

Zadatak

Primjer sustava drugog reda opisanog jednačbom diferencija:

$$y(n) + a_1 y(n-1) + a_2 y(n-2) = b_0 u(n)$$

$$a_1 = -1/2, \quad a_2 = 1/4, \quad b_0 = 1$$

uz početne uvjete $y(-1)=2$, $y(-2)=8$

pobuđen sa $u(n) = \cos(2\pi/3 \cdot n + \pi/3)$

Potrebno je analitički odrediti :

- a) odziv nepobuđenog sustava na početna stanja,
- b) prisilni odziv sustava,
- c) odziv mirnog sustava vlastitim frekvencijama,
- d) ukupni odziv mirnog sustava,
- e) prirodni odziv sustava,
- f) ukupni odziv sustava,

Napisati Matlab program koji racuna i prikazuje sve analiticki izvedene odzive za $n=0,1,2,\dots,30$

Napisati Matlab program koji odredjuje odziv metodom korak po korak.

Usporediti analitička i numerička rješenja.

Opisati oblik jed. dif. za sustav II reda

$$y(n+2) + a_1 y(n+1) + a_2 y(n) = b_0 u(n+2) + b_1 u(n+1) + b_2 u(n) \quad (1)$$

$$y(n') + a_1 y(n'-1) + a_2 y(n'-2) = b_0 u(n') + b_1 u(n'-1) + b_2 u(n'-2) \quad n'=n+2$$

homogeni dio rješenja $u(n') = 0 \quad \forall n'$

$$y(n') + a_1 y(n'-1) + a_2 y(n'-2) = 0$$

Pretpostavimo rji. obliku $c \cdot z^{n'} = y(n')$

$$c \cdot z^{n'} + a_1 \cdot c \cdot z^{n'-1} + a_2 \cdot c \cdot z^{n'-2} = 0$$

$$c \cdot z^{n-2} (z^2 + a_1 z + a_2) = 0$$

karakteristični polinom

$$z^2 + a_1 z + a_2 = 0$$

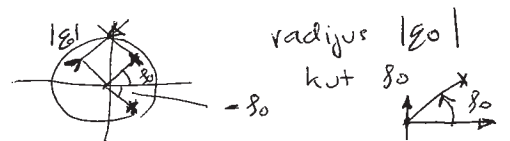
karakteristična jednačina

Potražujemo rješenja:

$$z_{1,2} = \frac{-a_1 \pm \sqrt{a_1^2 - 4a_2}}{2}$$

Odaberemo primjer sa konj. komp. parom $z_1 = z_0, z_2 = z_0^*$

$$z_1 = |z_0| \cdot e^{j\theta_0} \quad z_2 = |z_0| \cdot e^{-j\theta_0}$$



Kako glase koeficijenti karakter. polinoma za ovaj par karakterističnih frekvencija?

Prikažimo polinom kao produkt kompleksu faktora

$$(z - z_1)(z - z_2) = (z - |z_0| \cdot e^{j\theta_0})(z - |z_0| \cdot e^{-j\theta_0})$$

$$= z^2 - z|z_0| \cdot (\underbrace{e^{j\theta_0} + e^{-j\theta_0}}_{2\cos\theta_0}) + \underbrace{|z_0|^2 \cdot e^{j\theta_0} \cdot e^{-j\theta_0}}_1$$

$$= z^2 - \underbrace{2|z_0| \cdot \cos\theta_0}_{a_1} z + \underbrace{|z_0|^2}_{a_2}$$

Npr. odaberimo $|g_0| = \frac{1}{2}$, $\theta_0 = \frac{\pi}{3} \Rightarrow \cos \theta_0 = \frac{1}{2}$

(2)

$$a_1 = -2|g_0| \cdot \cos \theta_0 = -2 \cdot \frac{1}{2} \cdot \frac{1}{2} = -\frac{1}{2}$$

$$a_2 = |g_0|^2 = \left(\frac{1}{2}\right)^2 = \frac{1}{4}$$

(jednadba)

Dalje karakteristični polinom χ_f glasi:

$$z^2 - \frac{1}{2}z + \frac{1}{4} = 0$$

Za proučavanje nadjemo rješenje:

$$z_{1,2} = \frac{\frac{1}{2} \pm \sqrt{\frac{1}{4} - 1}}{2} = \frac{\frac{1}{2} \pm \frac{1}{2}j\sqrt{3}}{2} = \frac{1}{2} \left(\frac{1}{2} \pm \frac{j\sqrt{3}}{2} \right)$$

$|g_0|$

par konjug. kompl.
karakterističnih
frekvencija

$$e^{j\frac{\pi}{3}} = \left(\cos\left(\frac{\pi}{3}\right) + j \sin\left(\frac{\pi}{3}\right) \right)$$

$$\frac{1}{2} \pm j \frac{\sqrt{3}}{2} \text{ u OK}$$

Kada smo odredili karakteristične
(ili vlastite) frekvencije sustava z_1, z_2

možemo odrediti odziv neposrednog sustava

(početna stanja $\neq 0$, ali $u(n) = 0 \forall n$)

Pretpostavimo rješenje oblike

$$y_H(n) = c_1 \cdot z_1^n + c_2 \cdot z_2^n$$

(3)

$$y(u) = -a_1 y(u-1) - a_2 y(u-2)$$

Moramo postaviti početne uvjete

$$\left. \begin{array}{l} y(u-1) \\ y(u-2) \end{array} \right\} \text{ npr. za } u=0$$

$$y(0) = -a_1 y(-1) - a_2 y(-2)$$

$$y(1) = -a_1 y(0) - a_2 y(-1)$$

$$\left. \begin{array}{l} \text{npr. postavimo} \\ y(-1) \\ y(-2) \end{array} \right\} \text{ treba nam}$$

$$y(0) = -a_1 y(-1) - a_2 y(-2)$$

$$\begin{aligned} y(1) &= a_1^2 y(-1) + a_1 a_2 y(-2) - a_2 y(-1) \\ &= (a_1^2 - a_2) y(-1) + a_1 a_2 y(-2) \end{aligned}$$

12 homog. rješenje

za homog. rješenje mogli smo odabrati i bilo koji drugi par u -ova uključujući $u=-1, u=-2$

$$\begin{aligned} C_1 q_1^u + C_2 q_2^u &= /u=0/ = C_1 + C_2 = y(0) \\ &= /u=1/ = C_1 q_1 + C_2 q_2 = y(1) \end{aligned}$$

↓
svi daju isto rješenje za a_1, a_2

Dobivamo 2 jedn. sa 2 nepoz. (C_1, C_2)

$$C_1 + C_2 = -a_1 y(-1) - a_2 y(-2)$$

$$a_1 = -\frac{1}{2}$$

$$C_1 q_1 + C_2 q_2 = (a_1^2 - a_2) y(-1) + a_1 a_2 y(-2)$$

$$a_2 = \frac{1}{4}$$

$$C_1 + C_2 = \frac{1}{2} y(-1) - \frac{1}{4} y(-2)$$

$$C_1 q_1 + C_2 q_2 = \left(\frac{1}{4} - \frac{1}{4}\right) y(-1) - \frac{1}{2} \cdot \frac{1}{4} y(-2) = -\frac{1}{8} y(-2)$$

Npr. odaberemo početne uvjete $y(-1) = 2$
 $y(-2) = 8$

(4)

$$C_1 + C_2 = \frac{1}{2} \cdot 2 - \frac{1}{4} \cdot 8 = 1 - 2 = -1 \quad C_1 = -C_2 - 1$$

$$C_1 q_1 + C_2 q_2 = -\frac{1}{8} y(-2) = -1$$

$$q_1(-C_2 - 1) + C_2 q_2 = -1$$

$$q_1 = \frac{1}{4} + \frac{\sqrt{3}}{4}j$$

$$C_2(q_2 - q_1) = q_1 - 1$$

$$q_2 = \frac{1}{4} - \frac{\sqrt{3}}{4}j$$

$$C_2 = \frac{q_1 - 1}{q_2 - q_1}$$

$$q_2 - q_1 = -\frac{\sqrt{3}}{2}j$$

$$C_2 = \frac{-\frac{3}{4} + \frac{\sqrt{3}}{4}j}{-\frac{\sqrt{3}}{2}j} \cdot \frac{\frac{2}{\sqrt{3}}j}{\frac{2}{\sqrt{3}}j} = -\frac{6}{4\sqrt{3}}j - \frac{1}{2}$$

$$= -\frac{1}{2} - \frac{\sqrt{3}}{2}j$$

$$C_1 = -C_2 - 1 = \frac{1}{2} + \frac{\sqrt{3}}{2}j - 1 = -\frac{1}{2} + \frac{\sqrt{3}}{2}j = C_2^*$$

