



# 雷达信号处理及MATLAB仿真

## 第四章 杂波抑制与动目标检测

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2023年3月20日

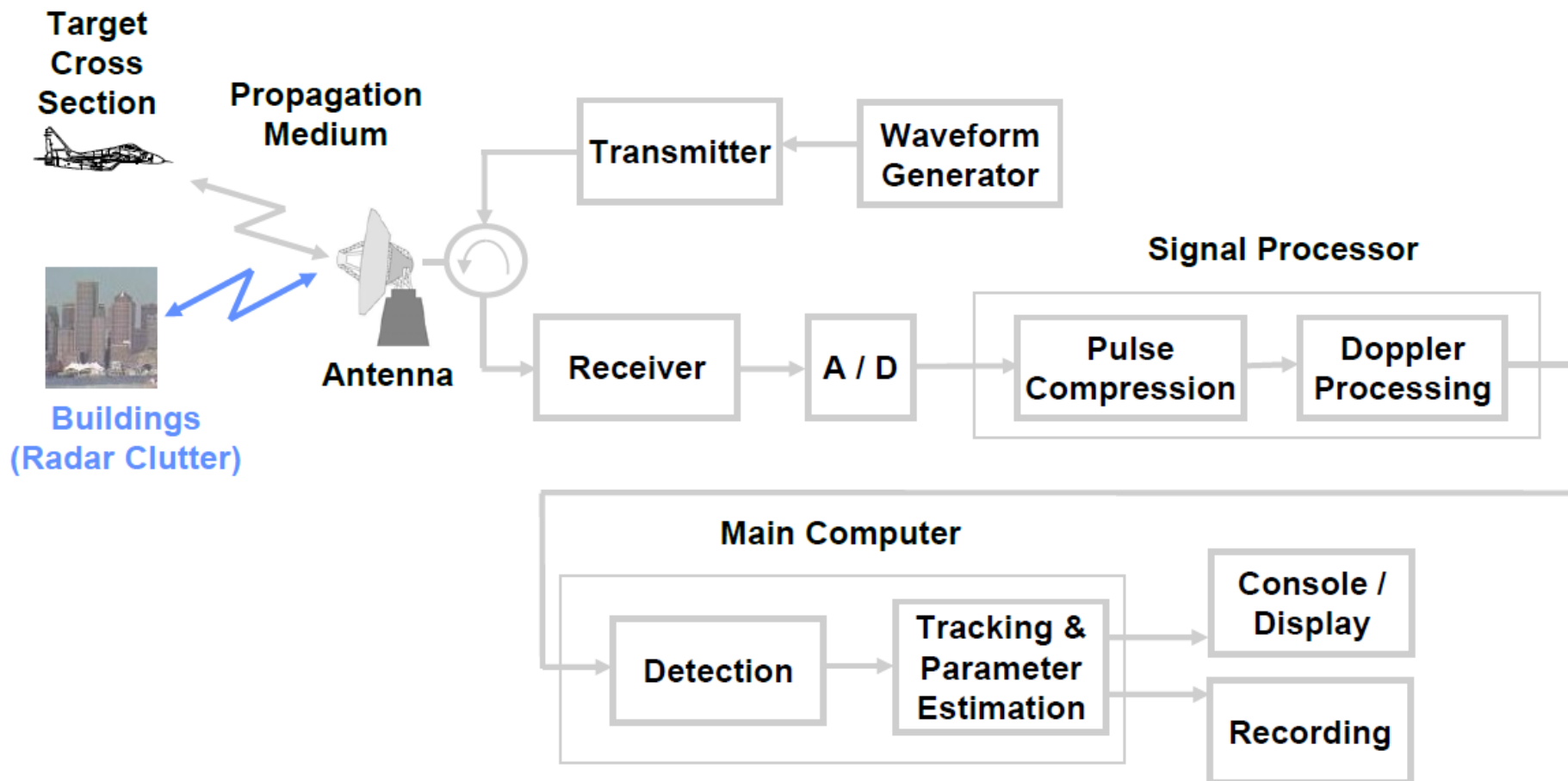


# 杂波抑制与动目标检测

- 杂波的定义
- 杂波抑制技术
- 动目标检测
- 本章小结



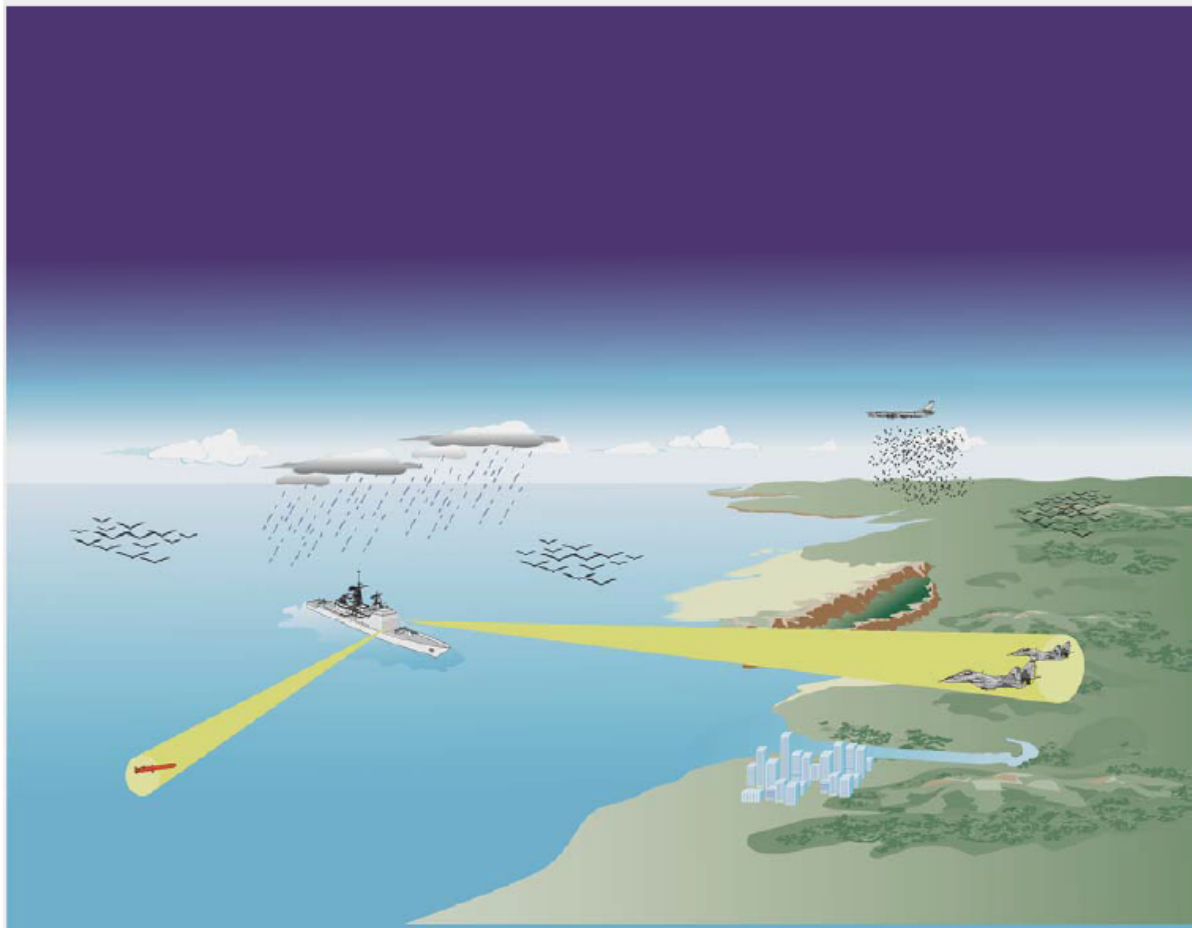
## 一、杂波的定义





## 一、杂波的定义

### Naval Air Defense Scenario



Radar echo is composed of:

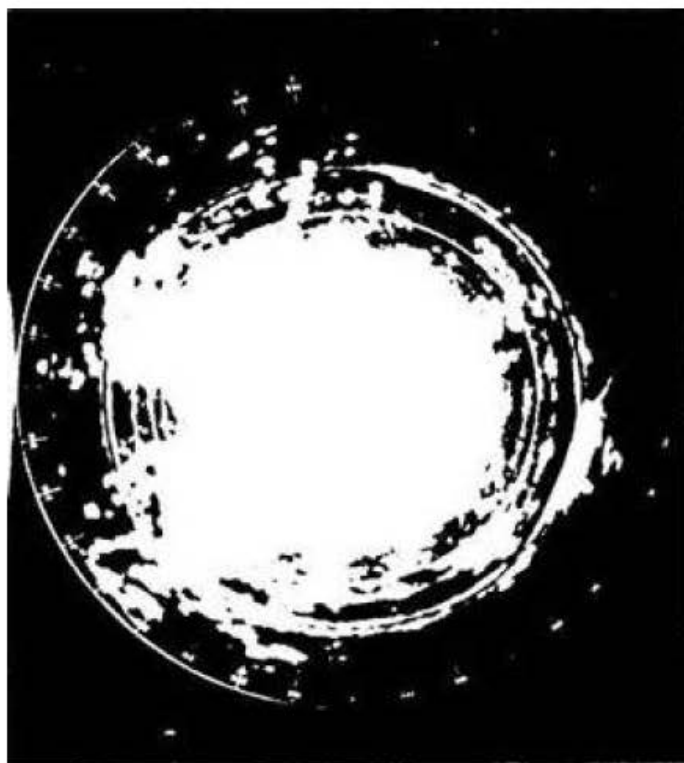
- Backscatter from target of interest
- Receiver noise
- Atmospheric noise
- Interference
  - From other radars
  - Jammers
- Backscatter from unwanted objects
  - Ground
  - Sea
  - Rain
  - Chaff
  - Birds
  - Ground traffic



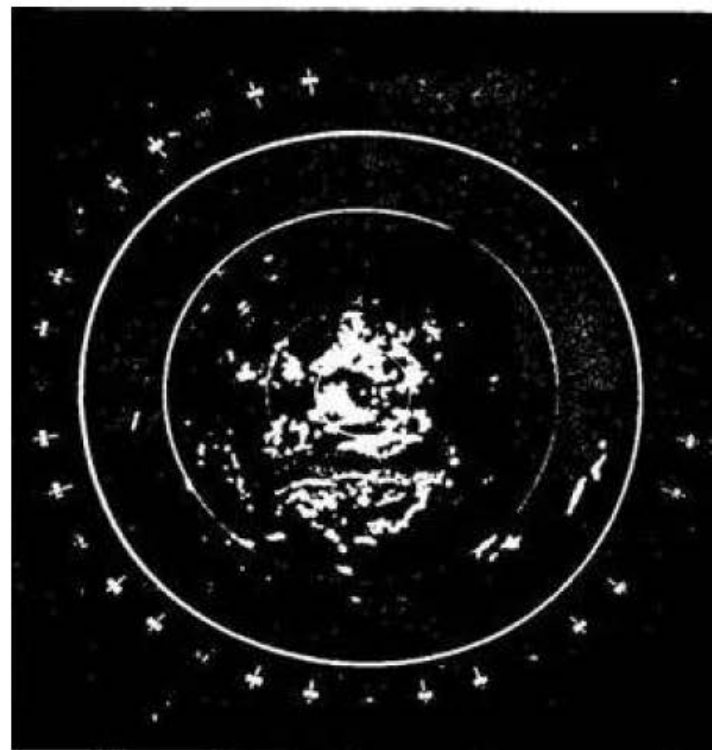
## 一、杂波的定义 - 地杂波

地面的山丘、树林、农田、沙漠、城市建筑等

Mountainous Region of Lakehead, Ontario, Canada  
PPI Set for 30 nmi.



0 dB

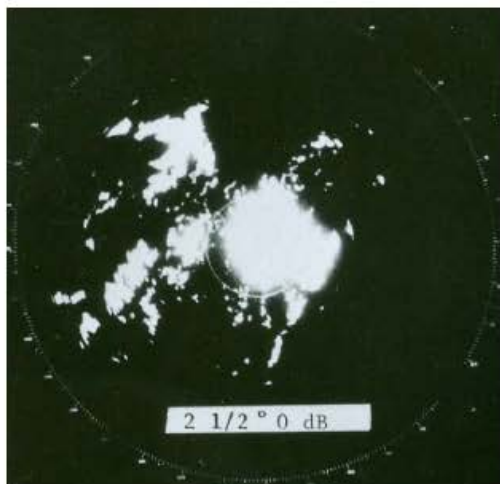


60 dB

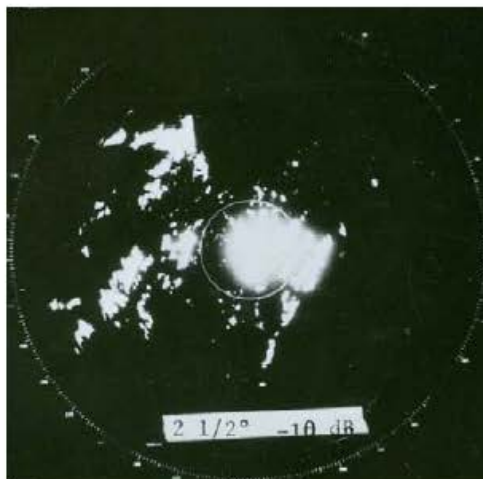


## 一、杂波的定义

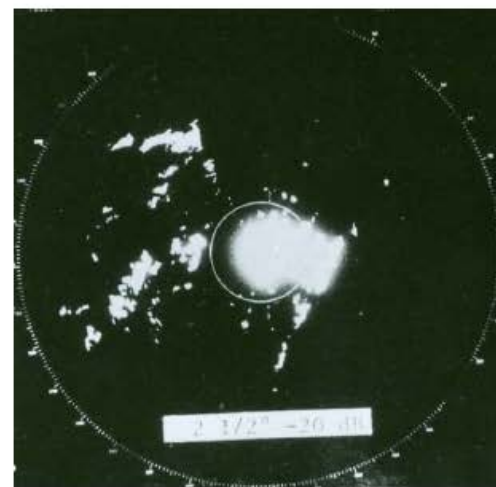
0 dB



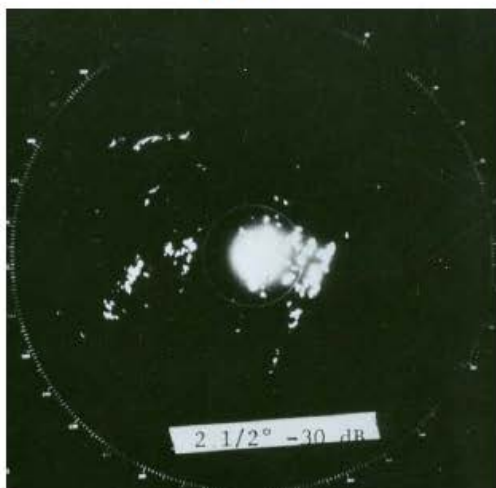
10 dB



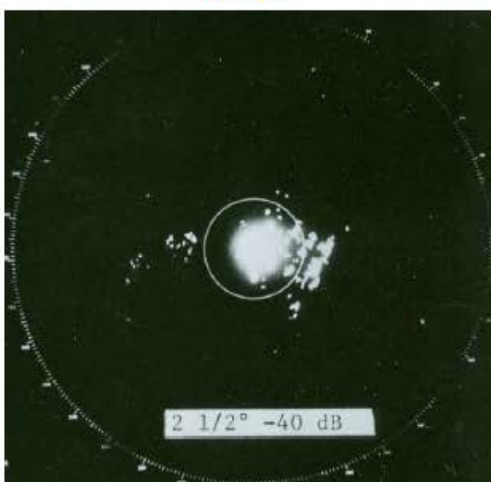
20 dB



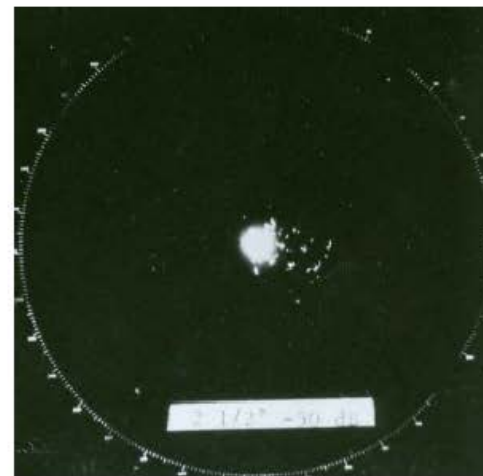
30 dB



40 dB



50 dB



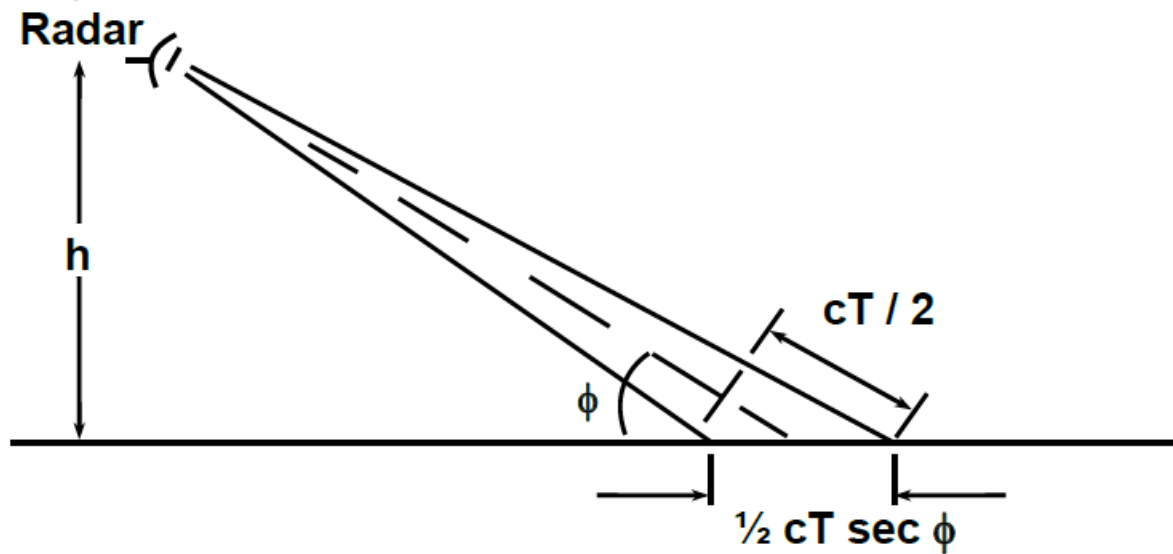




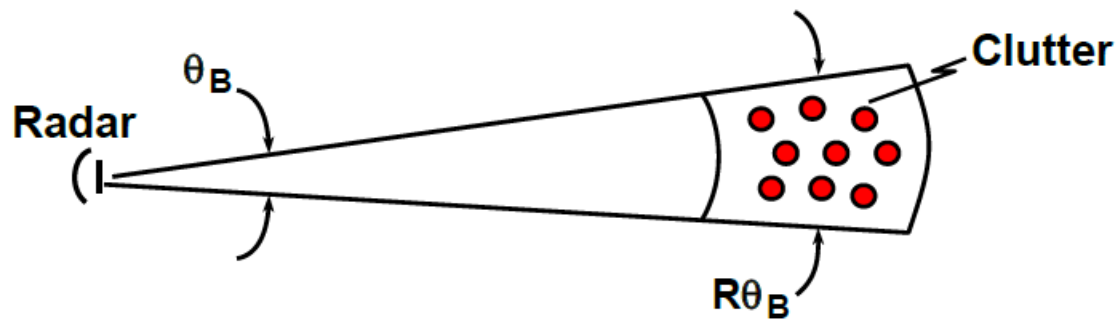
## 一、杂波的定义

### 地杂波几何结构

Elevation View



Plan View



$$\sigma_0 = \frac{\sigma}{A}$$

$$A = R \theta_B [\frac{1}{2} cT \sec \phi]$$



## 一、杂波的定义

- Typical Value of  $\sigma_0 = -20 \text{ dB} = \frac{0.01 \text{ m}^2}{\text{m}^2}$

