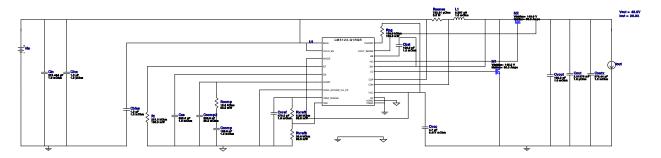


WEBENCH® Design Report

VinMin = 15.0V VinMax = 28.0V Vout = 48.0V Iout = 20.0A Device = LM5123QRGRRQ1 Topology = Boost Created = 2024-06-13 13:09:00.752 BOM Cost = NA BOM Count = 22 Total Pd =

Design: 1 LM5123QRGRRQ1 LM5123QRGRRQ1 15V-28V to 48.00V @ 20A



Design Alerts

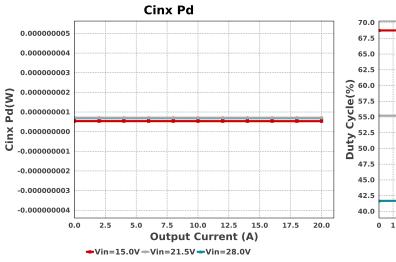
Component Selection Information

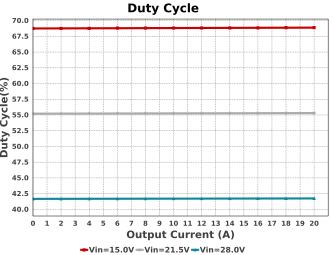
The LM5123-Q1 is qualified for Automotive applications. All passives and other components selected in this design may not be qualified for Automotive applications. The user is required to verify that all components in the design meet the qualification and safety requirements for their specific application.