Provjera

$$C_1 q_1 + C_2 q_2 = \left(-\frac{1}{2} + \frac{\sqrt{3}}{2}j\right)\left(\frac{1}{4} + \frac{\sqrt{3}}{4}j\right) + \left(-\frac{1}{2} - \frac{\sqrt{3}}{2}j\right)\left(\frac{1}{4} - \frac{\sqrt{3}}{4}j\right)$$

$$= \left(-\frac{1}{8} - \frac{3}{8} + j\left(\frac{\sqrt{3}}{8} - \frac{\sqrt{3}}{8}\right)\right) + \left(-\frac{1}{8} - \frac{3}{8} + j\left(-\frac{\sqrt{3}}{8} + \frac{\sqrt{3}}{8}\right)\right) = -\frac{1}{2} - \frac{1}{2} = -1$$

Dalje: rješenje glasi:

$$y_n(u) = C_1 q_1^n + C_2 q_2^n = \left(-\frac{1}{2} + \frac{\sqrt{3}}{2}j\right)\left(\frac{1}{4} + \frac{\sqrt{3}}{4}j\right)^n + \left(-\frac{1}{2} - \frac{\sqrt{3}}{2}j\right)\left(\frac{1}{4} - \frac{\sqrt{3}}{4}j\right)^n$$

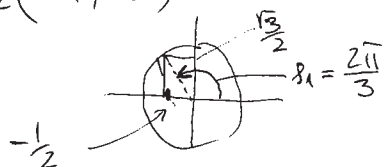
ili u polarnom obliku:

$$|C_1| = |C_2| = \frac{1}{4} + \frac{3}{4} = 1$$

$$C_1 = |C_1| \cdot e^{j\theta_1} \quad C_2 = |C_2| \cdot e^{j\theta_2}$$

$$\theta_1 = \arctan_2(\operatorname{Im}, \operatorname{Re})$$

$$C_1 = 1 \cdot e^{j\frac{2\pi}{3}} \quad C_2 = C_1^* = 1 \cdot e^{-j\frac{2\pi}{3}}$$



$$q_1 = |q_0| \cdot e^{j\theta_0}$$

$$q_2 = q_1^* = |q_0| \cdot e^{-j\theta_0}$$

$$|q_0| = \frac{1}{2}, \quad \theta_0 = \pi/3$$

Dalje u polarnom obliku:

(5)

$$\begin{aligned}
 y_H(n) &= e^{j\frac{2n}{3}} \left(\frac{1}{2} e^{j\frac{\pi}{3}} \right)^n + e^{-j\frac{2n}{3}} \left(\frac{1}{2} e^{-j\frac{\pi}{3}} \right)^n \\
 &= \left(\frac{1}{2} \right)^n \left[e^{j(n \cdot \frac{\pi}{3} + \frac{2n}{3})} + e^{-j(n \cdot \frac{\pi}{3} + \frac{2n}{3})} \right] \\
 &= \left(\frac{1}{2} \right)^n \cdot 2 \cos \left(n \frac{\pi}{3} + \frac{2n}{3} \right) \\
 &= 2^{(1-n)} \cdot \cos \left(n \frac{\pi}{3} + \frac{2n}{3} \right)
 \end{aligned}$$

Proverimo paritet uvrste:

$$y_H(-1) = 2^{(1-(-1))} \cdot \cos \left(-\frac{\pi}{3} + \frac{2(-1)}{3} \right) = 4 \cdot \cos \left(\frac{\pi}{3} \right) = 4 \cdot \frac{1}{2} = 2$$

$$y_H(-2) = 2^{(1+2)} \cdot \cos \left(-\frac{2\pi}{3} + \frac{2(-2)}{3} \right) = 8 \cdot \cos(\pi) = -8$$



Metoda koraka po koraku ... uvrstimo $y(0)$, $y(1)$

$$\begin{aligned}
 y(0) &= -a_1 y(-1) - a_2 y(-2) = \left/ \begin{matrix} a_1 = -\frac{1}{2} \\ a_2 = \frac{1}{4} \end{matrix} \right/ = \frac{1}{2} y(-1) - \frac{1}{4} y(-2) \\
 &= \frac{1}{2} \cdot 2 - \frac{1}{4} \cdot 8 = -1
 \end{aligned}$$

$$\begin{aligned}
 y(1) &= -a_1 y(0) - a_2 y(-1) = \left/ -1 \right/ = \frac{1}{2} \cdot (-1) - \frac{1}{4} \cdot 2 \\
 &= -\frac{1}{2} - \frac{1}{2} = -1
 \end{aligned}$$

Proverimo opie isprave za $n=0, 1$

$$y_H(0) = 2^1 \cdot \cos \left(\frac{2\pi}{3} \right) = -1 \text{ u OK}$$

$$y_H(1) = 2^0 \cdot \cos \left(\frac{\pi}{3} + \frac{2\pi}{3} \right) = 1 \cdot (-1) = -1 \text{ u OK}$$

Poslednec ovaj sustav drugog reda sa pobudom harmoničkog oblika

(6)

$$u[n] = A \cos\left(\frac{2\pi}{3}n + \frac{\pi}{3}\right) \quad \text{upr. vel. } A=1$$

$\swarrow \quad \searrow$
 $\omega_0 = \frac{2\pi}{3} \quad \theta = \frac{\pi}{3}$

općenito:

$$u[n] = A \cdot \cos(\omega_0 n + \theta)$$

možemo je prikazati i u exp. obliku (stvar. delje komp. exp.)

$$u[n] = \frac{A}{2} \cdot e^{j\theta} \cdot e^{j\omega_0 n} + \frac{A}{2} \cdot e^{-j\theta} \cdot e^{-j\omega_0 n}, \text{ jer}$$

$$= \frac{A}{2} \left(\underbrace{e^{j(\theta + \omega_0 n)} + e^{-j(\theta + \omega_0 n)}}_{2 \cos(\omega_0 n + \theta)} \right) = A \cdot \cos(\omega_0 n + \theta)$$

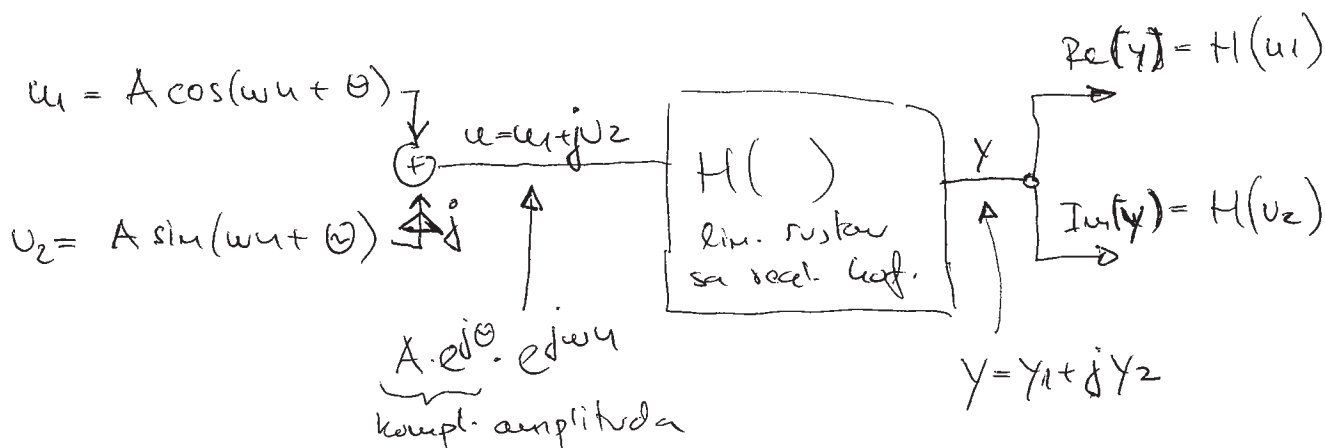
ili

$$u[n] = \operatorname{Re}(A \cdot e^{j\theta} \cdot e^{j\omega_0 n}) =$$

$$= \operatorname{Re}(A \cdot e^{j(\omega_0 n + \theta)}) =$$

$$= \operatorname{Re}(A \cdot \cos(\omega_0 n + \theta) + jA \sin(\omega_0 n + \theta))$$

$$= A \cdot \cos(\omega_0 n + \theta)$$



$$\text{Odeir na pobudu } H(u_1 + ju_2) = y_1 + jy_2$$

Zbog linearnosti! ako je u_1 realan i u_2 realan i H realan tada je $\operatorname{Re}(y) = y_1$ a $\operatorname{Im}(y) = y_2$

Pri określaniu particularnej metody
nieodreżony koef. ze powodu obliczeń:

(7)

$$u[n] = \underbrace{A}_{A... \text{komp! amplitude}} \cdot e^{j\omega_0 n}$$

możemy założyć: reakcja na obliczenia albo drugie
komp! amplitude K

$$y_p[n] = K \cdot e^{j\omega_0 n}$$

K ... oznacza ^{realna} i ^{amplitude} $|K|$ i ^{fazowy} kąt (pojemność)

Uwzględniamy założenia $y[n]$ i $u[n]$ jednor. dif.