- $\sigma_{\text{Clutter}} = \sigma_0 A = \sigma_0 \frac{c T}{2} R \theta_B$

– For ASR-9 (Airport Surveillance Radar)

$$\frac{c T}{2} = 100 \text{ m}$$

$$R = 60 \text{ km}$$

$$\theta_B = 1.5^\circ = 0.026 \text{ radians}$$

- $\sigma_{\text{Clutter}} = \frac{0.01 \text{ m}^2}{\text{m}^2} \times 100 \text{ m} \times 60,000 \text{ m} \times 0.026 \text{ radians} = 1500 \text{ m}^2$

For  $\sigma_{\text{Target}} = 1 \text{ m}^2$   $\Rightarrow \frac{\sigma_{\text{Target}}}{\sigma_{\text{Clutter}}} = \frac{1}{1500} \Rightarrow \frac{\sigma_{\text{Target}}}{\sigma_{\text{Clutter}}} = 20$

Small  
single-engine  
aircraft

$\therefore$  Must suppress clutter by a factor of  
 $1500 \times 20 = 30,000 = 45 \text{ dB}$

For good  
detection

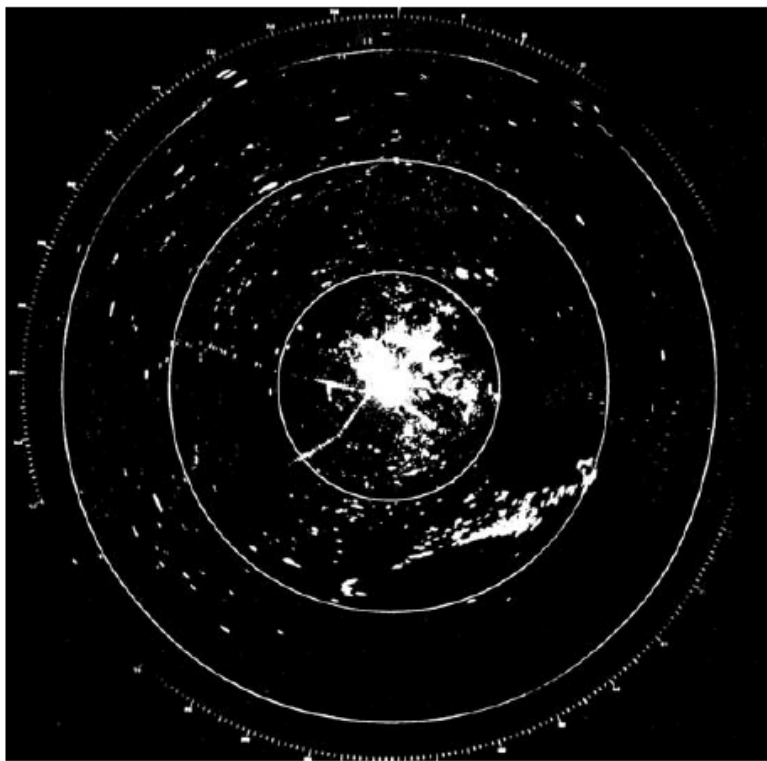




## 一、杂波的定义

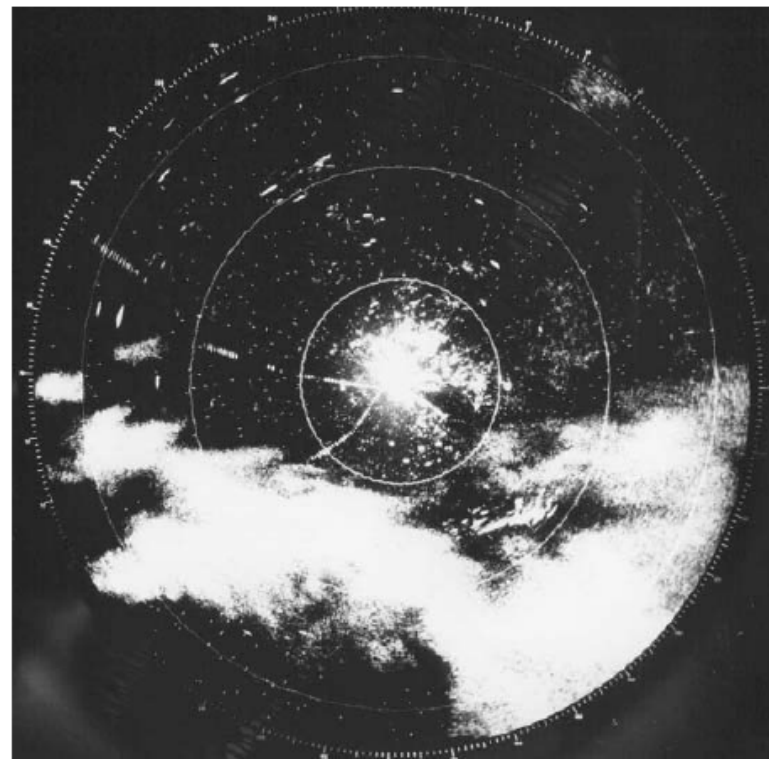
### 云雨杂波

**Clear Day (No Rain)**



**Airport Surveillance Radar  
S Band  
Detection Range - 60 nmi on  
a 1 m<sup>2</sup> target**

**Day of Heavy Rain**



**10 nmi Range Rings on PPI  
Display  
August 1975, FAA Test  
Center  
Atlantic City, New Jersey**



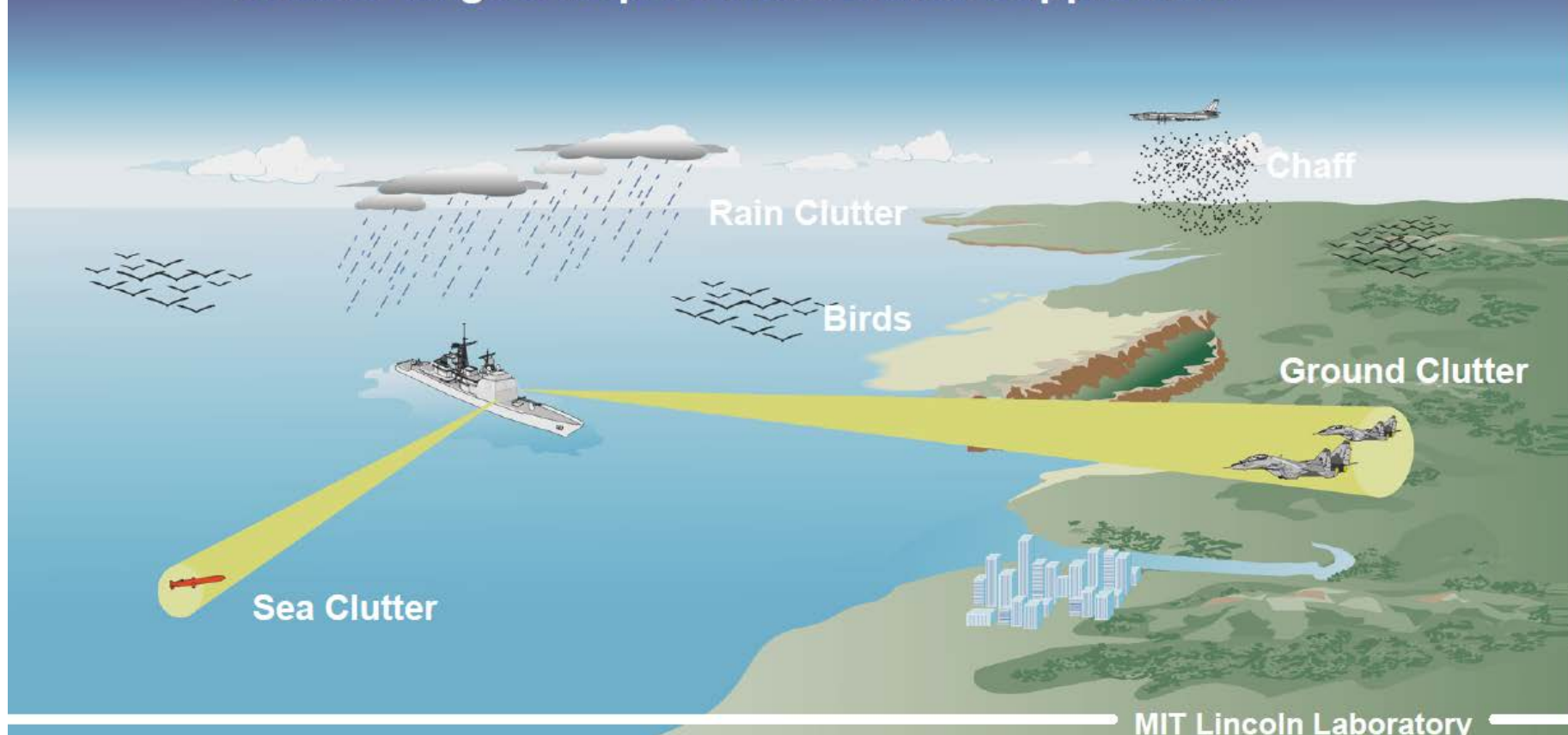
# 杂波抑制与动目标检测

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## 二、杂波抑制

- Moving Target Indicator (MTI) and Pulse-Doppler (PD) processing use Doppler to reject clutter and enhance detection of moving targets
- Smaller targets require more clutter suppression





## 二、杂波抑制

### Moving Target Indicator (MTI) Techniques

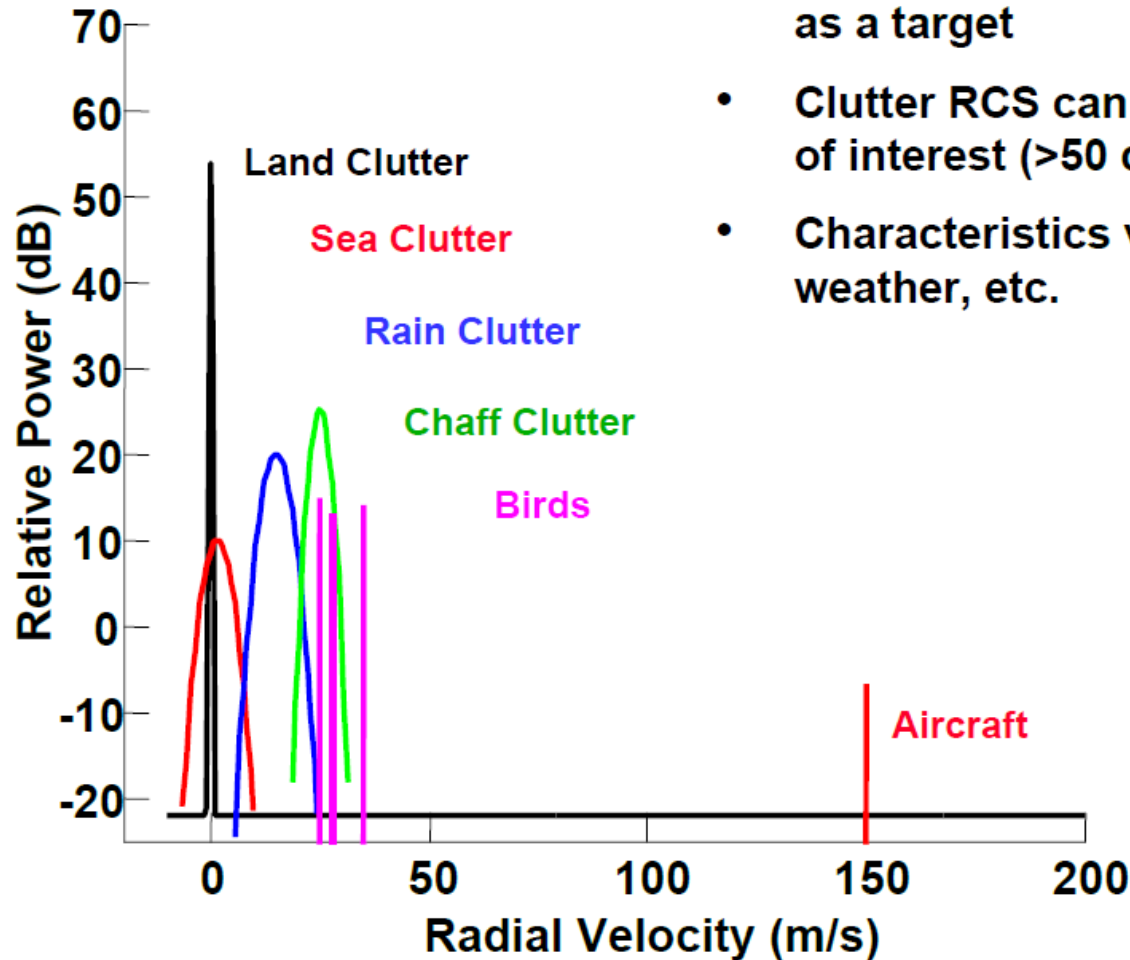
- Just separate moving targets from clutter
- Use short waveforms (two or three pulses)
- Do not provide target velocity estimation

### Pulsed Doppler (PD) Techniques

- Separate targets into different velocity regimes in addition to canceling clutter
- Provide good estimates of target velocity
- Use long waveforms -- (many pulses, tens to thousands of pulses)

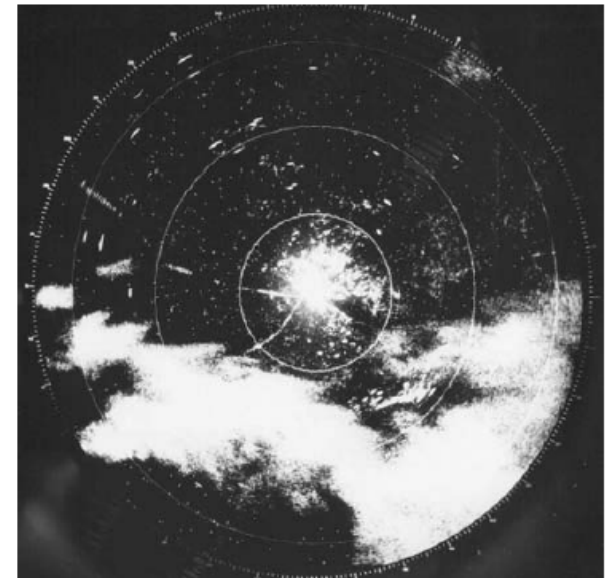


## 二、杂波抑制



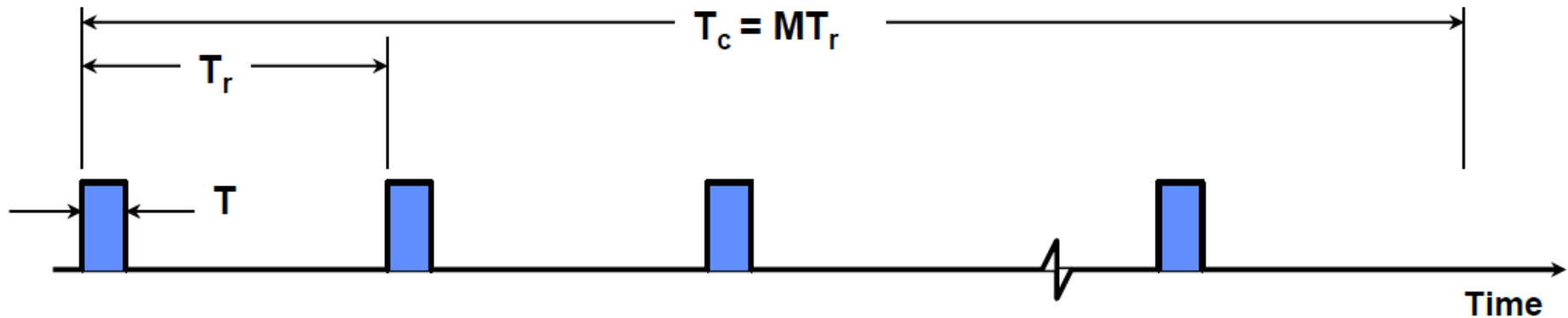
- Clutter comes from same range/angle cell as a target
- Clutter RCS can be much larger than targets of interest ( $>50$  dB)
- Characteristics vary with terrain (land/sea), weather, etc.

