

a relatively narrow frequency band in the neighbourhood of f_0 .

→ Problems :

Prove that ① $E[n_c^2(t)] = E[n_s^2(t)]$

② $E[n_c(t) n_s(t)] = 0$.

Ans) ① $E[n_c^2(t)] = E[n_s^2(t)]$

$n_c(t)$ & $n_s(t)$ are stationary random processes &

we know that $E[a_k^2] = E[b_k^2] \rightarrow$ equal variance

$E[a_k b_k] = 0 \rightarrow$ uncorrelated to each other.

$$n_c(t) = \lim_{\Delta f \rightarrow 0} \sum_{k=1}^{\infty} \left[a_k \cos 2\pi(k-K)\Delta f t + b_k \sin 2\pi(k-K)\Delta f t \right]$$

$$n_s(t) = \lim_{\Delta f \rightarrow 0} \sum_{k=1}^{\infty} \left[a_k \sin 2\pi(k-K)\Delta f t - b_k \cos 2\pi(k-K)\Delta f t \right]$$

$$E[n_c^2(t)] = \lim_{\Delta f \rightarrow 0} \sum_{k=1}^{\infty} E \left[a_k^2 \cos^2 2\pi(k-K)\Delta f t + b_k^2 \sin^2 2\pi(k-K)\Delta f t + 2a_k b_k \cos 2\pi(k-K)\Delta f t \sin 2\pi(k-K)\Delta f t \right]$$

$$= \lim_{\Delta f \rightarrow 0} \sum_{k=1}^{\infty} E \left[a_k^2 (\cos^2 2\pi(k-K)\Delta f t + \sin^2 2\pi(k-K)\Delta f t) + 0 \right]$$

$$\because E[a_k^2] = E[b_k^2] \text{ \& } E[a_k b_k] = 0$$

$$= \lim_{\Delta f \rightarrow 0} \sum_{k=1}^{\infty} E[a_k^2] \quad \left\{ \because \cos^2 \theta + \sin^2 \theta = 1 \right\}$$

$$E[n_s^2(t)] = \lim_{\Delta f \rightarrow 0} \sum_{k=1}^{\infty} E \left[a_k^2 \sin^2 2\pi(k-K)\Delta f t + b_k^2 \cos^2 2\pi(k-K)\Delta f t - 2a_k b_k \cos 2\pi(k-K)\Delta f t \sin 2\pi(k-K)\Delta f t \right]$$

$$= \lim_{\Delta f \rightarrow 0} \sum_{k=1}^{\infty} E \left[a_k^2 (\sin^2 2\pi(k-K)\Delta f t + \cos^2 2\pi(k-K)\Delta f t) + 0 \right]$$

$$\Rightarrow E[n_c^2(t)] = \lim_{\Delta t \rightarrow 0} \sum_{k=1}^{\infty} E[a_k^2]$$

$$\therefore E[n_c^2(t)] = E[n_c^2(t)]$$

$$(2) E[n_c(t) n_s(t)]$$

$$\Rightarrow \lim_{\Delta t \rightarrow 0} \sum_{k=1}^{\infty} E \left[\begin{aligned} &a_k \cos 2\pi(k-K)\Delta t + b_k \sin 2\pi(k-K)\Delta t \\ &\left[a_k \sin 2\pi(k-K)\Delta t - b_k \cos 2\pi(k-K)\Delta t \right] \end{aligned} \right]$$

$$\Rightarrow \lim_{\Delta t \rightarrow 0} \sum_{k=1}^{\infty} E \left[\begin{aligned} &a_k^2 \cos 2\pi(k-K)\Delta t \sin 2\pi(k-K)\Delta t \\ &+ b_k a_k \sin^2 2\pi(k-K)\Delta t \\ &- a_k b_k \cos^2 2\pi(k-K)\Delta t - b_k^2 \sin^2 2\pi(k-K)\Delta t \end{aligned} \right]$$

