

8.3 A requirement exists to simulate in a digital computer an analogue system with the following normalized characteristic:

$$H(s) = \frac{1}{s^2 + \sqrt{2}s + 1}$$

Obtain a suitable transfer function using:

- (1) the impulse invariant method, and
- (2) the bilinear transform method.

Assume a sampling frequency of 5 kHz and a 3 dB cutoff frequency of 1 kHz.

### Solution

(1) The transfer function of the denormalised analog filter is obtained by replacing  $s$  by  $s/\alpha$  (where  $\alpha$  is the cutoff frequency relative to the sampling frequency, i.e.  $\alpha = 2\pi \times 1/5 = 1.25395$ ):

$$H'(s) = H(s) \Big|_{s \rightarrow \frac{s}{\alpha}} = \frac{\alpha^2}{s^2 + \sqrt{2}\alpha s + \alpha^2}$$

For convenience,  $H(s)$  will be written in the general form:

$$H'(s) = \frac{B_0}{s^2 + A_1 s + A_0}$$

To apply the impulse invariant method, the s-plane transfer function is first expanded into partial fractions as described in section 8.6.

$$H'(s) = \frac{B_0}{s^2 + A_1 s + A_0} = \frac{c_1}{s - p_1} + \frac{c_2}{s - p_2}$$

where  $c_1$  and  $c_2$  are partial fraction coefficients and  $p_1$  and  $p_2$  are the poles of  $H'(s)$ , given by:

$$p_{1,2} = -\frac{A_1}{2} \pm \left[ \left( \frac{A_1}{2} \right)^2 - A_0 \right]^{\frac{1}{2}}$$

$$p_r = -\frac{A_1}{2}; \quad p_i = \left\{ - \left[ \left( \frac{A_1}{2} \right)^2 - A_0 \right] \right\}^{\frac{1}{2}}$$

Multiplying both sides of  $H'(s)$  by  $(s - p_1)(s - p_2)$  and equating coefficients of  $s$  and the constant terms we have:

$$B_0 = -(c_1 p_2 + c_2 p_1)$$

$$0 = c_1 + c_2$$

Solving for  $c_1$  and  $c_2$  we have:  $c_1 = \frac{B_0}{p_1 - p_2}$  and  $c_2 = -c_1$

Applying the impulse invariant transformation to Equation we obtain the transfer function:

$$H(z) = \frac{c_1}{1 - e^{p_r T} z^{-1}} + \frac{c_2}{1 - e^{p_i T} z^{-1}}$$

$$= \frac{2c_r - [c_r \cos(p_r T) + c_i \sin(p_r T)]2e^{p_r T} z^{-1}}{1 - 2e^{p_r T} \cos(p_r T) z^{-1} + e^{2p_r T} z^{-2}} \quad (\text{see Equation 8A.8}).$$

where  $p_r$  and  $p_i$  are the real and imaginary parts of  $p_1$ ,  $c_r$  and  $c_i$  are the real and imaginary parts of  $c_1$ . From the expressions for  $p_1$  above, the real and imaginary parts of  $p_1$  are:

$$p_r = -\frac{A_1}{2}; \quad p_i = \left\{ -\left[ \left( \frac{A_1}{2} \right)^2 - A_0 \right] \right\}^{\frac{1}{2}}$$

The real and imaginary parts of  $c_1$  are:  $c_r = 0$ ;  $c_i = -\frac{B_0}{2p_i} j$ . Now for the problem,

$A_0 = 1.25395$ ,  $A_1 = \sqrt{2}\alpha =$ ;  $B_0 = \alpha^2$ . Using these values in the equation above, we have:

$$p_r = -0.888576, \quad p_i = 0.888576, \quad 2c_r = 0, \quad c_i = -0.888576j$$

$$[c_r \cos(p_r T) + c_i \sin(p_r T)]2e^{p_r T} = 0.567258, \quad 2e^{p_r T} \cos(p_r T) = 0.518588 \quad \text{and}$$

$$e^{2p_r T} = 0.169119 \quad (T=1 \text{ is assumed}). \quad \text{Using these values leads to:}$$

$$H(z) = \frac{0.567258z^{-1}}{1 - 0.518588z^{-1} + 0.169119z^{-2}}$$

(2) The pre-warped cutoff frequency,  $\omega_p = \tan(\omega_p T / 2) = \tan(\pi \times 1000 / 5000) = 0.72654$ . The scaled s-plane transfer function,  $H'(s)$ , is:

$$H'(s) = H(s) \Big|_{s=\omega_p} = \frac{1}{\left( s / \omega_p \right)^2 + \sqrt{2}\alpha s / \omega_p + 1}$$

$$= \frac{\omega_p^2}{s^2 + \sqrt{2}\omega_p s + \omega_p^2} = \frac{0.52786}{s^2 + 1.02748s + 0.52786}$$

Applying the BZT and simplifying gives

$$H(z) = H'(s) \Big|_{s=\frac{z-1}{z+1}} = \frac{0.52786}{\left( \frac{z-1}{z+1} \right)^2 + 1.02748 \left( \frac{z-1}{z+1} \right) + 0.52786}$$

$$= \frac{0.20657(1 - 2z^{-1} + z^{-2})}{1 - 0.36953z^{-1} + 0.19582z^{-2}}$$

8.4 Determine, using the BZT method, the transfer function and difference equation for the digital equivalent of the resistance-capacitance (RC) filter shown in Figure 8.43 Assume a sampling frequency of 150 Hz and a cut off frequency of 30 Hz.

**Solution**

The normalized transfer function for the RC filter is  $H(s) = \frac{1}{s+1}$ . The cutoff frequency for the digital filter is

$\omega_p = 2\pi \times 30 \text{ rad}$ . The cutoff frequency for the equivalent analog filter, after pre-warping, is

$\omega'_p = \tan(\omega_p T / 2)$ . With  $T = 1/150 \text{ Hz}$ ,  $\omega'_p = \tan(\pi / 5) = 0.7265$ . The denormalized analog filter transfer function is obtained from  $H(s)$  as:

$$H'(s) = H(s) \big|_{s=0.7265} = \frac{1}{(s/0.7265)+1} = \frac{0.7265}{s+0.7265}$$

$$\begin{aligned} H(z) &= H'(s) \big|_{s=(z-1)/(z+1)} = \frac{0.7265}{\left(\frac{z-1}{z+1}\right) + 0.7265} = \frac{0.7265(z+1)}{(z-1) + 0.7265(z+1)} \\ &= \frac{0.4208(1+z^{-1})}{1-0.1584z^{-1}} \end{aligned}$$

The difference equation may be written by inspection of

$$H(z) \quad y(n) = 0.1584y(n-1) + 0.4208[x(n) + x(n-1)] .$$