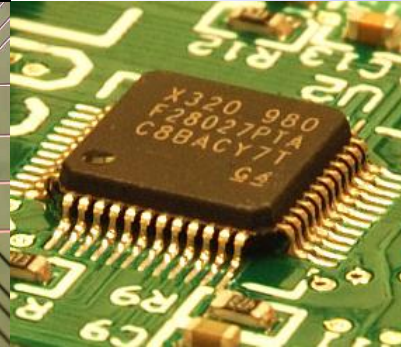
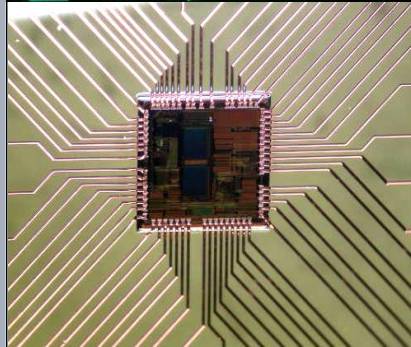
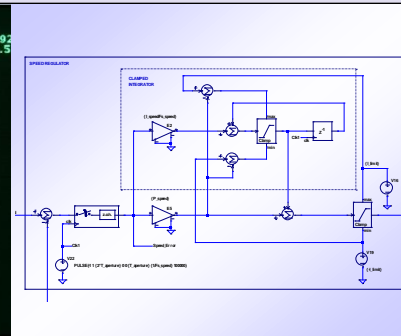
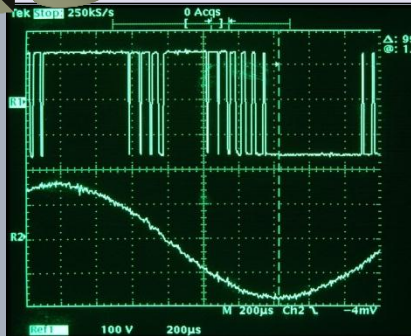
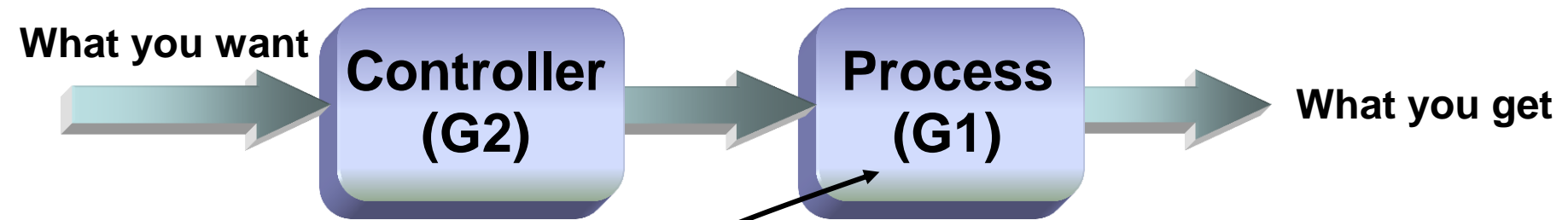


Teaching Old Motors New Tricks



Dave Wilson

Feed-Forward Approach

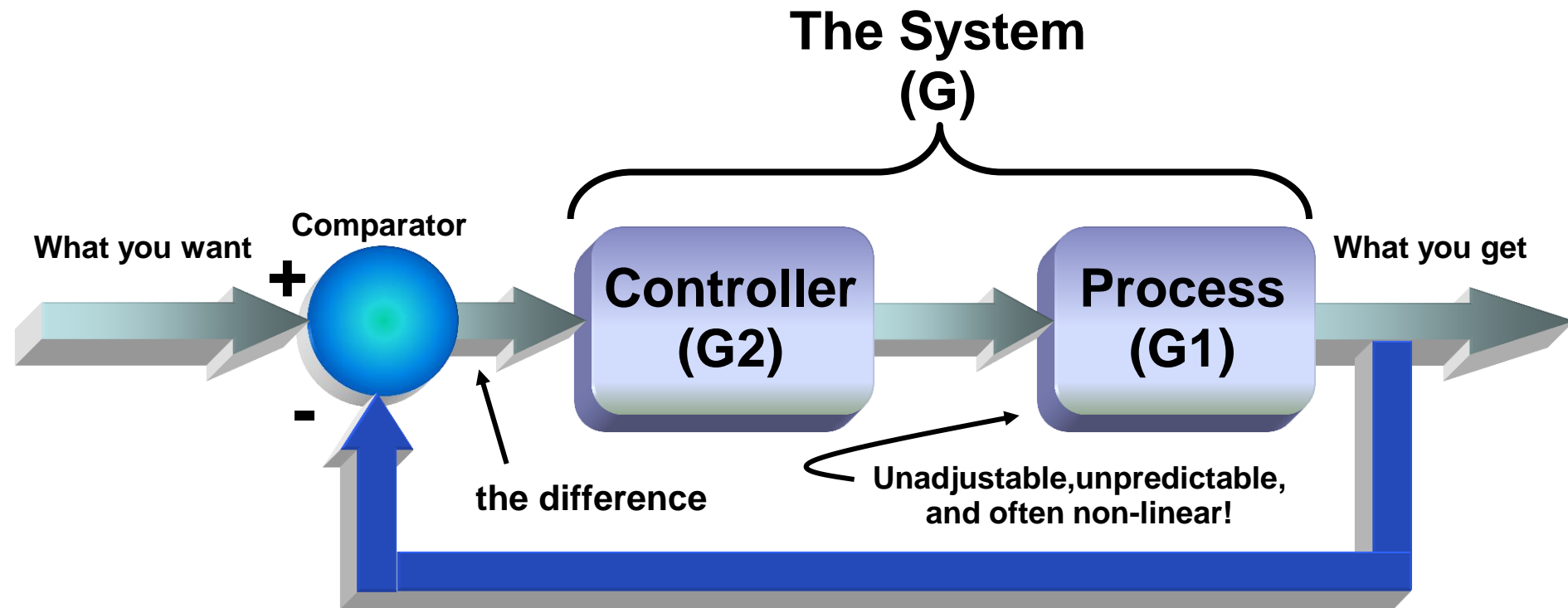


Unadjustable, but perfectly predictable

$$G2 = \frac{1}{G1}$$

**Everything is
beautiful and linear!**

Feedback Approach

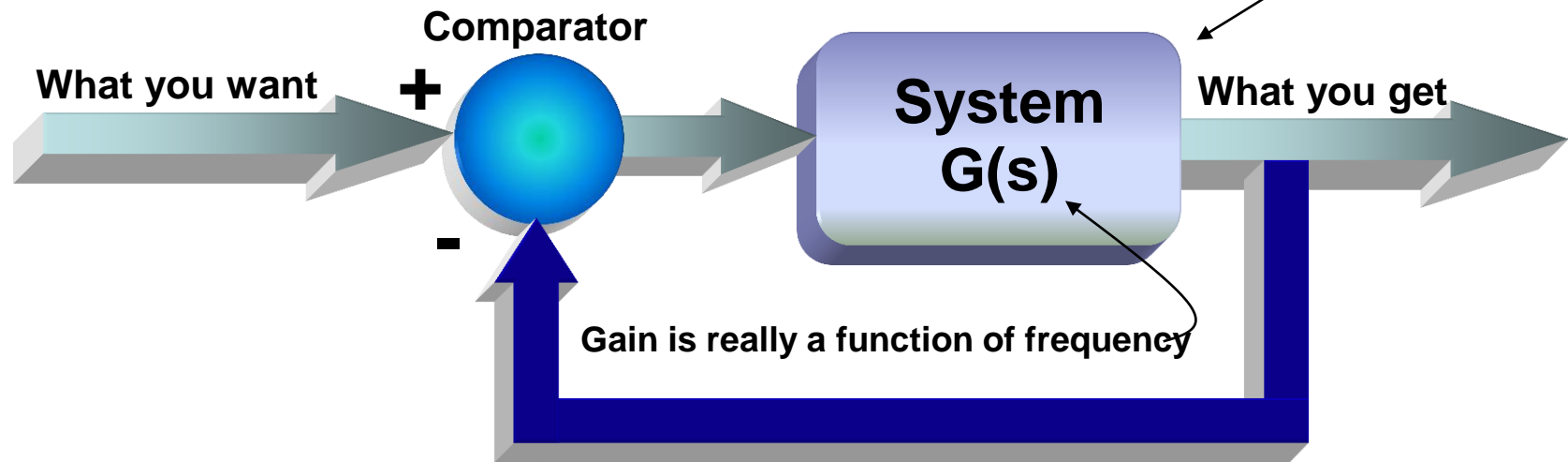


$$\text{What you get} = \left[\text{What you want} \right] \left(\frac{G}{1 + G} \right)$$

What if $G = 1$? What about 100?

