

1a) This is a type 1 filter so

$$\Theta(\omega) = -\alpha \omega$$

$$\text{where } \alpha = \frac{M}{2}$$

$$M = N - 1 = 5 - 1 = 4$$

$$\Theta(\omega) = -2\alpha$$

$$\begin{aligned} 1b) \quad H(e^{j\omega}) &= h_0 + h_1 e^{-j\omega} + h_2 e^{-j2\omega} + h_3 e^{-j3\omega} + h_4 e^{-j4\omega} \\ &= e^{-j2\omega} (h_0 e^{j2\omega} + h_1 e^{j\omega} + h_2 + h_3 e^{-j\omega} + h_4 e^{-j2\omega}) \end{aligned}$$

This is a type 1 filter thus  $h[n] = h[M-n]$

$$\begin{aligned} &= e^{-j2\omega} (h_0 e^{j2\omega} + h_1 e^{j\omega} + h_2 + h_1 e^{-j\omega} + h_0 e^{-j2\omega}) \\ &= e^{-j2\omega} (h_0 (e^{j2\omega} + e^{-j2\omega}) + h_1 (e^{j\omega} + e^{-j\omega}) + h_2) \\ &= e^{-j2\omega} (h_0 2 \cos 2\omega + h_1 2 \cos \omega + h_2) \\ &= (2h_0 \cos 2\omega + 2h_1 \cos \omega + h_2) e^{-j2\omega} \end{aligned}$$

Thus

$$A(e^{j\omega}) = 2h_0 \cos 2\omega + 2h_1 \cos \omega + h_2.$$

## EE461 Assignment Question 3

This m-file script generates the length 31 impulse response using impulse response truncation to approximate an ideal LPF filter with a cutoff frequency of  $f=1/8$  cycles/sample.

### Contents

---

- Define initial values
- Generate and plot the impulse response, h
- Generate and plot the frequency response

### Define initial values

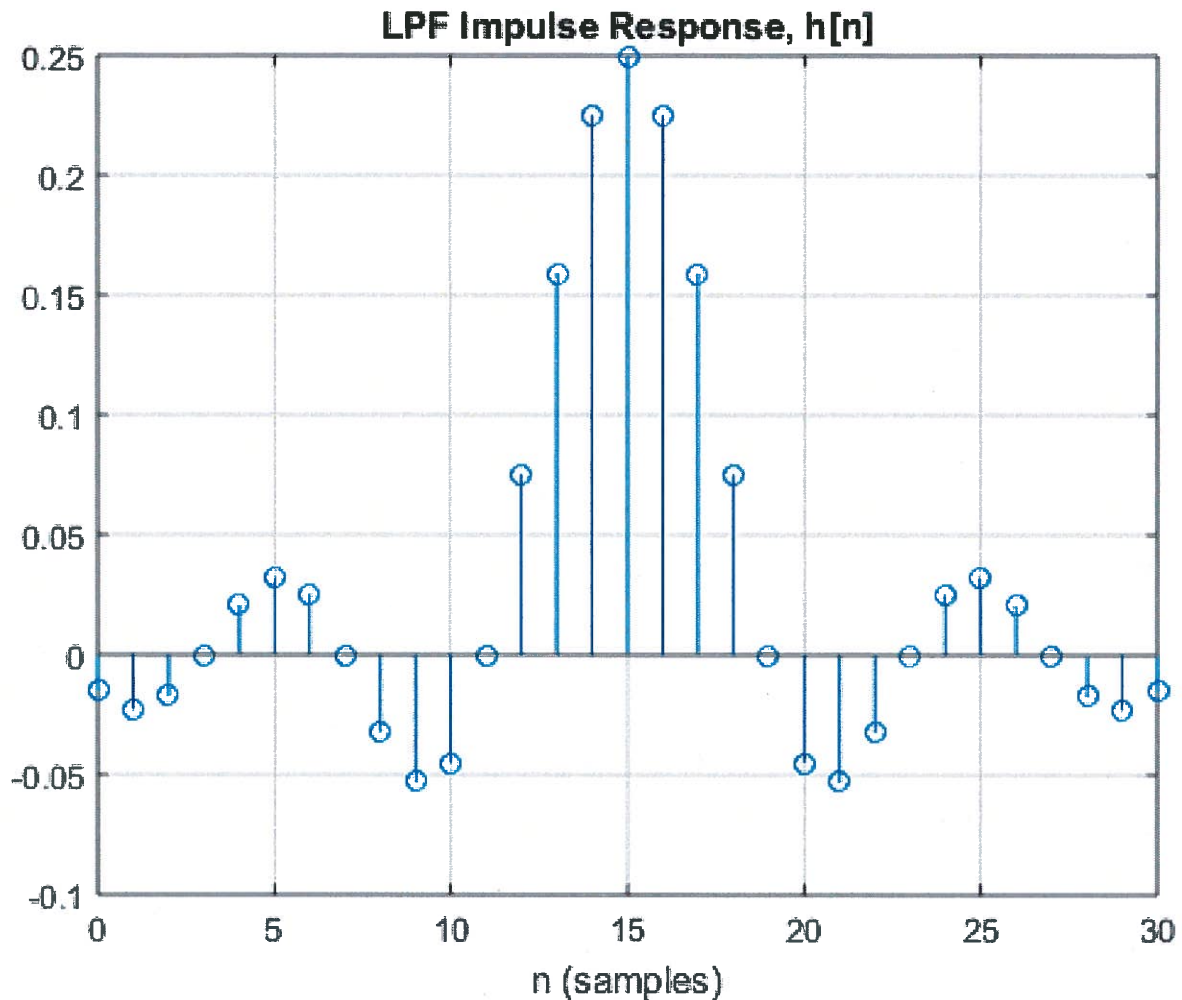
---

```
fc=1/8; % cutoff freq (cycles/sample)
M=30; % order of filter
n=0:M; % sample indices
```

