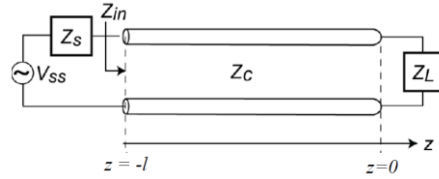


HARMONIC SIGNAL



$$\beta = \omega \sqrt{L'C'} = \omega \sqrt{\epsilon \mu} = \frac{\omega}{v_p} = \frac{2\pi}{\lambda} \left[\frac{1}{m} \right]$$

$$V^+(z) = V_0^+ e^{-j\beta z}$$

$$V^-(z) = V_0^- e^{+j\beta z}$$

Reflection coefficient and entrance impedance:

$$\Gamma(0) \equiv \frac{V_0^-}{V_0^+} = \Gamma_L = \frac{Z_L - Z_c}{Z_L + Z_c} \quad Z_{in}(z) = Z_c \frac{1 + \Gamma(z)}{1 - \Gamma(z)}$$

$$\Gamma(z) \equiv \frac{V^-(z)}{V^+(z)} = \Gamma_L \exp(2j\beta z) \quad \Gamma(z) = \frac{Z_{in}(z) - Z_c}{Z_{in}(z) + Z_c}$$

$$Z_{in}(z = -L) = Z_c \frac{Z_L + jZ_c \tan(\beta L)}{Z_c + jZ_L \tan(\beta L)}$$

$$Z_{in}\left(z = -\frac{\lambda}{4}\right) = \frac{Z_c^2}{Z_L}$$

$$\begin{cases} V(z) = V_0^+ e^{-j\beta z} [1 + \Gamma(z=0) e^{j2\beta z}] \\ I(z) = \frac{V_0^+}{Z_c} e^{-j\beta z} [1 - \Gamma(z=0) e^{j2\beta z}] \end{cases}$$

POWER AND ENERGY CALCULATIONS

$$P = P_{av} = \langle \mathcal{V}(z, t) \mathcal{I}(z, t) \rangle = \frac{1}{2} \text{Re}[V(z) I^*(z)]$$

$$= \frac{1}{2} |V(z)|^2 \text{Re} \left[\frac{1}{Z(z)} \right] = \frac{1}{2} |I(z)|^2 \text{Re}[Z(z)]$$

Power of the wave:

$$P^\pm = \frac{1}{2} \text{Re}\{V^\pm I^{\pm*}\} = \pm \frac{1}{2} |V^\pm|^2 \text{Re} \left[\frac{1}{Z_c} \right]$$

$$P = P^+ + P^-$$

Lossless circuit: entrance power equals power on the load

$$P_L = P_{in} = \frac{1}{2} \text{Re}[V_{in} I_{in}^*] = \frac{1}{2} |V_{in}|^2 \text{Re} \left[\frac{1}{Z_{in}} \right]$$

From the Telegraph Eq. we get the wave Eq.:

$$\frac{d^2 V}{dz^2} = L' C' \frac{d^2 V}{dt^2}$$

$$\frac{d^2 I}{dz^2} = L' C' \frac{d^2 I}{dt^2}$$

The general solution:

$$\begin{cases} V(z, t) = V^+ \left(t - \frac{z}{v_p} \right) + V^- \left(t + \frac{z}{v_p} \right) \\ I(z, t) = I^+ \left(t - \frac{z}{v_p} \right) + I^- \left(t + \frac{z}{v_p} \right) \end{cases}$$

The propagation velocity:

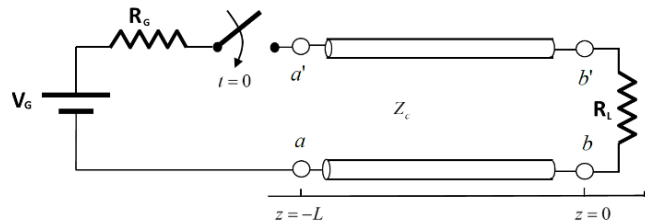
$$v_p = \frac{1}{\sqrt{L' C'}}$$

The characteristic Impedance:

$$Z_c = \sqrt{\frac{L'}{C'}} = \frac{V^+}{I^+} = -\frac{V^-}{I^-}$$

Time Domain solution:

$$\begin{cases} V(z, t) = V_0 + \sum_{n=1}^{\infty} V_n^+ \left(t - \frac{z}{v_p} \right) + \sum_{n=1}^{\infty} V_n^- \left(t + \frac{z}{v_p} \right) \\ I(z, t) = I_0 + \sum_{n=1}^{\infty} I_n^+ \left(t - \frac{z}{v_p} \right) + \sum_{n=1}^{\infty} I_n^- \left(t + \frac{z}{v_p} \right) \end{cases}$$



