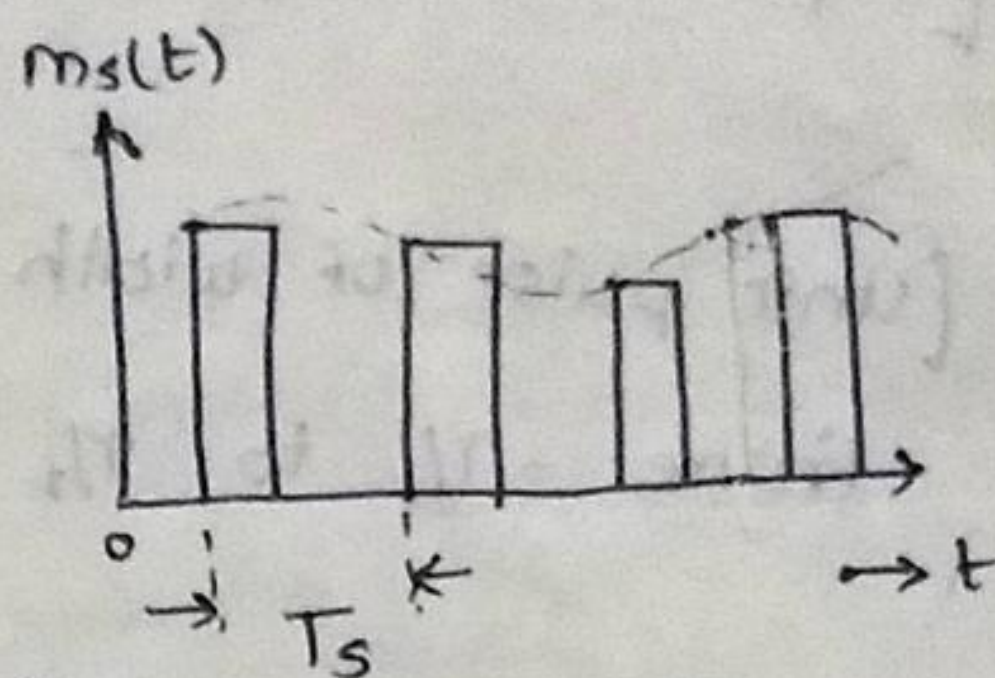


① By using instantaneous sampling, more no. of s/gs can be multiplexed since  $\gamma \rightarrow 0$ ,  $N = \frac{T_s}{\gamma}$

## ② Flat-Top Sampling :-

The flat-topped samples constitute the samples in the form of pulses whose top portion is flat & within pulse interval at one



instant of time only the pulse amplitude is equal to the base band s/g amplitude (i.e., it consists of pulses having constant amplitude value established by the sample value of the base band s/g at some point within the pulse interval).

Disadvantage :- In this type of sampling, the base band s/g  $m(t)$  can't be recovered exactly by simply passing the samples through an ideal LPF & at the Rxing end, when these samples are used to reconstruct the s/g, more distortion occurs.

### Advantages :-

- 1) Simplifies the design of electronic cktry used to perform the sampling operation.
- 2) Design of flat topped samples is relatively easier & cost of implementation is less.

## Spectrum for sampling methods :-

$\delta(t)$   
(instantaneous samples)

↓  
impulse of strength  $dt$

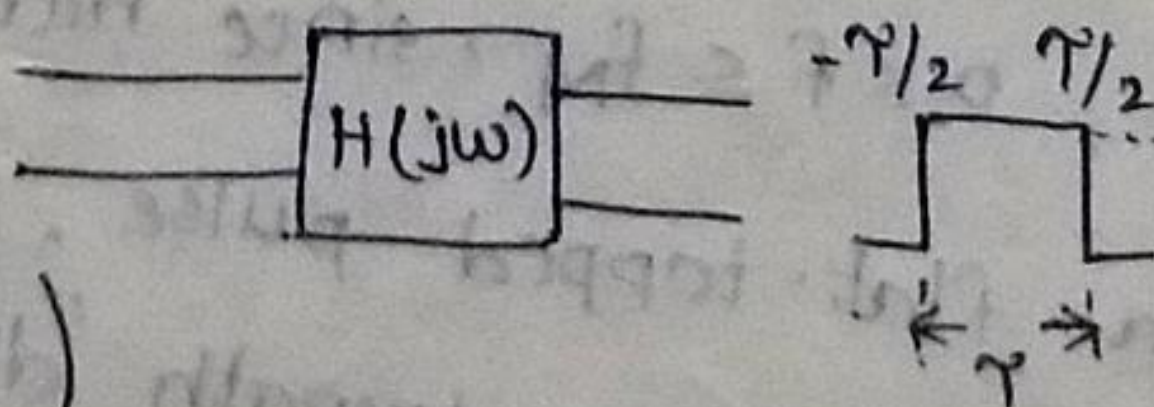


Fig - A  $\eta/w$  with transform  $H(jw)$  which converts a pulse of width  $dt$  into a rectangular pulse of same amplitude, but of duration  $\gamma$ .