### Generate and plot the impulse response, h

---

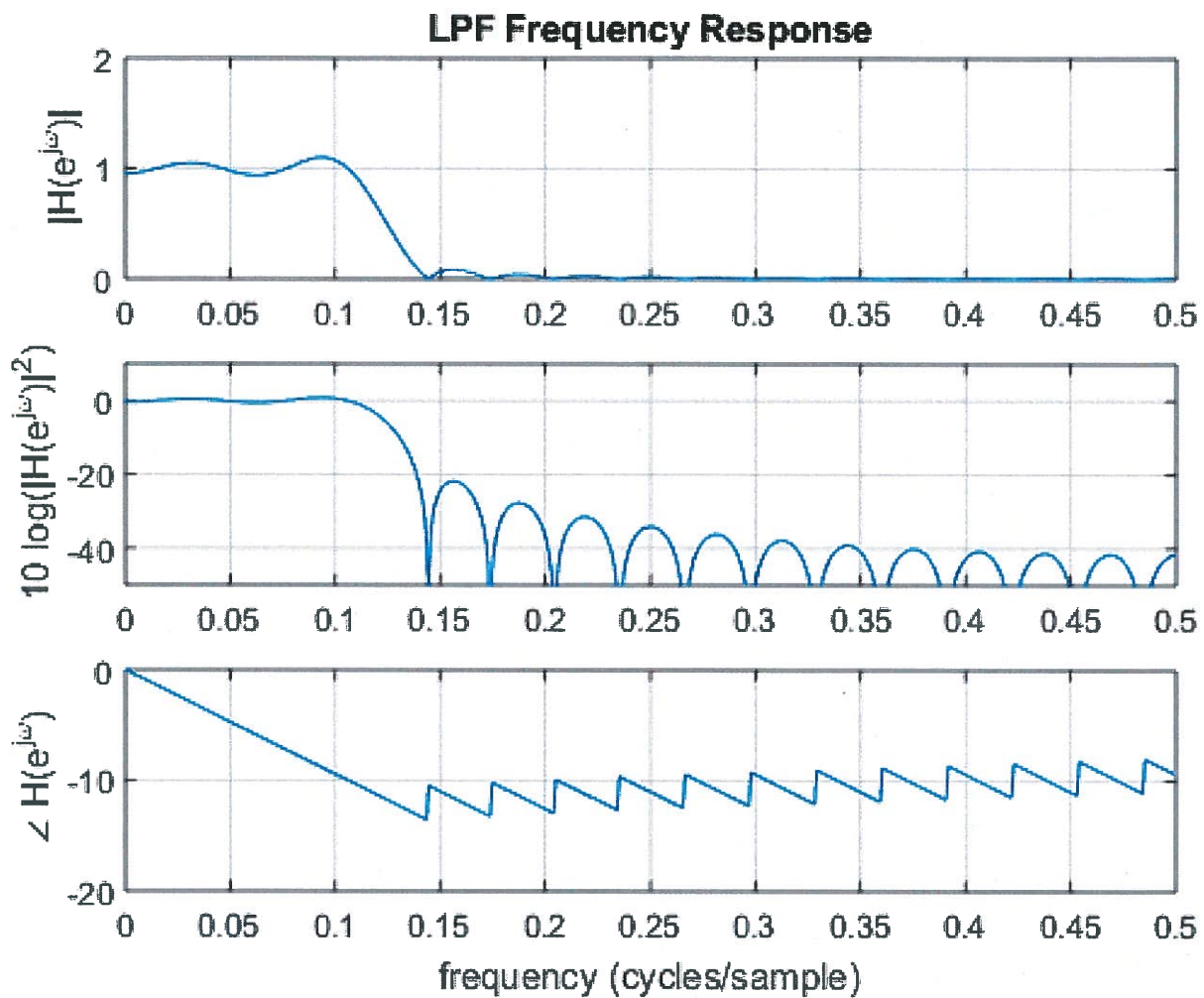
```
h=2*fc*sinc(2*fc*(n-M/2));
figure(1)
clf
stem(n,h)
title('LPF Impulse Response, h[n]')
xlabel('n (samples)')
grid
```