Frequency Response of a Feedback Loop

Every system in nature has poles in its transfer function

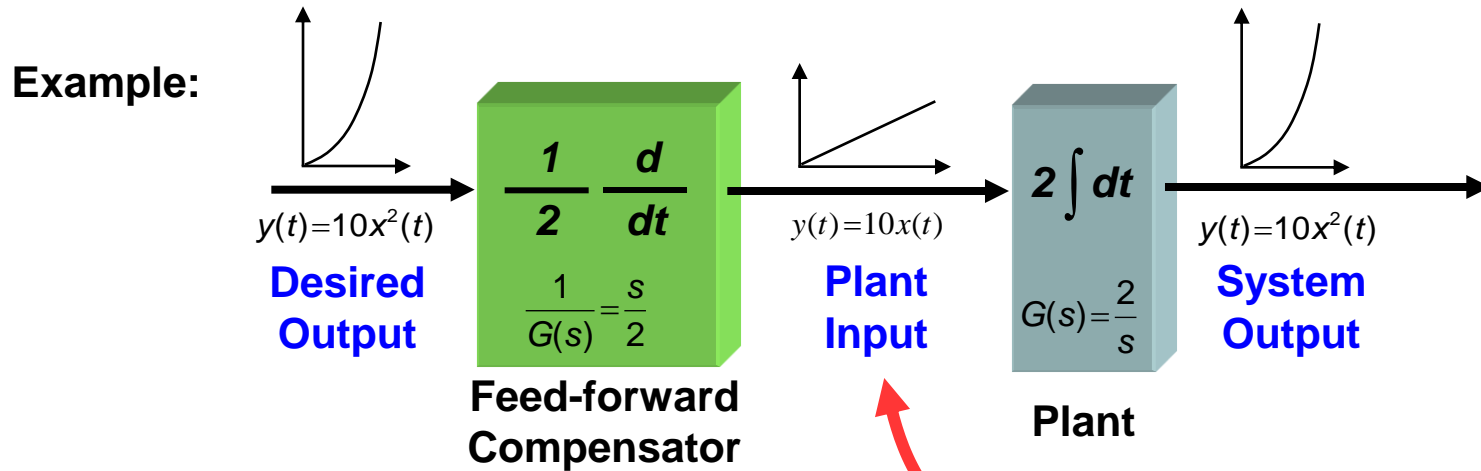


Poles negatively impact the control system in the following ways:

1. Gain is reduced at higher frequencies
2. Phase delay is introduced (90° for every real pole)

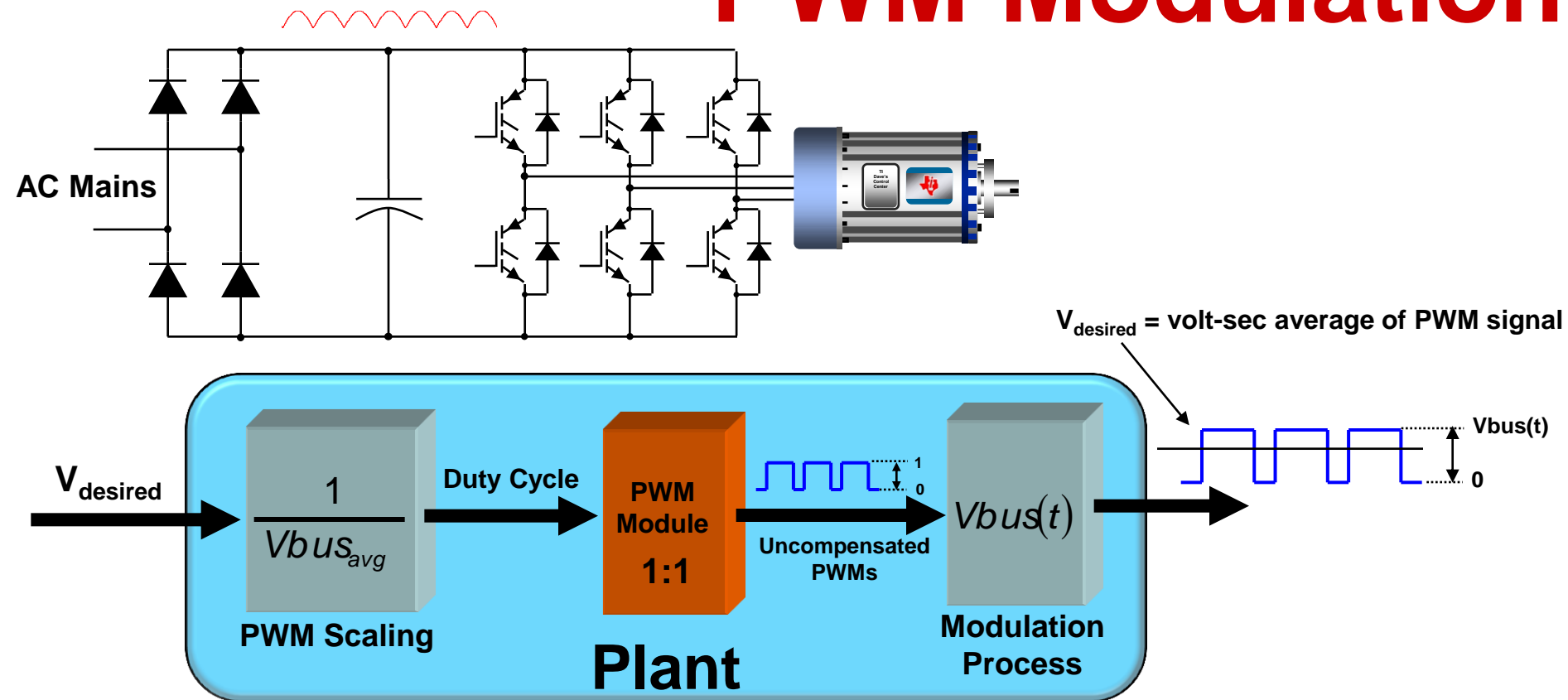
What happens if the phase delay = 180° and gain = 1 ?

Designing a Feed-Forward Compensator



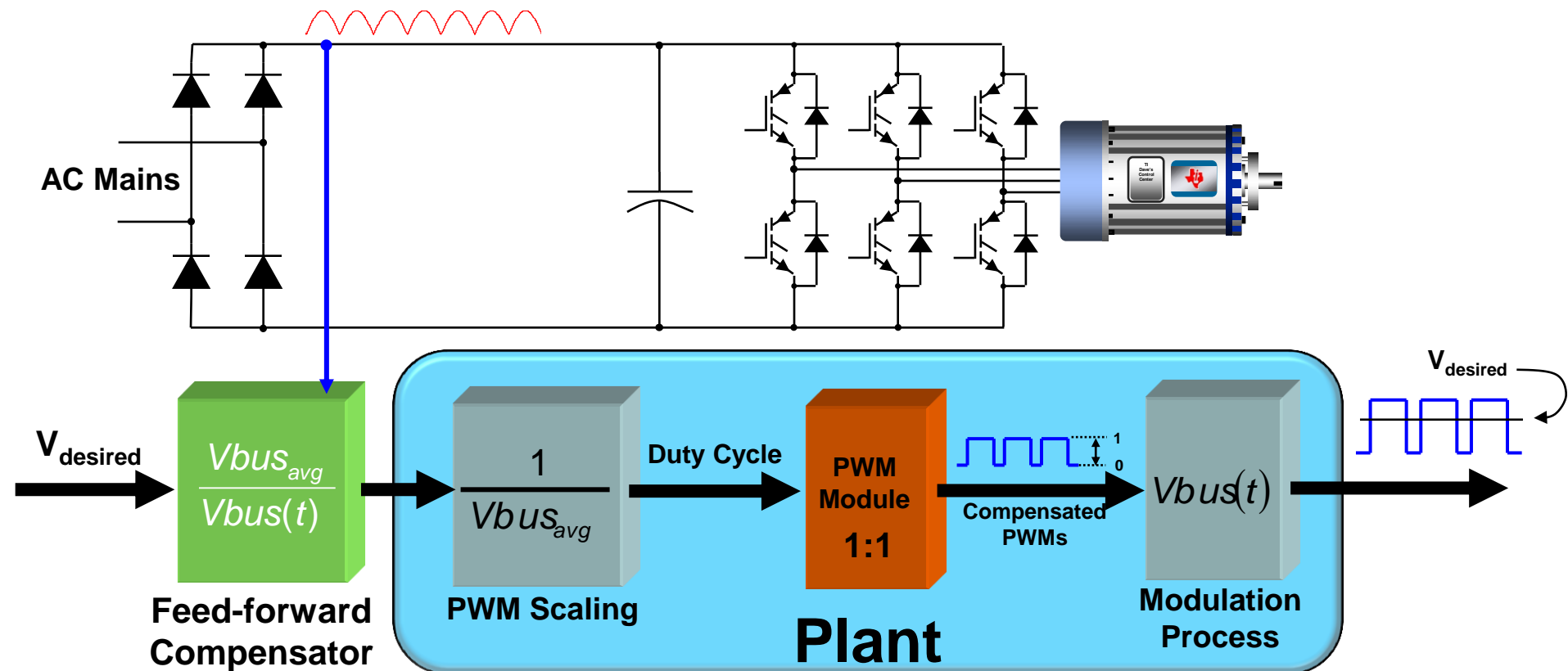
The thought process goes something like this: “To achieve a desired output, what stimulation signal is required at the plant input to get that output? By knowing the plant transfer function in the forward direction $[G(s)]$, I find the transfer function looking backwards through the plant $[1/G(s)]$, and that becomes the transfer function of my feedforward filter.”

