

# CHANNEL CODING - FINAL PROJECT

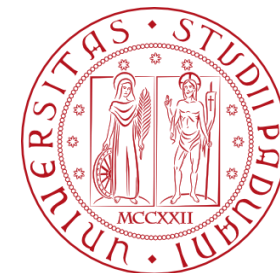
## UNIFIED HIGH-SPEED WIRELINE-BASE HOME NETWORKING TRANSCEIVERS

Access networks – In premises networks

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UNIVERSITÀ  
DEGLI STUDI  
DI PADOVA

- Standard generality
- Encoder
- Decoder
- Rationale Decoder
- BICM
- Results

- The **G.hn ITU G9960** is a standard proposed by the ITU-T.
- **Designed for:** the transmission of data over premises' wiring.
- The standard **defines:**
  1. the home network architecture and reference models.
  2. the physical layer specification.



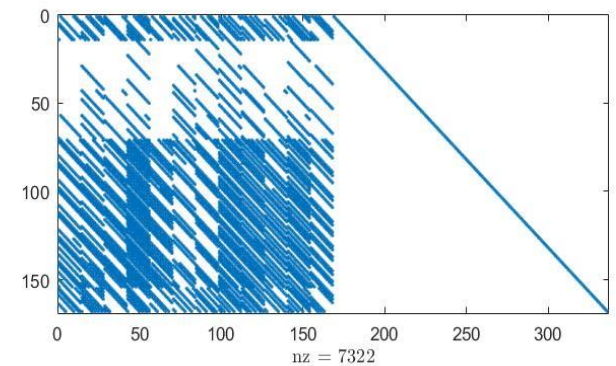
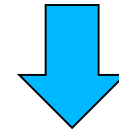
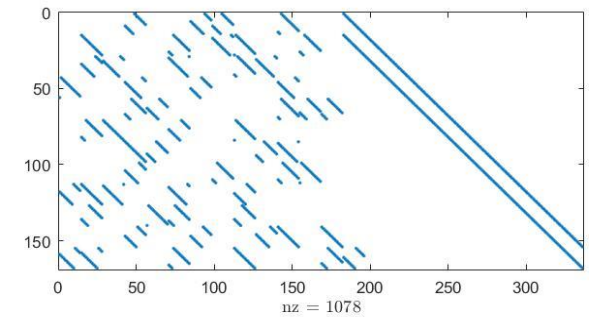
- Each node of the home-network has a different **profile**.
- The standard refers to two different profile:
  1. **Low-complexity profile** (L-CP).
  2. **Standard profile** (SP).
- A node of the network is required to support one profile, at minimum.

Profile name	Domain type	Valid bandplans
L-CP	Power-line baseband	25 MHz
SP	Power-line baseband	50 MHz, 100 MHz
	Telephone-line baseband	50 MHz, 100 MHz
	Coax baseband	50 MHz, 100 MHz
	Coax RF	50 MHz, 100 MHz, 200 MHz

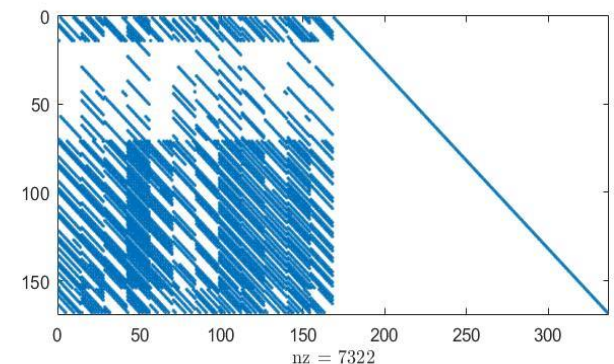
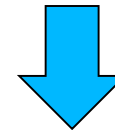
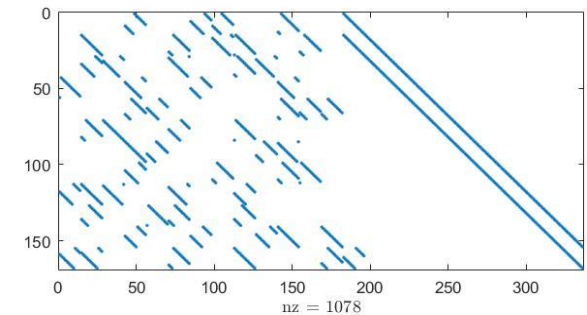
- The parity check matrix  $H$  of size  $(N-K) \times N$  will have:
  - ❑ Rate  $1/2$  :  $N = 1920$  ( $LCP, SP$ ), 8640 ( $SP$ ).
  - ❑ Rate  $2/3$  :  $N = 1440, 6480$  ( $SP$ )
  - ❑ Rate  $5/6$  :  $N = 1152, 5184$  ( $SP$ )
- Rate  $16/18$  ,  $20/21$  are obtained puncturing the code with rate  $5/6$ , through different puncturing patterns.