### Generate and plot the frequency response

```
figure(2)
clf
[H,w]=freqz(h,1);
subplot(3,1,1)
plot(w/(2*pi),abs(H))
ylabel('|H(e^{j\omega})|')
title('LPF Frequency Response')
grid
subplot(3,1,2)
plot(w/(2*pi),10*log10(abs(H).^2))
grid
axis([0 .5 -50 10])
ylabel('10 log{|H(e^{j\omega})|^2}')
subplot(3,1,3)
plot(w/(2*pi),unwrap(angle(H)))
ylabel('\angle H(e^{j\omega})')
```

```
xlabel('frequency (cycles/sample)')  
grid
```



Published with MATLAB® R2015a

## EE461 Assignment Question 5

This m-file script generates the length 31 impulse response using impulse response truncation to approximate an ideal LPF filter with a cutoff frequency of  $f=1/8$  cycles/sample.

### Contents

---

- Define initial values
- Generate and plot the impulse response,  $h$
- Generate and plot  $10 \log |H(e^{j\omega})|^2$  in dB
- Generate complete response
- What is the cutoff frequency?

### Define initial values

---

```
fc=1/8; % cutoff freq (cycles/sample)
M=30; % order of filter
n=0:M; % sample indices
```

### Generate and plot the impulse response, $h$

---

```
h=2*fc*sinc(2*fc*(n-M/2));

% Convert the LPR to a HPF using technique 1
h=-h;
h(M/2+1)=1+h(M/2+1);

figure(1)
clf
stem(n,h)
title('HPF Impulse Response, h[n]')
xlabel('n (samples)')
grid
```

### Generate and plot $10 \log |H(e^{j\omega})|^2$ in dB

---

```
figure(2)
clf
[H,w]=freqz(h,1);
plot(w/(2*pi),10*log10(H.*conj(H)))
```

```
ylabel('10 log{(|H(e^{j\omega})|^2)}')
title('HPF Response')
grid
```