Real Feed-Forward Example: PWM Modulation

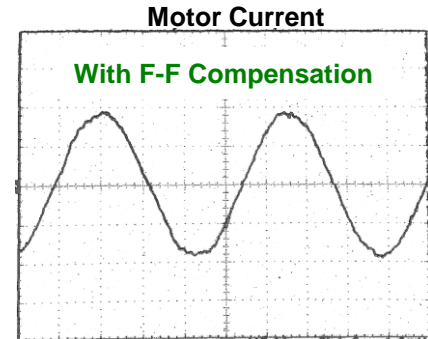
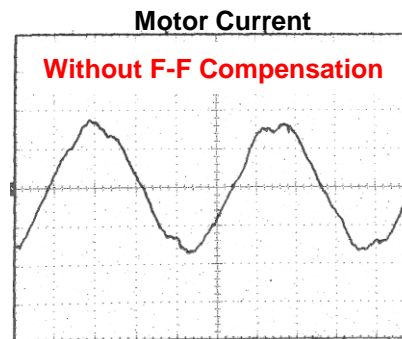


When $V_{bus}(t)$ is equal to $V_{bus_{avg}}$, then “what you want” and “what you get” are equal. However, when they are NOT equal, the output deviates from the desired input.

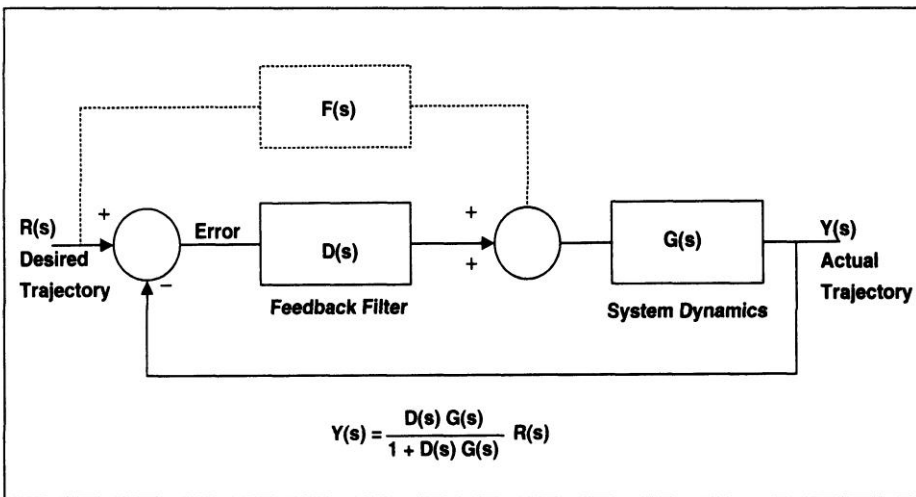
Solution: Feed-forward Compensation



When you multiply through the gain blocks, you will see that the composite gain = 1, which again implies that “what you get” equals “what you want”.

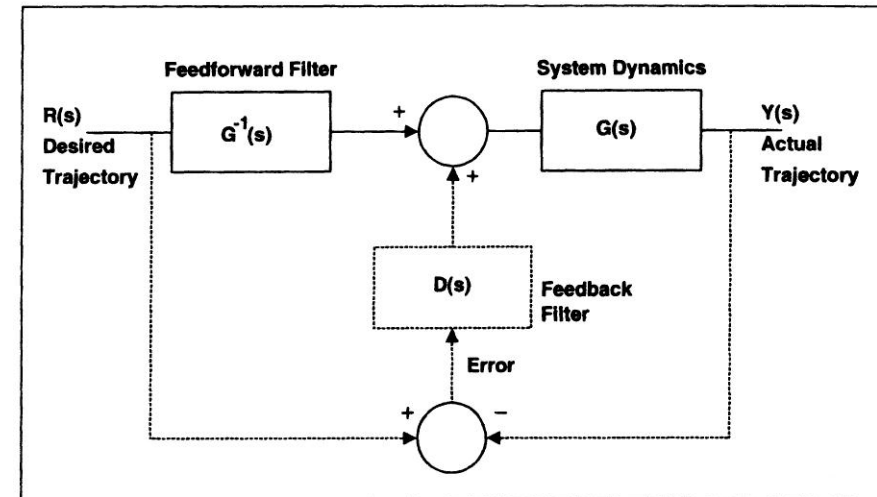


Feedback or Feedforward... Which is Best?



Feedback Philosophy.

More traditional approach
Best for disturbance rejection

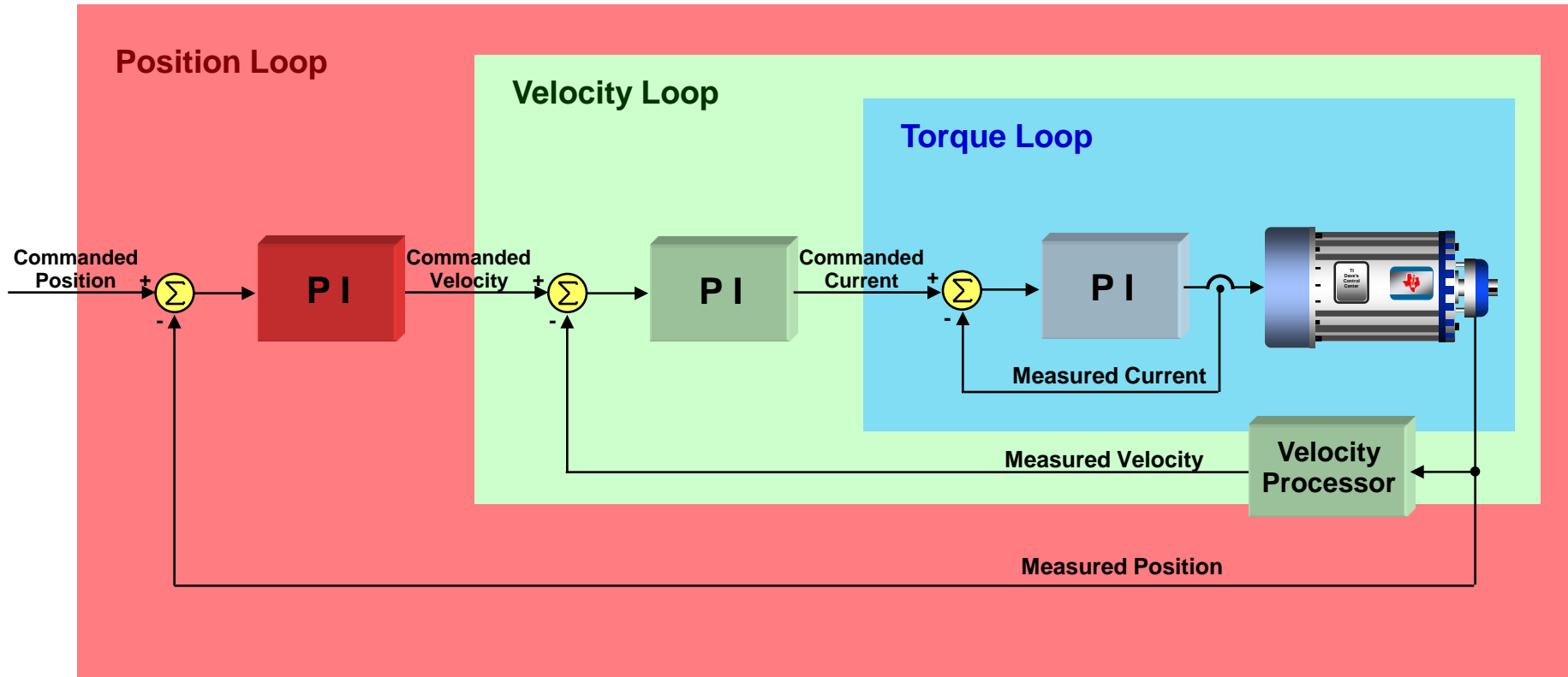


Feedforward Philosophy.