PPI Display of Heavy Rain





## 二、杂波抑制

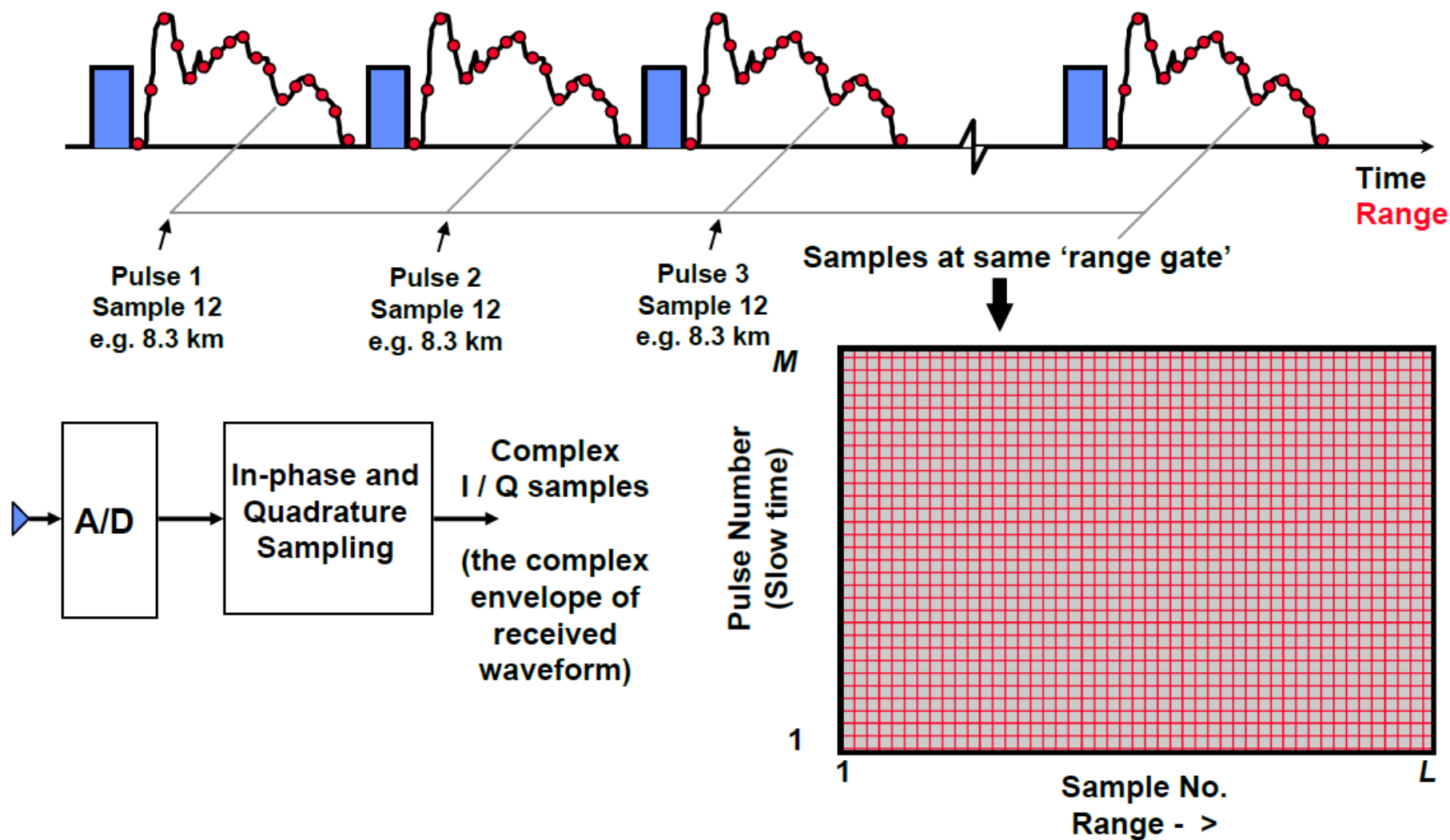


$T$	$=$	Pulse length
$B$	$= 1/T$	Bandwidth
$T_r$	$=$	Pulse repetition interval (PRI)
$f_r$	$= 1/T_r$	Pulse repetition frequency (PRF)
$\delta$	$= T/T_r$	Duty Factor
$T_c$	$= MT_r$	Coherent processing interval (CPI)
$M$	$=$	Number of pulses in the CPI
		$M = 2, 3$ , or sometimes 4 for MTI
		$M$ usually much greater for Pulse Doppler





## 二、杂波抑制



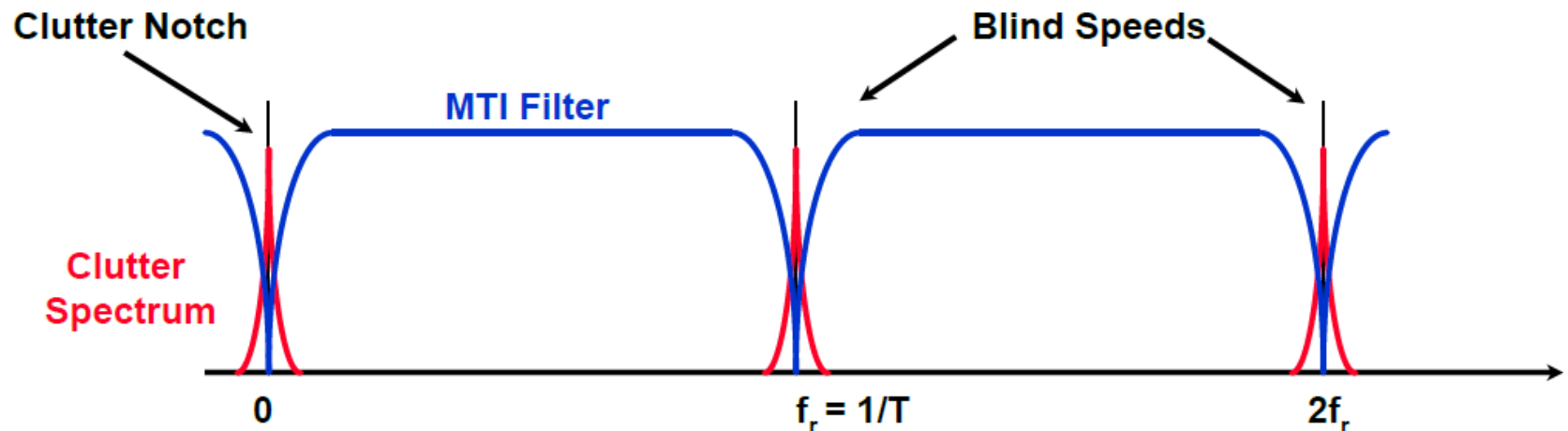




## 二、杂波抑制 - Moving Target Indicator (MTI)

动目标显示指利用杂波抑制滤波器抑制各种杂波，提高雷达的信杂噪比 (SCNR)，以利于运动目标检测的技术。

- Notch out Doppler spectrum occupied by clutter
- Provide broad Doppler passband everywhere else
- Blind speeds occur at multiples of the pulse repetition frequency
  - When sample frequency (PRF) equals a multiple of the Doppler frequency

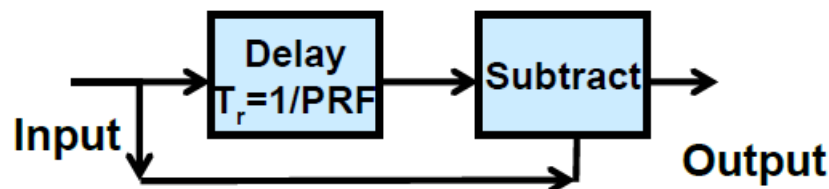




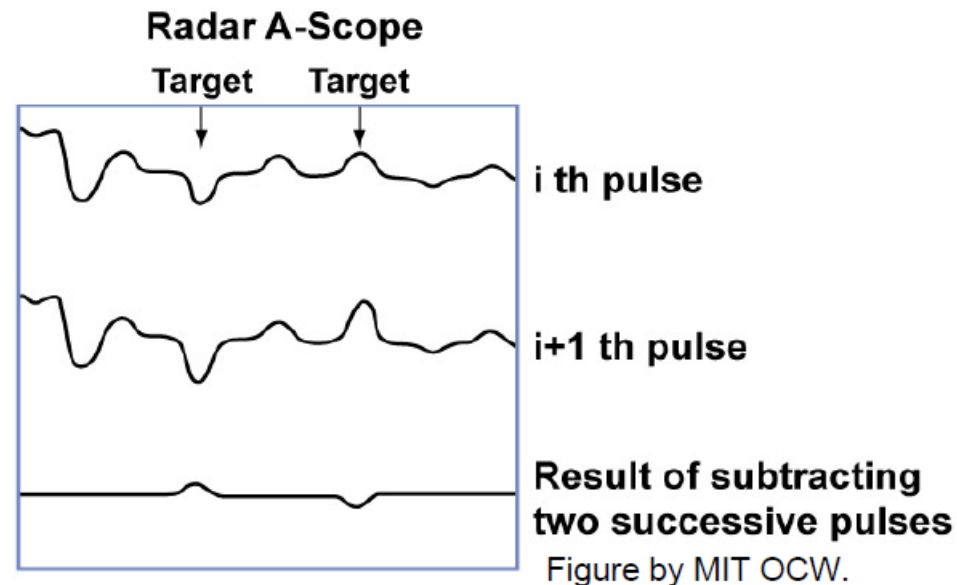
## 二、杂波抑制 - Moving Target Indicator (MTI)

- Fixed Clutter echoes
  - If one pulse is subtracted from the previous pulse, fixed clutter echoes will cancel and will not be detected
- Moving targets
  - Moving targets change in amplitude from one pulse to the next because of their Doppler frequency shift.
  - If one pulse is subtracted from the other, the result will be an uncanceled residue

### Block Diagram



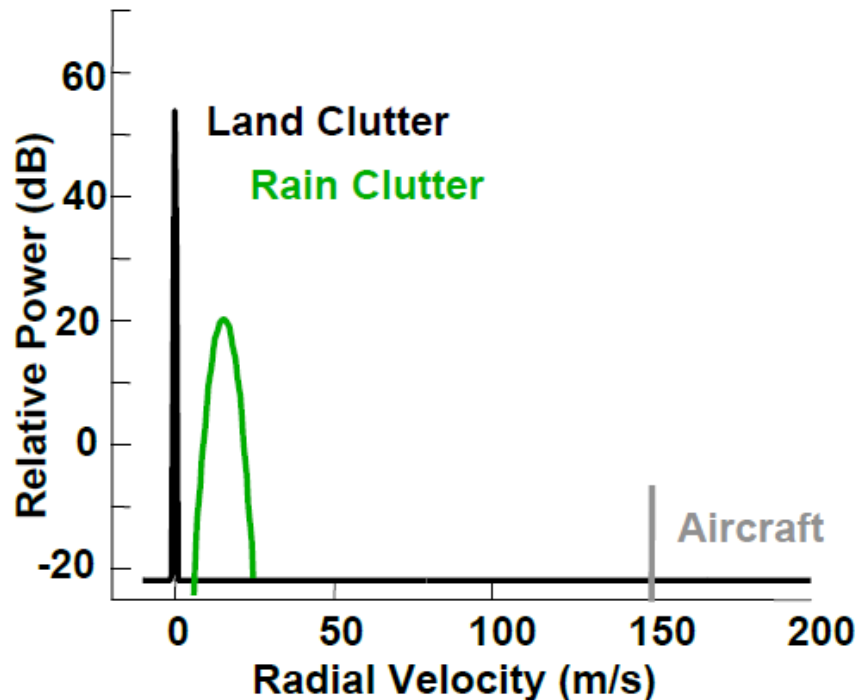
$$V_{\text{output}} = V_{i+1} - V_i$$





## 二、杂波抑制 - 评估指标 改善因子

- $S_{in}$  and  $C_{in}$  - Input target and clutter power per pulse
- $S_{out}(f_d)$  and  $C_{out}(f_d)$  - Output target and clutter power from processor at Doppler frequency,  $f_d$
- MTI Improvement Factor =  $I(f_d) = \frac{(\text{Signal} / \text{Clutter})_{out}}{(\text{Signal} / \text{Clutter})_{in}} \Big|_{f_d}$



MTI Improvement Factor

$$I(f_d) = \frac{C_{in}}{C_{out}} \times \frac{S_{out}}{S_{in}} \Big|_{f_d}$$

Clutter Attenuation

Signal Gain



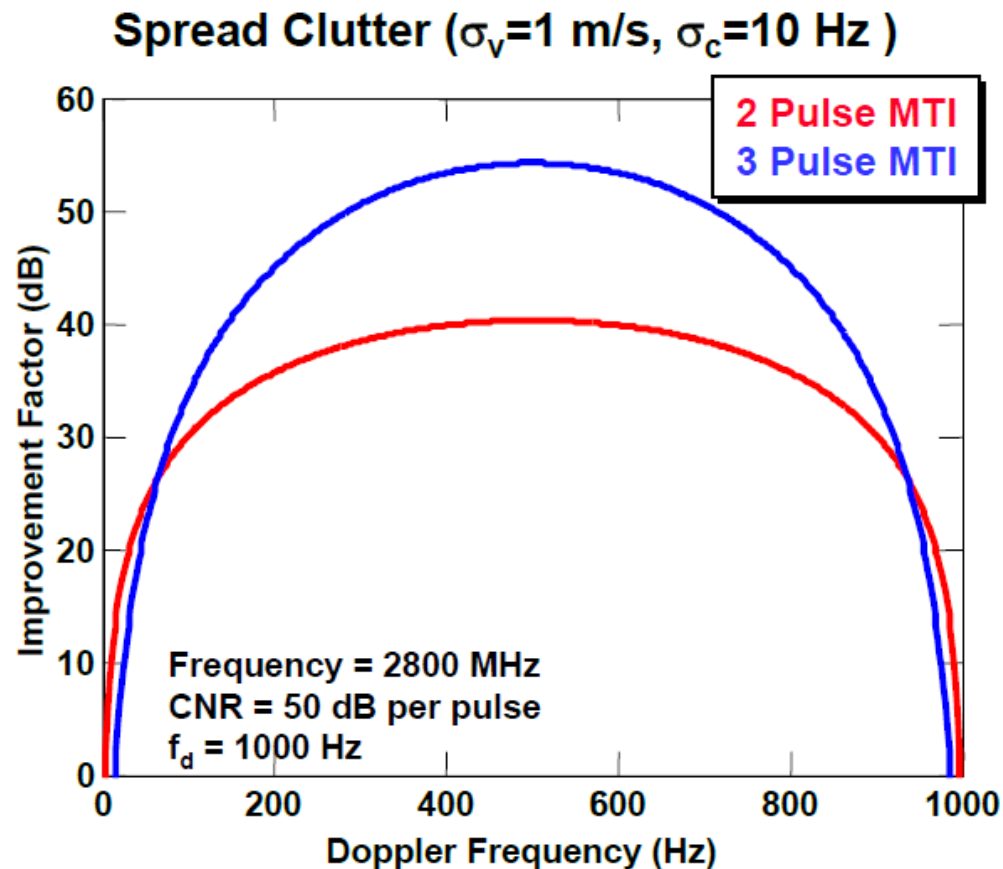
## 二、杂波抑制

### 2-Pulse MTI

$$V_{\text{output}} = V_i - V_{i-1}$$

### 3-Pulse MTI

$$V_{\text{output}} = V_i - 2V_{i-1} + V_{i-2}$$

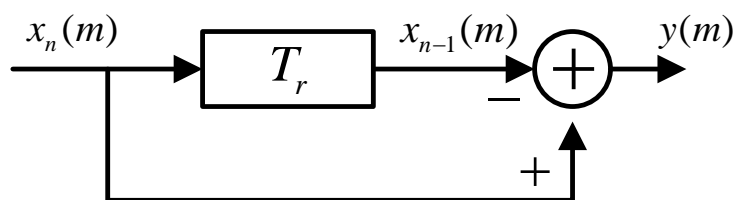


Three-pulse canceller provides wider clutter notch and greater clutter attenuation



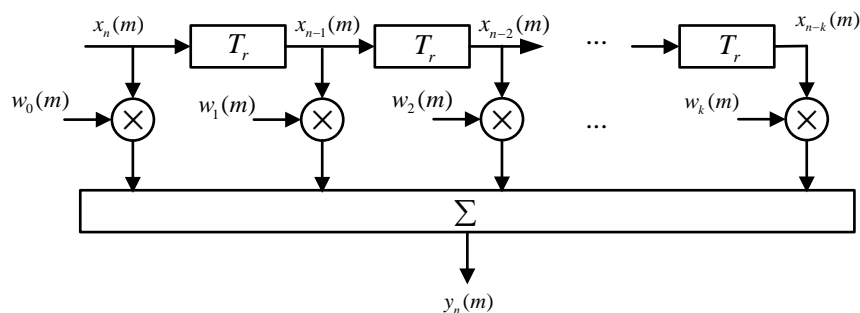
## 二、杂波抑制 - Moving Target Indicator (MTI)