---

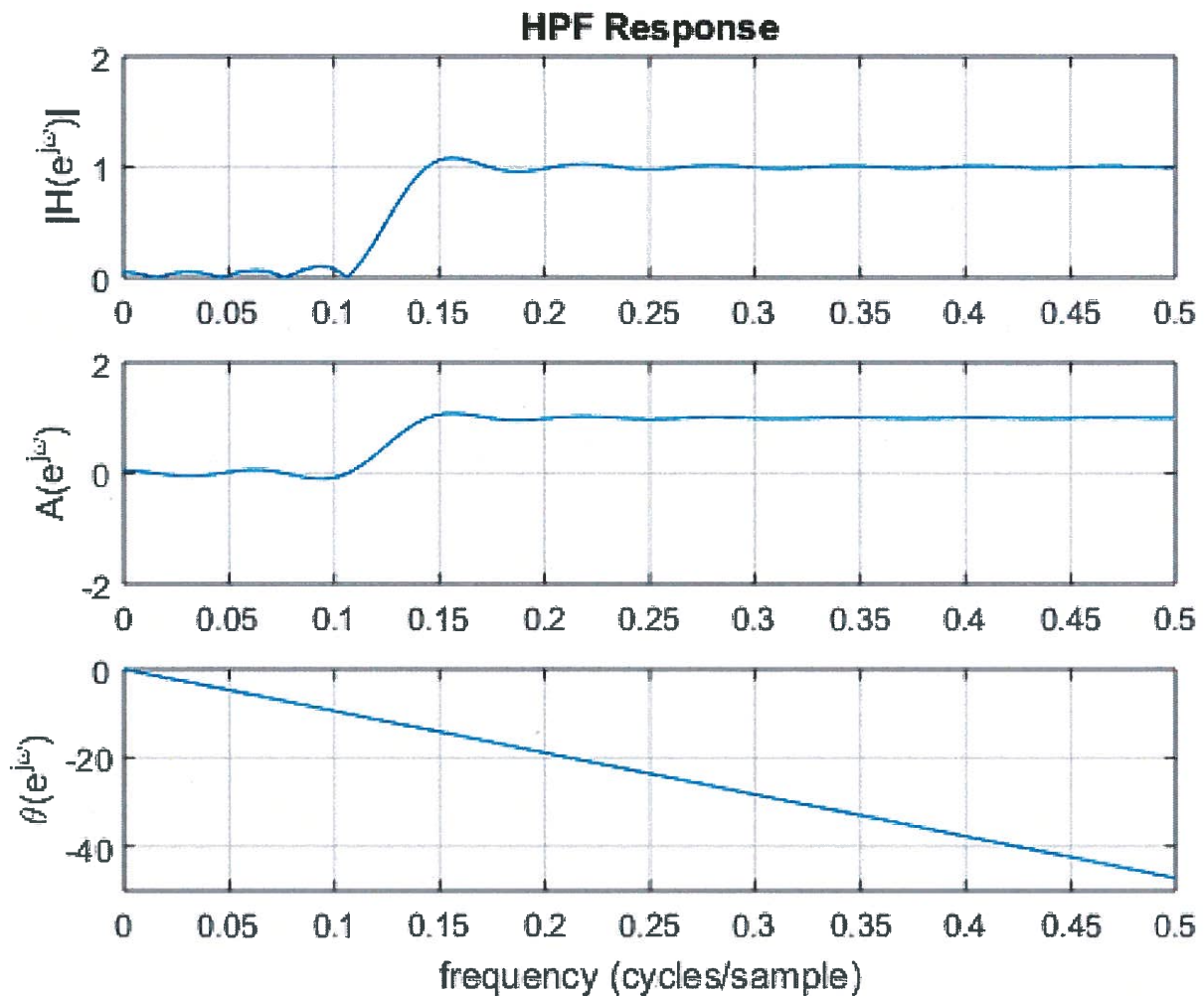
## Generate complete response

---

```
A=H.*exp(j*w*M/2);
figure(3)
subplot(3,1,1)
plot(w/(2*pi),abs(H))
ylabel('|H(e^{j\omega})|')
title('HPF Response')
grid
subplot(3,1,2)
plot(w/(2*pi),real(A))
grid
ylabel('A(e^{j\omega})')
subplot(3,1,3)
plot(w/(2*pi),-(M/2)*w)
ylabel('\theta(e^{j\omega})')
xlabel('frequency (cycles/sample)')
grid
```

---





### What is the cutoff frequency?

The cutoff frequency is the same as the LPF,  $f_c = 1/8$ .

Published with MATLAB® R2015a

$$7a) \quad Ad(e^{j\omega}) = \cos\left(\frac{\omega}{2}\right) (1 + \cos \omega)$$

Using Euler's

$$= \left( \frac{e^{j\frac{\omega}{2}} + e^{-j\frac{\omega}{2}}}{2} \right) \left( 1 + \frac{e^{j\omega} + e^{-j\omega}}{2} \right)$$

$$= \frac{1}{2} e^{j\omega/2} + \frac{1}{2} e^{-j\omega/2} + \frac{1}{4} e^{j3\omega/2} + \frac{1}{4} e^{j\omega/2} + \frac{1}{4} e^{-j\omega/2} + \frac{1}{4} e^{-j3\omega/2}$$

$$= \left( \frac{1}{2} e^{-j\omega} + \frac{1}{2} e^{-j2\omega} + \frac{1}{4} + \frac{1}{4} e^{-j\omega} + \frac{1}{4} e^{-j2\omega} + \frac{1}{4} e^{-j3\omega} \right) e^{j3\omega/2}$$

$$Ad(e^{j\omega}) = \left( \frac{1}{4} + \frac{3}{4} e^{-j\omega} + \frac{3}{4} e^{-j2\omega} + \frac{1}{4} e^{-j3\omega} \right) e^{j3\omega/2}$$