Better stability
Best for trajectory tracking
Requires System Dynamics Knowledge

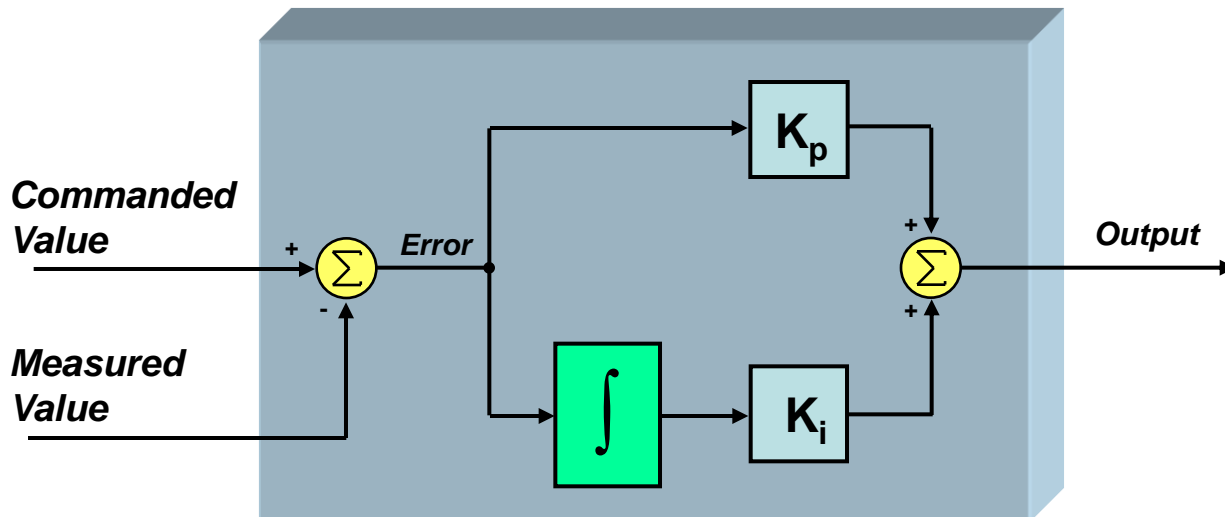
*Source: Myths and Realities of Feedforward for Motion Control Systems,
Curtis Wilson, Delta Tau Data Systems, PCIM – August, 1994*

Cascaded Control Structures



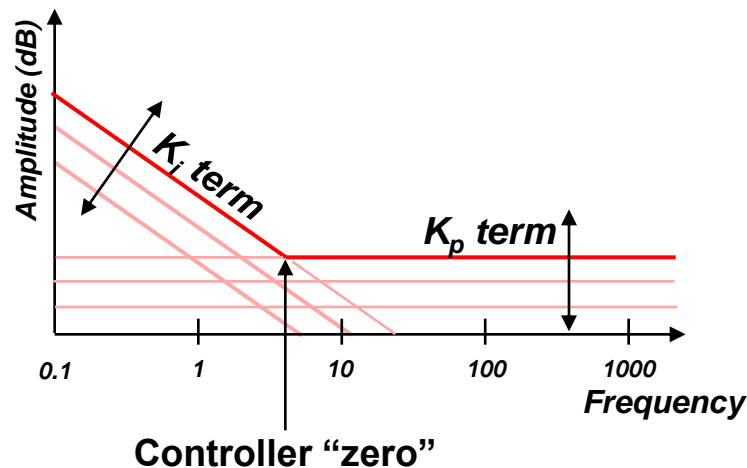
What do you think the bandwidth requirements are for each cascaded loop?

Parallel PI Controller

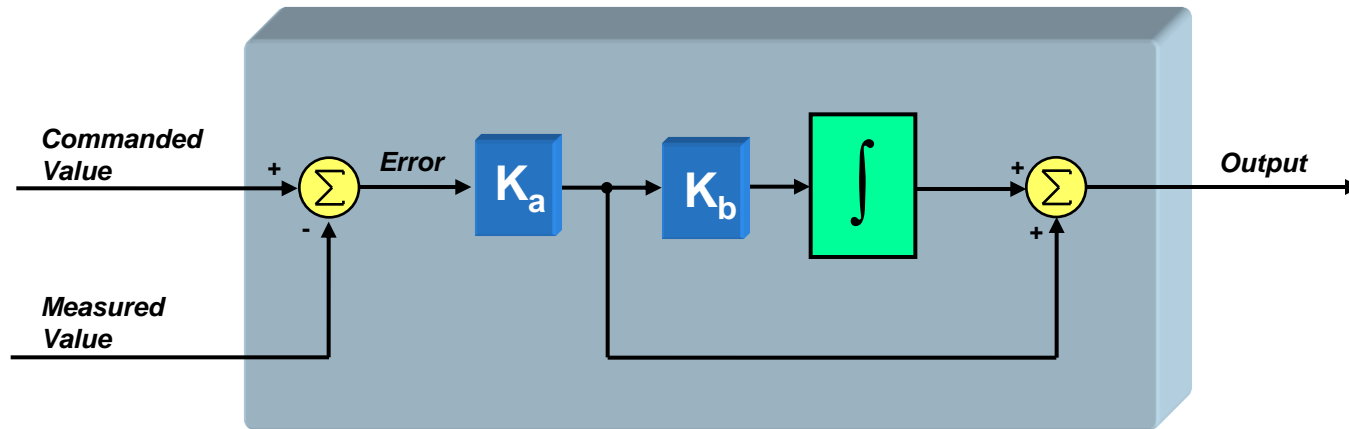


K_p term specifies the gain at higher frequencies

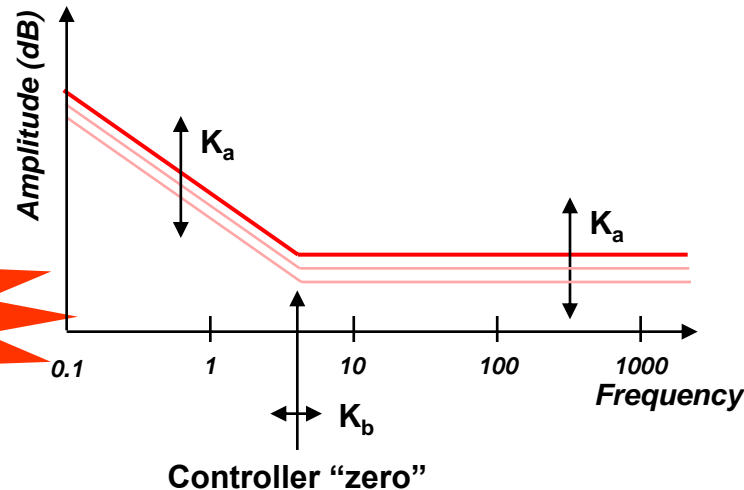
K_i term specifies the gain at lower frequencies



Series PI Controller



K_a is simply a gain term for all frequencies
 K_b is equal to the controller's "zero" in rad/sec

