

# The Chip Designer's **Guide** to the **Galaxy**

Chronicles of **RTL Engineering**: From University Theory  
to Professional SystemVerilog Workflow



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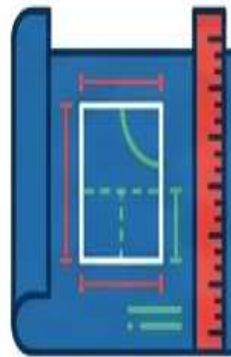
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# The Mission

To demystify the transition to professional RTL design by building a functional library of digital modules, emphasizing architecture, hardware-centric coding, and rigorous verification.



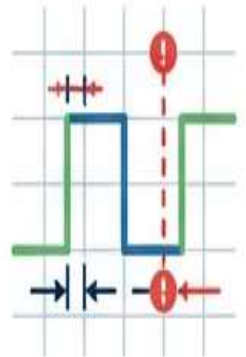
Architecture  
Before RTL



Hardware-Centric  
Coding

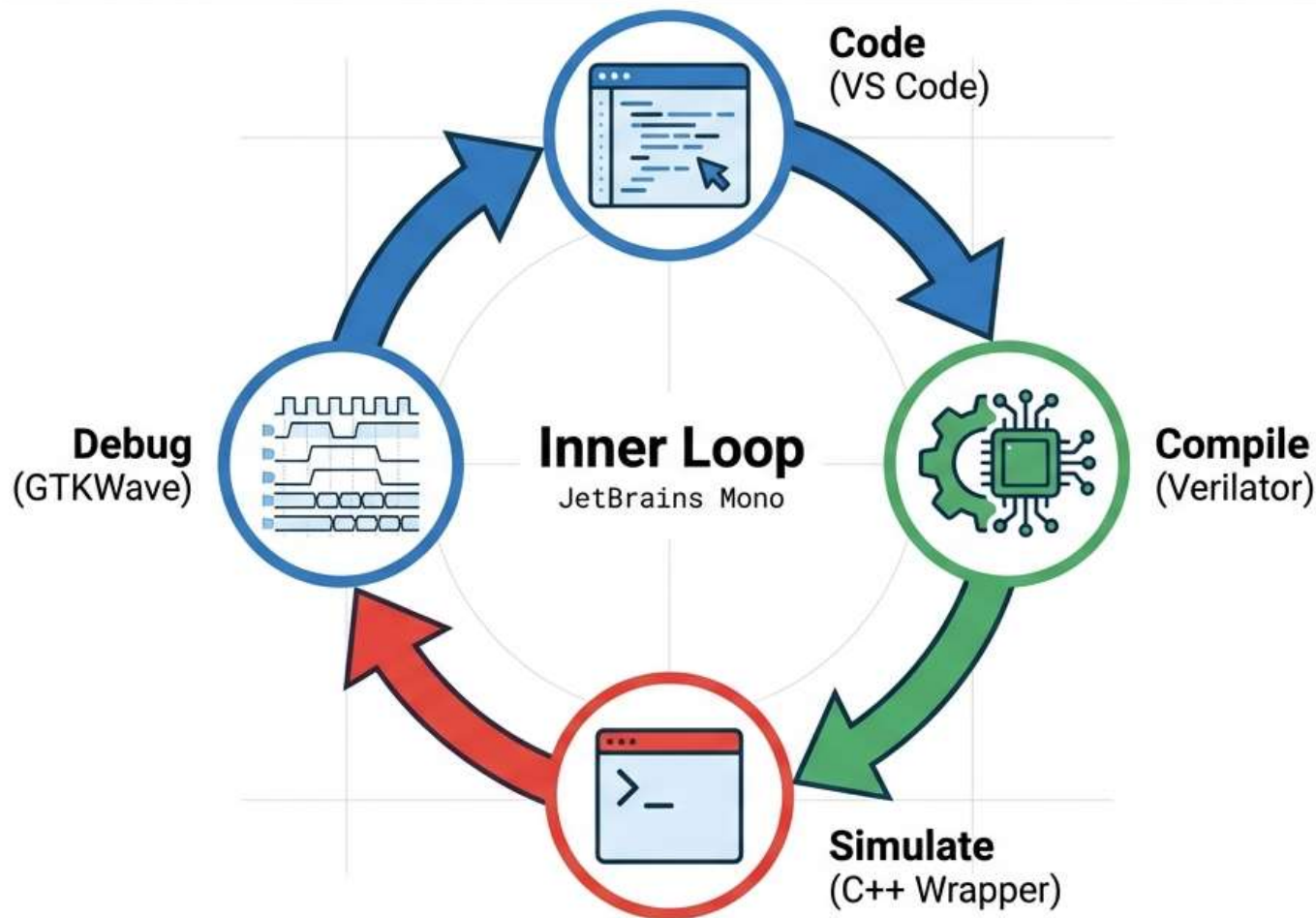


High-Level  
Verification



Waveform  
Debugging

# Constructing the Laboratory: The Toolchain



## The Stack

### OS

**Ubuntu via WSL2 (Windows Subsystem for Linux)**  
Industry standard environment.

### The Compiler

#### Verilator

Converts SystemVerilog to optimized C++ models.  
