









ISO6740-Q1, ISO6741-Q1, ISO6742-Q1 SLLSFG4E - DECEMBER 2019 - REVISED MAY 2022

ISO674x-Q1 General-Purpose Reinforced Quad-Channel Automotive Digital Isolators with Robust EMC

1 Features

- Functional Safety-Capable
 - Documentation available to aid functional safety system design: ISO6740-Q1, ISO6741-Q1, ISO6742-Q1
- AEC-Q100 qualified with the following results:
 - Device temperature Grade 1: –40°C to +125°C ambient operating temperature range
- Meets VDA320 isolation requirements
- 50 Mbps data rate
- Robust isolation barrier:
 - High lifetime at 1500 V_{RMS} working voltage
 - Up to 5000 V_{RMS} isolation rating
 - Up to 10 kV surge capability
 - ±150 kV/µs typical CMTI
- Wide supply range: 1.71 V to 1.89 V and 2.25 V to
- 1.71 V to 5.5 V level translation
- Default output high (ISO674x-Q1) and low (ISO674xF-Q1) options
- 1.6 mA per channel typical at 1 Mbps
- Low propagation delay: 11 ns typical
- Robust electromagnetic compatibility (EMC)
 - System-level ESD, EFT, and surge immunity
 - ±8 kV IEC 61000-4-2 contact discharge protection across isolation barrier
 - Low emissions
- Wide-SOIC (DW-16) Package
- Safety-Related Certifications:
 - DIN VDE V 0884-11:2017-01
 - UL 1577 component recognition program
 - IEC 62368-1, IEC 61010-1, IEC 60601-1
 - GB 4943.1-2011 (pending)

2 Applications

- Hybrid, electric and power train system (EV/HEV)
 - Battery management system (BMS)
 - On-board charger
 - DC/DC converter
 - Inverter and motor control

3 Description

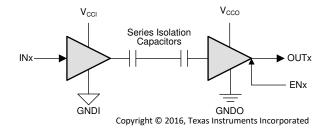
The ISO674x-Q1 devices are high-performance. quad-channel digital isolators ideal for cost-sensitive applications requiring up to 5000 V_{RMS} isolation ratings per UL 1577. These devices are also certified by VDE, TUV, CSA, and CQC.

The ISO674x-Q1 devices provide hiah electromagnetic immunity and low emissions at low power consumption, while isolating CMOS or LVCMOS digital I/Os. Each isolation channel has a logic input and output buffer separated by TI's double capacitive silicon dioxide (SiO₂) insulation barrier. These devices come with enable pins which can be used to put the respective outputs in high impedance for multi-master driving applications. The ISO6740-Q1 device has all four channels in the same direction, the ISO6741-Q1 device has three forward and one reverse-direction channels, and the ISO6742-Q1 device has two forward and two reversedirection channels. In the event of input power or signal loss, the default output is high for devices without suffix F and low for devices with suffix F. See Device Functional Modes section for further details.

Device Information

PART NUMBER ⁽¹⁾	PACKAGE	BODY SIZE (NOM)			
ISO6740-Q1, ISO6740F-Q1	SOIC (DW)	10.30 mm × 7.50 mm			
ISO6741-Q1, ISO6741F-Q1					
ISO6742-Q1, ISO6742F-Q1					

For all available packages, see the orderable addendum at the end of the data sheet.



V_{CCI}=Input supply, V_{CCO}=Output supply GNDI=Input ground, GNDO=Output ground

Simplified Schematic



Typical Characteristics
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Changes from Revision D (July 2021) to Revision E (May 2022)	Page
 Updated CMTI typical to 150 kV/us and minimum to 100 kV/us. Switched the labels for V_{CC1} falling and V_{CC2} rising in the graph legend of <i>Power Supply Undervoltage</i> 	<mark>7</mark>
Threshold vs Free-Air Temperature	25
Changes from Revision C (March 2021) to Revision D (July 2021)	Page
Updated high lifetime working voltage	
 Insulation Specifications table 7.6 has been updated with VIOWM 1500Vrms, VIORM at 2121Vpk 	
Updated Safety-Related Certification table	<mark>7</mark>
 Updated switching characteristics tables with test conditions for "Default output delay time from input por 	
loss" line item	
Updated Typical Application diagram to reflect 5.5Viso	
Updated Insulation Lifetime Projection Data image	34
Updated SN6505A reference with SN6505B in Power Supply Recommendations	36
Changes from Revision B (February 2021) to Revision C (March 2021)	Page
Added ISO6742-Q1 data under Specifications	7
Updated Typical Application figure	



Changes from Revision A (January 2021) to Revision B (February 2021)		
Updated device status to Production Data	1	
Changes from Revision * (August 2020) to Revision A (January 2021)	Page	
Added ISO674x-Q1 to APL data sheet.	1	



5 Description Continued

Used in conjunction with isolated power supplies, the ISO674x-Q1 devices help prevent noise currents on data buses, such as CAN and LIN from damaging sensitive circuitry. Through innovative chip design and layout techniques, the electromagnetic compatibility of the ISO674x-Q1 devices has been significantly enhanced to ease system-level ESD, EFT, surge, and emissions compliance. The ISO674x-Q1 family of devices is available in a 16-pin SOIC wide-body (DW) package and is a pin-to-pin upgrade to the older generations.



6 Pin Configuration and Functions

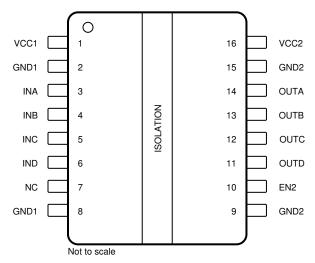


Figure 6-1. ISO6740-Q1 DW Package 16-Pin SOIC-WB Top View

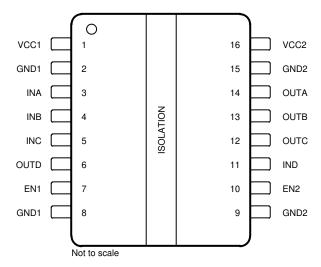


Figure 6-2. ISO6741-Q1 DW Package 16-Pin SOIC-WB Top View

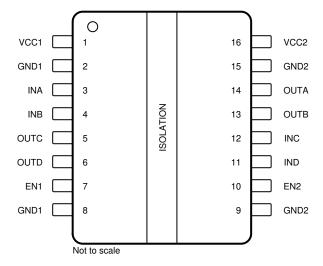


Figure 6-3. ISO6742-Q1 DW Package 16-Pin SOIC-WB Top View



Table 6-1. Pin Functions

Table 6-1.1 III I directions						
PIN				I/O	DESCRIPTION	
NAME	ISO6740-Q1	ISO6741-Q1	ISO6742-Q1	"/0	DESCRIPTION	
EN1	-	7	7	I	Output enable 1. Output pins on side 1 are enabled when EN1 is high or open and in high-impedance state when EN1 is low.	
EN2	10	10	10	I	Output enable 2. Output pins on side 2 are enabled when EN2 is high or open and in high-impedance state when EN2 is low.	
GND1	2, 8	2,8	2,8	_	Ground connection for V _{CC1}	
GND2	9, 15	9,15	9,15	_	Ground connection for V _{CC2}	
INA	3	3	3	ı	Input, channel A	
INB	4	4	4	ı	Input, channel B	
INC	5	5	12	I	Input, channel C	
IND	6	11	11	I	Input, channel D	
NC	7	-	-		Not connected	
OUTA	14	14	14	0	Output, channel A	
OUTB	13	13	13	0	Output, channel B	
OUTC	12	12	5	0	Output, channel C	
OUTD	11	6	6	0	Output, channel D	
V _{CC1}	1	1	1	_	Power supply, side 1	
V _{CC2}	16	16	16	_	Power supply, side 2	



7 Specifications

Updated switching characteristics tables with test conditions for "Default output delay time from input power loss" line item

7.1 Absolute Maximum Ratings

See(1)

		MIN	MAX	UNIT
Supply voltage (2)	V _{CC1} to GND1	-0.5	6	V
Supply voltage V	V _{CC2} to GND2	-0.5	6	V
Input/Output	INx to GNDx	-0.5	$V_{CCX} + 0.5$ (3)	V
Voltage	OUTx to GNDx	-0.5	V _{CCX} + 0.5 (3)	V
Output current	lo	-15	15	mA
Temperature	Operating junction temperature, T _J		150	°C
	Storage temperature, T _{stg}	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values
- (3) Maximum voltage must not exceed 6 V.

7.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ ESDA/JEDEC JS-001, all pins ⁽¹⁾	±6000	
V _(ESD)	V _(ESD) Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±1500	V
		Contact discharge per IEC 61000-4-2; Isolation barrier withstand test ⁽³⁾ (4)	±8000	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.
- (3) IEC ESD strike is applied across the barrier with all pins on each side tied together creating a two-terminal device.
- (4) Testing is carried out in air or oil to determine the intrinsic contact discharge capability of the device.