Given  $H_d(e^{j\omega}) = Ad(e^{j\omega}) e^{-j\omega M/2}$

then  $Ad(e^{j\omega}) = H_d(e^{j\omega}) e^{j\omega M/2}$  where  $M=3$ .

The integral square error is

$$E = \frac{1}{2} \int_{-\pi}^{\pi} (H_d(e^{j\omega}) - H(e^{j\omega}))^2 d\omega.$$

Thus  $E$  will be 0 if  $H(e^{j\omega}) = H_d(e^{j\omega})$ ,

let  $H(e^{j\omega}) = H_d(e^{j\omega}) = \frac{1}{4} + \frac{3}{4} e^{-j\omega} + \frac{3}{4} e^{-j2\omega} + \frac{1}{4} e^{-j3\omega}$

which gives

$$h[n] = \left[ \frac{1}{4}, \frac{3}{4}, \frac{3}{4}, \frac{1}{4} \right]$$

b)  $M = N-1 = 4-1 = 3$

c)  $h[n] = h[3-n]$  and  $M$  is odd  $\rightarrow$  Type 2

d)  $\phi(\omega) = -\alpha\omega = -\frac{M}{2}\omega = -\frac{3}{2}\omega$



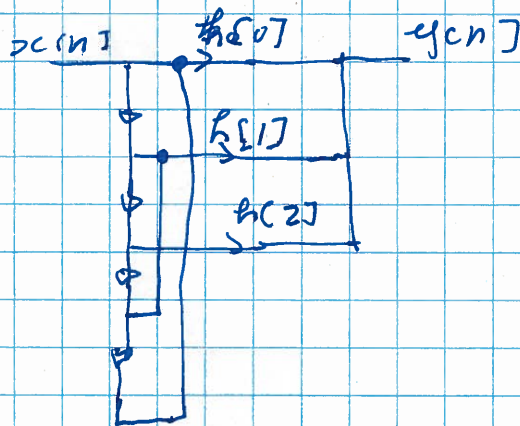
9)  $h(n) = h[4-n] \quad 0 \leq n \leq 4$

$$y(n) = \sum_{k=0}^4 h(k) x(n-k)$$

$$= h[0]x[n] + h[1]x[n-1] + h[2]x[n-2] + h[3]x[n-3] + h[4]x[n-4]$$

Using  $h[0] = h[4]$   
 $h[1] = h[3]$  (from  $h(n) = h[4-n]$ )

$$y(n) = h[0](x[n] + x[n-4]) + h[1](x[n-1] + x[n-3]) + h[2]x[n-2]$$



- a) Multiplexers  
 Direct form - 5  
 Reduced Direct form - 3

- b) Adders  
 Direct form - 4  
 Reduced Direct form - 4

11) a) Given  $H(z) = a + bz^{-1} + cz^{-2}$

then  $h[n] = [a, b, c]$

This is a type 3 filter, since

$$H(e^{j\pi}) = H(e^{j0}) = 0$$

A type 3 filter has an antisymmetric

$h[n]$ , which force  $b$  to be 0.

Using  $H(e^{j0}) = a + b\bar{e}^{j0} + c\bar{e}^{-j0} = 0$

$$= a + b + c = 0$$

with  $b=0$

$$a + c = 0 \quad (1)$$

Using the impulse response has unit energy

$$a^2 + b^2 + c^2 = 1$$

with  $b=0$ ,  $a^2 + c^2 = 1 \quad (2)$

Substituting (1) into (2)