### 一次对消器



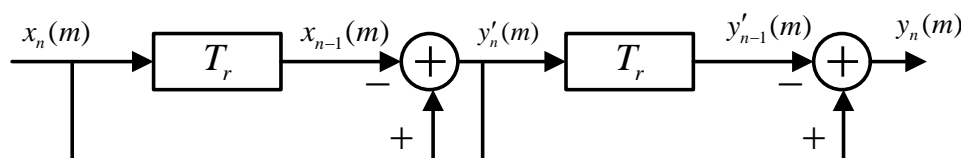
$$y_n(m) = x_n(m) - x_{n-1}(m)$$

### K次对消器

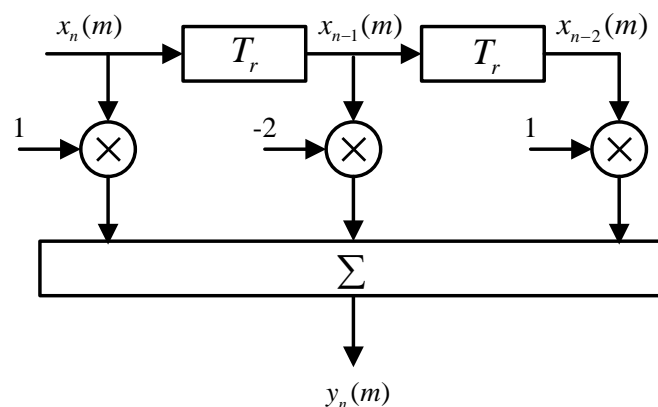


$$y_n(m) = \sum_{i=0}^k w_i x_{n-i}(m)$$

### 二次对消器



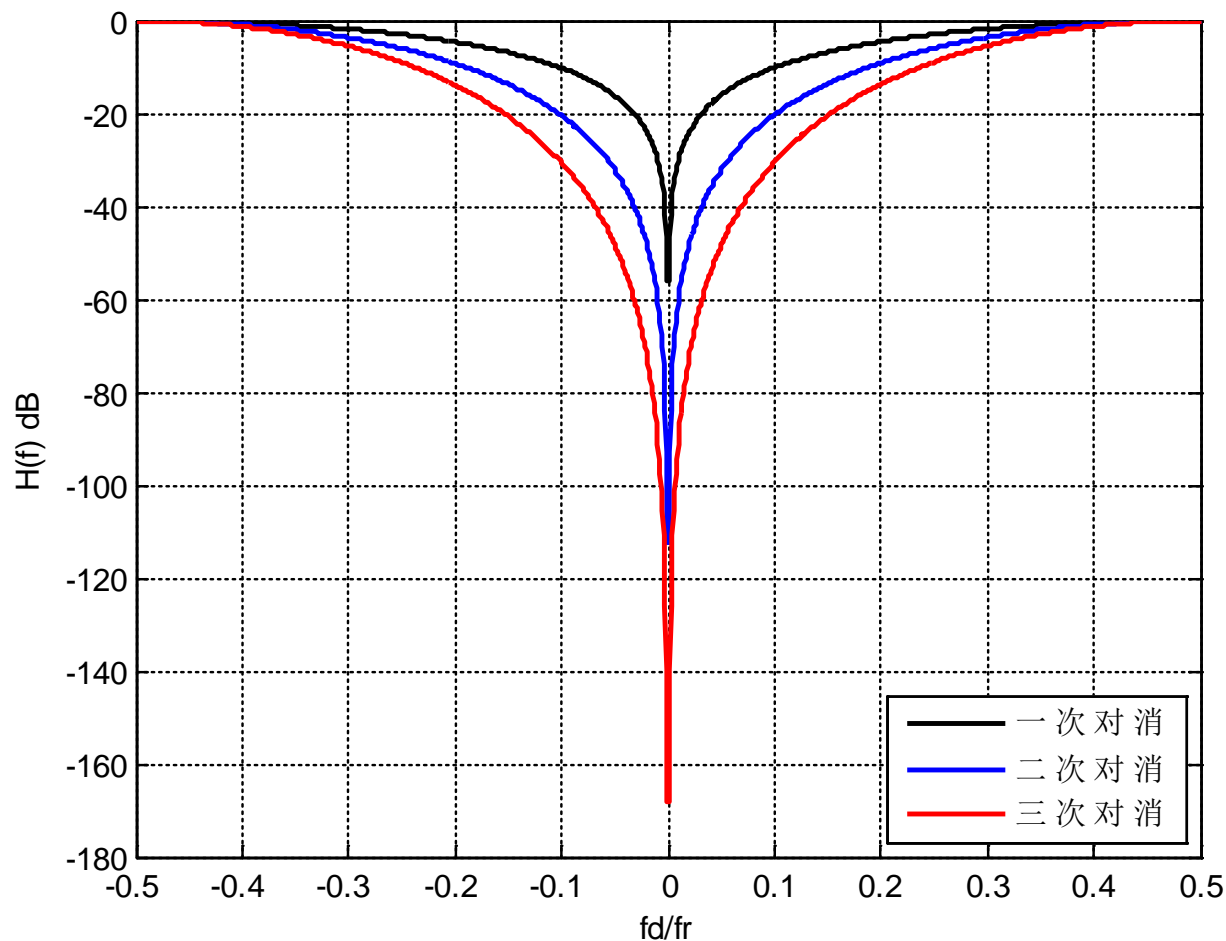
$$\begin{aligned} y_n(m) &= y'_n(m) - y'_{n-1}(m) \\ &= x_n(m) - x_{n-1}(m) - x_{n-1}(m) + x_{n-2}(m) \\ &= x_n(m) - 2x_{n-1}(m) + x_{n-2}(m) \end{aligned}$$





## 二、杂波抑制 - Moving Target Indicator (MTI)

滤波器相应特性





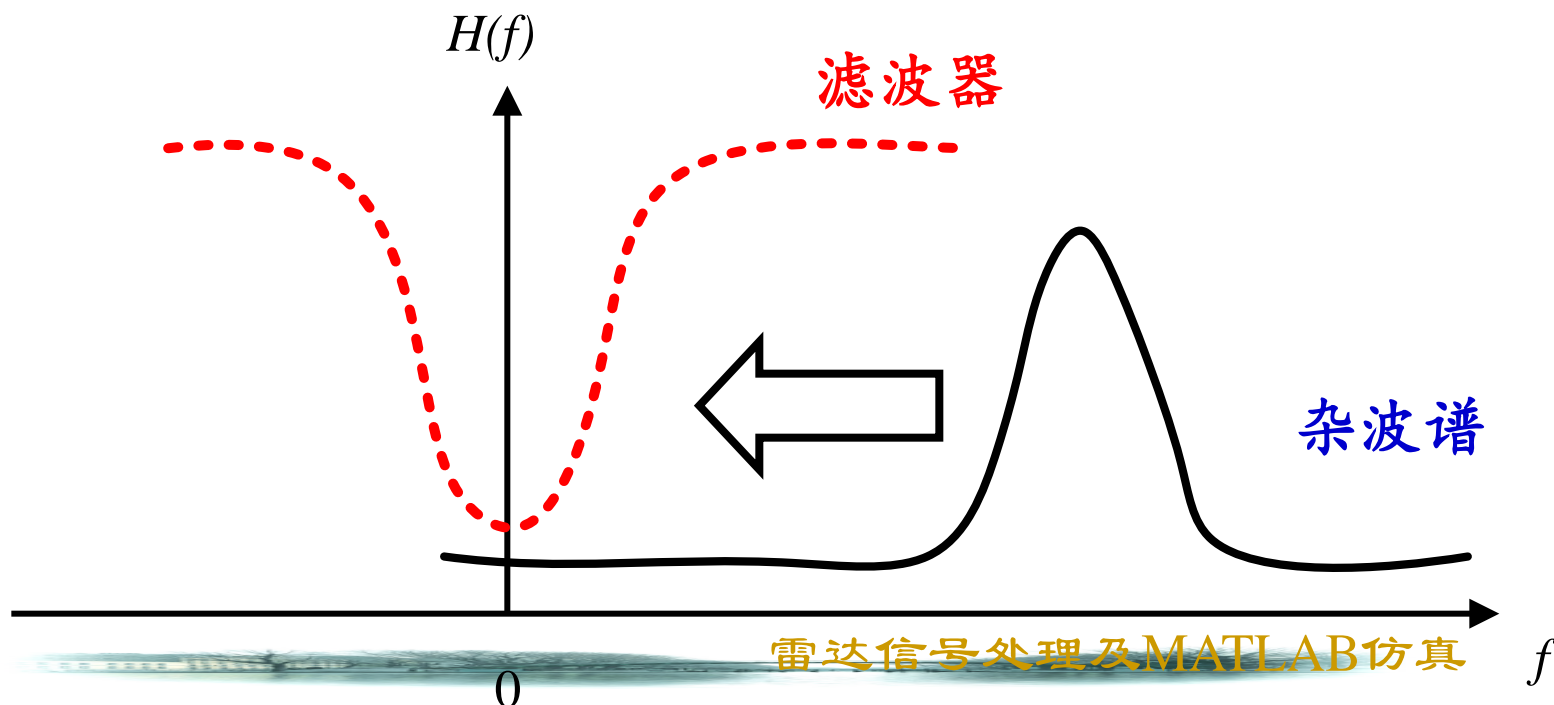
## 二、杂波抑制 - MTI滤波器

### 优点:

权值为实数权，其滤波结构为FIR滤波器，因此结构简单，计算量小。

### 缺点:

- 滤波器的凹口在0频处，即传统滤波器只能对消静止的杂波
- 滤波器的凹口宽度是固定的，不能根据杂波谱的宽度自适应的调节







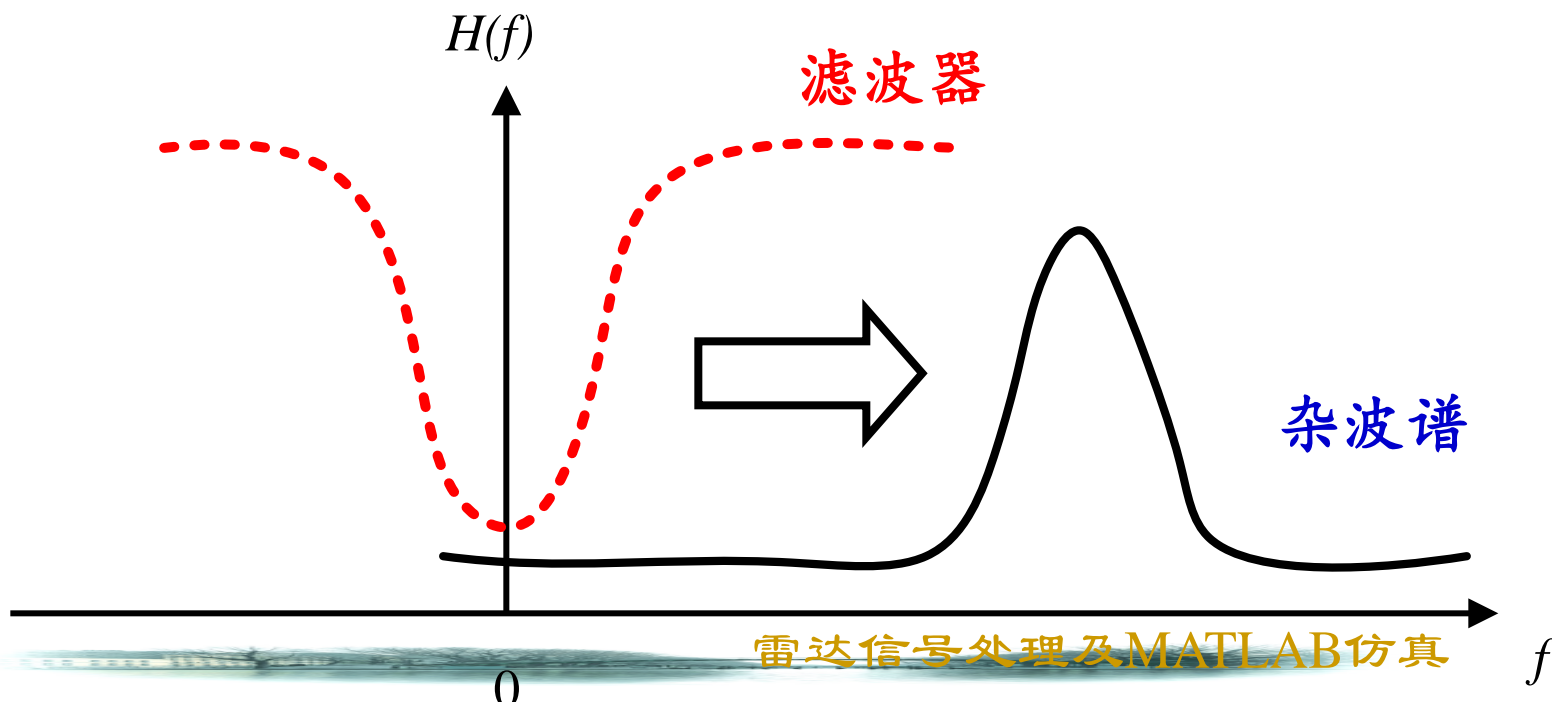
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## 二、杂波抑制 - MTI

### IIR滤波器

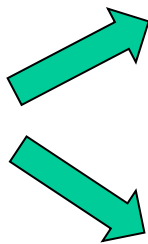
优点： 可以用相对较少的阶数达到滤波器响应；

缺点： 相位非线性 （在实际中很少采用）

### FIR滤波器

优点： 具有线性相位

### MTI滤波器的设计



有效抑制杂波

保证目标信号能良好通过



## 二、杂波抑制 - 特征矢量法

假设MTI滤波器输入的杂波数据和目标数据分别为

$$\begin{aligned} \mathbf{C} &= [c(t_1), c(t_2), \dots, c(t_n)] \\ \mathbf{S} &= [s(t_1), s(t_2), \dots, s(t_n)] \end{aligned} \quad \xrightarrow{\text{MTI 滤波器}} \quad \begin{aligned} C_0 &= E[|\mathbf{C}^H \mathbf{W}|^2] = C_{in} \mathbf{W}^H \mathbf{R}_c \mathbf{W} \\ S_0 &= E[|\mathbf{S}^H \mathbf{W}|^2] = C_{in} \mathbf{W}^H \mathbf{R}_s \mathbf{W} \end{aligned}$$

根据MTI改善因子的定义

$$\begin{aligned} I &= \frac{S_o/C_o}{S_{in}/C_{in}} = \frac{S_o}{S_{in}} \times \frac{C_{in}}{C_o} \\ &= \frac{S_{in} \mathbf{W}^H \mathbf{R}_s \mathbf{W}}{S_{in}} \times \frac{C_{in}}{C_{in} \mathbf{W}^H \mathbf{R}_c \mathbf{W}} \\ &= \frac{\mathbf{W}^H \mathbf{R}_s \mathbf{W}}{\mathbf{W}^H \mathbf{R}_c \mathbf{W}} \end{aligned}$$

$$I = \frac{\mathbf{W}^H \mathbf{R}_s \mathbf{W}}{\mathbf{W}^H \mathbf{R}_c \mathbf{W}} = \frac{\mathbf{W}^H \mathbf{W}}{\mathbf{W}^H \mathbf{R}_c \mathbf{W}}$$

对于目标回波，假设在区间  $(-B/2, B/2)$  上均匀分布

$$S(f) = \begin{cases} 1 & -B/2 \leq f \leq B/2 \\ 0 & \text{其它} \end{cases}$$

目标信号的自相关函数

$$\begin{aligned} r_s(i, j) &= \frac{1}{B} \int_{-B/2}^{B/2} e^{-j2\pi f \tau_{ij}} df \\ &= \frac{\sin(\pi B \tau_{ij})}{\pi B \tau_{ij}} \end{aligned}$$

$$r_s(i, j) = \begin{cases} 1 & i = j \\ 0 & i \neq j \end{cases}$$

$$\tau_{ij} = t_i - t_j$$



## 二、杂波抑制

假设杂波具有高斯分布

$$C(f) = \frac{1}{2\pi\sigma_f} \exp\left\{-\frac{(f-f_0)^2}{2\sigma_f^2}\right\}$$

傅里叶  
逆变换



$$\begin{aligned} r_c(i, j) &= \int_{-\infty}^{\infty} C(f) e^{j2\pi f(t_i - t_j)} df \\ &= \int_{-\infty}^{\infty} \frac{1}{2\pi\sigma_f} \exp\left[-\frac{(f-f_0)^2}{2\sigma_f^2}\right] e^{j2\pi f(t_i - t_j)} df \\ &= e^{-2\pi\sigma_f^2\tau_{ij}^2} \left[ \cos(2\pi f_0\tau_{ij}) + j \sin(2\pi f_0\tau_{ij}) \right] \end{aligned}$$

**Step3:** MTI权值确定

**准则:** 最小特征值所对应的特征矢量作为  
MTI滤波器的权系数矢量



**Step1:** 构建杂  
波协方差矩阵

$$\lambda_1 < \lambda_2 < \lambda_3 < \dots < \lambda_N$$

$$W_1, W_2, W_3, \dots, W_N$$

**Step2:**  
特征值分解



$$R_c = \begin{bmatrix} r_c(0,0) & r_c(0,1) & \dots & r_c(0,N) \\ r_c(1,0) & r_c(1,1) & \dots & r_c(1,N) \\ \vdots & \vdots & \vdots & \vdots \\ r_c(N,0) & r_c(N,1) & \dots & r_c(N,N) \end{bmatrix}$$



## 二、杂波抑制 - 特征矢量法

### 改善因子最大准则

$$I = \frac{S_o / C_o}{S_{in} / C_{in}} = \frac{S_o}{S_{in}} \times \frac{C_{in}}{C_o}$$

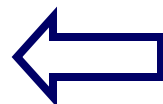


$$C_o = E \left[ \left[ C^H W \right]^2 \right] = C_{in} W^H R_c W$$

$$S_o = E \left[ \left[ S^H W \right]^2 \right] = S_{in} W^H R_s W$$



$$I = \frac{S_{in} W^H R_s W}{S_{in}} \times \frac{C_{in}}{C_{in} W^H R_c W} = \frac{W^H R_s W}{W^H R_c W}$$

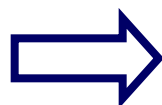


$$R_s = I \text{ (单位阵)}$$

$$R_c W_i = \lambda_i W_i \quad i = 1, 2, \dots, N$$



$$I = \frac{W_i^H W_i}{W_i^H \lambda_i W_i} = \frac{W_i^H W_i}{\lambda_i W_i^H W_i} = \frac{1}{\lambda_i}$$

