# life.augmented

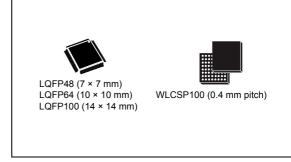
## STM32F303xB STM32F303xC

Arm<sup>®</sup>-based Cortex<sup>®</sup>-M4 32b MCU+FPU, up to 256KB Flash+ 48KB SRAM, 4 ADCs, 2 DAC ch., 7 comp, 4 PGA, timers, 2.0-3.6 V

Datasheet - production data

#### **Features**

- Core: Arm<sup>®</sup> Cortex<sup>®</sup>-M4 32-bit CPU with FPU (72 MHz max), single-cycle multiplication and HW division, 90 DMIPS (from CCM), DSP instruction and MPU (memory protection unit)
- · Operating conditions:
  - V<sub>DD</sub>, V<sub>DDA</sub> voltage range: 2.0 V to 3.6 V
- Memories
  - 128 to 256 Kbytes of Flash memory
  - Up to 40 Kbytes of SRAM, with HW parity check implemented on the first 16 Kbytes.
  - Routine booster: 8 Kbytes of SRAM on instruction and data bus, with HW parity check (CCM)
- CRC calculation unit
- · Reset and supply management
  - Power-on/power-down reset (POR/PDR)
  - Programmable voltage detector (PVD)
  - Low-power modes: Sleep, Stop and Standby
  - V<sub>BAT</sub> supply for RTC and backup registers
- · Clock management
  - 4 to 32 MHz crystal oscillator
  - 32 kHz oscillator for RTC with calibration
  - Internal 8 MHz RC with x 16 PLL option
  - Internal 40 kHz oscillator
- Up to 87 fast I/Os
  - All mappable on external interrupt vectors
  - Several 5 V-tolerant
- Interconnect matrix
- 12-channel DMA controller
- Four ADCs 0.20 µS (up to 39 channels) with selectable resolution of 12/10/8/6 bits, 0 to 3.6 V conversion range, single ended/differential input, separate analog supply from 2 to 3.6 V
- Two 12-bit DAC channels with analog supply from 2.4 to 3.6 V



- Seven fast rail-to-rail analog comparators with analog supply from 2 to 3.6 V
- Four operational amplifiers that can be used in PGA mode, all terminals accessible with analog supply from 2.4 to 3.6 V
- Up to 24 capacitive sensing channels supporting touchkey, linear and rotary touch sensors
- Up to 13 timers
  - One 32-bit timer and two 16-bit timers with up to 4 IC/OC/PWM or pulse counter and quadrature (incremental) encoder input
  - Two 16-bit 6-channel advanced-control timers, with up to 6 PWM channels, deadtime generation and emergency stop
  - One 16-bit timer with 2 IC/OCs, 1 OCN/PWM, deadtime generation and emergency stop
  - Two 16-bit timers with IC/OC/OCN/PWM, deadtime generation and emergency stop
  - Two watchdog timers (independent, window)
  - SysTick timer: 24-bit downcounter
  - Two 16-bit basic timers to drive the DAC
- Calendar RTC with Alarm, periodic wakeup from Stop/Standby
- · Communication interfaces
  - CAN interface (2.0B Active)
  - Two I<sup>2</sup>C Fast mode plus (1 Mbit/s) with 20 mA current sink, SMBus/PMBus, wakeup from STOP

- Up to five USART/UARTs (ISO 7816 interface, LIN, IrDA, modem control)
- Up to three SPIs, two with multiplexed half/full duplex I2S interface, 4 to 16 programmable bit frames
- USB 2.0 full speed interface
- Infrared transmitter
- Serial wire debug, Cortex<sup>®</sup>-M4 with FPU ETM, JTAG
- 96-bit unique ID

Table 1. Device summary

Reference	Part number	
STM32F303xB	STM32F303CB, STM32F303RB, STM32F303VB	
STM32F303xC	STM32F303CC, STM32F303RC, STM32F303VC	

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#### 1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32F303xB/STM32F303xC microcontrollers.

This STM32F303xB/STM32F303xC datasheet should be read in conjunction with the STM32F303x, STM32F358xC and STM32F328x4/6/8 reference manual (RM0316). The reference manual is available from the STMicroelectronics website <a href="https://www.st.com">www.st.com</a>.

For information on the Arm<sup>®</sup>(a) Cortex<sup>®</sup>-M4 core with FPU, refer to:

- Cortex<sup>®</sup>-M4 with FPU Technical Reference Manual, available from the http://www.arm.com website.
- STM32F3xxx and STM32F4xxx Cortex<sup>®</sup>-M4 programming manual (PM0214) available from our website www.st.com.



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## 2 Description

The STM32F303xB/STM32F303xC family is based on the high-performance Arm<sup>®</sup> Cortex<sup>®</sup>-M4 32-bit RISC core with FPU operating at a frequency of up to 72 MHz, and embedding a floating point unit (FPU), a memory protection unit (MPU) and an embedded trace macrocell (ETM). The family incorporates high-speed embedded memories (up to 256 Kbytes of Flash memory, up to 40 Kbytes of SRAM) and an extensive range of enhanced I/Os and peripherals connected to two APB buses.

The devices offer up to four fast 12-bit ADCs (5 Msps), seven comparators, four operational amplifiers, up to two DAC channels, a low-power RTC, up to five general-purpose 16-bit timers, one general-purpose 32-bit timer, and two timers dedicated to motor control. They also feature standard and advanced communication interfaces: up to two I<sup>2</sup>Cs, up to three SPIs (two SPIs are with multiplexed full-duplex I2Ss), three USARTs, up to two UARTs, CAN and USB. To achieve audio class accuracy, the I2S peripherals can be clocked via an external PLL.

The STM32F303xB/STM32F303xC family operates in the -40 to +85 °C and -40 to +105 °C temperature ranges from a 2.0 to 3.6 V power supply. A comprehensive set of power-saving mode allows the design of low-power applications.

The STM32F303xB/STM32F303xC family offers devices in four packages ranging from 48 pins to 100 pins.

The set of included peripherals changes with the device chosen.



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Table 2. STM32F303xB/STM32F303xC family device features and peripheral counts

Peri	STM32F303Cx		STM32F303Rx			STM32F303Vx	
Flash (Kbytes)		128	256	128	256	128	256
SRAM (Kbytes) on data bus			40		40		40
CCM (Core Coup RAM (Kbytes)		32         40         32         40         32         40           8					
	Advanced control		2 (16-bit)				
Timers	General purpose		5 (16-bit) 1 (32-bit)				
	Basic			2 (16	6-bit)		
PWM channels (	all) <sup>(1)</sup>	3	1		3	3	
PWM channels (complementary)	except	2	2		2	<u>!</u> 4	
	SPI (I2S) <sup>(2)</sup>			3(	2)		
	I <sup>2</sup> C			2	2		
Communication	USART	3					
interfaces	UART	0 2					
	CAN	1					
	USB	1					
GPIOs	Normal I/Os (TC, TTa)	20		2	45 in LQFP10 37 in WLCSP10		
GFIOS	5-volt tolerant I/Os (FT, FTf)	17		25 42 in LQFP1 40 in WLCSP			
DMA channels		12					
Capacitive sensi	ng channels	17 18		2	:4		
12-bit ADCs		4					
Number of chann	nels	1 15 1 22 1			QFP100 _CSP100		
12-bit DAC chan	nels	2					
Analog comparator		7					
Operational amplifiers		4					
CPU frequency	72 MHz						
Operating voltage		2.0 to 3.6 V					
Operating temperature		Ambient operating temperature: - 40 to 85 °C / - 40 to 105 °C Junction temperature: - 40 to 125 °C			o 105 °C		
Packages	Packages		P48	LQF	P64		P100 SP100

<sup>1.</sup> This total number considers also the PWMs generated on the complementary output channels



<sup>2.</sup> The SPI interfaces can work in an exclusive way in either the SPI mode or the  $I^2S$  audio mode.

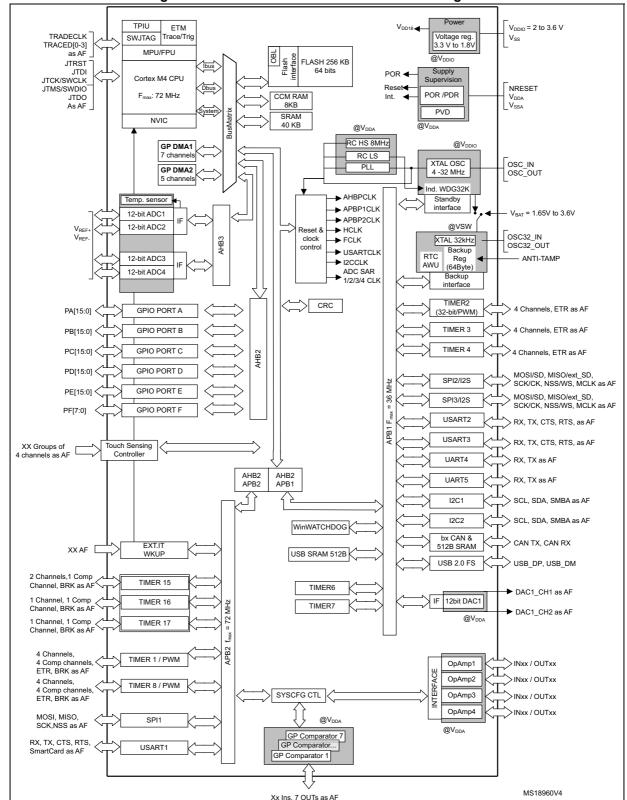


Figure 1. STM32F303xB/STM32F303xC block diagram

1. AF: alternate function on I/O pins.



#### 3 Functional overview

# 3.1 Arm<sup>®</sup> Cortex<sup>®</sup>-M4 core with FPU with embedded Flash and SRAM

The Arm Cortex-M4 processor with FPU is the latest generation of Arm processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The Arm Cortex-M4 32-bit RISC processor with FPU features exceptional code-efficiency, delivering the high-performance expected from an Arm core in the memory size usually associated with 8- and 16-bit devices.

The processor supports a set of DSP instructions which allow efficient signal processing and complex algorithm execution.

Its single precision FPU speeds up software development by using metalanguage development tools, while avoiding saturation.

With its embedded Arm core, the STM32F303xB/STM32F303xC family is compatible with all Arm tools and software.

*Figure 1* shows the general block diagram of the STM32F303xB/STM32F303xC family devices.

#### 3.2 Memory protection unit (MPU)

The memory protection unit (MPU) is used to separate the processing of tasks from the data protection. The MPU can manage up to 8 protection areas that can all be further divided up into 8 subareas. The protection area sizes are between 32 bytes and the whole 4 gigabytes of addressable memory.

The memory protection unit is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

## 3.3 Embedded Flash memory

All STM32F303xB/STM32F303xC devices feature up to 256 Kbytes of embedded Flash memory available for storing programs and data. The Flash memory access time is adjusted to the CPU clock frequency (0 wait state from 0 to 24 MHz, 1 wait state from 24 to 48 MHz and 2 wait states above).



#### 3.4 Embedded SRAM

STM32F303xB/STM32F303xC devices feature up to 48 Kbytes of embedded SRAM with hardware parity check. The memory can be accessed in read/write at CPU clock speed with 0 wait states, allowing the CPU to achieve 90 Dhrystone Mips at 72 MHz (when running code from the CCM (Core Coupled Memory) RAM).

- 8 Kbytes of CCM RAM mapped on both instruction and data bus, used to execute critical routines or to access data (parity check on all of CCM RAM).
- 40 Kbytes of SRAM mapped on the data bus (parity check on first 16 Kbytes of SRAM).

#### 3.5 Boot modes

At startup, Boot0 pin and Boot1 option bit are used to select one of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

The boot loader is located in the system memory. It is used to reprogram the Flash memory by using USART1 (PA9/PA10), USART2 (PD5/PD6) or USB (PA11/PA12) through DFU (device firmware upgrade).

#### 3.6 Cyclic redundancy check (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator polynomial value and size.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at linktime and stored at a given memory location.

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#### 3.7 Power management

#### 3.7.1 Power supply schemes

- $V_{SS}$ ,  $V_{DD}$  = 2.0 to 3.6 V: external power supply for I/Os and the internal regulator. It is provided externally through  $V_{DD}$  pins.
- V<sub>SSA</sub>, V<sub>DDA</sub> = 2.0 to 3.6 V: external analog power supply for ADC, DACs, comparators operational amplifiers, reset blocks, RCs and PLL. The minimum voltage to be applied to V<sub>DDA</sub> differs from one analog peripheral to another. *Table 3* provides the summary of the V<sub>DDA</sub> ranges for analog peripherals. The V<sub>DDA</sub> voltage level must be always greater or equal to the V<sub>DD</sub> voltage level and must be provided first.
- V<sub>BAT</sub> = 1.65 to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V<sub>DD</sub> is not present.

Analog peripheral	Minimum V <sub>DDA</sub> supply	Maximum V <sub>DDA</sub> supply	
ADC / COMP	2.0 V	3.6 V	
DAC / OPAMP	2.4 V	3.6V	

Table 3. External analog supply values for analog peripherals

#### 3.7.2 Power supply supervision

The device has an integrated power-on reset (POR) and power-down reset (PDR) circuits. They are always active, and ensure proper operation above a threshold of 2 V. The device remains in reset mode when the monitored supply voltage is below a specified threshold, VPOR/PDR, without the need for an external reset circuit.

- The POR monitors only the V<sub>DD</sub> supply voltage. During the startup phase it is required that V<sub>DDA</sub> should arrive first and be greater than or equal to V<sub>DD</sub>.
- The PDR monitors both the V<sub>DD</sub> and V<sub>DDA</sub> supply voltages, however the V<sub>DDA</sub> power supply supervisor can be disabled (by programming a dedicated Option bit) to reduce the power consumption if the application design ensures that V<sub>DDA</sub> is higher than or equal to V<sub>DD</sub>.

The device features an embedded programmable voltage detector (PVD) that monitors the  $V_{DD}$  power supply and compares it to the VPVD threshold. An interrupt can be generated when  $V_{DD}$  drops below the  $V_{PVD}$  threshold and/or when  $V_{DD}$  is higher than the  $V_{PVD}$  threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

#### 3.7.3 Voltage regulator

The regulator has three operation modes: main (MR), low-power (LPR), and power-down.

- The MR mode is used in the nominal regulation mode (Run)
- The LPR mode is used in Stop mode.
- The power-down mode is used in Standby mode: the regulator output is in high impedance, and the kernel circuitry is powered down thus inducing zero consumption.

The voltage regulator is always enabled after reset. It is disabled in Standby mode.



#### 3.7.4 Low-power modes

The STM32F303xB/STM32F303xC supports three low-power modes to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

Stop mode

Stop mode achieves the lowest power consumption while retaining the content of SRAM and registers. All clocks in the 1.8 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low-power mode.

The device can be woken up from Stop mode by any of the EXTI line. The EXTI line source can be one of the 16 external lines, the PVD output, the USB wakeup, the RTC alarm, COMPx, I2Cx or U(S)ARTx.

Standby mode

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.8 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, SRAM and register contents are lost except for registers in the Backup domain and Standby circuitry.

The device exits Standby mode when an external reset (NRST pin), an IWDG reset, a rising edge on the WKUP pin or an RTC alarm occurs.

Note: The RTC, the IWDG and the corresponding clock sources are not stopped by entering Stop or Standby mode.

#### 3.8 Interconnect matrix

Several peripherals have direct connections between them. This allows autonomous communication between peripherals, saving CPU resources thus power supply consumption. In addition, these hardware connections allow fast and predictable latency.

Table 4. STM32F303xB/STM32F303xC peripheral interconnect matrix

Interconnect source destination Interconnect action

Interconnect source destination		Interconnect action	
	TIMx	Timers synchronization or chaining	
TIMx	ADCx DAC1	Conversion triggers	
	DMA	Memory to memory transfer trigger	
	Compx	Comparator output blanking	
COMPx TIMx		Timer input: OCREF_CLR input, input capture	
ADCx	TIMx	Timer triggered by analog watchdog	



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Table 4. STM32F303xB/STM32F303xC peripheral interconnect matrix (continued)

Interconnect source	Interconnect destination	Interconnect action
GPIO RTCCLK HSE/32 MC0	TIM16	Clock source used as input channel for HSI and LSI calibration
CSS CPU (hard fault) COMPx PVD GPIO	TIM1, TIM8, TIM15, 16, 17	Timer break
	TIMx	External trigger, timer break
GPIO	ADCx DAC1	Conversion external trigger
DAC1	COMPx	Comparator inverting input

Note:

For more details about the interconnect actions, please refer to the corresponding sections in the reference manual (RM0316).

#### 3.9 Clocks and startup

System clock selection is performed on startup, however the internal RC 8 MHz oscillator is selected as default CPU clock on reset. An external 4-32 MHz clock can be selected, in which case it is monitored for failure. If failure is detected, the system automatically switches back to the internal RC oscillator. A software interrupt is generated if enabled. Similarly, full interrupt management of the PLL clock entry is available when necessary (for example with failure of an indirectly used external oscillator).

Several prescalers allow to configure the AHB frequency, the high speed APB (APB2) and the low speed APB (APB1) domains. The maximum frequency of the AHB and the high speed APB domains is 72 MHz, while the maximum allowed frequency of the low speed APB domain is 36 MHz.



FLITFCLK to Flash programming interface HSI → to I2Cx (x = 1,2) SYSCLK 12SSRC **SYSCLK** →to I2Sx (x = 2,3) I2S\_CKIN USB USBCLK prescaler to USB interface 8 MHz HSI /1,1.5 HSI RC /2 HCLK to AHB bus, core, memory and DMA PLLSRC | PLLMUL to cortex System timer SW /8 ➤ FHCLK Cortex free HSIN running clock

▶ to APB1 peripherals PLL AHB APB1 PLLCLK PCLK1 x2,x3, prescaler prescaler x16 /1,2,..512 /1,2,4,8,16 HSE SYSCLK If (APB1 prescaler css → to TIM 2,3,4,6,7 /2,/3,.. =1) x1 else x2 /16 PCLK1 SYSCLK HSI to U(S)ARTx (x = 2..5)  $\mathsf{OSC}\_\mathsf{OUT}$ 4-32 MHz HSE OSC LSE OSC IN APB2 PCLK2 prescaler → to APB2 peripherals /1,2,4,8,16 /32 RTCCLK to RTC OSC32\_IN LSE OSC If (APB2 prescaler 32.768kHz → to TIM 15,16,17 OSC32\_OUT =1) x1 else x2 RTCSEL[1:0] PCLK2 → IWDGCLK LSI LSI RC SYSCLK HSI to USART1 to IWDG 40kHz LSE /2 -PLLCLK x2 TIM1/8 - HSI - LSI MCO HSE ADC rescale /1,2,4 SYSCLK to ADCxy мсо Main clock (xy = 12, 34)output ADC Prescaler 1,2,4,6,8,10,12,16, 32,64,128,256 MS19989V5

Figure 2. Clock tree

#### 3.10 General-purpose input/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high current capable except for analog inputs.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

Fast I/O handling allows I/O toggling up to 36 MHz.

#### 3.11 Direct memory access (DMA)

The flexible general-purpose DMA is able to manage memory-to-memory, peripheral-tomemory and memory-to-peripheral transfers. The DMA controller supports circular buffer management, avoiding the generation of interrupts when the controller reaches the end of the buffer.

Each of the 12 DMA channels is connected to dedicated hardware DMA requests, with software trigger support for each channel. Configuration is done by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals: SPI, I<sup>2</sup>C, USART, general-purpose timers, DAC and ADC.

#### 3.12 Interrupts and events

#### 3.12.1 **Nested vectored interrupt controller (NVIC)**

The STM32F303xB/STM32F303xC devices embed a nested vectored interrupt controller (NVIC) able to handle up to 66 maskable interrupt channels and 16 priority levels.

The NVIC benefits are the following:

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface
- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

The NVIC hardware block provides flexible interrupt management features with minimal interrupt latency.

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#### 3.13 Fast analog-to-digital converter (ADC)

four fast analog-to-digital converters 5 MSPS, with selectable resolution between 12 and 6 bit, are embedded in the STM32F303xB/STM32F303xC family devices. The ADCs have up to 39 external channels. Some of the external channels are shared between ADC1&2 and between ADC3&4. Channels can be configured to be either single-ended input or differential input. The ADCs can perform conversions in single-shot or scan modes. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADCs have also internal channels: Temperature sensor connected to ADC1 channel 16,  $V_{BAT/2}$  connected to ADC1 channel 17, Voltage reference  $V_{REFINT}$  connected to the 4 ADCs channel 18, VOPAMP1 connected to ADC1 channel 15, VOPAMP2 connected to ADC2 channel 17, VREFOPAMP3 connected to ADC3 channel 17 and VREFOPAMP4 connected to ADC4 channel 17.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold
- Single-shunt phase current reading techniques.

The ADC can be served by the DMA controller. 3 analog watchdogs per ADC are available.

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers and the advanced-control timers (TIM1 and TIM8) can be internally connected to the ADC start trigger and injection trigger, respectively, to allow the application to synchronize A/D conversion and timers.

#### 3.13.1 Temperature sensor

The temperature sensor (TS) generates a voltage  $V_{\mbox{\footnotesize SENSE}}$  that varies linearly with temperature.

The temperature sensor is internally connected to the ADC1\_IN16 input channel which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.

#### 3.13.2 Internal voltage reference (V<sub>REFINT</sub>)

The internal voltage reference ( $V_{REFINT}$ ) provides a stable (bandgap) voltage output for the ADC and Comparators.  $V_{REFINT}$  is internally connected to the ADCx\_IN18, x=1...4 input channel. The precise voltage of  $V_{REFINT}$  is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.



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#### 3.13.3 V<sub>BAT</sub> battery voltage monitoring

This embedded hardware feature allows the application to measure the  $V_{BAT}$  battery voltage using the internal ADC channel ADC1\_IN17. As the  $V_{BAT}$  voltage may be higher than  $V_{DDA}$ , and thus outside the ADC input range, the  $V_{BAT}$  pin is internally connected to a bridge divider by 2. As a consequence, the converted digital value is half the  $V_{BAT}$  voltage.

#### 3.13.4 OPAMP reference voltage (VREFOPAMP)

Every OPAMP reference voltage can be measured using a corresponding ADC internal channel: VREFOPAMP1 connected to ADC1 channel 15, VREFOPAMP2 connected to ADC2 channel 17, VREFOPAMP3 connected to ADC3 channel 17, VREFOPAMP4 connected to ADC4 channel 17.

#### 3.14 Digital-to-analog converter (DAC)

Two 12-bit buffered DAC channels can be used to convert digital signals into analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in inverting configuration.

This digital interface supports the following features:

- Two DAC output channels
- 8-bit or 10-bit monotonic output
- Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation
- Dual DAC channel independent or simultaneous conversions
- DMA capability (for each channel)
- External triggers for conversion

## 3.15 Operational amplifier (OPAMP)

The STM32F303xB/STM32F303xC embeds four operational amplifiers with external or internal follower routing and PGA capability (or even amplifier and filter capability with external components). When an operational amplifier is selected, an external ADC channel is used to enable output measurement.

The operational amplifier features:

- 8.2 MHz bandwidth
- 0.5 mA output capability
- Rail-to-rail input/output
- In PGA mode, the gain can be programmed to be 2, 4, 8 or 16.

#### 3.16 Fast comparators (COMP)

The STM32F303xB/STM32F303xC devices embed seven fast rail-to-rail comparators with programmable reference voltage (internal or external), hysteresis and speed (low speed for low-power) and with selectable output polarity.

The reference voltage can be one of the following:

- External I/O
- DAC output pin
- Internal reference voltage or submultiple (1/4, 1/2, 3/4). Refer to *Table 28: Embedded internal reference voltage on page 63* for the value and precision of the internal reference voltage.

All comparators can wake up from STOP mode, generate interrupts and breaks for the timers and can be also combined per pair into a window comparator

#### 3.17 Timers and watchdogs

The STM32F303xB/STM32F303xC includes two advanced control timers, up to six general-purpose timers, two basic timers, two watchdog timers and a SysTick timer. The table below compares the features of the advanced control, general purpose and basic timers.

**DMA** Capture/ Counter Counter **Prescaler** Complementary Timer type **Timer** request compare resolution type factor outputs generation Channels Any integer TIM1, Up, Down, Advanced 16-bit between 1 Yes 4 Yes TIM8 Up/Down and 65536 Any integer General-Up, Down, TIM2 32-bit between 1 Yes No Up/Down purpose and 65536 Any integer General-Up, Down, TIM3. TIM4 16-bit Yes between 1 No Up/Down purpose and 65536 Any integer General-TIM15 16-bit Up between 1 Yes 2 1 purpose and 65536 Any integer General-**TIM16, TIM17** 16-bit Up between 1 Yes 1 1 purpose and 65536 Any integer TIM6, Basic 16-bit Up between 1 Yes 0 No TIM7 and 65536

Table 5. Timer feature comparison

Note: TIM1/8 can have PLL as clock source, and therefore can be clocked at 144 MHz.



#### 3.17.1 Advanced timers (TIM1, TIM8)

The advanced-control timers (TIM1 and TIM8) can each be seen as a three-phase PWM multiplexed on six channels. They have complementary PWM outputs with programmable inserted dead-times. They can also be seen as complete general-purpose timers. The four independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes) with full modulation capability (0-100%)
- One-pulse mode output

In debug mode, the advanced-control timer counter can be frozen and the PWM outputs disabled to turn off any power switches driven by these outputs.

Many features are shared with those of the general-purpose TIM timers (described in Section 3.17.2 using the same architecture, so the advanced-control timers can work together with the TIM timers via the Timer Link feature for synchronization or event chaining.

#### 3.17.2 General-purpose timers (TIM2, TIM3, TIM4, TIM15, TIM16, TIM17)

There are up to six synchronizable general-purpose timers embedded in the STM32F303xB/STM32F303xC (see *Table 5* for differences). Each general-purpose timer can be used to generate PWM outputs, or act as a simple time base.

TIM2, 3, and TIM4

These are full-featured general-purpose timers:

- TIM2 has a 32-bit auto-reload up/downcounter and 32-bit prescaler
- TIM3 and 4 have 16-bit auto-reload up/downcounters and 16-bit prescalers.

These timers all feature 4 independent channels for input capture/output compare, PWM or one-pulse mode output. They can work together, or with the other general-purpose timers via the Timer Link feature for synchronization or event chaining.

The counters can be frozen in debug mode.

All have independent DMA request generation and support quadrature encoders.

TIM15, 16 and 17

These three timers general-purpose timers with mid-range features:

They have 16-bit auto-reload upcounters and 16-bit prescalers.

- TIM15 has 2 channels and 1 complementary channel
- TIM16 and TIM17 have 1 channel and 1 complementary channel

All channels can be used for input capture/output compare, PWM or one-pulse mode output.

The timers can work together via the Timer Link feature for synchronization or event chaining. The timers have independent DMA request generation.

The counters can be frozen in debug mode.

#### 3.17.3 Basic timers (TIM6, TIM7)

These timers are mainly used for DAC trigger generation. They can also be used as a generic 16-bit time base.



#### 3.17.4 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 40 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

#### 3.17.5 Window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

#### 3.17.6 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0.
- Programmable clock source

#### 3.18 Real-time clock (RTC) and backup registers

The RTC and the 16 backup registers are supplied through a switch that takes power from either the  $V_{DD}$  supply when present or the  $V_{BAT}$  pin. The backup registers are sixteen 32-bit registers used to store 64 bytes of user application data when  $V_{DD}$  power is not present.

They are not reset by a system or power reset, or when the device wakes up from Standby mode.

The RTC is an independent BCD timer/counter. It supports the following features:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format.
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision.
- Automatic correction for 28, 29 (leap year), 30 and 31 days of the month.
- Two programmable alarms with wake up from Stop and Standby mode capability.
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Digital calibration circuit with 1 ppm resolution, to compensate for quartz crystal inaccuracy.
- Three anti-tamper detection pins with programmable filter. The MCU can be woken up from Stopand Standby modes on tamper event detection.
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event. The MCU can be woken up from Stop and Standby modes on timestamp event detection.



 17-bit Auto-reload counter for periodic interrupt with wakeup from STOP/STANDBY capability.

The RTC clock sources can be:

- A 32.768 kHz external crystal
- A resonator or oscillator
- The internal low-power RC oscillator (typical frequency of 40 kHz)
- The high-speed external clock divided by 32.

## 3.19 Inter-integrated circuit interface (I<sup>2</sup>C)

Up to two I<sup>2</sup>C bus interfaces can operate in multimaster and slave modes. They can support standard (up to 100 KHz), fast (up to 400 KHz) and fast mode + (up to 1 MHz) modes.

Both support 7-bit and 10-bit addressing modes, multiple 7-bit slave addresses (2 addresses, 1 with configurable mask). They also include programmable analog and digital noise filters.

Table 6. Comparison of I2C analog and digital filters

	Analog filter	Digital filter		
Pulse width of suppressed spikes	50 ns	Programmable length from 1 to 15 I2C peripheral clocks		
Benefits	Available in Stop mode	Extra filtering capability vs. standard requirements.     Stable length		
Drawbacks	Variations depending on temperature, voltage, process	Wakeup from Stop on address match is not available when digital filter is enabled.		

In addition, they provide hardware support for SMBUS 2.0 and PMBUS 1.1: ARP capability, Host notify protocol, hardware CRC (PEC) generation/verification, timeouts verifications and ALERT protocol management. They also have a clock domain independent from the CPU clock, allowing the I2Cx (x=1,2) to wake up the MCU from Stop mode on address match.