$$a^2 + ((-a)^2) = 1$$

$$a^2 = \frac{1}{2}$$

$$a = \frac{1}{\sqrt{2}} \quad (a > 0 \text{ as required})$$

and  $c = -\frac{1}{\sqrt{2}}$

$$h[n] = \left[ \frac{1}{\sqrt{2}}, 0, -\frac{1}{\sqrt{2}} \right]$$



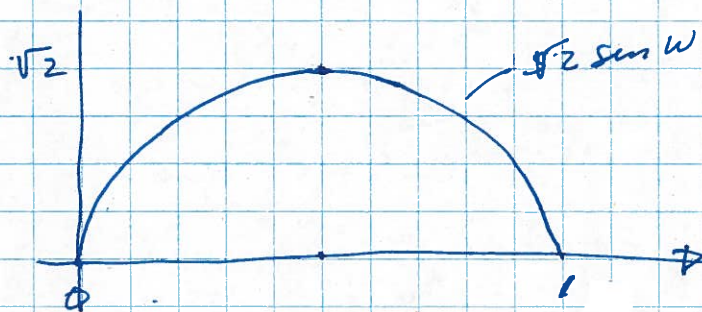
$$b) \quad 1/(e^{j\omega}) = \frac{1}{\sqrt{2}} \Rightarrow \frac{1}{\sqrt{2}} e^{-j2\omega}$$

$$= \left( \frac{1}{\sqrt{2}} e^{j\omega} - \frac{1}{\sqrt{2}} e^{-j\omega} \right) e^{-j\omega}$$

$$= \frac{2j}{\sqrt{2}} \sin(\omega) e^{-j\omega}$$

$$= \sqrt{2} \sin \omega e^{j(\frac{\pi}{2} - \omega)}$$

$$A(e^{j\omega}) = \sqrt{2} \sin \omega$$



$$c) \quad \Theta(\omega) = \frac{\pi}{2} - \omega$$

---

# Solution to Question 13

## Table of Contents

Define initial values .....	1
Generate and plot the impulse response, h .....	1
Generate and plot the amplitude response .....	2

Design a two band filter, using the impulse response truncation technique, with  $M=80$ ,  $f_{11}=0.1$ ,  $f_{21}=0.3$ ,  $f_{12}=0.35$ ,  $f_{22}=0.4$ ,  $C_1=1$ ,  $C_2=0.5$ . Plot the impulse response on one figure and the amplitude response on a second figure for  $f$  from  $-0.5$  to  $0.5$  cycles/sample. Use the MATLAB function `fft` to generate the amplitude response.

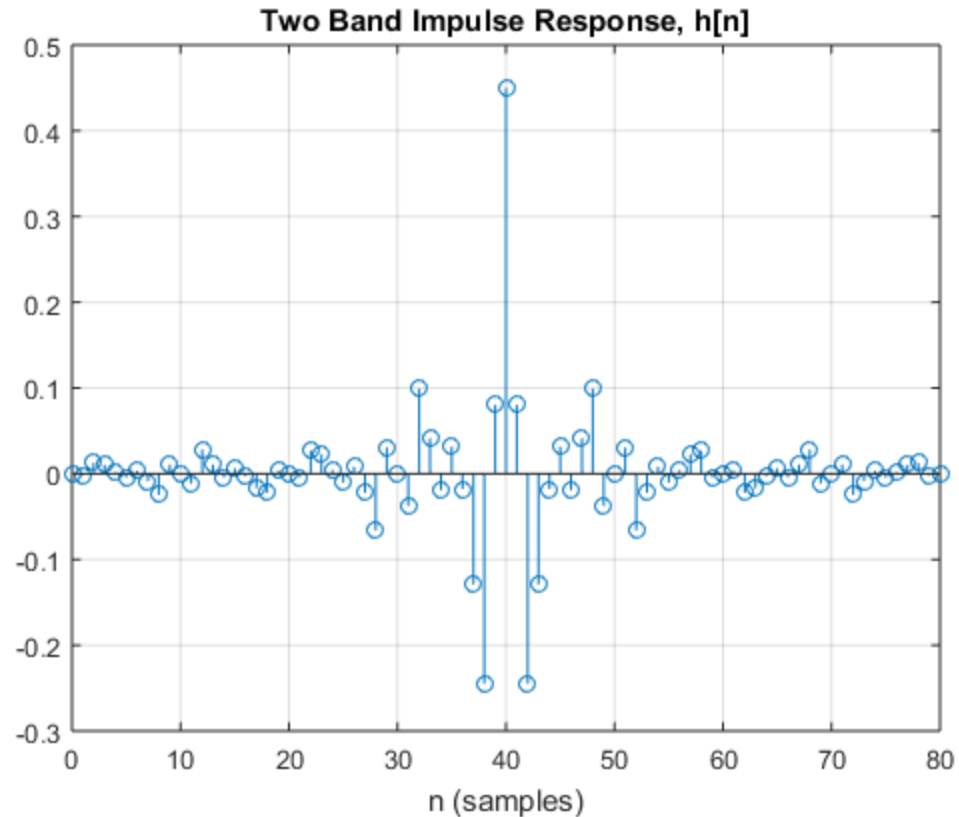
## Define initial values

```
f11=.1; % (cycles/sample)
f21=.3;
f12=.35;
f22=.4;
C1=1;
C2=.5;
M=80; % order of filter
n=0:M; % sample indices
```