Reflection coefficient at the load:

$$\Gamma_L = \frac{V_n^-(t, z=0)}{V_n^+(t, z=0)} = \frac{R_L - Z_c}{R_L + Z_c}$$

Reflection coefficient at the source:

$$\Gamma_G = \frac{V_{n+1}^+(t, z=-L)}{V_n^-(t, z=-L)} = \frac{R_G - Z_c}{R_G + Z_c}$$

GENERAL

$$\eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}} = 377 \Omega$$

$$\mu_0 = 4\pi \cdot 10^{-7} \frac{\text{Henry}}{\text{m}}, \quad \epsilon_0 = 8.85 \cdot 10^{-12} \frac{\text{Farad}}{\text{m}}$$

$$c = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = 2.9979 \cdot 10^8 \frac{\text{m}}{\text{s}}$$

$$S_n \equiv \sum_{k=0}^n r^k = \frac{1 - r^{n+1}}{1 - r}; \quad S_\infty = \sum_{k=0}^{\infty} r^k = \frac{1}{1 - r}$$

$$\sin(jx) = j \sinh(x); \quad \cos(jx) = \cosh(x)$$

$$e^{jx} = \cos(x) + j \sin(x)$$

$$z = x + jy = |z| e^{j\theta}$$

$$|z| = \sqrt{x^2 + y^2} \quad \theta = \arctan\left(\frac{y}{x}\right)$$

PHASORS

$$V(t) = V_0 \cos(\omega t + \theta) = \text{Re}\{V_0 e^{j\omega t} e^{j\theta}\} = \text{Re}\{\tilde{V} e^{j\omega t}\}$$

$$\tilde{V} \equiv V_0 e^{j\theta}$$

$$\frac{d}{dt} V(t) \Leftrightarrow j\omega \tilde{V}$$

DISTRIBUTED LINE IN LC MODEL (WITHOUT LOSSES)

Inductance and Capacitance per unit length

$$L = L' \Delta z \quad L' [\text{H/m}] \quad C' [\text{F/m}]$$

$$C = C' \Delta z$$

The Telegraph Eq. in the limit of $\Delta z \rightarrow 0$

$$\frac{dV}{dz} = -L' \frac{dI}{dt}$$

$$\frac{dI}{dz} = -C' \frac{dV}{dt}$$

For this solution we get:

$$\vec{E} \perp \vec{H} \perp \vec{k}$$

$$\frac{|\vec{E}|}{|\vec{H}|} = \eta = \sqrt{\frac{\mu}{\epsilon}} = \sqrt{\frac{\mu_0 \mu_r}{\epsilon_0 \epsilon_r}} = \eta_0 \sqrt{\frac{\mu_r}{\epsilon_r}} = \frac{\eta_0}{n}$$

$$\vec{H} = \frac{\hat{k} \times \vec{E}}{\eta} = \frac{1}{\eta} (\hat{k} \times \vec{E}_0) e^{-j\vec{k} \cdot \vec{r}}$$

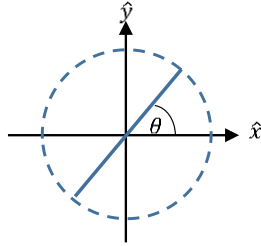
POLARIZATION OF a PLANE WAVE

In frequency:

$$\vec{E}(z_0) = \hat{x}|E_x| + \hat{y}|E_y|e^{j\phi_y - \phi_x}$$

In time:

$$\vec{E}(z_0, t) = \hat{x}|E_x| \cos(\omega t) + \hat{y}|E_y| \cos(\omega t + \phi_y - \phi_x)$$



LINEAR POLARIZATION

$$\phi_y - \phi_x = m \cdot \pi \quad ; \quad m \in \mathbb{Z}$$

$$\vec{E}(z_0, t) = \hat{x}|E_x| \cos(\omega t) \pm \hat{y}|E_y| \cos(\omega t)$$

$$\vec{E}(z_0) = |E_x|\hat{x} \pm |E_y|\hat{y}$$

$$\theta = \tan^{-1} \left(\pm \frac{|E_y|}{|E_x|} \right); \quad 0 < \theta < \pi$$

$$|E_x| = E_0 \cos(\theta), \quad \pm |E_y| = E_0 \sin(\theta)$$

REFLECTION WITH LOSSES

$$\Gamma(0) = \frac{Z_L - Z_C}{Z_L + Z_C}, \quad \Gamma(z) = \Gamma(0)e^{+2\gamma z}$$

$$\Gamma(-L) = \Gamma(0)e^{-2\gamma L} = \Gamma(0)e^{-2j\beta L - 2\alpha L}$$

$$Z_{in}(-L) = Z_C \frac{1 + \Gamma(-L)}{1 - \Gamma(-L)} = Z_C \frac{Z_L + Z_C \tanh(j\beta L + \alpha L)}{Z_C + Z_L \tanh(j\beta L + \alpha L)}$$

$$\begin{aligned} V(z) &= V_0^+ [e^{-\alpha z} e^{-j\beta z} + \Gamma_L e^{+\alpha z} e^{+j\beta z}] \\ &= V_0^+ e^{-\alpha z} e^{-j\beta z} [1 + \Gamma_L e^{+2\alpha z} e^{+2j\beta z}] \end{aligned}$$

$$\begin{aligned} I(z) &= \frac{V_0^+}{Z_0} [e^{-\alpha z} e^{-j\beta z} - \Gamma_L e^{+\alpha z} e^{+j\beta z}] \\ &= \frac{V_0^+}{Z_0} e^{-\alpha z} e^{-j\beta z} [1 - \Gamma_L e^{+2\alpha z} e^{+2j\beta z}] \end{aligned}$$