7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V _{CC1} (1)	Supply Voltage Side 1	V _{CC} = 1.8 V ⁽³⁾	1.71		1.89	V
V _{CC1} (1)	Supply Voltage Side 1	$V_{CC} = 2.5 \text{ V to 5 V}^{(3)}$	2.25		5.5	V
V _{CC2} (1)	Supply Voltage Side 2	$V_{CC} = 1.8 \text{ V}^{(3)}$	1.71		1.89	V
V _{CC2} (1)	Supply Voltage Side 2	V _{CC} = 2.5 V to 5 V ⁽³⁾	2.25		5.5	V
Vcc (UVLO+)	UVLO threshold when supply	voltage is rising		1.53	1.71	V
Vcc (UVLO-)	UVLO threshold when supply	voltage is falling	1.1	1.41		V
Vhys (UVLO)	Supply voltage UVLO hystere	sis	0.08	0.13		V
V _{IH}	High level Input voltage		0.7 x V _{CCI}		V _{CCI}	V
V _{IL}	Low level Input voltage		0		0.3 x V _{CCI}	V
		$V_{CCO} = 5 V^{(2)}$	-4			mA
		V _{CCO} = 3.3 V	-2			mA
ОН	High level output current	V _{CCO} = 2.5 V	-1			mA
		V _{CCO} = 1.8 V	-1			mA
		V _{CCO} = 5 V			4	mA
	Low lovel output ourrent	V _{CCO} = 3.3 V			2	mA
OL L	Low level output current	V _{CCO} = 2.5 V			1	mA
		V _{CCO} = 1.8 V			1	mA
OR .	Data Rate		0		50	Mbps
ΓΑ	Ambient temperature		-40	25	125	°C

 ⁽¹⁾ V_{CC1} and V_{CC2} can be set independent of one another
 (2) V_{CCI} = Input-side V_{CC}; V_{CCO} = Output-side V_{CC}
 (3) The channel outputs are in undetermined state when 1.89 V < V_{CC1}, V_{CC2} < 2.25 V and 1.05 V < V_{CC1}, V_{CC2} < 1.71 V



7.4 Thermal Information

		ISO674x	
	THERMAL METRIC(1)	DW (SOIC)	UNIT
		16 PINS	
R _{0JA}	Junction-to-ambient thermal resistance	73	°C/W
R _{0JC(top)}	Junction-to-case (top) thermal resistance	36.1	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	40.4	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	17	°C/W
ΨЈВ	Junction-to-board characterization parameter	39.9	°C/W
R _{0JC(bot)}	Junction-to-case (bottom) thermal resistance	_	°C/W

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

7.5 Power Ratings

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
ISO674	10					
P _D	Maximum power dissipation (both sides)	V _{CC1} = V _{CC2} = 5.5 V, T _J = 150°C, C _L =			130.9	mW
P _{D1}	Maximum power dissipation (side-1)	15 pF, Input a 25-MHz 50% duty cycle			33	mW
P _{D2}	Maximum power dissipation (side-2)	square wave			97.9	mW
ISO674	11		•			
P _D	Maximum power dissipation (both sides)	V _{CC1} = V _{CC2} = 5.5 V, T _{.I} = 150°C, C _I =			134.9	mW
P _{D1}	Maximum power dissipation (side-1)	15 pF, Input a 25-MHz 50% duty cycle			50.8	mW
P _{D2}	Maximum power dissipation (side-2)	square wave			84.1	mW
ISO674	12		•			
P_D	Maximum power dissipation (both sides)	V _{CC1} = V _{CC2} = 5.5 V, T _{.I} = 150°C, C _I =			137.5	mW
P _{D1}	Maximum power dissipation (side-1)	15 pF, Input a 25-MHz 50% duty cycle			68.75	mW
P _{D2}	Maximum power dissipation (side-2)	square wave			68.75	mW



7.6 Insulation Specifications

		VALUE		
	PARAMETER	TEST CONDITIONS	DW-16	UNIT
CLR	External clearance ⁽¹⁾	Shortest terminal-to-terminal distance through air	>8	mm
CPG	External creepage ⁽¹⁾	Shortest terminal-to-terminal distance across the package surface	>8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	um
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	>600	V
	Material group	According to IEC 60664-1	I	
	Over-18 to 18 and 18 to	Rated mains voltage ≤ 600 V _{RMS}	I-IV	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 1000 V _{RMS}	1-111	1
DIN VDE	V 0884-11:2017-01 ⁽²⁾		-	
V _{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	2121	V _{PK}
V _{IOWM}	Maximum working isolation voltage	AC voltage; Time dependent dielectric breakdown (TDDB) Test; See Figure 10-8	1500	V _{RMS}
		DC voltage	2121	V _{DC}
V_{IOTM}	Maximum transient isolation voltage	V _{TEST} = V _{IOTM} , t = 60 s (qualification); V _{TEST} = 1.2 x V _{IOTM} , t= 1 s (100% production)	7071	V _{PK}
V _{IOSM}	Maximum surge isolation voltage ⁽³⁾	Test method per IEC 62368-1, 1.2/50 μ s waveform, V_{TEST} = 1.6 x V_{IOSM} = 10,000 V_{PK} (qualification)	6250	V _{PK}
		Method a, After Input-output safety test subgroup 2/3, $V_{ini} = V_{IOTM}$, $t_{ini} = 60 \text{ s}$; $V_{pd(m)} = 1.2 \text{ x } V_{IORM}$, $t_m = 10 \text{ s}$	≤5	
q _{pd}	Apparent charge ⁽⁴⁾	Method a, After environmental tests subgroup 1, $V_{ini} = V_{IOTM}$, $t_{ini} = 60 \text{ s}$; $V_{pd(m)} = 1.6 \text{ x } V_{IORM}$, $t_m = 10 \text{ s}$	≤5	pC
		Method b; At routine test (100% production) and preconditioning (type test) $V_{\text{ini}} = 1.2 \text{ x } V_{\text{IOTM}}, t_{\text{ini}} = 1 \text{ s}; \\ V_{\text{pd(m)}} = 1.875 \text{ x } V_{\text{IORM}}, t_{\text{m}} = 1 \text{ s}$	≤5	
C _{IO}	Barrier capacitance, input to output ⁽⁵⁾	$V_{IO} = 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}$	~1	pF
		V _{IO} = 500 V, T _A = 25°C	>10 ¹²	
R_{IO}	Isolation resistance ⁽⁵⁾	V _{IO} = 500 V, 100°C ≤ T _A ≤ 125°C	>10 ¹¹	Ω
		V _{IO} = 500 V at T _S = 150°C	>109	1
	Pollution degree		2	
	Climatic category		40/125/21	
UL 1577	•		'	
V _{ISO}	Maximum withstanding isolation voltage	$V_{TEST} = V_{ISO}$, t = 60 s (qualification), $V_{TEST} = 1.2 \text{ x } V_{ISO}$, t = 1 s (100% production)	5000	V _{RMS}

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed-circuit board are used to help increase these specifications.
- (2) This coupler is suitable for *safe electrical insulation* only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Testing is carried out in air or oil to determine the intrinsic surge immunity of the isolation barrier.
- (4) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (5) All pins on each side of the barrier tied together creating a two-terminal device.



7.7 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Certified according to DIN VDE V 0884-11:2017- 01	Certified according to IEC 62368-1, IEC 61010-1 and IEC 60601	Certified according to UL 1577 Component Recognition Program	Plan to certify according to GB4943.1-2011	Certified according to EN 61010-1:2010/ A1:2019 and EN 62368-1:2014
Maximum transient isolation voltage, 7071 V _{PK} ; Maximum repetitive peak isolation voltage, 2121 V _{PK} ; Maximum surge isolation voltage, 6250 V _{PK}	5000 V _{RMS} insulation per CSA 62368-1:19, IEC 62368-1:2018, CSA 61010-1-12+A1 and IEC 61010-1 3rd Ed., 1000 V _{RMS} basic and 600 V _{RMS} reinforced working voltage (pollution degree 2, material group I); 5000 V _{RMS} insulation per CSA 60601-1-14 and IEC 60601-1 Ed.3+A1, 2 MOPP for 250 V _{RMS}	Single protection, 5000 V _{RMS}	Reinforced insulation, Altitude ≤ 5000 m, Tropical Climate, 700 V _{RMS} maximum working voltage	5000 V _{RMS} reinforced insulation per EN 61010-1:2010/A1:2019 and EN 62368-1:2014 up to working voltage of 600 V _{RMS}
Certificate number: 40040142	Master contract number: 220991	File number: E181974	Certificate planned	Client ID number: 077311

7.8 Safety Limiting Values

Safety limiting(1) intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DW-16 PA	ACKAGE					
		R _{0JA} =73°C/W, V _I = 5.5 V, T _J = 150°C, T _A = 25°C			311.4	mA
	Safety input, output, or supply current	R _{θJA} = 73°C/W, V _I = 3.6 V, T _J = 150°C, T _A = 25°C			475.7	A
I _S		$R_{\theta JA} = 73^{\circ}\text{C/W}, V_I = 2.75 \text{ V}, T_J = 150^{\circ}\text{C}, T_A = 25^{\circ}\text{C}$			622	mA
		$R_{\theta JA} = 73^{\circ}\text{C/W}, V_I = 1.89 \text{ V}, T_J = 150^{\circ}\text{C}, T_A = 25^{\circ}\text{C}$			905.1	mA
Ps	Safety input, output, or total power	R _{0JA} = 73°C/W, T _J = 150°C, T _A = 25°C			1712.4	mW
T _S	Maximum safety temperature				150	°C

The maximum safety temperature, T_S, has the same value as the maximum junction temperature, T_J, specified for the device. The I_S and P_S parameters represent the safety current and safety power respectively. The maximum limits of I_S and P_S should not be exceeded. These limits vary with the ambient temperature, T_A.