## Generate and plot the impulse response, h

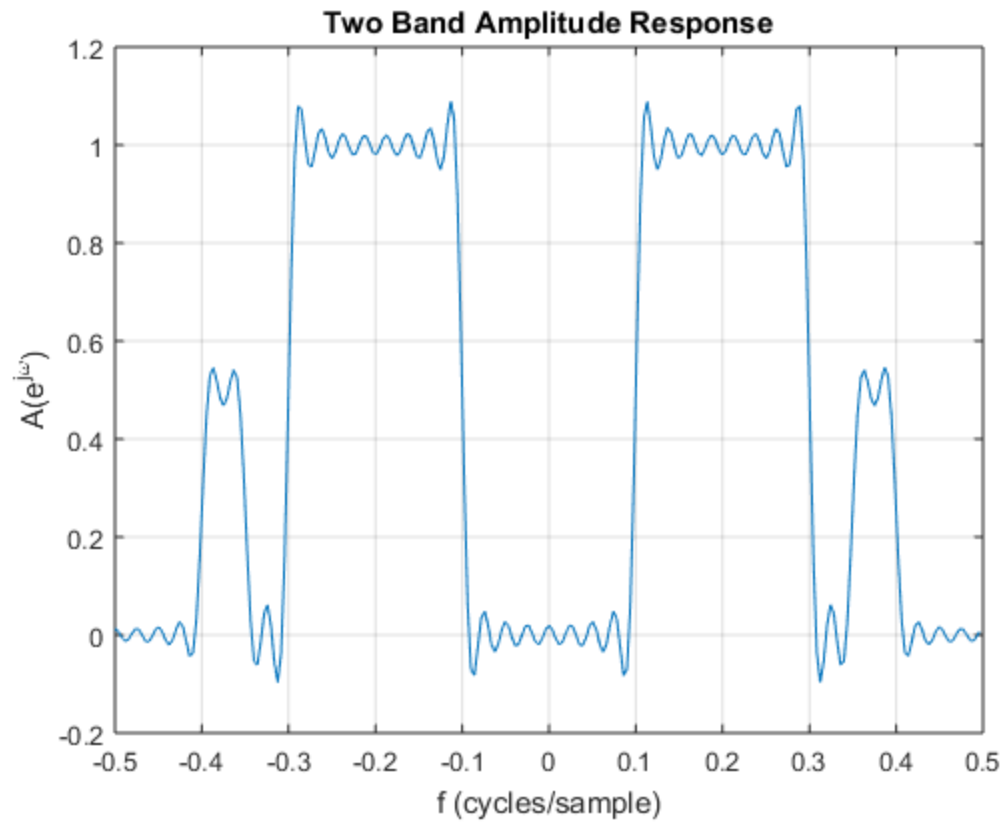
```
h1=2*f21*sinc(2*f21*(n-M/2))-2*f11*sinc(2*f11*(n-M/2));
h2=2*f22*sinc(2*f22*(n-M/2))-2*f12*sinc(2*f12*(n-M/2));
h=C1*h1+C2*h2;
figure(1)
clf
stem(n,h)
title('Two Band Impulse Response, h[n]')
xlabel('n (samples)')
grid
```





## Generate and plot the amplitude response

```
figure(2)
clf
L=256;
hz=[h,zeros(1,L-length(h))];
H=fft(hz);
k=0:L-1;
% M is even - thus this a type 1 filter with a linear phase response
% of
%  $-(M/2)*2*\pi*f$ , where  $f=k/L$ 
W=exp(-j*(-(M/2)*2*pi)*k/L);
A=H.*W;
A=real(A);
plot(k/L-0.5,fftshift(A))
ylabel('A(e^{j\omega})')
title('Two Band Amplitude Response')
xlabel('f (cycles/sample)')
grid
```



*Published with MATLAB® R2015a*