H(jw) calculation :  $\left\{ v(f) = F[v(t)] = \int_{-\infty}^{\infty} e^{-j2\pi ft} dt \right\}$

$F[\text{impulse of strength } dt \text{ at } t=0] = 1 \times dt = dt$   
 $\downarrow$   
 $\text{strength}$   
 $F[\text{impulse}] = \int_{-\infty}^{\infty} e^{-j2\pi ft} dt \Big|_{t=0} = \int_{-\infty}^{\infty} e^0 dt = 1$

$F[\text{unit pulse of width } \tau \text{ extending from } -\tau/2 \text{ to } \tau/2]$   
 $= \int_{-\tau/2}^{\tau/2} e^{-j2\pi ft} dt$

$= \left[ \frac{e^{-j2\pi ft}}{-j2\pi f} \right]_{-\tau/2}^{\tau/2}$

$= \frac{-2j \sin 2\pi f \tau/2}{-j2\pi f}$

$= \frac{\sin \omega \tau/2 \times \omega \tau/2}{\pi f \times \omega \tau/2}$

$= \frac{\sin \omega \tau/2}{\omega \tau/2}$

$\therefore \frac{1}{H(jw)} = \frac{F[\text{impulse of strength } dt \text{ at } t=0]}{F[\text{unit pulse of width } \tau \text{ extending from } -\tau/2 \text{ to } \tau/2]}$

$H(jw) = \frac{F[\text{unit pulse of width } \tau \text{ extending from } -\tau/2 \text{ to } \tau/2]}{F[\text{impulse of strength } dt \text{ at } t=0]}$

$\Rightarrow \boxed{H(jw) = \frac{\tau}{dt} \frac{\sin \omega \tau/2}{\omega \tau/2}}$

$\left\{ \begin{aligned} H(jw) &= \frac{F[\text{inp}]}{F[\text{ip}]} \end{aligned} \right\}$

Consider a base band s/g whose Fourier transform is  $F[m(t)] = M(jw)$ ,  $0 \leq f \leq f_m$ , since  $m(t)$  is band limited to  $f_m$ . To generate a flat-topped pulse, pass the instantaneously sampled s/g of strength 'dt' through a n/w having transfer fn  $H(jw) = \frac{\tau}{dt} \frac{\sin \omega \tau/2}{\omega \tau/2}$ . This n/w stretches



impulse of strength 'dt' to form a unit pulse of width  $\tau$  sec extending from  $-\tau/2$  to  $\tau/2$

$$\therefore F[\text{instantaneous samples}] = F\left[\frac{dt}{T_s}\right] F[m(t)]$$

$$= \frac{dt}{T_s} M(j\omega)$$

$$\therefore F[\text{flat topped sampled } m(t)] = H(j\omega) F[\text{instantaneous samples}]$$

$$= \frac{\tau}{T_s} M(j\omega) \frac{\sin \omega\tau/2}{\omega\tau/2}$$

The term  $\frac{\sin \omega\tau/2}{\omega\tau/2}$  is termed as aperture factor which is

responsible for more distortion in flat topped samples.

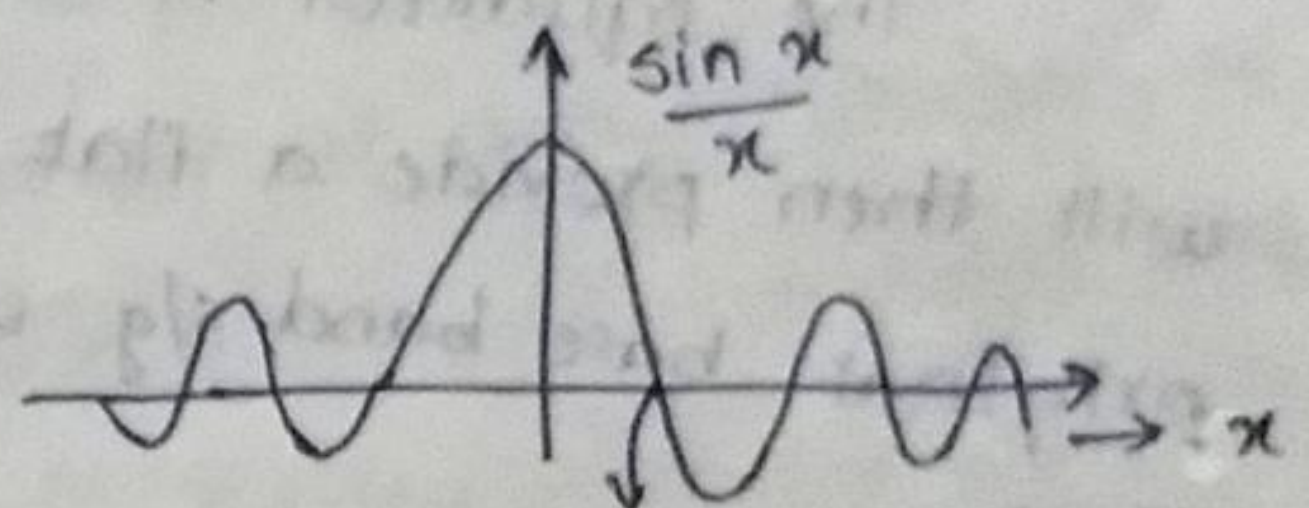
Null freq : The freq at which spectral amplitude becomes zero is called null freq.