The I2C interfaces can be served by the DMA controller.

Refer to *Table 7* for the features available in I2C1 and I2C2.

Table 7. STM32F303xB/STM32F303xC I<sup>2</sup>C implementation

I2C features <sup>(1)</sup>	I2C1	I2C2
7-bit addressing mode	Х	Х
10-bit addressing mode	Х	Х
Standard mode (up to 100 kbit/s)	Х	Х
Fast mode (up to 400 kbit/s)	Х	Х
Fast Mode Plus with 20mA output drive I/Os (up to 1 Mbit/s)	Х	Х
Independent clock	Х	Х



rable it of mozi occasion continuous					
I2C features <sup>(1)</sup>	I2C1	I2C2			
SMBus	Х	Х			
Wakeup from STOP	Х	Х			

Table 7. STM32F303xB/STM32F303xC I<sup>2</sup>C implementation (continued)

# 3.20 Universal synchronous/asynchronous receiver transmitter (USART)

The STM32F303xB/STM32F303xC devices have three embedded universal synchronous/asynchronous receiver transmitters (USART1, USART2 and USART3).

The USART interfaces are able to communicate at speeds of up to 9 Mbits/s.

They provide hardware management of the CTS and RTS signals, they support IrDA SIR ENDEC, the multiprocessor communication mode, the single-wire half-duplex communication mode and have LIN Master/Slave capability. The USART interfaces can be served by the DMA controller.

#### 3.21 Universal asynchronous receiver transmitter (UART)

The STM32F303xB/STM32F303xC devices have 2 embedded universal asynchronous receiver transmitters (UART4, and UART5). The UART interfaces support IrDA SIR ENDEC, multiprocessor communication mode and single-wire half-duplex communication mode. The UART4 interface can be served by the DMA controller.

Refer to *Table 8* for the features available in all U(S)ART interfaces.

USART modes/features<sup>(1)</sup> **USART1 USART2 USART3 UART4 UART5** Hardware flow control for modem Х Х Х Continuous communication using DMA Х Х Х Χ Multiprocessor communication Х Х Χ Χ Х Synchronous mode Х Χ Х Smartcard mode Χ Х Χ Single-wire half-duplex communication Χ Χ Х Χ Χ IrDA SIR ENDEC block Χ Х Х Х Х I IN mode Χ Χ Х Χ Χ Dual clock domain and wakeup from Stop mode Χ Х Χ Χ Х Х Х Х Χ Χ Receiver timeout interrupt Modbus communication Х Х Х Х Χ Χ Auto baud rate detection Χ Χ Х Χ Х **Driver Enable** 

Table 8. USART features

1. X = supported.



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<sup>1.</sup> X = supported.

# 3.22 Serial peripheral interface (SPI)/Inter-integrated sound interfaces (I2S)

Up to three SPIs are able to communicate up to 18 Mbits/s in slave and master modes in full-duplex and half-duplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame size is configurable from 4 bits to 16 bits.

Two standard I2S interfaces (multiplexed with SPI2 and SPI3) supporting four different audio standards can operate as master or slave at half-duplex and full duplex communication modes. They can be configured to transfer 16 and 24 or 32 bits with 16-bit or 32-bit data resolution and synchronized by a specific signal. Audio sampling frequency from 8 kHz up to 192 kHz can be set by 8-bit programmable linear prescaler. When operating in master mode it can output a clock for an external audio component at 256 times the sampling frequency.

Refer to *Table 9* for the features available in SPI1, SPI2 and SPI3.

SPI features <sup>(1)</sup>	SPI1	SPI2	SPI3
Hardware CRC calculation	Х	Х	X
Rx/Tx FIFO	Х	Х	Х
NSS pulse mode	Х	Х	Х
I2S mode	-	Х	Х
TI mode	Х	Х	Х

Table 9. STM32F303xB/STM32F303xC SPI/I2S implementation

### 3.23 Controller area network (CAN)

The CAN is compliant with specifications 2.0A and B (active) with a bit rate up to 1 Mbit/s. It can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers. It has three transmit mailboxes, two receive FIFOs with 3 stages and 14 scalable filter banks.

## 3.24 Universal serial bus (USB)

The STM32F303xB/STM32F303xC devices embed an USB device peripheral compatible with the USB full-speed 12 Mbs. The USB interface implements a full-speed (12 Mbit/s) function interface. It has software-configurable endpoint setting and suspend/resume support. The dedicated 48 MHz clock is generated from the internal main PLL (the clock source must use a HSE crystal oscillator). The USB has a dedicated 512-bytes SRAM memory for data transmission and reception.



<sup>1.</sup> X = supported.

#### 3.25 Infrared Transmitter

The STM32F303xB/STM32F303xC devices provide an infrared transmitter solution. The solution is based on internal connections between TIM16 and TIM17 as shown in the figure below.

TIM17 is used to provide the carrier frequency and TIM16 provides the main signal to be sent. The infrared output signal is available on PB9 or PA13.

To generate the infrared remote control signals, TIM16 channel 1 and TIM17 channel 1 must be properly configured to generate correct waveforms. All standard IR pulse modulation modes can be obtained by programming the two timers output compare channels.

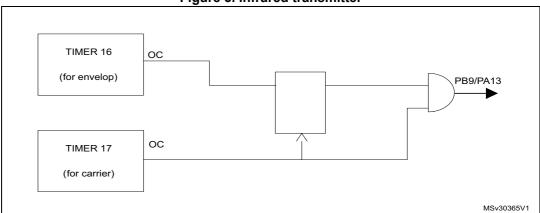


Figure 3. Infrared transmitter

### 3.26 Touch sensing controller (TSC)

The STM32F303xB/STM32F303xC devices provide a simple solution for adding capacitive sensing functionality to any application. These devices offer up to 24 capacitive sensing channels distributed over 8 analog I/O groups.

Capacitive sensing technology is able to detect the presence of a finger near a sensor which is protected from direct touch by a dielectric (glass, plastic, ...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle. It consists of charging the sensor capacitance and then transferring a part of the accumulated charges into a sampling capacitor until the voltage across this capacitor has reached a specific threshold. To limit the CPU bandwidth usage this acquisition is directly managed by the hardware touch sensing controller and only requires few external components to operate.

The touch sensing controller is fully supported by the STMTouch touch sensing firmware library which is free to use and allows touch sensing functionality to be implemented reliably in the end application.

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Table 10. Capacitive sensing GPIOs available on STM32F303xB/STM32F303xC devices

Group	Capacitive sensing signal name	Pin name	Group	Capacitive sensing signal name	Pin name
	TSC_G1_IO1	PA0		TSC_G5_IO1	PB3
1	TSC_G1_IO2	PA1	5	TSC_G5_IO2	PB4
'	TSC_G1_IO3	PA2	3	TSC_G5_IO3	PB6
	TSC_G1_IO4	PA3		TSC_G5_IO4	PB7
	TSC_G2_IO1	PA4		TSC_G6_IO1	PB11
2	TSC_G2_IO2	PA5	6	TSC_G6_IO2	PB12
2	TSC_G2_IO3	PA6		TSC_G6_IO3	PB13
	TSC_G2_IO4	PA7		TSC_G6_IO4	PB14
	TSC_G3_IO1	PC5		TSC_G7_IO1	PE2
3	TSC_G3_IO2	PB0	7	TSC_G7_IO2	PE3
3	TSC_G3_IO3	PB1	,	TSC_G7_IO3	PE4
	TSC_G3_IO4	PB2		TSC_G7_IO4	PE5
	TSC_G4_IO1	PA9		TSC_G8_IO1	PD12
4	TSC_G4_IO2	PA10	8	TSC_G8_IO2	PD13
4	TSC_G4_IO3	PA13	0	TSC_G8_IO3	PD14
	TSC_G4_IO4	PA14		TSC_G8_IO4	PD15

Table 11. No. of capacitive sensing channels available on STM32F303xB/STM32F303xC devices

Analan I/O amaun	Number of capacitive sensing channels					
Analog I/O group	STM32F303Vx	STM32F303Rx	STM32F303Cx			
G1	3	3	3			
G2	3	3	3			
G3	3	3	2			
G4	3	3	3			
G5	3	3	3			
G6	3	3	3			
G7	3	0	0			
G8	3	0	0			
Number of capacitive sensing channels	24	18	17			



### 3.27 Development support

#### 3.27.1 Serial wire JTAG debug port (SWJ-DP)

The Arm SWJ-DP Interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

The JTAG TMS and TCK pins are shared respectively with SWDIO and SWCLK and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

#### 3.27.2 Embedded trace macrocell™

The Arm embedded trace macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32F303xB/STM32F303xC through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. The TPA is connected to a host computer using a high-speed channel. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer running debugger software. TPA hardware is commercially available from common development tool vendors. It operates with third party debugger software tools.



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#### Pinouts and pin description 4

BOOTO PB7 PB6 PB4 PB3 PA15 <u>\_\_\_\_\_\_</u> 48 47 46 45 44 43 42 41 40 39 38 37 VBAT □ 2 □ vss PC13 □ 35 34 PA13 PC14/OSC32 IN 33 PA12 PC15/OSC32\_OUT☐ 4 32 PA11 PF0/OSC\_IN ☐ 5 31 PA10 PF1/OSC\_OUT ☐ 6 LQFP48 30 PA9 NRST □ 7 29 PA8 VSSA/VREF- 8 28 PB15 VDDA □ 27 PB14 PA0 🗖 10 PA1 🗖 11 26 PB13 25 PB12 PA2 PA3 PA4 PA5 PA6 PB0 PB1 PB2 PB10 VSS

Figure 4. STM32F303xB/STM32F303xC LQFP48 pinout

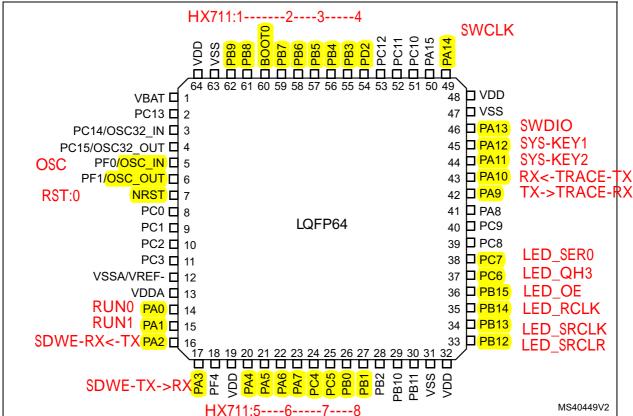


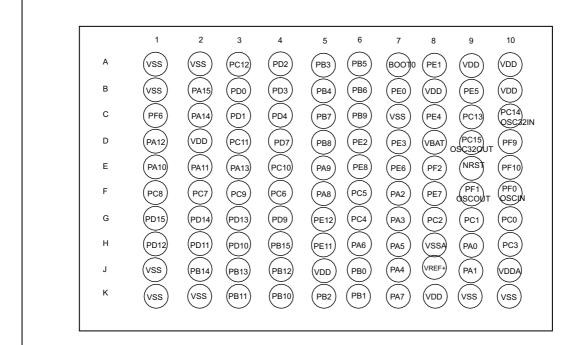
Figure 5. STM32F303xB/STM32F303xC LQFP64 pinout

75 UVDD
74 UVSS
73 PF6
72 PA13
71 PA12
70 PA11
69 PA10 PC13 🗖 PA9
PA9
PA8
PC9
PC8 PC14/OSC32\_IN 68 PC15/OSC32\_OUT = 9 PF9 = 10 PF10 = 11 67 66 65 65 | PC8 64 | PC7 63 | PC6 62 | PD15 61 | PD14 60 | PD13 59 | PD12 58 | PD11 57 | PD10 PF0/OSC\_IN 12 LQFP100 58 PD11 57 PD10 56 PD9 55 PD8 54 PB15 53 PB14 52 PB13 51 PB12 VREF+ □ 21 VDDA □ 22 PA0 □ 23 PA1 □ 24 PA2 □ 25 MS40450V1

Figure 6. STM32F303xB/STM32F303xC LQFP100 pinout

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Figure 7. STM32F303xB/STM32F303xC WLCSP100 pinout



MSv40453V1

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Table 12. Legend/abbreviations used in the pinout table

Name		Abbreviation	Definition	
Pin r	name	Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name		
		S	Supply pin	
Pin	type	I	Input only pin	
		I/O	Input / output pin	
		FT	5 V tolerant I/O	
		FTf	5 V tolerant I/O, FM+ capable	
I/O etr	ucture	TTa 3.3 V tolerant I/O directly connected to ADC		
1/0 511	ucture	TC Standard 3.3V I/O		
		B Dedicated BOOT0 pin		
		RST Bidirectional reset pin with embedded weak pull-up res		
Notes		Unless otherwise specified by a note, all I/Os are set as floating inputs during and after reset		
Pin functions	Alternate functions	Functions selected through GPIOx_AFR registers		
	Additional functions	Functions directly selected/enabled through peripheral registers		

Table 13. STM32F303xB/STM32F303xC pin definitions

	Pin number						Pin functions		
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
D6	1	-	-	PE2	I/O	FT	(1)	TRACECK, TIM3_CH1, TSC_G7_IO1, EVENTOUT	-
D7	2	-	-	PE3	I/O	FT	(1)	TRACED0, TIM3_CH2, TSC_G7_IO2, EVENTOUT	-
C8	3	-	-	PE4	I/O	FT	(1)	TRACED1, TIM3_CH3, TSC_G7_IO3, EVENTOUT	-
В9	4	-	-	PE5	I/O	FT	(1)	TRACED2, TIM3_CH4, TSC_G7_IO4, EVENTOUT	-
E7	5	-	-	PE6	I/O	FT	(1)	TRACED3, EVENTOUT	WKUP3, RTC_TAMP3
D8	6	1	1	$V_{BAT}$	S	-	-	Backup power supply	



Table 13. STM32F303xB/STM32F303xC pin definitions (continued)

	Pin nu	umber						Pin fui	nctions
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
С9	7	2	2	PC13 <sup>(2)</sup>	I/O	TC	-	TIM1_CH1N	WKUP2, RTC_TAMP1, RTC_TS, RTC_OUT
C10	8	3	3	PC14 <sup>(2)</sup> OSC32_IN (PC14)	I/O	TC	-	-	OSC32_IN
D9	9	4	4	PC15 <sup>(2)</sup> OSC32_ OUT (PC15)	I/O	тс	i	-	OSC32_OUT
D10	10	-	-	PF9	I/O	FT	(1)	TIM15_CH1, SPI2_SCK, EVENTOUT	-
E10	11	-	-	PF10	I/O	FT	(1)	TIM15_CH2, SPI2_SCK, EVENTOUT	-
F10	12	5	5	PF0- OSC_IN (PF0)	I/O	FTf	ı	TIM1_CH3N, I2C2_SDA,	OSC_IN
F9	13	6	6	PF1- OSC_OUT (PF1)	I/O	FTf	ı	I2C2_SCL	OSC_OUT
E9	14	7	7	NRST	I/O	RS T		Device reset input / intern	al reset output (active low)
G10	15	8	ı	PC0	I/O	TTa	(1)	EVENTOUT	ADC12_IN6, COMP7_INM
G9	16	9	ı	PC1	I/O	TTa	(1)	EVENTOUT	ADC12_IN7, COMP7_INP
G8	17	10	-	PC2	I/O	TTa	(1)	COMP7_OUT, EVENTOUT	ADC12_IN8
H10	18	11	-	PC3		TTa		TIM1_BKIN2, EVENTOUT	ADC12_IN9
E8	19	-	-	PF2	I/O	TTa	(1)	EVENTOUT	ADC12_IN10
Н8	20	12	8	VSSA/ VREF-	S	-	1	Analog ground/Nega	tive reference voltage
J8	21	-	-	VREF+ <sup>(3)</sup>	S	-	-	Positive refe	rence voltage
J10	22	-	-	VDDA	S	-	•	Analog po	wer supply
-	ı	13	9	VDDA/ VREF+	S	-	-	Analog power supply/Po	ositive reference voltage
H9	23	14	10	PA0	I/O	TTa	USART2_CTS, TIM2_CH1_ETR_TIM8_BKIN_ADC1_IN1, COMP1_INM,		RTC_TAMP2, WKUP1,



Table 13. STM32F303xB/STM32F303xC pin definitions (continued)

	Pin nu	umber						Pin fu	nctions		
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions		
J9	24	15	11	PA1	I/O	TTa	(4)	USART2_RTS_DE, TIM2_CH2, TSC_G1_IO2, TIM15_CH1N, RTC_REFIN, EVENTOUT	ADC1_IN2, COMP1_INP, OPAMP1_VINP, OPAMP3_VINP		
F7	25	16	12	PA2	I/O	ТТа	(4) (5)	USART2_TX, TIM2_CH3, TIM15_CH1, TSC_G1_IO3, COMP2_OUT, EVENTOUT	ADC1_IN3, COMP2_INM, OPAMP1_VOUT		
G7	26	17	13	PA3	I/O	ТТа	(4)	USART2_RX, TIM2_CH4, TIM15_CH2, TSC_G1_IO4, EVENTOUT	ADC1_IN4, OPAMP1_VINP, COMP2_INP, OPAMP1_VINM		
-	27	18	-	PF4	I/O	ТТа	<b>(1)</b> (4)	COMP1_OUT, EVENTOUT	ADC1_IN5		
K9, K10	-	-	-	VSS	S	-	-	Digital	l ground		
K8	28	19	-	VDD	S	-	-	Digital po	wer supply		
J7	29	20	14	PA4	I/O	ТТа	(4) (5)	SPI1_NSS, SPI3_NSS,I2S3_WS, USART2_CK, TSC_G2_IO1, TIM3_CH2, EVENTOUT	ADC2_IN1, DAC1_OUT1, OPAMP4_VINP, COMP1_INM, COMP2_INM, COMP3_INM, COMP4_INM, COMP5_INM, COMP6_INM, COMP7_INM		
H7	30	21	15	PA5	I/O	ТТа	(4) (5)	SPI1_SCK, TIM2_CH1_ETR, TSC_G2_IO2, EVENTOUT	ADC2_IN2, DAC1_OUT2 OPAMP1_VINP, OPAMP2_VINM, OPAMP3_VINP COMP1_INM, COMP2_INM, COMP3_INM, COMP4_INM, COMP5_INM, COMP6_INM, COMP7_INM		
H6	31	22	16	PA6	I/O	TTa	(4) SPI1_MISO, TIM3_CH1, TIM8_BKIN, TIM1_BKIN, TIM16_CH1, COMP1_OUT, TSC_G2_IO3, EVENTOUT		ADC2_IN3, OPAMP2_VOUT		
K7	32	23	17	PA7	I/O	TTa	COMP2_OUT, EVENTOUT  OPAMP1_VINP				
G6	33	24	-	PC4	I/O	ТТа	(1)				

Table 13. STM32F303xB/STM32F303xC pin definitions (continued)

	Pin nı	ımber						Pin fu	nctions		
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions		
F6	34	25	-	PC5	I/O	тта	(1)	USART1_RX, TSC_G3_IO1, EVENTOUT	ADC2_IN11, OPAMP2_VINM, OPAMP1_VINM		
J6	35	26	18	PB0	I/O	ТТа	-	TIM3_CH3, TIM1_CH2N, TIM8_CH2N,TSC_G3_IO2, EVENTOUT	ADC3_IN12, COMP4_INP, OPAMP3_VINP, OPAMP2_VINP		
K6	36	27	19	PB1	I/O	ТТа	(4) (5)	TIM3_CH4, TIM1_CH3N, TIM8_CH3N, COMP4_OUT, TSC_G3_IO3, EVENTOUT	ADC3_IN1, OPAMP3_VOUT-		
K5	37	28	20	PB2	I/O	ТТа	(1) TIM1_ETR, EVENTOUT ADC3_IN13, COMP4_INI				
F8	38	-	-	PE7	I/O	TTa	Ta (1) TIM1_ETR, EVENTOUT ADC3_IN13, COMP4_INP Ta (1) TIM1_CH1N, EVENTOUT COMP4_INM, ADC34_IN6				
E6	39	-	-	PE8	I/O	TTa	(1)	(1) TIM1_CH1N, EVENTOUT COMP4_INM, ADC34			
-	40	-	-	PE9	I/O	ТТа	(4) (1)	TIM1_CH1, EVENTOUT	ADC3_IN2		
-	41	-	-	PE10	I/O	TTa	(1)	TIM1_CH2N, EVENTOUT	ADC3_IN14		
H5	42	-	-	PE11	I/O	TTa	(1)	TIM1_CH2, EVENTOUT	ADC3_IN15		
G5	43	-	-	PE12	I/O	TTa	(1)	TIM1_CH3N, EVENTOUT	ADC3_IN16		
-	44	-	-	PE13	I/O	TTa	(1)	TIM1_CH3, EVENTOUT	ADC3_IN3		
-	45	-	-	PE14	I/O	ТТа	(4) (1)	TIM1_CH4, TIM1_BKIN2, EVENTOUT	ADC4_IN1		
-	46	-	-	PE15	I/O	ТТа	(4) (1)	USART3_RX, TIM1_BKIN, EVENTOUT	ADC4_IN2		
K4	47	29	21	PB10	I/O	ТТа	-	USART3_TX, TIM2_CH3, TSC_SYNC, EVENTOUT	COMP5_INM, OPAMP4_VINM, OPAMP3_VINM		
K3	48	30	22	PB11	I/O	ТТа	-	USART3_RX, TIM2_CH4, TSC_G6_IO1, EVENTOUT	COMP6_INP, OPAMP4_VINP		
K1, J1, K2	49	31	23	VSS	S	-	- Digital ground				
J5	50	32	24	VDD	S	-	- Digital power supply				
J4	51	33	25	PB12	I/O	ТТа	(4) SPI2_NSS,I2S2_WS,I2C2_S MBA, USART3_CK, TIM1_BKIN, TSC_G6_IO2, EVENTOUT ADC4_IN3, COMP3_INM, OPAMP4_VOUT				



Table 13. STM32F303xB/STM32F303xC pin definitions (continued)

	Pin nu	ımber						•	nctions	
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	Notes  Alternate functions			Additional functions	
J3	52	34	26	PB13	I/O	ТТа	(4)		ADC3_IN5, COMP5_INP, OPAMP4_VINP, OPAMP3_VINP	
J2	53	35	27	PB14	I/O	TTa	(4)	USART3_RTS_DE, TIM1_CH2N, TIM15_CH1,	COMP3_INP, ADC4_IN4, OPAMP2_VINP	
H4	54	36	28	PB15	I/O	ТТа	(4)	TIM1_CH3N, RTC_REFIN, TIM15_CH1N, TIM15_CH2,	ADC4_IN5, COMP6_INM	
-	55	-	-	PD8	I/O	TTa	(1)	USART3_TX, EVENTOUT	ADC4_IN12, OPAMP4_VINM	
G4	56	-	-	PD9	I/O	TTa	(1)	USART3_RX, EVENTOUT	ADC4_IN13	
Н3	57	ı	ı	PD10	9	TTa	(1)	USART3_CK, EVENTOUT	ADC34_IN7, COMP6_INM	
H2	58	ı	ı	PD11	I/O	TTa	(1)	USART3_CTS, EVENTOUT	ADC34_IN8, COMP6_INP, OPAMP4_VINP	
H1	59	ı	1	PD12	1/0	TTa	(1)	TIM4_CH1, TSC_G8_IO1,	ADC34_IN9, COMP5_INP	
G3	60	1	-	PD13	I/O	TTa	(1)		ADC34_IN10, COMP5_INM	
G2	61	1	-	PD14	I/O	TTa	(1)		COMP3_INP, ADC34_IN11, OPAMP2_VINP	
G1	62	ı	ı	PD15	1/0	TTa	(1)	SPI2_NSS, TIM4_CH4, TSC_G8_IO4, EVENTOUT	COMP3_INM	
F4	63	37	1	PC6	I/O	FT	(1)	I2S2_MCK, COMP6_OUT, TIM8_CH1, TIM3_CH1, EVENTOUT	-	
F2	64	38	1	PC7	I/O	FT	(1)	I2S3_MCK, TIM8_CH2, TIM3_CH2, COMP5_OUT, EVENTOUT	-	
F1	65	39	ı	PC8	I/O	FT	(1)	TIM8_CH3, TIM3_CH3, COMP3_OUT, EVENTOUT	-	
F3	66	40	-	PC9	I/O	FT	(1)	TIM8_CH4, TIM8_BKIN2,TIM3_CH4, I2S_CKIN, EVENTOUT	-	

Table 13. STM32F303xB/STM32F303xC pin definitions (continued)

	Pin nu	umber						<u>-</u>	2_MCK, M1_CH1, ), VENTOUT  MCK, M1_CH2, 5_BKIN, DMP5_OUT,  RT1_RX, CCH4, 17_BKIN, DMP6_OUT,  USB_DM, CH1N,			
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions			
F5	67	41	29	PA8	I/O	FT	-	I2C2_SMBA,I2S2_MCK, USART1_CK, TIM1_CH1, TIM4_ETR, MCO, COMP3_OUT, EVENTOUT	-			
E5	68	42	30	PA9	I/O	FTf	-	I2C2_SCL,I2S3_MCK, USART1_TX, TIM1_CH2, TIM2_CH3, TIM15_BKIN, TSC_G4_IO1, COMP5_OUT, EVENTOUT	-			
E1	69	43	31	PA10	I/O	FTf	1	I2C2_SDA, USART1_RX, TIM1_CH3, TIM2_CH4, TIM8_BKIN, TIM17_BKIN, TSC_G4_IO2, COMP6_OUT, EVENTOUT	-			
E2	70	44	32	PA11	I/O	FT	1	USART1_CTS, USB_DM, CAN_RX, TIM1_CH1N, TIM1_CH4, TIM1_BKIN2, TIM4_CH1, COMP1_OUT, EVENTOUT	-			
D1	71	45	33	PA12	I/O	FT	1	USART1_RTS_DE, USB_DP, CAN_TX, TIM1_CH2N, TIM1_ETR, TIM4_CH2, TIM16_CH1, COMP2_OUT, EVENTOUT	-			
E3	72	46	34	PA13	I/O	FT	-	USART3_CTS, TIM4_CH3, TIM16_CH1N, TSC_G4_IO3, IR_OUT, SWDIO-JTMS, EVENTOUT	-			
C1	73	-	-	PF6	I/O	FTf	(1)	I2C2_SCL, USART3_RTS_DE, TIM4_CH4, EVENTOUT	-			
A1, A2, B1	74	47	35	VSS	S	-	1	Gro	und			
D2	75	48	36	VDD	S	-	-	Digital pov	wer supply			
C2	76	49	37	PA14	I/O	FTf	I2C1_SDA, USART2_TX,					



Table 13. STM32F303xB/STM32F303xC pin definitions (continued)

	Pin nu	umber						Pin fur	,
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
B2	77	50	38	PA15	I/O	O FTf - SPI3_NSS, I2S3_WS, JTDI, USART2_RX, TIM1_BKIN, TIM2_CH1_ETR, TIM8_CH1, EVENTOUT  O FT (1) SPI3_SCK, I2S3_CK, USART3_TX, UART4_TX, TIM8_CH1N, EVENTOUT		USART2_RX, TIM1_BKIN, TIM2_CH1_ETR, TIM8_CH1, EVENTOUT	-
E4	78	51	-	PC10	I/O	FT	(1)	USART3_TX, UART4_TX,	-
D3	79	52	-	PC11	I/O	FT	(1)	SPI3_MISO, I2S3ext_SD, USART3_RX, UART4_RX, TIM8_CH2N, EVENTOUT	-
А3	80	53	-	PC12	I/O	FT	(1)	SPI3_MOSI, I2S3_SD, USART3_CK, UART5_TX, TIM8_CH3N, EVENTOUT	-
В3	81	-	-	PD0	I/O	FT	(1)	CAN_RX, EVENTOUT	-
С3	82	-	-	PD1	I/O	FT	(1)	CAN_TX, TIM8_CH4, TIM8_BKIN2,EVENTOUT	-
A4	83	54	-	PD2	I/O	FT	(1)	UART5_RX, TIM3_ETR, TIM8_BKIN, EVENTOUT	-
B4	84	-	-	PD3	I/O	FT	(1)	USART2_CTS, TIM2_CH1_ETR, EVENTOUT	-
C4	85	-	-	PD4	I/O	FT	(1)	USART2_RTS_DE, TIM2_CH2, EVENTOUT	-
-	86	-	-	PD5	I/O	FT	(1)	USART2_TX, EVENTOUT	-
-	87	-	-	PD6	I/O	FT	(1)	USART2_RX, TIM2_CH4, EVENTOUT	-
D4	88	-	-	PD7	I/O	FT	(1)	USART2_CK, TIM2_CH3, EVENTOUT	-
A5	89	55	39	PB3	I/O	FT	-	SPI3_SCK, I2S3_CK, SPI1_SCK, USART2_TX, TIM2_CH2, TIM3_ETR, TIM4_ETR, TIM8_CH1N, TSC_G5_IO1, JTDO- TRACESWO, EVENTOUT	<u>-</u>

Table 13. STM32F303xB/STM32F303xC pin definitions (continued)

	Pin nu	ımber						Pin fur	nctions		
WLCSP100	LQFP100	LQFP64	LQFP48	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions		
B5	90	56	40	PB4	I/O	FT	-	SPI3_MISO, I2S3ext_SD, SPI1_MISO, USART2_RX, TIM3_CH1, TIM16_CH1, TIM17_BKIN, TIM8_CH2N, TSC_G5_IO2, NJTRST, EVENTOUT	-		
A6	91	57	41	PB5	I/O	FT	TIM3_CH2, TIM8_CH3N, TIM17_CH1, EVENTOUT  I2C1_SCL, USART1_TX, TIM16_CH1N, TIM4_CH1,				
В6	92	58	42	PB6	I/O	FTf	-		-		
C5	93	59	43	PB7	I/O	FTf	-	I2C1_SDA, USART1_RX, TIM3_CH4, TIM4_CH2, TIM17_CH1N, TIM8_BKIN, TSC_G5_IO4, EVENTOUT	-		
A7	94	60	44	воото	I	В	-	Boot memo	ry selection		
D5	95	61	45	PB8	I/O	FTf	-	I2C1_SCL, CAN_RX, TIM16_CH1, TIM4_CH3, TIM8_CH2, TIM1_BKIN, TSC_SYNC, COMP1_OUT, EVENTOUT	-		
C6	96	62	46	PB9	I/O	FTf	-	I2C1_SDA, CAN_TX, TIM17_CH1, TIM4_CH4, TIM8_CH3, IR_OUT, COMP2_OUT, EVENTOUT	-		
В7	97	1	-	PE0	I/O	FT	(1)	USART1_TX, TIM4_ETR, TIM16_CH1, EVENTOUT	-		
A8	98	ı	ı	PE1	I/O	FT	(1)	USART1_RX, TIM17_CH1, EVENTOUT	-		
C7	99	63	47	VSS	S	-	-	Gro	und		
A9, A10, B10, B8	100	64	48	VDD	S	-	- Digital power supply				



- Function availability depends on the chosen device. When using the small packages (48 and 64 pin packages), the GPIO pins which are not present on these packages, must not be configured in analog mode.
- PC13, PC14 and PC15 are supplied through the power switch. Since the switch sinks only a limited amount of current (3 mA), the use of GPIO PC13 to PC15 in output mode is limited:

   The speed should not exceed 2 MHz with a maximum load of 30 pF
   These GPIOs must not be used as current sources (e.g. to drive an LED).