The junction-to-air thermal resistance, R_{0,JA}, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

 $T_J = T_A + R_{\theta JA} \times P$, where P is the power dissipated in the device. $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$, where $T_{J(max)}$ is the maximum allowed junction temperature. $P_S = I_S \times V_I$, where V_I is the maximum input voltage.



Electrical Characteristics—5-V Supply

V_{CC1} = V_{CC2} = 5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -4 mA; See Figure 8-1	V _{CCO} - 0.4 ⁽¹⁾			V
V _{OL}	Low-level output voltage	I _{OL} = 4 mA; See Figure 8-1			0.4	V
V _{IT+(IN)}	Rising input switching threshold			0.7	x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}			V
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}			V
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx			10	μΑ
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10			μΑ
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx			28	uA
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-28			uA
CMTI	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 8-4	100	150		kV/us
C _i	Input Capacitance (2)	$V_1 = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 5 \text{ V}$		2.8		pF

⁽¹⁾ V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} Measured from input pin to same side ground.

⁽²⁾



7.9 Supply Current Characteristics—5-V Supply

V_{CC1} = V_{CC2} = 5 V ±10% (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	S	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6740							
	$V_I = V_{CC1}$ (1)(ISO6740); $V_I = 0 \text{ V}$ (ISO	06740 with F	I _{CC1}		1.6	2.2	
Supply current - DC signal	suffix)		I _{CC2}		2.1	3.4	
Supply current - DC signal Supply current - AC signal Supply current - DC signal Supply current - DC signal Supply current - DC signal Supply current - AC signal	V _I = 0 V (ISO6740); V _I = V _{CC1} (ISO6	740 with F suffix)	I _{CC1}		5.8	8	
O6741 upply current - DC signal	V = 0 V (1300740), V = V661 (1300	740 WIIIT SUIIX)	I _{CC2}		2.3	3.7	
		1 Mbps	I _{CC1}		3.7	5.1	mA
		1 Mbps	I _{CC2}		2.4	3.8	ША
Supply current - AC signal	All channels switching with square	10 Mbps	I _{CC1}		3.8	5.3	
(3)	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}		4.8	6.4	
		50 Mbps	I _{CC1}		4.4	6	
		oo wapa	I _{CC2}		15	17.8	
ISO6741							
Supply current - DC signal (2)	$ V_1 = V_{CC1} ^{1/2} (1506/41); V_1 = 0 V_1 (1506/41) With F suffix) F$		I _{CC1}		1.9	2.8	
			I _{CC2}		2.2	3.5	
	$V_1 = 0 \text{ V (ISO6741)}; V_1 = V_{CC1} \text{ (ISO6741 with F suffix)}$		I _{CC1}		5.1	7.2	
	V = 0 V (1880/41), V = VEEI (1880)	THI WILLII SUIIX)	uffix) $\frac{I_{CC1}}{I_{CC2}}$ 5.1 7.2 3.4 5.1	5.1			
		1 Mbps	I _{CC1}		3.6	5.1	mA
		1 Mbps	I _{CC2}		3	4.5	ША
Supply current - AC signal	All channels switching with square	10 Mbps	I _{CC1}		4.2	5.8	
(3)	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}		4.8	6.5	
		50 Mbps	I _{CC1}		7.3	9.3	
		30 Mbps	I _{CC2}		12.6	15.3	ı
ISO6742							
Supply current - DC	$V_I = V_{CCI}$ (1)(ISO6742); $V_I = 0 V$ (ISO	6742 with F suffix)	I _{CC1} , I _{CC2}		2.2	3.3	
Supply current - DC signal ⁽²⁾	V _I = 0 V (ISO6742); V _I = V _{CCI} (ISO67	742 with F suffix)	I _{CC1} , I _{CC2}		4.4	6.3	
		1 Mbps	I _{CC1} , I _{CC2}		3.4	5	mA
Supply current - AC signal ⁽³⁾	All channels switching with square wave clock input; C _L = 15 pF	10 Mbps	I _{CC1} , I _{CC2}		4.7	6.4	
3	inate state input, or 10 pi	50 Mbps	I _{CC1} , I _{CC2}		10.2	12.5	

 ⁽¹⁾ V_{CCI} = Input-side V_{CC}
 (2) Supply current valid for ENx = V_{CCx} and ENx = 0V
 (3) Supply current valid for ENx = V_{CCx}



7.10 Electrical Characteristics—3.3-V Supply

V_{CC1} = V_{CC2} = 3.3 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -2mA; See Figure 8-1	V _{CCO} - 0.2 ⁽¹⁾			V
V_{OL}	Low-level output voltage	I _{OL} = 2mA; See Figure 8-1			0.2	V
V _{IT+(IN)}	Rising input switching threshold				0.7 x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}			٧
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}			٧
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx			10	μA
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10			μΑ
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx			30	uA
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-30			uA
CMTI	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 8-4	100	150		kV/us
C _i	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 3.3 \text{ V}$		2.8		pF

 $[\]begin{array}{ll} \text{(1)} & \text{V_{CCI} = Input-side V_{CC}; V_{CCO} = Output-side V_{CC}} \\ \text{(2)} & \text{Measured from input pin to same side ground.} \\ \end{array}$



7.11 Supply Current Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	s	SUPPLY CURRENT	MIN TYP	MAX	UNIT
ISO6740						
	V _I = V _{CC1} (1)(ISO6740); V _I = 0 V (IS	O6740 with F	I _{CC1}	1.6	2.2	
Supply current - DC	suffix)		I _{CC2}	2.1	3.3	
Supply current - DC ignal (2) Supply current - AC ignal (3) SO6741 Supply current - DC signal (2) Supply current - AC signal (3) Supply current - AC signal (3)	V _I = 0 V (ISO6740); V _I = V _{CC1} (ISO6	740 with F suffix)	I _{CC1}	5.7	8	
	V = 0 V (1300740); V = V6C1 (1300	740 WILL SUIIX)	I _{CC2}	1.6 2.2 2.1 3.3 5.7 8 2.3 3.6 3.7 5.1 2.4 3.7 3.8 5.2 4 5.6 4.2 5.7 11.2 13.8 1.9 2.7 2.2 3.4 5.1 3.5 5 2.9 4.4 4 5.5 4.2 5.8 6.1 8 9.7 12.1 12.1 13.4 1.9		
		1 Mbps	I _{CC1}	3.7	5.1	mA
		1 Wibps	I _{CC2}	2.4	3.7	ША
Supply current - AC	All channels switching with square	10 Mbps	I _{CC1}	3.8	5.2	
signal ⁽³⁾	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}	4	5.6	
		50 Mbps	I _{CC1}	4.2	5.7	
		30 Mbps	I _{CC2}	11.2	13.8	
ISO6741						
Supply current - DC signal	ouffix)		I _{CC1}	1.9	2.7	
			I _{CC2}	2.2	3.4	
(2)	ont - DC signal $V_{L} = 0 \text{ V (ISO6741); } V_{L} = V_{CCL} \text{ (ISO6741 with F suffix)}$ I_{CC2} 2.2 I_{CC1} 5	7.1				
	V = 0 V (1300741), V = VCCI (1300	741 WILLI F SUIIX)	I _{CC2}	3.4	5.1 3.7 5.2 5.6 5.7 13.8 2.7 3.4 7.1 5.1 5 4.4 5.5 5.8 8 12.1	
		1 Mbps	I _{CC1}	3.5	5	mA
		1 Mbps	I _{CC2}	2.9	4.4	ША
Supply current - DC signal $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (ISO6741 \text{ with F suffix})$ $V_{I} = 0 \text{ V (ISO6741)}; V_{I} = V_{CCI} (IS$	5.5					
(3)	wave clock input; C _L = 15 pF	TO Mbps	I _{CC2}	4.2	5.8	
		50 Mbps	I _{CC1}	6.1	8	
		50 Mbps	I _{CC2}	9.7	12.1	
ISO6742						
Supply current - DC	V _I = V _{CCI} ⁽¹⁾ (ISO6742); V _I = 0 V (ISO	06742 with F suffix)	I _{CC1} , I _{CC2}	2.2	2.7 3.4 7.1 5.1 5 4.4 5.5 5.8 8 12.1 3.3 6.3 4.9 5.9	
Supply current - DC signal ⁽²⁾	V _I = 0 V (ISO6742); V _I = V _{CCI} (ISO6	742 with F suffix)	I _{CC1} , I _{CC2}	4.4	6.3	
		1 Mbps	I _{CC1,} I _{CC2}	3.4	4.9	mA
Supply current - AC signal ⁽³⁾	All channels switching with square wave clock input; $C_L = 15 \text{ pF}$	10 Mbps	I _{CC1} , I _{CC2}	4.2	5.9	
oigilai ·	Tate door input, of 10 pi	50 Mbps	I _{CC1,} I _{CC2}	8.2	10.3	