POWERS

$$P(z) = \frac{1}{2} \text{Re}\{VI^*\} =$$

$$= \frac{1}{2} \text{Re}\{ |V_0^+|^2 Y_C^* e^{-2\alpha z} (1 - |\Gamma_L|^2 e^{+4\alpha z}) \}$$

$$P_{in}(z = -l) = \frac{1}{2} \text{Re}\{ |V_0^+|^2 Y_C^* e^{2\alpha l} (1 - |\Gamma_L|^2 e^{-4\alpha l}) \}$$

$$P_{load}(z = 0) = \frac{1}{2} \text{Re}\{ |V_0^+|^2 Y_C^* (1 - |\Gamma_L|^2) \}$$

PLANE WAVES

$$\begin{aligned} \text{Notating: } \left\{ \begin{array}{l} \nabla \times \vec{E} = -j\omega\mu\vec{H} \\ \nabla \times \vec{H} = j\omega\epsilon\vec{E} \\ \nabla \cdot \vec{E} = 0 \\ \nabla \cdot \vec{H} = 0 \end{array} \right\} &\stackrel{\nabla \times \nabla \times \vec{A} = -\nabla^2 \vec{A} + \nabla(\nabla \cdot \vec{A})}{\Rightarrow} \left\{ \begin{array}{l} \nabla^2 \vec{E} + \omega^2 \mu\epsilon \vec{E} = 0 \\ \nabla^2 \vec{H} + \omega^2 \mu\epsilon \vec{H} = 0 \end{array} \right. \end{aligned}$$

The solution is:

$$\vec{E} = \vec{E}_0 e^{-j\vec{k} \cdot \vec{r}}$$

$$\vec{k} \cdot \vec{r} = k_x x + k_y y + k_z z$$

$$k = |\vec{k}| = 2\pi n / \lambda_0 = \omega \sqrt{\mu\epsilon} = \omega n / c = \sqrt{k_x^2 + k_y^2 + k_z^2}$$

STANDING WAVE RATIO (SWR)

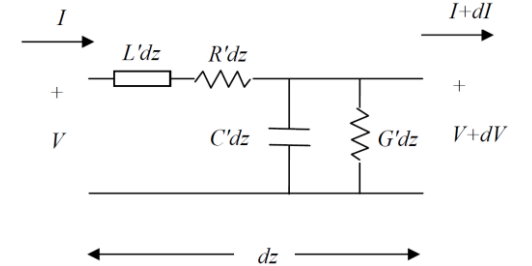
$$S = \frac{|V_{\max}|}{|V_{\min}|} = \frac{1 + |\Gamma|}{1 - |\Gamma|} \geq 1$$

$$|V_{\max}| = |V_0^+|(1 + |\Gamma|)$$

$$|V_{\min}| = |V_0^+|(1 - |\Gamma|)$$

$$|\Gamma(z)| = |\Gamma(0)e^{2j\beta z}| = |\Gamma(0)|$$

TRANSMISSION LINE WITH LOSSES



Conductivity (heat) losses – R'

Dielectric losses – G'

WAVE EQUATION WITH LOSSES

$$\begin{cases} \frac{dV}{dz} = -IR' - L' \frac{dI}{dt} \\ \frac{dI}{dz} = -VG' - C' \frac{dV}{dt} \end{cases}$$

$$\frac{d^2 V}{dz^2} = L'C' \frac{d^2 V}{dt^2} + (C'R' + L'G') \frac{dV}{dt} + R'G'V$$

The solution to the equation:

$$V = \text{Re} \{ V_0 e^{-\gamma z} e^{j\omega t} \}$$

$$\gamma = \alpha + j\beta = \pm \sqrt{(R' + j\omega L')(G' + j\omega C')} \quad [1/\text{m}]$$

$$Z_C(\omega) = \frac{V^+}{I^+} = R_0 + jX_0 = \sqrt{\frac{R' + j\omega L'}{G' + j\omega C'}}$$

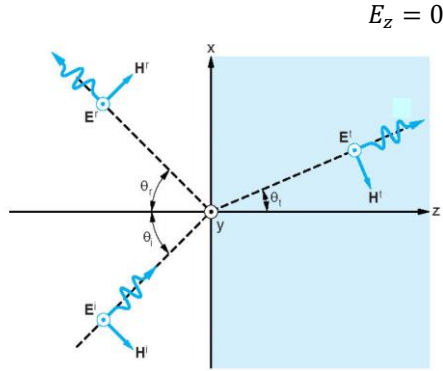
Small losses: $R' \ll \omega L', G' \ll \omega C'$

In first order approximation:

$$\beta \approx \omega \sqrt{L'C'} \quad \alpha \approx \frac{1}{2} \sqrt{L'C'} \left(\frac{R'}{L'} + \frac{G'}{C'} \right)$$

Under this approximation, the decay coefficient is not frequency dependent and the propagation coefficient has a linear dependency (no dispersion).