After the first backup domain power-up, PC13, PC14 and PC15 operate as GPIOs. Their function then depends on the content of the Backup registers which is not reset by the main reset. For details on how to manage these GPIOs, refer to the Battery backup domain and BKP register description sections in the RM0316 reference manual.

- The VREF+ functionality is available only on the 100 pin package. On the 64-pin and 48-pin packages, the VREF+ is internally connected to VDDA.
- 4. Fast ADC channel.
- 5. These GPIOs offer a reduced touch sensing sensitivity. It is thus recommended to use them as sampling capacitor I/O.





Table 14. Alternate functions for port A

Port & Pin Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF14	AF15
PA0	-	TIM2_ CH1_ ETR	-	TSC_ G1_IO1	-	-	-	USART2_ CTS	COMP1 _OUT	TIM8_ BKIN	TIM8_ ETR	-	-	-	EVENT OUT
PA1	RTC_ REFIN	TIM2_ CH2	-	TSC_ G1_IO2	-	-	-	USART2_ RTS_DE		TIM15_ CH1N	-	-	-	-	EVENT OUT
PA2	-	TIM2_ CH3	-	TSC_ G1_IO3	-	-	-	USART2_ TX	COMP2 _OUT	TIM15_ CH1	-	-	-	-	EVENT OUT
PA3	-	TIM2_ CH4	1	TSC_ G1_IO4	-	-	-	USART2_ RX	-	TIM15_ CH2	-	-	-	-	EVENT OUT
PA4	-	-	TIM3_ CH2	TSC_ G2_IO1	-	SPI1_ NSS	SPI3_NSS, I2S3_WS	USART2_ CK	-	-	-	-	-	-	EVENT OUT
PA5	-	TIM2_ CH1_ ETR	-	TSC_ G2_IO2	-	SPI1_ SCK	-	-	-	-	-	-	-	-	EVENT OUT
PA6	-	TIM16_ CH1	TIM3_ CH1	TSC_ G2_IO3	TIM8_ BKIN	SPI1_ MISO	TIM1_BKIN	-	COMP1 _OUT	-	-	-	-	-	EVENT OUT
PA7	-	TIM17_ CH1	TIM3_ CH2	TSC_ G2_IO4	TIM8_ CH1N	SPI1_ MOSI	TIM1_CH1N	-	COMP2 _OUT	-	-	-	-	-	EVENT OUT
PA8	МСО	-	-	-	I2C2_ SMBA	I2S2_ MCK	TIM1_CH1	USART1_ CK	COMP3 _OUT	-	TIM4_ ETR	-	-	-	EVENT OUT
PA9	-	-	-	TSC_ G4_IO1	I2C2_ SCL	I2S3_ MCK	TIM1_CH2	USART1_ TX	COMP5 _OUT	TIM15_ BKIN	TIM2_ CH3	-	-	-	EVENT OUT
PA10	-	TIM17_ BKIN	-	TSC_ G4_IO2	I2C2_ SDA	-	TIM1_CH3	USART1_ RX	COMP6 _OUT	-	TIM2_ CH4	TIM8_BKIN	-	-	EVENT OUT
PA11	-	-	-	-	-	-	TIM1_CH1N	USART1_ CTS	COMP1 _OUT	CAN_RX	TIM4_ CH1	TIM1_CH4	TIM1_ BKIN2	USB_ DM	EVENT OUT

Tabl	e 14. A	lternate func	tions for p	ort A (co	ntinued)	
						7

Port & Pin Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF14	AF15
PA12	-	TIM16_ CH1	-	-	-	-	TIM1_CH2N	USART1_ RTS_DE	COMP2 _OUT	CAN_TX	TIM4_ CH2	TIM1_ETR	-	USB_ DP	EVENT OUT
PA13	SWDIO -JTMS	TIM16_ CH1N	-	TSC_ G4_IO3	-	IR_ OUT	-	USART3_ CTS	-	-	TIM4_ CH3	-	-	-	EVENT OUT
PA14	SWCLK -JTCK	-	-	TSC_ G4_IO4	I2C1_ SDA	TIM8_ CH2	TIM1_BKIN	USART2_ TX	-	-	-	-	-	-	EVENT OUT
PA15	JTDI	TIM2_ CH1_ ETR	TIM8_ CH1	-	I2C1_ SCL	SPI1_ NSS	SPI3_NSS, I2S3_WS	USART2_ RX	-	TIM1_ BKIN	-	-	-	-	EVENT OUT



Table 15. Alternate functions for port B

		1	1	1			e iulicuolis i	<b> </b>	ı	1		,	1
Port & Pin Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF12	AF15
PB0	-	-	TIM3_ CH3	TSC_ G3_IO2	TIM8_ CH2N	-	TIM1_CH2N	-	-	-	-	-	EVENT OUT
PB1	-	-	TIM3_ CH4	TSC_ G3_IO3	TIM8_ CH3N	-	TIM1_CH3N	-	COMP4_ OUT	-	-	-	EVENT OUT
PB2	-	-	-	TSC_ G3_IO4	-	-	-	-	-	-	-	-	EVENT OUT
PB3	JTDO- TRACES WO	TIM2_ CH2	TIM4_ ETR	TSC_ G5_IO1	TIM8_ CH1N	SPI1_ SCK	SPI3_SCK, I2S3_CK	USART2_ TX	-	-	TIM3_ ETR	-	EVENT OUT
PB4	NJTRST	TIM16_ CH1	TIM3_ CH1	TSC_ G5_IO2	TIM8_ CH2N	SPI1_ MISO	SPI3_MISO, I2S3ext_SD	USART2_ RX	-	-	TIM17_ BKIN	-	EVENT OUT
PB5	-	TIM16_ BKIN	TIM3_ CH2	TIM8_ CH3N	I2C1_ SMBA	SPI1_ MOSI	SPI3_MOSI, I2S3_SD	USART2_ CK	-	-	TIM17_ CH1	-	EVENT OUT
PB6	-	TIM16_ CH1N	TIM4_ CH1	TSC_ G5_IO3	I2C1_SCL	TIM8_CH1	TIM8_ ETR	USART1_ TX	-	-	TIM8_ BKIN2	-	EVENT OUT
PB7	-	TIM17_ CH1N	TIM4_ CH2	TSC_ G5_IO4	I2C1_ SDA	TIM8_ BKIN	-	USART1_ RX	-	-	TIM3_ CH4	-	EVENT OUT
PB8	-	TIM16_ CH1	TIM4_ CH3	TSC_ SYNC	I2C1_SCL	-	-	-	COMP1_ OUT	CAN_RX	TIM8_ CH2	TIM1_ BKIN	EVENT OUT
PB9	-	TIM17_ CH1	TIM4_ CH4		I2C1_ SDA	-	IR_OUT	-	COMP2_ OUT	CAN_TX	TIM8_ CH3	-	EVENT OUT
PB10	-	TIM2_ CH3	-	TSC_ SYNC	-	-	-	USART3_ TX	-	-	-	-	EVENT OUT
PB11	-	TIM2_ CH4	-	TSC_ G6_IO1	-	-	-	USART3_ RX	-	-	-	-	EVENT OUT
PB12	-	-	-	TSC_ G6_IO2	I2C2_ SMBA	SPI2_NSS, I2S2_WS	TIM1_ BKIN	USART3_ CK	-	-	-	-	EVENT OUT

Pinouts and pin description

Table 15. Alternate functions for port B (continued)

Port & Pin Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF12	AF15
PB13	-	-	-	TSC_ G6_IO3	-	SPI2_SCK, I2S2_CK	TIM1_ CH1N	USART3_ CTS	-	-	-	-	EVENT OUT
PB14	-	TIM15_ CH1	-	TSC_ G6_IO4	-	SPI2_MISO, I2S2ext_SD	TIM1_ CH2N	USART3_ RTS_DE	-	-	-	-	EVENT OUT
PB15	RTC_ REFIN	TIM15_ CH2	TIM15_ CH1N	-	TIM1_ CH3N	SPI2_MOSI, I2S2_SD	-	-	-	-	-	-	EVENT OUT



Table 16. Alternate functions for port C

					<u> </u>		
Port & Pin Name	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PC0	EVENTOUT	-	-	-	-	-	-
PC1	EVENTOUT	-	-	-	-	-	-
PC2	EVENTOUT	-	COMP7_OUT	-	-	-	-
PC3	EVENTOUT	-	-	-	-	TIM1_BKIN2	-
PC4	EVENTOUT	-	-	-	-	-	USART1_TX
PC5	EVENTOUT	-	TSC_G3_IO1	-	-	-	USART1_RX
PC6	EVENTOUT	TIM3_CH1	-	TIM8_CH1	-	I2S2_MCK	COMP6_OUT
PC7	EVENTOUT	TIM3_CH2	-	TIM8_CH2	-	12S3_MCK	COMP5_OUT
PC8	EVENTOUT	TIM3_CH3	-	TIM8_CH3	-	-	COMP3_OUT
PC9	EVENTOUT	TIM3_CH4	-	TIM8_CH4	I2S_CKIN	TIM8_BKIN2	-
PC10	EVENTOUT	-	-	TIM8_CH1N	UART4_TX	SPI3_SCK, I2S3_CK	USART3_TX
PC11	EVENTOUT	-	-	TIM8_CH2N	UART4_RX	SPI3_MISO, I2S3ext_SD	USART3_RX
PC12	EVENTOUT	-	-	TIM8_CH3N	UART5_TX	SPI3_MOSI, I2S3_SD	USART3_CK
PC13	-	-	-	TIM1_CH1N	-	-	-
PC14	-	-	-	-	-	-	-
PC15	-	-	-	-	-		-

Table 17. Alternate functions for port D

Port & Pin Name	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PD0	EVENTOUT	-	-	-	-	-	CAN_RX
PD1	EVENTOUT	-	-	TIM8_CH4	-	TIM8_BKIN2	CAN_TX
PD2	EVENTOUT	TIM3_ETR	-	TIM8_BKIN	UART5_RX	-	-
PD3	EVENTOUT	TIM2_CH1_ETR	-	-	-	-	USART2_CTS
PD4	EVENTOUT	TIM2_CH2	-	-	-	-	USART2_RTS_DE
PD5	EVENTOUT	-	-	-	-	-	USART2_TX
PD6	EVENTOUT	TIM2_CH4	-	-	-	-	USART2_RX
PD7	EVENTOUT	TIM2_CH3	-	-	-	-	USART2_CK
PD8	EVENTOUT	-	-	-	-	-	USART3_TX
PD9	EVENTOUT	-	-	-	-	-	USART3_RX
PD10	EVENTOUT	-	-	-	-	-	USART3_CK
PD11	EVENTOUT	-	-	-	-	-	USART3_CTS
PD12	EVENTOUT	TIM4_CH1	TSC_G8_IO1	-	-	-	USART3_RTS_DE
PD13	EVENTOUT	TIM4_CH2	TSC_G8_IO2	-	-	-	-
PD14	EVENTOUT	TIM4_CH3	TSC_G8_IO3	-	-	-	-
PD15	EVENTOUT	TIM4_CH4	TSC_G8_IO4	-	-	SPI2_NSS	-





Table 18. Alternate functions for port E

Port & Pin Name	AF0	AF1	AF2	AF3	AF4	AF6	AF7
PE0	-	EVENTOUT	TIM4_ETR	-	TIM16_CH1	-	USART1_TX
PE1	-	EVENTOUT	-	-	TIM17_CH1	-	USART1_RX
PE2	TRACECK	EVENTOUT	TIM3_CH1	TSC_G7_IO1	-	-	-
PE3	TRACED0	EVENTOUT	TIM3_CH2	TSC_G7_IO2	-	-	-
PE4	TRACED1	EVENTOUT	TIM3_CH3	TSC_G7_IO3	-	-	-
PE5	TRACED2	EVENTOUT	TIM3_CH4	TSC_G7_IO4	-	-	-
PE6	TRACED3	EVENTOUT		-	-	-	-
PE7	-	EVENTOUT	TIM1_ETR	-	-	-	-
PE8	-	EVENTOUT	TIM1_CH1N	-	-	-	-
PE9	-	EVENTOUT	TIM1_CH1	-	-	-	-
PE10	-	EVENTOUT	TIM1_CH2N	-	-	-	-
PE11	-	EVENTOUT	TIM1_CH2	-	-	-	-
PE12	-	EVENTOUT	TIM1_CH3N	-	-	-	-
PE13	-	EVENTOUT	TIM1_CH3	-	-	-	-
PE14	-	EVENTOUT	TIM1_CH4	-	-	TIM1_BKIN2	-
PE15	-	EVENTOUT	TIM1_BKIN	-	-		USART3_RX

**EVENTOUT** 

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Table 19. Alternate functions for port F Port & AF2 AF4 AF7 AF1 AF3 AF5 AF6 Pin Name I2C2\_SDA TIM1\_CH3N PF0 PF1 I2C2\_SCL PF2 **EVENTOUT** PF4 **EVENTOUT** COMP1\_OUT PF6 **EVENTOUT** TIM4\_CH4 I2C2\_SCL USART3\_RTS\_DE PF9 **EVENTOUT** TIM15\_CH1 SPI2\_SCK

SPI2\_SCK

TIM15\_CH2

## 5 Memory mapping

0x5000 07FF AHB3 0xFFFF FFFF 0x5000 0000 Cortex-M4 Reserved with FPU 7 0x4800 1800 Internal Peripherals AHB2 0xE000 0000 0x4800 0000 Reserved 6 0x4002 43FF AHB1 0xC000 0000 0x4002 0000 Reserved 5 0x4001 6C00 APB2 0xA000 0000 0x4001 0000 Reserved 4 0x4000 A000 APB1 0x8000 0000 0x4000 0000 3 ,0x1FFF FFFF Option bytes 0x6000 0000 0x1FFF F800 System memory 2 0x1FFF D800 Reserved 0x1000 2000 0x4000 0000 Peripherals **CCM RAM** 0x1000 0000 Reserved 1 0x0804 0000 0x2000 0000 **SRAM** Flash memory 0x0800 0000 0 CODE Reserved 0x0004 0000 Flash, system 0x0000 0000 memory or SRAM, depending on BOOT configuration Reserved 0x0000 00000 MSv30355V2

Figure 8. STM32F303xB/STM32F303xC memory map

Table 20. STM32F303xB/STM32F303xC memory map, peripheral register boundary addresses  $^{(1)}$ 

Bus	Boundary address	Size (bytes)	Peripheral
AHB3	0x5000 0400 - 0x5000 07FF	1 K	ADC3 - ADC4
Ands	0x5000 0000 - 0x5000 03FF	1 K	ADC1 - ADC2
	0x4800 1800 - 0x4FFF FFFF	~132 M	Reserved
	0x4800 1400 - 0x4800 17FF	1 K	GPIOF
	0x4800 1000 - 0x4800 13FF	1 K	GPIOE
AHB2	0x4800 0C00 - 0x4800 0FFF	1 K	GPIOD
ANDZ	0x4800 0800 - 0x4800 0BFF	1 K	GPIOC
	0x4800 0400 - 0x4800 07FF	1 K	GPIOB
	0x4800 0000 - 0x4800 03FF	1 K	GPIOA
	0x4002 4400 - 0x47FF FFFF	~128 M	Reserved
	0x4002 4000 - 0x4002 43FF	1 K	TSC
	0x4002 3400 - 0x4002 3FFF	3 K	Reserved
	0x4002 3000 - 0x4002 33FF	1 K	CRC
	0x4002 2400 - 0x4002 2FFF	3 K	Reserved
ALIDA	0x4002 2000 - 0x4002 23FF	1 K	Flash interface
AHB1	0x4002 1400 - 0x4002 1FFF	3 K	Reserved
	0x4002 1000 - 0x4002 13FF	1 K	RCC
	0x4002 0800 - 0x4002 0FFF	2 K	Reserved
	0x4002 0400 - 0x4002 07FF	1 K	DMA2
	0x4002 0000 - 0x4002 03FF	1 K	DMA1
	0x4001 8000 - 0x4001 FFFF	32 K	Reserved
	0x4001 4C00 - 0x4001 7FFF	13 K	Reserved
	0x4001 4800 - 0x4001 4BFF	1 K	TIM17
	0x4001 4400 - 0x4001 47FF	1 K	TIM16
	0x4001 4000 - 0x4001 43FF	1 K	TIM15
	0x4001 3C00 - 0x4001 3FFF	1 K	Reserved
	0x4001 3800 - 0x4001 3BFF	1 K	USART1
APB2	0x4001 3400 - 0x4001 37FF	1 K	TIM8
	0x4001 3000 - 0x4001 33FF	1 K	SPI1
	0x4001 2C00 - 0x4001 2FFF	1 K	TIM1
	0x4001 0800 - 0x4001 2BFF	9 K	Reserved
	0x4001 0400 - 0x4001 07FF	1 K	EXTI
	0x4001 0000 - 0x4001 03FF	1 K	SYSCFG + COMP + OPAMP

Table 20. STM32F303xB/STM32F303xC memory map, peripheral register boundary addresses<sup>(1)</sup> (continued)

Bus	Boundary address	Size (bytes)	Peripheral
	0x4000 8000 - 0x4000 FFFF	32 K	Reserved
	0x4000 7800 - 0x4000 7FFF	2 K	Reserved
	0x4000 7400 - 0x4000 77FF	1 K	DAC (dual)
	0x4000 7000 - 0x4000 73FF	1 K	PWR
	0x4000 6800 - 0x4000 6FFF	2 K	Reserved
	0x4000 6400 - 0x4000 67FF	1 K	bxCAN
	0x4000 6000 - 0x4000 63FF	1 K	USB SRAM 512 bytes
	0x4000 5C00 - 0x4000 5FFF	1 K	USB device FS
	0x4000 5800 - 0x4000 5BFF	1 K	I2C2
	0x4000 5400 - 0x4000 57FF	1 K	I2C1
	0x4000 5000 - 0x4000 53FF	1 K	UART5
	0x4000 4C00 - 0x4000 4FFF	1 K	UART4
	0x4000 4800 - 0x4000 4BFF	1 K	USART3
	0x4000 4400 - 0x4000 47FF	1 K	USART2
APB1	0x4000 4000 - 0x4000 43FF	1 K	I2S3ext
AFBI	0x4000 3C00 - 0x4000 3FFF	1 K	SPI3/I2S3
	0x4000 3800 - 0x4000 3BFF	1 K	SPI2/I2S2
	0x4000 3400 - 0x4000 37FF	1 K	I2S2ext
	0x4000 3000 - 0x4000 33FF	1 K	IWDG
	0x4000 2C00 - 0x4000 2FFF	1 K	WWDG
	0x4000 2800 - 0x4000 2BFF	1 K	RTC
	0x4000 1800 - 0x4000 27FF	4 K	Reserved
	0x4000 1400 - 0x4000 17FF	1 K	TIM7
	0x4000 1000 - 0x4000 13FF	1 K	TIM6
	0x4000 0C00 - 0x4000 0FFF	1 K	Reserved
	0x4000 0800 - 0x4000 0BFF	1 K	TIM4
	0x4000 0400 - 0x4000 07FF	1 K	TIM3
	0x4000 0000 - 0x4000 03FF	1 K	TIM2

<sup>1.</sup> The gray color is used for reserved Flash memory addresses.

### 6 Electrical characteristics

#### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

#### 6.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25$  °C and  $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean±30).

### 6.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A$  = 25 °C,  $V_{DD}$  =  $V_{DDA}$  = 3.3 V. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean±2 $\sigma$ ).

## 6.1.3 Typical curves

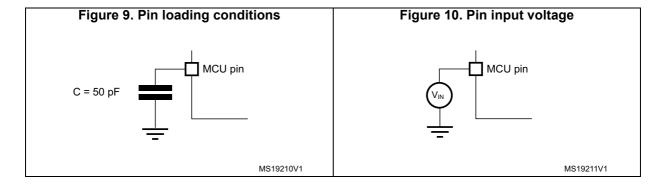
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

### 6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 9.

#### 6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 10.



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### 6.1.6 Power supply scheme

Backup circuitry Power (LSE, RTC, switch Wakeup logic, Backup registers) OUT GP I/Os I/O logic Kernel logic (CPU, digital & memories)  $4 \; x \; V_{DD}$ Regulator 4 x 100 nF 4 x V<sub>SS</sub> + 1 x 4.7  $\mu$ F  $V_{\text{DDA}}$  $V_{DDA}$ V<sub>REF+</sub> Analog: RCs, 10 nF ADC/DAC PLL,comparators, OPAMP, + 1 µF  $V_{REF}$ MS19875V5

Figure 11. Power supply scheme

 Dotted lines represent the internal connections on low pin count packages, joining the dedicated supply pins.

#### Caution:

Each power supply pair ( $V_{DD}/V_{SS}$ ,  $V_{DDA}/V_{SSA}$  etc..) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

### 6.1.7 Current consumption measurement

IDD\_VBAT
VBAT
IDD
VDD
VDD
VDD

MS19213V1

Figure 12. Current consumption measurement scheme

## 6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 21: Voltage characteristics*, *Table 22: Current characteristics*, and *Table 23: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Symbol	Ratings	Min	Max	Unit
V <sub>DD</sub> -V <sub>SS</sub>	External main supply voltage (including $V_{DDA,\ }V_{BAT}$ and $V_{DD})$	-0.3	4.0	
V <sub>DD</sub> -V <sub>DDA</sub>	Allowed voltage difference for V <sub>DD</sub> > V <sub>DDA</sub>	-	0.4	
V <sub>REF+</sub> -V <sub>DDA</sub> <sup>(2)</sup>	Allowed voltage difference for V <sub>REF+</sub> > V <sub>DDA</sub>	-	0.4	]
	Input voltage on FT and FTf pins	V <sub>SS</sub> -0.3	V <sub>DD</sub> + 4.0	V
V <sub>IN</sub> <sup>(3)</sup>	Input voltage on TTa pins	V <sub>SS</sub> -0.3	4.0	
VIN.	Input voltage on any other pin	V <sub>SS</sub> -0.3	4.0	
	Input voltage on Boot0 pin	0	9	
ΔV <sub>DDx</sub>	Variations between different V <sub>DD</sub> power pins	-	50	mV
V <sub>SSX</sub> -V <sub>SS</sub>	Variations between all the different ground pins <sup>(4)</sup>	-	50	1111
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	see Section 6.3.12: Electrical sensitivity characteristics		-

Table 21. Voltage characteristics<sup>(1)</sup>



All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range. The following relationship must be respected between V<sub>DDA</sub> and V<sub>DD</sub>: V<sub>DDA</sub> must power on before or at the same time as V<sub>DD</sub> in the power up sequence. V<sub>DDA</sub> must be greater than or equal to V<sub>DD</sub>.

- V<sub>REF+</sub> must be always lower or equal than V<sub>DDA</sub> (V<sub>REF+</sub> ≤V<sub>DDA</sub>). If unused then it must be connected to V<sub>DDA</sub>.
- V<sub>IN</sub> maximum must always be respected. Refer to Table 22: Current characteristics for the maximum allowed injected current values.
- 4. Include VREF- pin.

**Table 22. Current characteristics** 

Symbol	Ratings	Max.	Unit
$\Sigma I_{VDD}$	Total current into sum of all V <sub>DD</sub> power lines (source)	160	
Σl <sub>VSS</sub>	Total current out of sum of all V <sub>SS</sub> ground lines (sink)	-160	
I <sub>VDD</sub>	Maximum current into each V <sub>DD</sub> power line (source) <sup>(1)</sup>	100	
I <sub>VSS</sub>	Maximum current out of each V <sub>SS</sub> ground line (sink) <sup>(1)</sup>	-100	
	Output current sunk by any I/O and control pin	25	
I <sub>IO(PIN)</sub>	Output current source by any I/O and control pin	-25	mA
ΣΙ	Total output current sunk by sum of all IOs and control pins <sup>(2)</sup>	80	T IIIA
$\Sigma I_{IO(PIN)}$	Total output current sourced by sum of all IOs and control pins <sup>(2)</sup>	-80	
	Injected current on FT, FTf and B pins <sup>(3)</sup>	-5/+0	
I <sub>INJ(PIN)</sub>	Injected current on TC and RST pin <sup>(4)</sup>	± 5	
	Injected current on TTa pins <sup>(5)</sup>	± 5	
$\Sigma I_{\text{INJ(PIN)}}$	Total injected current (sum of all I/O and control pins) <sup>(6)</sup>	± 25	1

- All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub> and V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.
- This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count LQFP packages.
- Positive injection is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
- A positive injection is induced by V<sub>IN</sub> > V<sub>DD</sub> while a negative injection is induced by V<sub>IN</sub> < V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer to *Table 21: Voltage characteristics* for the maximum allowed input voltage values.
- 5. A positive injection is induced by V<sub>IN</sub> > V<sub>DDA</sub> while a negative injection is induced by V<sub>IN</sub> < V<sub>SS</sub>. I<sub>INJ</sub>(PIN) must never be exceeded. Refer also to *Table 21: Voltage characteristics* for the maximum allowed input voltage values. Negative injection disturbs the analog performance of the device. See note <sup>(2)</sup> below *Table 70*.
- When several inputs are submitted to a current injection, the maximum ΣI<sub>INJ(PIN)</sub> is the absolute sum of the positive and negative injected currents (instantaneous values).

Table 23. Thermal characteristics

Symbol	Ratings	Value	Unit
T <sub>STG</sub>	Storage temperature range	-65 to +150	°C
T <sub>J</sub>	Maximum junction temperature	150	°C



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# 6.3 Operating conditions

### 6.3.1 General operating conditions

Table 24. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit		
f <sub>HCLK</sub>	Internal AHB clock frequency	-	0	72			
f <sub>PCLK1</sub>	Internal APB1 clock frequency	-	0	36	MHz		
f <sub>PCLK2</sub>	Internal APB2 clock frequency	-	0	72			
$V_{DD}$	Standard operating voltage	-	2	3.6	V		
V	Analog operating voltage (OPAMP and DAC not used)	Must have a potential	2	3.6	V		
$V_{DDA}$	Analog operating voltage (OPAMP and DAC used)	V <sub>DD</sub>	2.4	3.6	V		
$V_{BAT}$	Backup operating voltage	-	1.65	3.6	V		
		TC I/O	-0.3	V <sub>DD</sub> +0.3			
W	I/O input voltage	TTa I/O	-0.3	V <sub>DDA</sub> +0.3	V		
$V_{IN}$		FT and FTf I/O <sup>(1)</sup>	-0.3	5.5	V		
		воото	0	5.5			
		WLCSP100	-	500	- mW		
$P_{D}$	Power dissipation at T <sub>A</sub> = 85 °C for suffix 6 or T <sub>A</sub> =	LQFP100	-	488			
гD	105 °C for suffix 7 <sup>(2)</sup>	LQFP64	-	444			
		- 0 72 - 0 36 - 0 72 - 0 72 - 2 3.6 - 2 3.6 - Must have a potential equal to or higher than VDD 2.4 3.6 - 1.65 3.6 - 1.65 3.6 - TC I/O -0.3 VDD+0.3 - TTa I/O -0.3 VDDA+0.3 - TTand FTf I/O(1) -0.3 5.5 - BOOTO 0 5.5 - WLCSP100 - 488					
	Ambient temperature for 6 suffix version		-40	85	°C		
TA	Sullix version	Low-power dissipation <sup>(3)</sup>	-40	105			
IA	Ambient temperature for 7 suffix version		-40	105	°C		
	Sullix version	Low-power dissipation <sup>(3)</sup>	-40	125			
TJ	lunction tomporature range	6 suffix version	-40	105	°C		
IJ	Junction temperature range	7 suffix version	-40	125	- "C		

<sup>1.</sup> To sustain a voltage higher than  $V_{DD}$ +0.3 V, the internal pull-up/pull-down resistors must be disabled.



If T<sub>A</sub> is lower, higher P<sub>D</sub> values are allowed as long as T<sub>J</sub> does not exceed T<sub>Jmax</sub> (see Section 7.5: Thermal characteristics).