 ⁽¹⁾ V_{CCI} = Input-side V_{CC}
 (2) Supply current valid for ENx = V_{CCx} and ENx = 0V
 (3) Supply current valid for ENx = V_{CCx}



7.12 Electrical Characteristics—2.5-V Supply

V_{CC1} = V_{CC2} = 2.5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -1mA; See Figure 8-1	V _{CCO} - 0.1 ⁽¹⁾			V
V _{OL}	Low-level output voltage	I _{OL} = 1mA; See Figure 8-1			0.1	V
V _{IT+(IN)}	Rising input switching threshold				0.7 x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}			V
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}			V
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx			10	μA
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10			μA
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx			30	uA
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-30			uA
CMTI	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 8-4	100	150		kV/us
Ci	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 2.5 \text{ V}$		2.8		pF

 $[\]begin{array}{ll} \text{(1)} & \text{V_{CCI} = Input-side V_{CC}; V_{CCO} = Output-side V_{CC}} \\ \text{(2)} & \text{Measured from input pin to same side ground.} \\ \end{array}$



7.13 Supply Current Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	S	SUPPLY CURRENT	MIN TYP	MAX	UNIT
ISO6740						
	$V_I = V_{CC1}$ (1)(ISO6740); $V_I = 0$ V (ISO	D6740 with F	I _{CC1}	1.6	2.2	
Supply current - DC signal	suffix)		I _{CC2}	2.1	3.3	
Supply current - DC signal Supply current - AC signal (3) SO6741 Supply current - DC signal 2) Supply current - AC signal 3)	$ V_1 = 0 \text{ V (ISO6740)}; V_1 = V_{CC1} \text{ (ISO6740 with F suffix)} $		I _{CC1}	5.7	7.9	
	V = 0 V (1300740), V = V661 (1300	740 Willi Sullix)	I _{CC2}	2.3	3.6	
		1 Mbps	I _{CC1}	3.7	5.1	mA
O6741		1 Mbps	I _{CC2}	2.3	3.6	ША
Supply current - AC	All channels switching with square	10 Mbps	I _{CC1}	3.7	5.1	
signal ⁽³⁾	wave clock input; C _L = 15 pF	TO IVIDPS	I _{CC2}	3.5	5.1	
		50 Mbps	I _{CC1}	4.1	5.6	
			I _{CC2}	9	11.2	
ISO6741						
Supply current - DC signal	V = V (1)(ISO6741): V = 0.V (ISO	$V_{i} = V_{0.01} (1)(15.06741)$: $V_{i} = 0.07 (15.06741)$ with E suffix)		1.9	2.7	
			I _{CC2}	2.2	3.4	
(2)	- DC signal $V_1 = V_{CCI} \text{ (ISO6741); } V_1 = 0 \text{ V (ISO6741 with F suffix)} $ I_{CC2} 2.2 3 $V_1 = 0 \text{ V (ISO6741); } V_1 = V_{CCI} \text{ (ISO6741 with F suffix)} $	7.1				
	V = 0 V (1300741), V = V _{CCI} (1300	741 WILLIF SUIIX)	I _{CC2}	3.4	5.1 5.1 5.6 11.2 2.7 3.4	
		1 Mbps	I _{CC1}	3.5	5	mA
		1 Mbps	I _{CC2}	2.9	4.4	IIIA
Supply current - AC signal	All channels switching with square	10 Mbps	I _{CC1}	3.9	5.4	
(3)	wave clock input; C _L = 15 pF	10 Mbps	I _{CC2}	3.8	5.4	
		EO Mbno	I _{CC1}	5.5	7.2	
		50 Mbps	I _{CC2}	8.1	10.2	
ISO6742					'	
Supply current - DC signal	V _I = V _{CCI} ⁽¹⁾ (ISO6742); V _I = 0 V (ISC	06742 with F suffix)	I _{CC1} , I _{CC2}	2.2	3.4 7.1 5.1 5 4.4 5.4 5.4 7.2	
(2)	V _I = 0 V (ISO6742); V _I = V _{CCI} (ISO6	742 with F suffix)	I _{CC1} , I _{CC2}	4.3	6.3	
		1 Mbps	I _{CC1} , I _{CC2}	3.3	4.8	mA
Supply current - AC signal ⁽³⁾	All channels switching with square wave clock input; $C_L = 15 \text{ pF}$	10 Mbps	I _{CC1,} I _{CC2}	4	5.6	
signal · /	wave clock input, of - 10 bi	50 Mbps	I _{CC1} , I _{CC2}	7	9	

 ⁽¹⁾ V_{CCI} = Input-side V_{CC}
 (2) Supply current valid for ENx = V_{CCx} and ENx = 0V
 (3) Supply current valid for ENx = V_{CCx}



Electrical Characteristics—1.8-V Supply

V_{CC1} = V_{CC2} = 1.8 V ±5% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -1mA; See Figure 8-1	V _{CCO} - 0.1 (1)		V
V _{OL}	Low-level output voltage	I _{OL} = 1mA; See Figure 8-1		0.1	V
V _{IT+(IN)}	Rising input switching threshold			0.7 x V _{CCI} ⁽¹⁾	V
V _{IT-(IN)}	Falling input switching threshold		0.3 x V _{CCI}		V
V _{I(HYS)}	Input threshold voltage hysteresis		0.1 x V _{CCI}		V
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at INx		10	μA
I _{IL}	Low-level input current	V _{IL} = 0 V at INx	-10		μA
I _{IH}	High-level input current	V _{IH} = V _{CCI} ⁽¹⁾ at ENx		30	μA
I _{IL}	Low-level input current	V _{IL} = 0 V at ENx	-30		μA
СМТІ	Common mode transient immunity	V _I = V _{CC} or 0 V, V _{CM} = 1200 V; See Figure 8-4	100	150	kV/us
C _i	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 1.8 \text{ V}$		2.8	pF

 V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} Measured from input pin to same side ground. (1)



7.14 Supply Current Characteristics—1.8-V Supply

V_{CC1} = V_{CC2} = 1.8 V ±5% (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITION	S	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6740							
	$V_I = V_{CC1}$ (1)(ISO6740); $V_I = 0 \text{ V}$ (ISO	06740 with F	I _{CC1}		1.2	1.8	
Supply current - DC	suffix)		I _{CC2}		2	3.4	
signal ⁽²⁾	V _I = 0 V (ISO6740); V _I = V _{CC1} (ISO6	740 with F suffix)	I _{CC1}		5.1	2 1.8 2 3.4 1 7.6 2 3.7 1 4.7 2 3.7 2 4.8 1 4.6 4 5.1 7 8.9 5 2.4 2 3.4 5 6.9 2 5 1 4.7 7 4.3 3 5 4 5 5 6.3 4 8.3	
	V = 0 V (1300740), V = V661 (1300	740 Willi Sullix)	I _{CC2}		2.2	1.8 3.4 7.6 3.7 4.7 3.7 4.8 4.6 5.1 8.9 2.4 3.4 6.9 5 4.7 4.3 5 6.3 8.3	
		1 Mbps	I _{CC1}		3.1	3.4 7.6 3.7 4.7 3.7 4.8 4.6 5.1 8.9 2.4 3.4 6.9 5 4.7 4.3 5 6.3 8.3	mA
		1 Mbp3	I _{CC2}		2.2		ША
Supply current - AC	All channels switching with square	10 Mbps	I _{CC1}		3.2	4.8	
signal ⁽³⁾	wave clock input; C _L = 15 pF	TO MIDPS	I _{CC2}	,	3.1	4.6	
		50 Mbps	I _{CC1}		3.4	5.1	
		30 Mbps	I _{CC2}		7	8.9	
ISO6741							
	$ V_1 = V_{CC1} \cup (1SO6/41); V_1 = 0 \ V (1SO6/41) $ with F suffix)		I _{CC1}		1.5	2.4	
			I _{CC2}		2	3.4	
(2)	V = 0 V (ISO6741); V = V (ISO6	744 with Fouffix)	I _{CC1}	2 3.4 c1 4.5 6.9			
	V ₁ = 0 V (1506741), V ₁ = V _{CCI} (1506	741 WILLI F SUIIIX)	I _{CC2}		3.2	4.6 5.1 8.9 2.4 3.4 6.9 5 4.7 4.3 5 6.3	
		1 Mbno	I _{CC1}		3.1	4.7	m 1
		1 Mbps	I _{CC2}		2.7	2 5 .1 4.7	mA
Supply current - DC signal (2) $V_{I} = 0 \text{ V (ISO6741); } V_{I} = V_{CCI} \text{ (ISO6741 with F suffix)} \qquad \begin{array}{c ccccccccccccccccccccccccccccccccccc$	5						
(3)	wave clock input; C _L = 15 pF	TO Mbps	I _{CC2}		3.4	5	
		50 Mb a c	I _{CC1}		4.5	6.3	
		50 Mbps	I _{CC2}		6.4	8.3	
ISO6742	1			,			
Supply current - DC	V _I = V _{CCI} ⁽¹⁾ (ISO6742); V _I = 0 V (ISC	06742 with F suffix)	I _{CC1} , I _{CC2}		1.9	3.1	
Supply current - DC signal ⁽²⁾	V _I = 0 V (ISO6742); V _I = V _{CCI} (ISO67	742 with F suffix)	I _{CC1} , I _{CC2}		4	6.1	
		1 Mbps	I _{CC1} , I _{CC2}		3	4.7	mA
Supply current - AC signal ⁽³⁾	All channels switching with square wave clock input; $C_L = 15 \text{ pF}$	10 Mbps	I _{CC1} , I _{CC2}		3.5	5.2	
signal 💛	wave clock illput, ot - 10 pr	50 Mbps	I _{CC1} , I _{CC2}		5.6	7.6	