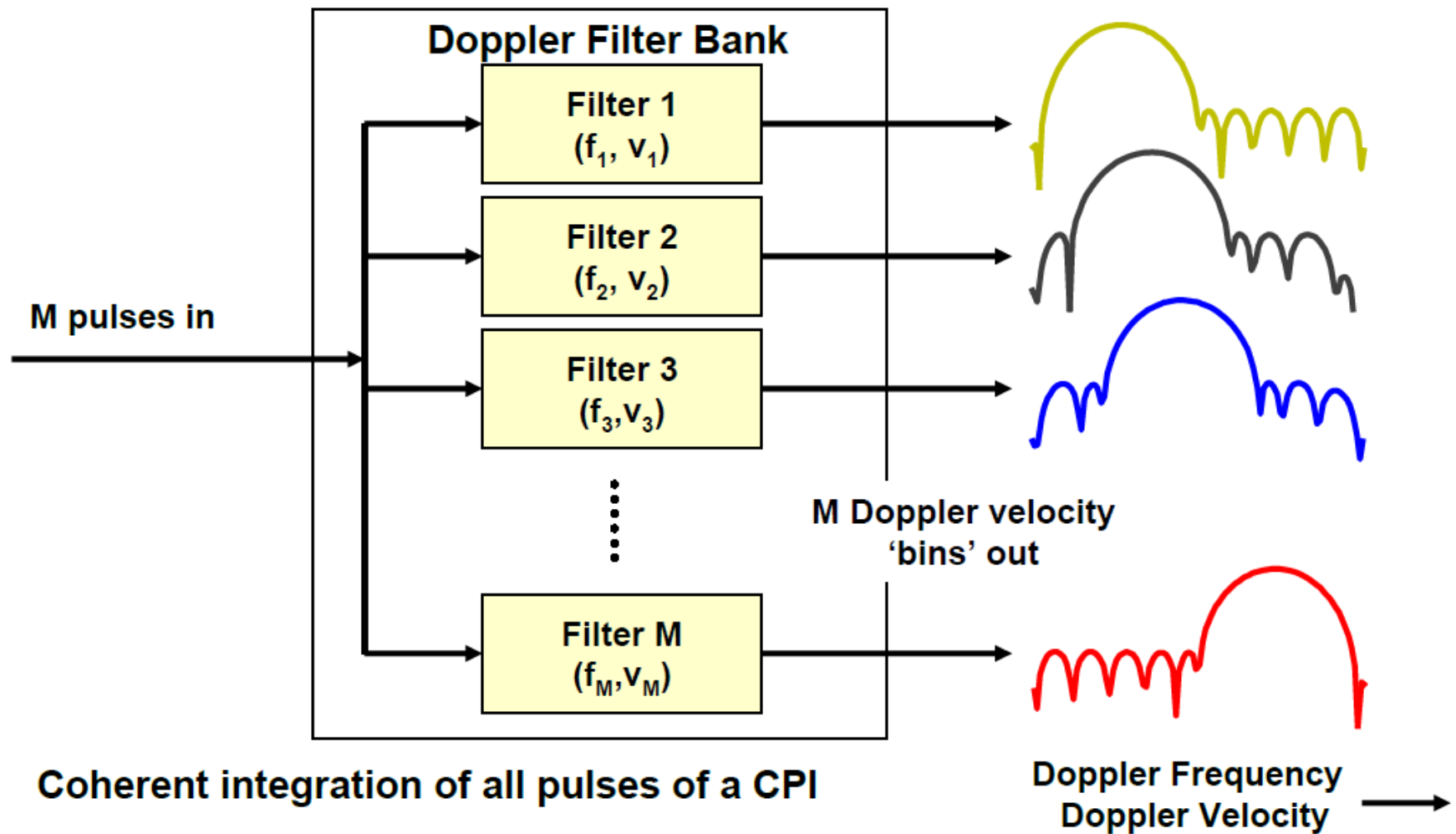


$$I \rightarrow \text{MAX} \Rightarrow \lambda \rightarrow \text{MIN}$$

噪声子空间和信号子空间是正交的，所以最小特征值所对应的特征向量作为MTI权系数矢量，就可以最大限度的抑制杂波



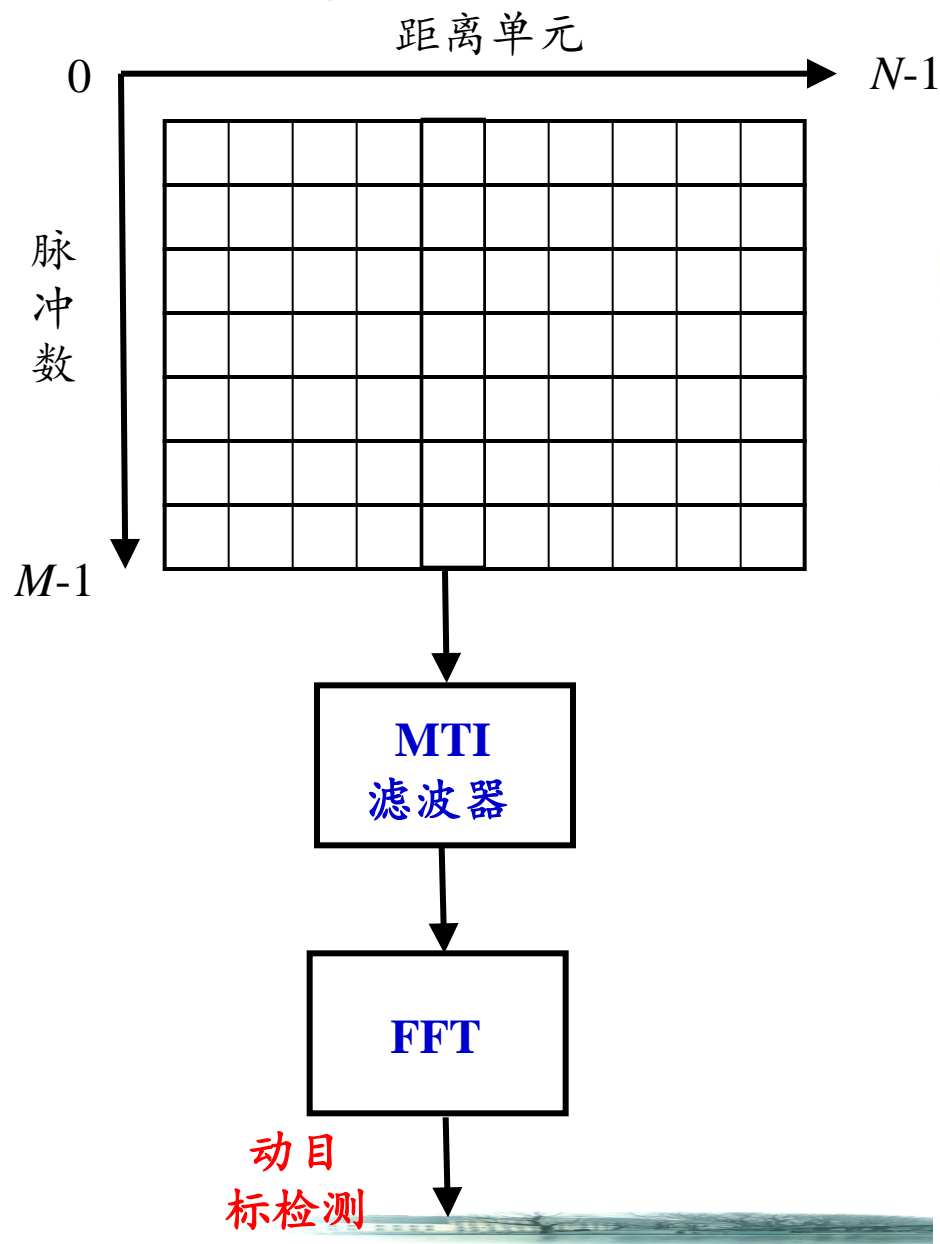
## 二、杂波抑制 - 动目标检测 (MTD)



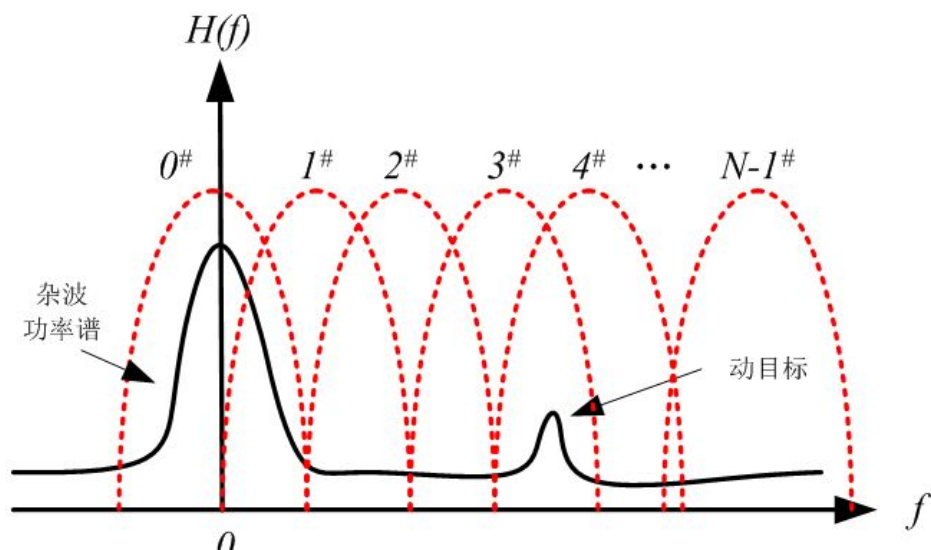
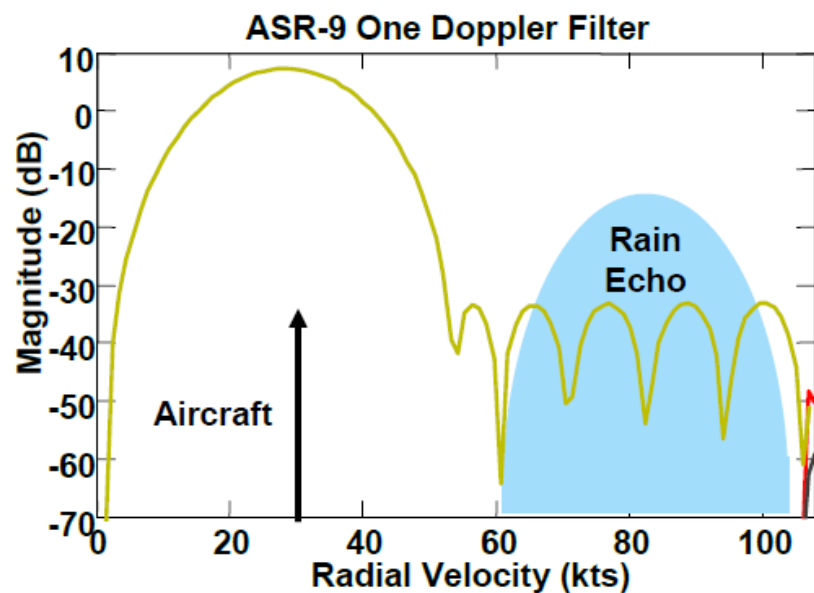
- Coherent integration of all pulses of a CPI
- Clutter rejection
- Resolving targets into different velocity segments and allowing for fine-grain target radial velocity estimation



