

TI0153 - Trabalho de Conclusão de Curso II  
Department of Teleinformatics Engineering  
Federal University of Ceará - UFC

# **Optimal Control: An application to a non-isothermal continuous reactor**

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## Summary

### Introduction

- Contextualization
- Problem Definition

### Dynamical System Analysis

- Dynamical Models for Chemical Reactors
- General Properties of State-Space Models
- Experiments

### State-Feedback Controllers

- Definitions
- Regulation vs. Tracking

### Optimal Control

- Formulation
- Linear Quadratic (LQ) Controllers
- Simulations

### Optimal State Estimation

- Formulation
- Kalman Filter and LQG Controllers
- Simulations

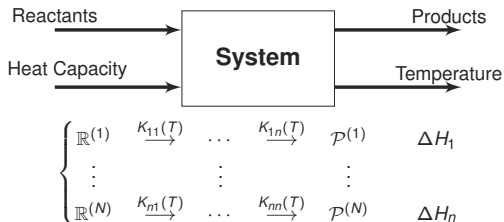
### Conclusion



# Introduction



- We are discussing the **automatic control** of **dynamical systems**;
- We are discussing **chemical reactor network systems**, for processes described as:

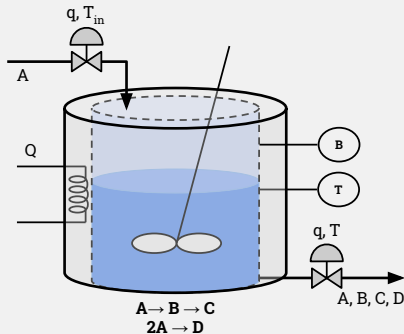


- Optimal control theory has been revisited from several innovative fields in the last years [Cairano et al., 2014, Tang et al., 2016, Eren et al., 2017]. Furthermore, reactor systems are subject of active research, with several open challenges [Gupta et al., 2012, Mulas et al., 2015].



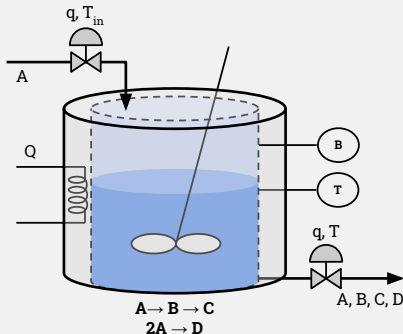
- The system used for the experiments was the non-isothermal Continuous Stirred Tank Reactor (CSTR) presented by [Klatt and Engell, 1998].

## Non-isothermal CSTR



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## Non-isothermal CSTR



### This system...

- characterizes a wide range of industrial applications;
- is a classical benchmark for multiple-input multiple-output (MIMO) control systems;
- is highly nonlinear, with non-minimum phase behavior and unmeasurable states;



# Dynamical System Analysis



- Dynamical systems are modelled through the Conservation Laws from physics;

## Mass Balance for Chemical Reactors

$$\begin{aligned} \left( \begin{array}{c} \text{Accumulation} \\ \text{of mass} \\ \text{in the system} \end{array} \right) &= \left[ \left( \begin{array}{c} \text{Mass flow} \\ \text{entering} \\ \text{system} \end{array} \right) + \left( \begin{array}{c} \text{Mass} \\ \text{produced} \\ \text{by reactions} \end{array} \right) \right] \\ &\quad - \left[ \left( \begin{array}{c} \text{Mass flow} \\ \text{leaving} \\ \text{system} \end{array} \right) + \left( \begin{array}{c} \text{Mass} \\ \text{consumed} \\ \text{by reactions} \end{array} \right) \right] \end{aligned} \quad (1)$$

## Conservation of Energy for Chemical Reactors

$$\left( \begin{array}{c} \text{Accumulation} \\ \text{of thermal energy} \\ \text{in the system} \end{array} \right) = \left( \begin{array}{c} \text{Heat flow} \\ \text{entering} \\ \text{the system} \end{array} \right) - \left( \begin{array}{c} \text{Heat flow} \\ \text{leaving} \\ \text{the system} \end{array} \right) + \left( \begin{array}{c} \text{Entropy} \\ \text{contribution} \\ \text{from reactions} \end{array} \right) \quad (2)$$





- Assuming a continuous reactor of constant volume, comprised by a dilute solution of reactant and products...

## Mass Balance for Chemical Reactors

$$\frac{d(\rho_A)}{dt} = q(\rho_{in}^{(A)} - \rho_{out}^{(A)}) + \left( \sum_{\alpha X \rightarrow \beta A} \frac{1}{\beta} K_{XA}(T) (\rho_X)^\alpha \right) - \left( \sum_{\alpha A \rightarrow \beta X} \frac{1}{\beta} K_{AX}(T) (\rho_A)^\alpha \right) \quad (3)$$

## Conservation of Energy for Chemical Reactors

$$\begin{cases} \frac{d(T)}{dt} = q(T_{in} - T_{out}) + \eta(T_C - T) + \delta \sum_{\alpha A \rightarrow \beta X} K_{AX}(T) (\rho_A)^\alpha \Delta H_{AX} \\ \frac{d(T_C)}{dt} = \gamma Q + \beta(T - T_C) \end{cases} \quad (4)$$

- **Remark:** this model is highly nonlinear!