 ⁽¹⁾ V_{CCI} = Input-side V_{CC}
 (2) Supply current valid for ENx = V_{CCx} and ENx = 0V
 (3) Supply current valid for ENx = V_{CCx}



7.15 Switching Characteristics—5-V Supply

V_{CC1} = V_{CC2} = 5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@100kbps		11	18	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} - t _{PLH}	See Figure 8-1		0.2	7	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾				6	ns
t _r	Output signal rise time	Con Figure 0.4		2.6	4.5	ns
t _f	Output signal fall time	See Figure 8-1		2.6	4.5	ns
t _{PHZ}	Disable propagation delay, high-to-high impedance output			18.6	25.8	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output			18.6	25.8	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO674x	out for See Figure 8-2		14.2	21.1	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO674x			14.2	21.1	ns
t _{PU}	Time from UVLO to valid output data				300	us
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2V. See Figure 8-3		0.1	0.3	us
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1		ns

⁽¹⁾ Also known as pulse skew.

⁽²⁾ $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ t_{sk(pp)} is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



7.16 Switching Characteristics—3.3-V Supply

V_{CC1} = V_{CC2} = 3.3 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@100kbps		11	18	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} - t _{PLH}	See Figure 8-1		0.5	7	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾				7	ns
t _r	Output signal rise time	Con Firmer 0.4		1.6	3.2	ns
t _f	Output signal fall time	See Figure 8-1		1.6	3.2	ns
t _{PHZ}	Disable propagation delay, high-to-high impedance output			23.2	34.4	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output			23.2	34.4	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO674x	tput put put for See Figure 8-2		16.6	23	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO674x			16.6	23	ns
t _{PU}	Time from UVLO to valid output data				300	us
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2V. See Figure 8-3		0.1	0.3	us
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1		ns

⁽¹⁾ Also known as pulse skew.

⁽²⁾ $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ t_{sk(pp)} is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



7.17 Switching Characteristics—2.5-V Supply

V_{CC1} = V_{CC2} = 2.5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@100kbps		12	20.5	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} - t _{PLH}	See Figure 8-1		0.6	7.1	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾				7	ns
t _r	Output signal rise time	Con Figure 0.4		2	4	ns
t _f	Output signal fall time	See Figure 8-1	12 20.5 0.6 7.1 6 7	ns		
t _{PHZ}	Disable propagation delay, high-to-high impedance output			28.1	43	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output			28.1	43	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO674x	put for See Figure 8-2		20.4	36.3	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO674x			20.4	36.3	ns
t _{PU}	Time from UVLO to valid output data				300	us
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2V. See Figure 8-3		0.1	0.3	us
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1		ns

⁽¹⁾ Also known as pulse skew.

⁽²⁾ $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ t_{sk(pp)} is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



7.18 Switching Characteristics—1.8-V Supply

V_{CC1} = V_{CC2} = 1.8 V ±5% (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH} , t _{PHL}	Propagation delay time	@100kbps		15	24	ns
PWD	Pulse width distortion ⁽¹⁾ t _{PHL} - t _{PLH}	See Figure 8-1		0.7	8.2	ns
t _{sk(o)}	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			6	ns
t _{sk(pp)}	Part-to-part skew time ⁽³⁾				8.8	ns
t _r	Output signal rise time	See Figure 9.4		2.7	5.3	ns
t _f	Output signal fall time	See Figure 8-1		2.7	5.3	ns
t _{PHZ}	Disable propagation delay, high-to-high impedance output			40.3	63	ns
t _{PLZ}	Disable propagation delay, low-to-high impedance output			40.3	63	ns
t _{PZH}	Enable propagation delay, high impedance-to-high output for ISO674x	See Figure 8-2		30	51.4	ns
t _{PZL}	Enable propagation delay, high impedance-to-low output for ISO674x			30	51.4	ns
t _{PU}	Time from UVLO to valid output data				300	us
t _{DO}	Default output delay time from input power loss	Measured from the time VCC goes below 1.2V. See Figure 8-3		0.1	0.3	us
t _{ie}	Time interval error	2 ¹⁶ – 1 PRBS data at 50 Mbps		1		ns

⁽¹⁾ Also known as pulse skew.

⁽²⁾ $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

⁽³⁾ $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



7.19 Insulation Characteristics Curves

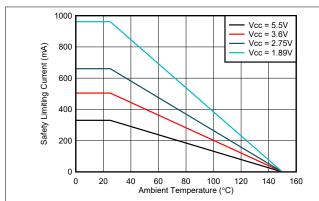


Figure 7-1. Thermal Derating Curve for Safety Limiting Current for DW-16 Package

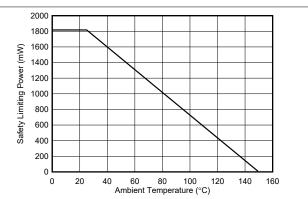
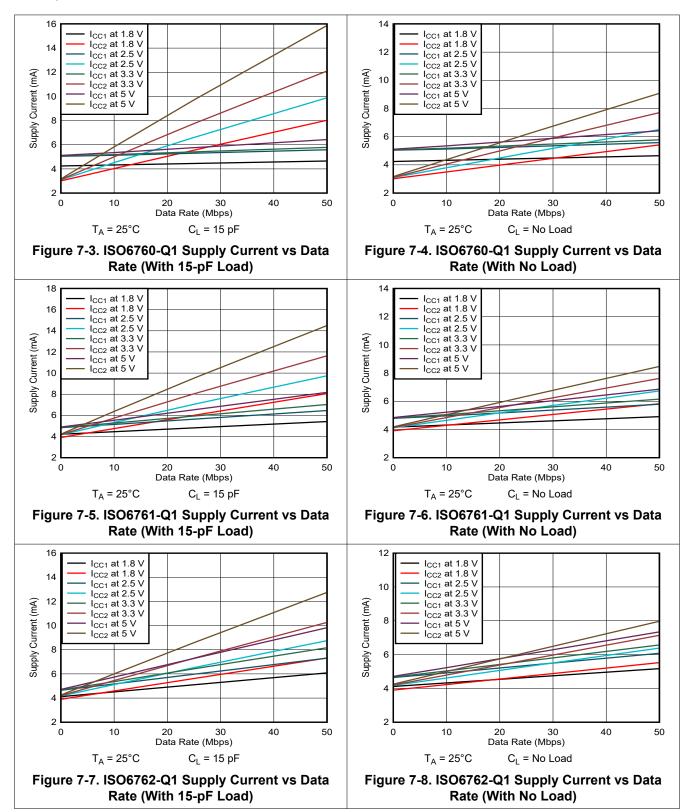


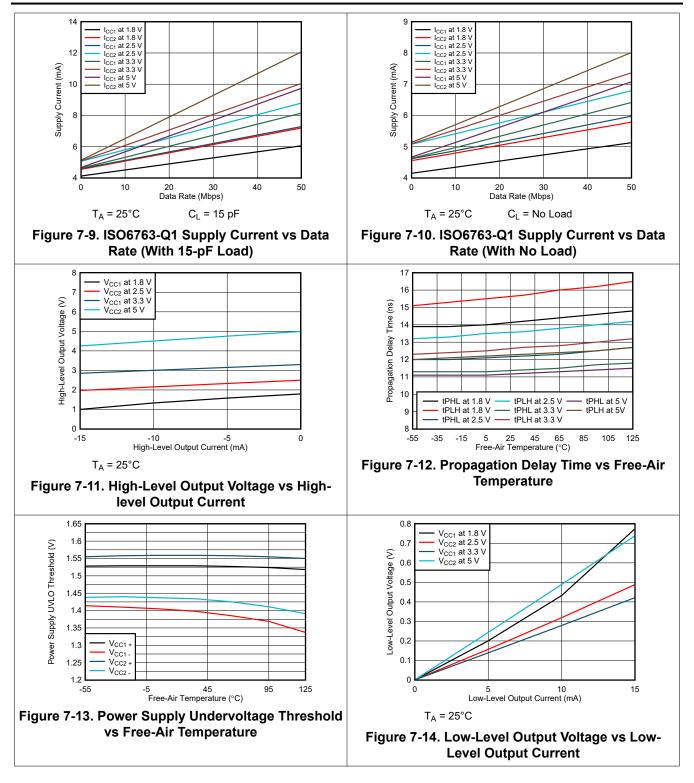
Figure 7-2. Thermal Derating Curve for Safety Limiting Power for DW-16 Package



