

Brief Papers

Passivity-Based Control for a Rolling-Balancing System: The Nonprehensile Disk-on-Disk

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Abstract—In this brief, we propose a passivity-based control design for a rolling-balancing system called the disk-on-disk (DoD). The stabilization of the desired equilibrium is obtained via energy shaping and damping injection. The DoD is an underactuated mechanical system composed of two disks arranged one on top of the other. The top disk, which we call the object, is free to roll without slipping on the lower disk, which we call the hand. The hand is actuated by a controlled torque, while the object is unactuated. The control objective is to balance the object at the upright position and drive the hand to a desired angle. We design an energy shaping controller without solving the partial differential equations, which rise from the matching equation. We assess the performance of the controller by both simulations and experiment results, which also verify the practical applicability of the design approach.

Index Terms—Energy shaping, nonprehensile systems, passivity-based control, rolling-balancing systems.

I. INTRODUCTION

MANIPULATING an object without grasping is a task known as *nonprehensile manipulation*. Although this class of manipulation problems has received great attention by the research community, it is still rather far from being fully solved for robotic applications [1]–[3]. There are several advantages in nonprehensile dynamic manipulation. Since the object is not caged between fingertips during the task, it is possible to manipulate the object outside the robot workspace by allowing both contact and noncontact interaction between the robot and the object, e.g., by throwing and catching the object [1]. Moreover, it is possible to control more

object degrees of freedom than the actuators of the robotic platform [3], [4]. In several industrial applications, it is not directly possible to manipulate the object through firm or fine/precise manipulation, and therefore, only nonprehensile manipulation is allowed to accomplish the task, e.g., using vibratory platforms [1]. Dynamic nonprehensile manipulation tasks are performed by surgeons with their instruments during operations, for example when they push away an organ or an artery. For this class of systems, the control design has to take into account the dynamics of both the robot and the object, which increases the complexity of the design. Thus, challenging problems in high-speed sensing and control fields arise in nonprehensile manipulation.

A classical approach for the control design of dynamic nonprehensile tasks is to divide a complex action into simpler primitives and subtasks, such as rolling, pushing, throwing, batting, and juggling, to mention some of them [1], [5], [6]. In this brief, we concentrate on a particular primitive of nonprehensile manipulations, that is *rolling*, considering the disk-on-disk (DoD) system. We address the stabilization-balancing problem of the DoD using passivity-based control and port-Hamiltonian (pH) systems (see [7] and [8] for a survey on these topics). The DoD controller proposed in this brief is designed under the assumption that the disks are in contact. Such assumption cannot be ensured in the physical setup, since the disks are not mechanically attached. However, the experiments show that the controller performs satisfactory well. Similar considerations have been used in [5], [6], and [9].

Among other examples of nonprehensile rolling primitives, we can mention the ball and plate (B&P), the ball and beam (B&B), and the “butterfly” system. The B&P is a nonholonomic system for which it has been shown that there exist an admissible path between any two configurations [10]. A PID-based control was proposed in [11] to stabilize the linearized model and a sliding mode controller is instead employed in [12]. The B&P problem is also related to the field of spherical robots [13]. Two planning methods for this class of systems are presented in [14], which are based on minimum energy and time approaches. The B&B system considers the problem of stabilizing the position of a ball along a beam. Since B&B system is not full-feedback linearizable, an approximated partial-feedback linearization (PFL) and output feedback controller were proposed in [15] and [16], respectively. Alternatively, an interconnection and damping assignment passivity-based control (IDA-PBC) design is proposed in [17], a backstepping controller is designed in [18], and a sliding control law is instead proposed in [19]. Nonprehensile

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This paper has supplementary downloadable multimedia material available at <http://ieeexplore.ieee.org> provided by the authors. The supplementary file contains a video that shows experiments of the disk-on-disk system in closed loop with the passivity-based control proposed in the paper. The total size of the file is 12.7 MB. Contact alejandro.donaire@unina.it for further questions about this work.

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rolling systems, where the ball's (or disk's) center of mass does not coincide with its geometric center, are more challenging [20]. The control of a dynamically balanced asymmetrical sphere with three internal rotors is described in [21]. Finally, the so-called "butterfly" juggling action, in which planar rolling is involved, has been investigated in [4], [6], and [22]. A controller for a ball rolling in an asymmetrical bowl that can be accelerated along one linear direction is proposed in [23]. Planning and control problems of a ball rolling on curved surfaces are also investigated in [24] and [25].

In particular, in this brief, we consider the rolling between two circular surfaces as the primitive. The case study is the balancing of a disk that is free to roll on an actuated disk. We refer to the former disk as the object, and the latter as the hand. This system, introduced in [9], can be considered as an example for nonprehensile rolling primitive. The theoretical and technological novelties introduced by this brief are as follows.