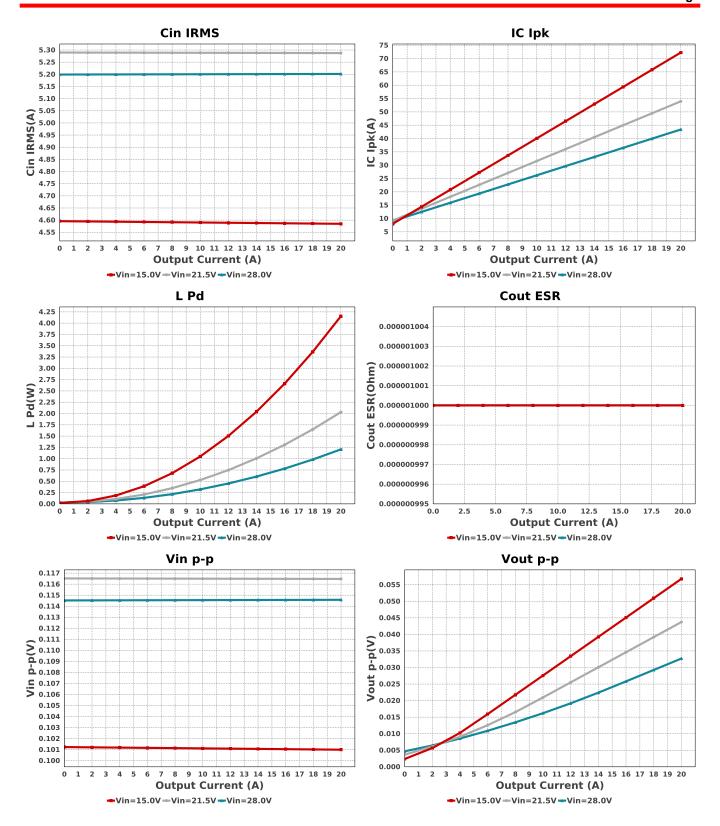
Electrical BOM

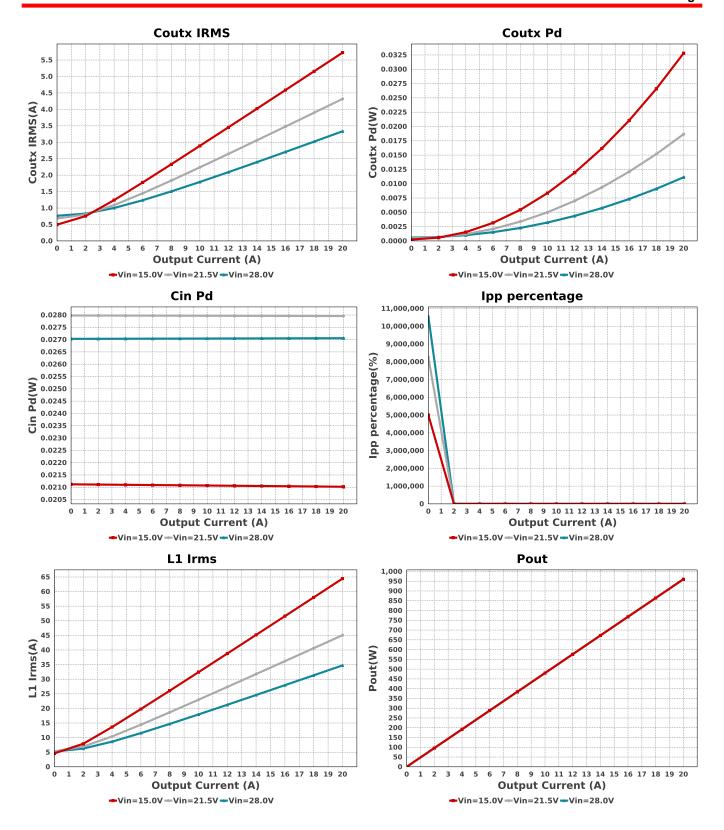
Name	Manufacturer	Part Number	Properties	Qty	Price	Footprint
Cbias	Kemet	C0603C105K8PACTU Series= X5R	Cap= 1.0 uF ESR= 1.0 mOhm VDC= 10.0 V IRMS= 0.0 A	1	\$0.01	0603 5 mm ²
Cbst	MuRata	GRM155R71A104KA01D Series= X7R	Cap= 100.0 nF ESR= 1.0 mOhm VDC= 10.0 V IRMS= 0.0 A	1	\$0.01	0402 3 mm ²
Ccomp	MuRata	GRM155R70J104KA01D Series= X7R	Cap= 100.0 nF ESR= 1.0 mOhm VDC= 6.3 V IRMS= 0.0 A	1	\$0.01	0402 3 mm ²
Ccomp2	Kemet	C0805C331J5GACTU Series= C0G/NP0	Cap= 330.0 pF ESR= 38.0 mOhm VDC= 50.0 V IRMS= 1.04 A	1	\$0.01	0805 7 mm ²
Cin	CUSTOM	CUSTOM Series= ?	Cap= 201.454 uF ESR= 1.0 mOhm VDC= 33.6 V IRMS= 6.5828 A	1	NA	CUSTOM 0 mm ²
Cinx	CUSTOM	CUSTOM Series= ?	Cap= 1.0 uF ESR= 1.0 uOhm VDC= 33.6 V IRMS= 6.5828 A	1	NA	CUSTOM 0 mm ²
Cout	CUSTOM	CUSTOM Series= ?	Cap= 2.31376 mF ESR= 1.0 uOhm VDC= 62.4 V IRMS= 19.4653 A	1	NA	CUSTOM 0 mm ²
Coutx	CUSTOM	CUSTOM Series= ?	Cap= 578.44 uF ESR= 1.0 mOhm VDC= 72.0 V IRMS= 3.31 A	1	NA	CUSTOM 0 mm ²
Css	Kemet	C0603C334K8RACTU Series= X7R	Cap= 330.0 nF ESR= 1.0 mOhm VDC= 10.0 V IRMS= 0.0 A	1	\$0.02	0603 5 mm ²