$$y[n] + a_1 y[n-1] + a_2 y[n-2] = b_0 u[n]$$

$$y[n] + \frac{1}{2} y[n-1] + \frac{1}{4} y[n-2] = u[n]$$

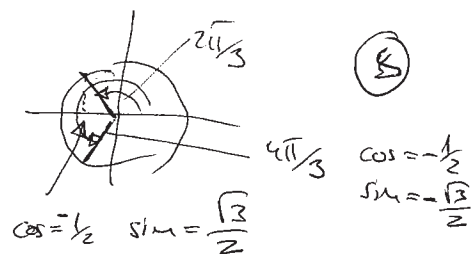
/upr. $b_1 = b_2 = 0$ /
/ $a_1 = -\frac{1}{2}$ $a_2 = \frac{1}{4}$ /
/ ulega że $b_0 = 1$ /

$$K \cdot e^{j\omega_0 n} - \frac{1}{2} K \cdot e^{j\omega_0 (n-1)} + \frac{1}{4} K \cdot e^{j\omega_0 (n-2)} = A \cdot e^{j\omega_0 n}$$

$$K \cdot e^{j\omega_0 n} \left(1 - \frac{1}{2} \cdot e^{-j\omega_0} + \frac{1}{4} \cdot e^{-2j\omega_0} \right) = A \cdot e^{j\omega_0 n}$$

$$K = \frac{A}{\left(1 - \frac{1}{2} e^{-j\omega_0} + \frac{1}{4} e^{-2j\omega_0} \right)}$$

Drugi primjer... za $\omega_0 = \frac{2\pi}{3}$, $A = 1 \cdot e^{j\pi/3}$



$$K = \frac{A}{\left(1 - \frac{1}{2}e^{-j\omega_0} + \frac{1}{4}e^{-2j\omega_0}\right)}$$

$$e^{-j\frac{4\pi}{3}} = \cos\left(\frac{4\pi}{3}\right) - j\sin\left(\frac{4\pi}{3}\right) = -\frac{1}{2} + j\frac{\sqrt{3}}{2}$$

$$K = \frac{e^{j\pi/3}}{1 - \frac{1}{2}\left(-\frac{1}{2} - j\frac{\sqrt{3}}{2}\right) + \frac{1}{4}\left(-\frac{1}{2} + j\frac{\sqrt{3}}{2}\right)}$$

$$e^{-j\frac{2\pi}{3}} = \cos\left(\frac{2\pi}{3}\right) - j\sin\left(\frac{2\pi}{3}\right) = -\frac{1}{2} - j\frac{\sqrt{3}}{2}$$

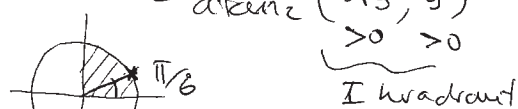
$$K = \frac{e^{j\pi/3}}{1 + \frac{1}{4} + j\frac{\sqrt{3}}{4} - \frac{1}{8} + j\frac{\sqrt{3}}{8}} = \frac{e^{j\pi/3}}{\frac{9}{8} + j\frac{3\sqrt{3}}{8}} \quad \text{on}$$

$$| |^2 = \frac{g^2}{8^2} + \frac{3^2 \cdot 3}{8^2} = \frac{81 + 27}{64} = \frac{108}{64} = \frac{54}{32} = \frac{27}{16}$$

$$K = \frac{4}{3\sqrt{3}} e^{j\pi/3} \cdot e^{-j\pi/6} = \frac{4}{3\sqrt{3}} e^{j\pi/6} = \frac{4\sqrt{3}}{9} e^{j\pi/6}$$

$$| | = \frac{3\sqrt{3}}{4}$$

$$\arg = \arctan_2(\text{Im}, \text{Re}) = \arctan_2(3\sqrt{3}, 9)$$



$$\arctan\left(\frac{3\sqrt{3}}{9}\right) =$$

$$\arctan\left(\frac{1}{\sqrt{3}}\right) = \frac{\pi}{6}$$

Dakle uz pretpostavljeno referenja oblika

$$y_p[u] = K \cdot e^{j\omega_0 u}$$

dobivamo

$$y_p[u] = \left(\frac{4\sqrt{3}}{9}\right) e^{j\frac{\pi}{6}} \cdot e^{j\frac{2\pi}{3}u}$$

↑
modul

↑
početna faza
(za $u=0$)

↑
frekvencija
odziva

na pobudu oblika:

$$u[u] = 1 \cdot e^{j\pi/3} \cdot e^{j\frac{2\pi}{3}u}$$

↑
frekvencija
pobudnog
signala

moraju biti
jednake zbog
linearosti sustava

$$y_p[u] = \text{Re}[y_p[u]] + j\text{Im}(y_p[u])$$

$$= y_{pr}[u] + j y_{pi}[u] \quad \text{su odziv na}$$

$$u[u] = \text{Re}[u[u]] + j\text{Im}(u[u])$$

$$= u_r[u] + j u_i[u]$$

pr tome $y_{pi}[u]$ je odziv na u_r
a $y_{pr}[u]$ je odziv na $u_i[u]$

(9)

$$u_r[u] = \cos\left(\frac{2\pi}{3}u + \frac{\pi}{3}\right)$$

$$\begin{aligned} y_{pr}[u] &= \operatorname{Re}\left(\frac{4\sqrt{3}}{9} \cdot e^{j\frac{\pi}{6}} \cdot e^{j\frac{2\pi}{3}u}\right) \\ &= \frac{4\sqrt{3}}{9} \cdot \operatorname{Re}\left(e^{j\left(\frac{2\pi}{3}u + \frac{\pi}{6}\right)}\right) \\ &= \frac{4\sqrt{3}}{9} \cdot \cos\left(\frac{2\pi}{3}u + \frac{\pi}{6}\right) \end{aligned}$$

Tako nas je interesirao odziv na $u_r[u]$, primenom kompleksne resp. automatski smo dobili i odziv na $u_i[u] = \sin\left(\frac{2\pi}{3}u + \frac{\pi}{3}\right)$ kao imaginarni dio odziva $\operatorname{Im}(y_{pr}[u])$

$$\begin{aligned} y_{pi}[u] &= \operatorname{Im}\left(\frac{4\sqrt{3}}{9} \cdot e^{j\frac{\pi}{6}} \cdot e^{j\frac{2\pi}{3}u}\right) \\ &= \frac{4\sqrt{3}}{9} \cdot \sin\left(\frac{2\pi}{3}u + \frac{\pi}{6}\right) \end{aligned}$$

Alternativan način rješavanja kod kojeg izbjegnemo rad sa kompleksnim brojevima je da pretpostavimo realno rješenje u obliku:

$$y_{pr}[u] = k_1 \cdot \cos(\omega_0 u) + k_2 \cdot \sin(\omega_0 u)$$

\Rightarrow u konstantama k_1, k_2 sadržana je informacija o realnoj amplitudi i početnoj fazi izlaze kosinusoida:

$$y_{pr}[u] = A_y \cdot \cos(\omega_0 u + \theta_0)$$

\uparrow realna ampl. \nwarrow početna faza

$$\begin{aligned} y_{pr}[u] &= A_y (\cos(\omega_0 u) \cos(\theta_0) - \sin(\omega_0 u) \cdot \sin(\theta_0)) \\ &= \underbrace{(A_y \cdot \cos(\theta_0))}_{k_1} \cdot \cos(\omega_0 u) + \underbrace{(-A_y \cdot \sin(\theta_0))}_{k_2} \cdot \sin(\omega_0 u) \end{aligned}$$

$$k_1 = A_y \cdot \cos(\theta_0)$$

$$k_2 = -A_y \cdot \sin(\theta_0)$$

kako odrediti A_y i θ_0 iz k_1 i k_2 ?

(10)

$$\begin{aligned} k_1^2 + k_2^2 &= A_y^2 \cdot \cos^2(\theta_0) + A_y^2 \cdot \sin^2(\theta_0) \\ &= A_y^2 (\cos^2 \theta_0 + \sin^2 \theta_0) \\ &= A_y^2 \end{aligned}$$

$$\Rightarrow A_y = \sqrt{k_1^2 + k_2^2} \quad A_y \in \mathbb{R}$$

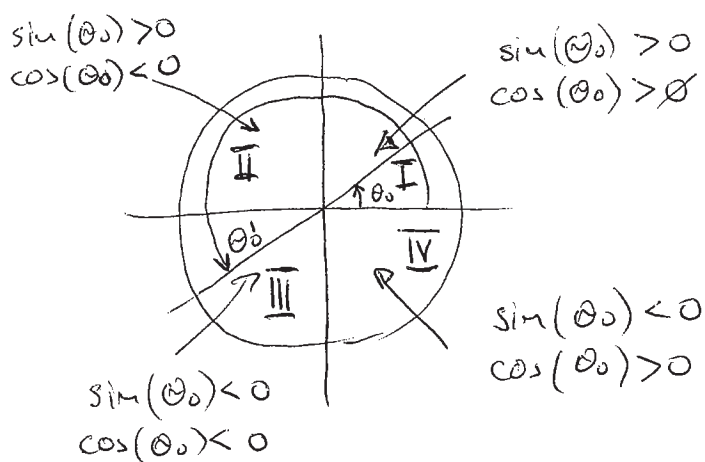
kada znamo A_y treba odrediti θ_0 :

$$\cos(\theta_0) = \frac{k_1}{A_y}, \quad \sin(\theta_0) = -\frac{k_2}{A_y}$$

$$\operatorname{tg}(\theta_0) = \frac{\sin(\theta_0)}{\cos(\theta_0)} \Rightarrow \theta_0 = \operatorname{atan}\left(\frac{\sin(\theta_0)}{\cos(\theta_0)}\right)$$

Mora mo koristiti četvero-kvadrantnu atan funkciju.