7.20 Typical Characteristics

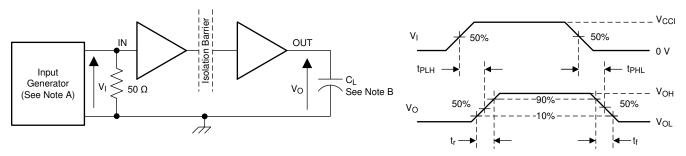








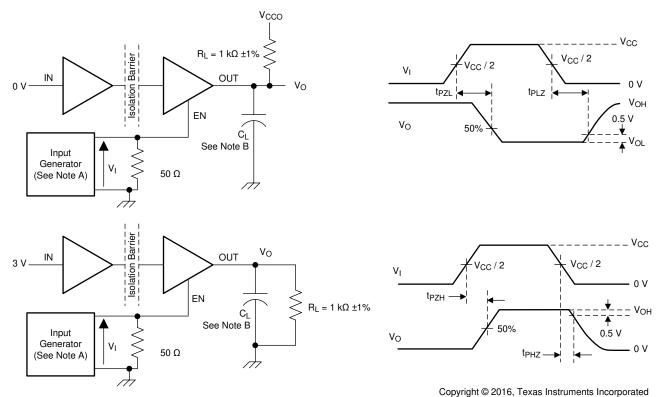
8 Parameter Measurement Information



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- A. The input pulse is supplied by a generator having the following characteristics: PRR \leq 50 kHz, 50% duty cycle, $t_r \leq$ 3 ns, $t_f \leq$ 3ns, $Z_O =$ 50 Ω. At the input, 50 Ω resistor is required to terminate Input Generator signal. It is not needed in actual application.
- B. C_L = 15 pF and includes instrumentation and fixture capacitance within ±20%.

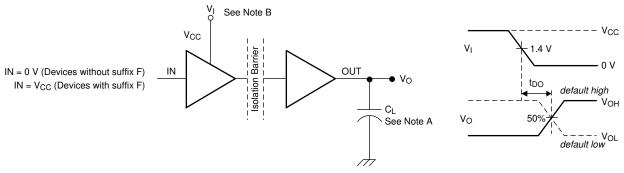
Figure 8-1. Switching Characteristics Test Circuit and Voltage Waveforms



- The input pulse is supplied by a generator having the following characteristics: PRR ≤ 10 kHz, 50% duty cycle,
- $t_r \le 3 \text{ ns, } t_f \le 3 \text{ ns, } Z_O = 50 \Omega.$
- B. $C_L = 15 \text{ pF}$ and includes instrumentation and fixture capacitance within $\pm 20\%$.

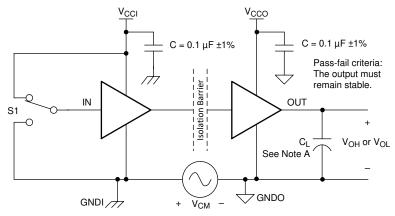
Figure 8-2. Enable/Disable Propagation Delay Time Test Circuit and Waveform





- C_L = 15 pF and includes instrumentation and fixture capacitance within ±20%.
- B. Power Supply Ramp Rate = 10 mV/ns

Figure 8-3. Default Output Delay Time Test Circuit and Voltage Waveforms



A. $C_L = 15 \text{ pF}$ and includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 8-4. Common-Mode Transient Immunity Test Circuit

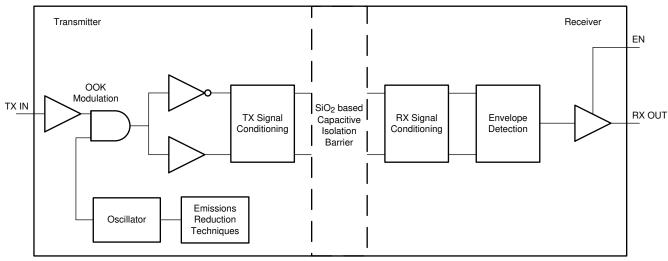


9 Detailed Description

9.1 Overview

The ISO674x-Q1 family of devices have an ON-OFF keying (OOK) modulation scheme to transmit the digital data across a silicon dioxide based isolation barrier. The transmitter sends a high frequency carrier across the barrier to represent one digital state and sends no signal to represent the other digital state. The receiver demodulates the signal after advanced signal conditioning and produces the output through a buffer stage. If the ENx pin is low then the output goes to high impedance. The ISO674x-Q1 devices also incorporate advanced circuit techniques to maximize the CMTI performance and minimize the radiated emissions due to the high frequency carrier and IO buffer switching. The conceptual block diagram of a digital capacitive isolator, Figure 9-1, shows a functional block diagram of a typical channel.

9.2 Functional Block Diagram



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Figure 9-1. Conceptual Block Diagram of a Digital Capacitive Isolator

Figure 9-2 shows a conceptual detail of how the ON-OFF keying scheme works.

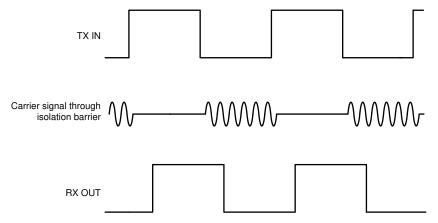


Figure 9-2. On-Off Keying (OOK) Based Modulation Scheme



9.3 Feature Description

Table 9-1 provides an overview of the device features.

Table 9-1. Device Features

PART NUMBER	CHANNEL DIRECTION	MAXIMUM DATA RATE	DEFAULT OUTPUT	PACKAGE	RATED ISOLATION(1)
ISO6740-Q1	4 Forward, 0 Reverse	50 Mbps	High	DW-16	5000 V _{RMS} / 8000 V _{PK}
ISO6740F-Q1	4 Forward, 0 Reverse	50 Mbps	Low	DW-16	5000 V _{RMS} / 8000 V _{PK}
ISO6741-Q1	3 Forward, 1 Reverse	50 Mbps	High	DW-16	5000 V _{RMS} / 8000 V _{PK}
ISO6741F-Q1	3 Forward, 1 Reverse	50 Mbps	Low	DW-16	5000 V _{RMS} / 8000 V _{PK}
ISO6742-Q1	2 Forward, 2 Reverse	50 Mbps	High	DW-16	5000 V _{RMS} / 8000 V _{PK}
ISO6742F-Q1	2 Forward, 2 Reverse	50 Mbps	Low	DW-16	5000 V _{RMS} / 8000 V _{PK}

⁽¹⁾ See Safety-Related Certifications for detailed isolation ratings.

9.3.1 Electromagnetic Compatibility (EMC) Considerations

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 25. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the ISO674x-Q1 family of devices incorporates many chip-level design improvements for overall system robustness. Some of these improvements include:

- Robust ESD protection cells for input and output signal pins and inter-chip bond pads.
- · Low-resistance connectivity of ESD cells to supply and ground pins.
- Enhanced performance of high voltage isolation capacitor for better tolerance of ESD, EFT and surge events.
- Bigger on-chip decoupling capacitors to bypass undesirable high energy signals through a low impedance path.
- PMOS and NMOS devices isolated from each other by using guard rings to avoid triggering of parasitic SCRs.
- Reduced common mode currents across the isolation barrier by ensuring purely differential internal operation.



9.4 Device Functional Modes

Table 9-2 lists the functional modes for the ISO674x-Q1 devices.

Table 9-2. Function Table

Table 3-2. I direction Table						
V _{CCI} ⁽¹⁾	V _{cco}	INPUT (INx) ⁽³⁾	OUTPUT ENABLE (ENx)	OUTPUT (OUTx)	COMMENTS	
		Н	H or open	Н	Normal Operation: A channel output assumes the logic state of its	
		L	H or open	L	input.	
PU	PU	Open	H or open	Default	Default mode: When INx is open, the corresponding channel output goes to its default logic state. Default is <i>High</i> for ISO674x-Q1 and <i>Low</i> for ISO674x-Q1 with F suffix.	
Х	PU	х	L	Z	A low value of output enable causes the outputs to be high-impedance.	
PD	PU	x	H or open	Default	Default mode: When $V_{\rm CCI}$ is unpowered, a channel output assumes the logic state based on the selected default option. Default is $High$ for ISO674x-Q1 and Low for ISO674x-Q1 with F suffix. When $V_{\rm CCI}$ transitions from unpowered to powered-up, a channel output assumes the logic state of the input. When $V_{\rm CCI}$ transitions from powered-up to unpowered, channel output assumes the selected default state.	
Х	PD	х	х	Undetermined	When $V_{\rm CCO}$ is unpowered, a channel output is undetermined ⁽²⁾ . When $V_{\rm CCO}$ transitions from unpowered to powered-up, a channel output assumes the logic state of the input.	