## Linearization by Taylor Series Expansion

Consider a nonlinear time-invariant system:

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{f}(\mathbf{x}(t), \mathbf{u}(t)) \\ \mathbf{y}(t) = \mathbf{g}(\mathbf{x}(t), \mathbf{u}(t)) \end{cases} \quad (5)$$

Given steady-state operating points  $\mathbf{x}_o$ ,  $\mathbf{y}_o$  and  $\mathbf{u}_o$ , the dynamics of the system in the neighborhood of these points can be represented by the linear model:

$$\begin{cases} \Delta \dot{\mathbf{x}}(t) = \mathbf{A} \Delta \mathbf{x}(t) + \mathbf{B} \Delta \mathbf{u}(t) \\ \mathbf{y}(t) = \mathbf{C} \Delta \mathbf{x}(t) + \mathbf{D} \Delta \mathbf{u}(t) \end{cases} \quad (6)$$

where

$$\mathbf{A} = \left. \frac{\partial \mathbf{f}}{\partial \mathbf{x}} \right|_{\mathbf{x}_o, \mathbf{u}_o}; \quad \mathbf{B} = \left. \frac{\partial \mathbf{f}}{\partial \mathbf{u}} \right|_{\mathbf{x}_o, \mathbf{u}_o}; \quad \mathbf{C} = \left. \frac{\partial \mathbf{g}}{\partial \mathbf{x}} \right|_{\mathbf{x}_o, \mathbf{u}_o}; \quad \mathbf{D} = \left. \frac{\partial \mathbf{g}}{\partial \mathbf{u}} \right|_{\mathbf{x}_o, \mathbf{u}_o} \quad (7)$$

and

$$\Delta \mathbf{x}(t) = \mathbf{x}(t) - \mathbf{x}_o; \quad \Delta \mathbf{u}(t) = \mathbf{u}(t) - \mathbf{u}_o. \quad (8)$$



- ▶ A linear and time-invariant model in State-Space representation has the form:

$$\begin{cases} \dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u} \\ \mathbf{y} = \mathbf{C}\mathbf{x} + \mathbf{D}\mathbf{u} \end{cases} \quad (9)$$

This representation has several analysis and control design advantages...

- ▶ The response of the system has a practical analytical solution:

$$\begin{cases} \mathbf{x}(t) = e^{\mathbf{A}(t-t_0)}\mathbf{x}(t_0) + \int_{t_0}^t e^{\mathbf{A}(t-\tau)}\mathbf{B}\mathbf{u}(\tau)d\tau \\ \mathbf{y}(t) = \mathbf{C}e^{\mathbf{A}(t-t_0)}\mathbf{x}(t_0) + \mathbf{C} \int_{t_0}^t e^{\mathbf{A}(t-\tau)}\mathbf{B}\mathbf{u}(\tau)d\tau + \mathbf{D}\mathbf{u}(t) \end{cases} ; \quad (10)$$

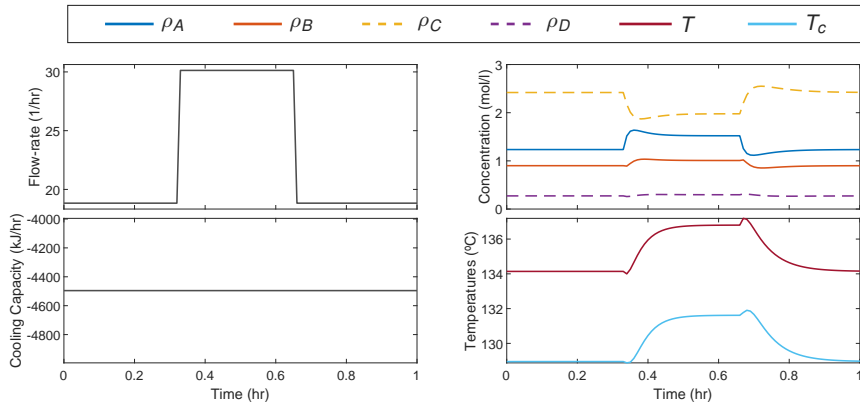
- ▶ The controllability property is directly related to the *Controllability Matrix*:

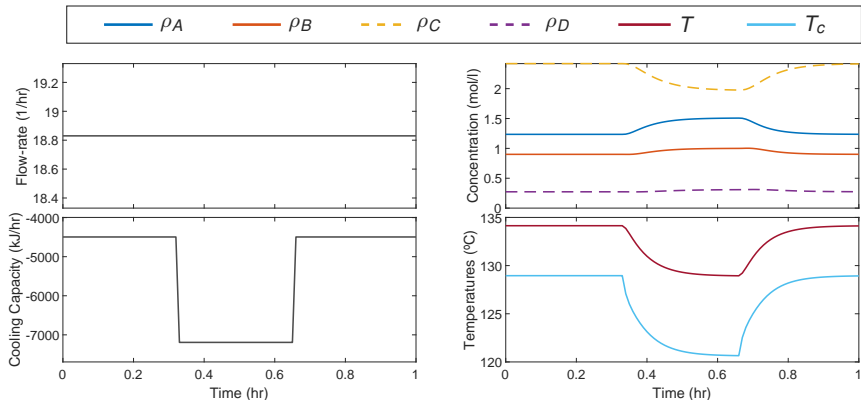
$$\mathbf{C} = [\mathbf{B} \quad \mathbf{A}\mathbf{B} \quad \mathbf{A}^2\mathbf{B} \quad \dots \quad \mathbf{A}^{n-1}\mathbf{B}] ; \quad (11)$$

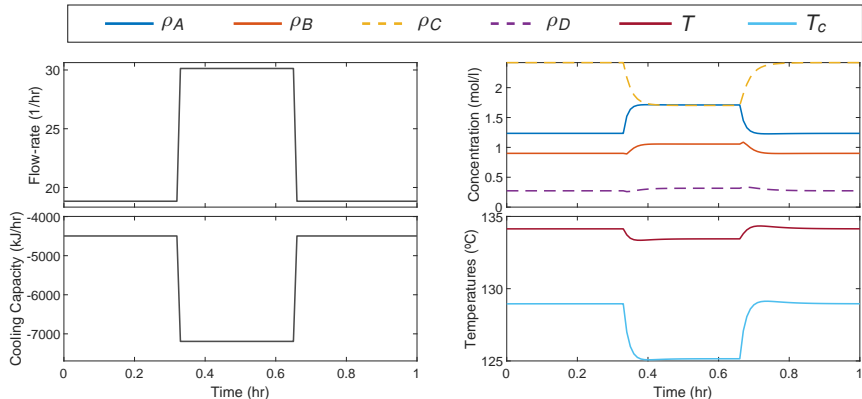
- ▶ The observability property is directly related to the *Observability Matrix*:

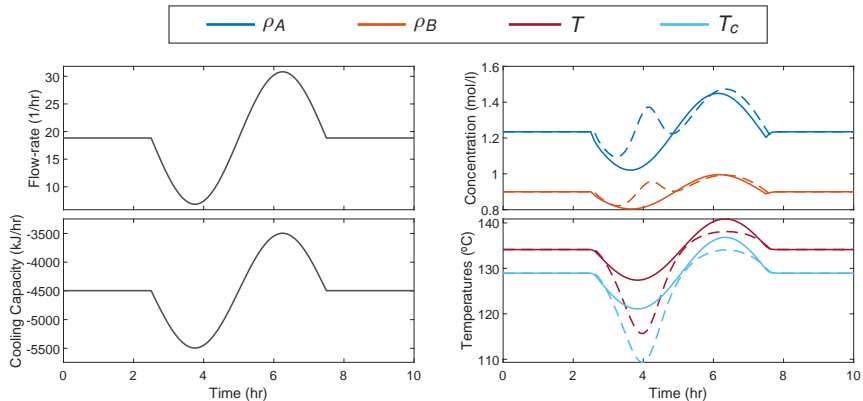
$$\mathcal{O} = [\mathbf{C} \quad \mathbf{C}\mathbf{A} \quad \mathbf{C}\mathbf{A}^2 \quad \dots \quad \mathbf{C}\mathbf{A}^{n-1}]^T ; \quad (12)$$

- ▶ The stability of the system is directly related to the eigenvalues of  $\mathbf{A}$ ;









# State-Feedback Controllers





## Full State-Feedback Controller

Given a linear system in State-Space representation, an input action  $\mathbf{u}(t)$  is calculated by the linear control law  $\pi(\cdot)$  through state-feedback as:

$$\mathbf{u}(t) = \pi(\mathbf{r}(t), \mathbf{x}(t)) = \mathbf{r}(t) - \mathbf{K}\mathbf{x}(t), \quad (13)$$

where  $\mathbf{r} : \mathbb{R} \rightarrow \mathbb{R}^n$  is a state reference signal that the system must follow and  $\mathbf{K} \in \mathbb{R}^{r \times n}$  is the *feedback gain matrix*.



## Full State-Feedback Controller

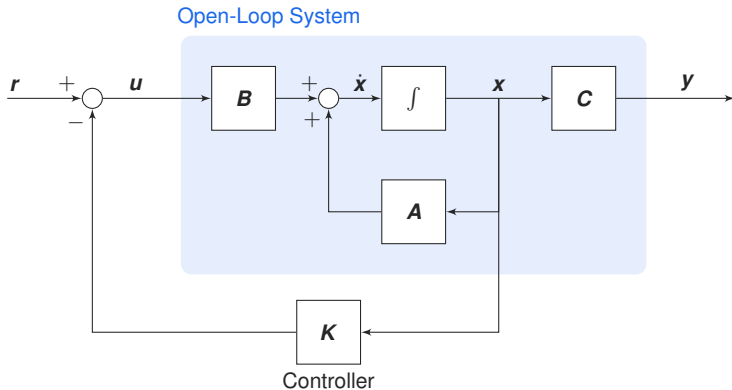
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$$\mathbf{u}(t) = \pi(\mathbf{r}(t), \mathbf{x}(t)) = \mathbf{r}(t) - \mathbf{K}\mathbf{x}(t), \quad (14)$$

where  $\mathbf{r} : \mathbb{R} \rightarrow \mathbb{R}^n$  is a state reference signal that the system must follow and  $\mathbf{K} \in \mathbb{R}^{r \times n}$  is the *feedback gain matrix*.

## Pole-Placement Property

If a system in State-Space representation is controllable, then by state feedback using a gain matrix  $\mathbf{K} \in \mathbb{R}^{r \times n}$  the eigenvalues of  $\mathbf{A}_{cl} = \mathbf{A} - \mathbf{B}\mathbf{K}$ , the poles of the closed-loop system, can be placed arbitrarily in the complex plane, as long as complex conjugate eigenvalues are assigned in pairs.

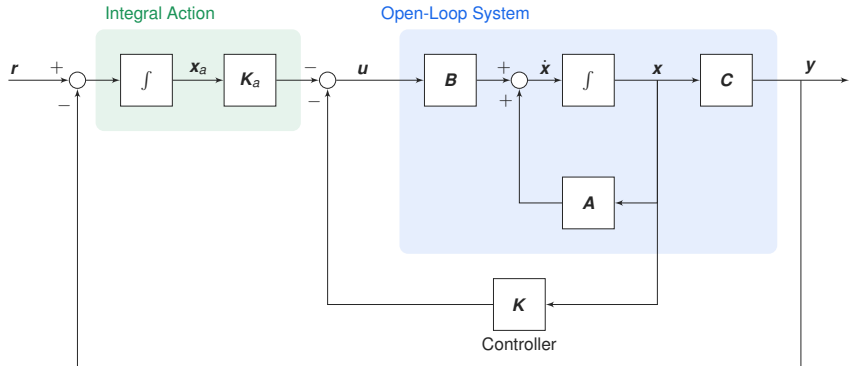




During the semester, we have discussed Conservation Laws of fluid elements...