Profile name	FEC rate	FEC block size
L-CP	$1/2$	120 bytes (Payload)
SP	$1/2, 2/3, 5/6$ $16/18, 20/21$	120 and 540 bytes (Payload)

1. Given  $H$ , get its systematic form  $H_{\text{sys}}$ .



1. Given  $H$ , get its systematic form  $H_{\text{sys}}$ .
  2. Encoding Procedure for a single word:
    - ✓ The  $K$  information bits of  $u$  are directly copied to the codeword  $v$ .
    - ✓ Let  $A$  be the submatrix obtained considering  $N-K$  rows and the first  $k$  columns of  $H_{\text{sys}}$ ; then, the  $N-K$  parity-check bits are computed as  $A * u$ .
    - ✓ The final codeword is  $v = [u | A * u]$
- Preprocessing is done once.



# Decoder – a first (naïf) approach



- All the minsum/ sumproduct algorithm written in c, and invoking via mex functions.
- At least three matrixes as big as the  $H$  matrix:



# Decoder – a first (naïf) approach



- All the minsum/ sumproduct algorithm were written in c, and invoked via mex functions.
- At least three matrixes as big as the  $H$  matrix

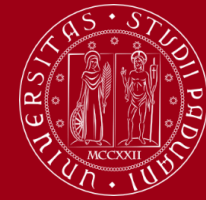


Unfeasible in terms of:

- Memory allocation.
- Computational efforts.



# Decoder – a first (naïf) approach



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# Decoder – a first (naïf) approach



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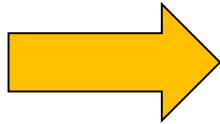
Unfeasible in terms of:

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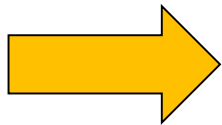
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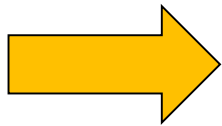


Do only once.

- From  $H$  we'll get:

$$\Psi_{v \rightarrow c} \quad \Psi_{c \rightarrow v} \quad \prod_{v \rightarrow c} \quad \prod_{c \rightarrow v} \quad \{l_i\}_{i=1}^{N-K} \quad \{j_i\}_{i=1}^N$$

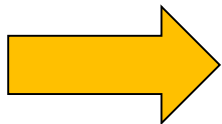
- Preprocess the matrix  $H$  in such a way to get all the patterns from it.



Do only once.

- From  $H$  we'll get:

$$\Psi_{v \rightarrow c} \quad \Psi_{c \rightarrow v} \quad \prod_{v \rightarrow c} \quad \prod_{c \rightarrow v} \quad \{l_i\}_{i=1}^{N-K} \quad \{j_i\}_{i=1}^N$$



No need to evaluate directly  $H$  in the message passing for the computation of LLRs.



- Let be  $\Psi_{c \rightarrow v}$  and  $\Psi_{v \rightarrow c}$  the check update matrix and the variable update matrix, respectively:

$$\Psi_{c \rightarrow v} = \begin{bmatrix} I_{l_1}^C & & 0 \\ & \ddots & \\ 0 & & I_{l_{N-K}}^C \end{bmatrix}$$

$$\Psi_{v \rightarrow c} = \begin{bmatrix} I_{j_1}^C & & 0 \\ & \ddots & \\ 0 & & I_{j_N}^C \end{bmatrix}$$

- $I_s^C$  denotes the complement of the Identity matrix of size  $s$ .
- $\{l_i\}_{i=1}^{N-K}$  and  $\{j_i\}_{i=1}^N$  are the number of ones in each row and column respectively.



- Let  $LLR_{c \rightarrow v}, LLR_{v \rightarrow c} \in \mathbb{F}_2^U$  with  $U$  the total numbers of ones (edges) in the H matrix.

$$LLR_{c \rightarrow v} = \begin{bmatrix} \text{---} \\ \vdots \\ \text{---} \end{bmatrix} \begin{matrix} \} j_1 \\ \\ \} j_N \end{matrix}$$

$$LLR_{v \rightarrow c} = \begin{bmatrix} \text{---} \\ \vdots \\ \text{---} \end{bmatrix} \begin{matrix} \} l_1 \\ \\ \} l_{N-K} \end{matrix}$$

- For both minsum and sumproduct decoder the **variable updates** will have the following expression:

$$LLR_{v \rightarrow c} = \Psi_{v \rightarrow c} \cdot LLR_{c \rightarrow v} + p$$

- In the sum product the **check updates** is:

$$LLR_{c \rightarrow v} = \left( \Psi_{c \rightarrow v} \cdot \Phi \left( |LLR_{v \rightarrow c}| \right) \right) * \text{sgn} \left( LLR_{v \rightarrow c} \right)$$

- Where  $\Phi(x) = \log \left( \frac{e^x + 1}{e^x - 1} \right)$ ,  $\text{sgn}(\cdot)$  is the function that maps each element of the vector in its corresponding sign and  $p$  is the vector of prior LLRs.