- 1) In [5] and [9], the DoD controller is designed using feedback linearization and backstepping, respectively. The backstepping design considers only the stabilization of the object at the upright position, leaving the hand uncontrolled. The feedback linearization controller stabilizes the positions of both the object and the hand. In this brief, we propose to shift the control design to the passivity approach to stabilize both the object and the hand to the desired equilibrium. An important feature of this approach is that the control design exploits the energy and interconnection properties of the physical system without the need of nonlinear cancelations, which in general compromise the robustness of the closed loop. This approach differs from the standard feedback linearization, where a linear dynamics is imposed at the expense of canceling all the nonlinear dynamics of the system.
- 2) The passivity-based controller is developed via energy shaping and damping injection [17]. We follow the approach in [26], where the energy is shaped without the need of solving partial differential equations (PDEs). In [26], the requirement of energy shaping that the closed loop should preserve the form of a mechanical system is dropped. However, we show that the controller proposed here satisfies the so-called *matching equation* [17]. Therefore, the controller belongs to the class of IDA-PBC controllers, and the closed loop can be written in the pH form. The control design here is simpler, since there is no need of solving PDEs.
- 3) In this brief, we develop a less demanding hardware with respect to [5] and [9]. Indeed, the camera and control frame-rate are downgraded from 800 to 75 Hz. In addition, the disks are vertically aligned, and thus the full-gravity field is considered. The experimental results show that the proposed controller design has a good performance on this less demanding setup.
- 4) In the experiments, we do not consider the addition of integral action to robustify the controller as done in [5]. However, the proposed passivity-based controller performs satisfactorily well and cope with

uncertainties without the need of an integral action redesign.

The outline of this brief is as follows. The pH framework is briefly revised in Section II. The dynamic model of the DoD system and the control design are discussed in Section III. Section IV presents simulations of the control system, whilst the experiment results are shown in Section V. Final discussions are provided in Section VI.

II. PORT-HAMILTONIAN FRAMEWORK

The dynamics of a general class of mechanical systems can be described as a pH system as follows:

$$\begin{bmatrix} \dot{q} \\ \dot{p} \end{bmatrix} = \begin{bmatrix} 0 & I_n \\ -I_n & 0 \end{bmatrix} \nabla H(q, p) + \begin{bmatrix} 0 \\ G(q) \end{bmatrix} \tau \quad (1)$$

where $q \in \mathbb{R}^n$ are the generalized coordinates, $p \in \mathbb{R}^n$ are the generalized momenta defined as $p = M\dot{q}$, $\tau \in \mathbb{R}^m$ are the generalized input forces, $G : \mathbb{R}^n \rightarrow \mathbb{R}^{n \times m}$ is the input matrix, and I_n is the n -dimensional identity matrix. The function $H : \mathbb{R}^{n \times n} \rightarrow \mathbb{R}$, which represents the total energy of the system, is the Hamiltonian given by

$$H(q, p) = \frac{1}{2} p^\top M^{-1}(q) p + V(q) \quad (2)$$

where $M : \mathbb{R}^n \rightarrow \mathbb{R}^{n \times n}$ is the mass matrix and $V : \mathbb{R}^n \rightarrow \mathbb{R}$ is the potential energy.

A feature of pH systems is that they are cyclopasive with input τ and output $y = G^\top(q) M^{-1} p$, and storage function $H(q, p)$. Moreover, if $H(q, p)$ is bounded from below, then the pH system is passive [8].

The stabilization problem of the mechanical system (1) using IDA-PBC is to find a control law $\tau = \bar{\tau}(q, p)$ such that the closed loop has a stable equilibrium at the desired point $(q, p) = (q_*, 0)$, with Lyapunov function

$$H_d(q, p) = \frac{1}{2} p^\top M_d^{-1}(q) p + V_d(q) \quad (3)$$

with $M_d > 0$ and $q_* = \arg \min V_d(q)$, and this minimum is isolated. The matrix $M_d : \mathbb{R}^n \rightarrow \mathbb{R}^{n \times n}$ and the function $V_d : \mathbb{R}^n \rightarrow \mathbb{R}$ are, respectively, the desired mass matrix and desired potential energy to be chosen. It is also required that the closed-loop dynamics retain the pH form

$$\begin{bmatrix} \dot{q} \\ \dot{p} \end{bmatrix} = \begin{bmatrix} 0 & M^{-1} M_d \\ -M_d M^{-1} & J_2(q, p) - R(q, p) \end{bmatrix} \nabla H_d(q, p) \quad (4)$$

where J_2 is a skew-symmetric matrix to be chosen, and $R = R^\top \geq 0$ is the damping injection matrix [17].

In the most general case, the control design using IDA-PBC for mechanical systems involves the task of solving a set of PDEs [17]. Indeed, the control law rendering the system (1) in the closed-loop dynamics of the form (4) should satisfy the so-called *matching equation*

$$-\nabla_q H + G\tau = -M_d M^{-1} \nabla_q H_d + (J_2 - R) M_d^{-1} p. \quad (5)$$

However, there exist constructive results to overcome the difficulty of solving the PDEs [26], [27].

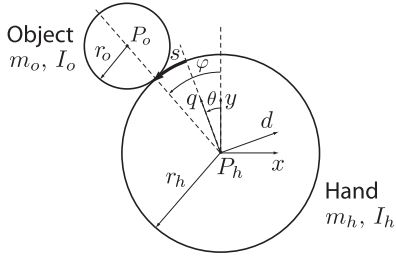


Fig. 1. Idealized physical scheme of the DoD system.

III. ROLLING MANIPULATION

The DoD balancing system is an example of a primitive for nonprehensile manipulation. The system is shown in Fig. 1. The bottom disk represents the manipulator, and the disk on the top is the object to be balanced in the upright position. For the purpose of control design, we make Assumptions A1–A4.

- 1) The hand rotates about its center P_h , whilst there is no translational motion.
- 2) The object is always in point contact with the hand.
- 3) The object rolls on the hand without slipping.

Notice that these assumptions imply that the object cannot depart from the hand. Only from the control design perspective, the problem thus becomes a prehensile manipulation problem. However, the real system does not necessarily satisfy these assumptions and it is intrinsically a nonprehensile system. As it will be shown in the experiments, the controller performs satisfactory even if the assumptions are not *a priori* ensured at all time.