## Conservation of Mass

$$\left( \begin{array}{c} \text{Time rate of} \\ \text{change of mass} \\ \text{in the system} \end{array} \right) = \left( \begin{array}{c} \text{Mass} \\ \text{entering} \\ \text{the system} \end{array} \right) - \left( \begin{array}{c} \text{Mass} \\ \text{leaving} \\ \text{the system} \end{array} \right)$$





# Optimal Control



## Optimal Controller

Given a system in State-Space formulation, with state signal  $\mathbf{x} : \mathbb{R} \rightarrow \mathbb{R}^n$ , and a reference signal  $\mathbf{r} : \mathbb{R} \rightarrow \mathbb{R}^n$ , the input signal  $\mathbf{u}(t) \in \mathbb{R}^r$ , for any time  $t$ , is optimal if an optimal control law  $\pi^* : \mathbb{R}^{n \times n \times 1} \rightarrow \mathbb{R}^r$  can be found as:

$$\mathbf{u}(t) = \pi^*(\mathbf{x}, \mathbf{r}, t) = \min_{\mathbf{u}} J(\mathbf{x}, \mathbf{r}, t), \quad (15)$$

where  $J : \mathbb{R}^{n \times n \times 1} \rightarrow \mathbb{R}$  is known as a *cost function* of the states and reference signals.



## Optimal Controller

Given a system in State-Space formulation, with state signal  $\mathbf{x} : \mathbb{R} \rightarrow \mathbb{R}^n$ , and a reference signal  $\mathbf{r} : \mathbb{R} \rightarrow \mathbb{R}^n$ , the input signal  $\mathbf{u}(t) \in \mathbb{R}^r$ , for any time  $t$ , is optimal if an optimal control law  $\pi^* : \mathbb{R}^{n \times n \times 1} \rightarrow \mathbb{R}^r$  can be found as:

$$\mathbf{u}(t) = \pi^*(\mathbf{x}, \mathbf{r}, t) = \min_{\mathbf{u}} J(\mathbf{x}, \mathbf{r}, t), \quad (16)$$

where  $J : \mathbb{R}^{n \times n \times 1} \rightarrow \mathbb{R}$  is known as a *cost function* of the states and reference signals.

## Finite-Horizon Optimal Regulators

A *Finite-Horizon Optimal Regulator* is defined as any controller whose optimal policy over a time interval  $t \in [t_0, T]$  minimizes the cost functional:

$$J(\mathbf{x}, \mathbf{u}, t_0) = \int_{t_0}^T l(\mathbf{x}, \mathbf{u}, \tau) d\tau + l_f(\mathbf{x}, T), \quad (17)$$

where  $l(\cdot) : \mathbb{R}^{n \times r \times 1} \rightarrow \mathbb{R}$  and  $l_f(\cdot) : \mathbb{R}^n \rightarrow \mathbb{R}$  are, respectively, the *trajectory* and *terminal loss functions*. In the case that  $t_0 = 0$ ,  $T$  is also known as the *control horizon*.





## Hamilton-Jacobi-Bellman Equation

Consider a finite-horizon cost function for a system described by  $\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}, t)$ :

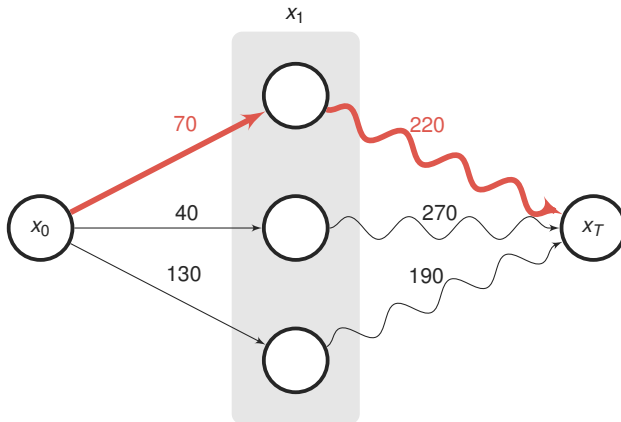
$$V(\mathbf{x}, \mathbf{u}, t_0) = \int_{t_0}^T l(\mathbf{x}, \mathbf{u}, \tau) d\tau + l_f(\mathbf{x}(T)) \quad (18)$$

Consider also that the loss  $l(\cdot)$  and state function  $\mathbf{f}$  are smooth on their parameters. Then, minimizing any functional in the form of  $V(\cdot)$  is equivalent to determining the solution of the *Hamilton-Jacobi equation*, which is given by the partial differential equation:

$$\frac{\partial V^*}{\partial t} = - \min_{\mathbf{u}(t)} \left[ l(\mathbf{x}, \mathbf{u}, t) + \left[ \frac{\partial V^*}{\partial \mathbf{x}} \right]^T \mathbf{f}(\mathbf{x}, \mathbf{u}, t) \right] \quad (19)$$

and the boundary condition:

$$V^*(\mathbf{x}, T) = l_f(\mathbf{x}(T)). \quad (20)$$



## Linear Quadratic Regulator (LQR)

Given a linear State-Space system in the form:

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) \\ \mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{D}\mathbf{u}(t) \end{cases} \quad (21)$$

A *Linear Quadratic Regulator* (LQR) for this system is an optimal controller defined by the quadratic cost function:

$$J(\mathbf{x}, \mathbf{u}, t_0) = \int_{t_0}^T \left( \mathbf{x}^T \mathbf{Q} \mathbf{x} + \mathbf{u}^T \mathbf{R} \mathbf{u} \right) dt + \mathbf{x}^T(T) \mathbf{Q}_f \mathbf{x}(T), \quad (22)$$

where is assumed that  $\mathbf{Q}, \mathbf{Q}_f \succ 0$  and  $\mathbf{R} \succ 0$  are matrices penalizing, respectively, the state-vector magnitude and the control effort.