Speed + Strict Linting.

### The Eyes

#### GTKWave

Visualization interface for debugging logic and timing.

### The Engine

#### Make

Automates build dependencies with a single command.

# From Behavioral Equations to Structural Architecture

## The 1-Bit Full Adder

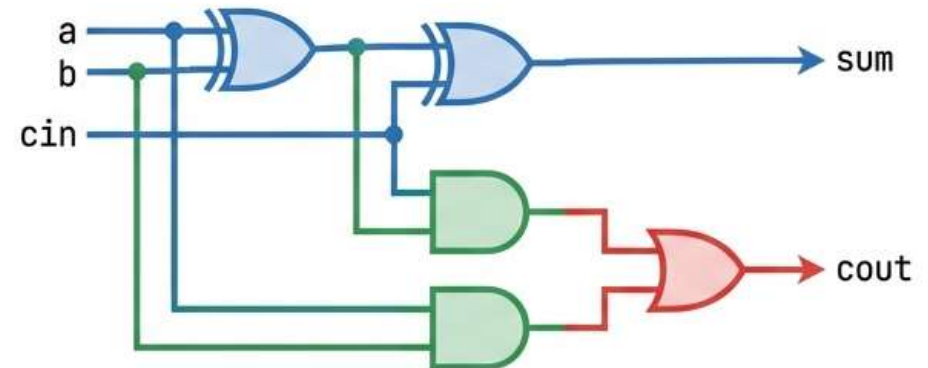
### The Theory (Behavioral)

$$\text{sum} = a \oplus b \oplus \text{cin}$$

$$\text{cout} = (a \& b) \mid (\text{cin} \& (a \oplus b))$$

⇒ Simulation Delay: assign #1 represents gate propagation time (Ripple Effect).

### The Architecture (Structural)

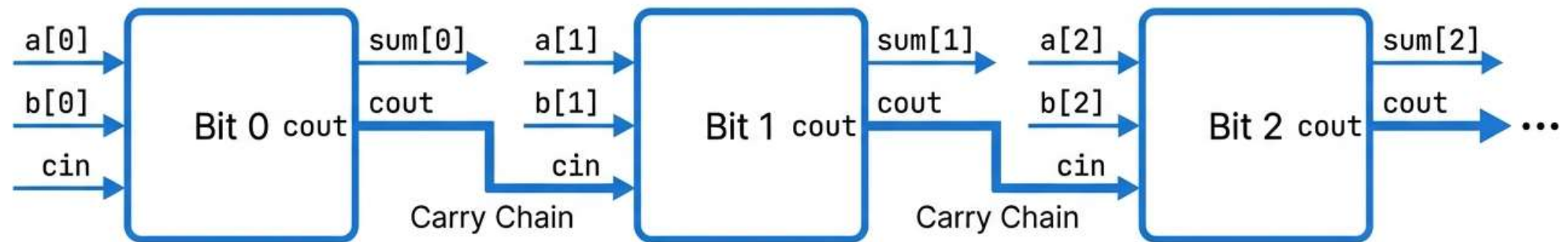


```
module adder_1bit (  
    input logic a, b, cin,  
    output logic sum, cout  
);  
    assign #1 sum = a ^ b ^ cin;  
    assign #1 cout = (a & b) | (cin & (a ^ b));  
endmodule
```



# Scaling Complexity: The Parameterizable Ripple Carry Adder

Using Generators to create scalable hardware IP.