In order to compute properly the updates of the LLRs, vectors must be properly ordered...

## HOW?

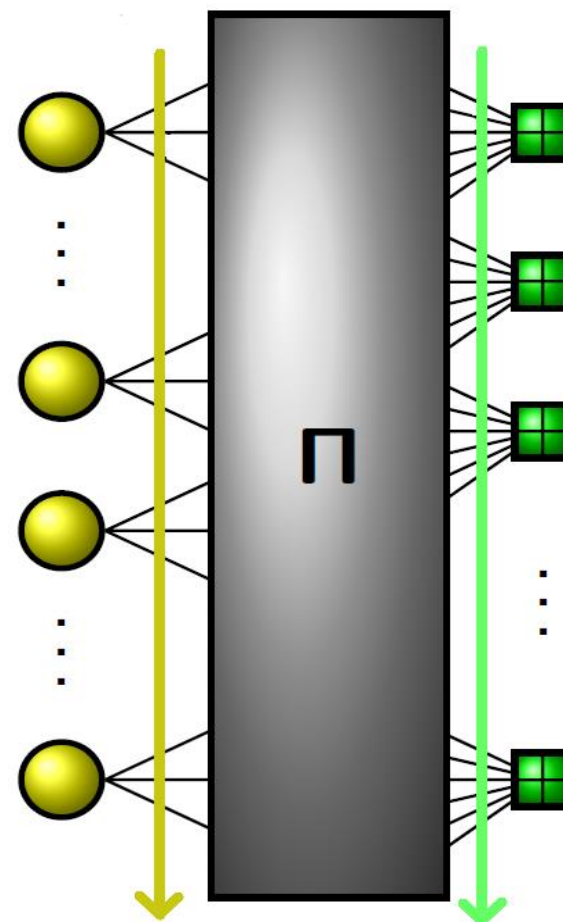
In order to compute properly the updates of the LLRs, vectors must be properly ordered...

## HOW?

We will refer to two different LLR representation:

**Check representation (CR)**

**Variable representation (VR)**

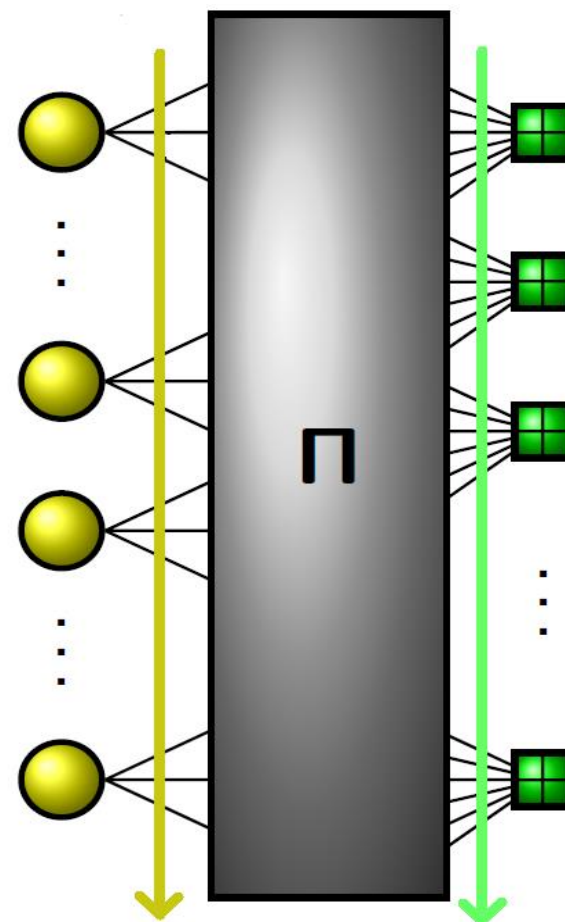


## CHECK REPRESENTATION (CR)

$$H = \begin{bmatrix} 1 & 0 & 1 & 1 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 1 & 1 & 1 & 0 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 & 1 & 1 & 1 & 0 & 1 \\ 1 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 1 & 0 \\ 1 & 0 & 1 & 0 & 0 & 1 & 1 & 1 & 1 & 0 \end{bmatrix}$$

## VARIABLE REPRESENTATION (VR)

$$H = \begin{bmatrix} 1 & 0 & 1 & 1 & 0 & 0 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 0 & 1 & 0 & 0 & 1 \\ 0 & 1 & 0 & 1 & 0 & 1 & 0 & 0 & 1 \\ 1 & 0 & 1 & 0 & 0 & 1 & 1 & 1 & 0 \\ 1 & 0 & 1 & 0 & 0 & 1 & 1 & 1 & 0 \end{bmatrix}$$



- Let be  $per : D_U \longrightarrow D_U$  the permutation map, that associate for each edge in **CR** the corresponding position in the **VR**.
- $D_U = \{1, \dots, U\}$ ,  $U$  the number of ones (edges) in the matrix  $H$ .

$$\prod_{c \rightarrow v} = \left[ \begin{array}{c} e_{per(1)}^\top \\ \hline \vdots \\ \hline e_{per(U)}^\top \end{array} \right]$$

- Where  $e_i \in \mathbb{F}_2^U$  denotes the  $i$ -th canonical vector in  $\mathbb{F}_2^U$ .