- V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} ; PU = Powered up ($V_{CC} \ge 1.71$ V); PD = Powered down ($V_{CC} \le 1.05$ V); X = Irrelevant; H = High level; L = Low level; Z = High Impedance (1)
- The outputs are in undetermined state when 1.89 V < V_{CCI} , V_{CCO} < 2.25 V and 1.05 V < V_{CCI} , V_{CCO} < 1.71 V A strongly driven input signal can weakly power the floating V_{CC} through an internal protection diode and cause undetermined output

9.4.1 Device I/O Schematics

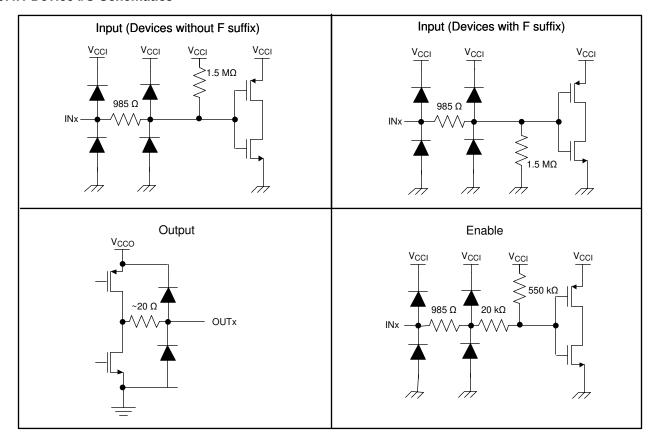


Figure 9-3. Device I/O Schematics



10 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

The ISO674x-Q1 devices are high-performance, quad-channel digital isolators. These devices come with enable pins on each side which can be used to put the respective outputs in high impedance for multi master driving applications. The ISO674x-Q1 devices use single-ended CMOS-logic switching technology. The supply voltage range is from 1.71 V to 5.5 V for both supplies, V_{CC1} and V_{CC2} . Since an isolation barrier separates the two sides, each side can be sourced independently with any voltage within recommended operating conditions. As an example, it is possible to supply ISO674x-Q1 V_{CC1} with 3.3 V (which is within 1.71 V to 5.5 V) and V_{CC2} with 5V (which is also within 1.71 V to 5.5 V). You can use the digital isolator as a logic-level translator in addition to providing isolation. When designing with digital isolators, keep in mind that because of the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is, MCU or FPGA), and a data converter or a line transceiver, regardless of the interface type or standard.

10.2 Typical Application

Figure 10-1 shows the typical isolated CAN interface implementation.

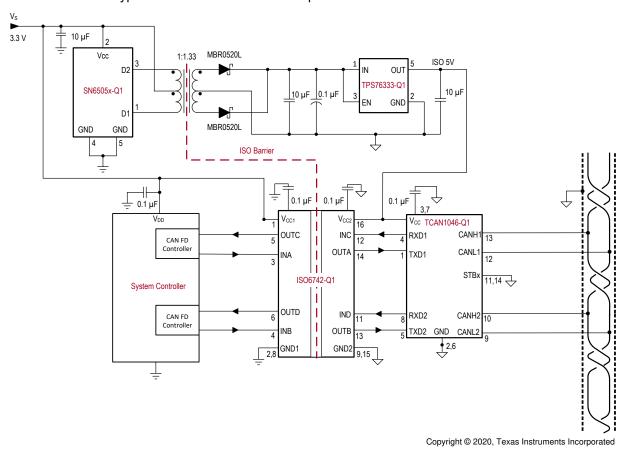


Figure 10-1. Typical Isolated CAN Application Circuit



10.2.1 Design Requirements

To design with these devices, use the parameters listed in Table 10-1.

Table 10-1. Design Parameters

PARAMETER	VALUE
Supply voltage, V _{CC1} and V _{CC2}	1.71 V to 1.89 V and 2.25 V to 5.5 V
Decoupling capacitor between V _{CC1} and GND1	0.1 µF
Decoupling capacitor from V _{CC2} and GND2	0.1 µF

10.2.2 Detailed Design Procedure

Unlike optocouplers, which require external components to improve performance, provide bias, or limit current, the ISO674x-Q1 family of devices only require two external bypass capacitors to operate.

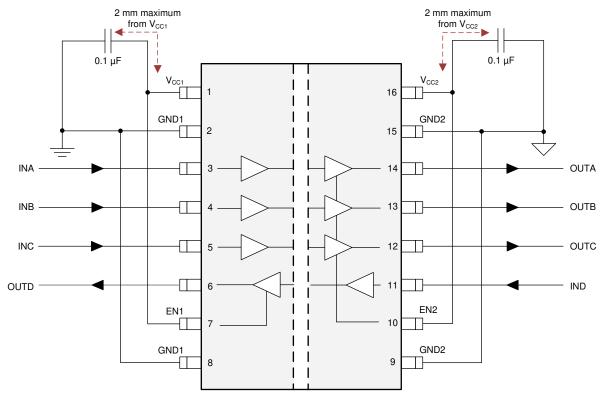


Figure 10-2. Typical ISO674x-Q1 Circuit Hook-up



10.2.3 Application Curve

The following typical eye diagrams of the ISO674x-Q1 family of devices indicates low jitter and wide open eye at the maximum data rate of 50 Mbps.

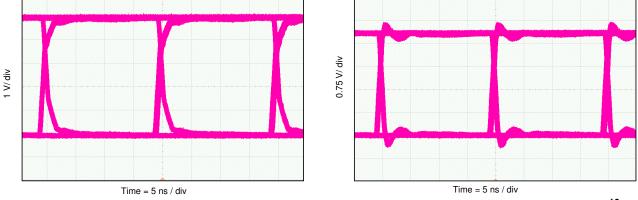


Figure 10-3. Eye Diagram at 50 Mbps PRBS $2^{16} - 1$, Figure 10-4. Eye Diagram at 50 Mbps PRBS $2^{16} - 1$, 5 V and 25° C 3.3 V and 25° C

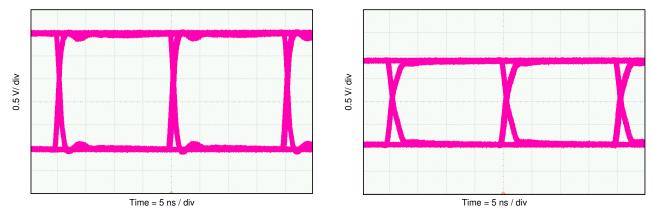


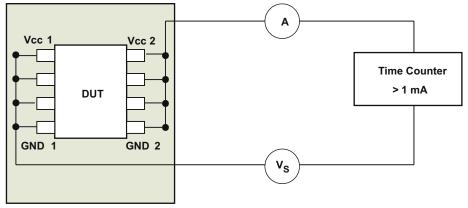
Figure 10-5. Eye Diagram at 50 Mbps PRBS $2^{16} - 1$, Figure 10-6. Eye Diagram at 50 Mbps PRBS $2^{16} - 1$, 2.5 V and 25°C 1.8 V and 25°C

10.2.3.1 Insulation Lifetime

Insulation lifetime projection data is collected by using industry-standard Time Dependent Dielectric Breakdown (TDDB) test method. In this test, all pins on each side of the barrier are tied together creating a two-terminal device and high voltage applied between the two sides; See Figure 10-7 for TDDB test setup. The insulation breakdown data is collected at various high voltages switching at 60 Hz over temperature. For reinforced insulation, VDE standard requires the use of TDDB projection line with failure rate of less than 1 part per million (ppm). Even though the expected minimum insulation lifetime is 20 years at the specified working isolation voltage, VDE reinforced certification requires additional safety margin of 20% for working voltage and 87.5% for lifetime which translates into minimum required insulation lifetime of 37.5 years at a working voltage that's 20% higher than the specified value.

Figure 10-8 shows the intrinsic capability of the isolation barrier to withstand high voltage stress over its lifetime. Based on the TDDB data, the intrinsic capability of the insulation is $1500 \, V_{RMS}$ with a lifetime of $135 \, \text{years}$. Other factors, such as package size, pollution degree, material group, etc. can further limit the working voltage of the component. The working voltage of DW-16 package is specified upto $1500 \, V_{RMS}$. At the lower working voltages, the corresponding insulation lifetime is much longer than $135 \, \text{years}$.