Običan atan daje samo dvo-kvadrantno rješenje



za običan atan imamo

θ_0 u I kvadrantu i

θ'_0 u III kvadrantu su

ekvivalentni

$$\theta'_0 = \theta_0 + \pi \quad \theta_0 \in (0, \pi/2)$$

$$\begin{aligned} \operatorname{tg}(\theta'_0) &= \frac{\sin(\theta_0 + \pi)}{\cos(\theta_0 + \pi)} = \frac{\overset{-1}{\sin(\theta_0)} \cdot \overset{-1}{\cos(\pi)} + \cos(\theta_0) \cdot \overset{\cancel{0}}{\sin(\pi)}}{\overset{-1}{\cos(\theta_0)} \cdot \overset{-1}{\cos(\pi)} - \sin(\theta_0) \cdot \overset{\cancel{0}}{\sin(\pi)}} \\ &= \frac{-\sin(\theta_0)}{-\cos(\theta_0)} = \operatorname{tg}(\theta_0) \end{aligned}$$

Jednako

"ekvivalencija"

vrjedbi za kutove

u II i IV kvadrantu

za $\theta_0 \in [0, \pi/2)$ i

$$\theta'_0 = \theta_0 + \pi$$

kako su predznaci sin i cos suprotni

imaemo istu vrijednost tg funkcije

Kako odrediti četvero-kvadrantno rješenje ako znamo vrijednost sinusa i kosinusa?

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1. Izračunamo običan atan($\frac{\sin}{\cos}$) = $\Phi_{I,IV}$
2. Uz pretpostavku da je kosinus običnog atan-a $[-\frac{\pi}{2}, \frac{\pi}{2}]$ tada automatski dobivamo traženo rješenje u slučaju da je Φ_0 u I ili IV kvadrantu.
3. Međutim, ako je $\cos < 0$ tada je sigurno rješenje u II ili III kvadrantu... dakle pravo rješenje je tako

$$\Phi_{\pm} = \begin{cases} \Phi_{I,IV} \text{ ako } \cos \geq 0 \\ \Phi_{I,IV} + \pi \text{ inače} \end{cases}$$

4. Ako želimo da konačna kosinusa četvero-kvadrantnog atan-a bude $[0, 2\pi)$ tada je još potrebno rješenja iz IV kvadranta pretvoriti u pozitivna dodavanjem 2π

$$\Phi_{\pm} = \begin{cases} \Phi \text{ ako } \Phi \geq 0 \\ 2\pi + \Phi \text{ ako } \Phi < 0 \end{cases}$$

Uvrtimo sada pretpostavljenu rješenje u jed. dif

(12)

$$y_p(u) = k_1 \cdot \cos(\omega_0 u) + k_2 \sin(\omega_0 u) \quad \omega_0 = \frac{2\pi}{3}$$

$$y(u) + a_1 y(u-1) + a_2 y(u-2) = u_f(u)$$

$$u_f(u) = \cos\left(\frac{2\pi}{3}u + \frac{\pi}{3}\right)$$

$$= \cos\frac{2\pi}{3}u \cos\frac{\pi}{3} - \sin\frac{2\pi}{3}u \sin\frac{\pi}{3}$$

$$= \frac{1}{2} \cos\left(\frac{2\pi}{3}u\right) - \frac{\sqrt{3}}{2} \sin\left(\frac{2\pi}{3}u\right)$$

$$k_1 \cdot \cos(\omega_0 u) + k_2 \cdot \sin(\omega_0 u) +$$

$$a_1 k_1 \cos(\omega_0(u-1)) + a_1 k_2 \sin(\omega_0(u-1)) +$$

$$a_2 k_1 \cos(\omega_0(u-2)) + a_2 k_2 \sin(\omega_0(u-2)) = \frac{1}{2} \cos\left(\frac{2\pi}{3}u\right) - \frac{\sqrt{3}}{2} \sin\left(\frac{2\pi}{3}u\right)$$

$$\left. \begin{aligned} \cos(\omega_0(u-k)) &= \cos(\omega_0 u) \cos(k\omega_0) + \sin(\omega_0 u) \sin(k\omega_0) \\ \sin(\omega_0(u-k)) &= \sin(\omega_0 u) \cos(k\omega_0) - \cos(\omega_0 u) \sin(k\omega_0) \end{aligned} \right\} \text{ za } \omega_0 = \frac{2\pi}{3}$$

$$k=1 \quad \cos(\omega_0(u-1)) = \cos(\omega_0 u) \cdot \left(-\frac{1}{2}\right) + \sin(\omega_0 u) \cdot \frac{\sqrt{3}}{2}$$

$$k=2 \quad \cos(\omega_0(u-2)) = \cos(\omega_0 u) \cdot \left(-\frac{1}{2}\right) + \sin(\omega_0 u) \cdot \left(-\frac{\sqrt{3}}{2}\right)$$

$$k=1 \quad \sin(\omega_0(u-1)) = \sin(\omega_0 u) \cdot \left(-\frac{1}{2}\right) - \cos(\omega_0 u) \cdot \frac{\sqrt{3}}{2}$$

$$k=2 \quad \sin(\omega_0(u-2)) = \sin(\omega_0 u) \cdot \left(-\frac{1}{2}\right) - \cos(\omega_0 u) \cdot \left(-\frac{\sqrt{3}}{2}\right)$$

Grupirajmo sve članove sa lijeve strane koji sadrže članje istog oblika, ali različitih amplituda (... dakle posebno $\cos(\omega_0 u)$, posebno $\sin(\omega_0 u)$)

$$\left(k_1 + a_1 k_1 \left(-\frac{1}{2}\right) + a_1 k_2 \left(-\frac{\sqrt{3}}{2}\right) + a_2 k_1 \left(-\frac{1}{2}\right) + a_2 k_2 \left(\frac{\sqrt{3}}{2}\right) \right) \cos(\omega_0 u) +$$

$$\left(k_2 + a_1 k_1 \left(\frac{\sqrt{3}}{2}\right) + a_1 k_2 \left(-\frac{1}{2}\right) + a_2 k_1 \cdot \left(-\frac{\sqrt{3}}{2}\right) + a_2 k_2 \left(-\frac{1}{2}\right) \right) \sin(\omega_0 u) =$$

$$\left/ \begin{aligned} a_1 &= -1/2 \\ a_2 &= 1/4 \end{aligned} \right/ = \left(k_1 + k_1 \cdot \frac{1}{4} + k_2 \cdot \frac{\sqrt{3}}{4} + k_1 \left(-\frac{1}{8}\right) + k_2 \frac{\sqrt{3}}{8} \right) \cdot \cos(\omega_0 u) +$$

$$\left(k_2 - k_1 \cdot \frac{\sqrt{3}}{4} + k_2 \cdot \frac{1}{4} + k_1 \left(-\frac{\sqrt{3}}{8}\right) + k_2 \left(-\frac{1}{8}\right) \right) \cdot \sin(\omega_0 u) =$$

$$= \frac{1}{2} \cos\left(\frac{2\pi}{3}u\right) - \frac{\sqrt{3}}{2} \sin\left(\frac{2\pi}{3}u\right) \Rightarrow$$

desna strana

Dobivamo sustav od dvije jednačine sa dvije nepoz.

(13)

$$k_1 \left(1 + \frac{1}{4} - \frac{1}{8}\right) + k_2 \left(\frac{\sqrt{3}}{4} + \frac{\sqrt{3}}{8}\right) = \frac{1}{2} \cdot 8$$

$$k_1 \left(-\frac{\sqrt{3}}{4} - \frac{\sqrt{3}}{8}\right) + k_2 \left(1 + \frac{1}{4} - \frac{1}{8}\right) = -\frac{\sqrt{3}}{2} \cdot 8$$

$$9k_1 + 3\sqrt{3}k_2 = 4 \Rightarrow k_1 = \frac{4}{9} - \frac{3\sqrt{3}}{9} \cdot k_2$$

$$-3\sqrt{3}k_1 + 9k_2 = -4\sqrt{3}$$

$$-\frac{12\sqrt{3}}{9} + \frac{27}{9}k_2 + 9k_2 = -4\sqrt{3}$$

$$12k_2 = -4\sqrt{3} + \frac{4}{3}\sqrt{3} = -\frac{8}{3}\sqrt{3}$$

$$k_2 = -\frac{2}{9}\sqrt{3}$$

$$k_1 = \frac{4}{9} + \frac{3\sqrt{3}}{9} \cdot \frac{8}{3} \sqrt{3} \cdot \frac{1}{12}$$

$$k_1 = \frac{4}{9} + \frac{2}{9} = \frac{6}{9} = \frac{2}{3}$$

Dakle odziv glasi:

$$y_{pr}(t) = \frac{2}{3} \cos(\omega_0 t) - \frac{2\sqrt{3}}{9} \sin(\omega_0 t)$$

$$= A_y \cdot \cos(\omega_0 t + \theta_0)$$

$$A_y = \sqrt{\left(\frac{2}{3}\right)^2 + \left(\frac{2\sqrt{3}}{9}\right)^2} = \sqrt{\frac{4}{9} + \frac{4 \cdot 3}{81}} = \sqrt{\frac{4 \cdot 9 + 4 \cdot 3}{81}} = \sqrt{\frac{4 \cdot 4 \cdot 3}{81}} = \frac{4}{9}\sqrt{3}$$

