

## Chapter 3 (Antenna Fundamentals) - Exercise Solutions

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1. First, the effective area of the aperture is

$$A_e = \eta A_p = 0.65\pi(15 \times 10^{-2}\text{m})^2 = 45.95 \times 10^{-3} \text{ m}^2.$$

Thus the gain  $G$  is

$$G = \frac{4\pi A_e}{\lambda^2} = \frac{4\pi(45.95 \times 10^{-3} \text{ m}^2)}{(7.5 \times 10^{-3} \text{ m})^2} = 10265.33.$$

In decibels,

$$10 \log 10265.33 = 40.11\text{dB}.$$

2. As before, we just use the formula

$$G = \frac{4\pi\eta A_p f^2}{c^2} = \frac{4\pi(0.6)0.2 \text{ m}^2(10 \times 10^9 \text{ s}^{-1})}{(3 \times 10^8 \text{ m s}^{-1})^2} = 1675.52 = 32.24 \text{ dB}.$$

3. A loss of 0.5 dB is equivalent to a loss of  $10^{-0.05} = 0.891$  in absolute terms. Thus, for this system  $P_t = 0.891P_i$ . Then, for  $(1 - \rho^2) = 0.891$  we have that  $\rho = \pm 0.33$ . Taking each value in turn:

$$\begin{aligned} 0.33(Z_1 + Z_0) &= Z_1 - Z_0 \\ \implies Z_1 &\approx \frac{1.33}{0.67}Z_0 = 99.25\Omega \end{aligned}$$

or

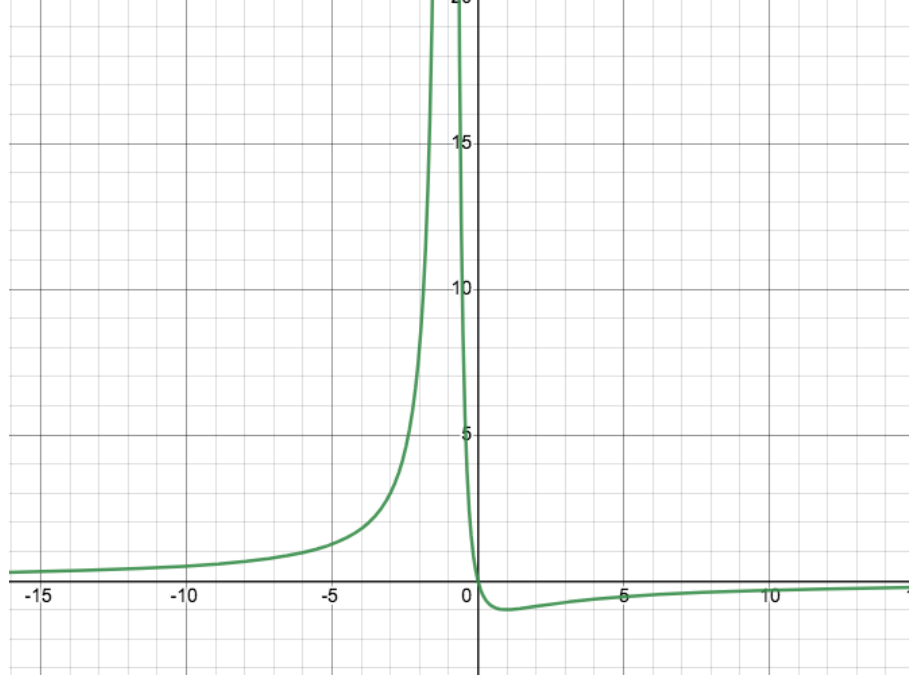
$$\begin{aligned} -0.33(Z_1 + Z_0) &= Z_1 - Z_0 \\ \implies Z_1 &\approx \frac{0.67}{1.33}Z_0 = 25.19\Omega. \end{aligned}$$

So as long as the driving point impedance is between  $25.19\Omega$  and  $99.25\Omega$ , then the system will stay within the 0.5 dB loss value.

4. We can show (by simply rearranging variables) that the loss can be rewritten as

$$L = -\left[1 - \left(\frac{\text{VSWR} - 1}{\text{VSWR} + 1}\right)^2\right].$$

So we simply plot VSWR (x axis) against loss (y axis), noting that there is formally undefined (infinite) loss when  $\text{VSWR} = -1$ .



5. Using the formula

$$F = \frac{(1 + \text{AR}_w^2)(1 + \text{AR}_r^2) + 4\text{AR}_w\text{AR}_r + (1 - \text{AR}_w^2)(1 + \text{AR}_r^2) \cos(2[\tau_w - \tau_r])}{2(1 + \text{AR}_w^2)(1 + \text{AR}_r^2)}$$

Since  $\text{AR}_{w,dB} = \text{AR}_{r,dB} = 2\text{db}$ , then  $\text{AR}_w = \text{AR}_r = 10^{\frac{2}{20}} = 1.259$ , and the above equation reduces to

$$F = 0.918 + 0.0256 \cos[2(\tau_w - \tau_r)].$$

Where  $P_r = FP_i$ , the minimum *value* (most loss) of  $F$  should therefore occur when  $\cos[2(\tau_w - \tau_r)] = -1$ , and the *max* (least loss) should occur when  $\cos[2(\tau_w - \tau_r)] = 1$ . So

$$F_{min} = 0.918 - 0.0256 = 0.8924 = -0.494 \text{ dB}$$

and

$$F_{max} = 0.918 + 0.0256 = 0.9436 = -0.252 \text{ dB}$$

So when the angle between the same axes of each antenna is rotated either  $90^\circ$  or  $270^\circ$ , we experience the maximum loss of  $-0.494 \text{ dB}$ . When  $0^\circ$  or  $180^\circ$ , we experience a minimum loss of  $-0.252 \text{ dB}$ . For all angles in between, the loss will vary sinusoidally between these two values.

6. To get the effective height,  $h_e$ , we first must consider the average value of a sin wave. By definition, this is an integral over the period  $[0, \pi]$

$$I_{avg} = \frac{I_0}{\pi} \int_0^\pi \sin x \, dx = \frac{I_0}{\pi} [-\cos x]_0^\pi = \frac{2}{\pi} I_0.$$

$h_e$  is then given by

$$h_{eff} = \frac{1}{I_0} \int_0^L I_{avg} dx = \frac{1}{I_0} \int_0^L \frac{2}{\pi} I_0 dx = \frac{2}{\pi} (L - 0) \approx 0.637L$$

where  $L$  is the physical length of the antenna. In this case,

$$L = 0.1 \frac{3 \times 10^8 \, \text{m s}^{-1}}{80 \times 10^6 \, \text{s}^{-1}} = 0.375 \, \text{m}.$$

Thus

$$h_e = 0.637 \times 0.375 = 0.239$$

and the effective area  $A_e$  is

$$A_e = \frac{h_e^2 Z_0}{4R_r} = \frac{(0.239)^2 377 \Omega}{4 \times 73 \Omega} = 0.074 \, \text{m}^2.$$

7. We simply use Fig. 3.2 of the text. The lowest point outside of the mainlobe is  $-80$  dB, and its independent of distance. Since

$$100 \, \text{dBW} - 80 \, \text{dB} = 20 \, \text{dBW}$$

we can then calculate the power as

$$P = 10^{\frac{100}{10}} = 10^{10} \, \text{W}.$$

Since we were asked for power *density*  $S$ , then

$$S = \frac{P}{4\pi d^2} = \frac{10^{10}}{4\pi 2^2} = 198.94 \times 10^6 \, \text{W m}^{-2}$$