<sup>3.</sup> In low-power dissipation state, T<sub>A</sub> can be extended to this range as long as T<sub>J</sub> does not exceed T<sub>Jmax</sub> (see *Section 7.5: Thermal characteristics*).

### 6.3.2 Operating conditions at power-up / power-down

The parameters given in *Table 25* are derived from tests performed under the ambient temperature condition summarized in *Table 24*.

Table 25. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit	
t <sub>VDD</sub>	V <sub>DD</sub> rise time rate		0	∞		
	V <sub>DD</sub> fall time rate	-	20	∞	μs/V	
	V <sub>DDA</sub> rise time rate		0	∞	μ5/ ν	
t <sub>VDDA</sub>	V <sub>DDA</sub> fall time rate	-	20	∞		

### 6.3.3 Embedded reset and power control block characteristics

The parameters given in *Table 26* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 24*.

Table 26. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>POR/PDR</sub> <sup>(1)</sup>	Power on/power down	Falling edge	1.8 <sup>(2)</sup>	1.88	1.96	V
	reset threshold	Rising edge	1.84	1.92	2.0	V
V <sub>PDRhyst</sub> <sup>(1)</sup>	PDR hysteresis	-	-	40	-	mV
t <sub>RSTTEMPO</sub> <sup>(3)</sup>	POR reset temporization	-	1.5	2.5	4.5	ms

<sup>1.</sup> The PDR detector monitors  $V_{DD}$  and also  $V_{DDA}$  (if kept enabled in the option bytes). The POR detector monitors only  $V_{DD}$ .

<sup>2.</sup> The product behavior is guaranteed by design down to the minimum V<sub>POR/PDR</sub> value.

<sup>3.</sup> Guaranteed by design.

Table 27. Programmable voltage detector characteristics

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Тур	Max <sup>(1)</sup>	Unit
V <sub>PVD0</sub>	PVD threshold 0	Rising edge	2.1	2.18	2.26	
	F VD tillesiloid 0	Falling edge	2	2.08	2.16	
V	PVD threshold 1	Rising edge	2.19	2.28	2.37	
V <sub>PVD1</sub>	F VD tillesiloid i	Falling edge	2.09	2.18	2.27	
V	PVD threshold 2	Rising edge	2.28	2.38	2.48	
V <sub>PVD2</sub>	PVD (illeshold 2	Falling edge	2.18	2.28	2.38	
V	DVD throubold 2	Rising edge	2.38	2.48	2.58	
V <sub>PVD3</sub>	PVD threshold 3	Falling edge	2.28	2.38	2.48	V
V	PVD threshold 4	Rising edge	2.47	2.58	2.69	V
$V_{PVD4}$		Falling edge	2.37	2.48	2.59	
V	D) (D) the reach and E	Rising edge	2.57	2.68	2.79	
V <sub>PVD5</sub>	PVD threshold 5	Falling edge	2.47	2.58	2.69	
V	PVD threshold 6	Rising edge	2.66	2.78	2.9	
V <sub>PVD6</sub>	PVD threshold 6	Falling edge	2.56	2.68	2.8	
V	D\/D throohold 7	Rising edge	2.76	2.88	3	
V <sub>PVD7</sub>	PVD threshold 7	Falling edge	2.66	2.78	2.9	
V <sub>PVDhyst</sub> <sup>(2)</sup>	PVD hysteresis	-	-	100	-	mV
IDD(PVD)	PVD current consumption	-	-	0.15	0.26	μΑ

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> Guaranteed by design.

### 6.3.4 Embedded reference voltage

The parameters given in *Table 28* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 24*.

**Conditions Symbol Parameter** Min Max Unit Typ -40 °C < T<sub>A</sub> < +105 °C 1.2 1.23 1.25 ٧  $V_{REFINT}$ Internal reference voltage 1.24<sup>(1)</sup> -40 °C < T<sub>A</sub> < +85 °C 1.2 1.23 V ADC sampling time when reading the internal 2.2 T<sub>S vrefint</sub> μs reference voltage Internal reference voltage spread over the  $V_{DD} = 3 V \pm 10 mV$  $10^{(2)}$ mV  $V_{RERINT}$ temperature range  $T_{Coeff}$ Temperature coefficient  $100^{(2)}$ ppm/°C

Table 28. Embedded internal reference voltage

<sup>2.</sup> Guaranteed by design.

Table 29. Internal refe	rence voltage calibration values
-------------------------	----------------------------------

Calibration value name	Description	Memory address
V <sub>REFINT_CAL</sub>	Raw data acquired at temperature of 30 °C V <sub>DDA</sub> = 3.3 V	0x1FFF F7BA - 0x1FFF F7BB

#### 6.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in *Figure 12: Current consumption measurement scheme*.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to CoreMark code.

#### Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in input mode with a static value at V<sub>DD</sub> or V<sub>SS</sub> (no load)
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted to the f<sub>HCLK</sub> frequency (0 wait state from 0 to 24 MHz,1 wait state from 24 to 48 MHz and 2 wait states from 48 to 72 MHz)
- Prefetch in ON (reminder: this bit must be set before clock setting and bus prescaling)
- When the peripherals are enabled f<sub>PCLK2</sub> = f<sub>HCLK</sub> and f<sub>PCLK1</sub> = f<sub>HCLK/2</sub>
- When f<sub>HCLK</sub> > 8 MHz, the PLL is ON and the PLL input is equal to HSI/2 (4 MHz) or HSE (8 MHz) in bypass mode.



<sup>1.</sup> Guaranteed by characterization results.

The parameters given in *Table 30* to *Table 34* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 24*.

Table 30. Typical and maximum current consumption from  $V_{DD}$  supply at  $V_{DD}$  = 3.6V

	,			All	periphe	erals en	abled	All	periphe	erals dis	abled				
Symbol	Parameter	Conditions	f <sub>HCLK</sub>	т	М	ах @ Т	A <sup>(1)</sup>	<b>T</b>	М	ах @ Т,	A <sup>(1)</sup>	Unit			
				Тур	25 °C	85 °C	105 °C	Тур	25 °C	85 °C	105 °C				
			72 MHz	61.2	65.8	67.6	68.5	27.8	30.3	30.7	31.5				
			64 MHz	54.7	59.1	60.2	61.1	24.6	27.2	27.6	28.3				
		External	48 MHz	41.7	45.1	46.2	47.2	19.2	21.1	21.4	21.8				
		clock (HSE	32 MHz	28.1	31.5	32.5	32.7	12.9	14.6	14.8	15.3				
	Supply	bypass)	24 MHz	21.4	23.7	24.4	25.2	10.0	11.4	11.4	12.1				
	current in Run mode, executing		8 MHz	7.4	8.4	8.6	9.4	3.6	4.1	4.4	5.0				
			1 MHz	1.3	1.6	1.8	2.6	0.8	1.0	1.2	2.1				
	from Flash		64 MHz	49.7	54.4	55.4	56.3	24.5	27.2	27.4	28.1				
		Internal clock (HSI)		48 MHz	37.9	42.2	43.0	43.5	18.9	21.4	21.5	21.6			
				32 MHz	25.8	29.2	29.2	30.0	12.7	14.2	14.6	15.2			
			24 MHz	19.7	22.3	22.6	23.2	6.7	7.7	7.9	8.5				
			8 MHz	6.9	7.8	8.3	8.8	3.5	4.0	4.4	5.0	mA			
I <sub>DD</sub>			72 MHz	60.8	66.2 <sup>(2)</sup>	69.7	70.4 <sup>(2)</sup>	27.4	31.7 <sup>(2)</sup>	32.2	32.5 <sup>(2)</sup>	IIIA			
			64 MHz	54.3	59.1	62.2	63.3	24.3	28.3	28.7	28.8				
		External	48 MHz	41.0	45.6	47.3	47.9	18.3	21.6	21.9	22.1				
		clock (HSE	32 MHz	27.6	32.4	32.4	32.9	12.3	15.0	15.2	15.4				
	Supply	bypass)	24 MHz	20.8	23.9	24.3	25.0	9.3	11.3	11.4	12.0				
	current in Run mode, executing from RAM		8 MHz	6.9	7.8	8.7	9.0	3.1	3.7	4.2	4.9				
			1 MHz	0.9	1.2	1.5	2.3	0.4	0.6	1.0	1.8				
			64 MHz	49.2	53.9	55.2	57.4	23.9	27.8	28.2	28.4				
			48 MHz	37.3	40.8	41.4	44.1	18.2	21.0	21.6	21.9				
		Internal clock (HSI)	32 MHz	25.1	27.6	29.1	30.1	12.0	14.0	14.5	15.1	1			
		CIOCK (HSI)	24 MHz	19.0	21.6	22.1	22.9	6.3	7.2	7.7	8.1				
						8 MHz	6.4	7.3	7.9	8.4	3.0	3.5	4.0	4.7	

Table 30. Typical and maximum current consumption from  $V_{DD}$  supply at  $V_{DD}$  = 3.6V (continued)

				All	periphe	erals en	abled	All	periphe	erals dis	abled					
Symbol Parameter	Parameter	Conditions	f <sub>HCLK</sub>	Тур	Max @ T <sub>A</sub> <sup>(1)</sup>			Tyrn	Max @ T <sub>A</sub> <sup>(1)</sup>			Unit				
			тур	25 °C	85 °C	105 °C	Тур	25 °C	85 °C	105 °C						
			72 MHz	44.0	48.4	49.4	50.5	6.6	7.5	7.9	8.7					
			64 MHz	39.2	43.3	44.0	45.2	6.0	6.8	7.2	7.9					
		External clock (HSE bypass)	48 MHz	29.6	32.7	33.3	34.3	4.5	5.2	5.6	6.3					
	Committee		32 MHz	19.7	23.3	23.3	23.5	3.1	3.5	4.0	4.8					
	Supply current in		24 MHz	14.9	17.6	17.8	18.3	2.4	2.8	3.3	3.9					
	Sleep mode,		8 MHz	4.9	5.7	6.1	6.9	0.8	1.0	1.4	2.2	mA				
I <sub>DD</sub>	executing		1 MHz	0.6	0.9	1.2	2.1	0.1	0.3	0.6	1.5	IIIA				
	from Flash or RAM		64 MHz	34.2	38.1	39.2	40.3	5.7	6.3	6.8	7.5					
	OI RAIVI		48 MHz	25.8	28.7	29.6	30.3	4.3	4.8	5.2	5.9					
		Internal clock (HSI)	32 MHz	17.4	19.4	19.9	20.7	2.9	3.2	3.7	4.5					
		3.001 (1101)	24 MHz	13.2	15.1	15.6	15.9	1.5	1.8	2.2	2.9					
								8 MHz	4.5	5.0	5.6	6.2	0.7	0.9	1.2	2.1

<sup>1.</sup> Guaranteed by characterization results unless otherwise specified.

Table 31. Typical and maximum current consumption from the  $V_{\text{DDA}}$  supply

					V <sub>DDA</sub>	= 2.4 V	,		V <sub>DDA</sub>	_ = 3.6 \	/														
Symbol	Parameter	Conditions (1)	f <sub>HCLK</sub>	Тур	Max @ T <sub>A</sub> <sup>(2)</sup>			Тур	Max @ T <sub>A</sub> <sup>(2)</sup>		Unit														
				тур	25 °C	85 °C	105 °C	iyp	25 °C	85 °C	105 °C														
			72 MHz	225	276	289	297	245	302	319	329														
			64 MHz	198	249	261	268	216	270	284	293														
			48 MHz	149	195	204	211	159	209	222	230														
	Supply	rrent in	32 MHz	102	145	152	157	110	154	162	169														
	current in Run/Sleep		24 MHz	80	119	124	128	86	126	131	135														
l	mode,		8 MHz	2	3	4	6	3	4	5	9	μΑ													
I <sub>DDA</sub>	code		1 MHz	2	3	5	7	3	4	6	9	μΛ													
	executing from Flash or RAM	executing from Flash	_	_	-	-	-	~	64 MHz	270	323	337	344	299	354	371	381								
			48 MHz	220	269	280	286	244	293	309	318														
		HSI clock	32 MHz	173	218	228	233	193	239	251	257	-													
			24 MHz	151	194	200	204	169	211	219	225														
																	8 MHz	73	97	99	103	88	105	110	116

Current consumption from the V<sub>DDA</sub> supply is independent of whether the peripherals are on or off. Furthermore when the PLL is off, I<sub>DDA</sub> is independent from the frequency.

<sup>2.</sup> Guaranteed by characterization results.



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<sup>2.</sup> Data based on characterization results and tested in production with code executing from RAM.

Table 32. Typical and maximum  $V_{\mbox{\scriptsize DD}}$  consumption in Stop and Standby modes

			Typ $@V_{DD}(V_{DD}=V_{DDA})$						Max <sup>(1)</sup>			
Symbol Parameter		Conditions	2.0 V	2.4 V	2.7 V	3.0 V	3.3 V	3.6 V	T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	Unit
	O	Regulator in run mode, all oscillators OFF	20.05	20.33	20.42	20.50	20.67	20.80	44.2 <sup>(2)</sup>	350	735 <sup>(2)</sup>	
	Stop mode	Regulator in low-power mode, all oscillators OFF	7.63	7.77	7.90	8.07	8.17	8.33	30.6 <sup>(2)</sup>	335	720 <sup>(2)</sup>	μA
	1- 1- 7	LSI ON and IWDG ON	0.80	0.96	1.09	1.23	1.37	1.51	-	-	-	
current in Standby mode		LSI OFF and IWDG OFF	0.60	0.74	0.83	0.93	1.02	1.11	5.0 <sup>(2)</sup>	7.8	13.3 <sup>(2)</sup>	

<sup>1.</sup> Guaranteed by characterization results unless otherwise specified.

Table 33. Typical and maximum  $V_{\text{DDA}}$  consumption in Stop and Standby modes

		er Conditions		DDA	Тур @	V <sub>DD</sub> (	V <sub>DD</sub> =	V <sub>DDA</sub> )			Max <sup>(1)</sup>						
Symbol	Parameter			2.0 V	2.4 V	2.7 V	3.0 V	3.3 V	3.6 V	T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	Unit				
	Supply	NO	Regulator in run mode, all oscillators OFF	1.81	1.95	2.07	2.20	2.35	2.52	3.7	5.5	8.8					
	current in Stop mode	gr.	Regulator in low-power mode, all oscillators OFF	1.81	1.95	2.07	2.20	2.35	2.52	3.7	5.5	8.8					
		current in Standby		Supply current in		LSI ON and IWDG ON	2.22	2.42	2.59	2.78	3.0	3.24	ı	ı	ı		
	Clariday			Standby >		otaliaby -		LSI OFF and IWDG OFF	1.69	1.82	1.94	2.08	2.23	2.40	3.5	5.4	9.2
I <sub>DDA</sub>	Supply	)FF	Regulator in run mode, all oscillators OFF	1.05	1.08	1.10	1.15	1.22	1.29	-	-	-	μΑ				
	current in Stop mode	=	Regulator in low-power mode, all oscillators OFF	1.05	1.08	1.10	1.15	1.22	1.29	-	-	-					
	Cupp.j   -		Cappiy		LSI ON and IWDG ON	1.44	1.52	1.60	1.71	1.84	1.98	į	-	-			
	current in Standby mode		current in Standby	current in Standby	Standby	V <sub>ро</sub> д	LSI OFF and IWDG OFF	0.93	0.95	0.98	1.02	1.08	1.15	-	- 1	-	

<sup>1.</sup> Guaranteed by characterization results.

The total consumption is the sum of IDD and IDDA.



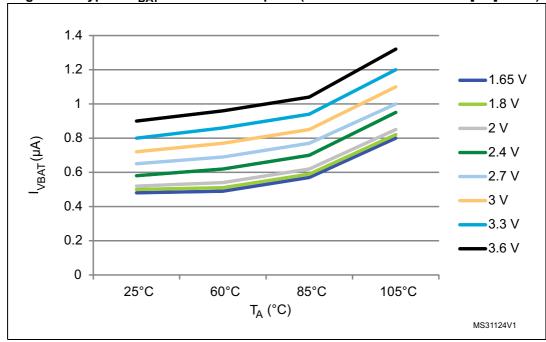
<sup>2.</sup> Data based on characterization results and tested in production.

Table 34. Typical and maximum current consumption from  $V_{\text{BAT}}$  supply

Symbol	Para	Conditions (1)	Typ @V <sub>BAT</sub>								Max @V <sub>BAT</sub> = 3.6 V <sup>(2)</sup>			Unit
	meter		1.65V	1.8V	2V	2.4V	2.7V	3V	3.3V	3.6V	T <sub>A</sub> = 25°C	T <sub>A</sub> = 85°C	T <sub>A</sub> = 105°C	Onic
	Backup domain	LSE & RTC ON; "Xtal mode" lower driving capability; LSEDRV[1: 0] = '00'	0.48	0.50	0.52	0.58	0.65	0.72	0.80	0.90	1.1	1.5	2.0	
I <sub>DD_VBAT</sub> supply	supply current	LSE & RTC ON; "Xtal mode" higher driving capability; LSEDRV[1: 0] = '11'	0.83	0.86	0.90	0.98	1.03	1.10	1.20	1.30	1.5	2.2	2.9	μА

- 1. Crystal used: Abracon ABS07-120-32.768 kHz-T with a CL of 6 pF for typical values.
- 2. Guaranteed by characterization results.

Figure 13. Typical  $V_{BAT}$  current consumption (LSE and RTC ON/LSEDRV[1:0] = '00')



#### **Typical current consumption**

The MCU is placed under the following conditions:

- V<sub>DD</sub> = V<sub>DDA</sub> = 3.3 V
- All I/O pins available on each package are in analog input configuration
- The Flash access time is adjusted to f<sub>HCLK</sub> frequency (0 wait states from 0 to 24 MHz, 1 wait state from 24 to 48 MHz and 2 wait states from 48 MHz to 72 MHz), and Flash prefetch is ON
- When the peripherals are enabled,  $f_{APB1} = f_{AHB/2}$ ,  $f_{APB2} = f_{AHB}$
- PLL is used for frequencies greater than 8 MHz
- AHB prescaler of 2, 4, 8,16 and 64 is used for the frequencies 4 MHz, 2 MHz, 1 MHz, 500 kHz and 125 kHz respectively.

Table 35. Typical current consumption in Run mode, code with data processing running from Flash

				Ту	/p	
Symbol	Parameter	Conditions	f <sub>HCLK</sub>	Peripherals enabled	Peripherals disabled	Unit
			72 MHz	61.3	28.0	
			64 MHz 54.8	25.4		
			48 MHz	41.9	19.3	
			32 MHz	28.5	13.3	
			24 MHz	21.8	10.4	
	Supply current in Run mode from		16 MHz	14.9	7.2	mA
$I_{DD}$	V <sub>DD</sub> supply		8 MHz	7.7	3.9	IIIA
	- 00		4 MHz	4.5	2.5	
		Running from HSE crystal clock 8 MHz, code executing from Flash	2 MHz	2.8	1.7	
			1 MHz	1.9	1.3	
			500 kHz	1.4	1.1	
			125 kHz	1.1	0.9	
			72 MHz	240.3	239.5	
			64 MHz	210.9	210.3	
			48 MHz	155.8	155.6	
			32 MHz	105.7	105.6	
			24 MHz	82.1	82.0	
(1)(2)	Supply current in		16 MHz	58.8	58.8	
I <sub>DDA</sub> <sup>(1) (2)</sup>	Run mode from V <sub>DDA</sub> supply		8 MHz	2.4	2.4	μA
	- DDA cabbil		4 MHz	2.4	2.4	
			2 MHz	2.4	2.4	
			1 MHz	2.4	2.4	
			500 kHz	2.4	2.4	
			125 kHz	2.4	2.4	

V<sub>DDA</sub> monitoring is ON.



When peripherals are enabled, the power consumption of the analog part of peripherals such as ADC, DAC, Comparators, OpAmp etc. is not included. Refer to the tables of characteristics in the subsequent sections.

Table 36. Typical current consumption in Sleep mode, code running from Flash or RAM

				Ty	ур			
Symbol	Parameter	Conditions	f <sub>HCLK</sub>	Peripherals enabled	Peripherals disabled	Unit		
			72 MHz	44.1	7.0			
			64 MHz	39.7	6.3			
			48 MHz	30.3	4.9			
			32 MHz	20.5	3.5			
			24 MHz	15.4	2.8			
	Supply current in		16 MHz	10.6	2.0	A		
I <sub>DD</sub>	Sleep mode from V <sub>DD</sub> supply		8 MHz	5.4	1.1	mA		
			4 MHz	3.2	1.0			
			2 MHz	2.1	0.9			
		Running from HSE	1 MHz	1.5	0.8			
			500 kHz	1.2	0.8			
		crystal clock 8 MHz,	125 kHz	1.0	0.8			
		code executing from		72 MHz	239.7	238.5		
		Flash or RAM	64 MHz	210.5	209.6			
					48 MHz	155.0	155.6	
			32 MHz	105.3	105.2			
			24 MHz	81.9	81.8			
I <sub>DDA</sub> <sup>(1) (2)</sup>	Supply current in		16 MHz	58.7	58.6			
IDDA'''	Sleep mode from V <sub>DDA</sub> supply	n	8 MHz	2.4	2.4	μA		
	DDA - FF 7		4 MHz	2.4	2.4			
			2 MHz	2.4	2.4			
			1 MHz	2.4	2.4			
		<u> </u>	500 kHz	2.4	2.4			
			125 kHz	2.4	2.4			

<sup>1.</sup>  $V_{DDA}$  monitoring is ON.

<sup>2.</sup> When peripherals are enabled, the power consumption of the analog part of peripherals such as ADC, DAC, Comparators, OpAmp etc. is not included. Refer to the tables of characteristics in the subsequent sections.

#### I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

#### I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 54: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

#### Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

#### I/O dynamic current consumption

In addition to the internal peripheral current consumption (see *Table 38: Peripheral current consumption*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the MCU supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DD} \times f_{SW} \times C$$

where

 $I_{SW}$  is the current sunk by a switching I/O to charge/discharge the capacitive load  $V_{DD}$  is the MCU supply voltage

f<sub>SW</sub> is the I/O switching frequency

C is the total capacitance seen by the I/O pin:  $C = C_{INT} + C_{EXT} + C_{S}$ 

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.



Table 37. Switching output I/O current consumption

Symbol	Parameter	Conditions <sup>(1)</sup>	I/O toggling frequency (f <sub>SW</sub> )	Тур	Unit
			2 MHz	0.90	
			4 MHz	0.93	
		$V_{DD} = 3.3 V$ $C_{ext} = 0 pF$	8 MHz	1.16	
		$C = C_{INT} + C_{EXT} + C_{S}$	18 MHz	1.60	
			36 MHz	2.51	
			48 MHz	2.97	
			2 MHz	0.93	
		V = 3 3 V	4 MHz	1.06	
		V <sub>DD</sub> = 3.3 V C <sub>ext</sub> = 10 pF	8 MHz	1.47	
		$C = C_{INT} + C_{EXT} + C_{S}$	18 MHz	2.26	
		36 MHz 3.39	3.39	l	
			48 MHz	5.99	
			2 MHz	1.03	
I <sub>SW</sub>	I/O current consumption	V <sub>DD</sub> = 3.3 V 4 MHz	1.30	mA	
	·	$C_{ext} = 22 pF$	8 MHz	1.79	
		$C = C_{INT} + C_{EXT} + C_{S}$ 18 MHz 3.	3.01		
			36 MHz	5.99	
			2 MHz	1.10	
		V <sub>DD</sub> = 3.3 V	4 MHz	1.31	
		C <sub>ext</sub> = 33 pF	8 MHz	2.06	
		$C = C_{INT} + C_{EXT} + C_{S}$	18 MHz	3.47	
			36 MHz	8.35	
			2 MHz	1.20	
		V <sub>DD</sub> = 3.3 V	4 MHz	1.54	
		C <sub>ext</sub> = 47 pF	8 MHz	2.46	
		$C = C_{INT} + C_{EXT} + C_{S}$	18 MHz	4.51	
			36 MHz	9.98	

<sup>1.</sup> CS = 5 pF (estimated value).



### On-chip peripheral current consumption

The MCU is placed under the following conditions:

- all I/O pins are in analog input configuration
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
  - with all peripherals clocked off
  - with only one peripheral clocked on
- ambient operating temperature at 25°C and V<sub>DD</sub> = V<sub>DDA</sub> = 3.3 V.

Table 38. Peripheral current consumption

Davinhaval	Typical consumption <sup>(1)</sup>	11-24
Peripheral	I <sub>DD</sub>	Unit
BusMatrix (2)	12.6	
DMA1	7.6	
DMA2	6.1	
CRC	2.1	
GPIOA	10.0	
GPIOB	10.3	
GPIOC	2.2	
GPIOD	8.8	
GPIOE	3.3	
GPIOF	3.0	
TSC	5.5	
ADC1&2	17.3	
ADC3&4	18.8	μΑ/MHz
APB2-Bridge (3)	3.6	μΑ/Ινίπ2
SYSCFG	7.3	
TIM1	40.0	
SPI1	8.8	
TIM8	36.4	
USART1	23.3	
TIM15	17.1	
TIM16	10.1	
TIM17	11.0	
APB1-Bridge (3)	6.1	
TIM2	49.1	
TIM3	38.8	
TIM4	38.3	



Table 38. Peripheral current consumption (continued)

Peripheral	Typical consumption <sup>(1)</sup>	Unit
Feripilerai	I <sub>DD</sub>	- Onit
TIM6	9.7	
TIM7	12.1	
WWDG	6.4	
SPI2	40.4	
SPI3	40.0	
USART2	41.9	
USART3	40.2	
UART4	36.5	μA/MHz
UART5	30.8	
I2C1	10.5	
12C2	10.4	
USB	26.2	
CAN	33.4	
PWR	5.7	
DAC	15.4	

The power consumption of the analog part (I<sub>DDA</sub>) of peripherals such as ADC, DAC, Comparators, OpAmp etc. is not included. Refer to the tables of characteristics in the subsequent sections.

<sup>2.</sup> BusMatrix is automatically active when at least one master is ON (CPU, DMA1 or DMA2).

<sup>3.</sup> The APBx bridge is automatically active when at least one peripheral is ON on the same bus.

# 6.3.6 Wakeup time from low-power mode

The wakeup times given in *Table 39* are measured starting from the wakeup event trigger up to the first instruction executed by the CPU:

- For Stop or Sleep mode: the wakeup event is WFE.
- WKUP1 (PA0) pin is used to wakeup from Standby, Stop and Sleep modes.

All timings are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 24*.

Table 39. Low-power mode wakeup timings

Symbol	Parameter	Conditions	Typ @VDD, V <sub>DD</sub> = V <sub>DDA</sub>						Max	Unit
Symbol	Farameter	Conditions	2.0 V	2.4 V	2.7 V	3 V	3.3 V	3.6 V	IVIAX	
twustop	Wakeup from	Regulator in run mode	4.1	3.9	3.8	3.7	3.6	3.5	4.5	
	Stop mode	Regulator in low-power mode	7.9	6.7	6.1	5.7	5.4	5.2	9	μs
t <sub>WUSTANDBY</sub> (1)	Wakeup from Standby mode	LSI and IWDG OFF	69.2	60.3	56.4	53.7	51.7	50	100	
t <sub>WUSLEEP</sub>	Wakeup from Sleep mode	-		6			-	CPU clock cycles		

<sup>1.</sup> Guaranteed by characterization results.

## 6.3.7 External clock source characteristics

# High-speed external user clock generated from an external source

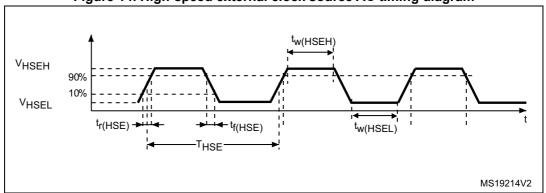
In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO. The external clock signal has to respect the I/O characteristics in *Section 6.3.14*. However, the recommended clock input waveform is shown in *Figure 14*.

Table 40. High-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSE_ext</sub>	User external clock source frequency <sup>(1)</sup>		1	8	32	MHz
V <sub>HSEH</sub>	OSC_IN input pin high level voltage		0.7V <sub>DD</sub>	-	$V_{DD}$	V
V <sub>HSEL</sub>	OSC_IN input pin low level voltage	-	$V_{SS}$	ı	0.3V <sub>DD</sub>	٧
t <sub>w(HSEH)</sub> t <sub>w(HSEL)</sub>	OSC_IN high or low time <sup>(1)</sup>		15	ı	-	ns
t <sub>r(HSE)</sub>	OSC_IN rise or fall time <sup>(1)</sup>		-	-	20	115

<sup>1.</sup> Guaranteed by design.

Figure 14. High-speed external clock source AC timing diagram



## Low-speed external user clock generated from an external source

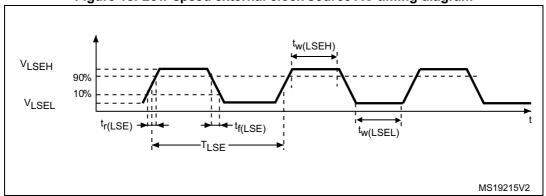
In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO. The external clock signal has to respect the I/O characteristics in *Section 6.3.14*. However, the recommended clock input waveform is shown in *Figure 15* 

Table 41. Low-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>LSE_ext</sub>	User External clock source frequency <sup>(1)</sup>		-	32.768	1000	kHz
V <sub>LSEH</sub>	OSC32_IN input pin high level voltage		0.7V <sub>DD</sub>	-	$V_{DD}$	>
V <sub>LSEL</sub>	OSC32_IN input pin low level voltage	-	V <sub>SS</sub>	ı	0.3V <sub>DD</sub>	V
t <sub>w(LSEH)</sub>	OSC32_IN high or low time <sup>(1)</sup>		450	ı	ı	ns
t <sub>r(LSE)</sub>	OSC32_IN rise or fall time <sup>(1)</sup>		-	-	50	110

<sup>1.</sup> Guaranteed by design.