Determination of null freq  $\therefore (f_0)$

$$\text{At } x = \omega\tau/2 = \pi$$

$$\Rightarrow 2\pi f_0 \tau/2 = \pi$$

$$\Rightarrow f_0 = 1/\tau$$



null points exists at multiples of  $\pi$  i.e.,  $\pi, 2\pi, 3\pi \dots$

fig a) An idealized spectrum of base band s/g

magnitude, instantaneously sampled  $\frac{dt}{T_s} M_0$   $m(t)$

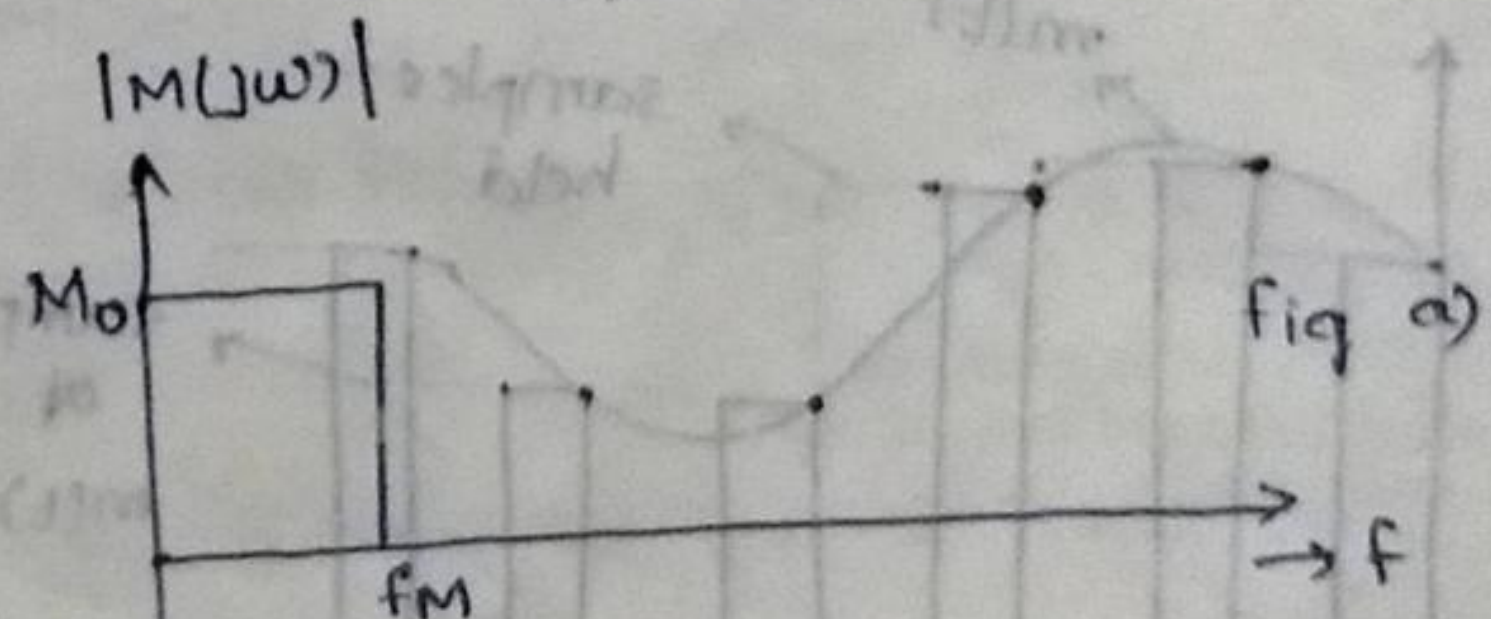


fig b) Spectrum of s/g with instantaneous sampling

$$\left| \frac{\sin \omega\tau/2}{\omega\tau/2} \right|$$

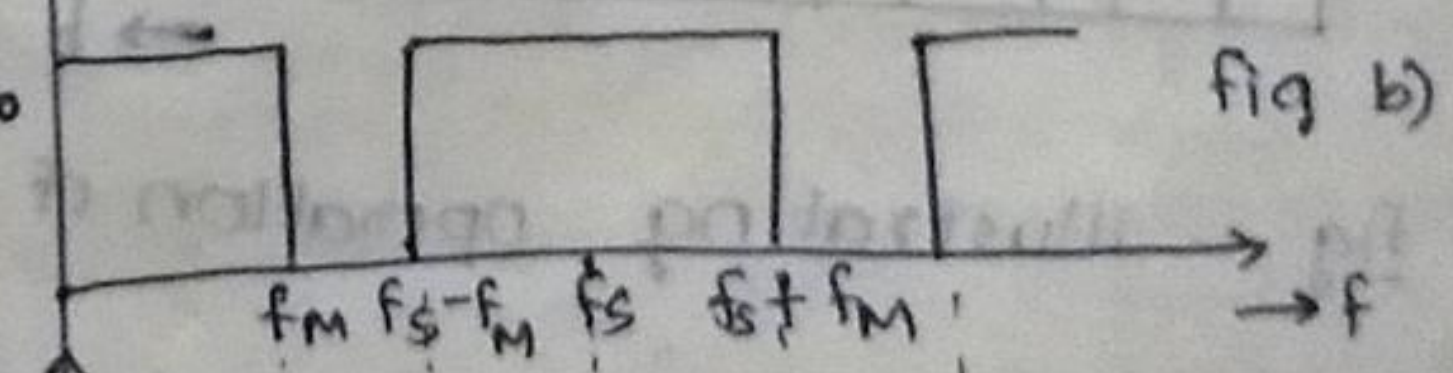


fig c) The form of distortion factor (aperture effect) introduced by flat-topped samples