## LQR Control Action from Dynamic Programming

Given a Linear Quadratic Regulator, the optimal action produced by this optimal controller at any time  $t \in [t_0, T]$  is given by:

$$\mathbf{u}^*(t) = -\mathbf{R}^{-1} \mathbf{B}^T \mathbf{P}(t) \mathbf{x}(t), \quad (23)$$

where  $\mathbf{P}(t)$  is the solution of the matrix Riccati differential equation:

$$-\dot{\mathbf{P}}(t) = \mathbf{A}^T \mathbf{P}(t) + \mathbf{P}(t) \mathbf{A} - \mathbf{P}(t) \mathbf{B} \mathbf{R}^{-1} \mathbf{B}^T \mathbf{P}(t) + \mathbf{Q}, \quad (24)$$

with terminal condition  $\mathbf{P}(T) = \mathbf{Q}_f$ .

## LQR with Integral Action

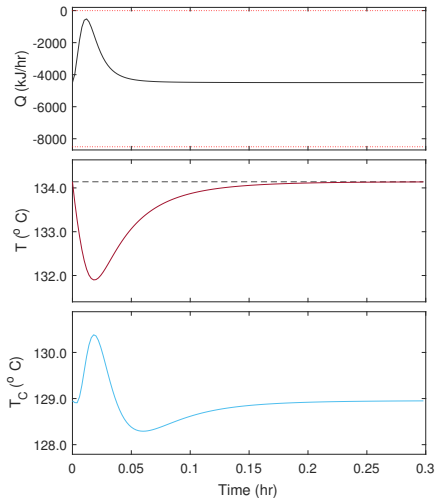
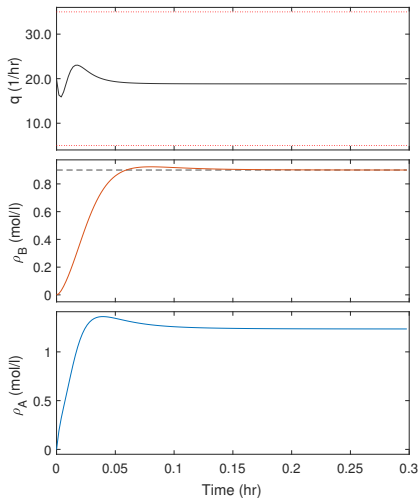
Given a linear State-Space system represented by matrices ( $\mathbf{A}$ ,  $\mathbf{B}$ ,  $\mathbf{C}$ ,  $\mathbf{D}$ ), augmented with state  $\dot{\mathbf{x}}_a(t) = \mathbf{r}(t) - \mathbf{C}\mathbf{x}(t)$ :

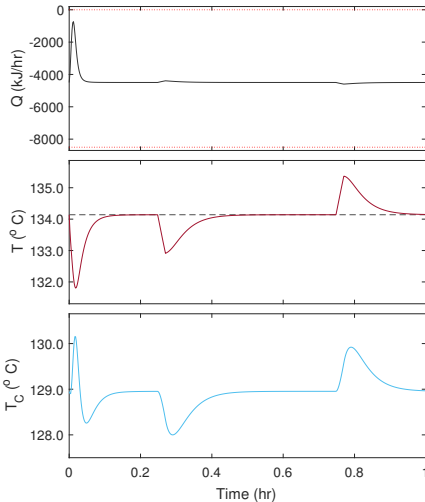
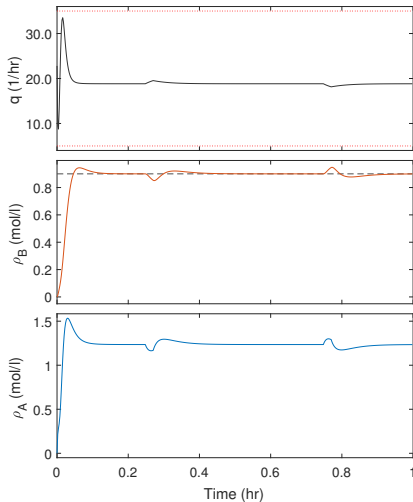
$$\begin{cases} \begin{bmatrix} \dot{\mathbf{x}}(t) \\ \dot{\mathbf{x}}_a(t) \end{bmatrix} = \underbrace{\begin{bmatrix} \mathbf{A} & \mathbf{0} \\ -\mathbf{C} & \mathbf{0} \end{bmatrix}}_{\tilde{\mathbf{A}}} \underbrace{\begin{bmatrix} \mathbf{x}(t) \\ \mathbf{x}_a(t) \end{bmatrix}}_{\tilde{\mathbf{x}}(t)} + \underbrace{\begin{bmatrix} \mathbf{B} \\ \mathbf{0} \end{bmatrix}}_{\tilde{\mathbf{B}}} \mathbf{u}(t) + \begin{bmatrix} \mathbf{0} \\ \mathbf{I} \end{bmatrix} \mathbf{r}(t) \\ \mathbf{y}(t) = \begin{bmatrix} \mathbf{C} & \mathbf{0} \end{bmatrix} \begin{bmatrix} \mathbf{x}(t) \\ \mathbf{x}_a(t) \end{bmatrix} \end{cases} \quad (25)$$

