Fields 2012 – Solutions

- 1a) Answer two parts out of three.
- i) The characteristic impedance Z_0 is the ratio between voltage and current (for waves travelling in a transmission line) or between electric field and magnetic field (for electromagnetic waves).

[2]

Most candidates answered this correctly.

For example for a transmission line, the governing equations are;

$$dV/dz = -j\omega L_p I$$
 and $dI/dz = -j\omega C_p V$

Hence,
$$d^2V/dz^2 = -i\omega L_p dI/dz = -\omega^2 L_p C_p V$$

Assuming the travelling wave solutions $V(z) = V_0 \exp(-jkz)$ and $I(z) = I_0 \exp(-jkz)$

We get
$$-k^2V = -\omega^2 L_p C_p V$$
, so that $k = \omega (L_p C_p)^{1/2}$

We also get $-jkV_0 = -j\omega L_p I_0$

Hence
$$Z_0 = V_0/I_0 = \omega L_p/k = \sqrt{(L_p/C_p)}$$

Most common mistake: failure to give sufficient detail, e.g derive a formula.

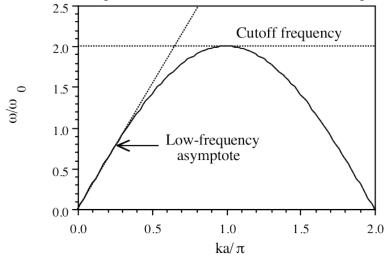
[3]

ii) A dispersion diagram is a plot of angular frequency (ω) against propagation constant (k). It is very useful way of displaying the important features of wave propagation, since it allows the phase velocity $v_p = \omega/k$, the group velocity $v_g = d\omega/dk$ and the frequency range of propagating waves to be read off.

Most candidates answered this correctly.

[2]

For example, the relation between w and k for a low pass ladder network is $\omega/\omega_0 = 2 \sin(ka/2)$. In this case the dispersion characteristic is a sinusoid with peak amplitude $2\omega_0$ as shown below



Most common mistake: failure to illustrate the answer with a dispersion graph.

[3]

iii) The phase velocity v_{ph} gives the speed of a single wave. Unfortunately a single wave cannot carry any information, since it never varies. To send some data, a carrier must be modulated. The envelope (which contains the information) then travels at a slightly different speed, the group velocity v_g (which represents the velocity of a group of waves).

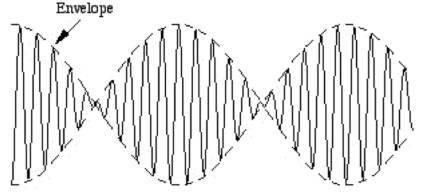
Most candidates answered this correctly.

[2]

To calculate v_g , consider the simplest possible AM signal, formed by beating together two signals of different angular frequencies $\omega + d\omega$ and ω - $d\omega$. The corresponding k-values at these frequencies are k+dk and k - dk. For equal amplitudes, the combined voltage is:

$$V = V_0 \left[\exp \left\{ j((\omega + d\omega)t - (k + dk)z) \right\} + \exp \left\{ j((\omega - d\omega)t - (k - dk)z) \right\} \right]$$

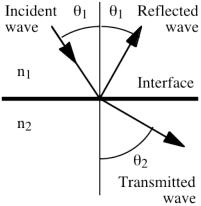
This result can be written alternatively as $V = 2V_0 \exp\{j\omega \ t - kz\}\} \cos\{d\omega \ t - dk \ z\}$. Hence, the wave is an amplitude-modulated carrier as shown below. The velocity of the carrier is $v_p = \omega/k$ as before. However, the velocity of the envelope is the group velocity $v_g = d\omega/dk$.



Most common mistake: failure to illustrate the answer with mathematics and/or a diagram showing the information-carrying envelope.

[3]

b) Snell's law $n_1 \sin(\theta_1) = n_2 \sin(\theta_2)$ relates the angles of incidence and refraction when an electromagnetic wave strikes an interface between two dielectric media with refractive indices n_1 and n_2 as shown below.



Most candidates got the formula correct. Most common mistake: failure to include a diagram.

[2]

At a critical angle $\theta_1 = \theta_C = \sin^{-1}(n_2/n_1)$, we obtain $\theta_2 = \pi/2$. In this case, the transmitted wave is parallel to the interface, and there is no real solution for θ_2 for $\theta_1 > \theta_c$. Under these conditions,

total internal reflection occurs and no energy can cross the interface. Note however that we must have $n_2 < n_1$ for total internal reflection, so that incidence must be from the high-index side.

Most candidates answered this correctly.

[2]

Assuming that $n_1 = 1.5$ (glass) and $n_2 = 1$ (air) we obtain $\theta_C = \sin^{-1}(1/1.5) = 41.81^\circ$.

Most candidates answered this correctly.

[1]

c) The electric field due to one plane wave travelling at an angle $+\theta$ to the z-axis is:

 $E_y = E_{y0} \left\{ \exp[-jk_0(z\cos(\theta) + x\sin(\theta))] \right\}$

Here $k_0 = 2\pi/\lambda$

Most candidates answered this correctly.

[1]

For two waves travelling at angles $\pm \theta$ to the z-axis, the combined field can be written as the sum: $E_v = E_{v0} \{ \exp[-jk_0(z\cos(\theta) + x\sin(\theta))] + \exp[-jk_0(z\cos(\theta) - x\sin(\theta))] \}$

Most candidates answered this correctly.

[1]

This field can be written alternatively as

 $E_v = 2E_{v0} \exp[-jk_0z \cos(\theta)] \cos[k_0x \sin(\theta)]j$

Most common mistake: failure to re-arrange the plane wave expression correctly.