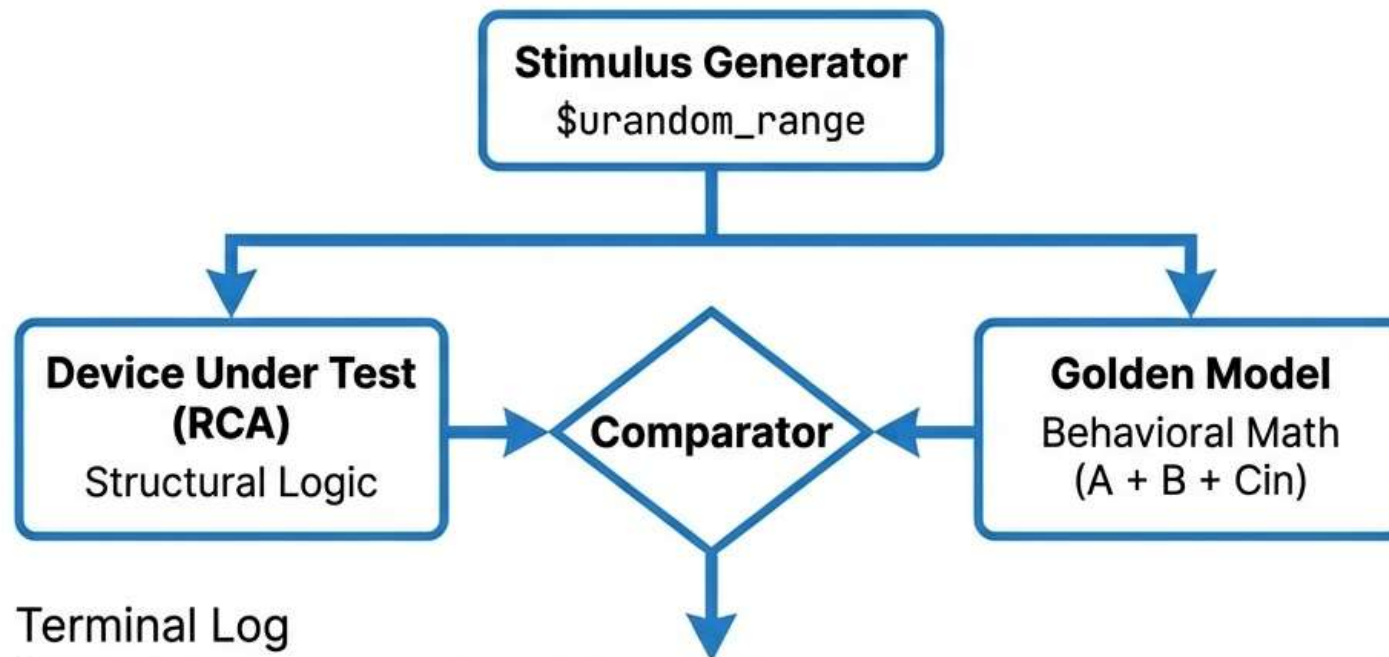
```
module rca #(parameter WIDTH = 8) (...);  
    genvar i;  
    generate  
        for (i = 0; i < WIDTH; i++) begin : adder_loop  
            adder_1bit u_adder (  
                .a(a[i]), .b(b[i]), .cin(carry_chain[i]),  
                .sum(sum[i]), .cout(carry_chain[i+1])  
            );  
        end  
    endgenerate  
endmodule
```