## 二、杂波抑制



$$X(k) = \sum_{n=0}^{N-1} x(n)e^{-j2\pi nk/N}$$

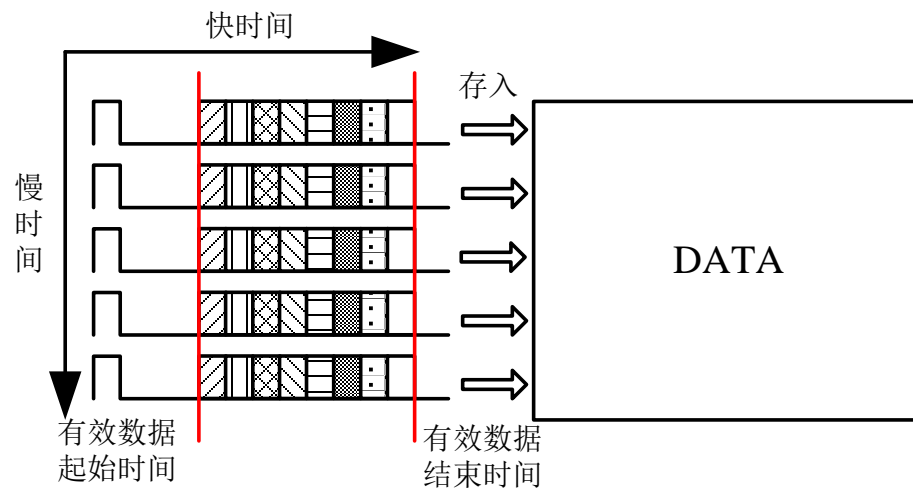




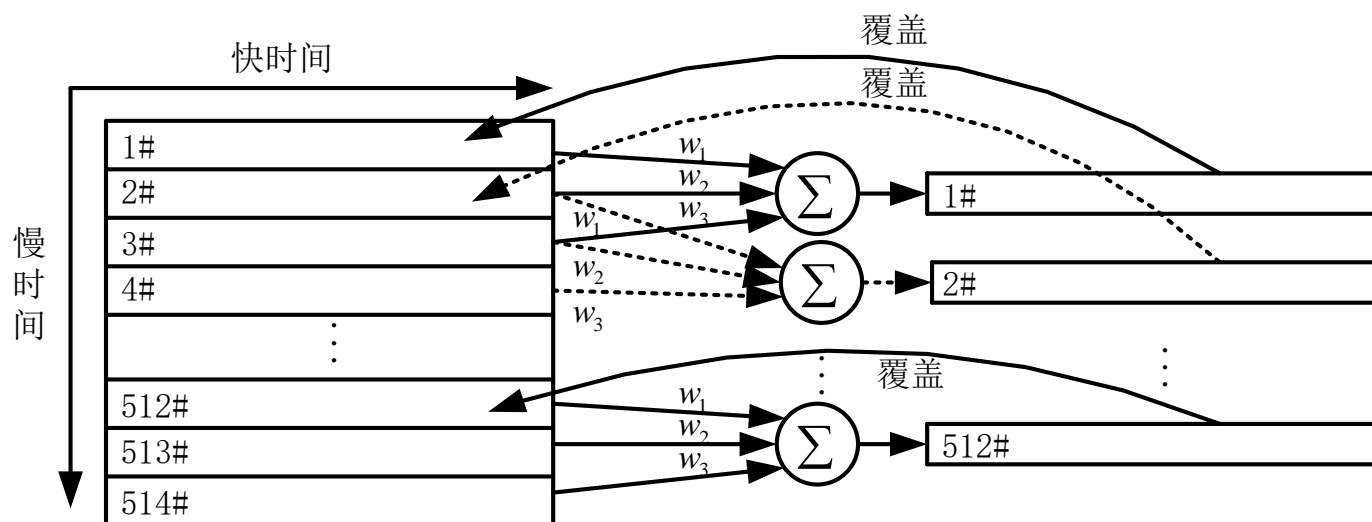


## 二、杂波抑制

### □ 数据采集与存储阶段



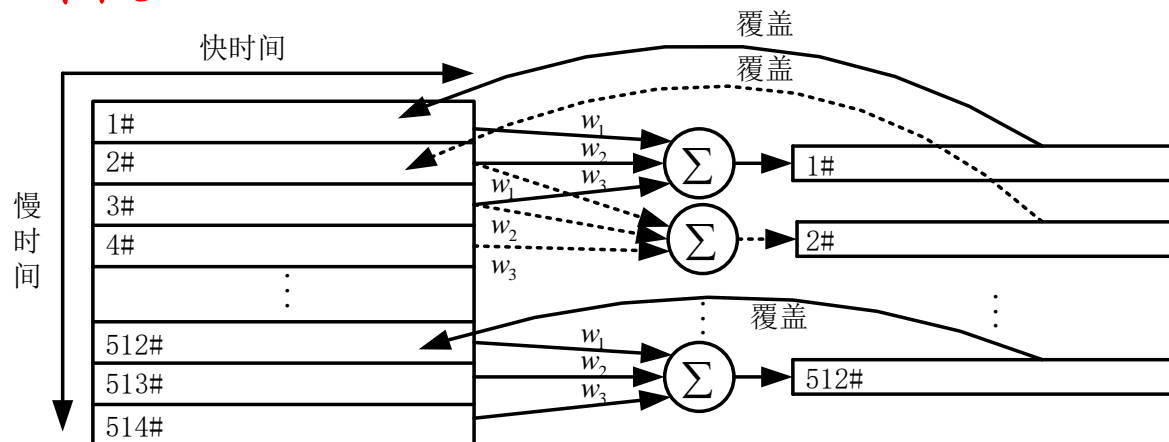
### □ MTI阶段



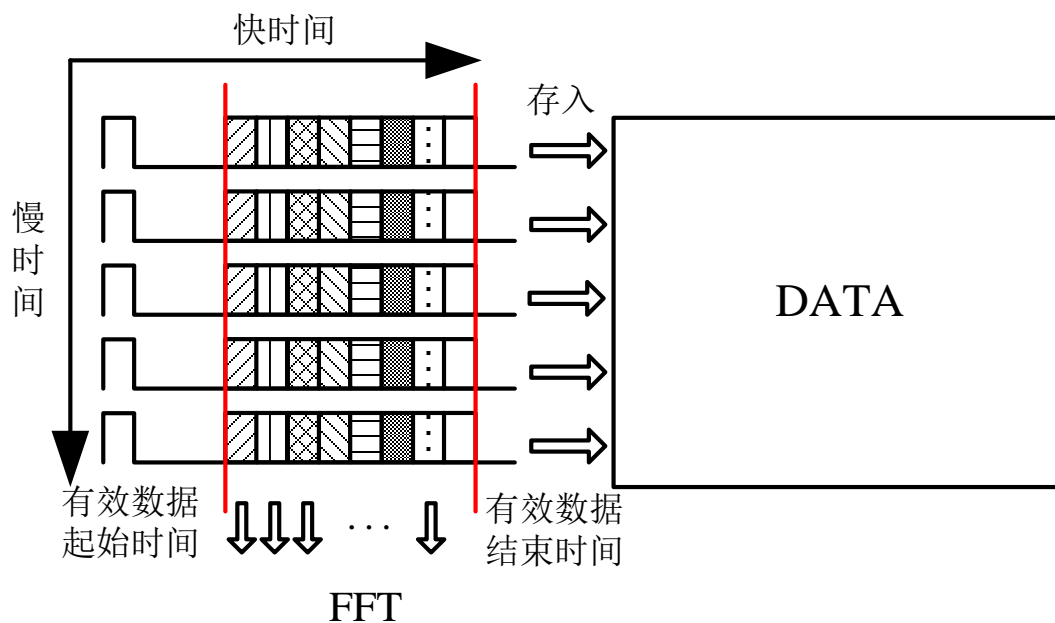


## 二、杂波抑制

### □ MTI阶段



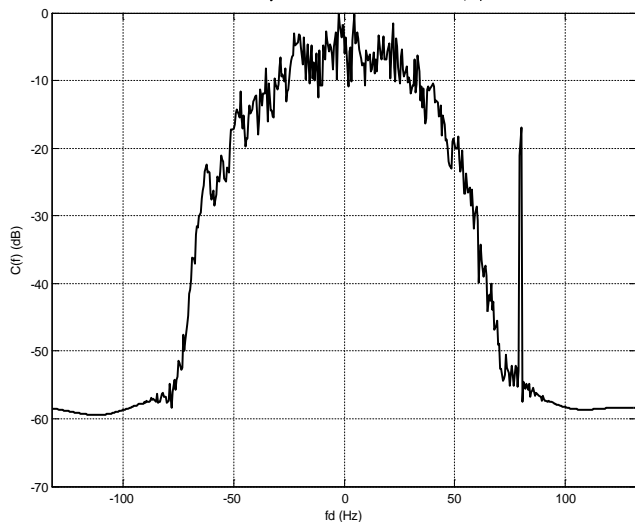
### □ MTD阶段



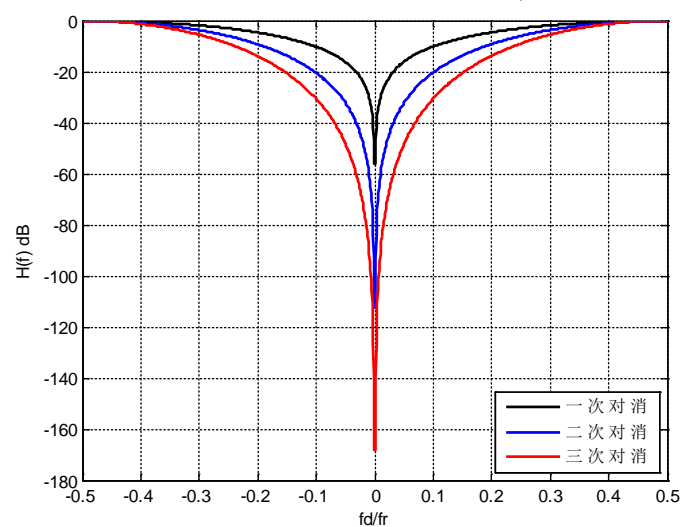


## 二、杂波抑制 - 传统方法

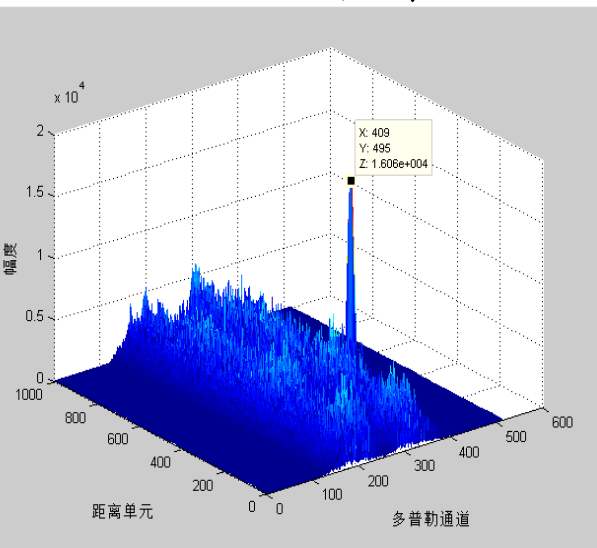
### 杂波频谱



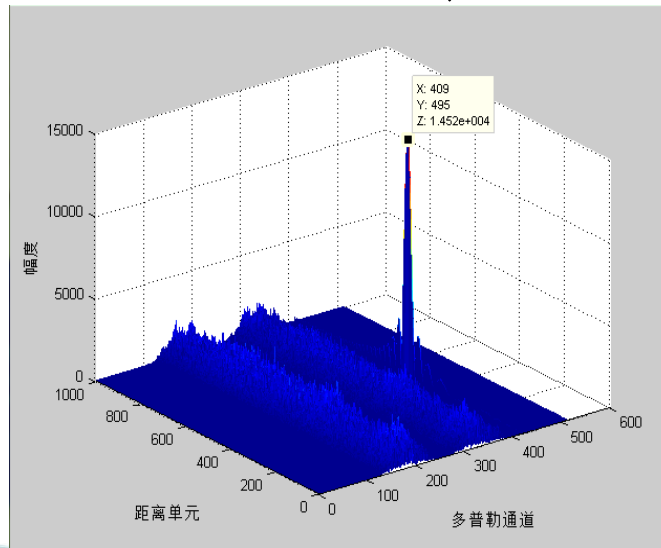
### 滤波器相应特性



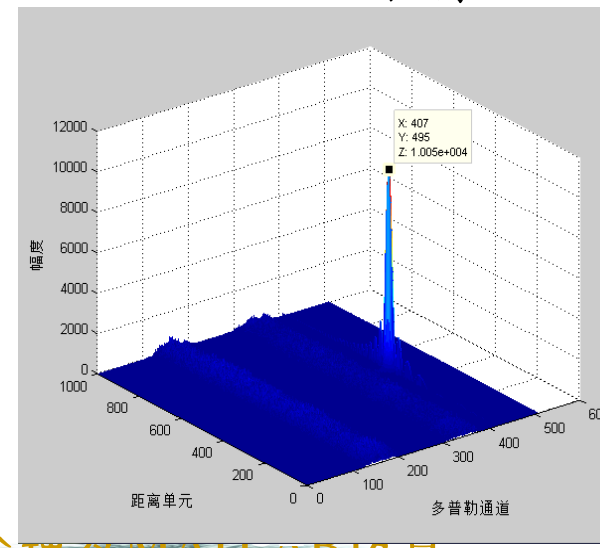
### 一次对消



### 二次对消

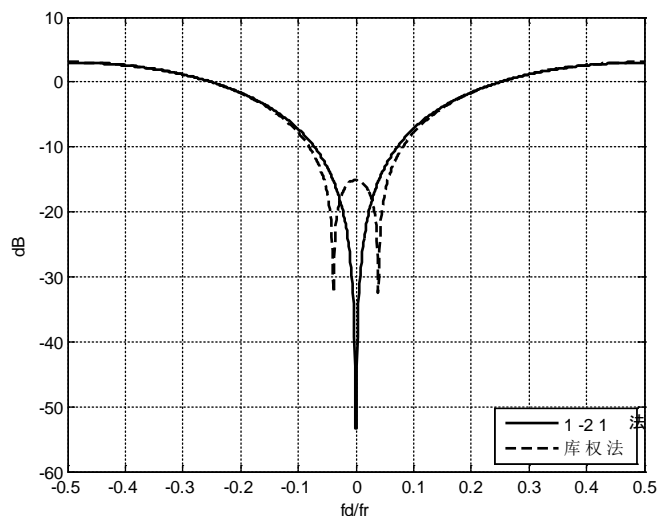


### 三次对消

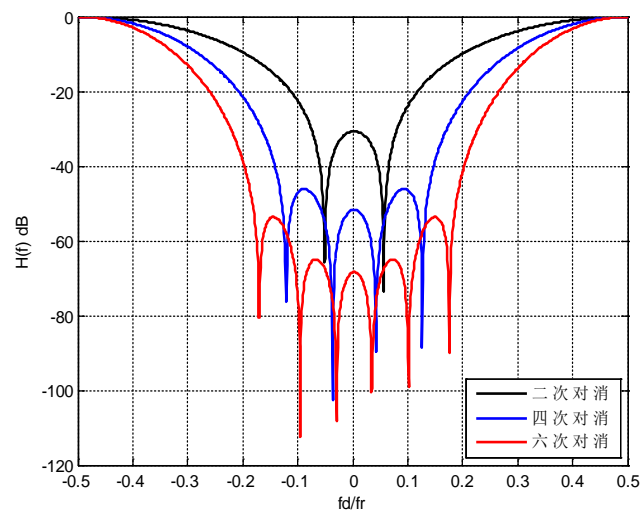




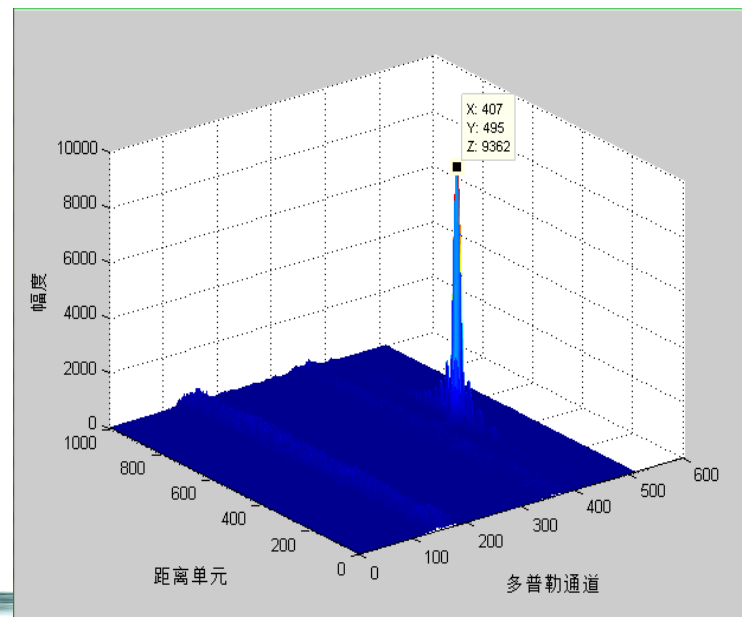
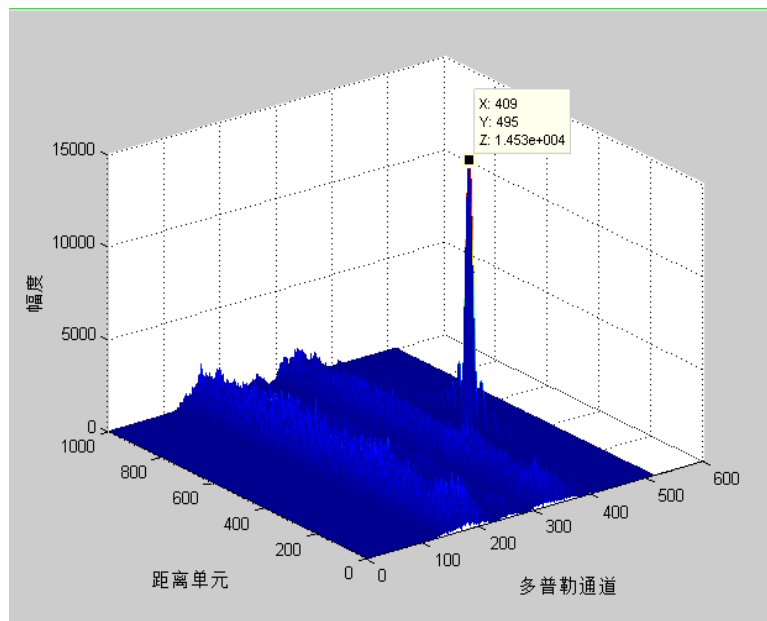
## 二、杂波抑制 - 特征矢量法



二次对消



四次对消





# 杂波抑制与动目标检测

- 杂波的定义
- 杂波抑制技术
- 动目标检测
- 本章小结



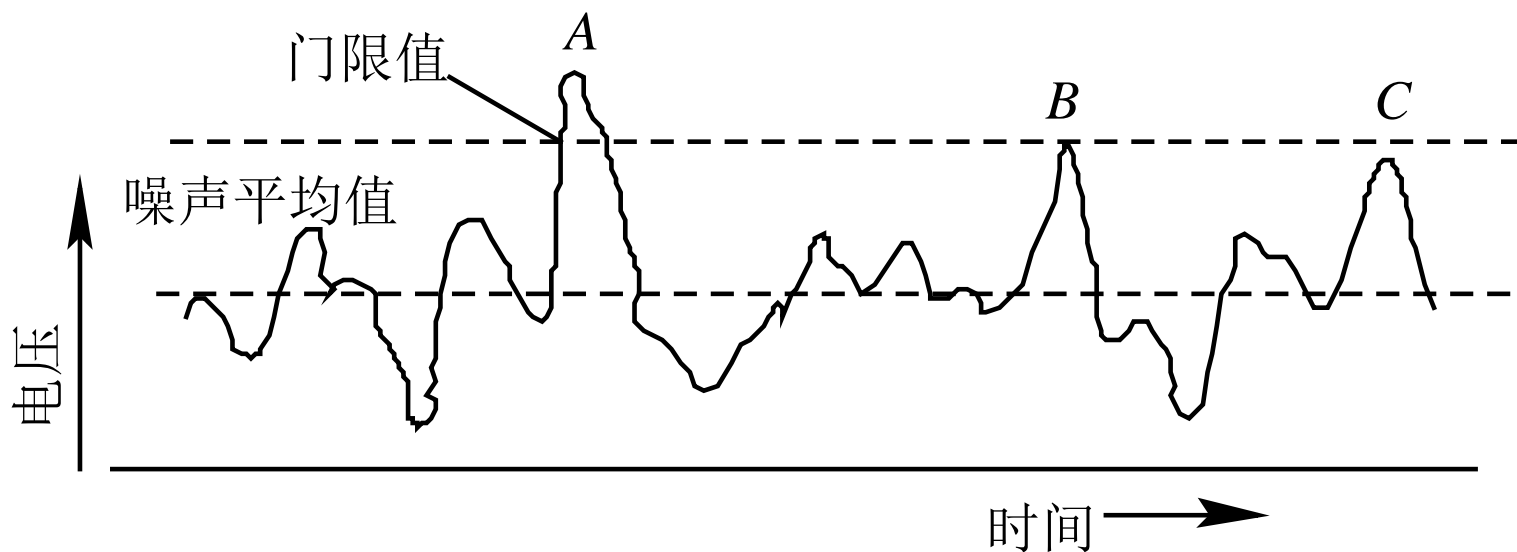
### 三、动目标检测

检测时门限电压的高低影响以下两种错误判断的多少: ☐

(1) 有信号而误判为没有信号(**漏警**); ☐

(2) 只有噪声时误判为有信号(**虚警**)。 ☐

应根据两种误判的影响大小来选择合适的门限。





### 三、动目标检测

门限检测是一种统计检测，由于信号叠加有噪声，所以总输出是一个随机量。在输出端根据输出振幅是否超过门限来判断有无目标存在，可能出现以下四种情况：□

- (1)发现概率 $P_d$ ：存在目标时，判为有目标；
- (2)漏报概率 $P_{la}$ ：存在目标时，判为无目标；□
- (3)正确不发现概率 $P_{an}$ ：不存在目标时判为无目标；□
- (4)虚警概率 $P_{fa}$ ：不存在目标时判为有目标；□

$$P_d + P_{la} = 1, \quad P_{an} + P_{fa} = 1$$





### 三、动目标检测

虚警概率  $P_{fa}$  □

通常加到接收机中频滤波器(或中频放大器)上的噪声是**宽带高斯噪声**,其概率密度函数由下式给出:

$$p(v) = \frac{1}{\sqrt{2\pi}\sigma} \exp\left(-\frac{v^2}{2\sigma^2}\right)$$

包络检波器输出端**噪声电压振幅**的概率密度函数为**瑞利分布**:

$$p(r) = \frac{r}{\sigma^2} \exp\left(-\frac{r^2}{2\sigma^2}\right) \quad r \geq 0$$

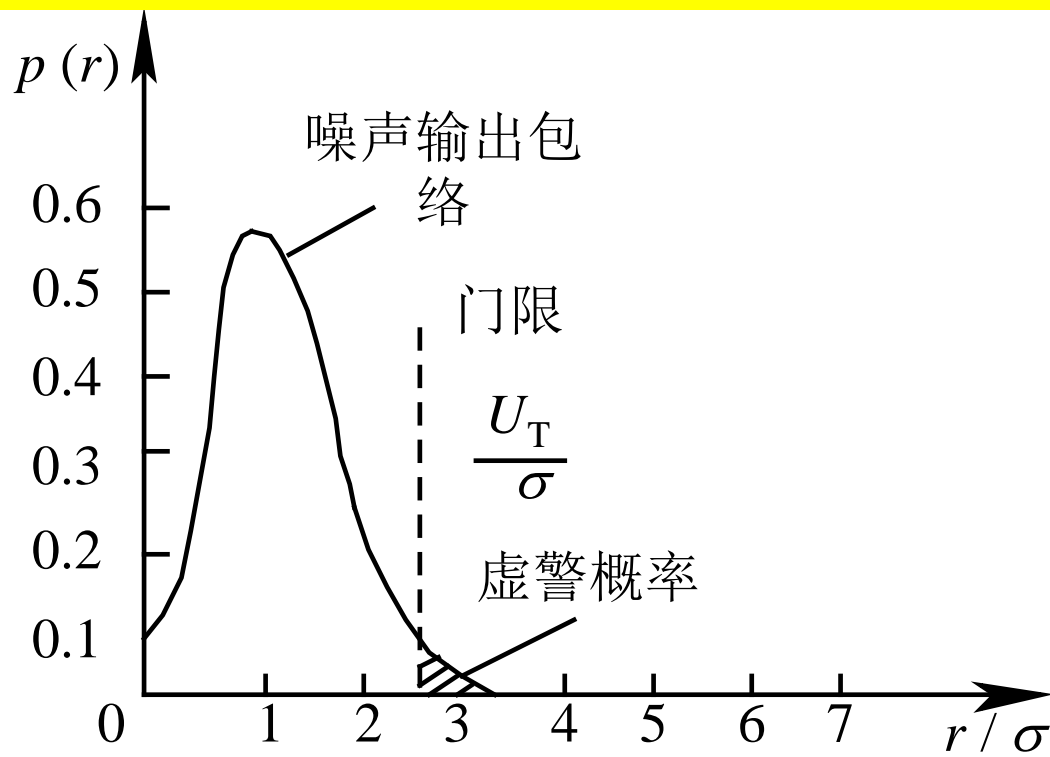


### 三、动目标检测

#### 虚警概率 $P_{fa}$

设置门限电平  $U_T$ , 噪声包络电压超过门限电平的概率就是虚警概率  $P_{fa}$ , 它可以由下式求出:

$$P_{fa} = P(U_T \leq r < \infty) = \int_{U_T}^{\infty} \frac{r}{\sigma^2} \exp\left(-\frac{r^2}{2\sigma^2}\right) dr = \exp\left(-\frac{U_T^2}{2\sigma^2}\right)$$