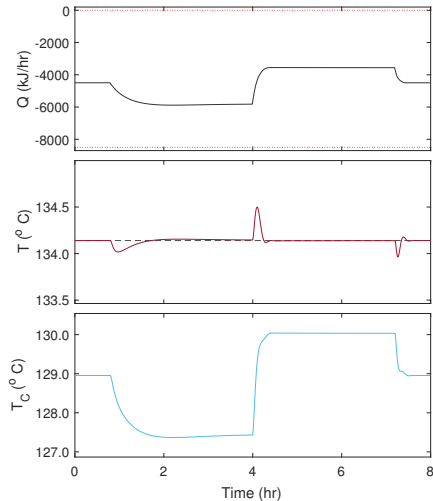
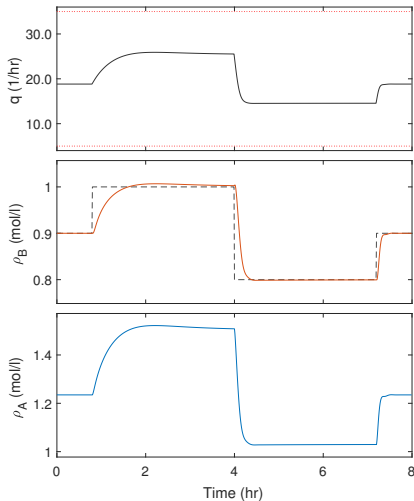
A *Linear Quadratic Servo* (LQ-Servo) for this system is an optimal controller defined by the quadratic cost function:

$$J(\mathbf{x}, \mathbf{u}, t_0) = \int_{t_0}^T \left( \tilde{\mathbf{x}}^T \tilde{\mathbf{Q}} \tilde{\mathbf{x}} + \mathbf{u}^T \mathbf{R} \mathbf{u} \right) dt + \tilde{\mathbf{x}} \tilde{\mathbf{Q}}_f \tilde{\mathbf{x}}(T), \quad (26)$$

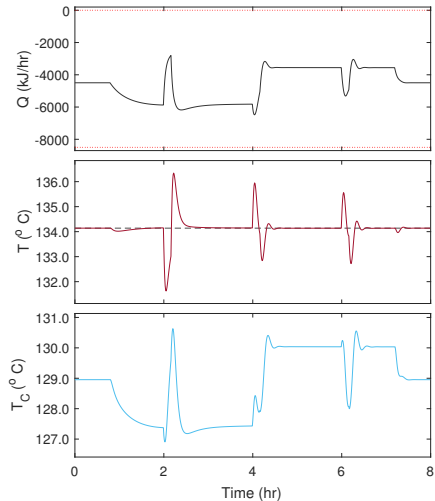
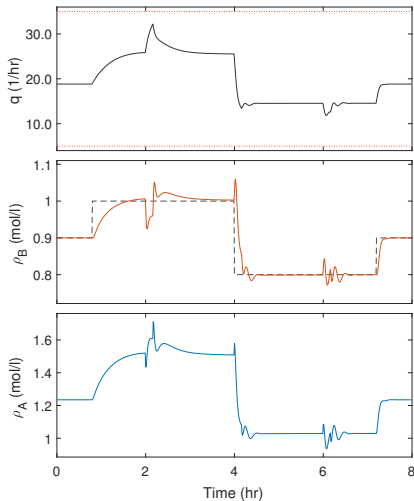
where is assumed that  $\tilde{\mathbf{Q}}$ ,  $\tilde{\mathbf{Q}}_f \succ 0$  and  $\mathbf{R} \succ 0$  are matrices penalizing, respectively, the state-vector magnitude and the control effort.











# Optimal State Estimation



## Closed-Loop Observer

Given a system in State-Space with output signal  $\mathbf{y}(t) : \mathbb{R} \rightarrow \mathbb{R}^p$  and an observer gain  $\mathbf{L} \in \mathbb{R}^{n \times p}$ , the estimated state-vector  $\hat{\mathbf{x}}(t)$  is represented by the system:

$$\dot{\hat{\mathbf{x}}}(t) = \mathbf{A}\hat{\mathbf{x}}(t) + \mathbf{B}\mathbf{u}(t) + \mathbf{L}(\mathbf{y}(t) - \mathbf{C}\hat{\mathbf{x}}(t)), \quad (27)$$

or, equivalently:

$$\dot{\hat{\mathbf{x}}}(t) = (\mathbf{A} - \mathbf{L}\mathbf{C})\hat{\mathbf{x}}(t) + \mathbf{B}\mathbf{u}(t) + \mathbf{L}\mathbf{y}(t). \quad (28)$$

- ▶ The observer system works as a parallel system that is simulated alongside the actual system;
- ▶ Alternatively, it is possible to create a variable  $\mathbf{e}(t) = \mathbf{x}(t) - \hat{\mathbf{x}}(t)$  such that:

$$\begin{aligned} \dot{\mathbf{e}} &= \dot{\mathbf{x}} - \dot{\hat{\mathbf{x}}} \\ &= (\mathbf{A} - \mathbf{L}\mathbf{C})(\mathbf{x} - \hat{\mathbf{x}}) \\ &= (\mathbf{A} - \mathbf{L}\mathbf{C})\mathbf{e}, \end{aligned} \quad (29)$$

implying that the observer tracks the actual state-vector if  $\mathbf{e}(t) = \mathbf{0}$  as  $t \rightarrow \infty$ .



## Pole-Placement Property of Observers

If a system in State-Space representation is observable, then by a closed-loop observer with gain matrix  $\mathbf{L} \in \mathbb{R}^{n \times p}$  the eigenvalues of  $\mathbf{A}_{obs} = \mathbf{A} - \mathbf{L}\mathbf{C}$  can arbitrarily be assigned anywhere in the complex plane, as long as that complex conjugate eigenvalues are assigned in pairs.

► a



## The Separation Principle

Given a system in State-Space with a Luenberger observer of gain  $\mathbf{L}$  and state-feedback controller of gain  $\mathbf{K}$ , the closed-loop eigenvalues contributions of  $(\mathbf{A} - \mathbf{BK})$  are independent from those of  $(\mathbf{A} - \mathbf{LC})$ .