Compile-time iteration variable

User-definable bit width

# Trust but Verify: Automation and the Golden Model

## Self-Checking Testbench



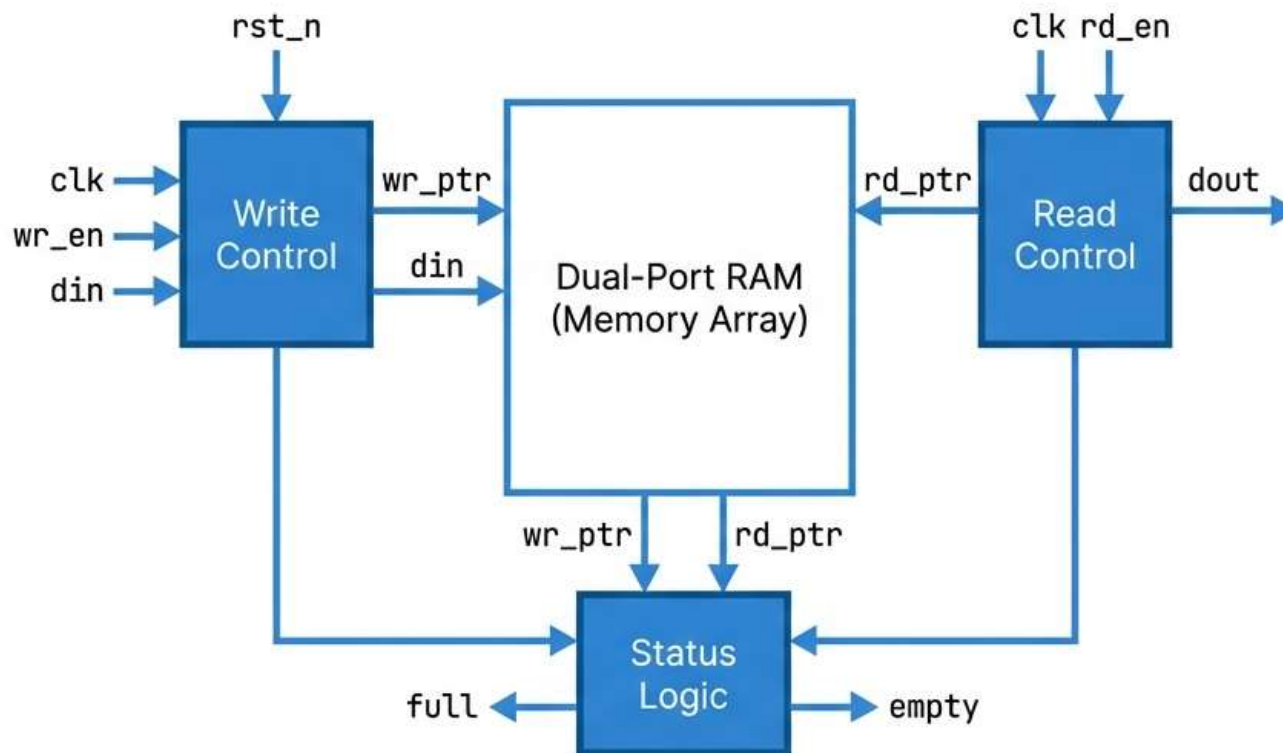
- **Stimulus:** Generates unpredictable inputs.
- **Golden Model:** The trusted mathematical truth (expected = a + b + cin).
- **The Check:**  

```
if (sum !== expected)
  $error("Mismatch!");
```

```
PASSED: a=104, b=22, cin=1 | Sum=127, Cout=0
PASSED: a=50, b=50, cin=0 | Sum=100, Cout=0
```

# The Waiting Room: Synchronous FIFO Architecture

## Managing Data Flow with Sequential Logic and Memory



### Interface:

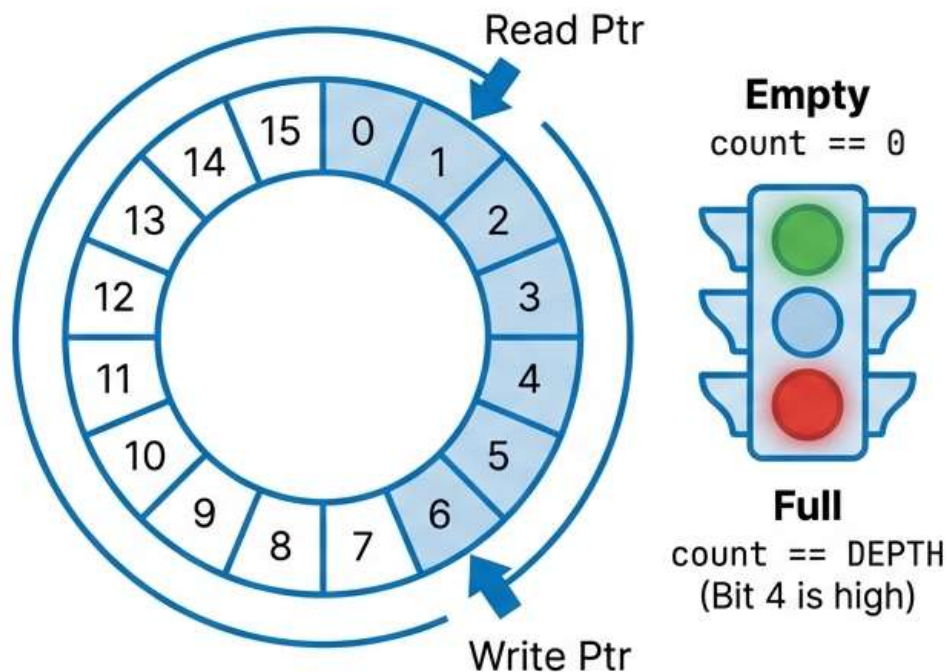
- Inputs:
  - `clk`
  - `rst_n`
  - `wr_en`
  - `rd_en`
  - `din`
- Outputs:
  - `dout`
  - `full`
  - `empty`