仿真



## 三、动目标检测

为获得可预知且稳定的检测性能，雷达系统一般倾向于设计雷达具有恒定的虚警概率。因此，实际干扰噪声功率电平必须实时的从数据中估计，从而调整雷达检测门限以获得期望的虚警概率。

可保持**恒定虚警概率**的检波处理器——**恒虚警检测(CFAR)**

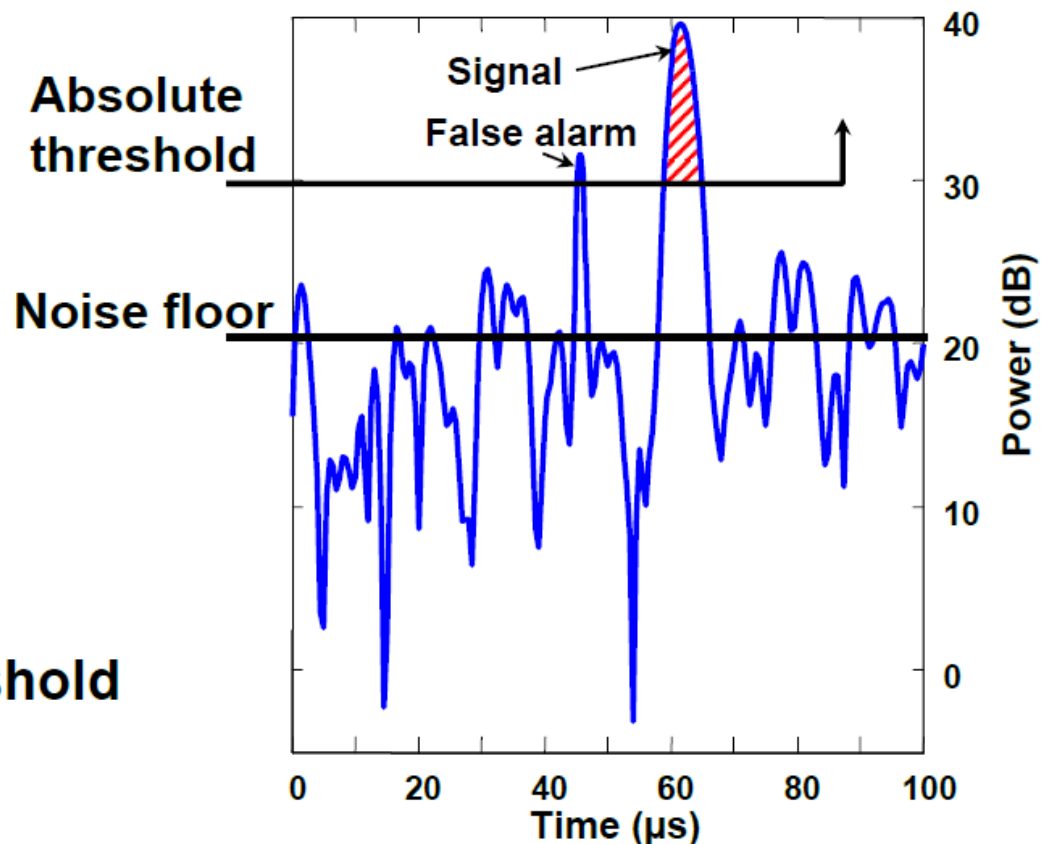
恒虚警检测的方法就是采用自适应门限代替固定门限，因此自适应门限能随着被检测点的背景噪声、杂波和干扰的大小自适应调整。



## 三、动目标检测

- Problem: Must know (or estimate) noise floor to set threshold
- Solution: Estimate noise floor using noise-only samples
  - Adaptive thresholding
- CFAR thresholding:

$$\frac{\text{test cell}}{\text{noise floor estimate}} > \text{threshold}$$

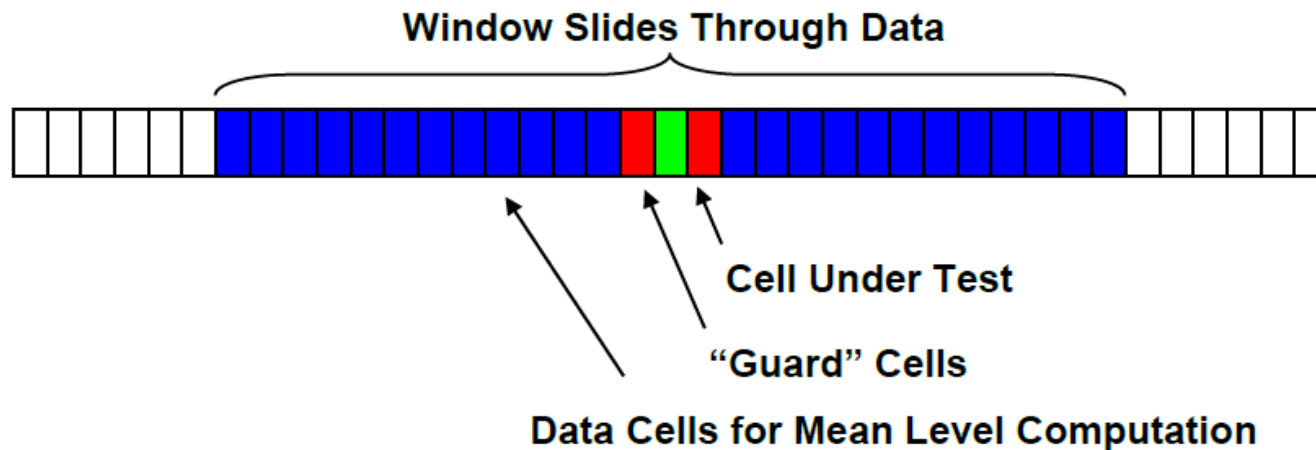




### 三、动目标检测

#### 平均单元恒虚警检测 (CA-CFAR)

- Use mean value of surrounding range cells to determine threshold for cell under test

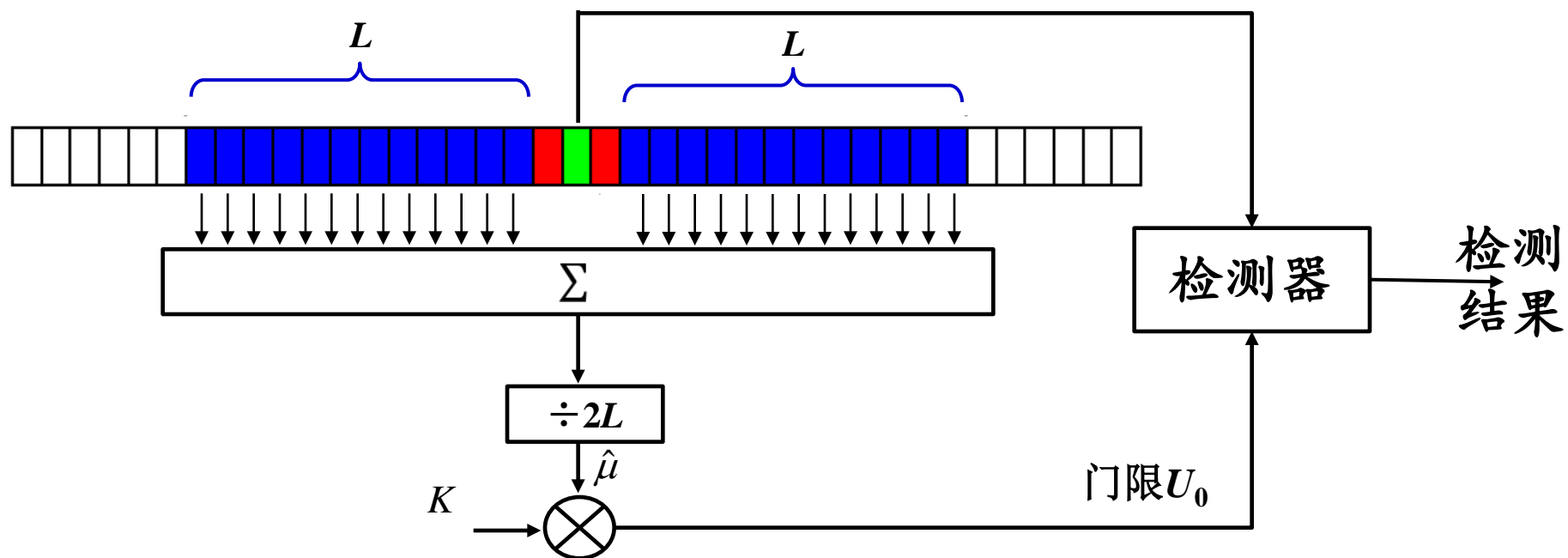


- Nearby targets can raise threshold and suppress detection



### 三、动目标检测

#### 平均单元恒虚警检测 (CA-CFAR)



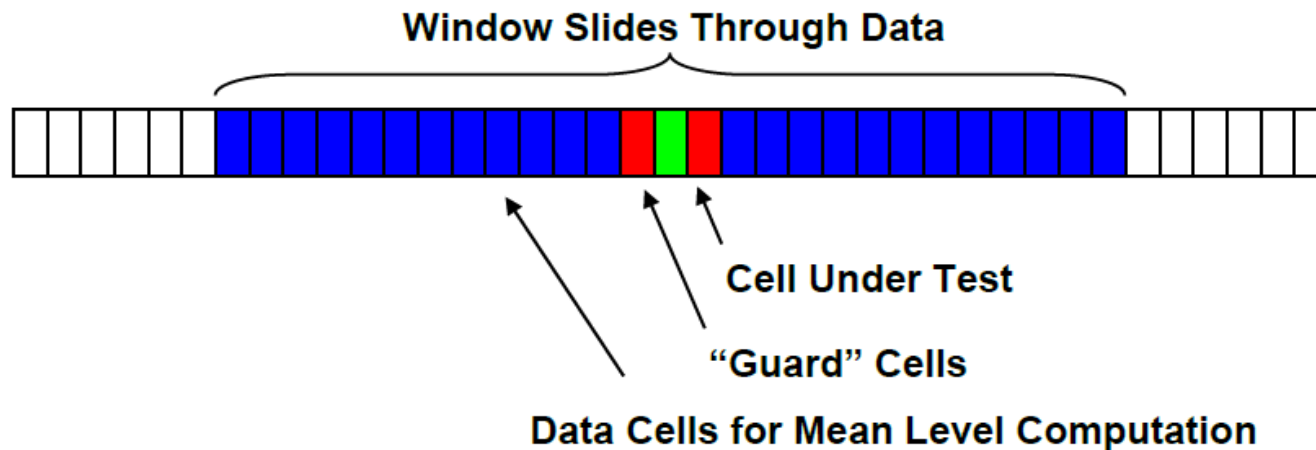
$$U_0 = K \hat{\mu}$$



### 三、动目标检测

#### 平均单元恒虚警检测 (CA-CFAR)

- Use mean value of surrounding range cells to determine threshold for cell under test



- Nearby targets can raise threshold and suppress detection

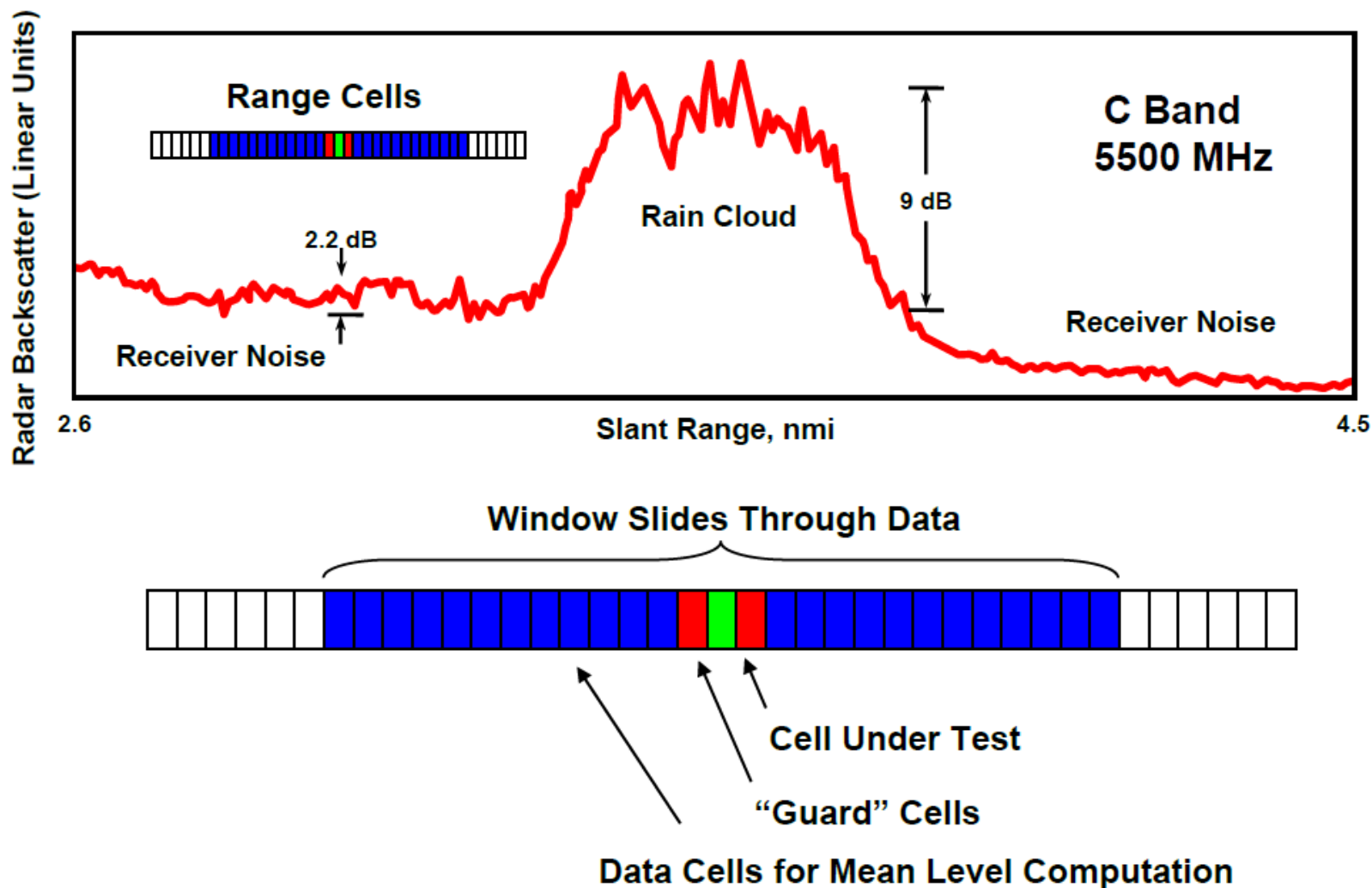




### 三、动目标检测

选大恒虚警 (GO-CFAR)

选小恒虚警 (SO-CFAR)

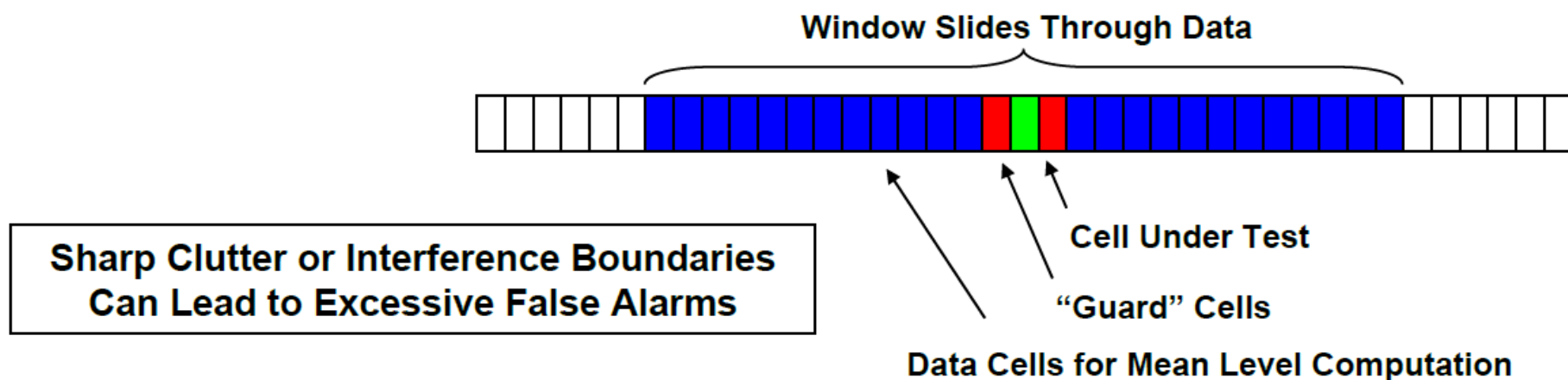
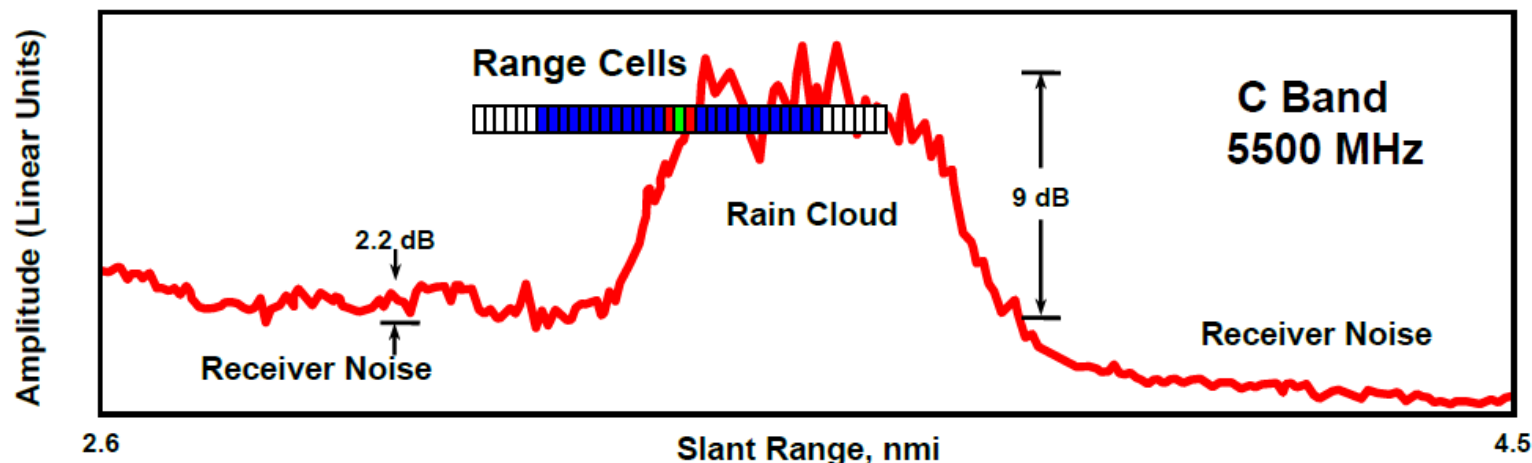