magnitude, flat-topped sampled  $m(t)$

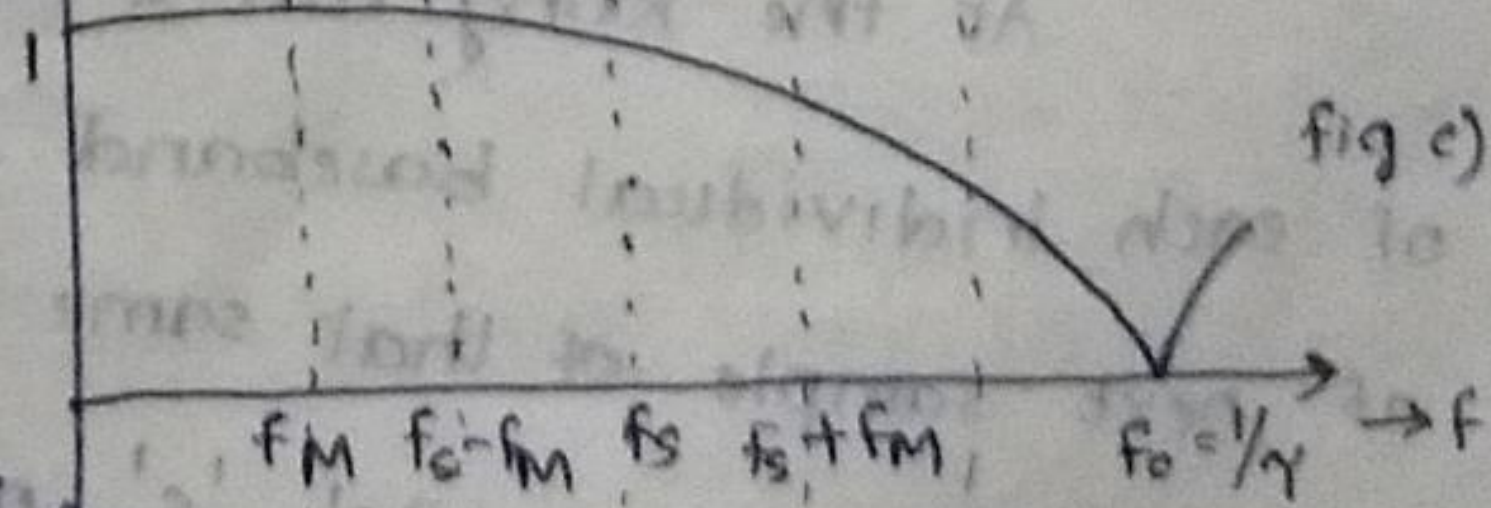
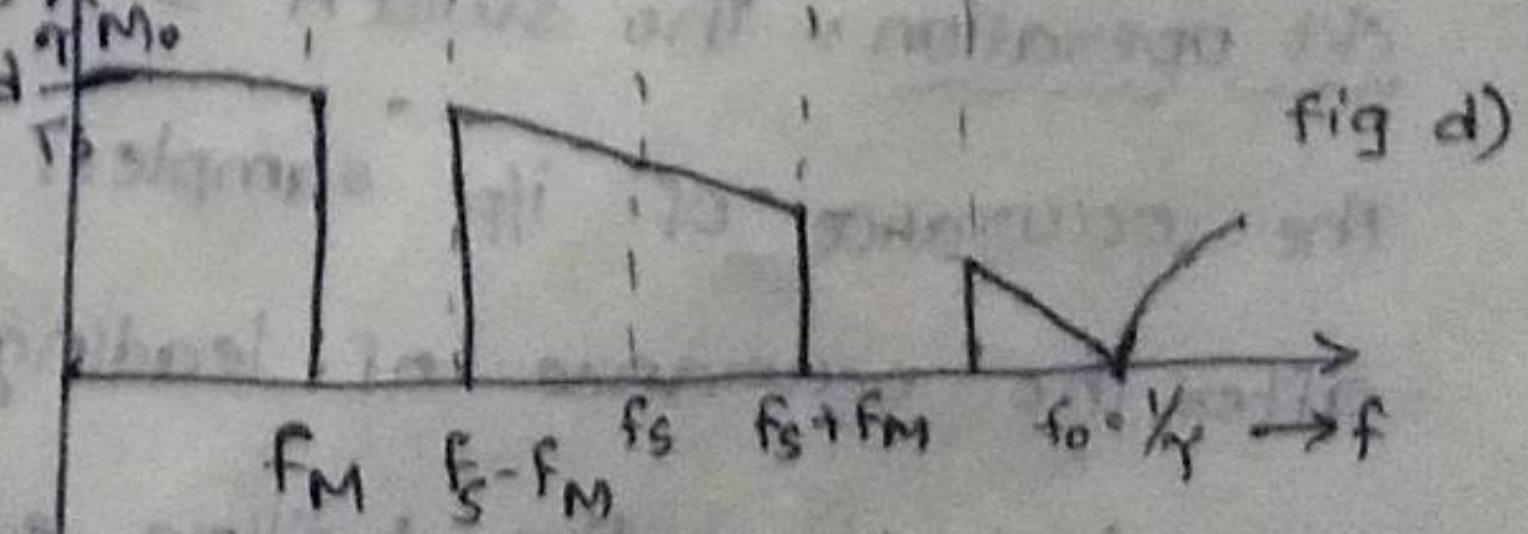


fig d) Spectrum of s/g with flat topped sampling.