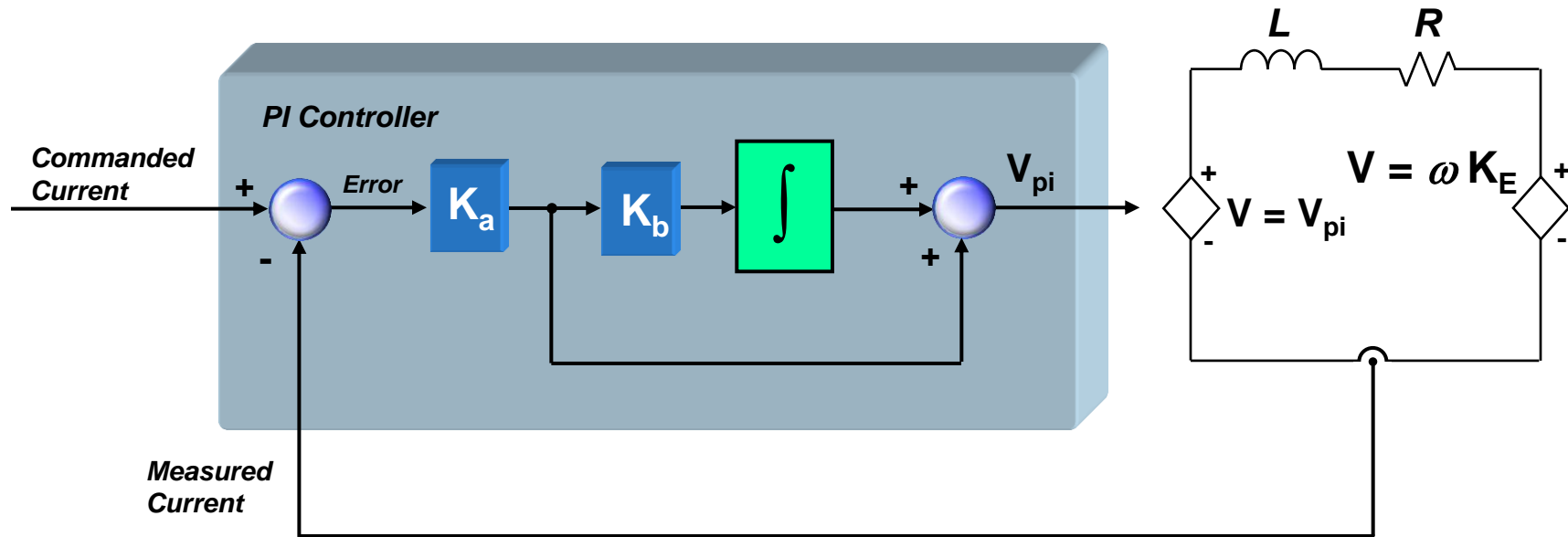


$$K_a = K_p$$

$$K_b = \frac{K_i}{K_p}$$

**Popular with
Current Controllers**

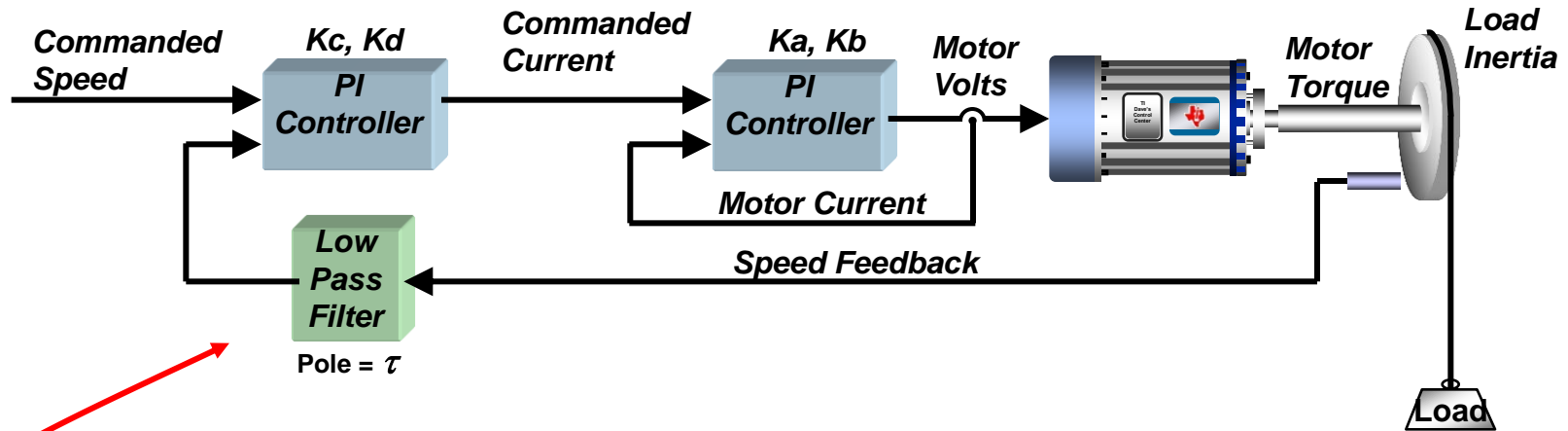
Current PI Controller Coefficients



$$K_a = L \cdot \text{Current Bandwidth (rad/sec)}, \quad K_b = \frac{R}{L}$$

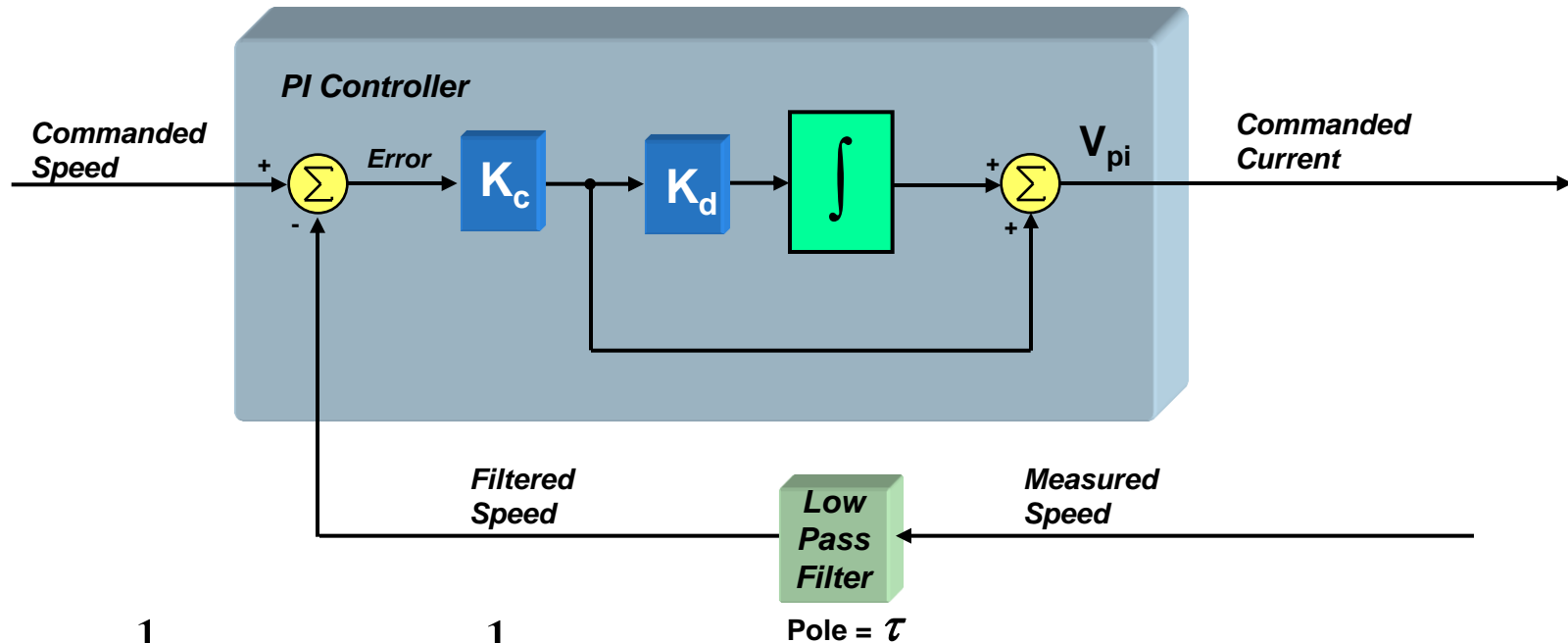
Motor	Rd	Rq	Ld	Lq
PMSM	R_s	R_s	L_s	L_s
ACIM	R_s	$R_s + R_r$	$L_s \left(1 - \frac{L_m^2}{L_r L_s} \right)$	$L_s \left(1 - \frac{L_m^2}{L_r L_s} \right)$
IPM	R_s	R_s	L_{s_d}	L_{s_q}

Cascaded Velocity Controller



Due to the techniques used for velocity feedback synthesis, the velocity signal must be filtered in most cases.