TE SOLUTION (PERPENDICULAR)



$$\vec{k} = k_x \hat{x} + k_z \hat{z} = k(\hat{x} \sin \theta + \hat{z} \cos \theta)$$

$$\hat{e}_{TE} = \hat{y}$$

$$\hat{h}_{TE} = -\hat{x} \cos \theta + \hat{z} \sin \theta$$

$$\vec{E}_{TE} = E_{TE} \hat{y} e^{-jk \sin \theta x} e^{-jk \cos \theta z}$$

$$E_y = e^{-jk \sin \theta x} [V_0^+ e^{-jk \cos \theta z} + V_0^- e^{jk \cos \theta z}], \quad V_0^+ = E_{TE}$$

TL MODEL FOR TE POLARIZATION:

$$E_y^+ \leftrightarrow V^+$$

$$-H_x^+ \leftrightarrow I^+$$

$$k_z = k \cos \theta \leftrightarrow \beta$$

$$\frac{\eta}{\cos \theta} \leftrightarrow Z_{CTE}$$

The longitudinal component:

$$H_z = \left[\frac{1}{Z_{CTE}} \tan \theta \right] E_y$$

FRESNEL EQUATIONS FOR TE:

(For the transverse component E_y):

$$r_{21}^{TE} = \frac{\frac{\eta_2}{\cos \theta_2} - \frac{\eta_1}{\cos \theta_1}}{\frac{\eta_2}{\cos \theta_2} + \frac{\eta_1}{\cos \theta_1}} = \frac{n_1 \cos \theta_1 - \frac{\mu_1}{\mu_2} n_2 \cos \theta_2}{n_1 \cos \theta_1 + \frac{\mu_1}{\mu_2} n_2 \cos \theta_2}$$

$$t_{21}^{TE} = \frac{2 \frac{\eta_2}{\cos \theta_2}}{\frac{\eta_2}{\cos \theta_2} + \frac{\eta_1}{\cos \theta_1}} = \frac{2 n_1 \cos \theta_1}{n_1 \cos \theta_1 + \frac{\mu_1}{\mu_2} n_2 \cos \theta_2}$$

$$t_{21}^{TE} = r_{21}^{TE} + 1$$

Skin depth:

$$\delta_s = \frac{1}{\alpha}$$

In a metal (good conductor):

$$\epsilon_{\text{eff}} = \epsilon' - j \left[\epsilon'' + \frac{\sigma}{\omega} \right] \xrightarrow{\frac{\sigma}{\omega} \gg |\epsilon'|, |\epsilon''|} -j \frac{\sigma}{\omega}$$

$$k = \omega \sqrt{\mu \epsilon_{\text{eff}}} = \frac{1-j}{\sqrt{2}} \sqrt{\omega \mu \sigma}$$

$$\alpha = \beta = \sqrt{\frac{\omega \mu \sigma}{2}}, \quad \delta_s = \frac{1}{\alpha} = \sqrt{\frac{2}{\omega \mu \sigma}}$$

$$\eta = (1+j) \sqrt{\frac{\omega \mu}{2\sigma}}$$

Usually: $|Z_{\text{metal}}| \ll Z_{\text{other}}$

FRESNEL EQUATIONS FOR PERPENDICULAR INCIDENT:

Refractive index in a dielectric:

$$n = \sqrt{\epsilon_r}$$

The incident wave is coming from medium 1 to medium 2:

$$\begin{aligned} n_1 &= \sqrt{\epsilon_{r1}} & r_{21} &= \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{n_1 - n_2}{n_1 + n_2} \\ n_2 &= \sqrt{\epsilon_{r2}} & \Rightarrow & \\ & & t_{21} &= 1 + r_{21} = \frac{2\eta_2}{\eta_2 + \eta_1} = \frac{2n_1}{n_2 + n_1} \end{aligned}$$