The dynamic model of the DoD, under the assumptions above, has been derived in [9] using the coordinates (θ, s) , where θ is the angle of the hand and s is the length of the arc from the q -axis to the contact point, with the positive convention taken on the counterclockwise direction (see Fig. 1). A detailed formulation of the DOD model in these coordinates can be found in [9].

A. Disk-on-Disk Model

A model, more convenient for our control design, is obtained by expressing the DoD dynamics given in [9] in coordinates (θ, φ) , where φ is the deviation angle of the object from the upright position. The deviation angle is related with θ and s as follows:

$$\varphi = \theta + \frac{s}{r_h}. \quad (6)$$

Using this change of coordinates, the dynamics of the DoD can be equivalently written in the pH form by defining¹ $q = \text{col}(\theta, \varphi)$ and momenta $p = \text{col}(p_1, p_2) = \mathcal{M}\dot{q}$, with

$$\mathcal{M} = \begin{bmatrix} M_{11} & M_{12} \\ M_{21} & M_{22} \end{bmatrix} \quad (7)$$

where $M_{11} = (m_o + m_h)r_h^2$, $M_{12} = M_{21} = -m_or_h(r_o + r_h)$, and $M_{22} = 2m_o(r_h + r_o)^2$, and potential energy function

$$V(\varphi) = c_g \cos(\varphi) \quad (8)$$

where $c_g = m_o g(r_h + r_o)$, g is the gravity constant, and I_h and r_h are the moment of inertia and the radius of the hand,

¹In this brief, a column vector $v \in \mathbb{R}^n$ with entries a_i with $i = 1, \dots, n$ is noted as $v = \text{col}(a_1, \dots, a_n)$.

respectively. The moment of inertia, radius, and mass of the object is noted as I_o , r_o , and m_o , respectively.

The dynamics in pH form is as follows:

$$\begin{bmatrix} \dot{q} \\ \dot{p} \end{bmatrix} = \begin{bmatrix} 0 & I_n \\ -I_n & 0 \end{bmatrix} \nabla \mathcal{H}(q, p) + \begin{bmatrix} 0 \\ \mathcal{G} \end{bmatrix} \tau_h \quad (9)$$

with $\mathcal{G} = \text{col}(1, 0)$ and Hamiltonian

$$\mathcal{H}(q, p) = \frac{1}{2} p^\top \mathcal{M}^{-1} p + V(\varphi). \quad (10)$$

Contact Forces: The DoD model (9) relies on the assumptions that the object and the hand are always in contact and there is no slipping. As discussed in [28, Ch. 5], by using a simple *Coulomb model* for frictional forces (see [29] for more elaborated models), the conditions on the normal and frictional forces that ensure rolling can be written as

$$f_n > 0 \quad (11)$$

$$|f_f| \leq \mu f_n \quad (12)$$

where f_n and f_f are the normal and frictional forces, respectively, and $\mu > 0$ is the frictional coefficient. Furthermore, the normal and frictional forces can be written as follows:

$$f_n = m_o(\ddot{y}_o + g) \cos(\varphi) - m_o\ddot{x}_o \sin(\varphi) \quad (13)$$

$$f_f = m_o(\ddot{y}_o + g) \sin(\varphi) + m_o\ddot{x}_o \cos(\varphi) \quad (14)$$

where x_o and y_o are the components of the center of the object P_o . Using the DoD dynamics, the expressions (13) and (14) can also be written as follows:

$$f_n = m_o g \cos(\varphi) - m_o(r_h + r_o)\dot{\varphi}^2 \quad (15)$$

$$f_f = \frac{m_o(r_h + r_o)}{M_{11}M_{22} - M_{12}^2} [M_{12}\tau_h - M_{11}c_g \sin(\varphi)] + m_o g \sin(\varphi). \quad (16)$$

The control design in this brief, as in [5], does not consider explicitly the constraints (11) and (12). However, we will show later that the model assumptions are satisfied by computing the normal and frictional forces using data from experiments and verifying that the constraints are met.

B. Control Objective

We aim at designing an IDA-PBC controller to stabilize the DoD system at the equilibrium given by $q_\star = \text{col}(\theta_\star, \varphi_\star)$ and $p_\star = \text{col}(0, 0)$. There are two classes of equilibrium points of interest. The first class corresponds to the equilibria, where the final position of the hand is not of interest, that is $q_{\star 1} = \text{col}(\bar{\theta}, 0)$ with $\bar{\theta}$ any constant angle of the hand. The second class of equilibria corresponds to the case, where the final position of the hand is specified $q_{\star 2} = \text{col}(\theta_\star, 0)$ with θ_\star the desired position of the hand. Note that the equilibrium on the momentum vector p implies $\dot{\theta}_\star = 0$ and $\dot{\varphi}_\star = 0$.

C. Control Design

In this brief, we follow the approach proposed in [26] and design the controller for the DoD system via IDA-PBC without solving PDEs in the pH framework. The main idea in [26] is to design a controller in two steps. First, a PFL as proposed in [30] is performed. This PFL controller preserves, under

certain assumptions, the Hamiltonian structure. This fact has been first shown in [31] and is fundamental for the design in [26]. Second, two passive outputs are identified to build a Lyapunov energy-like candidate function to design a controller that stabilizes the system about an equilibrium point.