- If we think of  $\prod_{c \rightarrow v}$  as a linear application and we exploit the property of the permutation matrixes:

$$\prod_{v \rightarrow c} = \prod_{c \rightarrow v}^{-1} = \prod_{c \rightarrow v}^{\top}$$

- Hence, there is no need to compute directly  $\prod_{v \rightarrow c}$  since:

$$\prod_{c \rightarrow v} \longleftrightarrow \prod_{v \rightarrow c}$$

```
for it=1:iter
    LLRvc = perMatrVC * LLRvc;
    curSign = sign(LLRvc);
    curSign(curSign == 0) = 1;
    signLLRcv = GetSign(curSign,nOR);
    tmpP = abs(LLRvc);
    tmpP(tmpP < minf) = 1e-10;
    temp = upMatrCV * PhiMap(tmpP);
    LLRcv = PhiMap(temp).* signLLRcv;
    LLRvc = perMatrCV * LLRcv;
    LLRvc = upMatrVC * LLRvc + LLRprior;
    curIndex = 1;
    for j=1:length(nOC)
        if(vPLLRs(j) + sum(LLRcv(curIndex:nOC(j)+curIndex-1)) < 0)
            u_hat(j) = 1;
        else
            u_hat(j) = 0;
        end
        curIndex = curIndex + nOC(j);
    end
    if( mod(Hreal * u_hat,2) == 0)
        break;
    end
end
```

← Get LLRvc in **CR**

← Update

← Get LLRcv in **VR**

← Update



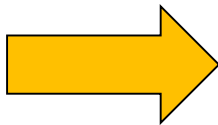
- Product between sparse matrixes is optimized in Matlab.  
(<https://it.mathworks.com/help/matlab/math/sparse-matrix-operations.html> )
- Only vectors proportional to the number of ones in the check matrix are used.

**Mex function** for specific tasks have been implemented:

- Arrayfunc command of matlab leads to slow computation.
- Mex function **PhiMap()** improves dramatically the performances (Make a number).
- Search in the minsum is not bottle neck since elements to be checked are few (consequence to the fact that  $H$  is sparse).

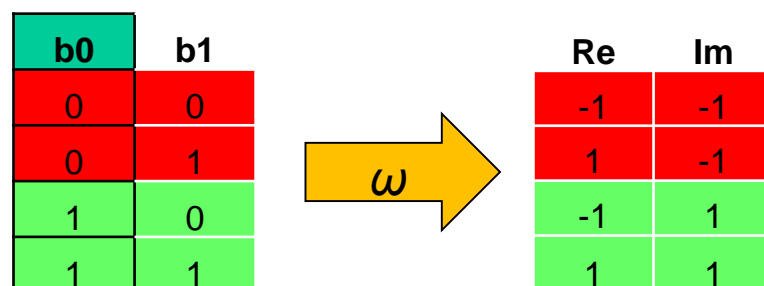
## ENCODING

- Grouping sequences of  $b = \log_2 M$  bits and then mapping according to the suitable constellation mappers.
- Both **constellation mappers** (4 and 16 symbols) realized with mex function.



Faster than define a matlab function and then using `arrayfunc`.

## COMPUTATION OF LLRs FOR DECODER

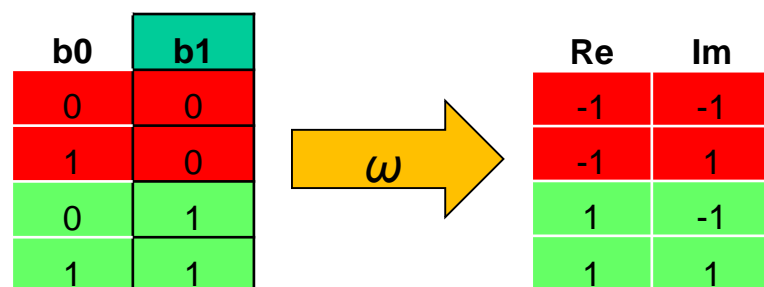


Given the  
*rx* symbol



$$LLR = \ln \left( \frac{P(\mathbf{0}, \mathbf{0}) + P(\mathbf{0}, \mathbf{1})}{P(\mathbf{1}, \mathbf{0}) + P(\mathbf{1}, \mathbf{1})} \right)$$