ANGLED INCIDENT:

$$\vec{E}(\vec{r}) = \vec{E}_0 e^{-j\vec{k} \cdot \vec{r}}$$

$$= [E_{TE} \hat{e}_{TE} + E_{TM} \hat{e}_{TM}] e^{-j\vec{k} \cdot \vec{r}}$$

$$\vec{H}(\vec{r}) = \frac{1}{\eta} (\hat{k} \times \vec{E}_0) e^{-j\vec{k} \cdot \vec{r}}$$

$$= \left[\frac{E_{TE}}{\eta} \hat{h}_{TE} + \frac{E_{TM}}{\eta} \hat{h}_{TM} \right] e^{-j\vec{k} \cdot \vec{r}}$$

$$\hat{e}_{TM} \perp \hat{e}_{TE} \perp \hat{k}$$

CIRCULAR POLARIZATION

$$\phi_y - \phi_x = \pm \frac{\pi}{2} + 2\pi m; \quad m \in \mathbb{Z}$$

And the modulus of the amplitudes are equal

$$|E_x| = |E_y| = |E|/\sqrt{2}$$

RIGHT CIRCULAR

$$\phi_y - \phi_x = -\pi/2$$

$$\{\text{RC}\}: \vec{E}(z_0, t) = \frac{|E|}{\sqrt{2}} (\cos(\omega t) \hat{x} + \sin(\omega t) \hat{y})$$

$$\{\text{RC}\}: \vec{E}(z_0) = \frac{|E|}{\sqrt{2}} \cdot (\hat{x} - j\hat{y})$$

LEFT CIRCULAR

$$\phi_y - \phi_x = +\pi/2$$

$$\{\text{LC}\}: \vec{E}(z_0, t) = \frac{|E|}{\sqrt{2}} (\cos(\omega t) \hat{x} - \sin(\omega t) \hat{y})$$

$$\{\text{LC}\}: \vec{E}(z_0) = \frac{|E|}{\sqrt{2}} \cdot (\hat{x} + j\hat{y})$$

PERPENDICULAR INCIDENT

$$\vec{E}(z) = \vec{E}_0 e^{-j\beta z} = [E_{0x} \hat{x} + E_{0y} \hat{y}] e^{-j\beta z}$$

$$\vec{H}(z) = \frac{1}{\eta} (\hat{z} \times \vec{E}_0) e^{-j\beta z} = \left[\frac{E_{0x}}{\eta} \hat{y} - \frac{E_{0y}}{\eta} \hat{x} \right] e^{-j\beta z}$$

TL MODEL FOR PERPENDICULAR INCIDENT

$$E_x^+ / E_y^+ \leftrightarrow V^+$$

$$H_y^+ / -H_x^+ \leftrightarrow I^+$$

$$k = \omega \sqrt{\epsilon \mu} \leftrightarrow \beta$$

$$\eta = \sqrt{\frac{\mu}{\epsilon}} \leftrightarrow Z_c$$

MEDIUM WITH LOSSES AND CONDUCTIVITY

$$\gamma = jk = j\beta + \alpha$$

Dielectric and magnetic losses:

$$\epsilon = \epsilon' - j\epsilon'', \quad \mu = \mu' - j\mu''$$

With a finite conductivity:

$$\eta = \sqrt{\frac{\mu}{\epsilon - j \frac{\sigma}{\omega}}}, \quad k_2 = \omega \sqrt{\mu \left(\epsilon - j \frac{\sigma}{\omega} \right)},$$

PLANE WAVE SPECTRUM

Spectrum of the field in the z plane:

$$E_T(x; z) = \int_{-\infty}^{\infty} \tilde{E}_T(k_x; z) \cdot e^{-jk_x x} dk_x$$

$$\tilde{E}_T(k_x; z) = \frac{1}{2\pi} \int_{-\infty}^{\infty} E_T(x; z) \cdot e^{jk_x x} dx$$

Spectrum of the field at $z=0$:

$$E_T(x; 0) = \int_{-\infty}^{\infty} \tilde{E}_T(k_x; 0) \cdot e^{-jk_x x} dk_x$$

$$\tilde{E}_T(k_x; 0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} E_T(x; 0) \cdot e^{jk_x x} dx$$

The connection between the spectrum at z and the spectrum at $z=0$:

$$\tilde{E}(k_x; z) = \tilde{E}(k_x; 0) \cdot e^{-jk_z z} = \tilde{E}_0(k_x) \cdot e^{-j\sqrt{k^2 - k_x^2} z}$$

The field in $z > 0$:

$$E(x; z) = \int_{-\infty}^{\infty} \tilde{E}_0(k_x) \cdot e^{-j\sqrt{k^2 - k_x^2} z} \cdot e^{-jk_x x} dk_x$$

For far field we can neglect the decay waves with $|k_x| > k$

$$E(x; z) = \int_{-k}^k \tilde{E}_0(k_x) \cdot e^{-jk_x x} \cdot e^{-j\sqrt{k^2 - k_x^2} z} dk_x$$

Useful functions:

$$\text{sinc}(x) = \frac{\sin(x)}{x}$$

$$\text{comb}\left(\frac{x}{a}\right) = |a| \cdot \sum_{-\infty}^{\infty} \delta(x - na)$$

