

## Advanced Antenna Engineering

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I, the undersigned Tong Lin  
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By submitting this course assignment material,  
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3. TL With reference to the above, in particular: I have neither used, nor taken inspiration from codes (like Matlab scripts) written by others (e.g. Classmates), or written by me in collaboration with others;
4. TL In carrying out the tasks of this submitted material I have used only: a) the material(s) provided by the official course instructor(s) via the official course webpage, that does not need referencing; b) materials publicly available<sup>3</sup> (published books, journal papers, etc.) as duly acknowledged below. Any material in b) above **MUST** be clearly listed and precisely referenced in a separate sheet, signed at bottom, to be attached to the submitted paper as an Appendix. Absence of such an Appendix is a declaration that only materials in a) have been used;
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<sup>3</sup> If in doubt about any material that could be used please ask the Instructor by email

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## Advanced Antenna Engineering

TONG LIN

(Complete name, please print)

全霖. Tong Lin

signature<sup>6</sup>

Torino, 01/12/2021 (date)

<sup>6</sup> If your National language is not written in Latin characters (e.g. Chinese, Arabic), you must sign both with your name in Latin characters and as you sign official documents in your National language  
HW declaration, rev. 3, June 2021



### Questions:

- a) Why do we consider that min complexity corresponds to min number of radiating elements? (1 line max)

Due to BFN, larger N means more a complex BFN, which will cause higher losses, etc.

- b) Explain why the presence of grating lobes makes the (max) directivity of the AF decrease (2 lines max)

The presence of the grating lobes when d is large enough that the side lobes become larger than the main lobe, thus decrease the main lobes' directivity.

- c) What is the main limitation of uniform arrays? (1 line max)

Uniform amplitude has no control over Side-Lobe Level.

- d) How can we overcome the main limitation of uniform arrays? (2 lines max)

By using non-uniform amplitudes or non-uniform spacing to improve the SLL.

- e) By using examples, explain the trade-off between (max) directivity and SLL of the AF. (2 lines max)

By setting a proper N, to guarantee a higher value above beam-width condition and a lower value below grating lobes condition, for which the value N shall larger than 7.

- f) In a tapered array, for a given N and SLL we choose the min value of tapering t (i.e. closest to 1): why?

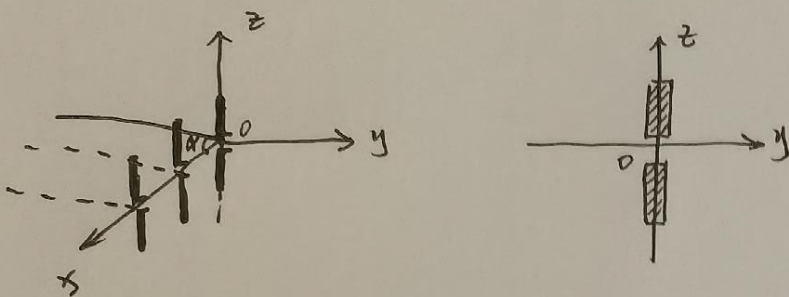
When  $t < 1$  reduces SLL, but also reduces directivity. Therefore, to have a trade-off, we set t closer to 1. Also, normally the amplitude is maximum at center, so t cannot be larger than 1.

- g) For a given spec on beamwidth (e.g. HPBW) and min n. of elements, a uniform array (with no option of SLL) will require less elements than a tapered array with SLL requirement better than 13dB; why? Hint: look at the SLL of a uniform amplitude array (in  $\phi$ ) for N=4-10 on the slides, or beyond 10 using your scripts (see below)

As in the case of uniform array, the increasing element number will increase the SLL level, but with a tapered array, the increasing of n will increasing the main lobe's directivity due to the mathematical explanation, thus a tapered array need more elements numbers to improve performance.

# Assignment 3 Problem 3

1. Envision antenna and chosen coordinate system.



2. Radiation Pattern.

For a single dipole, the radiation pattern is described as,

$$E_{\theta} = j \cdot 2 I_0 \tilde{z}_0 \cdot \frac{\cos(\frac{\pi}{2} \cos \theta)}{\sin \theta}$$

$$E_{\phi} = 0$$

For a array design,

$$E_{array, \theta} = E_{\theta} \cdot AF$$

$$= j \cdot 2 I_0 \tilde{z}_0 \cdot \frac{\cos(\frac{\pi}{2} \cos \theta)}{\sin \theta} \cdot \sum_{n=1}^N A_n \exp[j(n-1) \frac{2\pi d}{\lambda} \cos \alpha]$$

$$\text{where } A_n = \cos \left\{ \left[ \frac{2(n-1)}{N-1} - 1 \right] \cdot k \right\}$$