Odredimo početni
kut odziva θ_0

$$k_1 = \frac{2}{3} = A_y \cdot \cos(\theta_0) \Rightarrow \cos(\theta_0) = \frac{k_1}{A_y} = \frac{\frac{2}{3}}{\frac{4}{9}\sqrt{3}} = \frac{3}{2\sqrt{3}} = \frac{\sqrt{3}}{2}$$

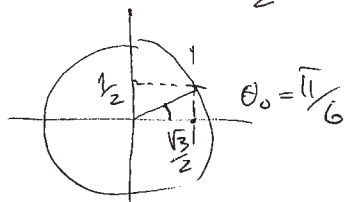
$$k_2 = -\frac{2\sqrt{3}}{9} = -A_y \cdot \sin(\theta_0) \Rightarrow \sin(\theta_0) = \frac{k_2}{-A_y} = \frac{-\frac{2\sqrt{3}}{9}}{-\frac{4\sqrt{3}}{9}} = \frac{1}{2}$$

$$\cos \theta_0 = \frac{\sqrt{3}}{2} > 0 \left. \begin{array}{l} \text{prvi kvadrant} \end{array} \right\}$$

$$\sin \theta_0 = \frac{1}{2} > 0$$

$$\theta_0 = \arctan\left(\frac{\sin \theta_0}{\cos \theta_0}\right) = \arctan\left(\frac{\frac{1}{2}}{\frac{\sqrt{3}}{2}}\right) =$$

$$= \arctan\left(\frac{1}{\sqrt{3}}\right) = \frac{\pi}{6}$$



Datle rekurzivne glasi:

14

$$\begin{aligned} y_p(t) &= A_y \cdot \cos(\omega_0 t + \theta_0) \\ &= \frac{4}{9} \sqrt{3} \cdot \cos\left(\frac{2\pi}{3} t + \frac{\pi}{6}\right) \end{aligned}$$

Dobili smo isto rješenje kao i metodom sa kompleksnom eksponencijalom, ali postupkom koji je nešto duži i složeniji jer zahtjeva razpisivanje sinusa i kosinusa, kao i rješavanje ^{sistema} jed. sa dužje nepoznate k_1, k_2 , te konačno pretvorbu rješenja $k_1 \cos t + k_2 \sin t$ u $A \cdot \cos(\omega_0 t + \theta_0)$

Sada uz poznato partikularno rješenje možemo odrediti odziv mirovog sustava (onog čija su početna stanja $= \emptyset$)
 Taj odziv $y_m[n]$ mora sadržavati i titranje vlastitim frekvencijama ξ_1, ξ_2 jer sustav „nije spreman“ na pobudu koja se pojavila u trenutku $n = \emptyset$, jer su njegova stanja jednaka nuli. Da je pobuda djelovala od $n = -\infty$ na dalje, tada bi stanja stabilnog sustava u trenutku $n = \emptyset$ bila sukladna pobudi, jer bi prikladan popravak davno već otkrila (ustala) te bi nastavio se odazivati frekvencijama pobude.

Međutim kako pobuda djeluje tek od $n = \emptyset$ na dalje... jer $u[n] = (\gamma \circ x)(n)$ dolazi do prikladne popravke pa pismo

$$y_m[n] = c_3 \xi_1^n + c_4 \xi_2^n + \underbrace{y_{pr}[n]}_{\text{ovo smo upravo odredili}}$$

Zbog činjenice da je sustav miran znamo da je

$y_m[n] = \emptyset$ za $n < \emptyset$, a zbog kausalnosti pobudnog signala znamo da je $u[n] = \emptyset$ također za $n < \emptyset$.

Partikularno rješenje $y_{pr}[n]$ vrijedi samo za $n \geq 0$, pa stoga u svrhu određivanja nepoznatih koef. c_3 i c_4 moramo odabrati bilo koja dva uzorka odziva $y_m[n]$ za $n \geq 0$, npr. $n = \emptyset$ i $n = 1$ ili npr. $n = 3$ i $n = 7$. Međutim $y_m[n]$ nije poznat za te n -ove, te ga trebamo odrediti metodom korak po korak.

Znamo da je $y_m(2) = y_m(-1) = 0$ $y_{pr}[n] = \frac{4\sqrt{3}}{9} \cos\left(\frac{2\pi}{3}n + \frac{\pi}{6}\right)$ (16)

Iz jed. dif.

$$u_r[n] = \cos\left(\frac{2\pi}{3}n + \frac{\pi}{3}\right)$$

$$\begin{aligned} y_m(0) &= -a_1 y_m(-1) - a_2 y_m(-2) + u_r(0) \leftarrow u_r(0) = \cos\frac{\pi}{3} = \frac{1}{2} \\ &= \frac{1}{2} y_m(-1) - \frac{1}{4} y_m(-2) + \frac{1}{2} \\ &= \frac{1}{2} \cdot 0 - \frac{1}{4} \cdot 0 + \frac{1}{2} = \frac{1}{2} \end{aligned}$$

$$u_r(1) = \cos(\pi) = -1$$

$$\begin{aligned} y_m(1) &= \frac{1}{2} y_m(0) - \frac{1}{4} y_m(-1) + u_r(1) \\ &= \frac{1}{2} \cdot \frac{1}{2} - \frac{1}{4} \cdot 0 + (-1) = -\frac{3}{4} \end{aligned}$$

Iz pretpostavljamoj opet rješenja za $y_m(n)$ za $n=0$ i 1 sledi:

za $n=0$ $y_m(0) = c_3 \cdot q_1^0 + c_4 \cdot q_2^0 + y_{pr}(0)$

$$y_{pr}(0) = \frac{4\sqrt{3}}{9} \cdot \frac{\sqrt{3}}{2} = \frac{2}{3}$$

za $n=1$ $y_m(1) = c_3 \cdot q_1^1 + c_4 \cdot q_2^1 + y_{pr}(1)$

$$y_{pr}(1) = \frac{4\sqrt{3}}{9} \cdot \cos\left(\frac{5\pi}{6}\right) = -\frac{2}{3}$$

$$\left. \begin{aligned} y_m(0) &= \frac{1}{2} = c_3 + c_4 + \frac{2}{3} \\ y_m(1) &= -\frac{3}{4} = c_3 q_1 + c_4 q_2 - \frac{2}{3} \end{aligned} \right\}$$

$$c_3 + c_4 = -\frac{1}{6}$$

$$c_3 q_1 + c_4 q_2 = -\frac{1}{12}$$

Dvije jedn. sa dvije nep.

$$c_3 = -c_4 - \frac{1}{6}$$

$$-c_4 q_1 - \frac{1}{6} q_1 + c_4 q_2 = -\frac{1}{12}$$

$$c_4 (q_2 - q_1) = -\frac{1}{12} + \frac{1}{6} q_1$$

$$\left. \begin{aligned} 2 \cdot q_1 &= \frac{1}{2} + j\frac{\sqrt{3}}{2} \\ 2 \cdot q_2 &= \frac{1}{2} - j\frac{\sqrt{3}}{2} \end{aligned} \right\} \quad q_2 - q_1 = -j\frac{\sqrt{3}}{2}$$

$$\begin{aligned} -1 + 2q_1 &= -1 + \frac{1}{2} + j\frac{\sqrt{3}}{2} \\ &= j\frac{\sqrt{3}}{2} - \frac{1}{2} \end{aligned}$$

$$c_4 = \frac{1}{12} \cdot \frac{(-1 + 2q_1)}{q_2 - q_1} =$$

$$c_3 = c_4^* = -\frac{\sqrt{3}}{18} \cdot e^{-j\frac{\pi}{6}}$$

$$= \frac{1}{12} \cdot \frac{-\frac{1}{2} + j\frac{\sqrt{3}}{2}}{-j\frac{\sqrt{3}}{2}} \cdot \frac{\frac{2}{\sqrt{3}}j}{\frac{2}{\sqrt{3}}j} =$$

$$= \frac{1}{12} \left(-\frac{1}{\sqrt{3}}j - 1 \right)$$

$$= -\frac{1}{12} \cdot \frac{2}{\sqrt{3}} \left(\frac{\sqrt{3}}{2} + \frac{1}{2}j \right)$$

$$= -\frac{\sqrt{3}}{18} \cdot e^{j\frac{\pi}{6}}$$

Provera:

$$q_1 = \frac{1}{2} \cdot e^{j\frac{\pi}{3}}$$

$$q_2 = \frac{1}{2} \cdot e^{-j\frac{\pi}{3}}$$

$$c_1 = -\frac{\sqrt{3}}{18} \cdot e^{j\frac{\pi}{6}}$$

$$c_3 = -\frac{\sqrt{3}}{18} \cdot e^{-j\frac{\pi}{6}}$$

(17)

$$\begin{aligned} q_1 \cdot c_3 + q_2 \cdot c_1 &= \frac{1}{2} \cdot \left(-\frac{\sqrt{3}}{18}\right) \cdot \left(e^{j\frac{\pi}{3}} \cdot e^{-j\frac{\pi}{6}} + e^{-j\frac{\pi}{3}} \cdot e^{j\frac{\pi}{6}}\right) \\ &= \frac{1}{2} \left(-\frac{\sqrt{3}}{18}\right) \cdot \left(e^{j\frac{\pi}{6}} + e^{-j\frac{\pi}{6}}\right) = \left(-\frac{\sqrt{3}}{18}\right) \cdot \cos\left(\frac{\pi}{6}\right) = -\frac{\sqrt{3}}{18} \cdot \frac{\sqrt{3}}{2} = -\frac{1}{12} \text{ u ok} \end{aligned}$$

$$c_3 + c_1 = -\frac{\sqrt{3}}{18} (e^{j\frac{\pi}{6}} + e^{-j\frac{\pi}{6}}) = -\frac{\sqrt{3}}{18} \cdot 2 \cos\left(\frac{\pi}{6}\right) = -\frac{\sqrt{3}}{18} \cdot 2 \cdot \frac{\sqrt{3}}{2} = -\frac{1}{6} \text{ ok}$$

Dobivamo odziv mirnog sustava vlastitim frekvencijama:

$$y_{m-vlast}[n] = c_3 \cdot q_1^n + c_1 \cdot q_2^n = -\frac{\sqrt{3}}{18} \cdot |g_0|^n \cdot \left(e^{-j\frac{\pi}{6}} \cdot e^{j\frac{\pi}{3} \cdot n} + e^{j\frac{\pi}{6}} \cdot e^{-j\frac{\pi}{3} \cdot n}\right)$$

$$= -\frac{\sqrt{3}}{9} \cdot |g_0|^n \cdot \cos\left(\frac{\pi}{3}n - \frac{\pi}{6}\right) \text{ u ok}$$

$$= \frac{\sqrt{3}}{9} \cdot |g_0|^n \cdot \cos\left(\frac{\pi}{3}n + \frac{5\pi}{6}\right)$$

$$= \frac{\sqrt{3}}{9} \cdot 2^{-n} \cdot \cos\left(\frac{\pi}{3}n + \frac{5\pi}{6}\right)$$

za vjezbu odredite početna stanja $y(-1)$ i $y(-2)$

talva da odziv mirnog sustava vlastitim frekvencijama bude jednak odzivu neprobudjenog sustava, ali suprotnog predznaka tj:

$$y_{m-vlast}[n] = -y_0[n]$$

$$\Rightarrow c_1 = -c_3$$

$$c_2 = -c_4$$

Konačno dobivamo izraz za odziv mirnog sustava

(18)

$$y_m[n] = \underbrace{\frac{\sqrt{3}}{9} 2^{-n} \cdot \cos\left(\frac{\pi}{3}n + \frac{5\pi}{6}\right)}_{\text{titraње vlastitih frekvencija zbog nesklada početnog stanja sustava (nol-stanje) i stanja koje bi odgovaralo sve-vremenskoj pobudi koja djeluje od } -\infty} + \underbrace{\frac{4\sqrt{3}}{9} \cdot \cos\left(\frac{2\pi}{3}n + \frac{\pi}{6}\right)}_{\text{titraње frekvencijama pobude}}$$

titraње
frekvencijama
pobude

⇒ prisilni
odziv

Odziv nepobudnog sustava sa početnim stanjima:

$y(-1) = 2$ & $y(-2) = 8$ bio je:

$$y_0[n] = 2^{1-n} \cdot \cos\left(n \cdot \frac{\pi}{3} + \frac{2\pi}{3}\right)$$

↑ ista frekvencija kao i ,
dli druga početna faza i
druga amplituda !!!

Označimo ovaj dio odziva sa $y_{m-vlast}[n]$

$y_{m-vlast}[n]$ i $y_0[n]$ se mogu „udružiti“ u jedan signal frekvencije $f_0 = \pi/3$, a neke nove amplitude i početne faze. Ovo udruživanje je najlakše provesti zbiranjem c-koeficijenta

prisilni odziv

$$y_{m-vlast}[n] + y_0[n] = (c_1 + c_3) z_1^n + (c_2 + c_4) z_2^n = y_{prisil}[n]$$

$$\left. \begin{matrix} c_1 = c_2^* \\ c_3 = c_4^* \end{matrix} \right\} \Rightarrow (c_1 + c_3) = (c_2 + c_4)^*$$

od prije sa strane 4 znamo da je

(19)

$$c_1 = -\frac{1}{2} + \frac{\sqrt{3}}{2}j$$

a na strani 16 imamo c_3 i c_4

$$c_3 = -\frac{\sqrt{3}}{18} \cdot e^{-j\frac{\pi}{6}} = -\frac{\sqrt{3}}{18} \left(\frac{\sqrt{3}}{2} - j\frac{1}{2} \right) = -\frac{1}{12} + j\frac{\sqrt{3}}{36}$$

Dakle $c_1 + c_3 = -\frac{1}{2} + \frac{\sqrt{3}}{2}j - \frac{1}{12} + j\frac{\sqrt{3}}{36} = -\frac{7}{12} + \frac{19}{36}j\sqrt{3}$ ili

$$|c_1 + c_3| = \sqrt{\frac{49}{144} + \frac{19 \cdot 19 \cdot 3}{36 \cdot 36 \cdot 12}} = \sqrt{\frac{3 \cdot 49 + 19 \cdot 19}{3 \cdot 144}} = \sqrt{\frac{508}{3 \cdot 144}} = \frac{2\sqrt{127}}{12 \cdot \sqrt{3}} = \frac{1}{6} \sqrt{\frac{127}{3}}$$

$$\angle(c_1 + c_3) = \operatorname{atan2}(\operatorname{Im}, \operatorname{Re}) = \operatorname{atan2}\left(\frac{19\sqrt{3}}{36}, -\frac{7}{12}\right) \\ = 0.6808 \pi$$

$$|c_2 + c_4| = |c_1 + c_3| = \frac{1}{6} \sqrt{\frac{127}{3}}$$

$$\angle(c_2 + c_4) = -\angle(c_1 + c_3)$$

$$y_{\text{pri}}(n) = |c_1 + c_3| \cdot e^{j\angle(c_1 + c_3)} |g_0|^n \cdot e^{j80n} + |c_2 + c_4| \cdot e^{j\angle(c_2 + c_4)} |g_0|^n \cdot e^{-j80n}$$

$$= \frac{1}{3 \cdot 6} \sqrt{\frac{127}{3}} \cdot 2^{-n} \cdot 2 \cos(80n + \angle(c_1 + c_3))$$

$$= \frac{1}{3} \sqrt{\frac{127}{3}} \cdot 2^{-n} \cos\left(80n + \operatorname{atan2}\left(\frac{19\sqrt{3}}{36}, -\frac{7}{12}\right)\right) \quad \text{ili}$$