TEM TRANSMISSION LINES

Structures supporting TEM mode:

$$E_z = H_z = 0$$

$$k \leftrightarrow \beta$$

$$\vec{E}(\vec{r}) = \vec{e}(x, y)[V_0^+ e^{-jkz} + V_0^- e^{+jkz}]$$

$$\vec{H}(\vec{r}) = \vec{h}(x, y)[I_0^+ e^{-jkz} + I_0^- e^{+jkz}]$$

$$\vec{h}(x, y) = \frac{Z_c}{\eta} \vec{z} \times \vec{e}(x, y)$$

BRUSTER ANGLE:

The angle for which there is no reflection for TM only:

$$\theta_B = \tan^{-1}\left(\frac{n_2}{n_1}\right) = \tan^{-1}\left(\sqrt{\frac{\epsilon_2}{\epsilon_1}}\right) \Rightarrow r_{21}^{\text{TM}} = 0$$

In general:

$$\theta_B = \sin^{-1}\left[\sqrt{\frac{\mu_2 \epsilon_2 (\eta_2^2 - \eta_1^2)}{\eta_2^2 \mu_1 \epsilon_1 - \eta_1^2 \mu_2 \epsilon_2}}\right]$$

SNELL'S LAW:

$$\frac{\sin \theta_1}{\sin \theta_2} = \frac{\sqrt{\mu_{r2} \epsilon_{r2}}}{\sqrt{\mu_{r1} \epsilon_{r1}}} = \frac{n_2}{n_1}$$

$$n_1 > n_2$$

The critic angle:

$$\sin \theta_{cr} = \frac{n_2}{n_1}$$

For incident angle greater than the critic:
: $\theta_i > \theta_{cr}$:

$$|\Gamma_{TE}| = |\Gamma_{TM}| = 1 = e^{j\phi}$$

$$\cos \theta_2 = -j \sqrt{\frac{\sin^2 \theta_1}{\sin^2 \theta_{cr}} - 1}$$

Plugging in: $\vec{E}_2(z) = \vec{E}_{02} e^{-jk_2(x \sin \theta_t + z \cos \theta_t)}$

leads to propagation in x and decay in z :

$$\vec{E}_2(z) = \vec{E}_{02} e^{-jk_2 x \frac{\sin \theta_1}{\sin \theta_{cr}}} e^{-k_2 z \sqrt{\frac{\sin^2 \theta_1}{\sin^2 \theta_{cr}} - 1}}$$

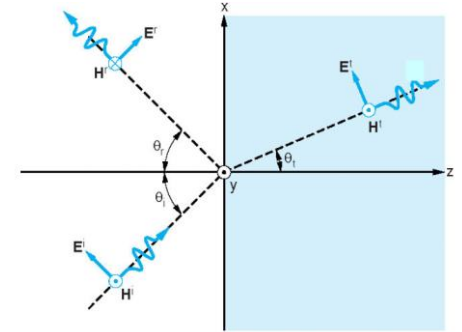
In a full reflection:

$$\Gamma_{TM} = e^{j\phi_{TM}} \quad \phi_{TM} = 2 \tan^{-1} \left(-\frac{\sqrt{\sin^2 \theta_1 - \sin^2 \theta_{cr}}}{\cos \theta_1 \sin^2 \theta_{cr}} \right)$$

$$\Gamma_{TE} = e^{j\phi_{TE}} \quad \phi_{TE} = 2 \tan^{-1} \left(\frac{\sqrt{\sin^2 \theta_1 - \sin^2 \theta_{cr}}}{\cos \theta_1} \right)$$

TM SOLUTION (PARALLEL)

$$H_z = 0$$



$$\vec{k} = k_x \hat{x} + k_z \hat{z} = k(\hat{x} \sin \theta + \hat{z} \cos \theta)$$

$$\hat{e}_{TM} = \hat{x} \cos \theta - \hat{z} \sin \theta$$

$$\hat{h}_{TM} = \hat{y}$$

$$\vec{E}_{TM} = E_{TM} (\hat{x} \cos \theta - \hat{z} \sin \theta) e^{-jk \sin \theta x} e^{-jk \cos \theta z} =$$

$$E_x = e^{-jk \sin \theta x} [V_0^+ e^{-jk \cos \theta z} + V_0^- e^{jk \cos \theta z}],$$

$$V_0^+ = E_{TM} \cos \theta$$

TL MODEL FOR TM POLARIZATION:

$$E_x^+ \leftrightarrow V^+$$

$$H_y^+ \leftrightarrow I^+$$

$$k_z = k \cos \theta \leftrightarrow \beta$$

$$\eta \cos \theta \leftrightarrow Z_{CTM}$$

The longitudinal component:

$$E_z = [-Z_{CTM} \tan \theta] H_y$$

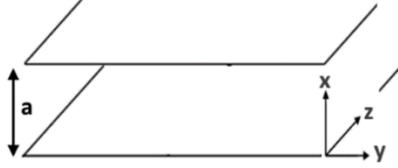
FRESNEL EQUATIONS FOR TM POLARIZATION:

(For the transverse component E_x):

$$r_{21}^{\text{TM}} = \frac{\eta_2 \cos \theta_2 - \eta_1 \cos \theta_1}{\eta_2 \cos \theta_2 + \eta_1 \cos \theta_1} = \frac{n_1 \cos \theta_2 - \frac{\mu_1}{\mu_2} n_2 \cos \theta_1}{n_1 \cos \theta_2 + \frac{\mu_1}{\mu_2} n_2 \cos \theta_1}$$

$$t_{21}^{\text{TM}} = \frac{2\eta_2 \cos \theta_2}{\eta_2 \cos \theta_2 + \eta_1 \cos \theta_1} = \frac{2n_1 \cos \theta_2}{n_1 \cos \theta_2 + \frac{\mu_1}{\mu_2} n_2 \cos \theta_1}$$

$$t_{21}^{\text{TM}} = r_{21}^{\text{TM}} + 1$$

PLATES WAVEGUIDE:*Wave vector:*

$$\beta_m = \sqrt{k^2 - k_{c,m}^2} = \omega \sqrt{\mu\epsilon \left[1 - \left(\frac{f_{c,m}}{f} \right)^2 \right]}$$

$$k_{c,m} = \left(\frac{m\pi}{a} \right)$$

Cut-off frequency:

$$f_{c,m} = \frac{1}{2\sqrt{\mu\epsilon}} \left(\frac{m}{a} \right)$$

TEM MODE:

$$E_x = A e^{-jkz}$$

$$H_y = \frac{E_x}{\eta} e^{-jkz}$$

TE MODES:

$$E_y = C_m \sin\left(\frac{m\pi}{a} x\right) e^{-j\beta_m z}$$

$$H_x = -\frac{\beta_m}{\omega\mu} C_m \sin\left(\frac{m\pi}{a} x\right) e^{-j\beta_m z}$$

$$H_z = j \frac{m\pi}{\omega\mu a} C_m \cos\left(\frac{m\pi}{a} x\right) e^{-j\beta_m z}$$

$$Z_m^{\text{TE}} = \frac{\omega\mu}{\beta_m} = \frac{k\eta}{\beta_m}$$

TM MODES:

$$H_y = C_m \cos\left(\frac{m\pi}{a} x\right) e^{-j\beta_m z}$$

$$E_x = \frac{\beta_m}{\omega\epsilon} C_m \cos\left(\frac{m\pi}{a} x\right) e^{-j\beta_m z}$$

$$E_z = \frac{j m \pi}{\omega \epsilon a} C_m \sin\left(\frac{m\pi}{a} x\right) e^{-j\beta_m z}$$

$$Z_m^{\text{TM}} = \frac{\beta_m}{\omega\epsilon} = \frac{\beta_m \eta}{k}$$

Wavelength:

$$\lambda_{nm} = \frac{2\pi}{\beta_{nm}}$$

TE MODES:

$$H_z = A_{nm} \cos\left(\frac{n\pi x}{a}\right) \cos\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$E_x = \frac{j\omega\mu}{k^2 - \beta_{nm}^2} \frac{m\pi}{b} A_{nm} \cos\left(\frac{n\pi x}{a}\right) \sin\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$E_y = -\frac{j\omega\mu}{k^2 - \beta_{nm}^2} \frac{n\pi}{a} A_{nm} \sin\left(\frac{n\pi x}{a}\right) \cos\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$H_x = \frac{j\beta_{nm}}{k^2 - \beta_{nm}^2} \frac{n\pi}{a} A_{nm} \sin\left(\frac{n\pi x}{a}\right) \cos\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$H_y = \frac{j\beta_{nm}}{k^2 - \beta_{nm}^2} \frac{m\pi}{b} A_{nm} \cos\left(\frac{n\pi x}{a}\right) \sin\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$Z_{nm}^{\text{TE}} = -\frac{E_y}{H_x} = \frac{E_x}{H_y} = \frac{k\eta}{\beta_{nm}} = \frac{\eta}{\sqrt{1 - \left(\frac{f_{c,nm}}{f}\right)^2}}$$

TM MODES:

$$E_z = B_{nm} \sin\left(\frac{n\pi x}{a}\right) \sin\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

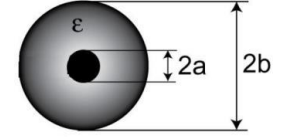
$$H_x = \frac{j\omega\epsilon}{k^2 - \beta_{nm}^2} \frac{m\pi}{b} B_{nm} \sin\left(\frac{n\pi x}{a}\right) \cos\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$H_y = -\frac{j\omega\epsilon}{k^2 - \beta_{nm}^2} \frac{n\pi}{a} B_{nm} \cos\left(\frac{n\pi x}{a}\right) \sin\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$E_x = -\frac{j\beta_{nm}}{k^2 - \beta_{nm}^2} \frac{n\pi}{a} B_{nm} \cos\left(\frac{n\pi x}{a}\right) \sin\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