$$\cos \alpha = \sin \theta \sin \phi$$

$$E_{array, \phi} = E_{\phi} \cdot AF = 0.$$

Also, with the approximation  $E_{\theta}(\theta, \phi) \approx \tilde{E}_{\theta}(\theta, \phi) = C(\sin \theta)^{\frac{3}{2}}$

$$C = E_{array, \theta} \left( \theta = \frac{\pi}{2} \right)$$

$$= j \cdot 2 I_0 \tilde{z}_0 \cdot \sum_{n=1}^N A_n \exp[j(n-1) \frac{2\pi d}{\lambda} \sin \theta]$$

$$\text{So, } E_{array, \theta} = \left\{ j \cdot 2 I_0 \tilde{z}_0 \cdot \sum_{n=1}^N A_n \exp[j(n-1) \frac{2\pi d}{\lambda} \sin \theta] \right\} \cdot (\sin \theta)^{\frac{3}{2}}$$

Therefore, the radiation pattern can be described as,

$$\underline{E}(r, \theta, \phi) = \frac{e^{-jk_0 r}}{4\pi r} \left\{ j \cdot 2 I_0 \tilde{z}_0 \cdot \sum_{n=1}^N A_n \exp[j(n-1) \frac{2\pi d}{\lambda} \sin \theta] \right\} (\sin \theta)^{\frac{3}{2}} \cdot \hat{\theta}$$

For the maximum radiation, where  $\alpha_{\max} = \frac{\pi}{2}$ ,

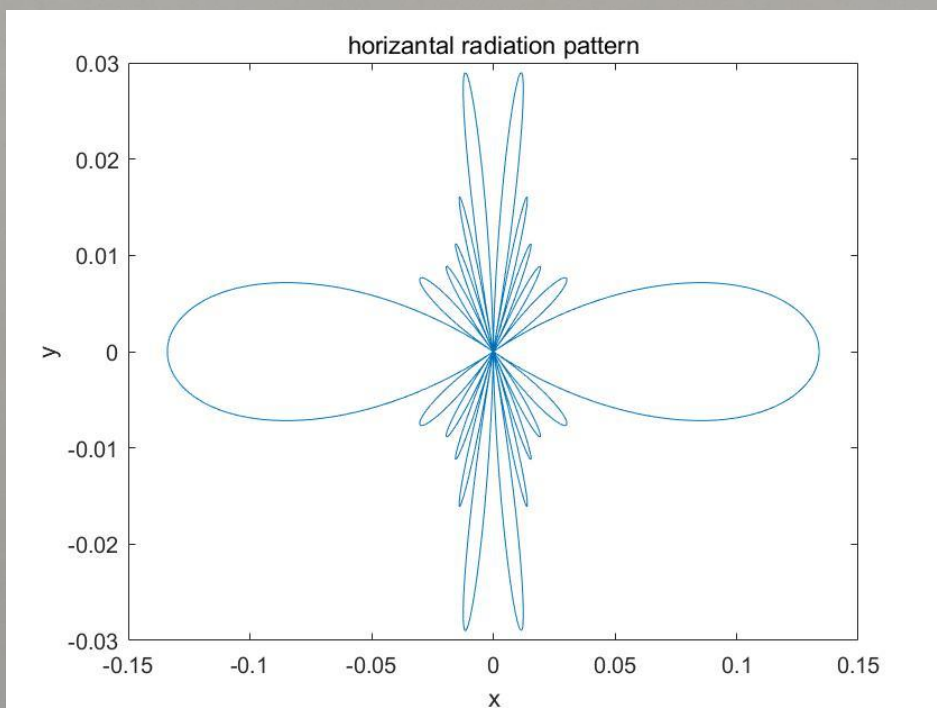
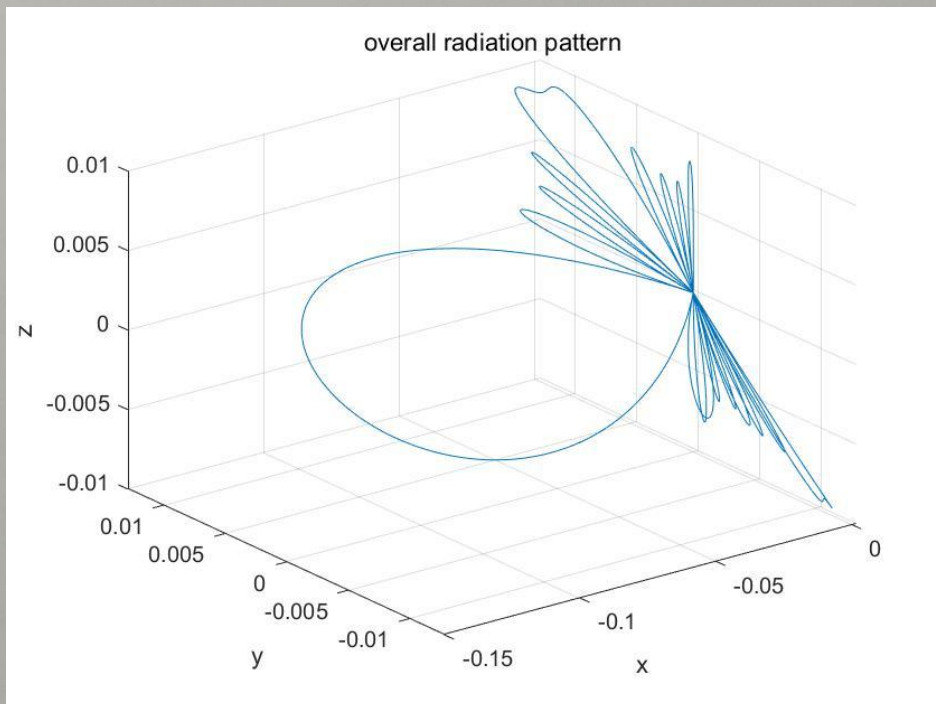
$$E(r, \theta, \varphi) = \frac{e^{-jk_0 r}}{4\pi r} \left\{ j 2 I_0 z_0 \sum_{n=1}^N A_n \right\} \cdot (\sin \theta)^{\frac{3}{2}} \hat{\theta}$$