We know that for a stationary random process

$$E[a_k b_k] = 0 \rightarrow \text{Uncorrelated to each other}$$

$$E[a_k^2] = E[b_k^2] \rightarrow \text{Equal variance.}$$

\therefore we have

$$E[n_c(t) n_s(t)] = \lim_{\Delta t \rightarrow 0} \sum_{k=1}^{\infty} E \left[\begin{aligned} &a_k^2 \cos 2\pi(k-K)\Delta t \sin 2\pi(k-K)\Delta t \\ &+ 0 - 0 - b_k^2 \sin^2 2\pi(k-K)\Delta t \end{aligned} \right]$$

$$= \lim_{\Delta t \rightarrow 0} \sum_{k=1}^{\infty} E \left[\begin{aligned} &a_k^2 \cos 2\pi(k-K)\Delta t \sin 2\pi(k-K)\Delta t \\ &- a_k^2 \sin 2\pi(k-K)\Delta t \cos 2\pi(k-K)\Delta t \end{aligned} \right]$$

$$= 0 \quad \therefore E[n_c(t) n_s(t)] = 0$$

Problem :-

\rightarrow Find the value of $E[n_c(t) n_s(t)]$ & $E[n_c(t) n_s(t)]$

$$\text{where } \dot{n}_c(t) = \frac{d}{dt} n_c(t) \quad \&$$

$$\dot{n}_s(t) = \frac{d}{dt} n_s(t)$$

Answer :- 0

→ Power Spectral Density of Quadrature Component

(or) [PSD of $n_c(t)$, $n_s(t)$]

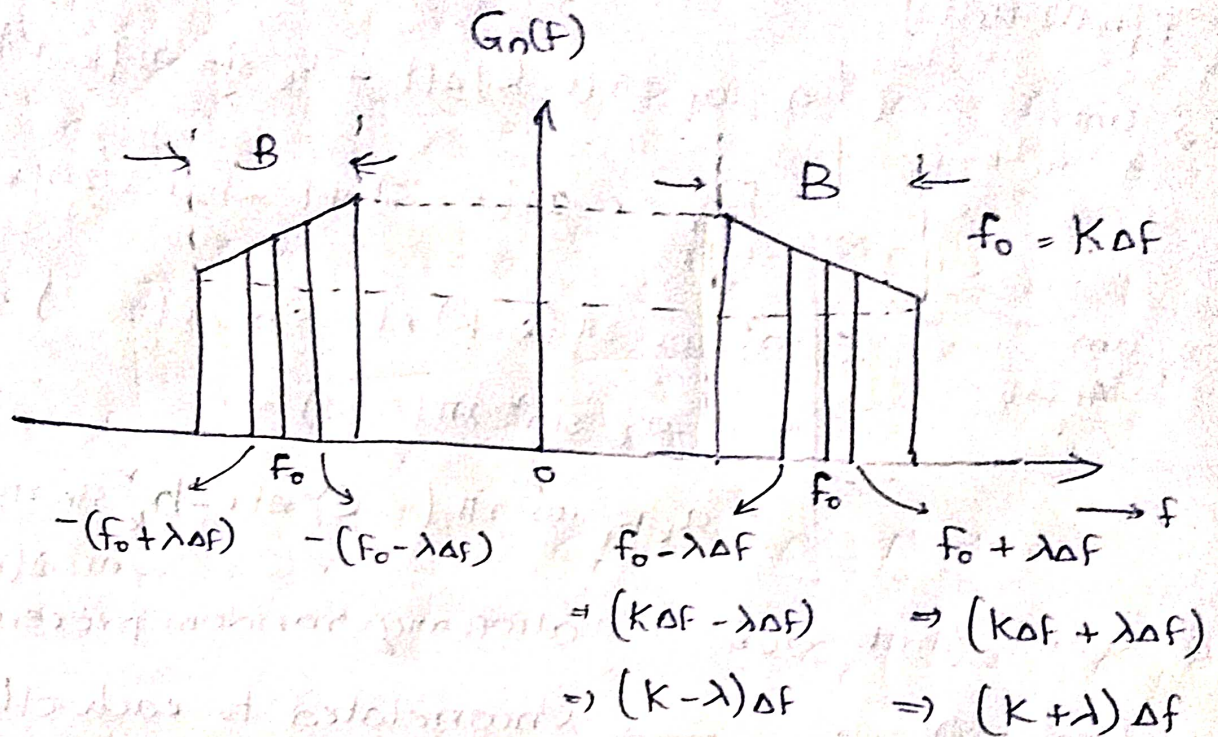


Fig: Power Spectrum of bandlimited noise

Consider non-white noise is having non-uniform noise PSD and f_0 need not be located at its centre.