$b_0 \rightarrow v_0$



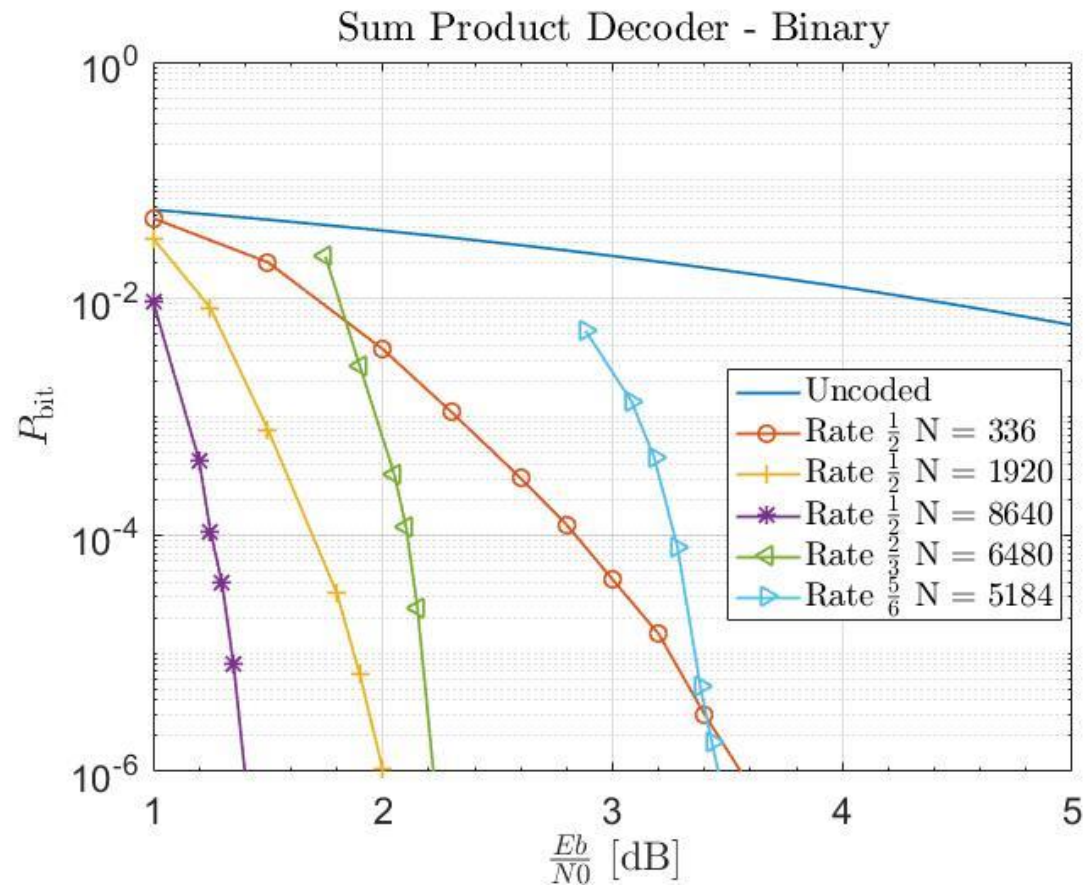
Given the  
*rx* symbol



$$LLR = \ln \left( \frac{P(\mathbf{0}, \mathbf{0}) + P(\mathbf{1}, \mathbf{0})}{P(\mathbf{0}, \mathbf{1}) + P(\mathbf{1}, \mathbf{1})} \right)$$