In the vertical plane:  $\theta = 0$ ,  $\varphi \in [0, 2\pi]$ ,

$$E(r, \theta, \varphi) = \frac{e^{-jk_0 r}}{4\pi r} \left\{ j 2 I_0 z_0 \sum_{n=1}^N A_n \cdot \exp \left[ j(n-1) \frac{2\pi d}{\lambda} \sin \varphi \right] \right\} \cdot (\sin \theta)^{\frac{3}{2}} \\ = 0$$

In the horizontal plane:  $\theta = \frac{\pi}{2}$ ,  $\varphi \in [0, 2\pi]$ ,

$$E(r, \theta, \varphi) = \frac{e^{-jk_0 r}}{4\pi r} \left\{ j 2 I_0 z_0 \sum_{n=1}^N A_n \exp \left[ j(n-1) \frac{2\pi d}{\lambda} \cdot \sin \varphi \right] \right\}$$



## Appendix: Matlab codes

```
clear all;
close all;
clc

j = sqrt(-1);
k0 = 1;    %normalized
r = 1;
I0 = 1;    %normalized
Z0 = 1;    %normalized
N = 7;
d_lamuda = 1-1/N;
lamuda0 = 3e8/850e6;
k = 2*pi/lamuda0;

theta0 = 0:0.0001:pi;
phi0 = 0:0.0002:2*pi;
PHI = 0;
cosalp = sin(theta0).*sin(phi0);
phi = 2*pi*d_lamuda*cosalp+PHI;
F_N = (1/N)*abs(sin(N*phi/2)./sin(phi/2));
E = exp(-j*k0*r)*j*2*I0*Z0*F_N.*sin(theta0).^1.5/(4*pi*r);
x = E.*sin(theta0).*cos(phi0);
y = E.*sin(theta0).*sin(phi0);
z = E.*cos(theta0);
figure
plot3(x,y,z);
title('overall radiation pattern');
xlabel('x');
ylabel('y');
zlabel('z');
grid on

%horizontal
theta0h = pi/2;
phi0 = 0:0.0002:2*pi;
PHI = 0;
cosalp = sin(theta0h).*sin(phi0);
phi = 2*pi*d_lamuda*cosalp+PHI;
F_N = (1/N)*abs(sin(N*phi/2)./sin(phi/2));
E = exp(-j*k0*r)*j*2*I0*Z0*F_N.*sin(theta0h).^1.5/(4*pi*r);
x = E.*sin(theta0h).*cos(phi0);
y = E.*sin(theta0h).*sin(phi0);
```

```

z = E.*cos(theta0h);
figure
plot(x,y);
title('horizontal radiation pattern');
xlabel('x');
ylabel('y');

%vertical
theta0v = 0;
phi0 = 0:0.0002:2*pi;
PHI = 0;
cosalp = sin(theta0v).*sin(phi0);
phi = 2*pi*d_lamuda*cosalp+PHI;
F_N = (1/N)*abs(sin(N*phi/2)./sin(phi/2));
E = exp(-j*k0*r)*j*2*I0*Z0*F_N.*sin(theta0v).^1.5/(4*pi*r);
x = E.*sin(theta0v).*cos(phi0);
y = E.*sin(theta0v).*sin(phi0);
z = E.*cos(theta0v);
figure
plot(y,z,'*');

```

## Reference:

1. linear\_AF\_7x3.pdf, in section 'Handouts'
2. 《天线原理与设计》王建, *Antenna Theory and Design*, by Wang Jian

**Discussed with student Zhang Zhifan.**