We recall here the assumptions made in [26] and [31]. We consider a mechanical system with generic coordinates $\mathbf{q} = (q_a, q_u)$, where q_a and q_u are the actuated and unactuated coordinates, respectively. The mass matrix is written as

$$\mathbf{M}(\mathbf{q}) = \begin{bmatrix} m_{aa} & m_{au} \\ m_{au}^\top & m_{uu} \end{bmatrix}. \quad (17)$$

Then, we make the following assumptions.

- A1.** The inertia matrix depends only on the unactuated variables q_u , i.e., $\mathbf{M}(\mathbf{q}) = \mathbf{M}(q_u)$.
- A2.** The subblock matrix m_{aa} is constant.
- A3.** The potential energy can be written as $\mathbf{V}(\mathbf{q}) = V_a(q_a) + V_u(q_u)$.
- A4.** The rows of the matrix $m_{au}(q_u)$ satisfy

$$\frac{\partial(m_{au})_k}{\partial q_{uj}} = \frac{\partial(m_{au})_j}{\partial q_{uk}} \quad \forall j \neq k, \quad j, k \in \mathcal{I} := \{1, \dots, n - m\}.$$

The model of the DoD in coordinates (θ, s) used in the previous works [5], [9] does not satisfy Assumption A3, and therefore, the approach in [26] cannot be applied. However, the change of coordinates $(\theta, s) \rightarrow (\theta, \varphi)$ allows transforming the dynamic model such that it satisfies Assumptions A1–A4. Therefore, we use the dynamics (9) to design the controller.

To design the control law, we first apply a PFL controller (see [30] for details)

$$\tau_h = \left[M_{11} - \frac{M_{12}^2}{M_{22}} \right] u + \frac{M_{12}}{M_{22}} c_g \sin(\varphi) \quad (18)$$

to obtain a pH system as follows:

$$\begin{bmatrix} \dot{q} \\ \dot{p} \end{bmatrix} = \begin{bmatrix} 0 & I_n \\ -I_n & 0 \end{bmatrix} \nabla H(q, p) + \begin{bmatrix} 0 \\ G(q) \end{bmatrix} u \quad (19)$$

where $q = \text{col}(\theta, \varphi)$, $p = \text{col}(p_\theta, p_\varphi) = M(q)\dot{q}$, and $G = \text{col}(1, -M_{12})$, and Hamiltonian

$$H = \frac{1}{2} p^\top M^{-1} p + V(q) \quad (20)$$

with $M = \text{diag}(1, M_{22})$. Note that there is a momentum transformation $\mathbf{p} \rightarrow p$ defined as $p = M\mathcal{M}^{-1}\mathbf{p}$, and a new (acceleration) control input u .

As observed in [26], the PFL produces, under Assumptions A1–A4, two passive outputs $y_\theta = p_\theta$ and $y_\varphi = -M_{12}M_{22}^{-1}p_\varphi$. Indeed, considering the storage functions $H_\theta = (1/2)p_\theta^2$ and $H_\varphi = (1/2)M_{22}^{-1}p_\varphi^2 + c_g \cos(\varphi)$, we obtain that their time derivatives are

$$\dot{H}_\theta = y_\theta u, \quad \dot{H}_\varphi = y_\varphi u,$$

which ensure passivity. Note that Assumptions A1–A4 are required to ensure the existence of these two new passive outputs via PFL.

We propose a desired closed-loop energy function H_d as follows:

$$H_d = k_e[k_a H_\theta + k_u H_\varphi] + \frac{1}{2} K_k (k_a y_\theta + k_u y_\varphi)^2 + \frac{1}{2} K_I \left[\int_t (k_a y_\theta + k_u y_\varphi) dt \right]^2 \quad (21)$$

which is built using the storages energies H_θ and H_φ , a weighted sum of the passive outputs y_θ and y_φ , and its integral. The expression (21) can be explicitly written as a state function by substituting H_θ , H_φ , y_θ , and y_φ by their expression as the functions of the states. After straightforward calculations, we obtain that the desired closed-loop energy can be written as in (3) with desired mass matrix

$$M_d^{-1} = \begin{bmatrix} k_e k_a + k_a^2 K_k & -k_a k_u K_k M_{12} M_{22}^{-1} \\ -k_a k_u K_k M_{22}^{-1} M_{12} & k_e k_u M_{22}^{-1} + k_u^2 K_k M_{12}^2 M_{22}^{-2} \end{bmatrix} \quad (22)$$

and desired potential function

$$V_d(q) := k_e k_u c_g \cos(\varphi) + \frac{K_I}{2} [k_a \theta - k_u M_{12}(\varphi + c)]^2 \quad (23)$$

with k_e , k_a , k_u , K_k , K_I , and c constant parameters to be chosen such that $M_d(q) > 0$, and $q_* = \arg \min V_d(q)$, and the minimum is isolated.

Proposition 1: Consider the system (19) in closed loop with the control law

$$u = -K^{-1} \left[k_u K_k \frac{M_{12}}{M_{22}} \nabla_\varphi V + K_I [k_a \theta - k_u M_{12}(\varphi + c)] \right] - K^{-1} K_p \left[k_a p_\theta - k_u \frac{M_{12}}{M_{22}} p_\varphi \right] \quad (24)$$

with $K = k_e + k_a K_k + k_u K_k ((M_{12}^2)/(M_{22}))$ and $K_p > 0$. The constants k_e , k_a , k_u , K_k , K_I , and c satisfy

$$k_a(k_e + k_a K_k) > 0, \quad k_a k_e k_u K > 0 \quad (25)$$

$$K_I k_a^2 > 0, \quad -k_e k_u > 0 \quad (26)$$

$$K \neq 0 \quad (27)$$

$$c = \frac{k_a}{k_u M_{12}} \theta_*. \quad (28)$$

Then, the following statements hold.