### Constraints:

- `wr_valid = wr_en && !full`  
(Prevents Overflow)
- `rd_valid = rd_en && !empty`  
(Prevents Underflow)

# Solving the Empty vs. Full Dilemma

The N+1 Bit Trick for Circular Buffers



**The Problem:** In a circular buffer,  $wr\_ptr == rd\_ptr$  could mean Empty OR Full.

**The Solution:** Use a counter with one extra bit.

**Example:** Depth 16 (4 bits). Counter is 5 bits.

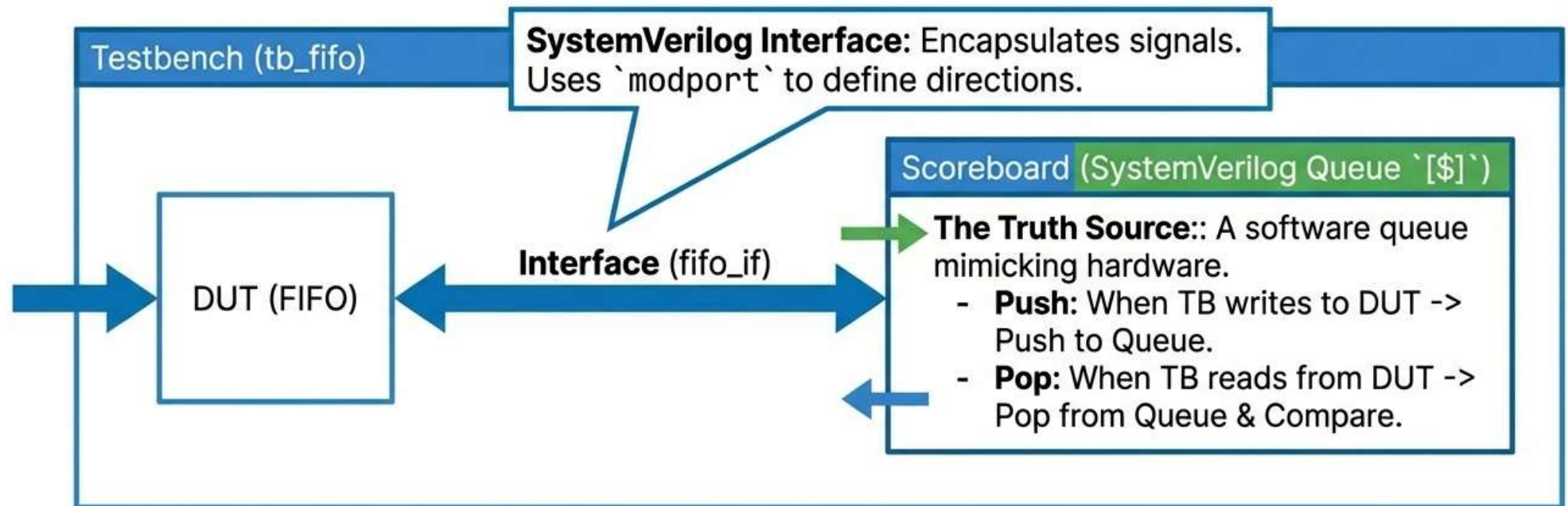
- 00000 = Empty
- 10000 = Full (16 items)

```
localparam PTR_WIDTH = $clog2(DEPTH);  
logic [PTR_WIDTH:0] count; // N+1 bits  
  
assign full = (count == DEPTH);  
assign empty = (count == 0);
```