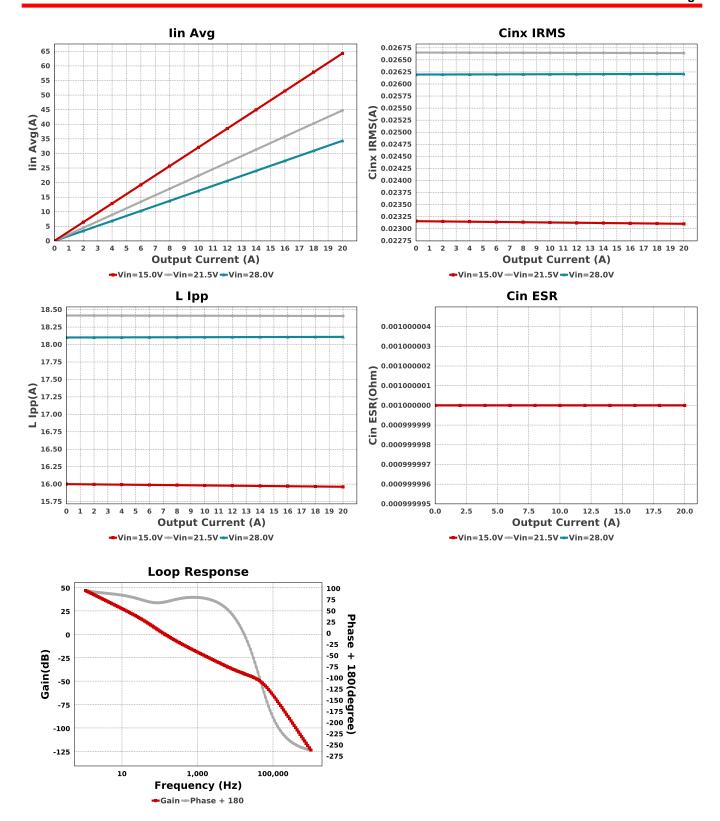
Name	Manufacturer	Part Number	Properties	Qty	Price	Footprint
Cvcc	TDK	C1005X5R1A475K050BC Series= X5R	Cap= 4.7 uF ESR= 3.417 mOhm VDC= 10.0 V IRMS= 2.7063 A	1	\$0.10	0402_065 3 mm ²
Cvout	MuRata	GRM155R71A104KA01D Series= X7R	Cap= 100.0 nF ESR= 1.0 mOhm VDC= 10.0 V IRMS= 0.0 A	1	\$0.01	0402 3 mm ²
Cvref	MuRata	GRM1555C1E471JA01D Series= C0G/NP0	Cap= 470.0 pF ESR= 1.0 mOhm VDC= 25.0 V IRMS= 0.0 A	1	\$0.02	0402 3 mm ²
L1	CUSTOM	CUSTOM	L= 6.507 μH 1.0 mOhm	1	NA	CUSTOM 0 mm ²
M1	NA	IdealFET	VdsMax= 148.0 V IdsMax= 30.0 Amps	1	NA	KCS0003B 80 mm ²
M2	NA	IdealFET	VdsMax= 148.0 V IdsMax= 30.0 Amps	1	NA	KCS0003B 80 mm ²
Rcomp	Yageo	RC0201FR-0710KL Series= ?	Res= 10.0 kOhm Power= 50.0 mW Tolerance= 1.0%	1	\$0.01	0201 2 mm ²
Rpg	Vishay-Dale	CRCW0603100KFKEA Series= CRCWe3	Res= 100.0 kOhm Power= 100.0 mW Tolerance= 1.0%	1	\$0.01	0603 5 mm ²
Rsense	CUSTOM	CUSTOM Series= ?	Res= 722.31 uOhm Power= 0.0 W Tolerance= 0.0%	1	NA	CUSTOM 0 mm ²
Rt	Vishay-Dale	CRCW0603221KFKEA Series= CRCWe3	Res= 221.0 kOhm Power= 100.0 mW Tolerance= 1.0%	1	\$0.01	0603 5 mm ²
Rvrefb	Vishay-Dale	CRCW040228K0FKED Series= CRCWe3	Res= 28.0 kOhm Power= 63.0 mW Tolerance= 1.0%	1	\$0.01	0402 3 mm ²
Rvreft	Vishay-Dale	CRCW04026K98FKED Series= CRCWe3	Res= 6.98 kOhm Power= 63.0 mW Tolerance= 1.0%	1	\$0.01	0402 3 mm ²
U1	Texas Instruments	LM5123QRGRRQ1	Switcher	1	\$1.59	RTE0016C-IPC_A 16 mm²