$b_1 \rightarrow v_1$



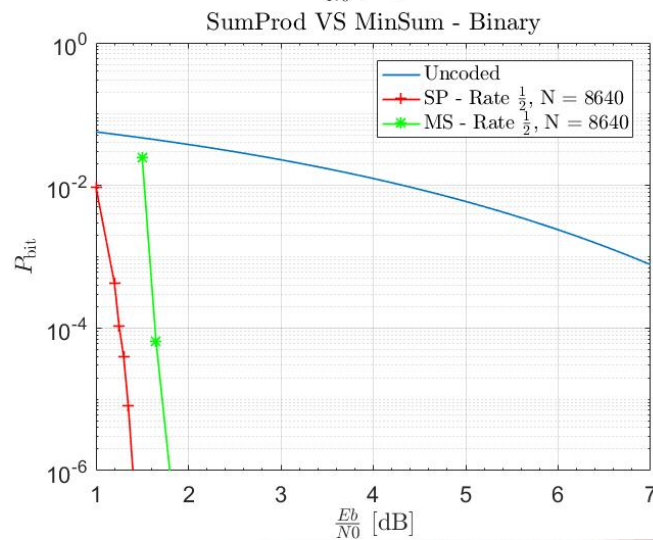
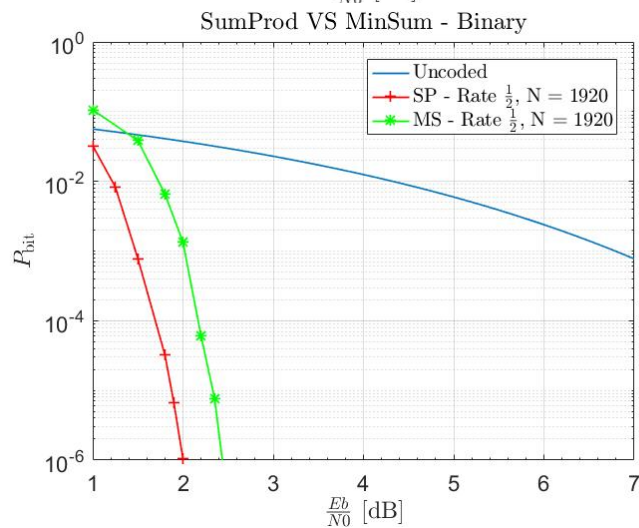
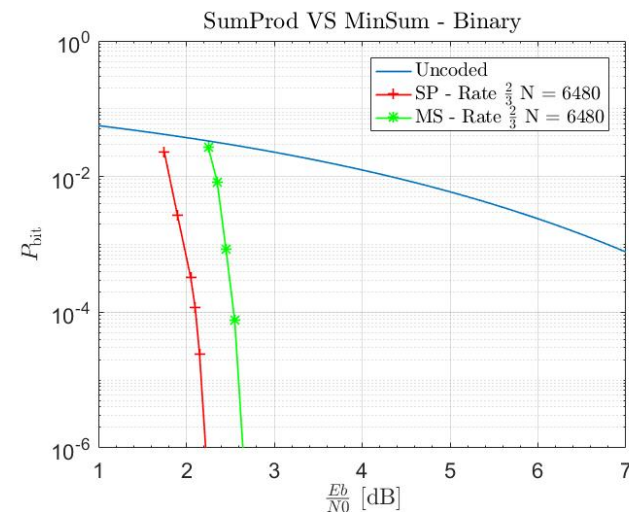
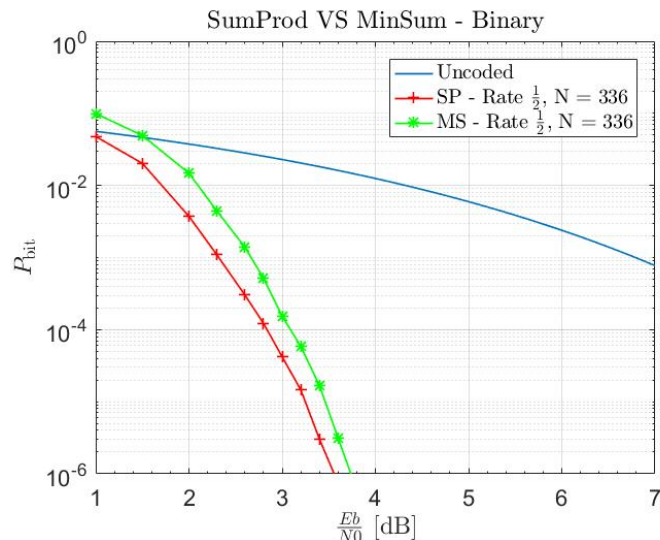