# Advanced Verification: Interfaces and Scoreboards

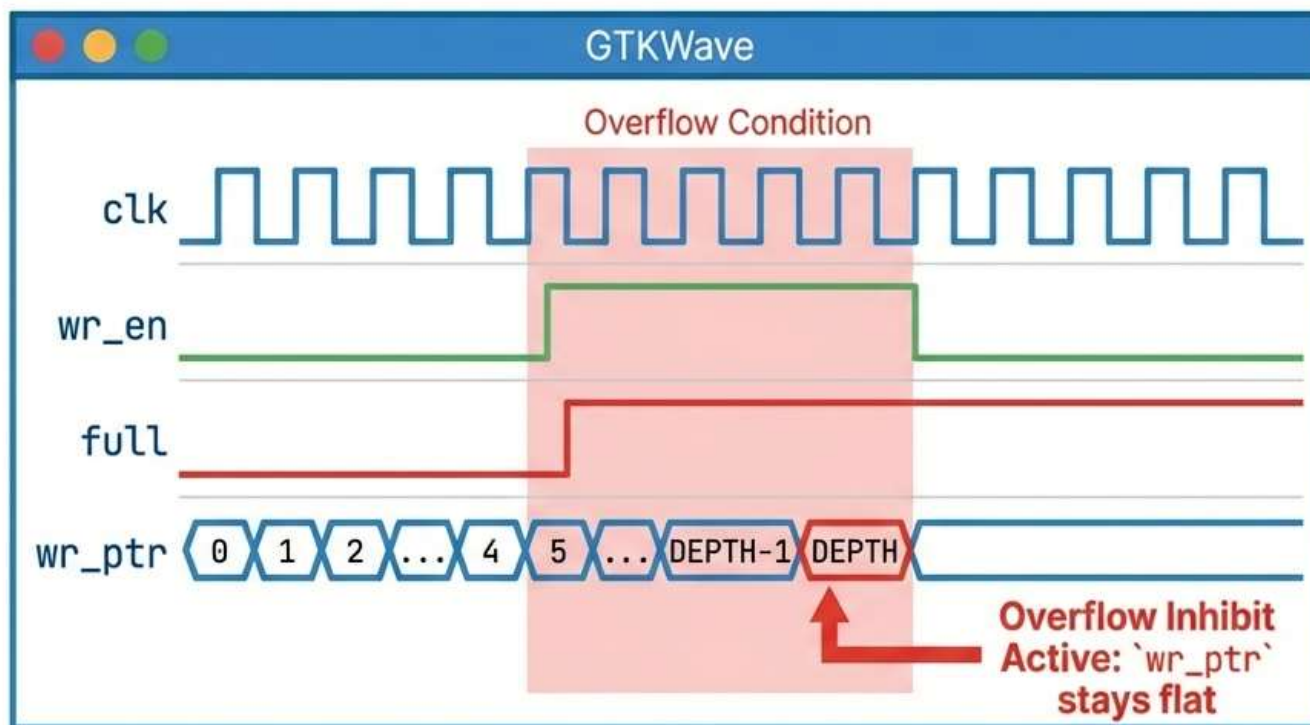
Systematic Approach to Ensuring FIFO Data Integrity and Ordering



**Benefit:** "Ensures data integrity and strict First-In-First-Out ordering."

# Trying to Break It: Corner Cases and Stress Testing

Systematic approach to identifying and validating boundary conditions and resilience