Oven at 150 °C

Figure 10-7. Test Setup for Insulation Lifetime Measurement

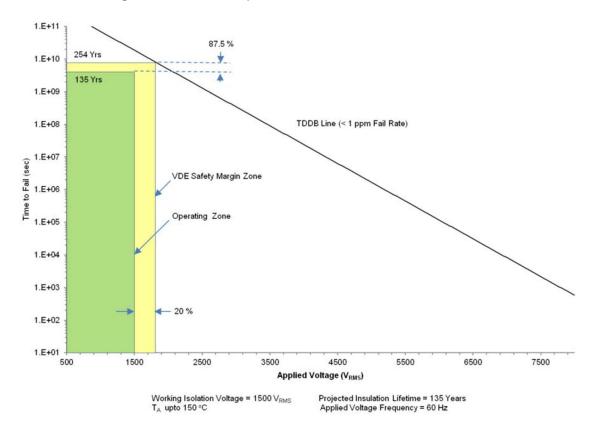


Figure 10-8. Insulation Lifetime Projection Data



11 Power Supply Recommendations

To help ensure reliable operation at data rates and supply voltages, a $0.1-\mu F$ bypass capacitor is recommended at the input and output supply pins (V_{CC1} and V_{CC2}). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver. For automotive applications, please use SN6501-Q1 or SN6505B-Q1. For such applications, detailed power supply design and transformer selection recommendations are available in SN6501-Q1 Transformer Driver for Isolated Power Supplies or SN6505B-Q1 Automotive, low-noise, 1-A, 420-kHz transformer driver with soft start for isolated power supplies.



12 Layout

12.1 Layout Guidelines

A minimum of two layers is required to accomplish a cost optimized and low EMI PCB design. To further improve EMI, a four layer board can be used (see Figure 12-2). Layer stacking for a four layer board should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/inch².
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, refer to the Digital Isolator Design Guide.

12.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit boards. This PCB is preferred over cheaper alternatives due to its lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and self-extinguishing flammability-characteristics.



12.2 Layout Example

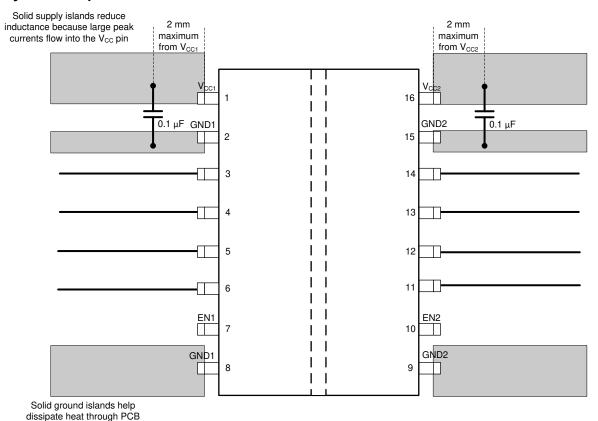


Figure 12-1. Layout Example

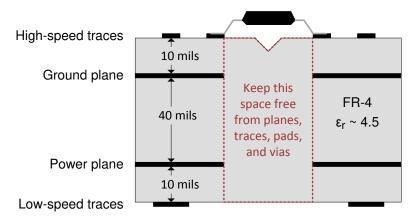


Figure 12-2. Layout Example Schematic



13 Device and Documentation Support

13.1 Documentation Support

13.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, Digital Isolator Design Guide
- · Texas Instruments, Digital Isolator Design Guide
- · Texas Instruments, Isolation Glossary
- Texas Instruments, How to use isolation to improve ESD, EFT, and Surge immunity in industrial systems application report
- Texas Instruments, SN6505x-Q1 Low-Noise 1-A Transformer Drivers for Isolated Power Supplies
- Texas Instruments, TCAN1044-Q1 Automotive Fault-Protected CAN FD Transceiver
- Texas Instruments, TPS763xx-Q1 Low-Power, 150-mA, Low-Dropout Linear Regulators data sheet
- Texas Instruments, TMS320F2803x Piccolo™ Microcontrollers data sheet

13.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

13.3 Support Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

13.4 Trademarks

TI E2E[™] is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

13.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

13.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



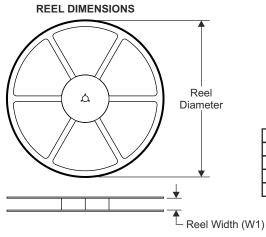
14.1 Package Option Addendum

Packaging Information

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/Ball Finish ⁽⁶⁾	MSL Peak Temp ⁽³⁾	Op Temp (°C)	Device Marking ^{(4) (5)}
ISO6740QDWR Q1	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	40 to 125	ISO6740
ISO6740FQDW RQ1	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	40 to 125	ISO6740F
ISO6741QDWR Q1	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	40 to 125	ISO6741
ISO6741FQDW RQ1	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	40 to 125	ISO6741F
ISO6742QDWR Q1	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	40 to 125	ISO6742
ISO6742FQDW RQ1	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	40 to 125	ISO6742F



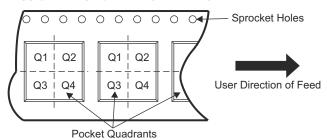
14.2 Tape and Reel Information



TAPE DIMENSIONS K0 P1 B0 Cavity A0

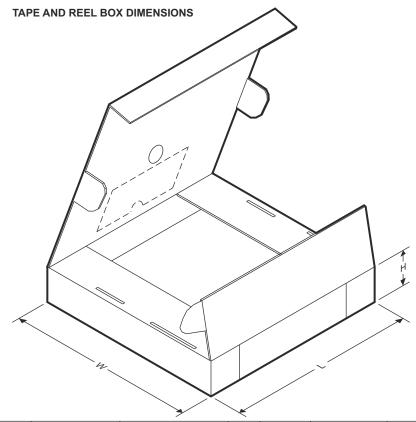
A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO6740QDWRQ1	SOIC	DW	16	2000	330.0	24.4	10.9	10.7	2.7	12.0	24.0	Q1
ISO6740FQDWRQ1	SOIC	DW	16	2000	330.0	24.4	10.9	10.7	2.7	12.0	24.0	Q1
ISO6741QDWRQ1	SOIC	DW	16	2000	330.0	24.4	10.9	10.7	2.7	12.0	24.0	Q1
ISO6741FQDWRQ1	SOIC	DW	16	2000	330.0	24.4	10.9	10.7	2.7	12.0	24.0	Q1
ISO6742QDWRQ1	SOIC	DW	16	2000	330.0	24.4	10.9	10.7	2.7	12.0	24.0	Q1
ISO6742FQDWRQ1	SOIC	DW	16	2000	330.0	24.4	10.9	10.7	2.7	12.0	24.0	Q1





Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO6740QDWRQ1	SOIC	DW	16	2000	367.0	367.0	45.0
ISO6740FQDWRQ1	SOIC	DW	16	2000	367.0	367.0	45.0
ISO6741QDWRQ1	SOIC	DW	16	2000	367.0	367.0	45.0
ISO6741FQDWRQ1	SOIC	DW	16	2000	367.0	367.0	45.0
ISO6742QDWRQ1	SOIC	DW	16	2000	367.0	367.0	45.0
ISO6742FQDWRQ1	SOIC	DW	16	2000	367.0	367.0	45.0

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PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
1000=10=00147001		2010	5111			5 110 0 0	(6)			10007107	
ISO6740FQDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6740F	Samples
ISO6740QDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6740	Samples
ISO6741FQDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6741F	Samples
ISO6741QDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6741	Samples
ISO6742FQDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6742F	Samples
ISO6742QDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6742	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

PACKAGE OPTION ADDENDUM

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OTHER QUALIFIED VERSIONS OF ISO6740-Q1, ISO6741-Q1, ISO6742-Q1:

• Catalog: ISO6740, ISO6741, ISO6742

NOTE: Qualified Version Definitions:

Catalog - TI's standard catalog product



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TAPE AND REEL INFORMATION



TAPE DIMENSIONS + K0 - P1 - B0 W Cavity - A0 -

A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO6740FQDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6740FQDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6740QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6740QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6740QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6741FQDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6741FQDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6741QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6741QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6742FQDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6742FQDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6742QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
ISO6742QDWRQ1	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1



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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO6740FQDWRQ1	SOIC	DW	16	2000	356.0	356.0	35.0
ISO6740FQDWRQ1	SOIC	DW	16	2000	356.0	356.0	35.0
ISO6740QDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6740QDWRQ1	SOIC	DW	16	2000	356.0	356.0	35.0
ISO6740QDWRQ1	SOIC	DW	16	2000	356.0	356.0	35.0
ISO6741FQDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6741FQDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6741QDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6741QDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6742FQDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6742FQDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6742QDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0
ISO6742QDWRQ1	SOIC	DW	16	2000	353.0	353.0	32.0

7.5 x 10.3, 1.27 mm pitch

SMALL OUTLINE INTEGRATED CIRCUIT

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





SOIC



NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing
- per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.



SOIC



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOIC



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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