The presence of aperture factor is the reason for getting distortion. By choosing null freq  $f_0$  as  $f_0 \gg f_m$ , the distortion can be reduced. But since  $f_0 = \frac{1}{T_s}$ , if  $f_0$  is large,  $\eta$  reduces which reduces the strength of Txed s/g.  $\left\{ S_0(t) = \frac{\eta}{T_s} m(t) \right\}$

### Equalizer :-

Alternately pass the s/g through an equalizer whose transfer fn is freq dependent in the form of  $\frac{x}{\sin x}$  and is placed in cascade with o/p LPF.

The equalizer in combination with the aperture effect will then provide a flat overall transfer characteristic b/w original base band s/g & the o/p at the Rxing end of the system.

→ Problems (2) [2, 3]

→ Signal Recovery through holding :-

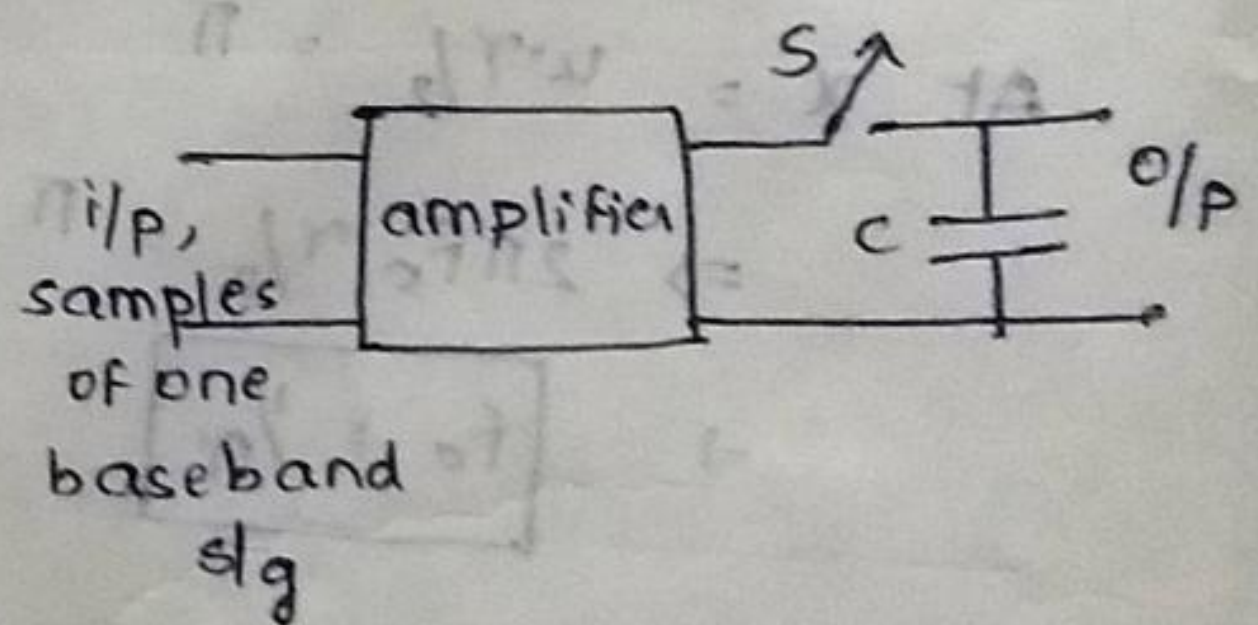
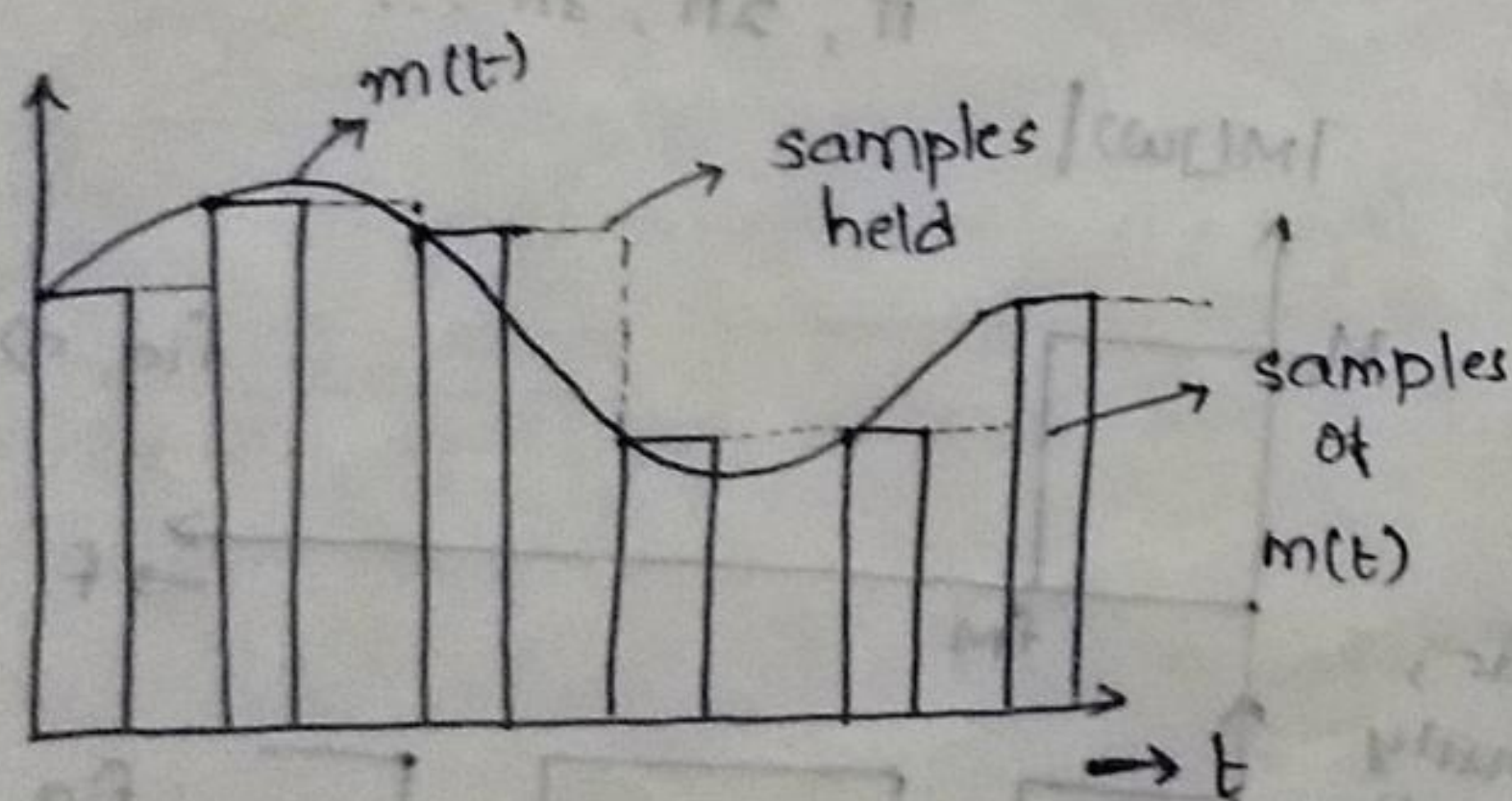