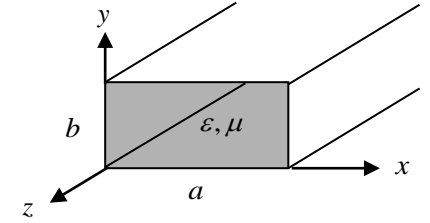
$$E_y = -\frac{j\beta_{nm}}{k^2 - \beta_{nm}^2} \frac{m\pi}{b} B_{nm} \sin\left(\frac{n\pi x}{a}\right) \cos\left(\frac{m\pi y}{b}\right) e^{-j\beta_{nm} z}$$

$$Z_{nm}^{\text{TM}} = -\frac{E_y}{H_x} = \frac{E_x}{H_y} = \frac{\beta_{nm}\eta}{k} = \eta \sqrt{1 - \left(\frac{f_{c,nm}}{f}\right)^2}$$

COAXIAL LINE:

$$\vec{e}(r, \phi) = \hat{r} \frac{1}{\ln(b/a)} \frac{1}{r}, \quad \vec{h}(r, \phi) = \hat{\phi} \frac{1}{2\pi r}$$

$$Z_c = \eta \frac{\ln(b/a)}{2\pi}$$

RECTANGULAR WAVEGUIDE:*Wave vector:*

$$\beta_{nm} = \sqrt{k^2 - k_{c,nm}^2}, \quad k = \omega \sqrt{\mu\epsilon}$$

$$k_{c,nm} = \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2}$$

Cut-off frequency:

$$f_{c,nm} = \frac{1}{2\sqrt{\mu\epsilon}} \sqrt{\left(\frac{n}{a}\right)^2 + \left(\frac{m}{b}\right)^2}$$

Phase velocity:

$$v_p = \frac{\omega}{\beta_{nm}} = \frac{\frac{1}{\sqrt{\mu\epsilon}}}{\sqrt{1 - \left(\frac{f_{c,nm}}{f}\right)^2}} \underset{\substack{\text{propagating} \\ f > f_{c,nm}}}{\geq} \frac{1}{\sqrt{\mu\epsilon}}$$

Group velocity:

$$v_g = \frac{\beta_{nm}}{k} \frac{1}{\sqrt{\mu\epsilon}} \leq \frac{1}{\sqrt{\mu\epsilon}}$$

$$\underline{E} \simeq \hat{\theta} j \omega \mu g(r) I_0 d \sin \theta e^{jk \hat{r} \cdot \underline{r}'}, \underline{H} \simeq \hat{\phi} \eta^{-1} E_{\theta}$$

Radiation Pattern:

$$F(\theta, \phi) = \frac{|E(\theta, \phi)|}{|E|_{\max}}$$

POINTING VECTOR:

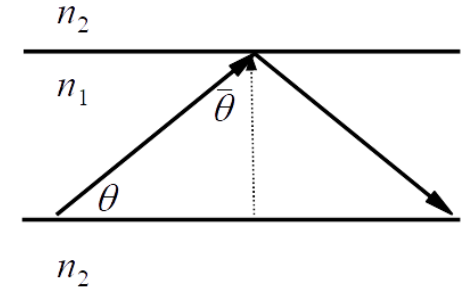
$$\underline{S} = \frac{1}{2} \underline{E} \times \underline{H}^* = \hat{r} \frac{|E|^2}{2\eta}$$

DISPERSION:

$$v_p = \frac{\omega}{\beta(\omega)}$$

$$v_g = \frac{1}{\partial \beta(\omega) / \partial \omega}$$

DIELECTRIC WAVEGUIDE:



$$\bar{\theta}_c = \cos^{-1} \left(\frac{n_2}{n_1} \right)$$

Solutions for TE modes:

$$\tan \left(\frac{\pi d}{\lambda} n_1 \sin \theta - m \frac{\pi}{2} \right) = \sqrt{\frac{\sin^2 \bar{\theta}_c}{\sin^2 \theta} - 1}$$

RADIATION:

$$\underline{J} = \hat{z} I_0 d \delta(\underline{r} - \underline{r}') e^{j\omega t}$$

The dipole location:

$$\underline{r}' = x' \hat{x} + y' \hat{y} + z' \hat{z}$$

The observation point:

$$\begin{aligned} \underline{r} &= x \hat{x} + y \hat{y} + z \hat{z} \\ &= r (\sin \theta \cos \phi \hat{x} + \sin \theta \sin \phi \hat{y} + \cos \theta \hat{z}) \end{aligned}$$

The magnetic vector potential of the dipole:

$$\underline{A} = \hat{z} \mu I_0 d \frac{e^{-jkR}}{4\pi R}, R = |\underline{r} - \underline{r}'|$$

Far field approximation:

$$R = r - \hat{r} \cdot \underline{r}'$$

$$\underline{A} = \hat{z} \mu g(r) I_0 d e^{jk \hat{r} \cdot \underline{r}'}, g(r) = \frac{e^{-jkr}}{4\pi r}$$