Operating Values

#	Name	Value	Category	Description
1.	Cin ESR	1.0 mOhm	Capacitor	Cin Capacitor ESR
2.	Cin IRMS	4.585 A	Capacitor	Input capacitor RMS ripple current
3.	Cin Pd	21.022 mW	Capacitor	Input capacitor power dissipation
4.	Cinx IRMS	22.938 mA	Capacitor	Bulk capacitor RMS ripple current
5.	Cinx Pd	526.14 pW	Capacitor	Bulk capacitor power dissipation
6.	Cout ESR	1.0 μOhm	Capacitor	Cout Capacitor ESR
7.	Coutx IRMS	5.688 A	Capacitor	Output capacitor_x RMS ripple current
8.	Coutx Pd	32.352 mW	Capacitor	Output capacitor_x power loss
9.	IC lpk	72.257 A	IC	Peak switch current in IC
10.	IC Tolerance	5.0 mV	IC	IC Feedback Tolerance
11.	ICThetaJA Effective	43.4 degC/W	IC	Effective IC Junction-to-Ambient Thermal Resistance

			_	
#	Name	Value	Category	Description
12.	lin Avg	64.302 A	IC	Average input current
13.	lpp percentage	24.834 %	Inductor	Inductor ripple current percentage (with respect to average inductor
				current)
14.	L lpp	15.962 A	Inductor	Peak-to-peak inductor ripple current
15.	L Pd	4.153 W	Inductor	Inductor power dissipation
16.	L1 DCR	1.0 mOhm	Inductor	L1 DCR
17.	L1 Irms	64.441 A	Inductor	Inductor ripple current
18.	Cin Pd	21.022 mW	Power	Input capacitor power dissipation
19.	Cinx Pd	526.14 pW	Power	Bulk capacitor power dissipation
20.	Coutx Pd L Pd	32.352 mW 4.153 W	Power Power	Output capacitor_x power loss Inductor power dissipation
21. 22.	BOM Count	4.133 W	System	Total Design BOM count
22.	DOW Count	22	Information	Total Design Bow Count
23.	Cross Freq	86.191 Hz	System	Bode plot crossover frequency
20.	010001104	00.101112	Information	Bodo plot dioddover frequency
24.	Duty Cycle	68.884 %	System	Duty cycle
	2 41, 0, 0.0	00.00 . 70	Information	zal, ojolo
25.	FootPrint	1.177 k mm²	System	Total Foot Print Area of BOM components
			Information	'
26.	Frequency	99.045 kHz	System	Switching frequency
			Information	
27.	Gain Marg	-38.687 dB	System	Bode Plot Gain Margin
			Information	
28.	lout	20.0 A	System	lout operating point
			Information	
29.	lout transient step used	110.0 A	System	Custom Transient current step requirement that was used for Cout
	for Cout calculations		Information	selection (A).
30.	Low Freq Gain	41.227 dB	System	Gain at 1Hz
0.4		0014	Information	
31.	Mode	CCM	System Information	Conduction Mode
32.	Overshoot Value	1.512 mV	System	Theoretical Vout Overshoot Value
32.	Oversiloot value	1.5121110	Information	Theoretical vout Overshoot value
33.	Phase Marg	67.822 deg	System	Bode Plot Phase Margin
00.	Tridoo Marg	01.022 409	Information	Bodo i locci naco margin
34.	Pout	960.0 W	System	Total output power
•			Information	
35.	Total BOM	NA	System	Total BOM Cost
			Information	
36.	Undershoot Value	1.455 mV	System	Theoretical Vout Undershoot Value
			Information	
37.	Vin	15.0 V	System	Vin operating point
			Information	
38.	Vin p-p	100.281 mV	System	Peak-to-peak input voltage
			Information	
39.	Vout	48.0 V	System	Operational Output Voltage
40		40.014	Information	W . A
40.	Vout Actual	48.0 V	System	Vout Actual calculated based on selected voltage divider resistors
44	Vout Dinnle	4.0.0/	Information	Custom maximum autaut ripple requirement that was used for Court
41.	'''	1.0 %	System Information	Custom maximum output ripple requirement that was used for Cout
	requirement used for Cout calculations		momation	selection(% of Vout).
42.	Vout Tolerance	18.286 m%	System	Vout Tolerance based on IC Tolerance (no load) and voltage divider
+∠.	vout roleiance	10.200 111/0	Information	resistors if applicable
43.	Vout p-p	56.513 mV	System	Peak-to-peak output ripple voltage
10.	. 501 P P	55.0101111	Information	. San to pour output rippio ronago
44.	Vout transient	3.0 %	System	Custom Transient voltage change requirement that was used for Cout
	requirement used for		Information	selection (% of Vout).
	Cout calculations			,