fig :- Illustrating operation of holding

At the Rxing end & after demultiplexing, sample value of each individual baseband s/g is held until the occurrence of next sample of that same base band s/g.

ckt operation :- The switch 'S' operates in synchronism with the occurrence of i/p samples. This switch closes somewhat after the occurrence of leading edge of a sample pulse & opens somewhat before the trailing edge. The amplifier has a low o/p impedance.



$\frac{dt}{T_s} M_0$  magnitude, instantaneously sampled s/g

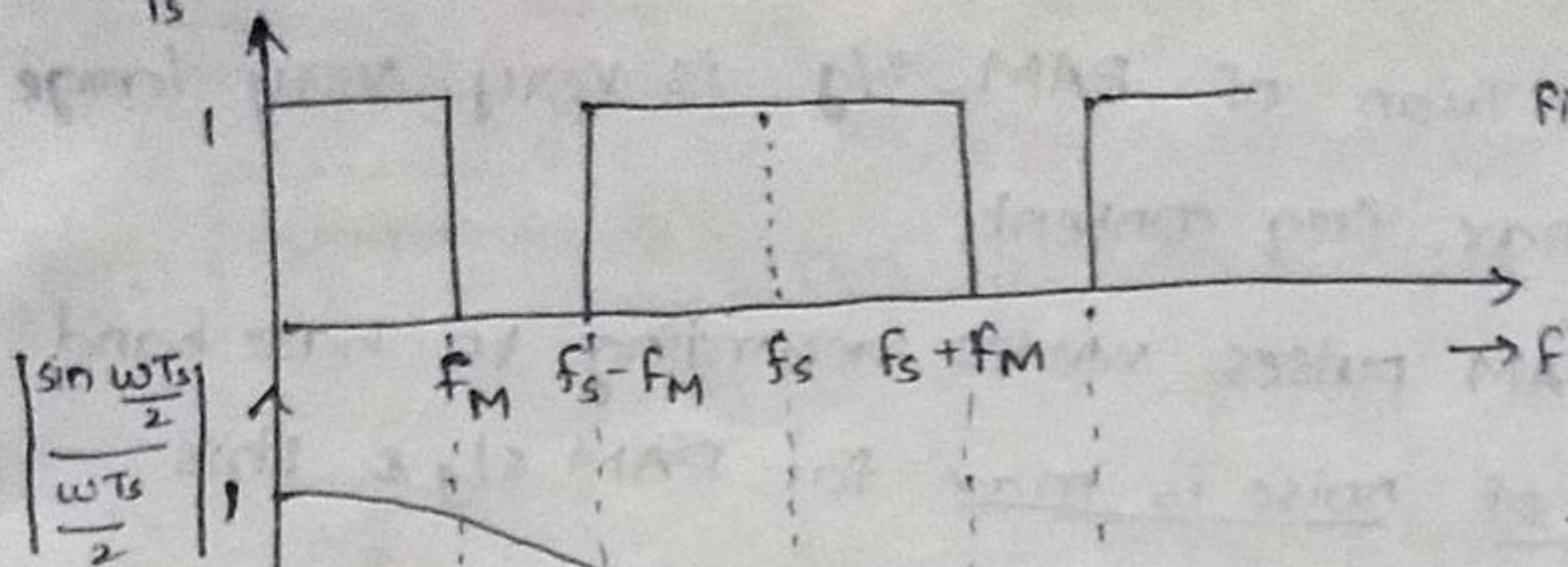


Fig a) Spectrum of instantaneously sampled s/g m(t)