Figure 15. Low-speed external clock source AC timing diagram



## High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 32 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 42*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

	140.0 1211102 000114101 011414010110110								
Symbol	Parameter	Conditions <sup>(1)</sup>	Min <sup>(2)</sup>	Тур	Max <sup>(2)</sup>	Unit			
f <sub>OSC_IN</sub>	Oscillator frequency	-	4	8	32	MHz			
$R_{F}$	Feedback resistor	-	-	200		kΩ			
		During startup <sup>(3)</sup>	-	-	8.5				
	HSE current consumption	V <sub>DD</sub> =3.3 V, Rm= 30Ω CL=10 pF@8 MHz	-	0.4	-				
		V <sub>DD</sub> =3.3 V, Rm= 45Ω CL=10 pF@8 MHz	-	0.5	-				
I <sub>DD</sub>		V <sub>DD</sub> =3.3 V, Rm= 30Ω CL=5 pF@32 MHz	-	0.8	-	mA			
		V <sub>DD</sub> =3.3 V, Rm= 30Ω CL=10 pF@32 MHz	-	1	-				
		V <sub>DD</sub> =3.3 V, Rm= 30Ω, CL=20 pF@32 MHz	-	1.5	-				
9 <sub>m</sub>	Oscillator transconductance	Startup	10	-	-	mA/V			
t <sub>SU(HSE)</sub> <sup>(4)</sup>	Startup time	V <sub>DD</sub> is stabilized	-	2	-	ms			

Table 42. HSE oscillator characteristics



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<sup>1.</sup> Resonator characteristics given by the crystal/ceramic resonator manufacturer.

<sup>2.</sup> Guaranteed by design.

<sup>3.</sup> This consumption level occurs during the first 2/3 of the  $t_{\mbox{\scriptsize SU(HSE)}}$  startup time.

t<sub>SU(HSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 16*).  $C_{L1}$  and  $C_{L2}$  are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing  $C_{L1}$  and  $C_{L2}$ .

Note:

For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

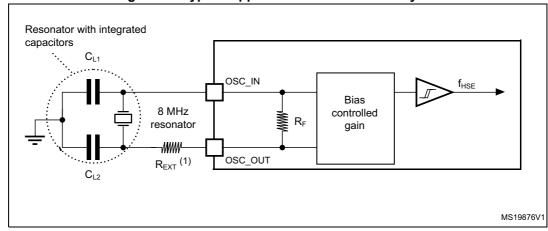


Figure 16. Typical application with an 8 MHz crystal

1.  $R_{\text{EXT}}$  value depends on the crystal characteristics.

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## Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 43*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 43	LSF	oscillator	characteristics	$(f_{1SE} = 32.768 \text{ kHz})$
I abic TJ.	ᆫᇰᆫ	OSCIIIALOI	Cital actel istics	III GE - JEII OU KIIEI

Symbol	Parameter	Conditions <sup>(1)</sup>	Min <sup>(2)</sup>	Тур	Max <sup>(2)</sup>	Unit
I <sub>DD</sub>		LSEDRV[1:0]=00 lower driving capability	-	0.5	0.9	
	LSE current consumption	LSEDRV[1:0]=10 medium low driving capability	-	-	1	^
		LSEDRV[1:0]=01 medium high driving capability	-	-	1.3	μА
		LSEDRV[1:0]=11 higher driving capability	-	-	1.6	
	Oscillator transconductance	LSEDRV[1:0]=00 lower driving capability	5	-	-	
g .		LSEDRV[1:0]=10 medium low driving capability	8	-	-	۸ /\ /
9 <sub>m</sub>		LSEDRV[1:0]=01 medium high driving capability	15	-	-	μΑ/V
		LSEDRV[1:0]=11 higher driving capability	25	-	-	
t <sub>SU(LSE)</sub> (3)	Startup time	V <sub>DD</sub> is stabilized	-	2	-	S

Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.



<sup>2.</sup> Guaranteed by design.

<sup>3.</sup> t<sub>SU(LSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer.

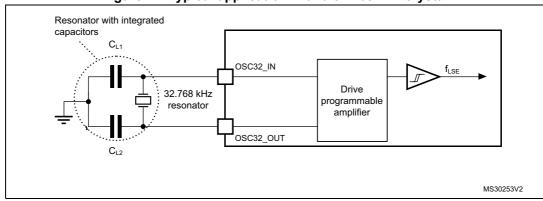


Figure 17. Typical application with a 32.768 kHz crystal

Note:

An external resistor is not required between OSC32\_IN and OSC32\_OUT and it is forbidden to add one.

### 6.3.8 Internal clock source characteristics

The parameters given in *Table 44* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 24*.

### High-speed internal (HSI) RC oscillator

**Conditions Symbol Parameter** Min Тур Unit Max 8 MHz Frequency  $f_{HSI}$ 1(2) **TRIM** HSI user trimming step % 45<sup>(2)</sup> 55<sup>(2)</sup> % Duty cycle DuCy(HSI)  $T_A = -40 \text{ to}$ -2.8<sup>(3)</sup>  $3.8^{(3)}$ 105°C -1.9<sup>(3)</sup>  $2.3^{(3)}$  $T_A = -10 \text{ to } 85^{\circ}\text{C}$  $2^{(3)}$ -1.9<sup>(3)</sup>  $T_A = 0$  to  $85^{\circ}C$ **ACC<sub>HSI</sub>** Accuracy of the HSI oscillator %  $2^{(3)}$ -1.3<sup>(3)</sup>  $T_A = 0 \text{ to } 70^{\circ}\text{C}$ -1<sup>(3)</sup> 2<sup>(3)</sup>  $T_{\Delta} = 0$  to  $55^{\circ}$ C  $T_A = 25^{\circ}C^{(4)}$ -1 1(2) 2(2) HSI oscillator startup time μs t<sub>su(HSI)</sub> HSI oscillator power 100<sup>(2)</sup> 80 μΑ I<sub>DDA(HSI)</sub> consumption

Table 44. HSI oscillator characteristics<sup>(1)</sup>

- 1.  $V_{DDA}$  = 3.3 V,  $T_A$  = -40 to 105 °C unless otherwise specified.
- 2. Guaranteed by design.
- 3. Guaranteed by characterization results.
- 4. Factory calibrated, parts not soldered.

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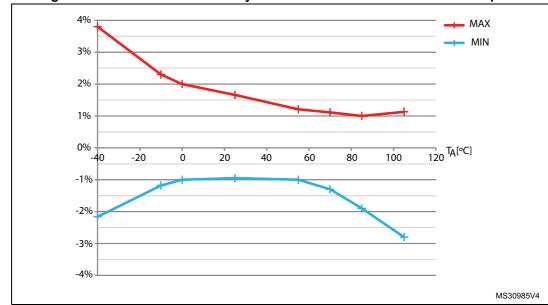


Figure 18. HSI oscillator accuracy characterization results for soldered parts

## Low-speed internal (LSI) RC oscillator

Table 45. LSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Тур	Max	Unit
f <sub>LSI</sub>	Frequency	30	40	50	kHz
t <sub>su(LSI)</sub> <sup>(2)</sup>	LSI oscillator startup time	-	-	85	μs
I <sub>DD(LSI)</sub> <sup>(2)</sup>	LSI oscillator power consumption	-	0.75	1.2	μA

<sup>1.</sup>  $V_{DDA}$  = 3.3 V,  $T_A$  = -40 to 105 °C unless otherwise specified.

<sup>2.</sup> Guaranteed by design.

### 6.3.9 PLL characteristics

The parameters given in *Table 46* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 24*.

Value **Symbol Parameter** Unit Min Тур Max 1<sup>(2)</sup> 24<sup>(2)</sup> PLL input clock<sup>(1)</sup> MHz f<sub>PLL\_IN</sub> 40<sup>(2)</sup>  $60^{(2)}$ PLL input clock duty cycle %  $16^{(2)}$  $f_{PLL\_OUT}$ PLL multiplier output clock 72 MHz 200<sup>(2)</sup> PLL lock time t<sub>LOCK</sub> μs Jitter  $300^{(2)}$ Cycle-to-cycle jitter ps

**Table 46. PLL characteristics** 

# 6.3.10 Memory characteristics

### Flash memory

The characteristics are given at  $T_A$  = -40 to 105 °C unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
t <sub>prog</sub>	16-bit programming time	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	40	53.5	60	μs
t <sub>ERASE</sub>	Page (2 KB) erase time	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	20	-	40	ms
t <sub>ME</sub>	Mass erase time	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	20	-	40	ms
	I <sub>DD</sub> Supply current	Write mode	-	-	10	mA
IDD		Erase mode	-	-	12	mA

Table 47. Flash memory characteristics

Table 48. Flash memory endurance and data retention

Cumbal	Doromotor	Conditions	Value	Unit	
Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Oilit	
N <sub>END</sub>	Endurance	$T_A = -40$ to +85 °C (6 suffix versions) $T_A = -40$ to +105 °C (7 suffix versions)	10	kcycles	
	- Data retention	1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 85 °C	30		
t <sub>RET</sub>		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 105 °C	10	Years	
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 55 °C	20		

<sup>1.</sup> Guaranteed by characterization results.

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Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by f<sub>PLL OUT</sub>.

<sup>2.</sup> Guaranteed by design.

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> Cycling performed over the whole temperature range.

#### 6.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

## Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 49*. They are based on the EMS levels and classes defined in the application note AN1709.

Level/ **Symbol Parameter Conditions** Class  $V_{DD} = 3.3 \text{ V, LQFP100, T}_{A} = +25^{\circ}\text{C,}$ Voltage limits to be applied on any I/O pin to  $V_{\text{FESD}}$ f<sub>HCLK</sub> = 72 MHz 3B induce a functional disturbance conforms to IEC 61000-4-2  $V_{DD} = 3.3 \text{ V, LQFP100, T}_{A} = +25^{\circ}\text{C,}$ Fast transient voltage burst limits to be f<sub>HCLK</sub> = 72 MHz  $\mathsf{V}_{\mathsf{EFTB}}$ applied through 100 pF on V<sub>DD</sub> and V<sub>SS</sub> 4A pins to induce a functional disturbance conforms to IEC 61000-4-4

Table 49. EMS characteristics

### Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

## Software recommendations

The software flowchart must include the management of runaway conditions such as:

- · Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)



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#### **Prequalification trials**

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

## **Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Symbol	Parameter	Conditions	Monitored	Max vs. [f <sub>HSE</sub> /f <sub>HCLK</sub> ]	Unit
Oymboi	i didilictor	Conditions	frequency band	8/72 MHz	
	Peak level	Peak level   V <sub>DD</sub> = 3.6 V, T <sub>A</sub> = 25 °C, LQFP100 package compliant with IEC 61967-2	0.1 to 30 MHz	7	
			30 to 130 MHz	20	dΒμV
S <sub>EMI</sub>			130 MHz to 1GHz	27	
			SAE EMI Level	4	-

Table 50. EMI characteristics

## 6.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### **Electrostatic discharge (ESD)**

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the JESD22-A114, ANSI/ESD STM5.3.1 standard.

Symbol	Ratings	Conditions	Packages	Class	Maximum value <sup>(1)</sup>	Unit				
V <sub>ESD(HBM)</sub>		T <sub>A</sub> = +25 °C, conforming to ANSI/ESDA/JEDEC JS-001	All	2	2000	V				
V <sub>ESD(CDM)</sub>		T <sub>A</sub> = +25 °C, conforming to ANSI/ESDA/JEDEC JS-002	All	C2a	500	V				

Table 51. ESD absolute maximum ratings

1. Guaranteed by characterization results.



## Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 52. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T <sub>A</sub> = +105 °C conforming to JESD78A	II level A

## 6.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DD}$  (for standard, 3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

## Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of  $-5 \,\mu\text{A}/+0 \,\mu\text{A}$  range), or other functional failure (for example reset occurrence or oscillator frequency deviation).

The test results are given in Table 53.

Table 53. I/O current injection susceptibility

		Functional s	usceptibility	
Symbol	Description	Negative injection	Positive injection	Unit
	Injected current on BOOT0	- 0	NA	
	Injected current on PC0, PC1, PC2, PC3, PF2, PA0, PA1, PA2, PA3, PF4, PA4, PA5, PA6, PA7, PC4, PC5, PB2 with induced leakage current on other pins from this group less than -50 $\mu$ A	- 5	-	
I <sub>INJ</sub>	Injected current on PB0, PB1, PE7, PE8, PE9, PE10, PE11, PE12, PE13, PE14, PE15, PB12, PB13, PB14, PB15, PD8, PD9, PD10, PD11, PD12, PD13, PD14 with induced leakage current on other pins from this group less than -50 µA	- 5	-	mA
	Injected current on PC0, PC1, PC2, PC3, PF2, PA0, PA1, PA2, PA3, PF4, PA4, PA5, PA6, PA7, PC4, PC5, PB2, PB0, PB1, PE7, PE8, PE9, PE10, PE11, PE12, PE13, PE14, PE15, PB12, PB13, PB14, PB15, PD8, PD9, PD10, PD11, PD12, PD13, PD14 with induced leakage current on other pins from this group less than $400~\mu\text{A}$	-	+5	
	Injected current on any other FT and FTf pins	<b>-</b> 5	NA	
	Injected current on any other pins	<b>-</b> 5	+5	

Note: It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.



# 6.3.14 I/O port characteristics

# General input/output characteristics

Unless otherwise specified, the parameters given in *Table 54* are derived from tests performed under the conditions summarized in *Table 24*. All I/Os are CMOS and TTL compliant.

Table 54. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		TC and TTa I/O	-	-	0.3 V <sub>DD</sub> +0.07 <sup>(1)</sup>	
	Low level input voltage	FT and FTf I/O	-	-	0.475 V <sub>DD</sub> -0.2 <sup>(1)</sup>	
VIL		BOOT0	-	-	0.3 V <sub>DD</sub> -0.3 <sup>(1)</sup>	
V <sub>IL</sub> V <sub>IH</sub>		All I/Os except BOOT0	-	-	0.3 V <sub>DD</sub> <sup>(2)</sup>	V
		TC and TTa I/O	0.445 V <sub>DD</sub> +0.398 <sup>(1)</sup>	-	-	V
	High level input	FT and FTf I/O	0.5 V <sub>DD</sub> +0.2 <sup>(1)</sup>	-	-	
V <sub>IH</sub>	voltage	воото	0.2 V <sub>DD</sub> +0.95 <sup>(1)</sup>	-	-	
		All I/Os except BOOT0	0.7 V <sub>DD</sub> <sup>(2)</sup>	-	-	
	Schmitt trigger hysteresis	TC and TTa I/O	-	200 (1)	-	
$V_{hys}$		FT and FTf I/O	-	100 (1)	-	mV
		воото	-	300 (1)	-	
		TC, FT and FTf I/O TTa I/O in digital mode $V_{SS} \le V_{IN} \le V_{DD}$	-	-	±0.1	
I <sub>lkg</sub>	Input leakage current <sup>(3)</sup>	TTa I/O in digital mode V <sub>DD</sub> ≤V <sub>IN</sub> ≤V <sub>DDA</sub>	-	-	1	μA
3	Current	TTa I/O in analog mode V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DDA</sub>	-	-	±0.2	
		FT and FTf I/O <sup>(4)</sup> V <sub>DD</sub> ≤V <sub>IN</sub> ≤5 V	-	-	10	
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(5)</sup>	$V_{IN} = V_{SS}$	25	40	55	kΩ
R <sub>PD</sub>	Weak pull-down equivalent resistor <sup>(5)</sup>	$V_{IN} = V_{DD}$	25	40	55	kΩ
C <sub>IO</sub>	I/O pin capacitance	-	-	5	-	pF

<sup>1.</sup> Data based on design simulation.

Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimum (~10% order).



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<sup>2.</sup> Tested in production.

<sup>3.</sup> Leakage could be higher than the maximum value. if negative current is injected on adjacent pins. Refer to *Table 53: I/O current injection susceptibility*.

<sup>4.</sup> To sustain a voltage higher than  $V_{DD}$  +0.3 V, the internal pull-up/pull-down resistors must be disabled.

All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in *Figure 19* and *Figure 20* for standard I/Os.

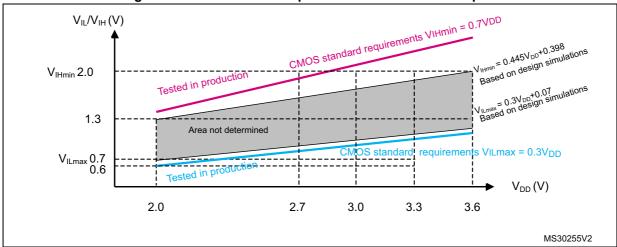
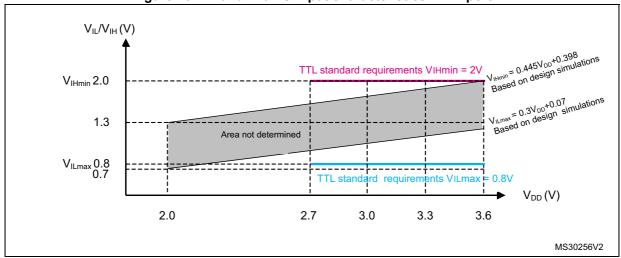


Figure 19. TC and TTa I/O input characteristics - CMOS port





V<sub>IL</sub>/V<sub>IH</sub> (V)

2.0

CMOS standard requirements VIHmin = 0.7VDD

V<sub>IHmin</sub> = 0.5Voo+0.2
Based on design simulations
Based on design simulations
Based on design simulations
CMOS standard requirements VILmax = 0.3VDD

V<sub>DD</sub> (V)

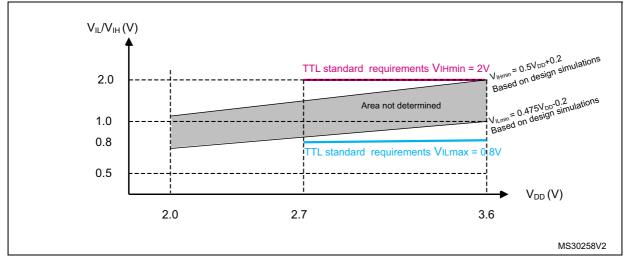
2.0

2.7

3.6

Figure 21. Five volt tolerant (FT and FTf) I/O input characteristics - CMOS port





## **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to +/-8 mA, and sink or source up to  $\pm$ -20 mA (with a relaxed  $V_{OL}/V_{OH}$ ).

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in Section 6.2:

- The sum of the currents sourced by all the I/Os on  $V_{DD,}$  plus the maximum Run consumption of the MCU sourced on V<sub>DD</sub> cannot exceed the absolute maximum rating ΣI<sub>VDD</sub> (see *Table 22*).
- The sum of the currents sunk by all the I/Os on  $V_{SS}$  plus the maximum Run consumption of the MCU sunk on V<sub>SS</sub> cannot exceed the absolute maximum rating  $\Sigma I_{VSS}$  (see *Table 22*).

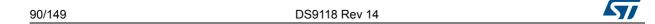
## **Output voltage levels**

Unless otherwise specified, the parameters given in *Table 55* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in Table 24. All I/Os (FT, TTa and TC unless otherwise specified) are CMOS and TTL compliant.

. ab. o os. o a.pat. voltago o nai acionotico							
Symbol	Parameter	Conditions	Min	Max	Unit		
V <sub>OL</sub> <sup>(1)</sup>	Output low level voltage for an I/O pin	CMOS port <sup>(2)</sup>	-	0.4			
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	I <sub>IO</sub> = +8 mA 2.7 V < V <sub>DD</sub> < 3.6 V	V <sub>DD</sub> -0.4	-			
V <sub>OL</sub> <sup>(1)</sup>	Output low level voltage for an I/O pin	TTL port <sup>(2)</sup>	-	0.4			
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	I <sub>IO</sub> = +8 mA 2.7 V < V <sub>DD</sub> < 3.6 V	2.4	-			
V <sub>OL</sub> <sup>(1)(4)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub> = +20 mA	-	1.3	V		
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin	2.7 V < V <sub>DD</sub> < 3.6 V	V <sub>DD</sub> -1.3	-			
V <sub>OL</sub> <sup>(1)(4)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub> = +6 mA	-	0.4			
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin	2 V < V <sub>DD</sub> < 2.7 V	V <sub>DD</sub> -0.4	-			
V <sub>OLFM+</sub> <sup>(1)(4)</sup>	Output low level voltage for an FTf I/O pin in FM+ mode	I <sub>IO</sub> = +20 mA 2.7 V < V <sub>DD</sub> < 3.6 V	-	0.4			

Table 55. Output voltage characteristics

4. Data based on design simulation.



The  $I_{|O}$  current sunk by the device must always respect the absolute maximum rating specified in *Table 22* and the sum of  $I_{|O|}$  (I/O ports and control pins) must not exceed  $\Sigma I_{|O(P|N)}$ .

<sup>2.</sup> TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

The I<sub>IO</sub> current sourced by the device must always respect the absolute maximum rating specified in Table 22 and the sum of I<sub>IO</sub> (I/O ports and control pins) must not exceed ΣI<sub>IO(PIN)</sub>.

## Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 23* and *Table 56*, respectively.

Unless otherwise specified, the parameters given are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 24*.

Table 56. I/O AC characteristics<sup>(1)</sup>

OSPEEDRy [1:0] value <sup>(1)</sup>	Symbol	Parameter	Conditions	Min	Max	Unit
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(2)</sup>	$C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 3.6 \text{ V}$	-	2 <sup>(3)</sup>	MHz
x0	t <sub>f(IO)out</sub>	Output high to low level fall time	-C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2 V to 3.6 V	ı	125 <sup>(3)</sup>	ns
	t <sub>r(IO)out</sub>	Output low to high level rise time	-CL = 30 μr, ν <sub>DD</sub> = 2 ν to 3.0 ν	1	125 <sup>(3)</sup>	115
	$f_{\text{max(IO)out}}$ Maximum frequency <sup>(2)</sup> $C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 3.6 \text{ V}$		-	10 <sup>(3)</sup>	MHz	
01	t <sub>f(IO)out</sub>	Output high to low level fall time	C = 50 pE // = 2 // to 2 6 //	-	25 <sup>(3)</sup>	20
	t <sub>r(IO)out</sub>	Output low to high level rise time	-C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2 V to 3.6 V	-	25 <sup>(3)</sup>	ns
			$C_L = 30 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	50 <sup>(3)</sup>	MHz
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(2)</sup>	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V	-	30 <sup>(3)</sup>	MHz
			C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2 V to 2.7 V	-	20 <sup>(3)</sup>	MHz
			$C_L = 30 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	5 <sup>(3)</sup>	
11	t <sub>f(IO)out</sub>	Output high to low level fall time	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	8 <sup>(3)</sup>	
			$C_L = 50 \text{ pF}, V_{DD} = 2 \text{ V to } 2.7 \text{ V}$	-	12 <sup>(3)</sup>	ne
			$C_L = 30 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	5 <sup>(3)</sup>	ns
	t <sub>r(IO)out</sub>	Output low to high level rise time	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	8 <sup>(3)</sup>	
			C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2 V to 2.7 V	-	12 <sup>(3)</sup>	
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(2)</sup>		-	2 <sup>(4)</sup>	MHz
FM+ configuration <sup>(4)</sup>	t <sub>f(IO)out</sub>	Output high to low level fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2 V to 3.6 V	-	12 <sup>(4)</sup>	20
22922	t <sub>r(IO)out</sub>	Output low to high level rise time		-	34 <sup>(4)</sup>	ns
-	t <sub>EXTIpw</sub>	Pulse width of external signals detected by the EXTI controller	-	10 <sup>(3)</sup>	-	ns

The I/O speed is configured using the OSPEEDRx[1:0] bits. Refer to the RM0316 reference manual for a description of GPIO Port configuration register.

<sup>4.</sup> The I/O speed configuration is bypassed in FM+ I/O mode. Refer to the STM32F303x STM32F313xx reference manual RM0316 for a description of FM+ I/O mode configuration.



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<sup>2.</sup> The maximum frequency is defined in *Figure 23*.

<sup>3.</sup> Guaranteed by design.

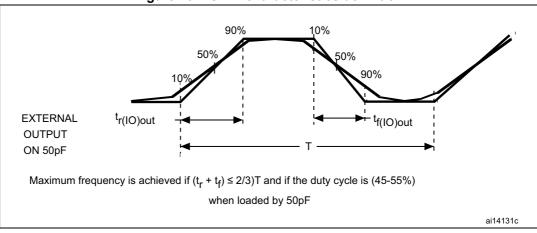


Figure 23. I/O AC characteristics definition

# 6.3.15 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R<sub>PU</sub> (see *Table 54*).

Unless otherwise specified, the parameters given in *Table 57* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 24*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL(NRST)</sub> <sup>(1)</sup>	NRST Input low level voltage	-	-	-	0.3V <sub>DD</sub> + 0.07 <sup>(1)</sup>	V
V <sub>IH(NRST)</sub> <sup>(1)</sup>	NRST Input high level voltage	-	0.445V <sub>DD</sub> + 0.398 <sup>(1)</sup>	-	-	V
V <sub>hys(NRST)</sub>	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(2)</sup>	$V_{IN} = V_{SS}$	25	40	55	kΩ
V <sub>F(NRST)</sub> <sup>(1)</sup>	NRST Input filtered pulse	-	-	-	100 <sup>(1)</sup>	ns
V <sub>NF(NRST)</sub> <sup>(1)</sup>	NRST Input not filtered pulse	-	500 <sup>(1)</sup>	-	-	ns

Table 57. NRST pin characteristics

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance must be minimum (~10% order).

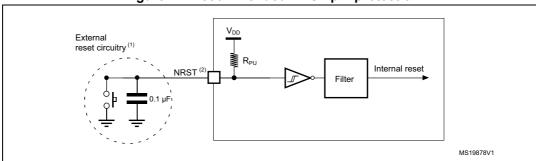


Figure 24. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- The user must ensure that the level on the NRST pin can go below the V<sub>IL(NRST)</sub> max level specified in Table 57. Otherwise the reset will not be taken into account by the device.

### 6.3.16 Timer characteristics

The parameters given in *Table 58* are guaranteed by design.

Refer to Section 6.3.14: I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Symbol	Parameter	Conditions	Min	Max	Unit
		-	1	-	t <sub>TIMxCLK</sub>
t <sub>res(TIM)</sub>	Timer resolution time	f <sub>TIMxCLK</sub> = 72 MHz	13.9	-	ns
res(Tiw)		f <sub>TIMxCLK</sub> = 144 MHz x=1.8	6.95	-	ns
f <sub>EXT</sub>	Timer external clock	-	0	f <sub>TIMxCLK</sub> /2	MHz
EXI	frequency on CH1 to CH4	f <sub>TIMxCLK</sub> = 72 MHz	0	36	MHz
Res <sub>TIM</sub>	Timer resolution	TIMx (except TIM2)	-	16	bit
I TCS/IM	Timer resolution	TIM2	-	32	Dit
	16-bit counter clock period	-	1	65536	t <sub>TIMxCLK</sub>
t <sub>COUNTER</sub>		f <sub>TIMxCLK</sub> = 72 MHz	0.0139	910	μs
COOMILIA		f <sub>TIMxCLK</sub> = 144 MHz x=1.8	0.0069	455	μs
		-	-	65536 × 65536	t <sub>TIMxCLK</sub>
t <sub>MAX_COUNT</sub>	Maximum possible count	f <sub>TIMxCLK</sub> = 72 MHz	-	59.65	S
	with 32-bit counter	f <sub>TIMxCLK</sub> = 144 MHz x=1.8	-	29.825	s

Table 58. TIMx<sup>(1)(2)</sup> characteristics

2. Guaranteed by design.

TIMx is used as a general term to refer to the TIM1, TIM2, TIM3, TIM4, TIM8, TIM15, TIM16 and TIM17 timers.

Table 59. IWDG min/max timeout period at 40 kHz (LSI) (1)

			· ·		
Prescaler divider	PR[2:0] bits	Min timeout (ms) RL[11:0]= 0x000	Max timeout (ms) RL[11:0]= 0xFFF		
/4	0	0.1	409.6		
/8	1	0.2	819.2		
/16	2	0.4	1638.4		
/32	3	0.8	3276.8		
/64	4	1.6	6553.6		
/128	5	3.2	13107.2		
/256	7	6.4	26214.4		

These timings are given for a 40 kHz clock but the microcontroller's internal RC frequency can vary from 30 to 60 kHz. Moreover, given an exact RC oscillator frequency, the exact timings still depend on the phasing of the APB interface clock versus the LSI clock so that there is always a full RC period of uncertainty.

Table 60. WWDG min-max timeout value @72 MHz (PCLK)<sup>(1)</sup>

Prescaler WDGTB  1 0		Min timeout value	Max timeout value
		0.05687	3.6409
2	1	0.1137	7.2817
4	2	0.2275	14.564
8	3	0.4551	29.127

<sup>1.</sup> Guaranteed by design.

# 6.3.17 Communications interfaces

## I<sup>2</sup>C interface characteristics

The I<sup>2</sup>C interface meets the timings requirements of the I<sup>2</sup>C-bus specification and user manual rev.03 for:

- Standard-mode (Sm): with a bit rate up to 100 Kbits/s
- Fast-mode (Fm): with a bit rate up to 400 Kbits/s
- Fast-mode Plus (Fm+): with a bit rate up to 1Mbits/s

The I<sup>2</sup>C timings requirements are guaranteed by design when the I<sup>2</sup>C peripheral is properly configured (refer to Reference manual).