Design Inputs

Name	Value	Description	
lout	20.0	Maximum Output Current	
SoftStart	10.0 ms	Soft Start Time (ms)	
VinMax	28.0	Maximum input voltage	
VinMin	15.0	Minimum input voltage	
Vout	48.0	Output Voltage	
base_pn	LM5123-Q1	Base Product Number	
source	DC	Input Source Type	
Ta	30.0	Ambient temperature	
UserFsw	100.0 k	Customer Selected Frequency	

WEBENCH® Assembly

Component Testing

Some published data on components in datasheets such as Capacitor ESR and Inductor DC resistance is based on conservative values that will guarantee that the components always exceed the specification. For design purposes it is usually better to work with typical values. Since this data is not always available it is a good practice to measure the Capacitance and ESR values of Cin and Cout, and the inductance and DC resistance of L1 before assembly of the board. Any large discrepancies in values should be electrically simulated in WEBENCH to check for instabilities and thermally simulated in WebTHERM to make sure critical temperatures are not exceeded.

Soldering Component to Board

If board assembly is done in house it is best to tack down one terminal of a component on the board then solder the other terminal. For surface mount parts with large tabs, such as the DPAK, the tab on the back of the package should be pre-tinned with solder, then tacked into place by one of the pins. To solder the tab town to the board place the iron down on the board while resting against the tab, heating both surfaces simultaneously. Apply light pressure to the top of the plastic case until the solder flows around the part and the part is flush with the PCB. If the solder is not flowing around the board you may need a higher wattage iron (generally 25W to 30W is enough).

Initial Startup of Circuit

It is best to initially power up the board by setting the input supply voltage to the lowest operating input voltage 15.0V and set the input supply's current limit to zero. With the input supply off connect up the input supply to Vin and GND. Connect a digital volt meter and a load if needed to set the minimum lout of the design from Vout and GND. Turn on the input supply and slowly turn up the current limit on the input supply. If the voltage starts to rise on the input supply continue increasing the input supply current limit while watching the output voltage. If the current increases on the input supply, but the voltage remains near zero, then there may be a short or a component misplaced on the board. Power down the board and visually inspect for solder bridges and recheck the diode and capacitor polarities. Once the power supply circuit is operational then more extensive testing may include full load testing, transient load and line tests to compare with simulation results.

Load Testing

The setup is the same as the initial startup, except that an additional digital voltmeter is connected between Vin and GND, a load is connected between Vout and GND and a current meter is connected in series between Vout and the load. The load must be able to handle at least rated output power + 50% (7.5 watts for this design). Ideally the load is supplied in the form of a variable load test unit. It can also be done in the form of suitably large power resistors. When using an oscilloscope to measure waveforms on the prototype board, the ground leads of the oscilloscope probes should be as short as possible and the area of the loop formed by the ground lead should be kept to a minimum. This will help reduce ground lead inductance and eliminate EMI noise that is not actually present in the circuit.



Design Assistance

- 1. Master key: B09A30C71D1B3D0ED38ABFDAA1D91EC8[v1]
- 2. LM5123-Q1 Product Folder: http://www.ti.com/product/LM5123%2DQ1: contains the data sheet and other resources.

Important Notice and Disclaimer

TI provides technical and reliability data (including datasheets), design resources (including reference designs), application or other design advice, web tools, safety information, and other resources AS IS and with all faults, and disclaims all warranties. These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

Providing these resources does not expand or otherwise alter TI's applicable Terms of Sale or other applicable terms available either on ti.com or provided in conjunction with TI products.