Velocity PI Controller Coefficients



$$K_c = \frac{1}{\delta K \tau}, \quad K_d = \frac{1}{\delta^2 \tau}$$

$$K = \frac{3PK_e}{4J} \quad \text{for permanent magnet motors}$$

$$K = \frac{3}{4} P \frac{Lm^2}{J Lr} I_d \quad \text{for AC Induction motors}$$

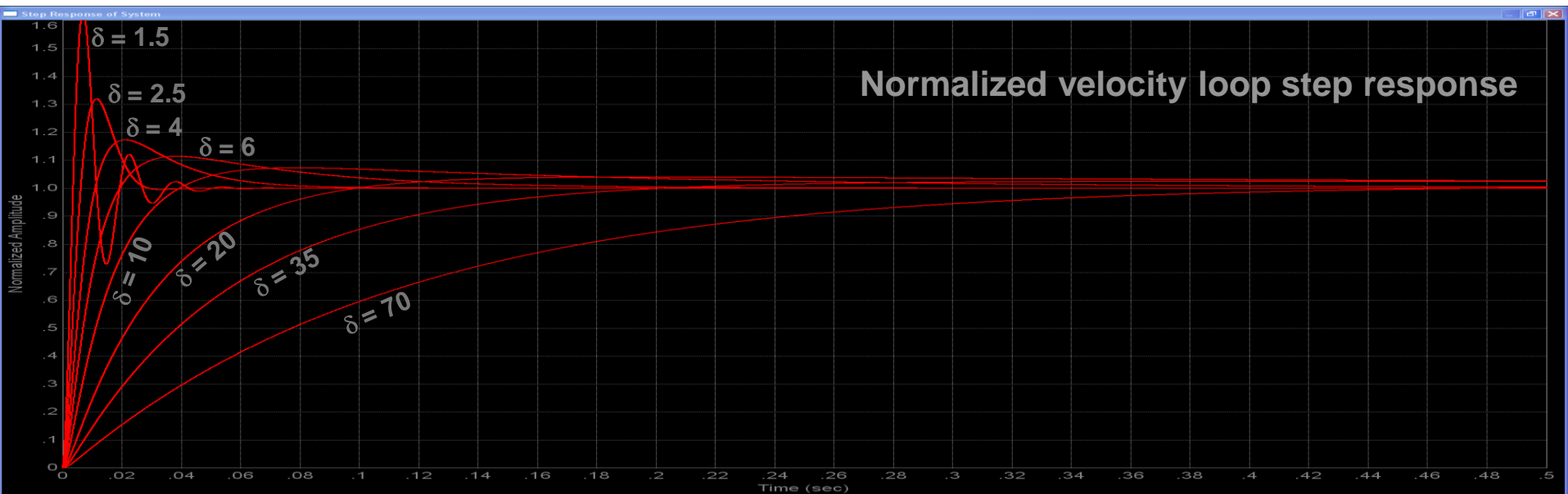
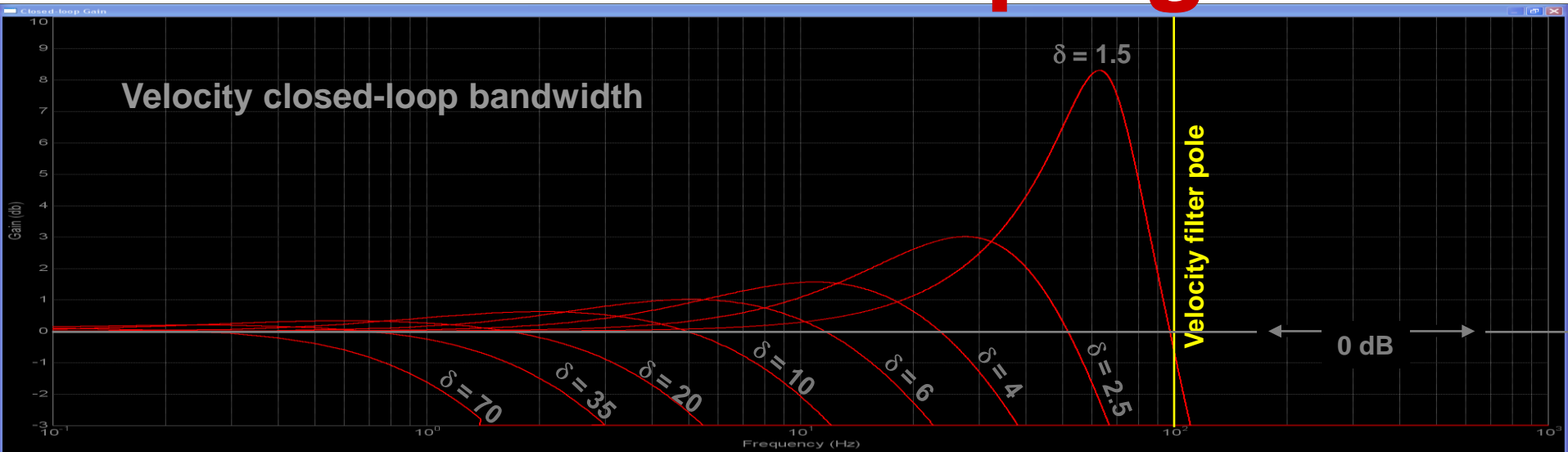
K_e = Back-EMF constant
 P = Number of motor poles
 J = System Inertia (as seen by the motor)
 τ = LPF Pole

δ = the Damping Factor

Single tuning adjustment!

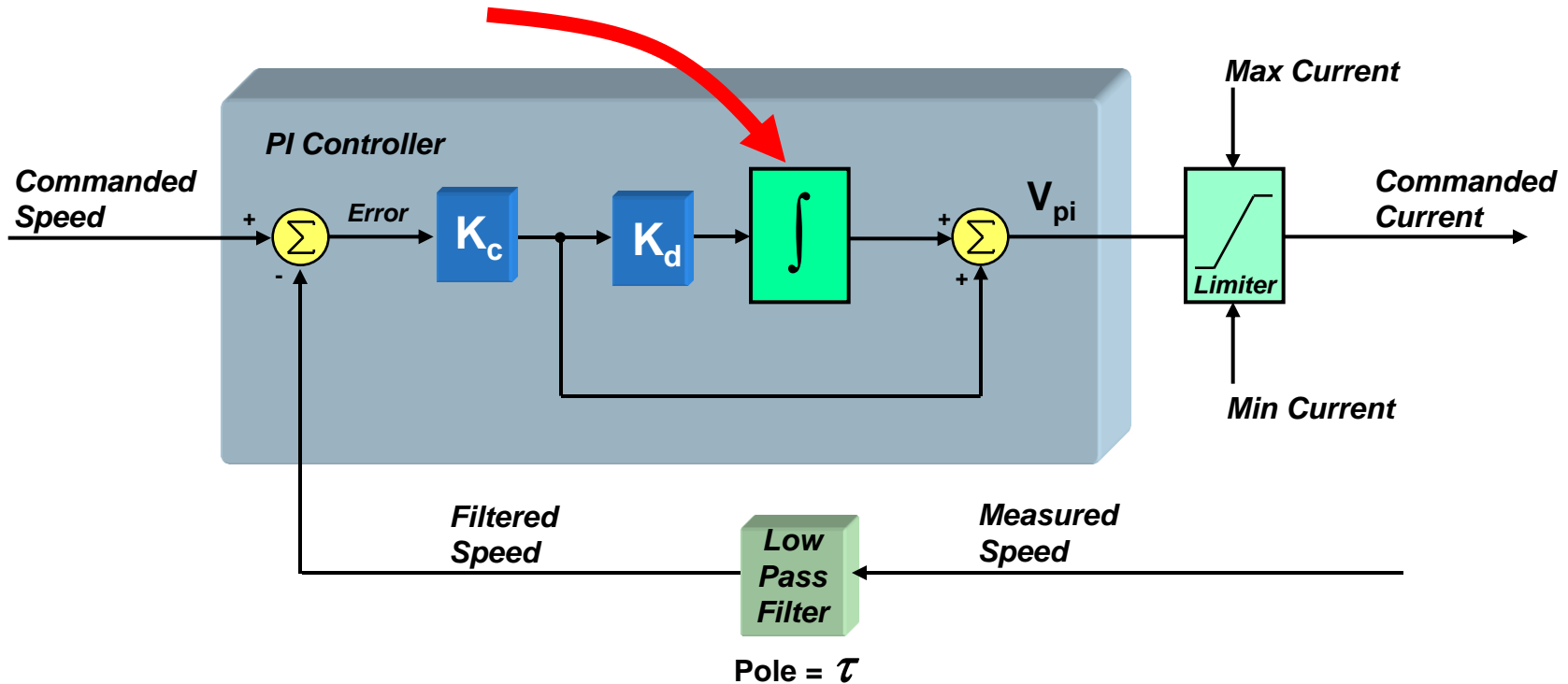
(Equations assume that the Current Controller closed-loop bandwidth is greater than 3τ)