The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and VDDIOx is disabled, but is still present. Only FTf I/O pins support Fm+ low level output current maximum requirement. Refer to Section 6.3.14: I/O port characteristics.

All I<sup>2</sup>C I/Os embed an analog filter. refer to the *Table 62: I2C analog filter characteristics*.

Table 61. I2C timings specification (see I2C specification, rev.03, June 2007)<sup>(1)</sup>

Symbol	Parameter	Standa	rd mode	Fast m	ode	Fast Mo	de Plus	Unit
Symbol	Parameter	Min	Max	Min	Max	Min	Max	Onit
f <sub>SCL</sub>	SCL clock frequency	0	100	0	400	0	1000	KHz
t <sub>LOW</sub>	Low period of the SCL clock	4.7	-	1.3	-	0.5	-	μs
t <sub>HIGH</sub>	High Period of the SCL clock	4		0.6		0.26	-	μs
t <sub>r</sub>	Rise time of both SDA and SCL signals	-	1000	-	300	-	120	ns
t <sub>f</sub>	Fall time of both SDA and SCL signals	-	300	-	300	-	120	ns
t <sub>HD;DAT</sub>	Data hold time	0	-	0	-	0	-	μs
t <sub>VD;DAT</sub>	Data valid time	-	3.45 <sup>(2)</sup>	-	0.9 <sup>(2)</sup>	-	0.45 <sup>(2)</sup>	μs
t <sub>VD;ACK</sub>	Data valid acknowledge time	-	3.45 <sup>(2)</sup>	-	0.9 <sup>(2)</sup>	-	0.45 <sup>(2)</sup>	μs
t <sub>SU;DAT</sub>	Data setup time	250	-	100	-	50	-	ns
t <sub>HD:STA</sub>	Hold time (repeated) START condition	4.0	-	0.6	-	0.26	-	μs
t <sub>SU:STA</sub>	Set-up time for a repeated START condition	4.7	-	0.6	-	0.26		μs
t <sub>SU:STO</sub>	Set-up time for STOP condition	4.0	-	0.6	-	0.26	-	μs
t <sub>BUF</sub>	Bus free time between a STOP and START condition	4.7	-	1.3	-	0.5	-	μs
C <sub>b</sub>	Capacitive load for each bus line	-	400	-	400	-	550	pF
t <sub>SP</sub>	Pulse width of spikes that are suppressed by the analog filter for Standard and Fast mode	0	50 <sup>(3)</sup>	0	50 <sup>(3)</sup>	-	-	ns



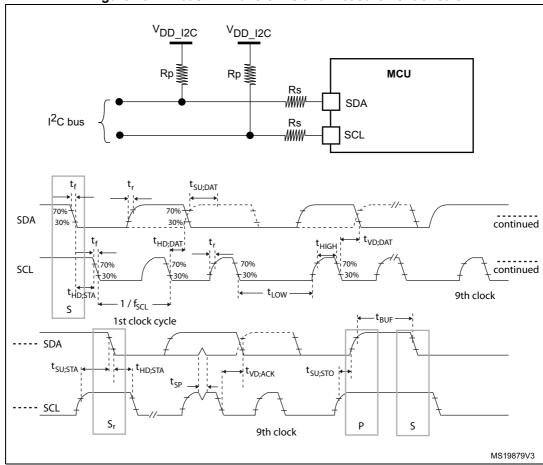
- The I2C characteristics are the requirements from I2C bus specification rev03. They are guaranteed by design when I2Cx\_TIMING register is correctly programmed (Refer to the RM0316 reference manual).
- The maximum tHD;DAT could be 3.45 µs, 0.9 µs and 0.45 µs for standard mode, fast mode and fast mode plus, but must be less than the maximum of tVD;DAT or tVD;ACK by a transition time.
- 3. The minimum width of the spikes filtered by the analog filter is above  $t_{SP}(max)$ .

Table 62. I2C analog filter characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>AF</sub>	Pulse width of spikes that are suppressed by the analog filter	50	260	ns

1. Guaranteed by design.

Figure 25. I<sup>2</sup>C bus AC waveforms and measurement circuit



1. Rs: Series protection resistors, Rp: Pull-up resistors, VDD\_I2C: I2C bus supply.



# SPI/I<sup>2</sup>S characteristics

Unless otherwise specified, the parameters given in *Table 63* for SPI or in *Table 64* for  $I^2S$  are derived from tests performed under ambient temperature,  $f_{PCLKX}$  frequency and  $V_{DD}$  supply voltage conditions summarized in *Table 24*.

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI and WS, CK, SD for I<sup>2</sup>S).

Table 63. SPI characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit			
		Master mode, SPI1 2.7 <v<sub>DD&lt;3.6</v<sub>			24				
f <sub>SCK</sub>	ODI ala di francisco	Slave mode, SPI1 2.7 <v<sub>DD&lt;3.6</v<sub>						24	NAL 1-
1/t <sub>c(SCK)</sub>	SPI clock frequency	Master mode, SPI1/2/3 2 <v<sub>DD&lt;3.6</v<sub>	-	-	18	MHz			
		Slave mode, SPI1/2/3 2 <v<sub>DD&lt;3.6</v<sub>			18				
DuCy(sck)	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%			
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI presc = 2	4*Tpclk	-	-				
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI presc = 2	2*Tpclk	-	-				
t <sub>w(SCKH)</sub>	SCK high and low time	Master mode	Tpclk-2	Tpclk	Tpclk+2				
t <sub>su(MI)</sub>	Data input actus time	Master mode	5.5	-	-				
t <sub>su(SI)</sub>	Data input setup time	Slave mode	6.5	-	-				
t <sub>h(MI)</sub>	Data input hold time	Master mode	5	-	-				
t <sub>h(SI)</sub>	Data input noid time	Slave mode	5	-	-	ns			
t <sub>a(SO)</sub>	Data output access time	Slave mode	0	-	4*Tpclk	113			
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	0	-	24				
		Slave mode	-	12	27				
t <sub>v(SO)</sub>	Data output valid time	Slave mode, SPI1 2.7 <v<sub>DD&lt;3.6V</v<sub>	-	12	18				
		Master mode	-	1.5	3				
t <sub>h(SO)</sub>	Data output hold time	Slave mode	11	-	-				
t <sub>h(MO)</sub>	Data output noid time	Master mode	0	-	-				

<sup>1.</sup> Guaranteed by characterization results.



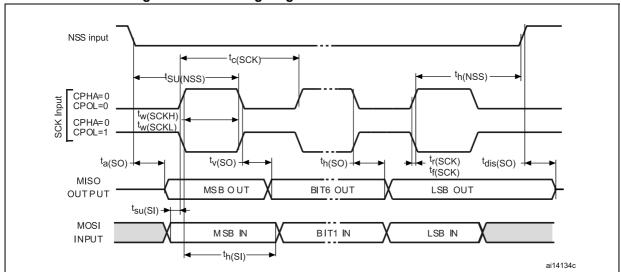
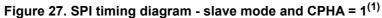
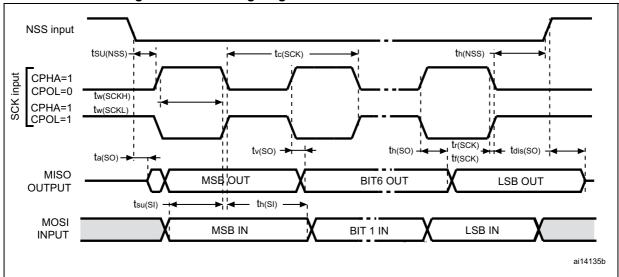


Figure 26. SPI timing diagram - slave mode and CPHA = 0





1. Measurement points are done at  $0.5V_{DD}$  and with external  $C_L$  = 30 pF.

57/

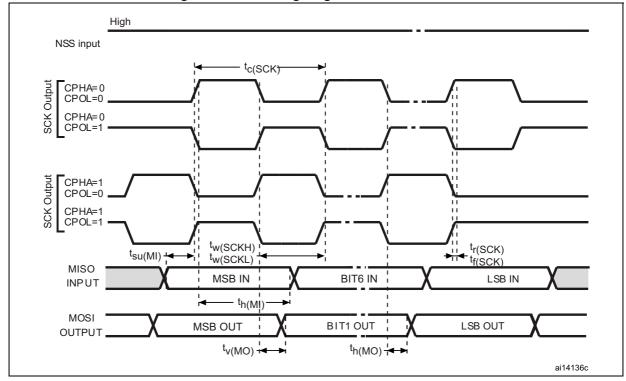


Figure 28. SPI timing diagram - master mode<sup>(1)</sup>

1. Measurement points are done at 0.5V $_{\rm DD}$  and with external C $_{\rm L}$  = 30 pF.



Table 64. I<sup>2</sup>S characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit	
f <sub>CK</sub>	I <sup>2</sup> S clock frequency	Master data: 16 bits, audio freq=48 kHz	1.496	1.503	MHz	
1/t <sub>c(CK)</sub>	1 3 clock frequency	Slave	0	12.288	IVITZ	
t <sub>r(CK)</sub>	I <sup>2</sup> S clock rise and fall time	Capacitive load C <sub>L</sub> = 30 pF	-	8		
t <sub>w(CKH)</sub>	I <sup>2</sup> S clock high time	Master f <sub>PCLK</sub> = 36 MHz,	331	-		
t <sub>w(CKL)</sub>	I <sup>2</sup> S clock low time	audio frequency = 48 kHz	332	-		
t <sub>v(WS)</sub>	WS valid time	Master mode	4	-	ns	
t <sub>h(WS)</sub>	WS hold time	Master mode	4	-		
t <sub>su(WS)</sub>	WS setup time	Slave mode	4	-		
t <sub>h(WS)</sub>	WS hold time	Slave mode	0	-		
Duty Cycle	I <sup>2</sup> S slave input clock duty cycle	Slave mode	30	70	%	
t <sub>su(SD_MR)</sub>	Data input setup time	Master receiver	9	-		
t <sub>su(SD_SR)</sub>	Data input setup time	Slave receiver	2	-		
t <sub>h(SD_MR)</sub>	- Data input hold time	Master receiver	0	-		
t <sub>h(SD_SR)</sub>		Slave receiver	0	-		
t <sub>v(SD_ST)</sub>	Data output valid time	Slave transmitter (after enable edge)	-	29	ns	
t <sub>h(SD_ST)</sub>	Data output hold time	Slave transmitter (after enable edge)	12	-		
t <sub>v(SD_MT)</sub>	Data output valid time	Master transmitter (after enable edge)	-	3		
t <sub>h(SD_MT)</sub>	Data output hold time	Master transmitter (after enable edge)	2	-		

<sup>1.</sup> Guaranteed by characterization results.

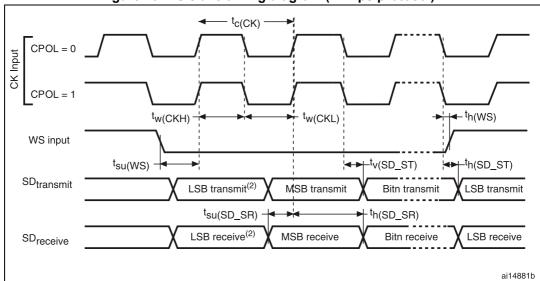


Figure 29. I<sup>2</sup>S slave timing diagram (Philips protocol)<sup>(1)</sup>

- Measurement points are done at  $0.5V_{DD}$  and with external  $C_L$ =30 pF.
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first

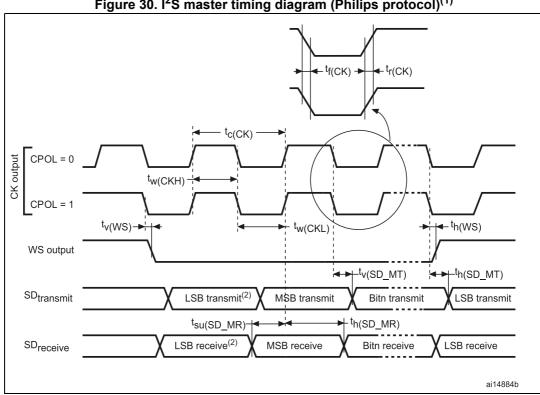


Figure 30. I<sup>2</sup>S master timing diagram (Philips protocol)<sup>(1)</sup>

- 1. Measurement points are done at  $0.5V_{DD}$  and with external  $C_L$ =30 pF.
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first

#### **USB** characteristics

Table 65. USB startup time

Symbol	Parameter	Max	Unit
t <sub>STARTUP</sub> <sup>(1)</sup>	USB transceiver startup time	1	μs

<sup>1.</sup> Guaranteed by design.

Table 66. USB DC electrical characteristics

Symbol	Parameter	Conditions	Min. <sup>(1)</sup>	Max. <sup>(1)</sup>	Unit		
Input levels							
V <sub>DD</sub>	USB operating voltage <sup>(2)</sup>	-	3.0 <sup>(3)</sup>	3.6	V		
V <sub>DI</sub> <sup>(4)</sup>	Differential input sensitivity	I(USB_DP, USB_DM)	0.2	-			
V <sub>CM</sub> <sup>(4)</sup>	Differential common mode range	Includes V <sub>DI</sub> range	0.8	2.5	٧		
V <sub>SE</sub> <sup>(4)</sup>	Single ended receiver threshold	-	1.3	2.0			
Output levels							
V <sub>OL</sub>	Static output level low	$R_L$ of 1.5 k $\Omega$ to 3.6 $V^{(5)}$	-	0.3	V		
V <sub>OH</sub>	Static output level high	$R_L$ of 15 k $\Omega$ to $V_{SS}^{(5)}$	2.8	3.6	'		

- 1. All the voltages are measured from the local ground potential.
- 2. To be compliant with the USB 2.0 full-speed electrical specification, the USB\_DP (D+) pin should be pulled up with a 1.5 k $\Omega$  resistor to a 3.0-to-3.6 V voltage range.
- 3. The STM32F303xB/STM32F303xC USB functionality is ensured down to 2.7 V but not the full USB electrical characteristics which are degraded in the 2.7-to-3.0 V  $V_{DD}$  voltage range.
- 4. Guaranteed by design.
- 5.  $R_L$  is the load connected on the USB drivers.

Figure 31. USB timings: definition of data signal rise and fall time

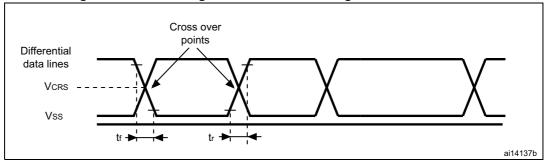




Table 67. USB: Full-speed electrical characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit		
Driver charac	Driver characteristics							
t <sub>r</sub>	Rise time <sup>(2)</sup>	C <sub>L</sub> = 50 pF	4	-	20	ns		
t <sub>f</sub>	Fall time <sup>(2)</sup>	C <sub>L</sub> = 50 pF	4	-	20	ns		
t <sub>rfm</sub>	Rise/ fall time matching	t <sub>r</sub> /t <sub>f</sub>	90	-	110	%		
V <sub>CRS</sub>	Output signal crossover voltage	-	1.3	-	2.0	V		
Output driver Impedance <sup>(3)</sup>	Z <sub>DRV</sub>	driving high and low	28	40	44	Ω		

<sup>1.</sup> Guaranteed by design.

## CAN (controller area network) interface

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (CAN\_TX and CAN\_RX).

Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).

<sup>3.</sup> No external termination series resistors are required on USB\_DP (D+) and USB\_DM (D-), the matching impedance is already included in the embedded driver.

# 6.3.18 ADC characteristics

Unless otherwise specified, the parameters given in *Table 68* to *Table 70* are guaranteed by design, with conditions summarized in *Table 24*.

**Table 68. ADC characteristics** 

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{DDA}$	Analog supply voltage for ADC	-	2	-	3.6	V
		Single-ended mode, 5 MSPS	-	907	1033.0	
		Single-ended mode, 1 MSPS	-	194	285.5	
	ADC current consumption on VDDA	Single-ended mode, 200 KSPS	-	51.5	70	] 
I <sub>DDA</sub>	pin (see <i>Figure 32</i> )	Differential mode, 5 MSPS	-	887.5	1009	- μΑ
		Differential mode, 1 MSPS	-	212	285	
		Differential mode, 200 KSPS	-	51	69.5	
V <sub>REF+</sub>	Positive reference voltage	-	2	-	$V_{DDA}$	V
V <sub>REF-</sub>	Negative reference voltage	-	-	0	-	
		Single-ended mode, 5 MSPS	-	104	139	
		Single-ended mode, 1 MSPS	-	20.4	37	
1	ADC current consumption on VREF+	Single-ended mode, 200 KSPS	-	3.3	11.3	μΑ
I <sub>REF</sub>	pin (see <i>Figure 33</i> )	Differential mode, 5 MSPS	-	174	235	μΛ
		Differential mode, 1 MSPS	-	34.6	52.6	
		Differential mode, 200 KSPS	-	6	13.6	





Table 68. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>ADC</sub>	ADC clock frequency	-	0.14	-	72	MHz
		Resolution = 12 bits, Fast Channel	0.01	-	5.14	
f <sub>S</sub> <sup>(1)</sup>		Resolution = 10 bits, Fast Channel	0.012	-	6	— MSPS
'S` '	Sampling rate	Resolution = 8 bits, Fast Channel	0.014	-	7.2	- IVISES
		Resolution = 6 bits, Fast Channel	0.0175	-	9	
f <sub>TRIG</sub> <sup>(1)</sup>	External trigger frequency	f <sub>ADC</sub> = 72 MHz Resolution = 12 bits	-	-	5.14	MHz
		Resolution = 12 bits	-	-	14	1/f <sub>ADC</sub>
V <sub>AIN</sub>	Conversion voltage range <sup>(2)</sup>	-	0	-	V <sub>REF+</sub>	V
R <sub>AIN</sub> <sup>(1)</sup>	External input impedance	-	-	-	100	kΩ
C <sub>ADC</sub> <sup>(1)</sup>	Internal sample and hold capacitor	-	-	5	-	pF
t <sub>STAB</sub> <sup>(1)</sup>	Power-up time	-	1			conversion cycle
t <sub>CAL</sub> <sup>(1)</sup>	Calibration time	f <sub>ADC</sub> = 72 MHz	1.56			μs
CAL`	Calibration time	-	112			1/f <sub>ADC</sub>
		CKMODE = 00	1.5	2	2.5	1/f <sub>ADC</sub>
t <sub>latr</sub> <sup>(1)</sup>	Trigger conversion latency Regular and injected channels without conversion abort	CKMODE = 01	-	-	2	1/f <sub>ADC</sub>
latr` ′		CKMODE = 10	-	-	2.25	1/f <sub>ADC</sub>
		CKMODE = 11	-	-	2.125	1/f <sub>ADC</sub>
		CKMODE = 00	2.5	3	3.5	1/f <sub>ADC</sub>
<b>4</b> (1)	Trigger conversion latency	CKMODE = 01	-	-	3	1/f <sub>ADC</sub>
t <sub>latrinj</sub> (1)	Injected channels aborting a regular conversion	CKMODE = 10	-	-	3.25	1/f <sub>ADC</sub>
		CKMODE = 11	-	-	3.125	1/f <sub>ADC</sub>

Table 68. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
ts <sup>(1)</sup>	Sampling time	f <sub>ADC</sub> = 72 MHz	0.021	-	8.35	μs
is.		-	1.5	-	601.5	1/f <sub>ADC</sub>
T <sub>ADCVREG_STUP</sub> (1)	ADC Voltage Regulator Start-up time	-	-	-	10	μs
t <sub>CONV</sub> <sup>(1)</sup>	Total conversion time (including sampling time)	f <sub>ADC</sub> = 72 MHz Resolution = 12 bits	0.19	-	8.52	μs
CONV, ,		Resolution = 12 bits	14 to 614 (t <sub>S</sub> for sampling + 12.5 for successive approximation)		1/f <sub>ADC</sub>	
CMIR <sup>(1)</sup>	Common Mode Input signal Range	ADC differential mode	(V <sub>SSA</sub> +V <sub>REF+</sub> )/2 -10%	(V <sub>SSA</sub> +V <sub>REF+</sub> )/2	(V <sub>SSA</sub> +V <sub>REF+</sub> )/2 + 10%	V

<sup>1.</sup> Data guaranteed by design.



<sup>2.</sup> V<sub>REF+</sub> can be internally connected to V<sub>DDA</sub> and V<sub>REF-</sub> can be internally connected to V<sub>SSA</sub>, depending on the package. Refer to Section 4: Pinouts and pin description for further details.

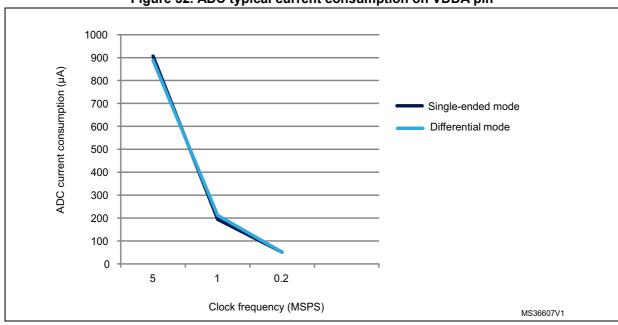


Figure 32. ADC typical current consumption on VDDA pin



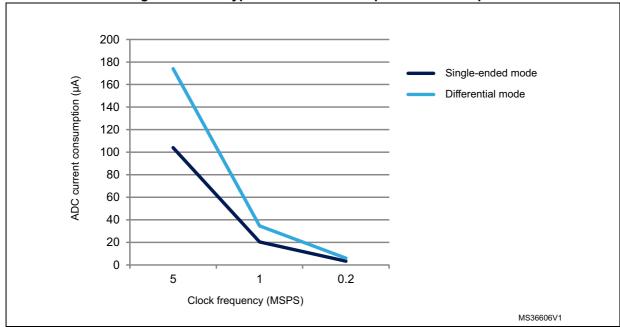


Table 69. Maximum ADC R<sub>AIN</sub> <sup>(1)</sup>

	Sampling	Sampling	ium ADC R <sub>AIN</sub> (*)	R <sub>AIN</sub> max (kΩ)	
Resolution	cycle @ 72 MHz	time [ns] @ 72 MHz	Fast channels <sup>(2)</sup>	Slow channels	Other channels <sup>(3)</sup>
	1.5	20.83	0.018	NA	NA
	2.5	34.72	0.150	NA	0.022
	4.5	62.50	0.470	0.220	0.180
40 6:4-	7.5	104.17	0.820	0.560	0.470
12 bits	19.5	270.83	2.70	1.80	1.50
	61.5	854.17	8.20	6.80	4.70
	181.5	2520.83	22.0	18.0	15.0
	601.5	8354.17	82.0	68.0	47.0
	1.5	20.83	0.082	NA	NA
	2.5	34.72	0.270	0.082	0.100
	4.5	62.50	0.560	0.390	0.330
40.15%	7.5	104.17	1.20	0.82	0.68
10 bits	19.5	270.83	3.30	2.70	2.20
	61.5	854.17	10.0	8.2	6.8
	181.5	2520.83	33.0	27.0	22.0
	601.5	8354.17	100.0	82.0	68.0
	1.5	20.83	0.150	NA	0.039
	2.5	34.72	0.390	0.180	0.180
	4.5	62.50	0.820	0.560	0.470
0 6:4-	7.5	104.17	1.50	1.20	1.00
8 bits	19.5	270.83	3.90	3.30	2.70
	61.5	854.17	12.00	12.00	8.20
	181.5	2520.83	39.00	33.00	27.00
	601.5	8354.17	100.00	100.00	82.00
	1.5	20.83	0.270	0.100	0.150
	2.5	34.72	0.560	0.390	0.330
	4.5	62.50	1.200	0.820	0.820
0.54	7.5	104.17	2.20	1.80	1.50
6 bits	19.5	270.83	5.60	4.70	3.90
	61.5	854.17	18.0	15.0	12.0
	181.5	2520.83	56.0	47.0	39.0
	601.5	8354.17	100.00	100.0	100.0

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> All fast channels, expect channels on PA2, PA6, PB1, PB12.

3. Channels available on PA2, PA6, PB1, PB12.

Table 70. ADC accuracy - limited test conditions, 100-pin packages (1)(2)

Symbol	Parameter	(	Conditions		Min (3)	Тур	Max (3)	Unit
			Single anded	Fast channel 5.1 Ms	-	±3.5	±4.5	
ET	Total unadjusted		Single ended	Slow channel 4.8 Ms	-	<u>±</u> 4	±4.5	
	error		Differential	Fast channel 5.1 Ms	-	±3	±3	
			Differential	Slow channel 4.8 Ms	-	±3	±3	
			Single anded	Fast channel 5.1 Ms	-	±1	±1.5	
FO	Officet error		Single ended	Slow channel 4.8 Ms	-	±1	±2.5	
EO	Offset error		Differential	Fast channel 5.1 Ms	-	±1	±1.5	
				Slow channel 4.8 Ms	-	±1	±1.5	
			Single ended Slov Differential		-	±3	±4	
EG	Cain arrar				-	±3.5	±4	LSB
EG	Gain error				-	±1.5	±2.5	LSB
			Dillerential	Slow channel 4.8 Ms	-	<u>+2</u>	±2.5	1
	ADC clock freq. ≤72 MHz	Single ended	Fast channel 5.1 Ms	-	±1	±1.5	-	
ED	Differential	Sampling freq. ≤ 5 Msps V <sub>DDA</sub> = V <sub>REF+</sub> = 3.3 V 25°C	Single ended	Slow channel 4.8 Ms	-	±1	±1.5	
ED	linearity error		Differential	Fast channel 5.1 Ms	-	±1	±1	
		100-pin package	Differential	Slow channel 4.8 Ms	-	±1	±1	
			Cinale ended	Fast channel 5.1 Ms	-	±1.5	<u>+2</u>	
-	Integral		Single ended	Slow channel 4.8 Ms	-	±1.5	±3	
EL	linearity error		Differential	Fast channel 5.1 Ms	-	±1	±1.5	
			Dillerential	Slow channel 4.8 Ms	-	±1	±1.5	
			Cinalo ondod	Fast channel 5.1 Ms	10.7	10.8	-	
ENOB <sup>(4)</sup>	Effective		Single ended	Slow channel 4.8 Ms	10.7	10.8	-	hita
ENOB	number of bits		Differential	Fast channel 5.1 Ms	11.2	11.3	-	bits
			Differential	Slow channel 4.8 Ms	11.1	11.3	-	
	Signal to		0	Fast channel 5.1 Ms	66	67	-	
SINAD <sup>(4)</sup>	Signal-to- noise and		Single ended	Slow channel 4.8 Ms	66	67	-	40
SINAD	distortion	on		Fast channel 5.1 Ms	69	70	-	dB
	ratio		Differential	Slow channel 4.8 Ms	69	70	-	



Table 70. ADC accuracy - limited test conditions, 100-pin packages (1)(2) (continued)

Symbol	Parameter	C	Min (3)	Тур	Max (3)	Unit		
			Single ended	Fast channel 5.1 Ms	66	67	-	
SNR <sup>(4)</sup> Signal-to- noise ratio		Single ended	Slow channel 4.8 Ms	66	67	-		
	noise ratio	ADC clock freq. ≤ 72 MHz Sampling freq ≤ 5 Msps	Differential	Fast channel 5.1 Ms	69	70	-	
				Slow channel 4.8 Ms	69	70	-	dB
		V <sub>DDA</sub> = V <sub>REF+</sub> = 3.3 V 25°C 100-pin package	0:	Fast channel 5.1 Ms	-	-76	-76	uБ
THD <sup>(4)</sup>	Total harmonic		Single ended	Slow channel 4.8 Ms	-	-76	-76	
	distortion		Differential	Fast channel 5.1 Ms	-	-80	-80	
	0.0.0.		Differential	Slow channel 4.8 Ms	ı	-80	-80	

<sup>1.</sup> ADC DC accuracy values are measured after internal calibration.

- 3. Guaranteed by characterization results.
- 4. Value measured with a -0.5 dB full scale 50 kHz sine wave input signal.

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<sup>2.</sup> ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.

Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and ΣI<sub>INJ(PIN)</sub> in Section 6.3.14 does not affect the ADC accuracy.