- 1) The equilibrium $q_{*1} = \text{col}(\bar{\theta}, 0)$ of the closed loop, with $K_I = 0$, is asymptotically stable.
- 2) The equilibrium $q_{*2} = \text{col}(\theta_*, 0)$ of the closed loop, with $K_I \neq 0$, is asymptotically stable.

Proof: First, we note that the conditions imposed on the parameters ensure that the desired mass matrix M_d is positive definite, and q_{1*} and q_{2*} are minima of the potential energy V_d with $K_I = 0$ and $K_I \neq 0$, respectively. Indeed, using Sylvester's criterion, the condition of the matrix M_d is satisfied if and only if (25) holds. The minimum conditions on V_d are as follows.

- 1) $\nabla V_d(q)|_{q=q_*} = 0$

$$\Leftrightarrow \begin{bmatrix} K_I k_a^2 \theta - K_I k_a k_u M_{12}(\varphi + c) \\ -k_e k_u c_g \sin(\varphi) - K_I k_u M_{12} [k_a \theta - k_u M_{12}(\varphi + c)] \end{bmatrix}_{q=q_*} = 0$$

which is satisfied for q_{*1} if $K_I = 0$, and for q_{*2} with c as in (28).

$$2) \nabla^2 V_d(q)|_{q=q_*} > 0$$

$$\Leftrightarrow \begin{bmatrix} K_I k_a^2 & -K_I k_a k_u M_{12} \\ -K_I k_a k_u M_{12} & -k_e k_u c_g \cos(\varphi) + K_I k_u^2 M_{12}^2 \end{bmatrix}_{q=q_*} > 0$$

which is satisfied provided that (26) holds true. The condition (27) ensures that the controller is well-defined.

The stability of the closed loop is shown by choosing H_d as a Lyapunov candidate function. Then, we compute its time derivative along the dynamics (19) as follows:

$$\begin{aligned} \dot{H}_d &= (k_a p_\theta - k_u M_{12} M_{22}^{-1} p_\varphi) \\ &\quad \times \left[k_e u + k_a K_k u - k_u K_k \frac{d}{dt} [-M_{12} M_{22}^{-1} p_\varphi] \right. \\ &\quad \left. + K_I [k_a \theta - k_u M_{12} (\varphi + c)] \right] \\ &= \left[k_a p_\theta - k_u \frac{M_{12}}{M_{22}} p_\varphi \right] \\ &\quad \times \left\{ \left[k_e + k_a K_k + k_u K_k \frac{M_{12}^2}{M_{22}} \right] u \right. \\ &\quad \left. + \left[k_u K_k \frac{M_{12}}{M_{22}} \nabla_\varphi V + K_I [k_a \theta - k_u M_{12} (\varphi + c)] \right] \right\}. \end{aligned}$$

By using the control law u as proposed in (24), we obtain

$$\dot{H}_d = -K_P \left[k_a p_\theta - k_u \frac{M_{12}}{M_{22}} p_\varphi \right]^2 \quad (29)$$

which ensures stability of the equilibrium. Asymptotic stability follows using the invariance principle and standard Lyapunov theory [32]. Indeed, consider the set $\mathcal{S} = \{(q, p) | H_d = 0\}$. Then, we obtain that $k_a M_{22} p_\theta = k_u M_{12} p_\varphi$ holds in \mathcal{S} , and by differentiating this equality respect to time, we obtain

$$k_a \dot{p}_\theta = k_u \frac{M_{12}}{M_{22}} \dot{p}_\varphi. \quad (30)$$

The input takes the form $u = ((-k_u M_{12} c_g) / (M_{22}(k_a + k_u M_{12}))) \sin(\varphi)$ from which, by comparing with (24), yields that φ is constant. Therefore, $p_\varphi = 0$ and $p_\theta = 0$. Also, the dynamics restricted to \mathcal{S} implies that θ is constant. Moreover, since $\dot{p}_\theta = 0$, then $u = 0$ and $\varphi = 0$, and from (24), we obtain that $\theta = \theta_*$ when $K_I \neq 0$, otherwise $\theta = \bar{\theta}$. This proves asymptotic stability of the desired equilibrium. $\square\square\square$

Remark 1: It can be shown that the complexity of the controller τ_h given in (18) with u as in (24) is the same as the one of a controller obtained by using the classical IDA-PBC design, which involves the task of solving the PDEs of the matching equations.

Remark 2: The structure of the proposed controller that asymptotically stabilizes both equilibria q_{*1} or q_{*2} is the same. The only difference being the gain K_I , which is set to zero to stabilize q_{*1} , and $K_I \neq 0$ to stabilize q_{*2} .

D. Closed-Loop Dynamics

In this section, we show that the closed-loop dynamics of the DoD has the form (4), therefore the control law design in Section III-C is an IDA-PBC controller. Notice that the requirement on the closed-loop dynamics was not considered in [26], however, here, we prove that the closed loop actually preserves the pH form.

Proposition 2: Consider the dynamics of the DoD (9) in closed loop with the controller (18), with u as in (24). Then, the closed-loop dynamics has the pH form (4).