Damping Factor

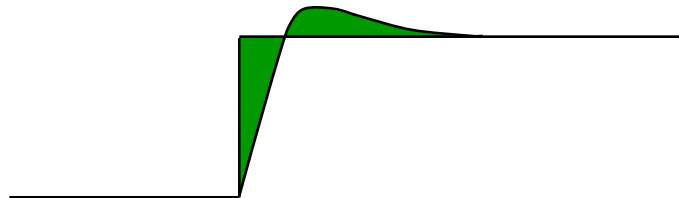


For more information on PI tuning, read my blog series at www.ti.com/motorblog

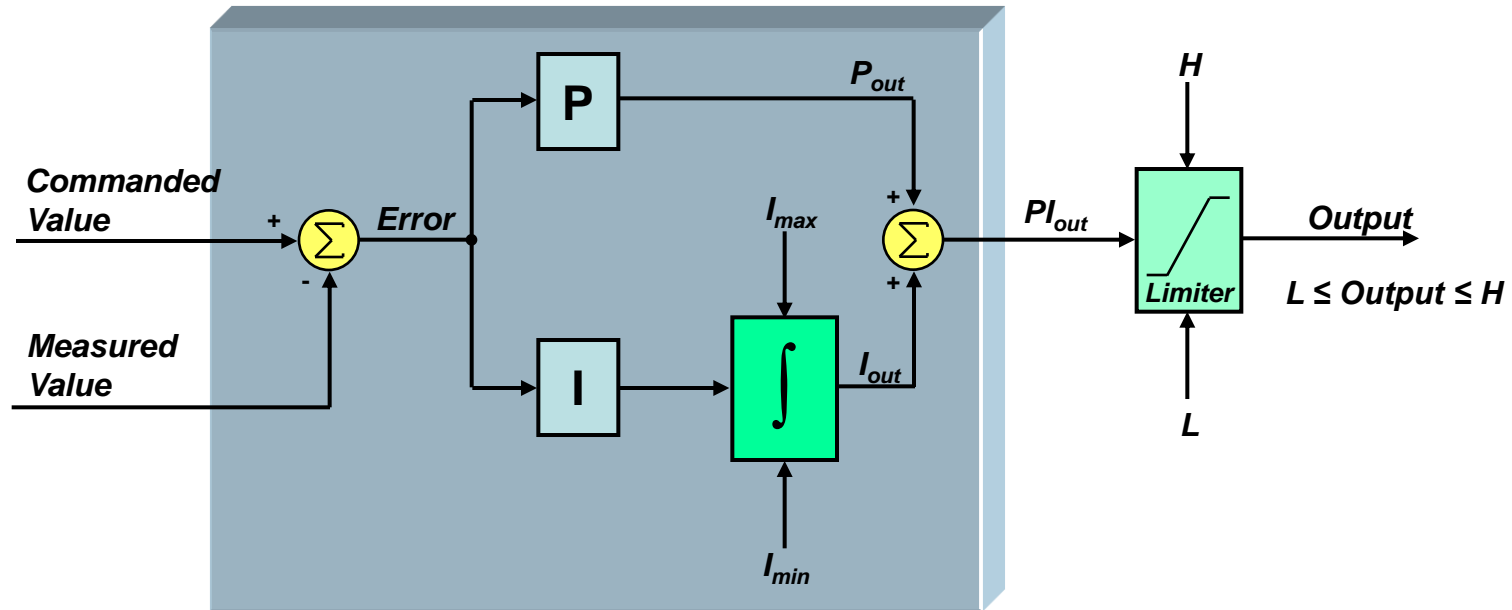
Integrator Windup



Unfortunately, the Integrator exhibits “windup” under sudden transient conditions.