ukupno titranje vlastitim frekvencijama
ili prirodni odziv sustava

Traženi Matlab program

```
clear;
n_max=30;
% vremenska os
n=[0:n_max];

% Odziv mirnog sustava vlastitim frekvencijama
ym_vlast=sqrt(3)/9*(2.^(-n)).*cos(pi/3*n+5*pi/6);

% Odziv nepobudjenog sustava vlastitim frekvencijama
y0=(2.^(1-n)).*cos(pi/3*n+2*pi/3);

% Pocetna faza ukupnog titranja vlastitim frekvencijama
an=atan2(19*sqrt(3)/36,-7/12);

% Ukupno titranje vlastitim frekvencijama
y_prir=sqrt(127/3)/6*(2.^(1-n)).*cos(pi/3*n+an);

% pokazi da je y_prir jednak sumi ym_vlast i y0
disp(max(abs(y_prir-(y0+ym_vlast))))

% partikularno rjesenje ... titranje frekvencijama pobude
y_par=4*sqrt(3)/9*cos(2*pi/3*n+pi/6);

% totalni odziv (suma titranja vlastitim frekvencijama i
% prisilnog odziva)
ytot=y_prir+y_par;

% Ukupni odziv mirnog sustava
y_mir=ym_vlast+y_par;

% Pobuda sustava
u=1*cos(2*pi/3*n+pi/3);

% Inicijaliziraj vektor odziva za metodu korak po korak
y=0*u;

% Odredi odziv metodom korak po korak
y_nm_2=8;      % pocetna vrijednost y(n-2) za n=0 ... y(-2)
y_nm_1=2;      % pocetna vrijednost y(n-1) za n=0 ... y(-1)

% koeficijenti jedn. dif. uz y(n-1) i y(n-2)
a1=-1/2;
a2=1/4;

% koef. jedn. dif. uz u(n)
b0=1;

for nn=0:n_max,
    % Zbog Matlab-a koji ne pozna indekse polja koji su manji od
    % jedan ... vremenski indeks n=0 pretvaramo u indeks polja 1