Proof: First, we note that the partial-feedback control in (18) renders the system in the form (19). Then, we analyze system (19) in closed loop with the inner controller u . The closed-loop dynamics has the form (4) if u satisfies

$$-\nabla_q H + G(q)u = -M_d M^{-1} \nabla_q H_d + (J_2 - R) \nabla_p H_d. \quad (31)$$

Using H from (20), H_d from (3) with desired mass matrix and potential function (22) and (23), respectively, and $G = \text{col}(1, -M_{12})$, we obtain

$$-\nabla_q V + Gu = -M_d M^{-1} \nabla_q V_d + (J_2 - R) M_d^{-1} p.$$

We split the control input (24) in $u = u_1 + u_2$ with

$$u_1 = -K^{-1} \left[k_u K_k \frac{M_{12}}{M_{22}} \nabla_\varphi V + K_I [k_a \theta - k_u M_{12} (\varphi + c)] \right] \quad (32)$$

and

$$u_2 = -K^{-1} K_p \left[k_a p_\theta - k_u \frac{M_{12}}{M_{22}} p_\varphi \right]. \quad (33)$$

Then, we will prove that

$$Gu_1 = \nabla_q V - M_d M^{-1} \nabla_q V_d \quad (34)$$

$$Gu_2 = (J_2 - R) M_d^{-1} p. \quad (35)$$

From (34), we obtain

$$\begin{aligned} M_d^{-1} Gu_1 &= M_d^{-1} \nabla_q V - M^{-1} \nabla_q V_d \\ \begin{bmatrix} k_a \\ -k_u \frac{M_{12}}{M_{22}} \end{bmatrix} K u_1 &= \begin{bmatrix} k_a \\ -k_u \frac{M_{12}}{M_{22}} \end{bmatrix} \\ &\quad \times \left[-k_u K_k \frac{M_{12}}{M_{22}} \nabla_\varphi V \right. \\ &\quad \left. - K_I [k_a \theta - k_u M_{12} (\varphi + c)] \right] \end{aligned}$$

which is clearly satisfied with u_1 as in (32). From (35) and by setting $J_2 = 0$ and $R = GK^{-1}K_p K^{-1}G^\top$, we obtain

$$\begin{aligned} Gu_2 &= GK^{-1}K_p K^{-1}G^\top M_d^{-1} p \\ Gu_2 &= GK^{-1}K_p \begin{bmatrix} k_a & -k_u \frac{M_{12}}{M_{22}} \end{bmatrix} p \end{aligned}$$

which is satisfied with u_2 as in (33). Therefore, the closed loop has the form (4) as claimed. $\square\square\square$

Remark 3: Notice that the DoD model used to design the controller does not include damping forces. These forces can compromise the passivity of the closed loop if they are not

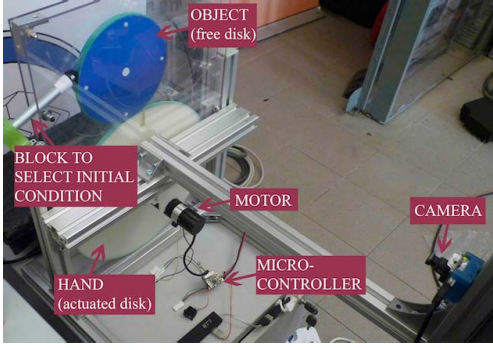


Fig. 2. Prototype of the DoD available at PRISMA Lab.

handled properly. Future research will aim to extend the design approach used here for the case when general damping forces are present in the model.

IV. SIMULATIONS

In this section, we present simulations to assess the performance of the controller proposed in Section III-C. We build a simulator of the dynamic model (9) in closed loop with the control law τ_h given in (18). The parameters of the DoD are $r_h = 0.15$ m, $r_o = 0.075$ m, $m_h = 0.335$ Kg, $m_o = 0.220$ Kg, and $g = 9.81$ Kg/m/s², which correspond to the prototype shown in Fig. 2 available at PRISMA Lab and whose detailed description is available in Section V.

The simulations are performed under the following scenario: the DoD starts at rest, and the initial conditions of the balancing and hand angles are $\varphi(0) = 7^\circ$ and $\theta = 0^\circ$, respectively. The desired equilibrium is $q_* = (0, 0)$, and the controller parameters are $k_a = 0.04$, $k_u = -100$, $k_e = 0.01$, $K_k = 1$, $K_p = 10$, and $K_I = 30$. To enhance the realism to the simulation, we have added noise, a zero-order hold and a time delay of about 0.013 s to the measurements. Also, we emulate parameter uncertainties by using different values for the model parameter in the controller and the DoD model (5% deviation). The time histories of the hand and balancing angles are shown in Fig. 3. The plots show that the hand angle converges to the desired set point, while the object is being balanced on the upright position. Notice that the convergence of the hand angle is slower than the settling time of the balancing angle. There is a small oscillation about the equilibrium due to the measurement noise; however, the control system remains stable. Fig. 4 shows the time histories of the balancing and hand angular velocities, as well as the control torque. It can be seen that the trajectories of the velocities are smooth and within acceptable values, and the torque demanded is reasonably smooth and remains bounded by values achievable in a realistic scenario.