magnitude, held sample  $M_0$

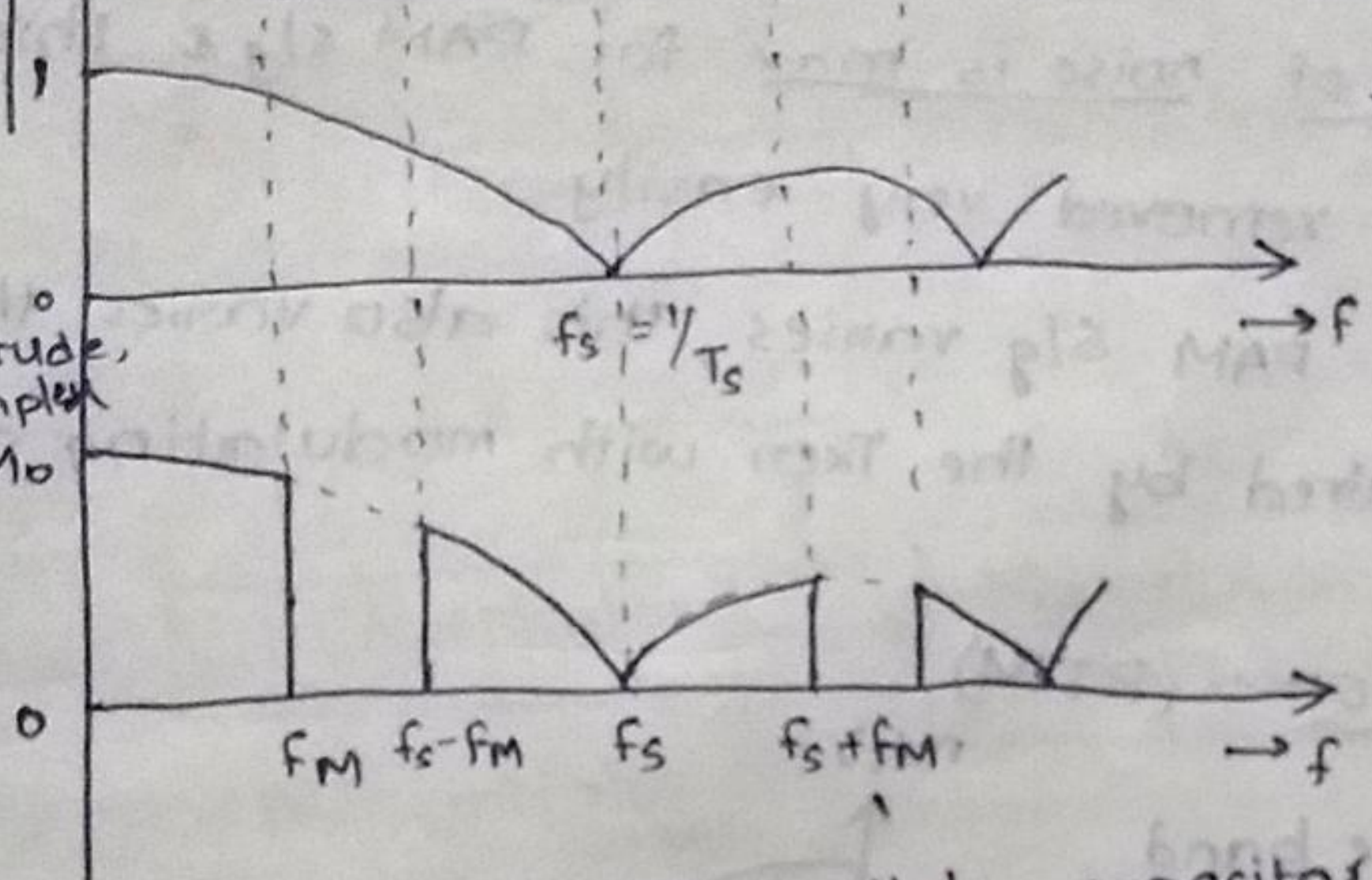


fig b) Magnitude of aperture effect factor

fig c) Spectrum of sampled & held s/g.

At the closing of the switch, capacitor 'c' charges abruptly to a voltage proportional to the sample value & the capacitor holds this voltage until the operation is repeated for the next sample.

let the base band s/g  $m(t)$  has spectral density

$$F[m(t)] = M(j\omega)$$

$$F[m(t), \text{sampled \& held}] = \frac{\sin wTs/2}{wTs/2} M(j\omega), \quad 0 \leq f \leq f_M$$

$$= \frac{T_s}{\tau} F[\text{flat topped samples at } \tau = T_s]$$

The sampled & held s/g has a magnitude larger by a factor  $\frac{T_s}{\tau}$  than the s/g of sample duration  $\tau$ . Thus this method is an alternate method of recovery of base band s/g that raises the level of o/p s/g.