    nM=nn+1;
    % Jednadzba diferencije ...
    ykk(nM)=-a1*y_nm_1 -a2*y_nm_2 +b0*u(nM);

    % za novi prolaz petlje (n+1)
    y_nm_2=y_nm_1;      % y((n+1)-2) = y(n-1)
    y_nm_1=ykk(nM);     % y((n+1)-1) = y(n)
end;

% Usporedi analiticko rjesenje i ono dobiveno metodom korak po korak
disp(max(abs(ykk-ytot)))
```

```

figure(1);
stem(n,u)
title('Pobuda sustava');
grid;

figure(2);
stem(n,y0)
title('Odziv nepobudjenog sustava');
grid;

figure(3);
stem(n,ym_vlast)
title('Odziv mirnog sustava vlastitim frekvencijama');
grid;

figure(4);
stem(n,y_par)
title('Prisilni odziv sustava');
grid;

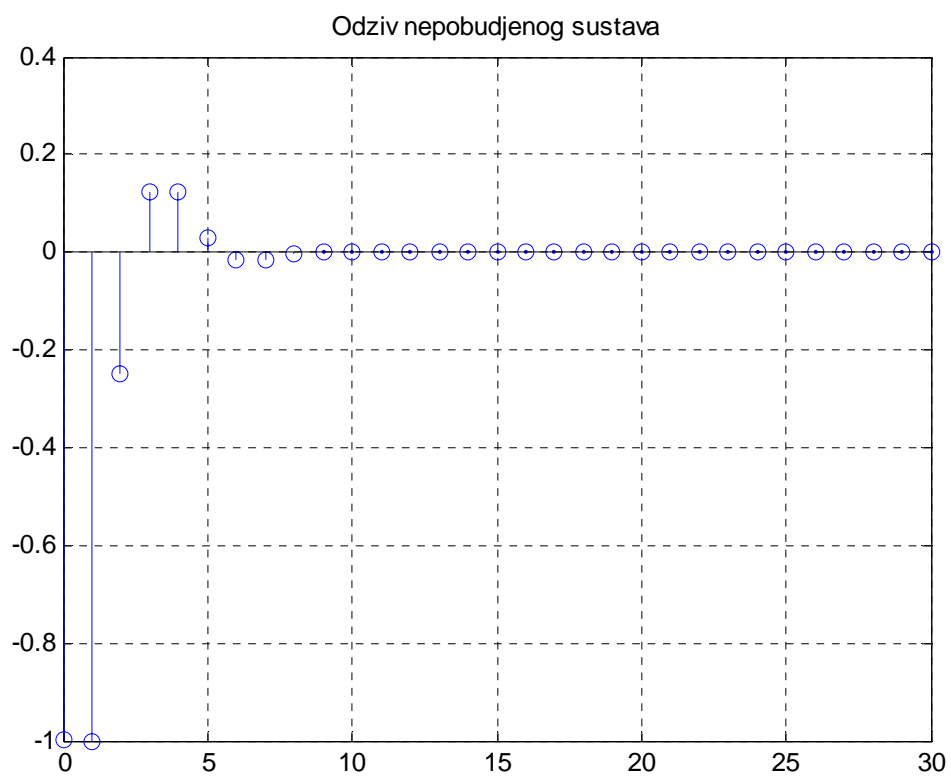
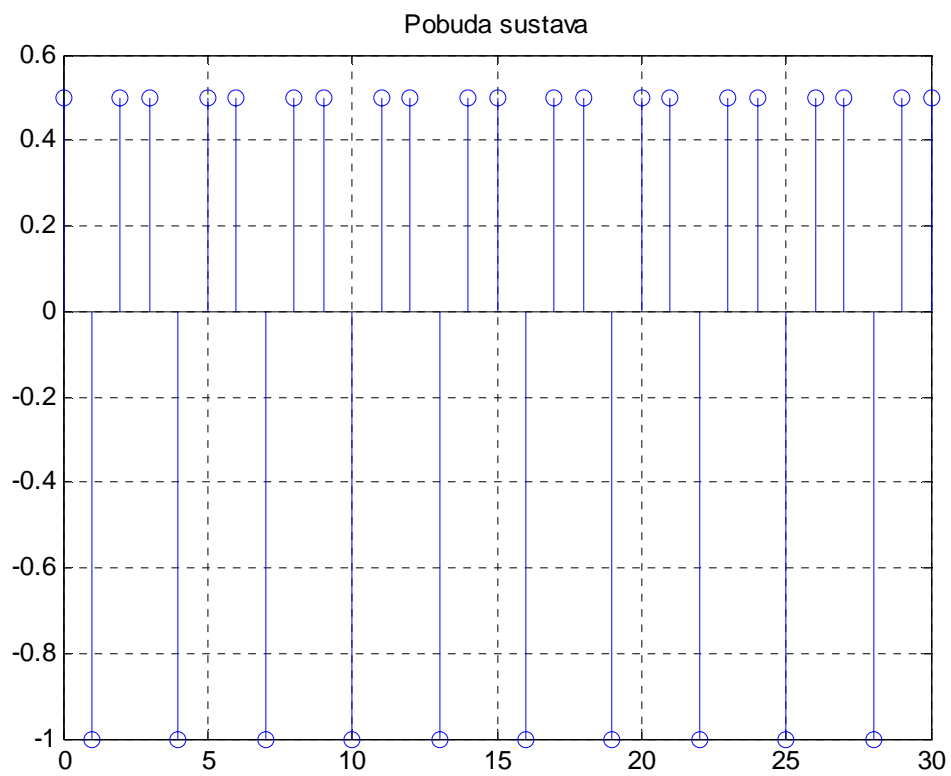
figure(5);
stem(n,y_prir)
title('Prirodni odziv sustava');
grid;

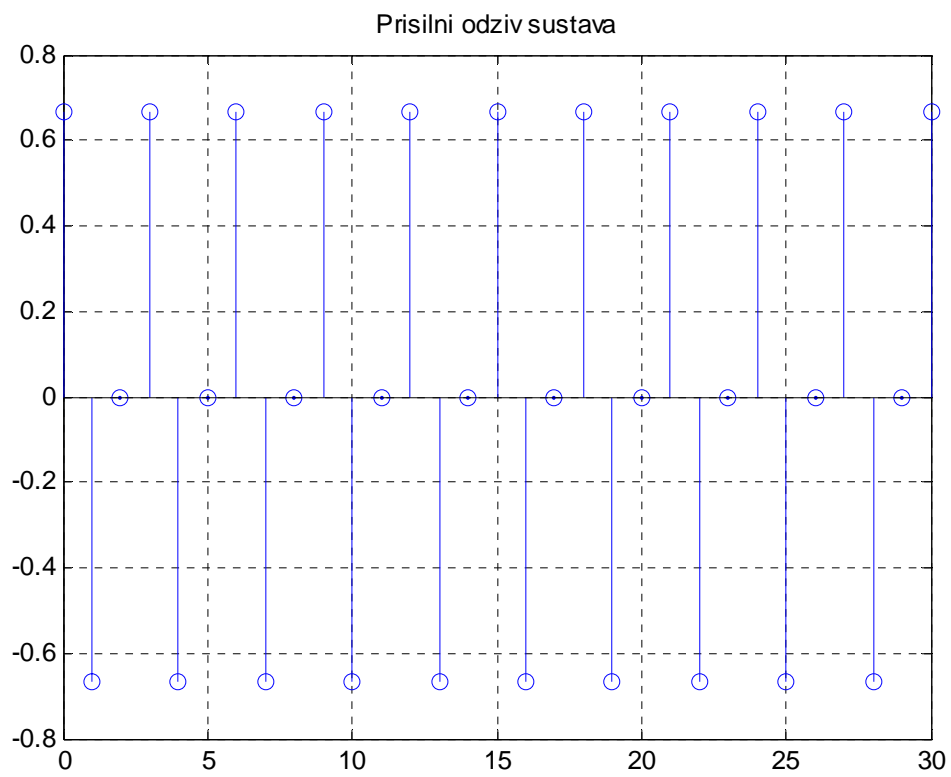
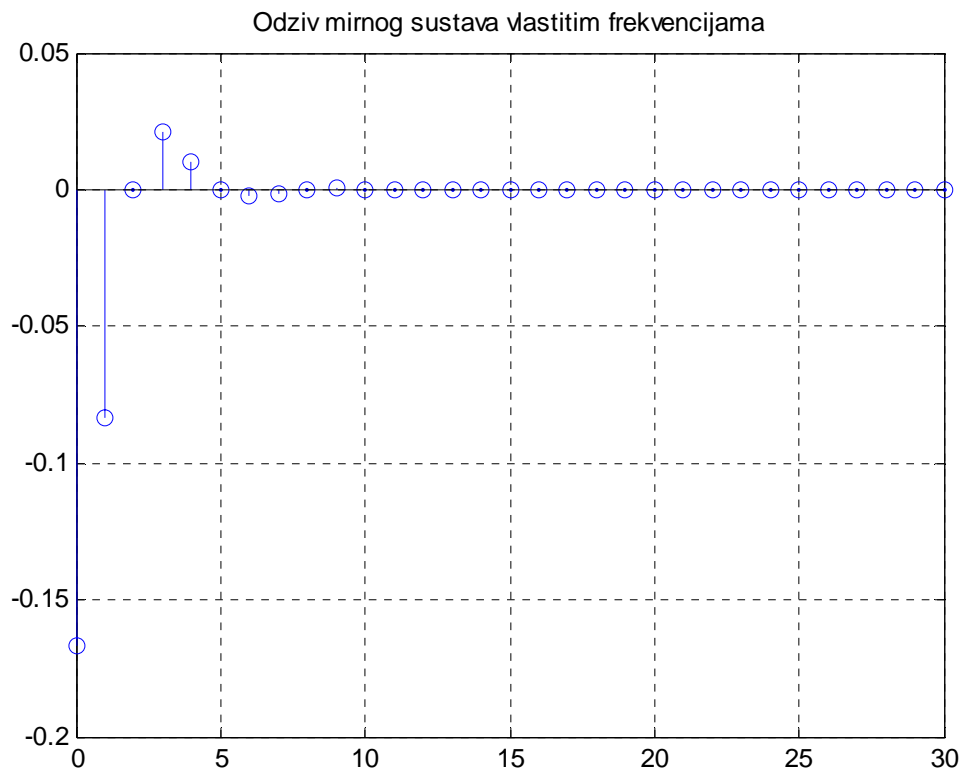
figure(6);
stem(n,y_mir)
title('Ukupni odziv mirnog sustava');
grid;

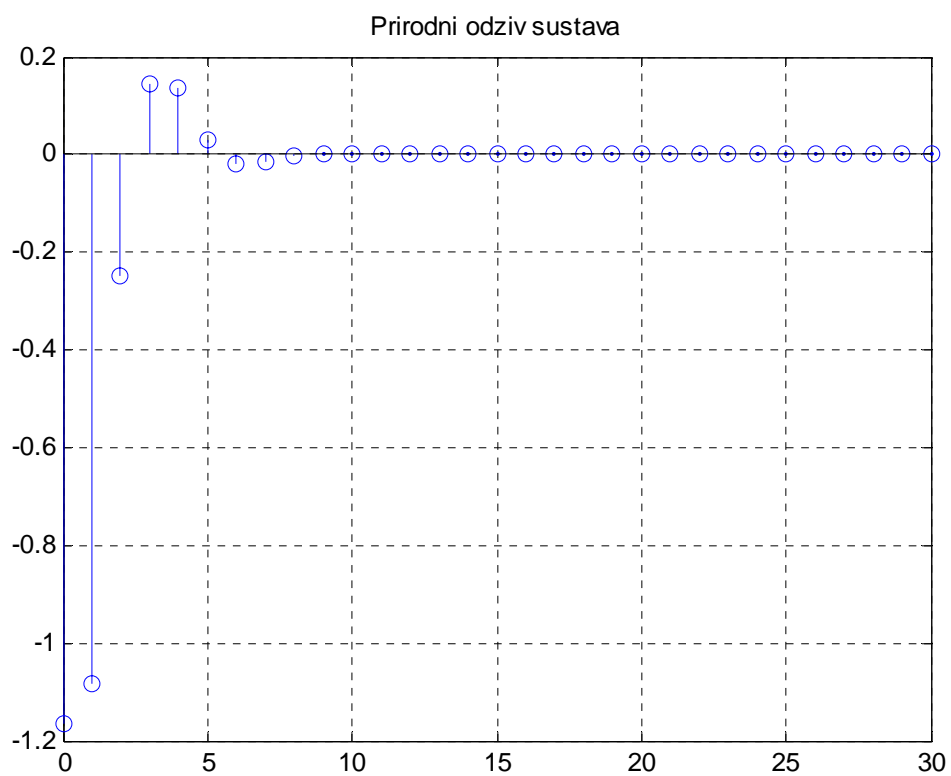
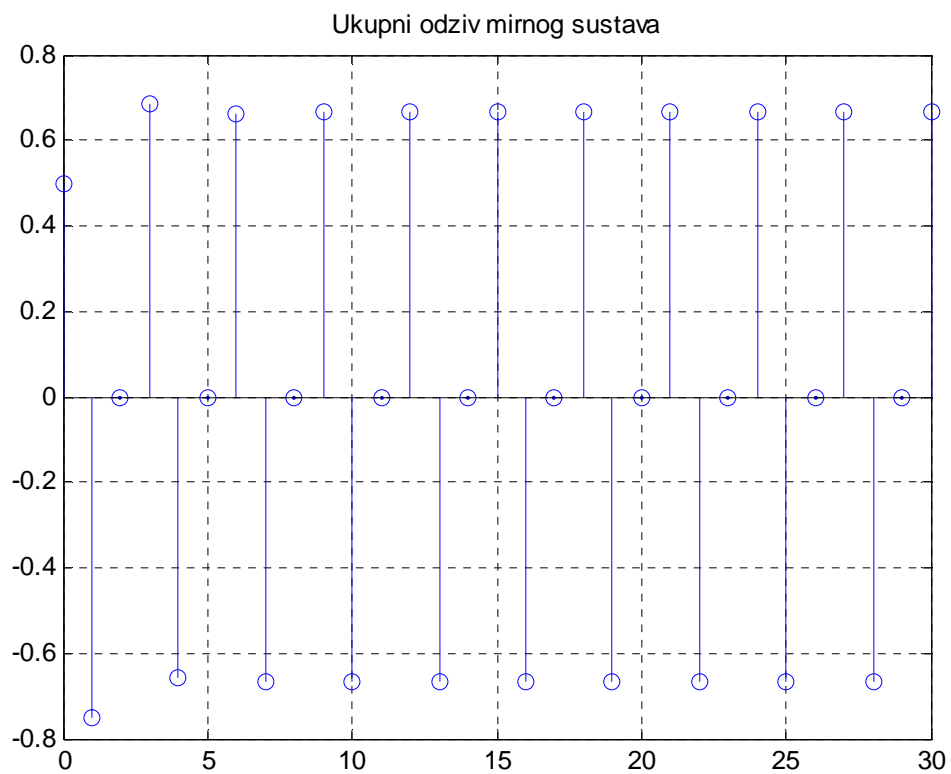
figure(7);
stem(n,ytot)
title('Ukupni odziv sustava - analiticko rjesenje');
grid;

figure(8);
stem(n,ykk)
title('Ukupni odziv sustava - korak po korak');
grid;

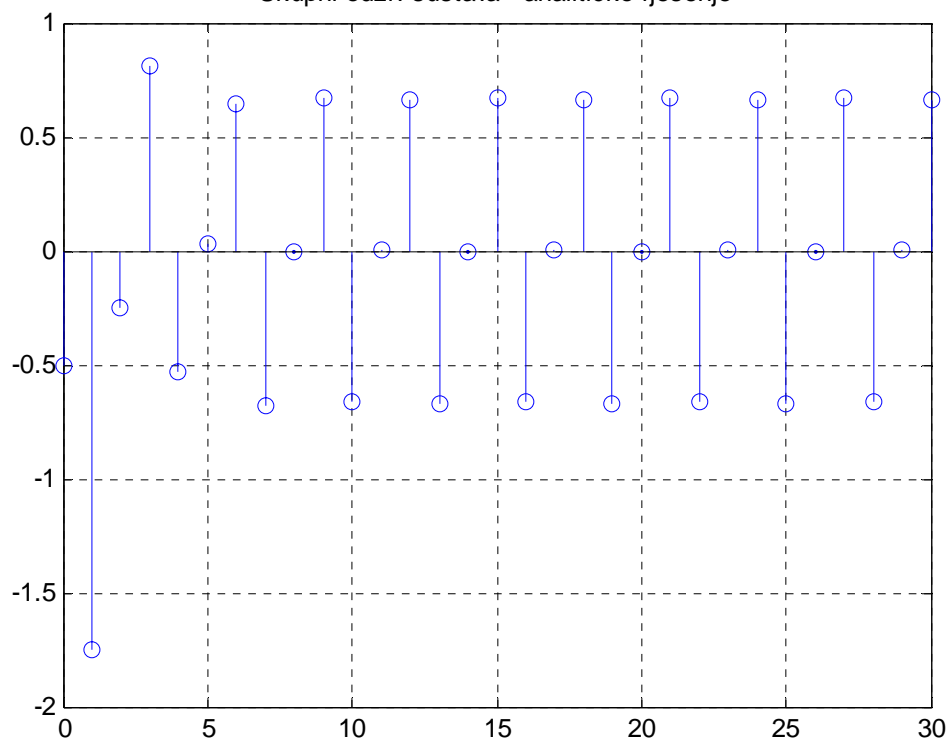
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Ukupni odziv sustava - analiticko rjesenje



Ukupni odziv sustava - korak po korak