V. EXPERIMENTS

A. Experimental Setup

We evaluate the performance of the controller in (18) by using the experimental prototype shown in Fig. 2. The lower disk is actuated by a dc motor (Harmonic Drive RH-8D 3006) equipped with a harmonic drive, whose gearhead ratio is 100 : 1, and a 500-p/r quadrature encoder. A rubber band

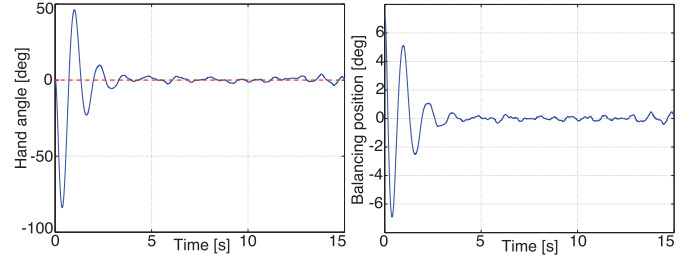


Fig. 3. Time histories of the hand angle $\theta(t)$ (left-hand side) and the balancing angle $\varphi(t)$ (right-hand side). The initial conditions are $\theta(0) = 0^\circ$ and $\varphi(0) = 7^\circ$. The hand reference is $\theta_* = 0$.

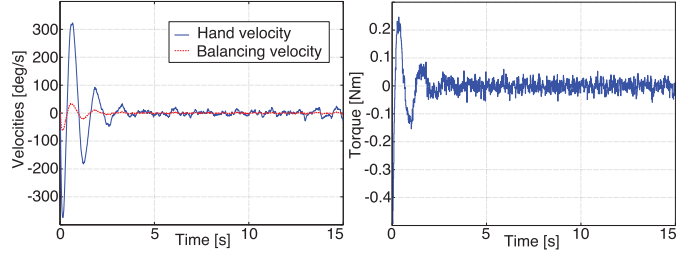


Fig. 4. Time histories of the balancing and hand angular velocities, $\dot{\varphi}(t)$ and $\dot{\theta}(t)$, respectively (left-hand side), and time history of the input torque $\tau_h(t)$ (right-hand side).

of about 1 mm encircles both disks to avoid slipping. The motor commands are provided by an ARM CORTEX M3 microcontroller (32 b, 75 MHz). This microcontroller receives current references from a PC through an universal serial bus. The microcontroller outputs the current reference for the motor servo, which provides the torque to the hand disk. Therefore, we transform the torque τ_h in (18) in a current control, and we implement an inner-loop current controller written as [5]

$$i_{com} = \frac{\tau_h + \mu_d \dot{\theta} + f_s \text{sgn}(\dot{\theta})}{k_m} + k_p(\hat{\theta} - \theta) + k_d(\dot{\hat{\theta}} - \dot{\theta}) \quad (36)$$

where $\hat{\theta}$ and $\dot{\hat{\theta}}$ are the desired hand position and velocity, respectively, obtained by integrating (24) at each sample time, while τ_h is given by (18). The parameters k_p and k_d are gains, which were set to 10 and 1, respectively, $k_m = 4.20$ is the motor constant available from the motor data sheet, $\mu_d = 0.29$ is a viscous friction coefficient, and $f_s = 0.3$ is the torque required to overcome friction from rest. The values of μ_d and f_s were found by experiments [33]. The inner-loop current controller of the servo motor runs at a sample rate of 4 KHz. In addition, the microcontroller also provides the measurement of the hand position. The control algorithm, which is written in C++, runs on the external PC with a Linux-based operating system. The position of the object is provided by an external visual system. This visual system consists of an uEye UI-122-xLE camera providing 376×240 pixel images to the PC at 75 Hz, which is also the controller sample rate. In order to speed up computations, a 15×15 pixel region of interest is employed by the image elaboration algorithm running on the same external PC. Notice that the setup is mounted in full gravity between two plexiglass plates. This differs from [5] and [9], where the gravity is weakened.

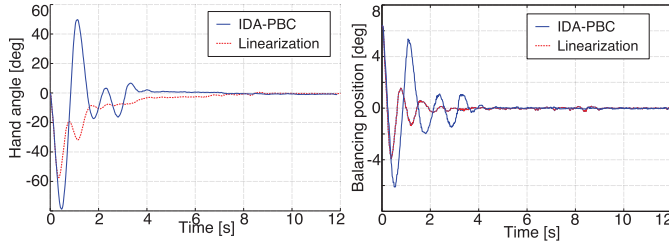


Fig. 5. Time histories of the hand angle $\theta(t)$ (left-hand side) and the balancing angle $\varphi(t)$ (right-hand side). The initial condition is $\theta(0) = 0^\circ$ and $\varphi(0) = 6.3^\circ$, and the hand reference is $\theta_\star = 0$.

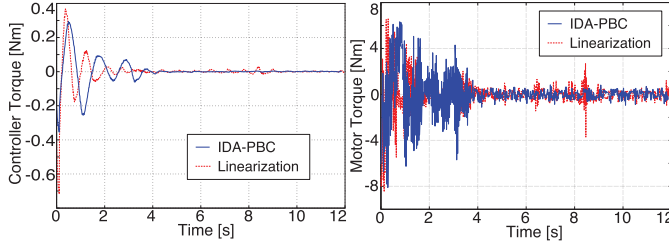


Fig. 6. Time histories of the torque commanded by the controller $\tau_h(t)$ (left-hand side) and the motor torque, including compensations, $\tau_m(t) = k_m i_{com}(t)$ (right-hand side).