### Advantages of PAM:

- 1) Easy to implement
  - 2) Greater peak power compared to AM & FM
  - 3) In PAM, pulses are of uniform duration & equally spaced.
- $\therefore$  Rxer synchronization is not a problem.



### → Disadvantages of PAM :-

- 1) The BW needed for Txn of PAM s/g is very very large compared to the max. freq content.
- 2) The amplitude of PAM pulses varies according to base band s/g.  $\therefore$  Interference of noise is max for PAM s/g & this noise cannot be removed very easily.
- 3) Since the amp. of PAM s/g varies, this also varies the peak power required by the Txn with modulating s/g.

### → Pulse Time Modulation :- (PTM)

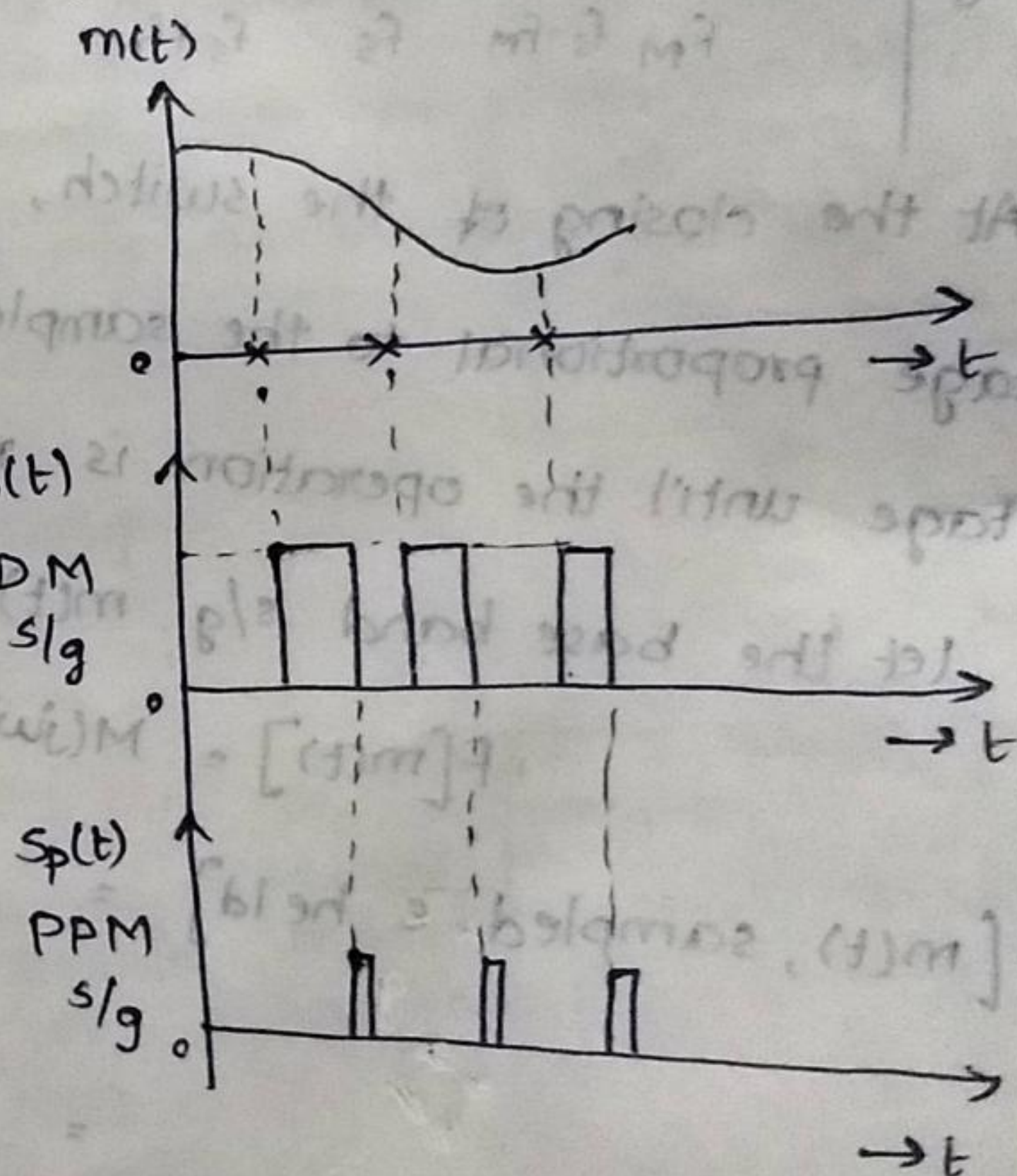
In PAM, the base band s/g modulates the amplitude of sampling s/g, but in PTM

one of the time characteristics (such as width, position) of the sampling s/g is varied according to the s/g

amplitude at that instant.

PTM is of 2 types :-

- 1) PWM/PDM/PLM
- 2) PPM



PWM : Pulse width modulation

PDM : Pulse Duration Modulation

PLM : Pulse Length Modulation

In PDM system, the duration of sampled pulse is proportional to sample value of base band s/g amplitude, whereas ppm system generates pulses of constant amplitude & width whose position is varied according to the base band amplitude at each sampling point.