### 三、动目标检测

选大恒虚警 (GO-CFAR)

选小恒虚警 (SO-CFAR)



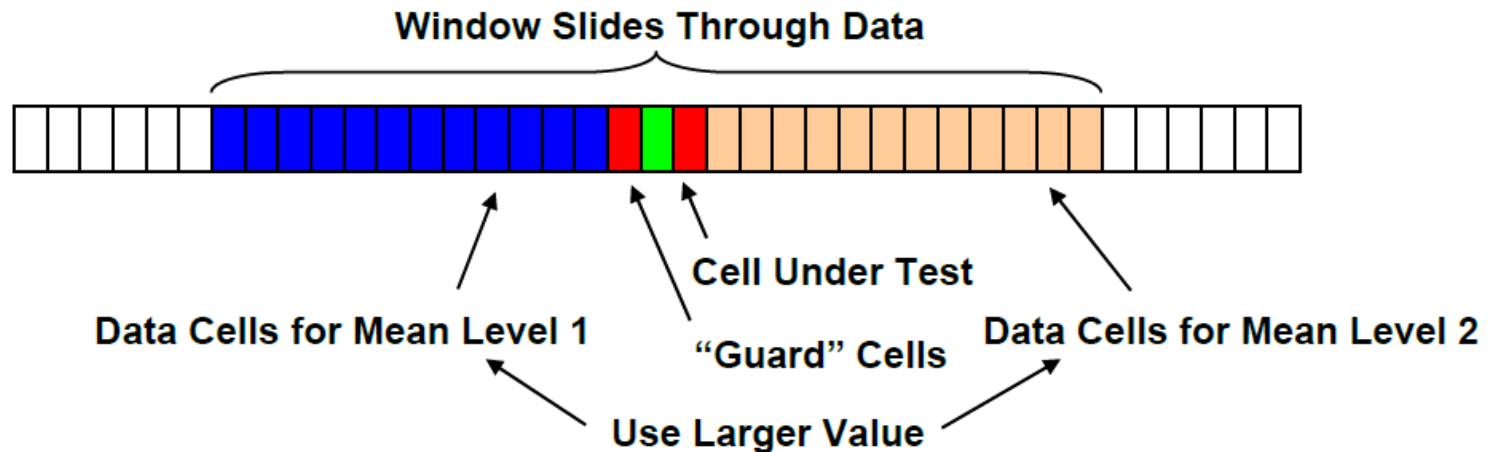


### 三、动目标检测

选大恒虚警 (GO-CFAR)

选小恒虚警 (SO-CFAR)

- Find mean value of  $N/2$  cells before and after test cell separately
- Use larger noise estimate to determine threshold



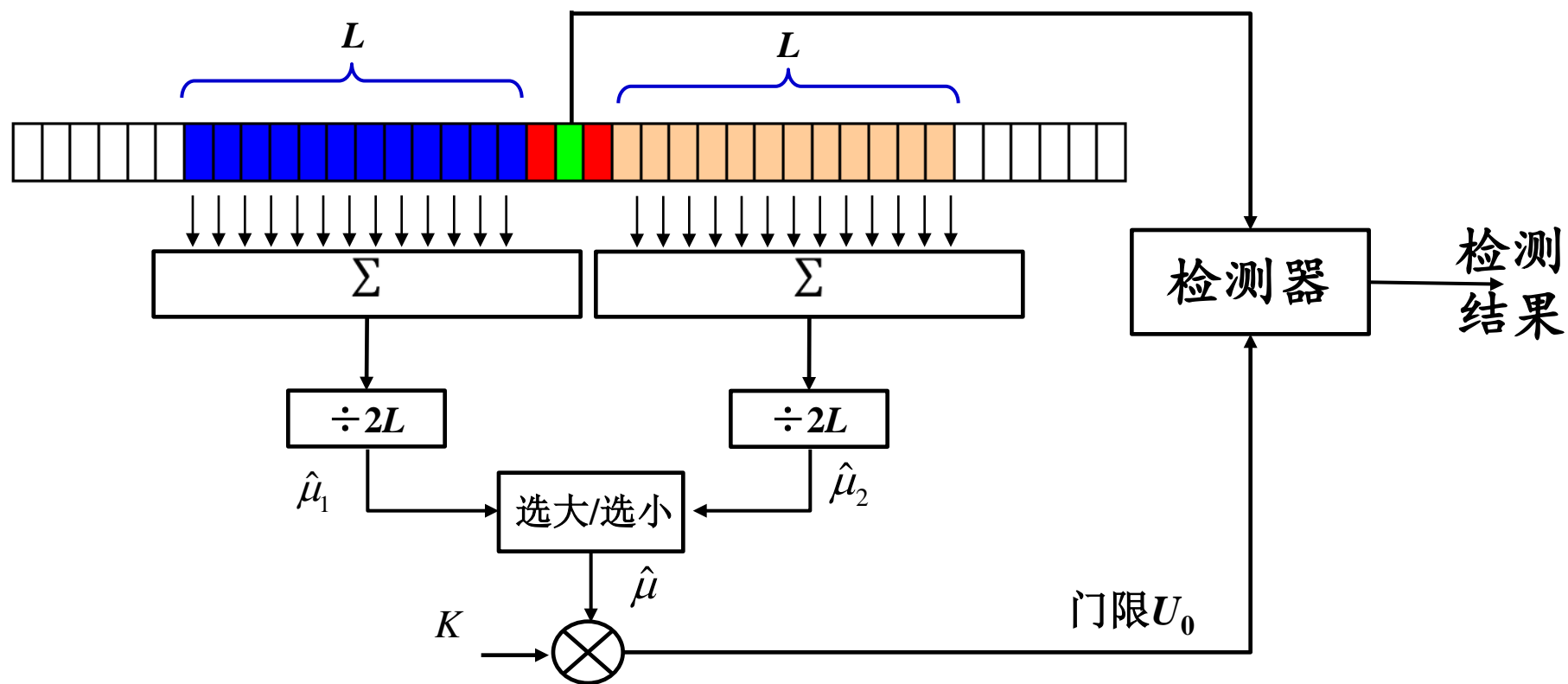
- Helps reduce false alarms near sharp clutter or interference boundaries
- Nearby targets still raise threshold and suppress detection



### 三、动目标检测

选大恒虚警 (GO-CFAR)

选小恒虚警 (SO-CFAR)



$$U_0 = K \max \{ \hat{\mu}_1, \hat{\mu}_2 \}$$

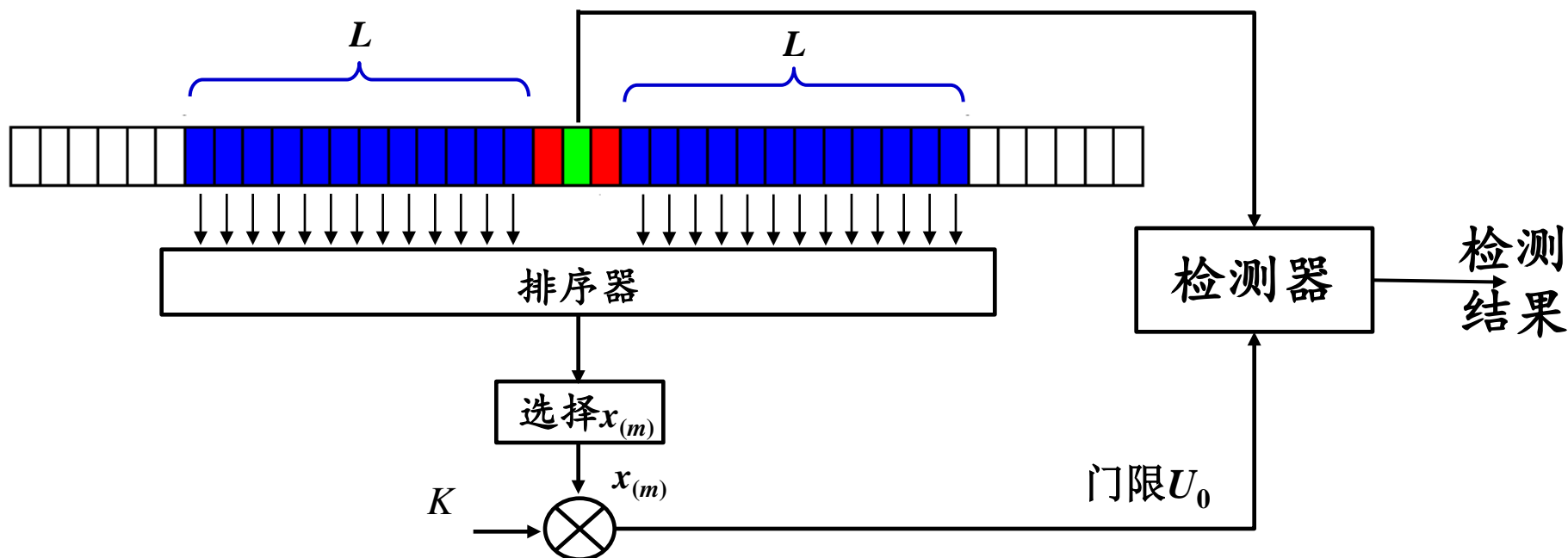
$$U_0 = K \min \{ \hat{\mu}_1, \hat{\mu}_2 \}$$



### 三、动目标检测

#### 有序恒虚警检测器 (OS-CFAR)

如果参考单元中出现其它信号（**干扰信号**），将引起恒虚警检测器性能的下降。为提高恒虚警抗其它干扰目标的能力——**有序恒虚警检测器**

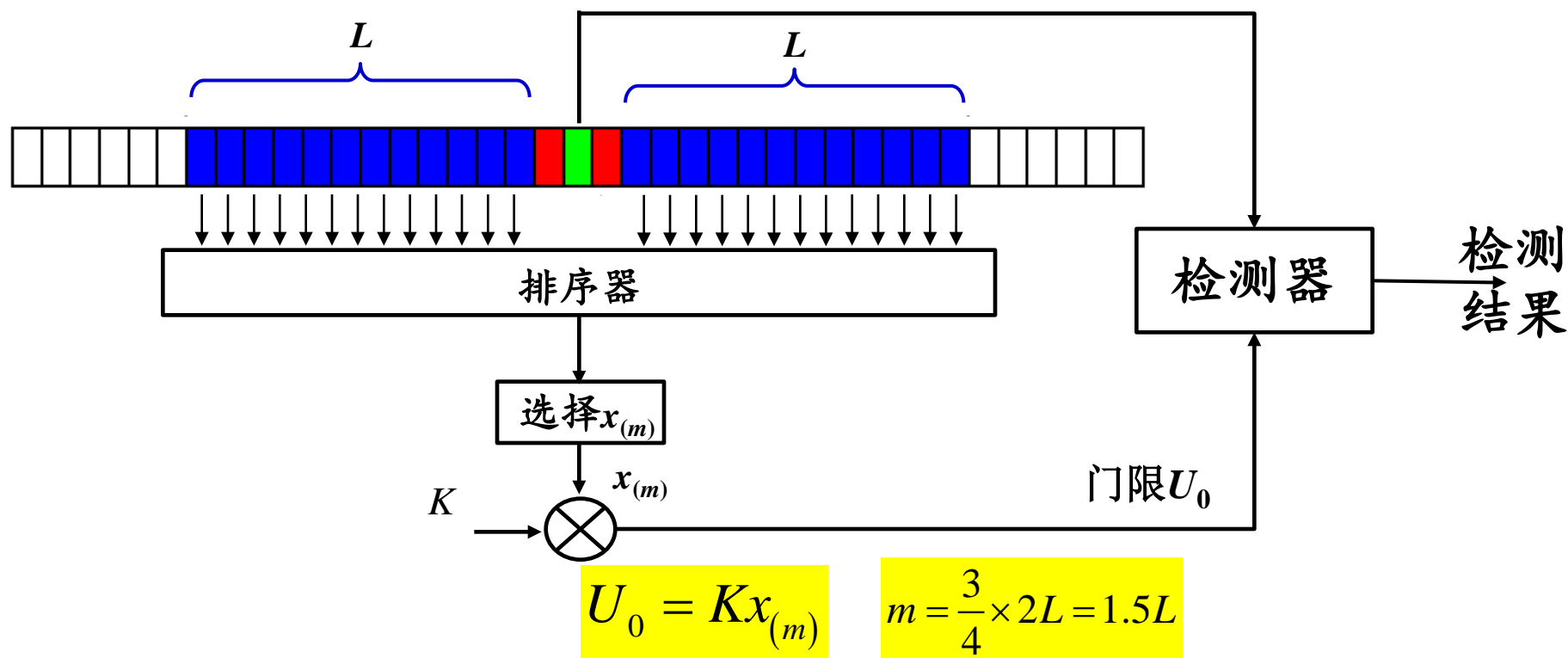




### 三、动目标检测

#### 有序恒虚警检测器 (OS-CFAR)

排序器对参考单元内的 $2L$ 个 $x$ 值进行排序,  $x_{(1)} \leq x_{(2)} \leq x_{(3)} \leq \dots \leq x_{(2L)}$   
再选择第 $m$ 个样本作为 $2L$ 个电平的一种估计



当有将强的1个或多个干扰目标进入 $2L$ 个参考单元时, 只会引起OS-CFAR检测器中排序结果变化



### 三、动目标检测

#### ➤ 剔除和平均恒虚警检测器

$r$

$$x_{(1)} \leq x_{(2)} \leq x_{(3)} \leq x_{(4)} \leq x_{(5)} \cdots \cdots \cdots \boxed{\cdots \leq x_{(2L-2)} \leq x_{(2L-1)} \leq x_{(2L)}}$$

剔除排序器内前 $r$ 最大值，再对剩余的求平均

$$\mu = \frac{1}{2L-r} \sum_{i=1}^{2L-r} x_{(i)}$$

#### ➤ 整理和平均恒虚警检测器

$r_2$

$r_1$

$$\boxed{x_{(1)} \leq x_{(2)} \leq x_{(3)} \leq x_{(4)} \leq x_{(5)}} \cdots \cdots \cdots \boxed{\cdots \leq x_{(2L-2)} \leq x_{(2L-1)} \leq x_{(2L)}}$$

剔除排序器内前 $r_1$ 最大值和后 $r_2$ 个最小值，再对剩余的求平均

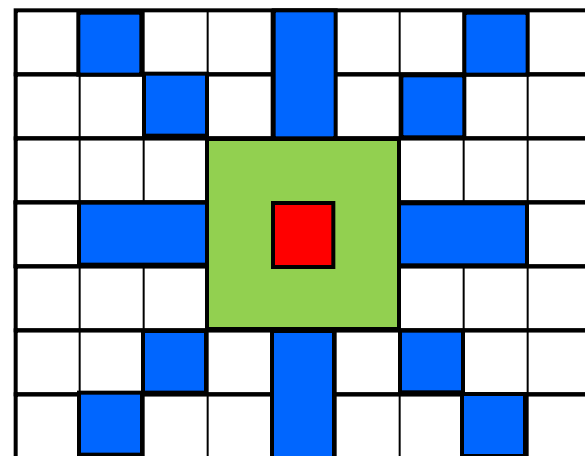
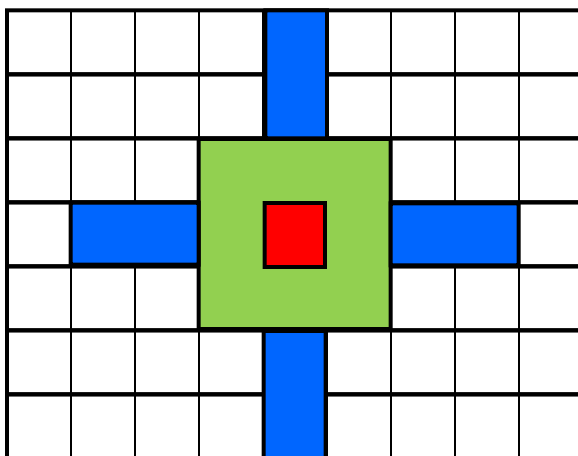
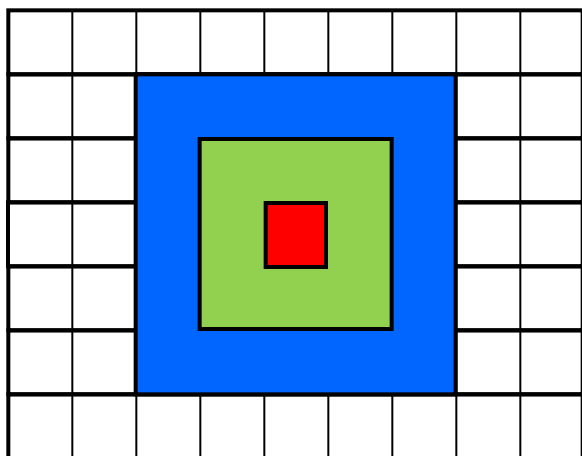
$$\mu = \frac{1}{2L-(r_1+r_2)} \sum_{i=r_1+1}^{2L-r_2} x_{(i)}$$





### 三、动目标检测

#### 二维恒虚警检测器



待检测单元



保护单元



参考单元



# 杂波抑制与动目标检测

- 杂波的定义
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- 动目标检测
- 本章小结



## 本章要点

### ◆ 基本要求

了解杂波对消器、特征矢量法以及恒虚警检测原理；

### ◆ 重点、难点

掌握MTI、MTD、CFAR仿真流程，能够熟练运用MATLAB工具进行仿真。

### ◆ 课堂作业

1, -2, 1 滤波器仿真、一维CFAR 仿真。