B. Case Studies

In the first experiment, we set the reference for the hand angle to zero, while the object is desired to be stabilized at the upright position. Under this scenario, we run the experiment using two controllers: the controller proposed in Section III-C, noted as IDA-PBC, and the full-state exact-feedback linearization controller proposed in [5], noted as linearization. We have tuned the parameters of the controllers at the best of our possibilities. The parameters of the IDA-PBC controller used in the experiments are $k_a = 0.004$, $k_u = -100$, $k_e = 0.01$, $K_k = 1$, $K_P = 5$, and $K_I = 30$; and the parameters of the feedback linearization controller are $k_1 = 100$, $k_2 = 55$, $k_3 = 117$, and $k_4 = 10$. The left-hand side of Fig. 5 shows that both controllers ensure that the hand converges to the desired position. The plot shows that the IDA-PBC controller drives the hand to the set-point reference faster than the feedback linearization controller, which produces a slow speed of convergence. This advantage is at the expense of a larger overshoot on the balance angle. Intuitively, the larger the hand angle motion, the larger the induced balancing angle motion. As can be seen on the right-hand side of Fig. 5, both controllers satisfactory balance the object at the upright position. It seems, however, that the closed-loop performance with fast hand convergence and large overshoot or slower hand convergence and small overshoot could be achieved by both controllers if their gain values are appropriately chosen. Fig. 6 shows the command torques computed by the IDA-PBC and feedback linearization controllers, and the torques produced by the motor and delivered to the hand. Notice that the inner-loop controller produced the necessary torque to compensate the friction forces and unmodeled dynamics of the dc motor. Finally, the normal and frictional forces, computed using (15) and (16) and the experimental data, are shown in Fig. 7. The same figure also displays the minimum

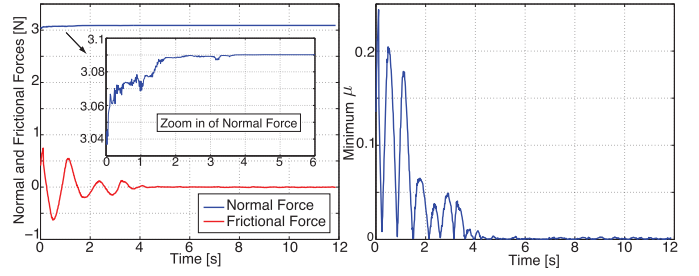


Fig. 7. Time histories of the normal and frictional forces $f_n(t)$ and $f_f(t)$ (left-hand side) and the lower bound for the frictional coefficient $\mu_{\min}(t)$ (right-hand side).

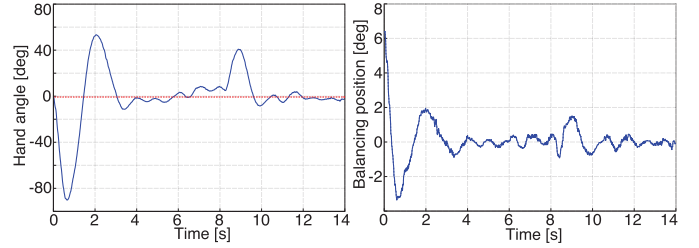


Fig. 8. Time histories of the hand angle $\theta(t)$ (left-hand side) and the balancing angle $\varphi(t)$ (right-hand side). The initial conditions are $\theta(0) = 0^\circ$ and $\varphi(0) = 6.3^\circ$, and the hand reference is $\theta_\star = 0$. A nonpersistent disturbance is applied at about $t = 8s$.

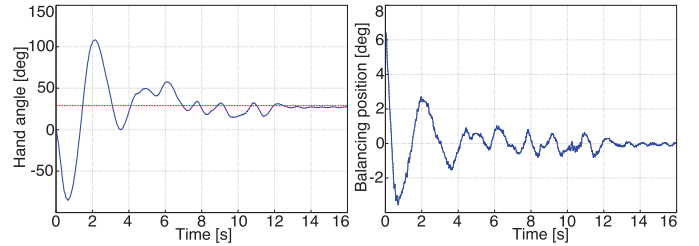


Fig. 9. Time histories of the hand angle $\theta(t)$ (left-hand side) and the balancing angle $\varphi(t)$ (right-hand side). The initial condition is $\theta(0) = 0^\circ$ and $\varphi(0) = 6.3^\circ$, and the hand reference is $\theta_\star = 30^\circ$.

frictional coefficient $\mu_{\min} = (|f_f|)/(f_n)$ needed to ensure the rolling assumption. Since the frictional coefficient for the setup is $\mu > 2$, which is obtained empirically, we infer that the constraints (11) and (12) are satisfied, and the rolling assumption is met.

In the second experiment, we test the IDA-PBC controller recovering when a nonpersistent disturbance acts on the object. The initial conditions are the same as in the first experiment, but now the object is pushed at time about 8 s. Fig. 8 shows the time history of the hand and balancing angles. Under this scenario, the plots show that the controller recovers the desired equilibrium.

In the third experiment, we change the set point of the hand angle, and we set the desired equilibrium to $(\theta_\star, \varphi_\star) = (30^\circ, 0)$. The initial conditions are the same as in the first experiment. The time histories of the hand and balancing angles are shown in Fig. 9.

The performance of the controller can be visualized in a multimedia video recorded while performing the experiments (available at https://youtu.be/oQ8hS6Hm_e4).

VI. CONCLUSION AND FUTURE WORK

We have investigated an IDA-PBC controller to balance the upright unstable position of the nonprehensile DoD system. We have developed the controller using passivity and pH theory. The controller has been designed without the need of solving PDEs. In addition, we have proved that the closed-loop dynamics retain the pH form. Simulations and experiments bolster the applicability of the presented theory in practice. In the future work, we will focus on the control design of other nonprehensile manipulation primitives and their integration using passivity properties.

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