**Uncoded case:**

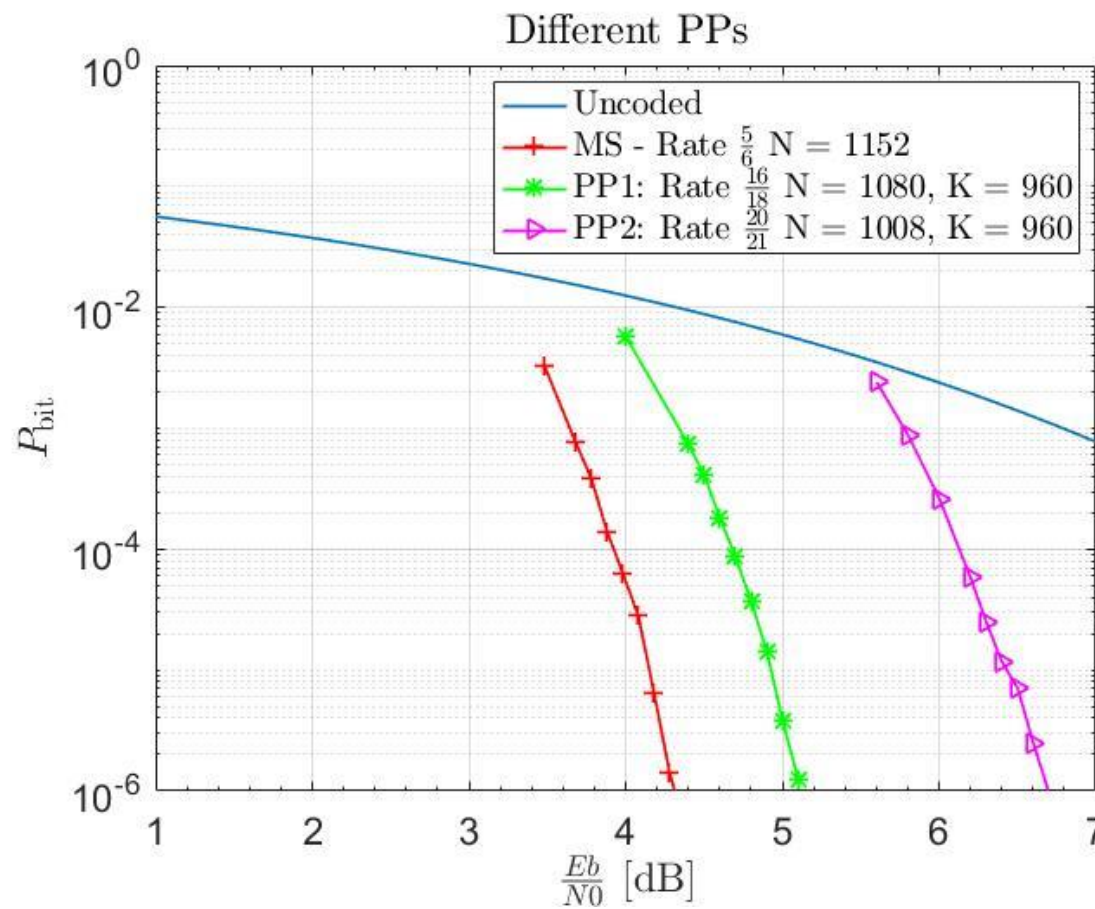
$P_{\text{bit}} = 10^{-6}$

for  $\frac{E_b}{N_0} \cong 10,5 \text{ dB}$

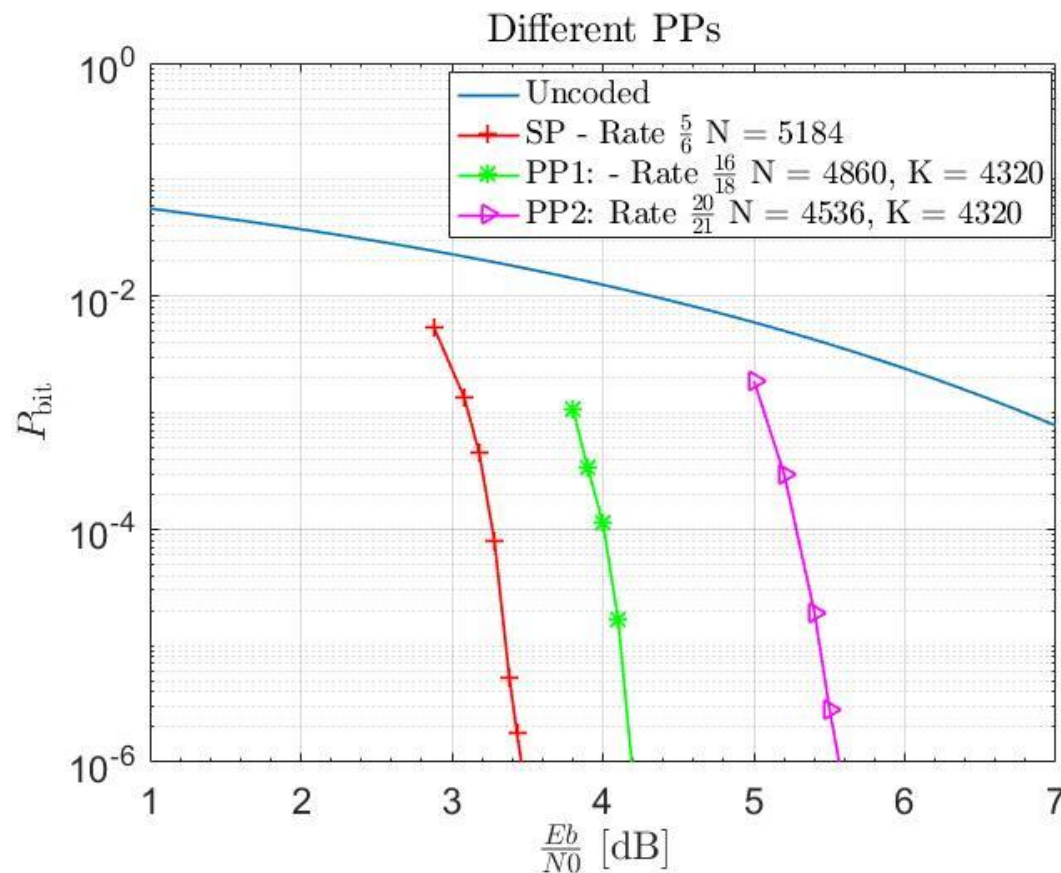
# Results – MinSum vs SumProd



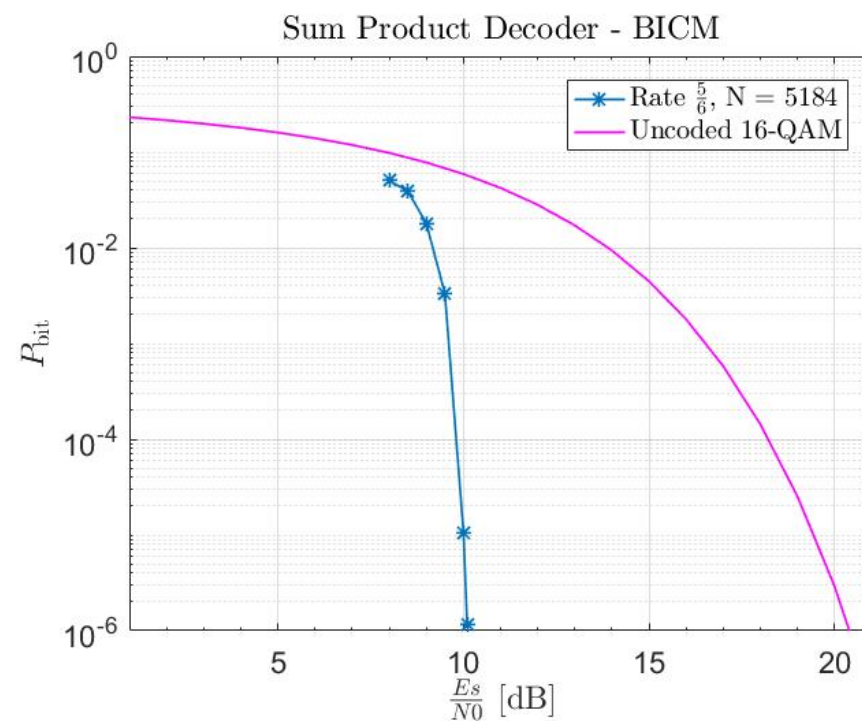
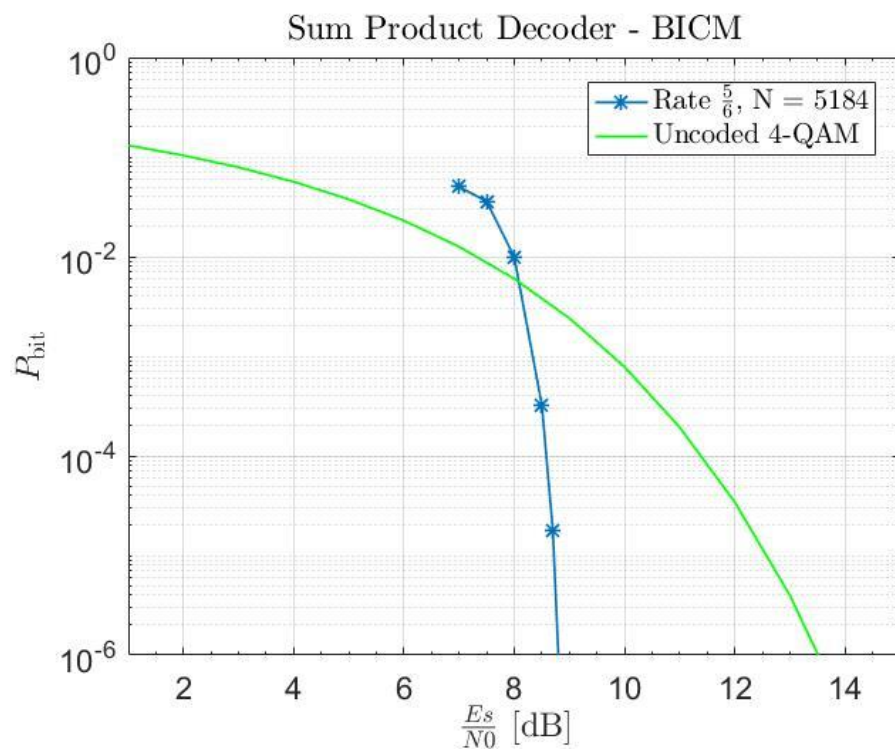
For  $K = 960$ :



For  $K = 4320$ :









- Possible extensions/improvements.

## Details

- All the images has been created with LaTeX or Matlab.
- The images reported in pag. 8 where taken from two superb books (According to Goodreads): The Pragmatic Programmer and Programming Pearls.
- The images of LDPC decoder have been select from PDFs of Channel Coding Course (Tomaso Erseghe).