Table 71. ADC accuracy, 100-pin packages<sup>(1)(2)(3)</sup>

Symbol	Parameter	С	onditions		Min <sup>(4)</sup>	Max <sup>(4)</sup>	Unit
			Single	Fast channel 5.1 Ms	-	±6.5	
ГТ	Total		Ended	Slow channel 4.8 Ms	-	±6.5	
ET	unadjusted error		Differential	Fast channel 5.1 Ms	-	±4	
			Differential	Slow channel 4.8 Ms	-	±4	
			Single	Fast channel 5.1 Ms	-	±3	
EO	O Offset error		Ended	Slow channel 4.8 Ms	-	±3	
EU			Differential	Fast channel 5.1 Ms	-	<u>+2</u>	
			טוווכופוונומו	Slow channel 4.8 Ms	ı	<u>+2</u>	
			Single	Fast channel 5.1 Ms	ı	<u>±</u> 6	
EG	Gain error	Ended	Slow channel 4.8 Ms	ı	<u>±</u> 6	LSB	
EG Gain error	Gain enoi	ADC clock freq. $\leq$ 72 MHz, Sampling freq. $\leq$ 5 Msps 2 V $\leq$ V <sub>DDA</sub> , V <sub>REF+</sub> $\leq$ 3.6 V	Differential	Fast channel 5.1 Ms	ı	±3	LSB
				Slow channel 4.8 Ms	ı	±3	
			Single Ended	Fast channel 5.1 Ms	ı	±1.5	
ED	Differential linearity			Slow channel 4.8 Ms	ı	±1.5	
LD	error		Differential	Fast channel 5.1 Ms	ı	±1.5	
			Dilicicitiai	Slow channel 4.8 Ms	ı	±1.5	
			Single	Fast channel 5.1 Ms	ı	<u>+2</u>	
EL	Integral linearity		Ended	Slow channel 4.8 Ms	ı	±3	
LL	error		Differential	Fast channel 5.1 Ms	ı	<u>+2</u>	
	05		Dillerential	Slow channel 4.8 Ms	ı	<u>+2</u>	
			Single	Fast channel 5.1 Ms	10.4	-	
ENOB	Effective number of		Ended	Slow channel 4.8 Ms	10.2	-	bits
(5)	bits		Differential	Fast channel 5.1 Ms	10.8	-	טונס
			Dilletetitial	Slow channel 4.8 Ms	10.8	-	



Table 71. ADC accuracy, 100-pin packages<sup>(1)(2)(3)</sup> (continued)

Symbol	Parameter	С	onditions		Min <sup>(4)</sup>	Max <sup>(4)</sup>	Unit
	Cianal to		Single	Fast channel 5.1 Ms	64	-	
SINAD	Signal-to- noise and		Ended	Slow channel 4.8 Ms	63	-	
(5) distortion ratio		Differential	Fast channel 5.1 Ms	67	-		
		Dillerential	Slow channel 4.8 Ms	67	-		
	SNR <sup>(5)</sup> Signal-to- noise ratio	ADC clock freq. $\leq$ 72 MHz, Sampling freq. $\leq$ 5 Msps, 2 V $\leq$ V <sub>DDA</sub> , V <sub>REF+</sub> $\leq$ 3.6 V	Single Ended	Fast channel 5.1 Ms	64	-	
CNID(5)				Slow channel 4.8 Ms	64	-	dB
SINK			Differential	Fast channel 5.1 Ms	67	-	uБ
		100-pin package		Slow channel 4.8 Ms	67	-	
			Single	Fast channel 5.1 Ms	-	-74	
THD <sup>(5)</sup>	Total harmonic		Ended	Slow channel 4.8 Ms	-	-74	
	distortion		Differential	Fast channel 5.1 Ms	-	-78	
			חוובובוונומו	Slow channel 4.8 Ms	-	-76	

- 1. ADC DC accuracy values are measured after internal calibration.
- ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
   Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and ΣI<sub>INJ(PIN)</sub> in Section 6.3.14 does not affect the ADC accuracy.
- 3. Better performance may be achieved in restricted  $V_{\text{DDA}}$ , frequency and temperature ranges.
- 4. Guaranteed by characterization results.
- 5. Value measured with a -0.5 dB full scale 50 kHz sine wave input signal.

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Table 72. ADC accuracy - limited test conditions, 64-pin packages<sup>(1)(2)</sup>

Symbol	Parameter	C	Conditions		Min (3)	Тур	Max (3)	Unit
			Single ended	Fast channel 5.1 Ms	-	±4	±4.5	
ET	Total		Sirigle efficed	Slow channel 4.8 Ms	-	±5.5	±6	
	unadjusted error		Differential	Fast channel 5.1 Ms	-	±3.5	±4	
			Differential	Slow channel 4.8 Ms	-	±3.5	±4	
		error	Cinala andad	Fast channel 5.1 Ms	-	±2	±2	
F0	Offeeter		Single ended	Slow channel 4.8 Ms	-	±1.5	±2	
EO	Offset error		Differential -	Fast channel 5.1 Ms	-	±1.5	±2	
				Slow channel 4.8 Ms	-	±1.5	±2	
			Cinale anded	Fast channel 5.1 Ms	-	±3	±4	
F0	0-:		Single ended	Slow channel 4.8 Ms	-	±5	±5.5	
EG	Gain error		Differential	Fast channel 5.1 Ms	-	±3	±3	LSB
		Differential	Slow channel 4.8 Ms	-	±3	±3.5	- -	
	ADC clock freq. ≤72 MHz	0:11-1	Fast channel 5.1 Ms	-	±1	±1		
	Differential	Sampling freq. ≤5 Msps V <sub>DDA</sub> = 3.3 V 25°C	Single ended	Slow channel 4.8 Ms	-	±1	±1	-
ED	linearity error		Differential	Fast channel 5.1 Ms	-	±1	±1	
		64-pin package	Differential -	Slow channel 4.8 Ms	-	±1	±1	
			Cinala andad	Fast channel 5.1 Ms	-	±1.5	±2	
	Integral		Single ended	Slow channel 4.8 Ms	-	±2	±3	
EL	linearity error		Differential	Fast channel 5.1 Ms	-	±1.5	±1.5	
			Differential	Slow channel 4.8 Ms	-	±1.5	±2	
			Single anded	Fast channel 5.1 Ms	10.8	10.8	-	
ENOB	Effective		Single ended	Slow channel 4.8 Ms	10.8	10.8	-	h:4
(4)	number of bits		D:#ti-l	Fast channel 5.1 Ms	11.2	11.3	-	bit
		Differential	Slow channel 4.8 Ms	11.2	11.3	-		
	Cignal to		Cinala	Fast channel 5.1 Ms	66	67	-	
SINAD	Signal-to- noise and		Single ended	Slow channel 4.8 Ms	66	67	-	40
(4)	distortion		F	Fast channel 5.1 Ms	69	70	-	dB
	ratio		Differential	Slow channel 4.8 Ms	69	70	-	



Table 72. ADC accuracy - limited test conditions, 64-pin packages<sup>(1)(2)</sup> (continued)

Symbol	Parameter	C	Min (3)	Тур	Max (3)	Unit		
			Cinale anded	Fast channel 5.1 Ms	66	67	-	
SNR <sup>(4)</sup> Signal-to- noise ratio	Signal-to-	ADC clock freq. ≤ 72 MHz Sampling freq ≤ 5 Msps	Single ended	Slow channel 4.8 Ms	66	67	-	-
	noise ratio		Differential	Fast channel 5.1 Ms	69	70	-	
				Slow channel 4.8 Ms	69	70	-	dB
		V <sub>DDA</sub> = 3.3 V 25°C 64-pin package	Circula and ad	Fast channel 5.1 Ms	-	-80	-80	uв
THD <sup>(4)</sup>	Total harmonic		Single ended	Slow channel 4.8 Ms	-	-78	-77	
	distortion		Differential	Fast channel 5.1 Ms	-	-83	-82	
			Dillerefillal	Slow channel 4.8 Ms	-	-81	-80	

<sup>1.</sup> ADC DC accuracy values are measured after internal calibration.

<sup>2.</sup> ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current. Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and ΣI<sub>INJ(PIN)</sub> in Section 6.3.14 does not affect the ADC accuracy.

<sup>3.</sup> Guaranteed by characterization results.

<sup>4.</sup> Value measured with a -0.5 dB full scale 50 kHz sine wave input signal.

Table 73. ADC accuracy, 64-pin packages<sup>(1)(2)(3)</sup>

Symbol	Parameter	C	Conditions		Min <sup>(4)</sup>	Max (4)	Unit
			Cinalo andod	Fast channel 5.1 Ms	-	±6.5	
	Total		Single ended	Slow channel 4.8 Ms	-	±6.5	
ET	unadjusted error		Differential	Fast channel 5.1 Ms	-	±4	
			Differential	Slow channel 4.8 Ms	-	±4.5	
			Cinalo andod	Fast channel 5.1 Ms	-	±3	
EO	Offeet error	Offset error	Single ended	Slow channel 4.8 Ms	-	±3	
EO	Oliset error		Differential	Fast channel 5.1 Ms	-	±2.5	
			Differential -	Slow channel 4.8 Ms	-	±2.5	
			Cinalo andod	Fast channel 5.1 Ms	-	±6	
F0	Cain aman		Single ended	Slow channel 4.8 Ms	-	±6	- LSB
EG	Gain error		Differential	Fast channel 5.1 Ms	-	±3.5	
				Slow channel 4.8 Ms	-	±4	
		ADC clock freq. $\leq$ 72 MHz, Sampling freq. $\leq$ 5 Msps $2.0 \text{ V} \leq \text{V}_{DDA} \leq 3.6 \text{ V}$	Single anded	Fast channel 5.1 Ms	-	±1.5	
ED	Differential		Single ended	Slow channel 4.8 Ms	-	±1.5	
ED	linearity error		Differential -	Fast channel 5.1 Ms	-	±1.5	
		64-pin package	Dillerential	Slow channel 4.8 Ms	-	±1.5	
			Cingle anded	Fast channel 5.1 Ms	-	±3	
EL	Integral		Single ended	Slow channel 4.8 Ms	-	±3.5	
EL	linearity error		Differential	Fast channel 5.1 Ms	-	±2	
			Dillerential	Slow channel 4.8 Ms	-	±2.5	
			Single ended	Fast channel 5.1 Ms	10.4	-	
ENOB	Effective		Sirigle ended	Slow channel 4.8 Ms	10.4	-	hito
(5)	number of bits		Differential	Fast channel 5.1 Ms	10.8	-	bits
		Dillerential	Slow channel 4.8 Ms	10.8	-		
	Signal-to- SINAD noise and		Single anded	Fast channel 5.1 Ms	64	-	
			Single ended -	Slow channel 4.8 Ms	63	-	dB
(5)	distortion ratio		Differential	Fast channel 5.1 Ms	67	-	ub
	Tallo		Dillerential	Slow channel 4.8 Ms	67	-	



Symbol	Parameter	C	Min <sup>(4)</sup>	Max (4)	Unit		
		ADC clock freq. ≤ 72 MHz, Sampling freq ≤ 5 Msps,	Single ended	Fast channel 5.1 Ms	64	-	
	Signal-to-		Sirigle efficed	Slow channel 4.8 Ms	64	-	
	noise ratio		Differential	Fast channel 5.1 Ms	67	-	
				Slow channel 4.8 Ms	67	-	dB
		onic	Cingle anded	Fast channel 5.1 Ms	-	-75	uБ
THD <sup>(5)</sup>	Total harmonic		Single ended	Slow channel 4.8 Ms	-	-75	]
י יטחו	distortion		Differential	Fast channel 5.1 Ms	-	-79	
			Dillerential	Slow channel 4.8 Ms	-	-78	

Table 73. ADC accuracy, 64-pin packages<sup>(1)(2)(3)</sup> (continued)

- 1. ADC DC accuracy values are measured after internal calibration.
- 2. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current. Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and ΣI<sub>INJ(PIN)</sub> in Section 6.3.14 does not affect the ADC accuracy.
- 3. Better performance may be achieved in restricted V<sub>DDA</sub>, frequency and temperature ranges.
- 4. Guaranteed by characterization results.
- 5. Value measured with a -0.5 dB full scale 50 kHz sine wave input signal.

Table 74. ADC accuracy at 1MSPS<sup>(1)(2)</sup>

Symbol	Parameter	Test condition	ıs	Тур	Max <sup>(3)</sup>	Unit
ЕТ	ET Total unadjusted error		Fast channel	±2.5	±5	
	Total urlaujusteu error		Slow channel	±3.5	±5	
EO	Offset error		Fast channel	±1	±2.5	
	Oliset elloi	ADC Freq $\leq$ 72 MHz Sampling Freq $\leq$ 1MSPS 2.4 V $\leq$ V <sub>DDA</sub> = V <sub>REF+</sub> $\leq$ 3.6 V	Slow channel	±1.5	±2.5	
EG	Gain error		Fast channel	±2	±3	LSB
EG	Gairrenoi		Slow channel	±3	±4	LOD
ED	Differential linearity error	Single-ended mode	Fast channel	±0.7	±2	
	Differential linearity error		Slow channel	±0.7	±2	
EL	Integral linearity error		Fast channel	±1	±3	
	Integral linearity error		Slow channel	±1.2	±3	

- 1. ADC DC accuracy values are measured after internal calibration.
- 2. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current. Any positive injection current within the limits specified for IINJ(PIN) and ∑IINJ(PIN) in Section 6.3.14: I/O port characteristics does not affect the ADC accuracy.
- 3. Guaranteed by characterization results.

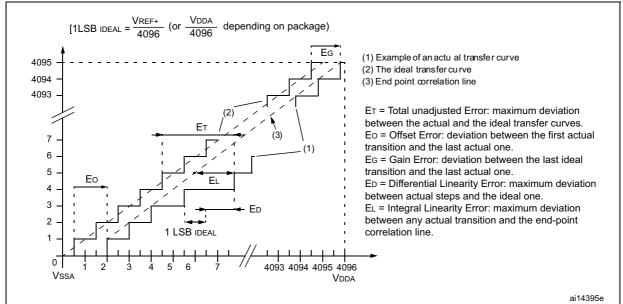
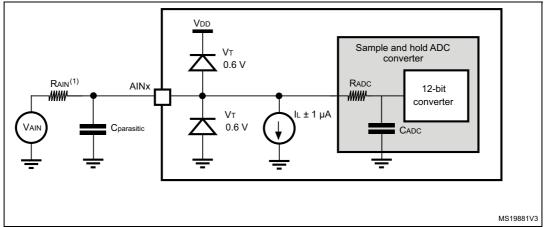


Figure 34. ADC accuracy characteristics





- Refer to Table 68 for the values of R<sub>AIN</sub>.
- C<sub>parasitic</sub> represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high C<sub>parasitic</sub> value will downgrade conversion accuracy. To remedy this, f<sub>ADC</sub> should be reduced.

### General PCB design guidelines

Power supply decoupling should be performed as shown in *Figure 11*. The 10 nF capacitor should be ceramic (good quality) and it should be placed as close as possible to the chip.



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# 6.3.19 DAC electrical specifications

Table 75. DAC characteristics

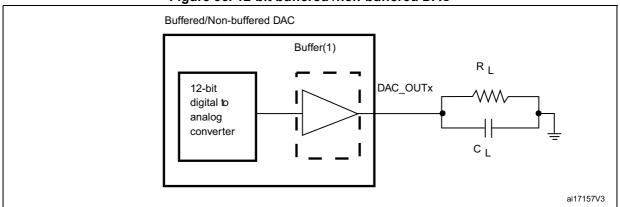
Symbol	Parameter	С	onditions	Min	Тур	Max	Unit
$V_{DDA}$	Analog supply voltage		-	2.4	-	3.6	V
R <sub>LOAD</sub> <sup>(1)</sup>	Resistive load		Connected to V <sub>SSA</sub>	5	-	-	kΩ
	resistive load	buffer ON	Connected to V <sub>DDA</sub>	25	-	-	1(32
R <sub>O</sub> <sup>(1)</sup>	Output impedance	DAC output	buffer OFF	-	-	15	kΩ
C <sub>LOAD</sub> <sup>(1)</sup>	Capacitive load	DAC output	buffer ON	-	-	50	pF
V <sub>DAC_OUT</sub> (1)	Voltage on DAC_OUT output	code (0x0E0 V <sub>DDA</sub> = 3.6 \ and (0x155)			1	V <sub>DDA</sub> – 0.2	V
		DAC output	buffer OFF	-	0.5	V <sub>DDA</sub> - 1LSB	mV
I <sub>DDA</sub> <sup>(3)</sup>	DAC DC current consumption in quiescent		Vith no load, middle code 0x800) on the input.		ı	380	μΑ
'DDA'	mode (Standby mode) <sup>(2)</sup>		(ith no load, worst code xF1C) on the input.		-	480	μΑ
(3)	Differential non linearity	Given for a	10-bit input code	-	-	±0.5	LSB
DNL <sup>(3)</sup>	Difference between two consecutive code-1LSB)	Given for a	12-bit input code	-	-	±2	LSB
	Integral non linearity	Given for a	10-bit input code	-	-	±1	LSB
INL <sup>(3)</sup>	(difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095)	Given for a 12-bit input code		1	-	±4	LSB
			-	-	-	±10	mV
Offset <sup>(3)</sup>	Offset error (difference between measured value at Code (0x800) and the ideal	Given for a V <sub>DDA</sub> = 3.6 '	10-bit input code at V	-	-	±3	LSB
	value = V <sub>DDA</sub> /2)	Given for a V <sub>DDA</sub> = 3.6 '	12-bit input code at V	-	-	±12	LSB
Gain error <sup>(3)</sup>	Gain error	Given for a	12-bit input code	-	-	±0.5	%
t <sub>SETTLING</sub> (3)		$C_{LOAD}$ ≤50 pF, $R_{LOAD}$ ≥ 5 kΩ		-	3	4	μs
Update rate <sup>(3)</sup>		C <sub>LOAD</sub> ⊴50 դ R <sub>LOAD</sub> ≥ 5		-	-	1	MS/s

Table 75. DAG	characteristics	(continued)
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Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	DAC Control register)	C <sub>LOAD</sub> ≤50 pF, R <sub>LOAD</sub> ≥ 5 kΩ	-	6.5	10	μs
PSRR+ (1)	Power supply rejection ratio (to V <sub>DDA</sub> ) (static DC measurement	$C_{LOAD} = 50 \text{ pF},$ No $R_{LOAD} \ge 5 \text{ k}\Omega,$	-	<b>–</b> 67	-40	dB

- 1. Guaranteed by design.
- 2. Quiescent mode refers to the state of the DAC a keeping steady value on the output, so no dynamic consumption is involved.
- 3. Guaranteed by characterization results.

Figure 36. 12-bit buffered /non-buffered DAC



 The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC\_CR register.

## 6.3.20 Comparator characteristics

Table 76. Comparator characteristics<sup>(1)</sup>

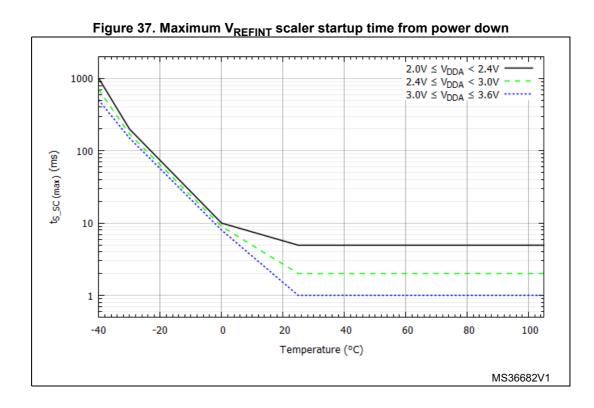
Symbol	Parameter	Condit	ions	Min	Тур	Max	Unit
$V_{DDA}$	Analog supply voltage	-		2	-	3.6	
V <sub>IN</sub>	Comparator input voltage range	-	0	-	V <sub>DDA</sub>	٧	
$V_{BG}$	Scaler input voltage	-		-	1.2	-	
V <sub>SC</sub>	Scaler offset voltage	-		-	±5	±10	mV
t <sub>S_SC</sub>	V <sub>REFINT</sub> scaler startup time from power down	First V <sub>REFINT</sub> scaler activation after device power on		-	-	1 <sup>(2)</sup>	s
	time from power down	Next activ	vations	-	-	0.2	ms
t <sub>START</sub>	Comparator startup time	Startup time to reach prespecification	-	-	60	μs	
		Ultra-low-power mode		-	2	4.5	
	Propagation delay for	Low-power mode		-	0.7	1.5	μs
	200 mV step with 100 mV	Medium power mode		-	0.3	0.6	
	overdrive	V <sub>DDA</sub> ≥ 2.7 V	$V_{DDA} \ge 2.7 \text{ V}$	-	50	100	
4		High speed mode	V <sub>DDA</sub> < 2.7 V	-	100	240	ns
$t_{D}$		Ultra-low-power mode		-	2	7	
	Propagation delay for full	Low-power mode		-	0.7	2.1	μs
	range step with 100 mV	Medium power mode		-	0.3	1.2	
	overdrive	l liab anad made	$V_{DDA} \ge 2.7 \text{ V}$	-	90	180	
		High speed mode	V <sub>DDA</sub> < 2.7 V	-	110	300	ns
V <sub>offset</sub>	Comparator offset error	-	·	-	±4	±10	mV
dV <sub>offset</sub> /dT	Offset error temperature coefficient	-		-	18	-	μV/° C
		Ultra-low-power mode		-	1.2	1.5	
	COMP current	Low-power mode		-	3	5	μΑ
I <sub>DD(COMP)</sub>	consumption	Medium power mode		-	10	15	
		High speed mode		-	75	100	



**Symbol Conditions** Min Unit **Parameter** Тур Max No hysteresis 0 (COMPxHYST[1:0]=00) High speed mode 13 Low hysteresis 8 All other power (COMPxHYST[1:0]=01) 10 modes Comparator hysteresis High speed mode 26 mV  $V_{hys}$ Medium hysteresis 15 All other power (COMPxHYST[1:0]=10) 19 modes High speed mode 18 49 High hysteresis 31 All other power (COMPxHYST[1:0]=11) 19 40 modes

Table 76. Comparator characteristics<sup>(1)</sup> (continued)

<sup>2.</sup> For more details and conditions, see Figure 37 Maximum  $V_{REFINT}$  scaler startup time from power down.



<sup>1.</sup> Data guaranteed by design.

## 6.3.21 Operational amplifier characteristics

Table 77. Operational amplifier characteristics<sup>(1)</sup>

Symbol	Param	neter	Condition	Min	Тур	Max	Unit
$V_{DDA}$	Analog supply volt	age	-	2.4	-	3.6	V
CMIR	Common mode in	out range	-	0	-	$V_{DDA}$	V
		Maximum calibration	25°C, No Load on output.	-	-	4	
VI <sub>OFFSET</sub>	Input offset	range	All voltage/Temp.	-	-	6	- mV
VIOFFSET	voltage	After offset	25°C, No Load on output.	-	-	1.6	IIIV
	calibrati	calibration	All voltage/Temp.	-	-	3	
ΔVI <sub>OFFSET</sub>	Input offset voltage	Input offset voltage drift		-	5	-	μV/°C
I <sub>LOAD</sub>	Drive current		-	-	-	500	μA
IDDOPAMP	Consumption	Consumption		-	690	1450	μA
TS_OPAMP_VOUT	ADC sampling time when reading the OPAMP output.		-	400	-	-	ns
CMRR	Common mode rejection ratio		-	-	90	-	dB
PSRR	Power supply rejection ratio		DC	73	117	-	dB
GBW	Bandwidth		-	-	8.2	-	MHz
SR	Slew rate		-	-	4.7	-	V/µs
R <sub>LOAD</sub>	Resistive load		-	4	-	-	kΩ
C <sub>LOAD</sub>	Capacitive load		-	-	-	50	pF
VOH <sub>SAT</sub>	High saturation vo	Itage(2)	R <sub>load</sub> = min, Input at V <sub>DDA</sub> .	V <sub>DDA</sub> -100	-	-	
VOITSAT	Tilgii saturation vo	itage	R <sub>load</sub> = 20K, Input at V <sub>DDA</sub> .	V <sub>DDA</sub> -20	-	-	- mV
VOL	High saturation vo	Itago(2)	Rload = min, input at 0V	-	-	100	IIIV
VOL <sub>SAT</sub>	Tilgii saturation vo	itage	Rload = 20K, input at 0V.	-	-	20	
φm	Phase margin		-	-	62	-	0
t <sub>OFFTRIM</sub>	Offset trim time: during calibration, minimum time needed between two steps to have 1 mV accuracy		-	-	-	2	ms
t <sub>WAKEUP</sub>	Wake up time from	n OFF state.	$C_{LOAD} \le 50 \text{ pf},$ $R_{LOAD} \ge 4 \text{ k}\Omega,$ Follower configuration	-	2.8	5	μs



Table 77. Operational amplifier characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Condition	Min	Тур	Max	Unit
			-	2	-	-
DCA gain	Non inverting gain value		-	4	-	-
PGA gain	Non inverting gain value	-	-	8	-	-
			-	16	-	-
		Gain=2	ı	5.4/5.4	i	
D	R2/R1 internal resistance values in	Gain=4	ı	16.2/5.4	i	kΩ
R <sub>network</sub>	PGA mode <sup>(3)</sup>	Gain=8	ı	37.8/5.4	i	NS2
		Gain=16	ı	40.5/2.7	i	
PGA gain error	PGA gain error	-	-1%	-	1%	
I <sub>bias</sub>	OPAMP input bias current	-	-	-	±0.2 <sup>(4)</sup>	μΑ
		PGA Gain = 2, Cload = 50pF, Rload = 4 K $\Omega$	-	4	-	
PGA BW	PGA bandwidth for different non inverting gain	PGA Gain = 4, Cload = 50pF, Rload = 4 K $\Omega$	-	2	-	MHz
FGA BW		PGA Gain = 8, Cload = 50pF, Rload = 4 KΩ	-	1	-	IVITIZ
		PGA Gain = 16, Cload = 50pF, Rload = 4 KΩ	-	0.5	-	
		@ 1KHz, Output loaded with 4 KΩ	-	109	-	
en	Voltage noise density	@ 10KHz, Output loaded with 4 KΩ	-	43	-	<u>nV</u> √Hz

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> The saturation voltage can be also limited by the Iload (drive current).

<sup>3.</sup> R2 is the internal resistance between OPAMP output and OPAMP inverting input. R1 is the internal resistance between OPAMP inverting input and ground. The PGA gain =1+R2/R1

<sup>4.</sup> Mostly TTa I/O leakage, when used in analog mode.

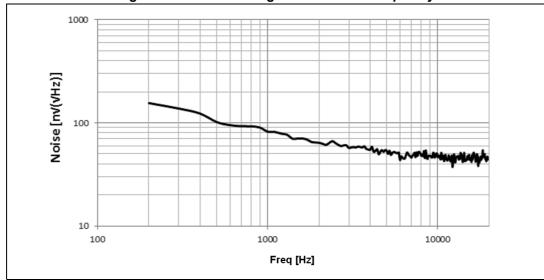


Figure 38. OPAMP voltage noise versus frequency

## 6.3.22 Temperature sensor characteristics

Table 78. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>SENSE</sub> linearity with temperature	-	±1	<u>+2</u>	°C
Avg_Slope <sup>(1)</sup>	Average slope	4.0	4.3	4.6	mV/°C
V <sub>25</sub>	Voltage at 25 °C	1.34	1.43	1.52	V
t <sub>START</sub> (1)	Startup time	4	-	10	μs
T <sub>S_temp</sub> <sup>(1)(2)</sup>	ADC sampling time when reading the temperature	2.2	-	-	μs

<sup>1.</sup> Guaranteed by design.

Table 79. Temperature sensor calibration values

Calibration value name	Description	Memory address
TS_CAL1	TS ADC raw data acquired at temperature of 30 °C, V <sub>DDA</sub> = 3.3 V	0x1FFF F7B8 - 0x1FFF F7B9
TS_CAL2	TS ADC raw data acquired at temperature of 110 °C V <sub>DDA</sub> = 3.3 V	0x1FFF F7C2 - 0x1FFF F7C3



 $<sup>2. \</sup>quad \text{Shortest sampling time can be determined in the application by multiple iterations}.$ 

## 6.3.23 V<sub>BAT</sub> monitoring characteristics

Table 80. V<sub>BAT</sub> monitoring characteristics

Symbol	Parameter	Min	Тур	Max	Unit
R	Resistor bridge for V <sub>BAT</sub>	-	50	-	ΚΩ
Q	Ratio on V <sub>BAT</sub> measurement	-	2	-	
Er <sup>(1)</sup>	Error on Q	-1	-	+1	%
T <sub>S_vbat</sub> <sup>(1)(2)</sup>	ADC sampling time when reading the V <sub>BAT</sub> 1mV accuracy	2.2	-	-	μs

<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> Shortest sampling time can be determined in the application by multiple iterations.

#### **Package information** 7

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com. ECOPACK® is an ST trademark.