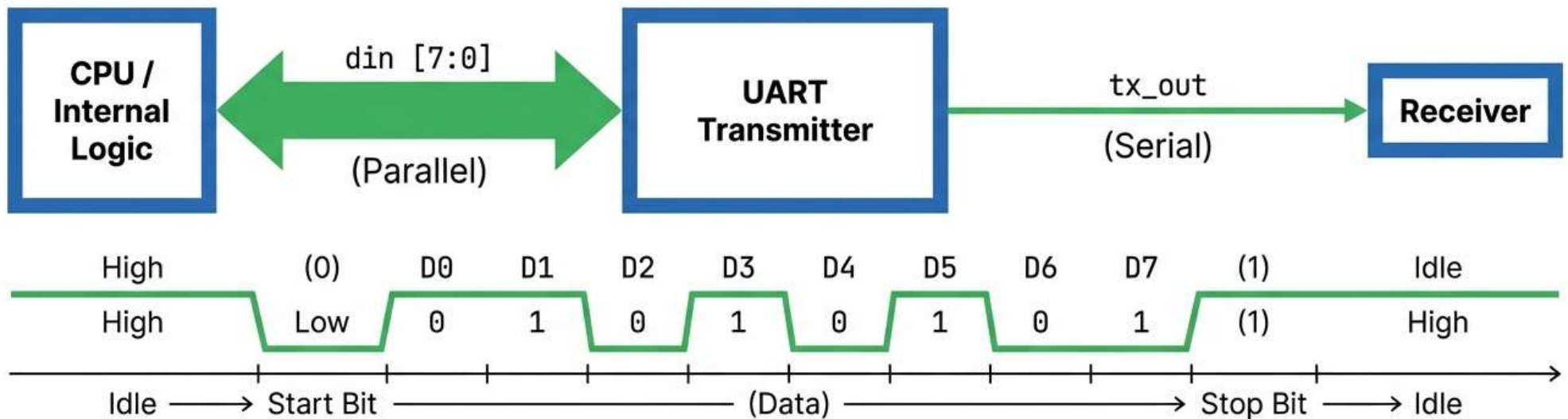


## 4 Verification Scenarios:

1. **Overflow (Fill Test):**  
Write `'DEPTH + 5'` items.  
Verify `'full'` flag blocks excess.
2. **Underflow (Empty Test):**  
Read from empty.  
Verify `'empty'` flag prevents garbage data.
3. **Sustained Throughput:**  
Simultaneous Read/Write ('fork-join'). Verify pointer independence.
4. **The "Surprise" Reset:**  
Trigger async reset mid-operation. Verify instant recovery.

# The Bridge to the Outside World: UART Transmitter

Universal Asynchronous Receiver-Transmitter

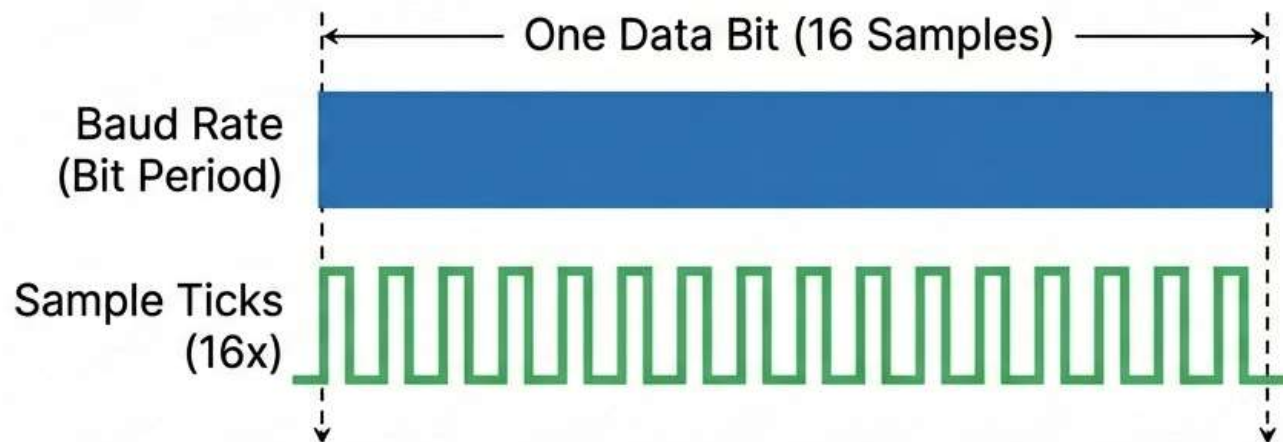


## Key Concepts

- **Asynchronous:** No shared clock. Relies on agreed "Baud Rate".
- **Serializer:** Converts 8-bit parallel to 1-bit serial.
- **Framing:** Wraps data in Start(0) and Stop(1) bits.

# The Heartbeat: Precision Baud Rate Generation

## Oversampling for Data Integrity



### The Math:

Formula:

$$\text{TICK\_LIMIT} = \frac{\text{CLOCK\_FREQ}}{(\text{BAUD\_RATE} * 16)}$$

Example:

$$100\text{MHz} / (115200 * 16) \\ \approx 54 \text{ cycles per tick}$$

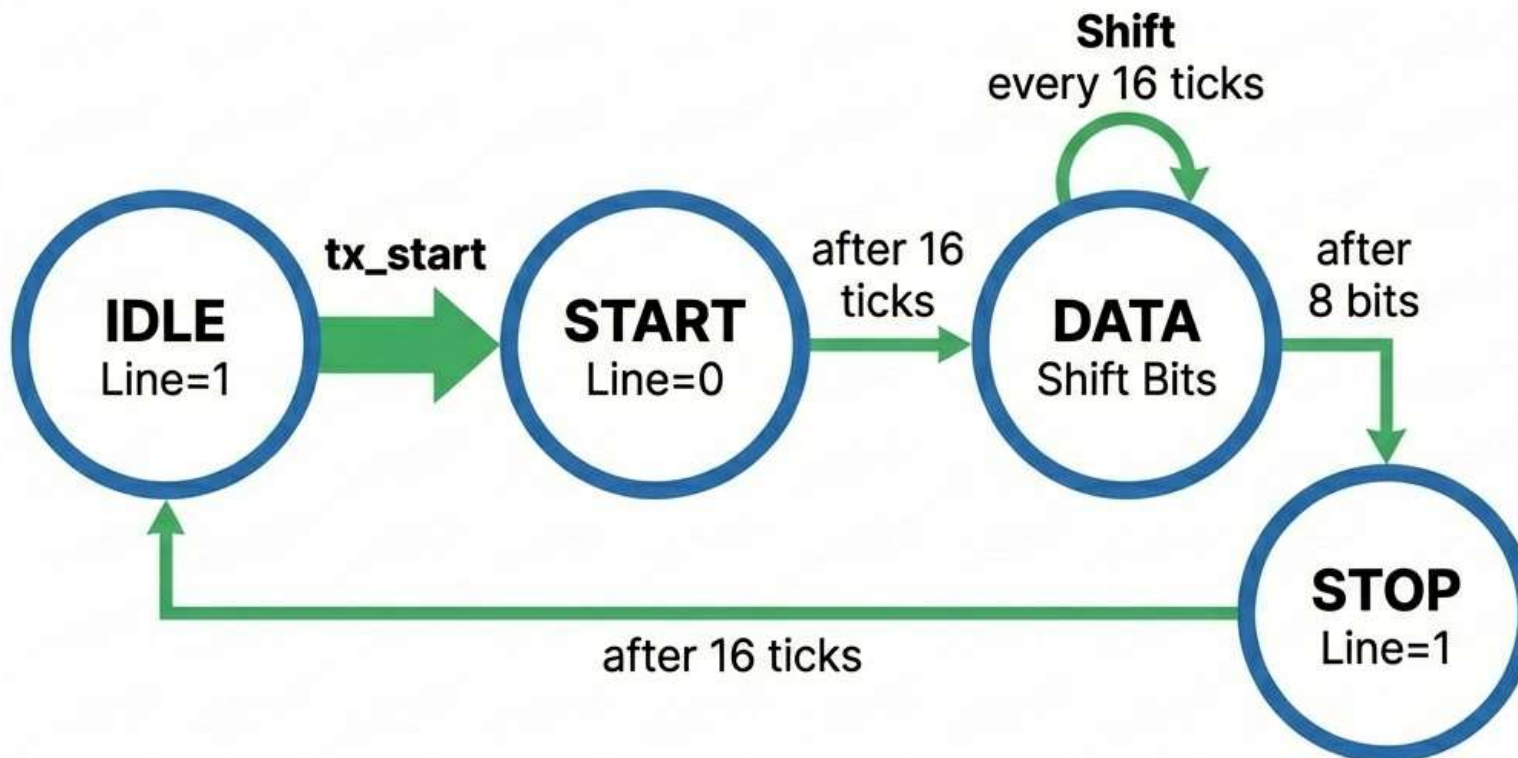
### Logic Flow:

- Counter increments on `clk`.
- At `TICK_LIMIT`, pulse `sample_tick` and Reset.
- **Sync Feature:** Generator is enabled (`en`) only when transmission starts (aligns phase to Start Bit).



# Orchestrating the Frame: The Finite State Machine

Finite State Machine (FSM) for UART Transmission Control



## Internal Counters:

- **tick\_count:** Tracks the 16 oversampling ticks.
- **bit\_count:** Tracks the 8 data bits transferred.

# Professional RTL Architecture: The Two-Block Style

## Sequential Logic (`always\_ff`)

```
always_ff @(posedge clk or negedge rst_n) begin
    if (!rst_n) state_reg <= IDLE;
    else      state_reg <= state_next;
end
```

Handles Reset & Clock updates.  
Defines *\*what\** the state is.

## Combinational Logic (`always\_comb`)

```
always_comb begin
    state_next = state_reg; // Default (Anti-Latch)
    case (state_reg)
        IDLE: if (start) state_next = START;
        // ... next state logic ...
    endcase
end
```

Calculates transitions.  
Defines *\*how\** the state changes.

Standard industry practice for timing stability and glitch prevention.

# The Integrated System