► a



## Kalman-Bucy Optimal Filter

Consider a continuous-time State-Space linear system subject to additive process noise variable  $\mathbf{w}_k \sim \mathcal{N}(\mathbf{0}, \mathbf{Q}_{kf})$  and measurement noise variable  $\mathbf{v}_k \sim \mathcal{N}(\mathbf{0}, \mathbf{R}_{kf})$ , where the covariances  $\mathbf{Q}_{kf} \in \mathbb{R}^{n \times n}$  and  $\mathbf{R}_{kf} \in \mathbb{R}^{p \times p}$  represents the *power spectral density* of the noises. In this case, for an estimated state  $\bar{\mathbf{x}}(t)$  at time  $t$ , the error covariance:

$$\mathbf{J}(\mathbf{x}, \bar{\mathbf{x}}, t) = \mathbb{E} \left\{ [\mathbf{x}(t) - \bar{\mathbf{x}}(t)][\mathbf{x}(t) - \bar{\mathbf{x}}(t)]^T \right\} \quad (30)$$

is minimized by  $\bar{\mathbf{x}}(t)$  obtained through the system:

$$\dot{\bar{\mathbf{x}}}(t) = \mathbf{A}\bar{\mathbf{x}}(t) + \mathbf{B}u(t) + \mathbf{K}_e(t) (\mathbf{y}(t) - \mathbf{C}\bar{\mathbf{x}}(t)), \quad (31)$$

where  $\mathbf{K}_e(t) = \mathbf{P}_e(t)\mathbf{C}\mathbf{R}^{-1}$ , being  $\mathbf{P}_e(t)$  the solution of the Riccati differential matrix equation:

$$\dot{\mathbf{P}}_e(t) = \mathbf{A}\mathbf{P}_e(t) + \mathbf{P}_e(t)\mathbf{A}^T - \mathbf{P}_e(t)\mathbf{C}^T\mathbf{R}_{kf}^{-1}\mathbf{C}\mathbf{P}_e(t) + \mathbf{Q}_{kf}, \quad (32)$$

with initial condition  $\mathbf{P}_e(t_0) = \mathbb{E} \left\{ [\mathbf{x}(t_0) - \bar{\mathbf{x}}(t_0)][\mathbf{x}(t_0) - \bar{\mathbf{x}}(t_0)]^T \right\}$  for  $t_0 > -\infty$ .



## Linear Quadratic Gaussian (LQG) Controller

Consider a stochastic system in State-Space representation:

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) + \mathbf{w}(t) \\ \mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{v}(t) \end{cases}, \quad (33)$$

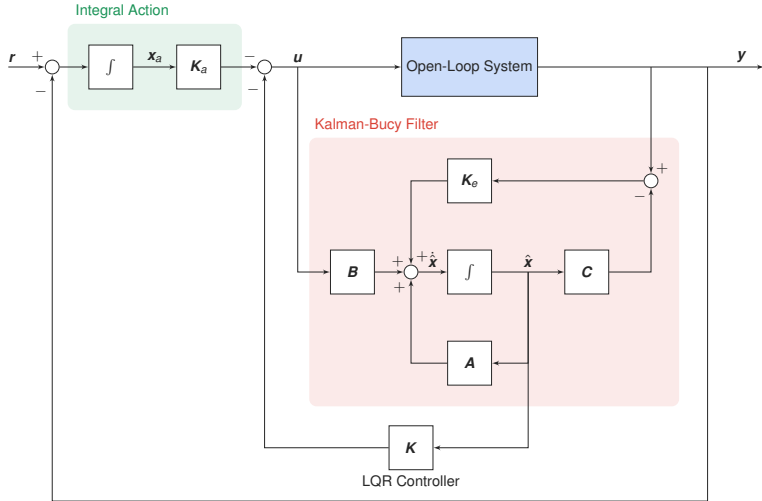
whose estimated state-vector  $\hat{\mathbf{x}}(t)$  is determined by a Kalman-Bucy filter and whose optimal input signal  $\mathbf{u}(t)$  is calculated through a finite-horizon LQR. The Linear Quadratic Gaussian (LQG) control for the horizon  $t \in [t_0, T]$ , with  $-\infty < t_0 \leq T < \infty$ , is defined as:

$$\dot{\hat{\mathbf{x}}}(t) = [\mathbf{A} - \mathbf{K}_e(t)\mathbf{C} - \mathbf{B}\mathbf{K}(t)] \hat{\mathbf{x}}(t) + \mathbf{K}_e(t)\mathbf{y}(t), \quad (34)$$

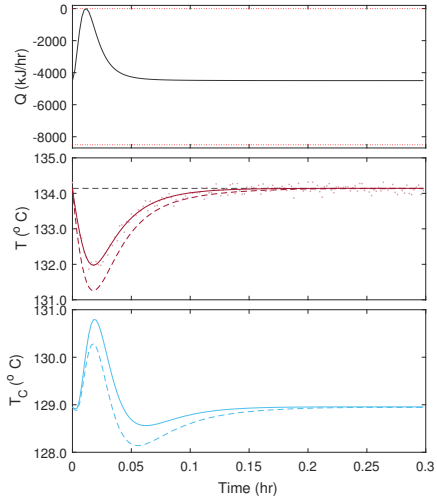
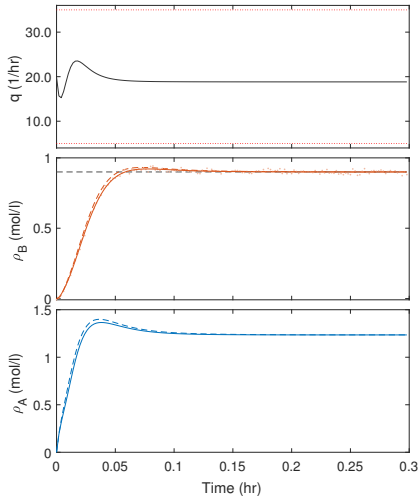
where  $\mathbf{K}(t) = \mathbf{R}^{-1}\mathbf{B}^T\mathbf{P}(t)$  and  $\mathbf{K}_e(t) = \mathbf{P}_e(t)\mathbf{C}\mathbf{R}^{-1}$  are, respectively, the LQR and Kalman-Bucy gains for matrices  $\mathbf{P}(t)$  and  $\mathbf{P}_e(t)$  that solve the Riccati differential equations:

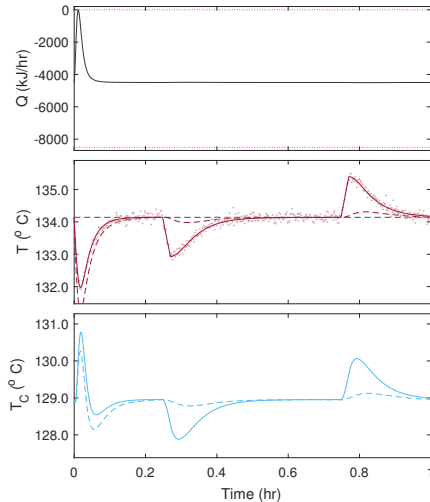
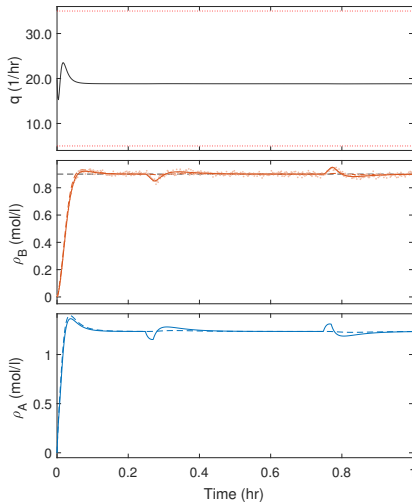
$$\begin{cases} -\dot{\mathbf{P}}(t) = \mathbf{A}^T\mathbf{P}(t) + \mathbf{P}(t)\mathbf{A} - \mathbf{P}(t)\mathbf{B}\mathbf{R}^{-1}\mathbf{B}^T\mathbf{P}(t) + \mathbf{Q} \\ \dot{\mathbf{P}}_e(t) = \mathbf{A}\mathbf{P}_e(t) + \mathbf{P}_e(t)\mathbf{A}^T - \mathbf{P}_e(t)\mathbf{C}^T\mathbf{R}_{kf}^{-1}\mathbf{C}\mathbf{P}_e(t) + \mathbf{Q}_{kf} \end{cases} \quad (35)$$

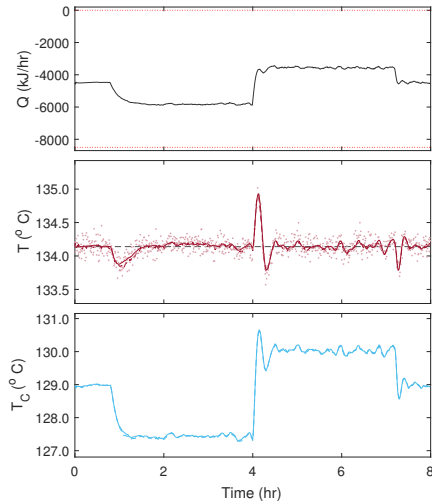
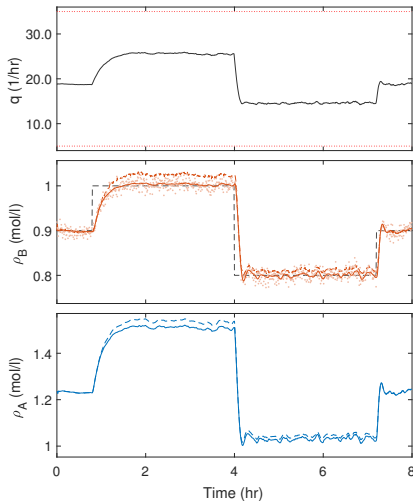
for boundary conditions  $\mathbf{P}(T) = \mathbf{Q}_f$  and  $\mathbf{P}_e(t_0) = \mathbb{E} \{ [\mathbf{x}(t_0) - \bar{\mathbf{x}}(t_0)][\mathbf{x}(t_0) - \bar{\mathbf{x}}(t_0)]^T \}$ , respectively.

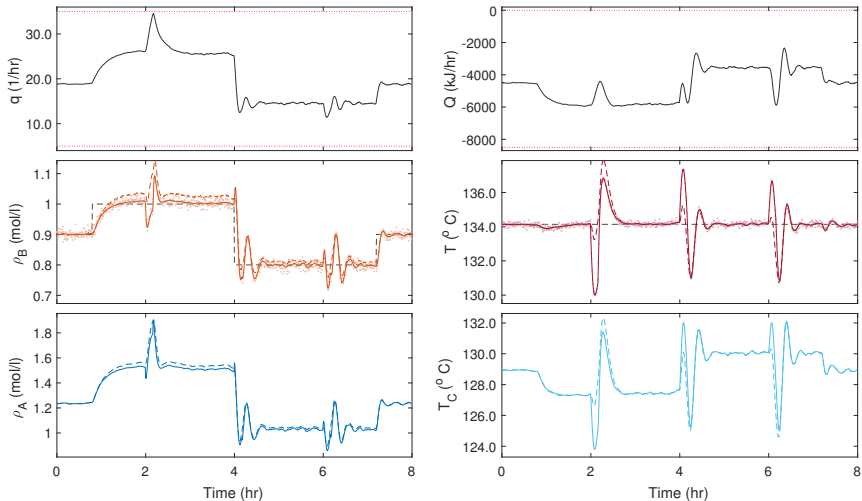












# Conclusion



- ▶ Past;
- ▶ Present;
- ▶ Future;

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**Thank you!**  
Questions?