### LQFP100 - 14 x 14 mm, low-profile quad flat package 7.1 information

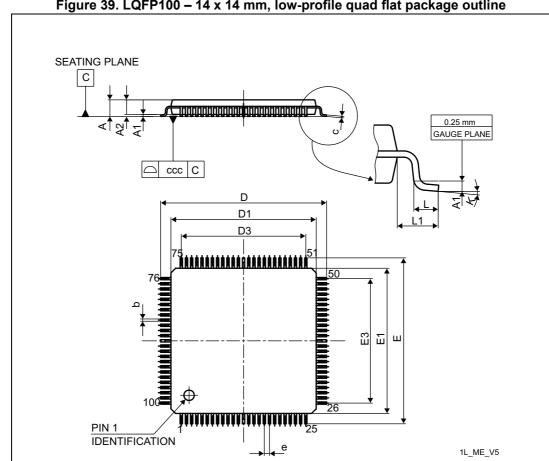


Figure 39. LQFP100 - 14 x 14 mm, low-profile quad flat package outline

1. Drawing is not to scale.

Table 81. LQPF100 - 14 x 14 mm, low-profile quad flat package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>			
Symbol	Min	Тур	Max	Min	Тур	Max	
Α	-	-	1.60	-	-	0.063	
A1	0.05	-	0.15	0.002	-	0.0059	

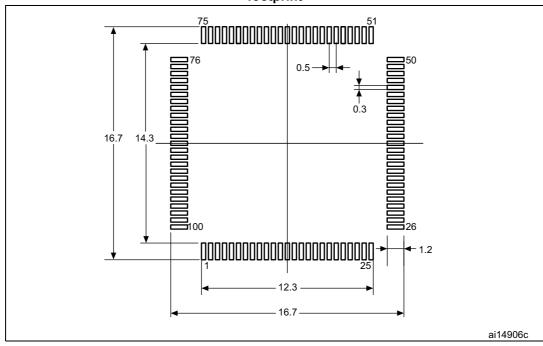


Table 81. LQPF100 – 14 x 14 mm, low-profile quad flat package mechanical data (continued)

Complete		millimeters			inches <sup>(1)</sup>	
Symbol	Min	Тур	Max	Min	Тур	Max
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571
b	0.17	0.22	0.27	0.0067	0.0087	0.0106
С	0.09	-	0.2	0.0035	-	0.0079
D	15.80	16.00	16.2	0.622	0.6299	0.6378
D1	13.80	14.00	14.2	0.5433	0.5512	0.5591
D3	-	12.00	-	-	0.4724	-
E	15.80	16.00	16.2	0.622	0.6299	0.6378
E1	13.80	14.00	14.2	0.5433	0.5512	0.5591
E3	-	12.00	-	-	0.4724	-
е	-	0.50	-	-	0.0197	-
L	0.45	0.60	0.75	0.0177	0.0236	0.0295
L1	-	1.00	-	-	0.0394	-
K	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.08	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 40. LQFP100 – 14 x 14 mm, low-profile quad flat package recommended footprint



1. Dimensions are in millimeters.

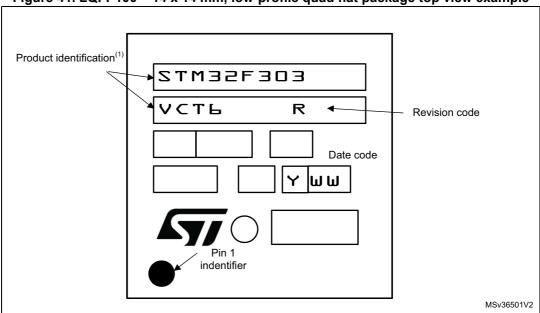
## LQFP100 device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 41. LQFP100 - 14 x 14 mm, low-profile quad flat package top view example



Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

# 7.2 LQFP64 – 10 x 10 mm, low-profile quad flat package information

Figure 42. LQFP64 - 10 x 10 mm, low-profile quad flat package outline

1. Drawing is not to scale.

Table 82. LQFP64 – 10 x 10 mm, low-profile quad flat package mechanical data

Symbol		millimeters		inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.60	-	-	0.0630
A1	0.05	-	0.15	0.0020	-	0.0059
A2	1.350	1.40	1.45	0.0531	0.0551	0.0571
b	0.17	0.22	0.27	0.0067	0.0087	0.0106
С	0.09	-	0.20	0.0035		0.0079
D	-	12.00	-	-	0.4724	-
D1	-	10.00	-	-	0.3937	-
D3	-	7.50	-	-	0.2953	-
Е	-	12.00	-	-	0.4724	-

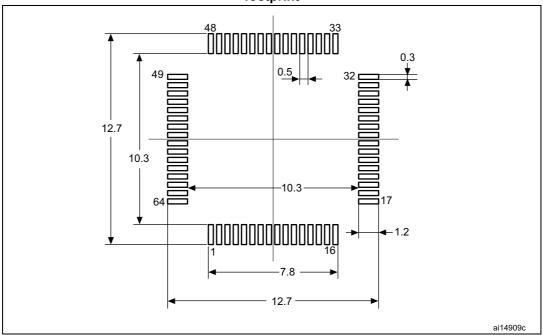
(continued)

Table 82. LQFP64 – 10 x 10 mm, low-profile quad flat package mechanical data

Symbol		millimeters		inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max
E1	-	10.00	-	-	0.3937	-
E3	-	7.50	-	-	0.2953	-
е	-	0.50	-	-	0.0197	-
K	0°	3.5°	7°	0°	3.5°	7°
L	0.45	0.60	0.75	0.0177	0.0236	0.0295
L1	-	1.00	-	-	0.0394	-
ccc	-	-	0.08	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 43. LQFP64 - 10 x 10 mm, low-profile quad flat package recommended footprint



1. Dimensions are in millimeters.

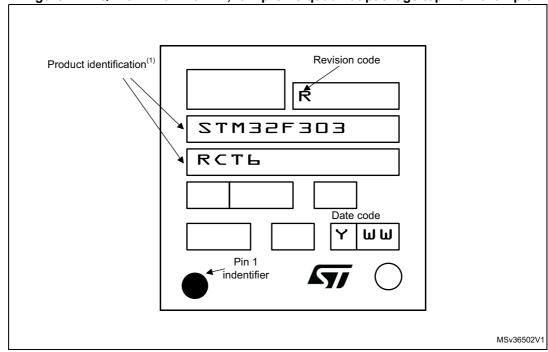
## LQFP64 device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 44. LQFP64 - 10 x 10 mm, low-profile quad flat package top view example



Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified
and therefore not approved for use in production. ST is not responsible for any consequences resulting
from such use. In no event will ST be liable for the customer using any of these engineering samples in
production. ST's Quality department must be contacted prior to any decision to use these engineering
samples to run a qualification activity.



# 7.3 LQFP48 – 7 x 7 mm, low-profile quad flat package information

SEATING PLANE С 0.25 mm GAUGE PLANE □ ccc C D Ā D1 D3 E3 핀 ш ш **IDENTIFICATION** е 5B\_ME\_V2

Figure 45. LQFP48 - 7 x 7 mm, low-profile quad flat package outline

1. Drawing is not to scale.

Table 83. LQFP48 – 7 x 7 mm, low-profile quad flat package mechanical data

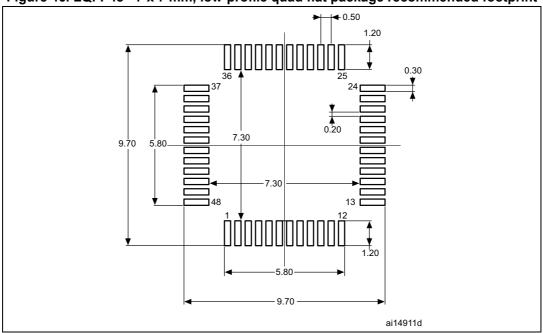
	*****							
Symbol		millimeters		inches <sup>(1)</sup>				
Symbol	Min	Тур	Max	Min	Тур	Max		
Α	-	-	1.60	-	-	0.0630		
A1	0.05	-	0.15	0.0020	-	0.0059		
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571		
b	0.17	0.22	0.27	0.0067	0.0087	0.0106		
С	0.09	-	0.20	0.0035	-	0.0079		
D	8.80	9.00	9.20	0.3465	0.3543	0.3622		
D1	6.80	7.00	7.20	0.2677	0.2756	0.2835		
D3	-	5.50	-	-	0.2165	-		
E	8.80	9.00	9.20	0.3465	0.3543	0.3622		

Table 83. LQFP48 – 7 x 7 mm, low-profile quad flat package mechanical data (continued)

Symbol		millimeters		inches <sup>(1)</sup>			
	Min	Тур	Max	Min	Тур	Max	
E1	6.80	7.00	7.20	0.2677	0.2756	0.2835	
E3	-	5.50	-	-	0.2165	-	
е	-	0.50	-	-	0.0197	-	
L	0.45	0.60	0.75	0.0177	0.0236	0.0295	
L1	-	1.00	-	-	0.0394	-	
K	0°	3.5°	7°	0°	3.5°	7°	
ccc	-	-	0.08	-	-	0.0031	

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 46. LQFP48 - 7 x 7 mm, low-profile quad flat package recommended footprint



1. Dimensions are in millimeters.

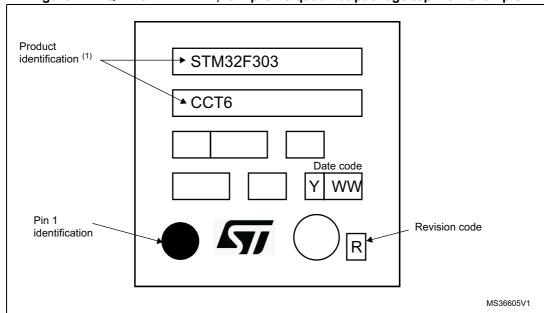
## LQFP48 device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

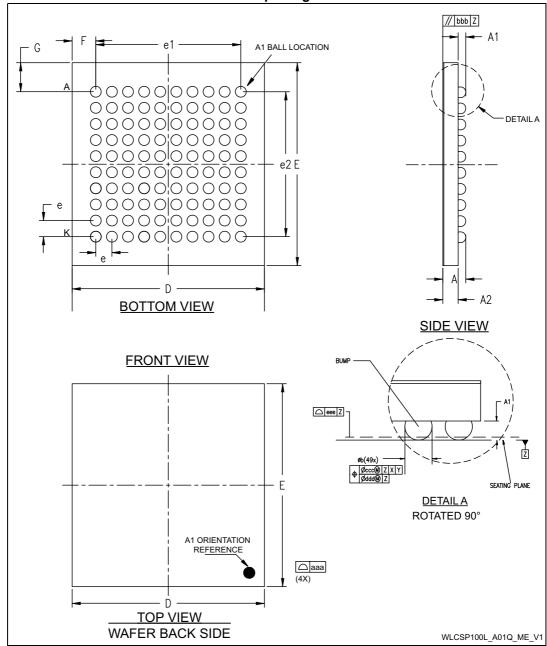
Figure 47. LQFP48 - 7 x 7 mm, low-profile quad flat package top view example



<sup>1.</sup> Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

# 7.4 WLCSP100 - 0.4 mm pitch wafer level chip scale package information

Figure 48. WLCSP100 – 100L, 4.166 x 4.628 mm 0.4 mm pitch wafer level chip scale package outline



1. Drawing is not to scale.

Table 84. WLCSP100 – 100L, 4.166 x 4.628 mm 0.4 mm pitch wafer level chip scale package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Тур	Min	Max
Α	0.525	0.555	0.585	0.0207	0.0219	0.0230
A1	-	0.17	-	-	0.0067	-
A2	-	0.38	-	-	0.0150	-
A3 <sup>(2)</sup>	-	0.025	-	-	0.0010	-
Ø b <sup>(3)</sup>	0.22	0.25	0.28	-	0.0098	0.0110
D	4.166	4.201	4.236	-	0.1654	0.1668
E	4.628	4.663	4.698	-	0.1836	0.1850
е	-	0.4	-	-	0.0157	-
e1	-	3.6	-	-	0.1417	-
e2	-	3.6	-	-	0.1417	-
F	-	0.3005	-	-	0.0118	-
G	-	0.5315	-	-	0.0209	-
N	-	100	-	-	3.9370	-
aaa	-	0.1	-	-	0.0039	-
bbb	-	0.1	-	-	0.0039	-
ccc	-	0.1	-	-	0.0039	-
ddd	-	0.05	-	-	0.0020	-
eee	-	0.05	-	-	0.0020	-

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

<sup>2.</sup> Back side coating.

<sup>3.</sup> Dimension is measured at the maximum bump diameter parallel to primary datum  ${\sf Z}$ .

Figure 49. WLCSP100 – 100L, 4.166 x 4.628 mm 0.4 mm pitch wafer level chip scale package recommended footprint

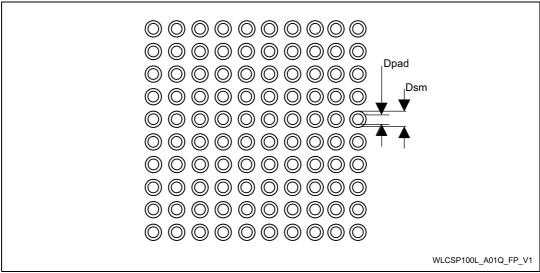


Table 85. WLCSP100 recommended PCB design rules (0.4 mm pitch)

Dimension	Recommended values
Pitch	0.4 mm
Dpad	0.225 mm
Dsm	0.290 mm
Stencil thickness	0.1 mm

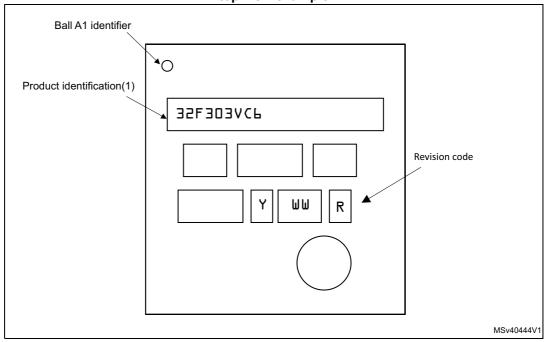
## WLCSP100 device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 50. WLCSP100, 0.4 mm pitch wafer level chip scale package top view example



Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified
and therefore not approved for use in production. ST is not responsible for any consequences resulting
from such use. In no event will ST be liable for the customer using any of these engineering samples in
production. ST's Quality department must be contacted prior to any decision to use these engineering
samples to run a qualification activity.

### 7.5 Thermal characteristics

The maximum chip junction temperature (T<sub>J</sub>max) must never exceed the values given in *Table 24: General operating conditions on page 60.* 

The maximum chip-junction temperature,  $T_J$  max, in degrees Celsius, may be calculated using the following equation:

$$T_J \max = T_A \max + (P_D \max x \Theta_{JA})$$

#### Where:

- T<sub>A</sub> max is the maximum ambient temperature in °C,
- Θ<sub>JA</sub> is the package junction-to-ambient thermal resistance, in °C/W,
- P<sub>D</sub> max is the sum of P<sub>INT</sub> max and P<sub>I/O</sub> max (P<sub>D</sub> max = P<sub>INT</sub> max + P<sub>I/O</sub>max),
- P<sub>INT</sub> max is the product of I<sub>DD</sub> and V<sub>DD</sub>, expressed in Watts. This is the maximum chip internal power.

P<sub>I/O</sub> max represents the maximum power dissipation on output pins where:

$$P_{I/O} \max = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual  $V_{OL}$  /  $I_{OL}$  and  $V_{OH}$  /  $I_{OH}$  of the I/Os at low and high level in the application.

Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient LQFP64 - 10 × 10 mm / 0.5 mm pitch	45	
$\Theta_{JA}$	Thermal resistance junction-ambient LQFP48 - 7 × 7 mm	55	°C/W
	Thermal resistance junction-ambient LQFP100 - 14 × 14 mm / 0.5 mm pitch	41	
	Thermal resistance junction-ambient WLCSP100 - 0.4 mm pitch	40	

Table 86. Package thermal characteristics

### 7.5.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org

## 7.5.2 Selecting the product temperature range

When ordering the microcontroller, the temperature range is specified in the ordering information scheme shown in *Section 8: Ordering information*.

Each temperature range suffix corresponds to a specific guaranteed ambient temperature at maximum dissipation and, to a specific maximum junction temperature.

As applications do not commonly use the STM32F303xB/STM32F303xC at maximum dissipation, it is useful to calculate the exact power consumption and junction temperature to determine which temperature range will be best suited to the application.

The following examples show how to calculate the temperature range needed for a given application.

### **Example 1: High-performance application**

Assuming the following application conditions:

Maximum ambient temperature  $T_{Amax}$  = 82 °C (measured according to JESD51-2),  $I_{DDmax}$  = 50 mA,  $V_{DD}$  = 3.5 V, maximum 3 I/Os used at the same time in output at low level with  $I_{OL}$  = 8 mA,  $V_{OL}$ = 0.4 V and maximum 2 I/Os used at the same time in output at low level with  $I_{OL}$  = 20 mA,  $V_{OL}$ = 1.3 V

 $P_{INTmax}$  = 50 mA × 3.5 V= 175 mW

 $P_{IOmax} = 3 \times 8 \text{ mA} \times 0.4 \text{ V} + 2 \times 20 \text{ mA} \times 1.3 \text{ V} = 61.6 \text{ mW}$ 

This gives: P<sub>INTmax</sub> = 175 mW and P<sub>IOmax</sub> = 61.6 mW:

 $P_{Dmax} = 175 + 61.6 = 236.6 \text{ mW}$ 

Thus:  $P_{Dmax} = 236.6 \text{ mW}$ 

Using the values obtained in *Table 86* T<sub>Jmax</sub> is calculated as follows:

For LQFP64, 45°C/W

 $T_{Jmax}$  = 82 °C + (45°C/W × 236.6 mW) = 82 °C + 10.65 °C = 92.65 °C

This is within the range of the suffix 6 version parts ( $-40 < T_{.l} < 105$  °C).

In this case, parts must be ordered at least with the temperature range suffix 6 (see Section 8: Ordering information).

### **Example 2: High-temperature application**

Using the same rules, it is possible to address applications that run at high ambient temperatures with a low dissipation, as long as junction temperature  $T_J$  remains within the specified range.

Assuming the following application conditions:

Maximum ambient temperature  $T_{Amax}$  = 115 °C (measured according to JESD51-2),  $I_{DDmax}$  = 20 mA,  $V_{DD}$  = 3.5 V, maximum 9 I/Os used at the same time in output at low level with  $I_{OL}$  = 8 mA,  $V_{OL}$ = 0.4 V

 $P_{INTmax}$  = 20 mA × 3.5 V= 70 mW

 $P_{IOmax} = 9 \times 8 \text{ mA} \times 0.4 \text{ V} = 28.8 \text{ mW}$ 

This gives:  $P_{INTmax} = 70 \text{ mW}$  and  $P_{IOmax} = 28.8 \text{ mW}$ :

 $P_{Dmax} = 70 + 28.8 = 98.8 \text{ mW}$ 

Thus: P<sub>Dmax</sub> = 98.8 mW

Using the values obtained in  $\it Table~86~T_{\it Jmax}$  is calculated as follows:

For LQFP100, 41°C/W

 $T_{Jmax}$  = 115 °C + (41°C/W × 98.8 mW) = 115 °C + 4.05 °C = 119.05 °C

This is within the range of the suffix 7 version parts ( $-40 < T_J < 125$  °C).

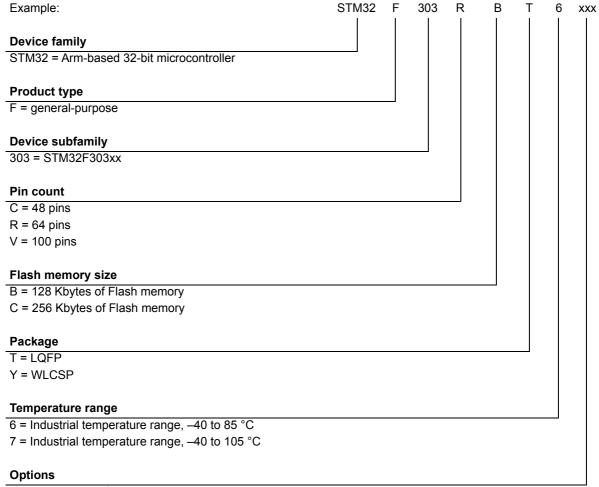
In this case, parts must be ordered at least with the temperature range suffix 7 (see Section 8: Ordering information).



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# 8 Ordering information

Table 87. Ordering information scheme



xxx = programmed parts

TR = tape and reel

For a list of available options (speed, package, etc.) or for further information on any aspect of this device, please contact your nearest ST sales office.

# 9 Revision history

Table 88. Document revision history

Date	Revision	Changes
22-Jun-2012	1	Initial release
07-Sep-2012	2	Modified Features on cover page.  Modified Table 2: STM32F301xx family device features and peripheral counts  Added clock tree to Section 3.9: Clocks and startup  Added Table 10: STM32F302xB/STM32F302xC 12C implementation  Added Table 11: USART features  Added Table 12: STM32F302xB/STM32F302xC SPI/12S implementation  Modified Table 13: Capacitive sensing GPIOs available on  STM32F302xB/STM32F302xC devices  Modified Figure 7, Figure 8 and Figure 9: STM32F302xB/STM32F302xC  LQFP100 pinout  Modified Table 16: STM32F302xB/STM32F302xC pin definitions  Modified Table 16: STM32F302xB/STM32F302xC pin definitions  Modified Table 21: Voltage characteristics  Modified Table 25: Operating conditions at power-up / power-down  Added footnote to Table 31: Typical and maximum current consumption from the V <sub>DDA</sub> supply  Added footnote to Table 35 and Table 36: Typical current consumption in  Sleep mode, code running from Flash or RAM  Removed table "Switching output I/O current consumption" and table "Peripheral current consumption"  Added note under Figure 17: Typical application with a 32.768 kHz  crystal  Updated Table 44: HSI oscillator characteristics  Updated Wakeup time from low-power mode and Table 39: Low-power mode wakeup timings  Updated Table 52: Electrical sensitivities  Updated Table 53: I/O current injection susceptibility  Updated Table 55: Output voltage characteristics  Updated Table 55: Output voltage characteristics  Updated Table 55: Output voltage characteristics  Updated Table 64: I <sup>2</sup> S characteristics  Updated Table 63: SPI characteristics  Updated Table 64: I <sup>2</sup> S characteristics  Corrected LQFP100 in Section 7.2.3: Selecting the product temperature range
21-Sep-2012	3	Updated Table 63: SPI characteristics



Table 88. Document revision history (continued)

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Date F
Date   F



Table 88. Document revision history (continued)

Date	Revision	Changes
08-Jan-2013	5	Updated V <sub>hys</sub> and I <sub>lkg</sub> in <i>Table 54: I/O static characteristics</i> .  Updated V <sub>IL(NRST)</sub> , V <sub>IH(NRST)</sub> , and V <sub>NF(NRST)</sub> in <i>Table 57: NRST pin characteristics</i> .  Updated <i>Table 70: ADC accuracy - limited test conditions, 100-pin packages</i> and <i>Table 64: ADC accuracy - limited test conditions 2</i> ).
24-Jun-2013	6	Replaced Cortex-M4F with Cortex M4 with FPU Updated Core, Memories and SPI bullet points in Features Removed 8KB CCM SRAM from STM32F302xx devices, updated Figure 2: STM32F303xB/STM32F303xC block diagram and Table 3: STM32F303xx family device features and peripheral counts Updated Section 3.4: Embedded SRAM Added VREF+ in Section 3.14: Digital-to-analog converter (DAC) Removed DMA support for UART5 in Table 11: USART features Added 'reference clock detection' bullet in Section 3.18: Real-time clock (RTC) and backup registers Added paragraph 'The touch sensing controller is fully' in Section 3.26: Touch sensing controller (TSC) Updated Comparison of I2C analog and digital filters Updated Section 3.10: General-purpose input/outputs (GPIOs) Added 'EVENTOUT' in Table 16: STM32F302xB/STM32F302xC pin definitions and added note to 'VREF+' pin Updated ΣI <sub>VDD</sub> in Table 22: Current characteristics and Output driving current Updated Table 61: I2C timings specification (see I2C specification, rev.03, June 2007) and Figure 25: I <sup>2</sup> C bus AC waveforms and measurement circuit Added VREF+ row to Table 68: ADC characteristics, replaced VDDA with VREF+, updated t <sub>conv</sub> and added note to 'conversion voltage range Added VREF+ row to Table 75: DAC characteristics and replaced VDDA with VREF+ Added 'PGA BW' and 'en' in Table 77: Operational amplifier characteristics
13-Nov-2013	7	Removed STM32F302xB/STM32F302xC products (now in a separate datasheet).  Added I2S feature for SPI2 and SPI3  Added t <sub>SP</sub> to <i>Table 61: I2C timings specification (see I2C specification, rev.03, June 2007).</i> Renamed t <sub>SP</sub> to t <sub>AN</sub> in <i>Table 62: I2C analog filter characteristics.</i> Added t <sub>STAB</sub> in <i>Table 68: ADC characteristics</i> Renamed V <sub>OPAMPx</sub> to <sub>VREFOPAMPx</sub> Updated <i>Table 71: ADC accuracy, 100-pin packages.</i> Updated ADC channel names in <i>Section 3.13.1</i> , <i>Section 3.13.2</i> and <i>Section 3.13.3</i> .



Table 88. Document revision history (continued)

Date	Revision	Changes		
Date	1/64191011			
		Updated <i>Table 50: EMI characteristics</i> conditions :3.3v replaced by 3.6V. Updated <i>Section 6.3.17: Communications interfaces</i> I <sup>2</sup> C interface.		
		Updated <i>Table 77: Operational amplifier characteristics</i> adding TS_OPAMP_VOUT row.		
		Updated Section 3.13: Fast analog-to-digital converter (ADC).		
		updated Arm and Cortex trademark.		
	8	Updated <i>Table 32: Typical and maximum</i> $V_{DD}$ <i>consumption in Stop and Standby modes</i> with Max value at 85°C and 105°C.		
18-Apr-2014		Updated <i>Table 70: ADC accuracy - limited test conditions, 100-pin packages</i> and <i>Table 71: ADC accuracy, 100-pin packages</i> for 100-pin package.		
		Added Table 72: ADC accuracy - limited test conditions, 64-pin		
		packages and Table 73: ADC accuracy, 64-pin packagesfor 64-pin package.		
		Added <i>Table 74: ADC accuracy at 1MSPS</i> for 1MSPS sampling frequency.		
		Updated Table 63: SPI characteristics.		
		Updated Table 75: DAC characteristics.		
	9	Updated core description in cover page.		
		Updated HSI characteristics <i>Table 44: HSI oscillator characteristics</i> and <i>Figure 18: HSI oscillator accuracy characterization results for soldered parts.</i>		
		Updated Table 58: TIMx characteristics.		
		Updated <i>Table 16:</i> STM32F302xB/STM32F302xC pin definitions adding note for I/Os featuring an analog output function (DAC_OUT,OPAMP_OUT).		
		Updated <i>Table 68: ADC characteristics</i> adding IDDA & IREF consumptions.		
09-Dec-2014		Added Figure 32: ADC typical current consumption on VDDA pin and Figure 33: ADC typical current consumption on VREF+ pin.		
09-Dec-2014		Added Section 3.8: Interconnect matrix.		
		Updated Figure 5: Clock tree.		
		Added note after Table 32: Typical and maximum $V_{\rm DD}$ consumption in Stop and Standby modes.		
		Updated Section: In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK <sup>®</sup> packages, depending on their level of environmental compliance. ECOPACK <sup>®</sup> specifications, grade definitions and product status are available at: www.st.com. ECOPACK <sup>®</sup> is an ST trademark. with new LQFP100, LQFP64, LQFP48 package marking.		
		Updated <i>Table 16: STM32F302xB/STM32F302xC pin definitions</i> and alternate functions tables replacing usart_rts by usart_rts_de.		
29-Jan-2015	10	Updated Section 6.3.20: Comparator characteristics modifying ts_sc characteristics in Table 76 and adding Figure 37: Maximum VREFINT scaler startup time from power down.		
		Updated I <sub>DD</sub> data in <i>Table 42: HSE oscillator characteristics</i> .		



Table 88. Document revision history (continued)

Date	Revision	Changes
17-Apr-2015	11	Updated Section 7: Package information: with new package information structure adding 1 sub paragraph for each package.  Updated Figure 41: LQFP100 – 14 x 14 mm, low-profile quad flat package top view example removing gate mark.  Added note for all packages about the device marking orientation: "the following figure gives an example of topside marking orientation versus pin 1 identifier location".  Updated Table 82: LQFP64 – 10 x 10 mm, low-profile quad flat package mechanical data.
11-Dec-2015	12	Added WLCSP100:  - Updated cover page.  - Updated Table 2: STM32F303xB/STM32F303xC family device features and peripheral counts.  - Added Figure 7: STM32F303xB/STM32F303xC WLCSP100 pinout.  - Updated Table 13: STM32F303xB/STM32F303xC pin definitions.  - Updated Table 24: General operating conditions.  - Added Section 7.4: WLCSP100 - 0.4 mm pitch wafer level chip scale package information.  - Updated Table 86: Package thermal characteristics.  - Updated Table 87: Ordering information scheme.  Updated Figure 4, Figure 5, Figure 6, Table 13 and Table 22 removing all VDD and VSS indexes.  Updated all the notes removing 'not tested in production'.  Updated Table 68: ADC characteristics adding V <sub>REF</sub> negative voltage reference.  Update Table 21: Voltage characteristics adding table note 4.



Table 88. Document revision history (continued)

Date	Revision	Changes
06-May-2016	13	Updated <i>Table 43: LSE oscillator characteristics (f<sub>LSE</sub> = 32.768 kHz)</i> LSEDRV[1:0] bits. Updated <i>Table 28: Embedded internal reference voltage</i> V <sub>REFINT</sub> internal reference voltage (min and typ values). Updated <i>Figure 5: STM32F303xB/STM32F303xC LQFP64 pinout</i> replacing VSS by PF4. Updated <i>Table 51: ESD absolute maximum ratings</i> ESD CDM at class 3 and 4 including WLCSP100 package information. Updated <i>Table 13: STM32F303xB/STM32F303xC pin definitions:</i> – Adding 'digital power supply' in the Pin function column at the line corresponding to K8/28/19 pins.  – Adding VSS digital ground line with WLCSP100 K9 and K10 pins connected.  – Replacing in VDD line for WLCSP100: 'A10, B10' by 'A9, A10, B10, B8'. Updated <i>Figure 21: Five volt tolerant (FT and FTf) I/O input characteristics - CMOS port.</i> Updated <i>Table 77: Operational amplifier characteristics</i> high saturation and low saturation voltages. Updated <i>Table 13: STM32F303xB/STM32F303xC pin definitions</i> adding note 'Fast ADC channel' for ADCx_IN15. Updated <i>Table 75: DAC characteristics</i> resistive load. Updated <i>Table 68: ADC characteristics</i> adding CMIR parameter and modifying tSTAB parameter characteristics.
30-Oct-2018	14	Updated Table 51: ESD absolute maximum ratings ESD class. Updated cover on 2 pages. Updated Section 1: Introduction with Arm logo. Updated Section 7: Package information adding information:  — Other optional marking or inset/upset marks.  — The printed markings may differ depending on the supply chain.  — Updated note 1 below all the package device marking figures.



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