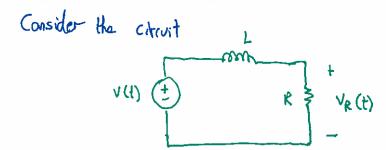
## Frequency dependent circuits



and let v(t) be as follows

$$V(t) = V_m \cos(\omega t + 0^\circ)$$

constant
value

(keep as a variable)

In terms of phasors and complex impedances,



This is a simple voltage divider

We may write  $V_R = \frac{1}{1 + j\omega(L/R)} \times V_m$ 

Here, the magnitude and phase of  $\overline{V}_R$  are frequency dependent. This dependence is called frequency response.

In the time domain,

$$V_R(t) = V_R \cos(\omega t + \sigma_R)$$
  
amplifude  $|V_R|$  at  $\omega$  phase of  $V_R(t)$   
frequency  $\omega$ 

The amplitude 
$$|\vec{v}_R| = V_R = \frac{1}{1+j(\omega L/R)} \times V_m$$

$$V_{R} = \frac{V_{m}}{\left[1 + j\omega(L/R)\right]} = \frac{V_{m}}{\left[\left(\omega L/R\right)^{2} + l^{2}\right]}$$

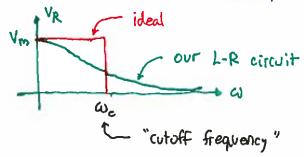
Graph this:



This is called a largess filter.

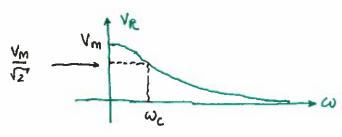
- · Situsoids with low frequencies come through strongly.
- . higher frequencies come through at reduced amplifudes,

An ideal lowpass filter has the frequency response



For the ideal filter,

We usually define the cutoff frequency as the frequency at which VR = (1/12) x Vm.



for our L-R circuit,

The cutoff frequency for this filter is when

$$V_{R} = \frac{V_{m}}{\sqrt{2}} = \frac{V_{m}}{\sqrt{(\omega_{c}/R)^{2} + 1}}$$

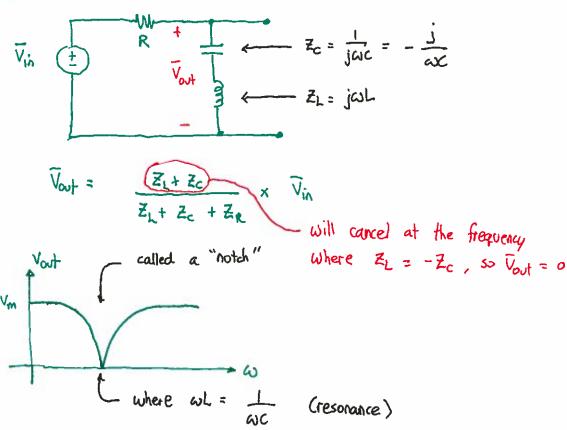
$$SO \left(\omega_{c}L/R\right)^{2} + 1 = 2$$

$$and \left(\omega_{c}L/R\right) = 1$$

Thus,  $\omega_c = R/L$  rads/sec, and  $f_c = \omega_c/2\Pi$ .

There are many other extremely useful filter circuits.

One such bandstop fitter circuit that exploits the series cancellation of impedances is:



This is called "notch filter" and has lots of important applications. E.g., "hum" in an audio system is due to 60 Hz power source.