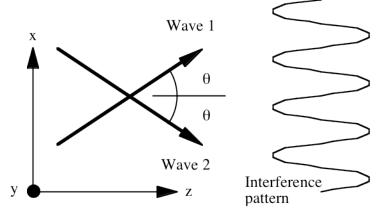
[1]

The time averaged power density is $\underline{S} = 1/2 (\underline{E} \times \underline{H}^*)$

In this case, $S_z = -1/2 E_v H_x *$

Since H_x is proportional to E_y , $S_z = C \cos^2[k_0 x \sin(\theta)]$

This pattern varies as cos squared in the x-direction, as shown in the RH figure.



Most common mistake: failure to draw a picture.

[2]

2a) A driven antenna will generate a characteristic pattern of time-averaged electric and magnetic field \underline{E} and \underline{H} . These fields in turn will give rise to a power flow $\underline{S} = 1/2$ Re($\underline{E} \times \underline{H}^*$). In the far field, \underline{S} is radial, so $\underline{S} = S \ \underline{r}$. At a given radius, S is a function only of the angular coordinates θ and φ , and is known as a radiation pattern. The normalised radiation pattern F is the ratio of S to its maximum value. The directivity D is then the ratio of the maximum of F to its average over 4π space.

Most common mistake: failure to explain directivity correctly, even though the expression is in the formula sheet.

[3]

The effective area A_e is the equivalent area from which an antenna can gather power, and deliver it to matched load.

Most candidates answered this correctly.

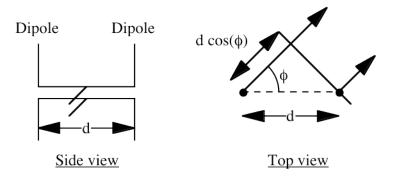
[2]

b) Assuming an average power P, the power density at radius r is $S = P/4\pi r^2$ Assuming an effective area A, the received power is $P_R = AS = P \ A/4\pi r^2$ If $P = 20 \ x \ 10^3 \ W$, $R = 10 \ x \ 10^3 \ m$ and $A = 10 \ m^2$, then: $P_R = 10^4 \ x \ 10 \ /(4 \ x \ \pi \ x \ 10^8) \ W = 7.96 \ x \ 10^{-5} \ W = 79.6 \ \mu W$

Most candidates answered this correctly.

[5]

c) The broadside antenna consists of two dipoles arranged thus:



Most candidates answered this correctly.

[5]

d) In the far field, suppose each antenna generates an electric field $E_{\omega}(\phi)$. A receiver will detect contributions from one, with phases that depend on the distance travelled. Signals from the LH dipole travel a distance $d \cos(\phi)$ further than those from the RH dipole.

Ignoring the common path, the combined field $E_{\iota}'(\phi)$ is:

$$E_{\text{\tiny e}}'(\phi) = E_{\text{\tiny e}}(\phi) \{1 + exp[-jk_0d \cos(\phi)]\} \text{ where } k_0 = 2\pi/\lambda.$$

[3]

If the antennae are individually isotropic, $E_{\theta}(\phi) = 1$. We then obtain:

 $E_{\text{\tiny o}}'(\phi) = \exp[-jk_0 d \, \cos(\phi)/2] \, \left\{ \exp[+jk_0 d \, \cos(\phi)/2] + \exp[-jk_0 d \, \cos(\phi)/2] \right\}, \, \text{or} \,$

 $E_{\text{\tiny o}}'(\varphi) = 2E_{\text{\tiny o}}(\varphi) \; exp[\text{-j}k_0 d \; cos(\varphi)/2] \; cos[k_0 d \; cos(\varphi)/2]$

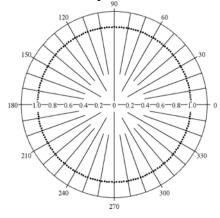
The normalised radiation pattern is $F'(\phi) = |E_{\omega}'(\phi)|^2$ divided by its maximum value In this case, $F'(\phi) = \cos^2[\cos(\phi) k_0 d/2] = \cos^2[\cos(\phi) \pi d/\lambda]$

Most common mistake: failure to work out $|E_{\bullet}'(\phi)|^2$ correctly.

[4]

e) If the frequency is f=1.5 MHz, the wavelength is $\lambda=c/f=3$ x 10^8 / (1.5 x $10^6)=200$ m If d=1 m, then $d/\lambda=0.005$ and $\pi d/\lambda=0.005$ π

Whatever the value of ϕ , $\cos(\phi) \pi d/\lambda$ is then very small, so $\cos^2[\cos(\phi) \pi d/\lambda] \approx 1$ The radiation pattern is therefore approximately isotropic, as shown below



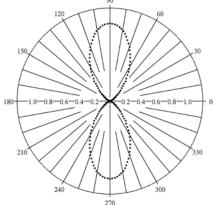
[4]

If the frequency is f = 150 MHz, λ = 3 x 10⁸ / (150 x 10⁶) = 2 m If d = 1 m, then d/λ = 0.5 and $\pi d/\lambda$ = $\pi/2$

The normalised radiation pattern is then $F'(\phi) = \cos^2[\cos(\phi) \pi/2]$

When $\phi = 0$, F'(ϕ) = $\cos^2(\pi/2) = 0$; when $\phi = \pi/2$, F'(ϕ) = $\cos^2(0) = 1$ etc

Hence the normalised radiation pattern is as shown below.



[4]

Most common mistake: most candidates who failed to work out $|E_{\bullet}'(\varphi)|^2$ correctly could not draw the radiation patterns, even though they are from the lecture notes.