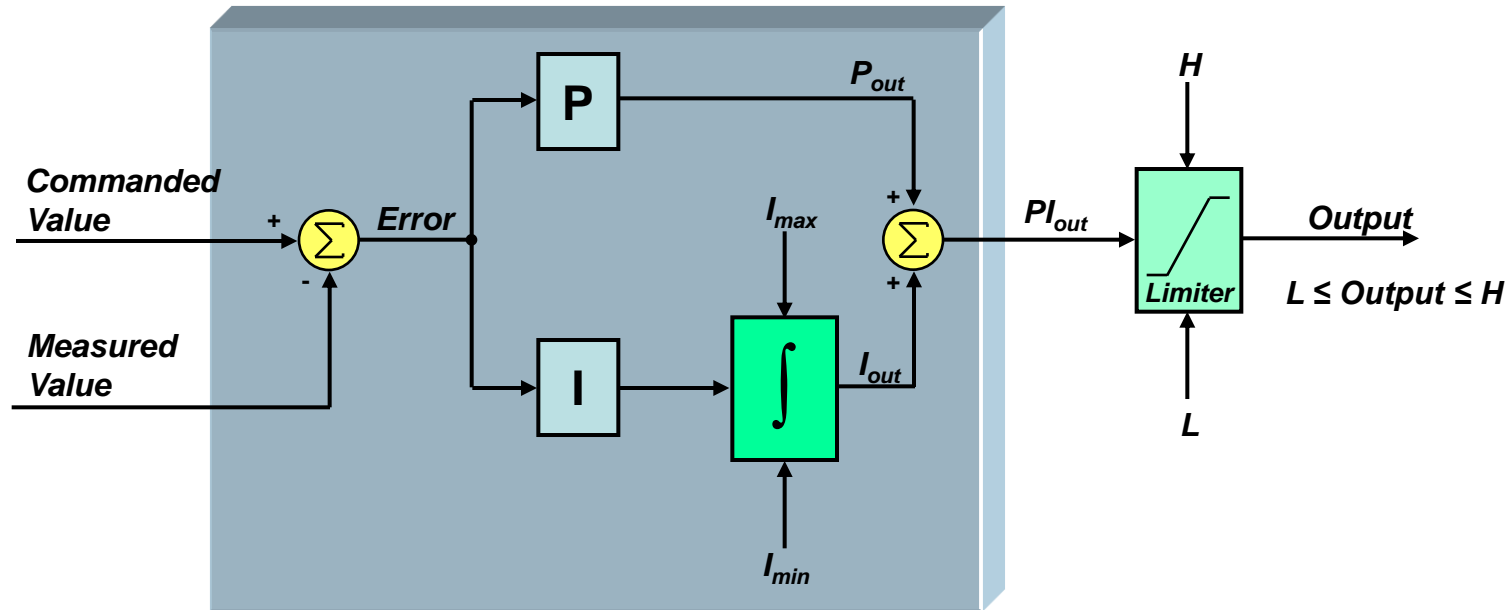


Simple Static Integrator Clamping



$$I_{min} = L, I_{max} = H$$

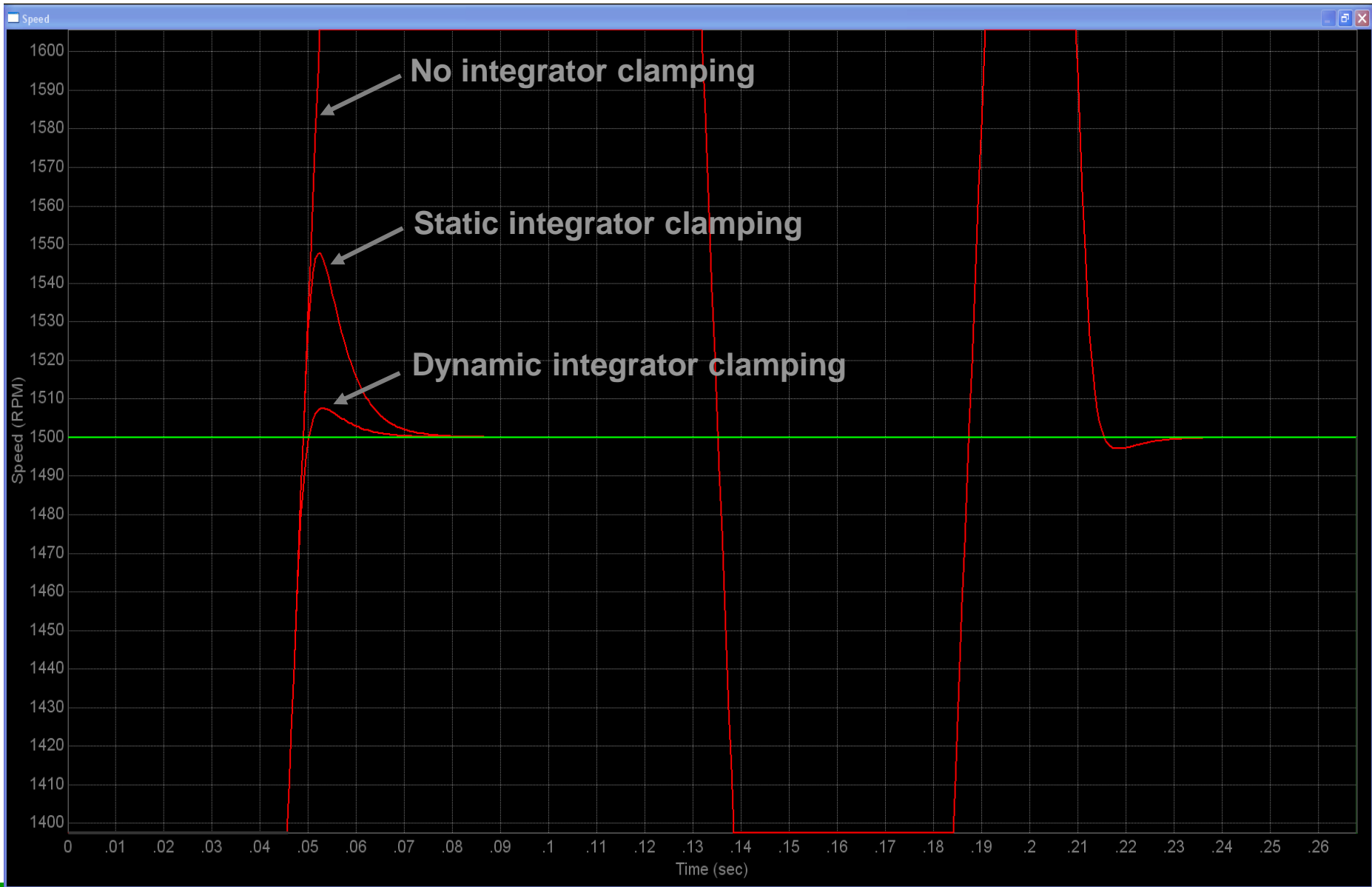
Dynamic Integrator Clamping



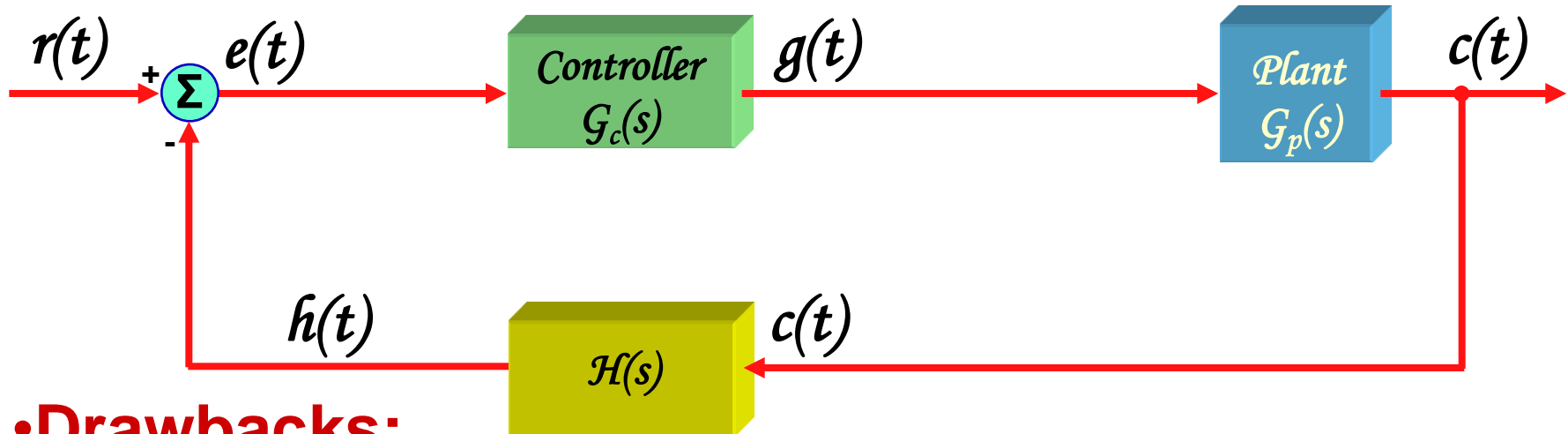
$$L \leq P_{out} + I_{out} \leq H$$

for minimum condition: $L = P_{out} + I_{min} \rightarrow I_{min} = \text{Min}(L - P_{out}, 0)$
 for maximum condition: $H = P_{out} + I_{max} \rightarrow I_{max} = \text{Max}(H - P_{out}, 0)$

Comparison of Clamping Techniques



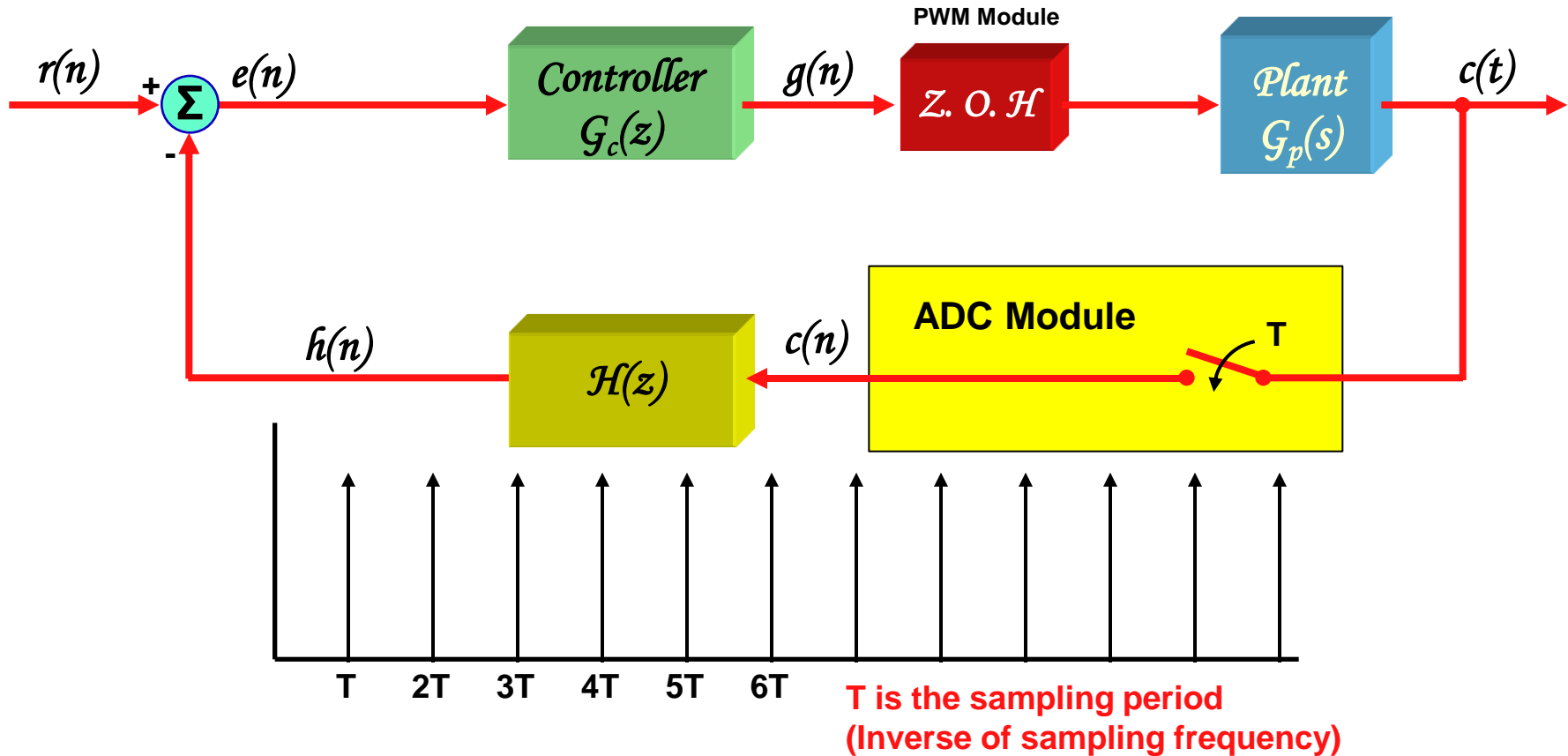
Typical Analog Control System



•Drawbacks:

- Low noise immunity
- Filter characteristics change with temperature
- Little flexibility
- Component aging
- Power supply variation
- Lot-to-lot manufacturing
- Requires adjustments
- Critical component specification, especially for high order filters

Typical Digital Control System



In most digital controllers, the control block ($G_c(z)$) is implemented as an IIR filter to minimize phase delay.

The 10 Commandments of Digital Control

There are 10 facts of life about digital control systems that your mother never told you!

If you can handle the truth, read about them on my blog site:

www.ti.com/motorblog