To determine the PSD of $n_c(t)$,

① select the spectral components of noise $n(t)$ from the value of $k = (K - \lambda)$ to $(K + \lambda)$ where λ is an integer.

Since $K\Delta f = f_0$, the selected components correspond to frequencies $(K - \lambda)\Delta f$ to $(K + \lambda)\Delta f$ i.e., from

$(f_0 - \lambda\Delta f)$ to $(f_0 + \lambda\Delta f)$. We know $n_c(t)$ as

$$n_c(t) = \lim_{\Delta f \rightarrow 0} \sum_{k=-\infty}^{\infty} \left[a_k \cos 2\pi(k - K)\Delta f t + b_k \sin 2\pi(k - K)\Delta f t \right]$$

Find $\Delta n_c(t)$ within this interval i.e.,

$$\Delta n_c(t) = n_c(t) \Big|_{k=k-\lambda} + n_c(t) \Big|_{k=k+\lambda} \rightarrow (1)$$

$$\Rightarrow \Delta n_c(t) = a_{k-\lambda} \cos 2\pi (k-\lambda - k) \Delta f t + b_{k-\lambda} \sin 2\pi (-\lambda) \Delta f t \\ + a_{k+\lambda} \cos 2\pi (k+\lambda - k) \Delta f t + b_{k+\lambda} \sin 2\pi (\lambda) \Delta f t$$

$$\Rightarrow \Delta n_c(t) = a_{k-\lambda} \cos 2\pi \lambda \Delta f t - b_{k-\lambda} \sin 2\pi \lambda \Delta f t + \\ a_{k+\lambda} \cos 2\pi \lambda \Delta f t + b_{k+\lambda} \sin 2\pi \lambda \Delta f t$$

These 4 terms are of same freq $\lambda \Delta f$ and are uncorrelated random processes.

(2) These random processes are stationary, consider at time $t = t_1$; $\cos 2\pi \lambda \Delta f t_1 = 1$ & $\sin 2\pi \lambda \Delta f t_1 = 0$ where $\lambda \Delta f t_1$ is then integer

$$\Delta n_c(t_1) = a_{k-\lambda}(1) - 0 + a_{k+\lambda}(1) + 0$$

$$\Rightarrow \Delta n_c(t_1) = a_{k-\lambda} + a_{k+\lambda} \rightarrow (2)$$

(3) Find the power P_λ associated with $\Delta n_c(t_1)$ and interrelate P_λ with PSD $G_{nc}(f)$ & $G_n(f)$

$$\therefore P_\lambda = E[\Delta n_c(t_1)^2]$$

$$= E[(a_{k-\lambda} + a_{k+\lambda})^2]$$

$$= E[a_{k-\lambda}^2] + E[a_{k+\lambda}^2] + 2 \underbrace{E[a_{k-\lambda} a_{k+\lambda}]}_{\downarrow 0}$$

$$\Rightarrow P_\lambda = E[a_{k-\lambda}^2] + E[a_{k+\lambda}^2] \rightarrow (3) \quad \because \text{uncorrelated to each other}$$

We know that $P_k = 2 G_n(k\Delta f) \Delta f$

$$\therefore P_\lambda = 2 G_n(\lambda \Delta f) \Delta f$$

$$E[a_{k-\lambda}^2] = P_{k-\lambda} = 2 G_n((k-\lambda) \Delta f) \Delta f$$

$$E[a_{k+\lambda}^2] = P_{k+\lambda} = 2 G_n((k+\lambda) \Delta f) \Delta f$$

$$\therefore \text{eq (3)} \Rightarrow P_\lambda = E[a_{k-\lambda}^2] + E[a_{k+\lambda}^2]$$

$$\Rightarrow P_\lambda = \overline{a_{k-\lambda}^2} + \overline{a_{k+\lambda}^2}$$

$$\Rightarrow 2 G_n(\lambda \Delta f) \Delta f = 2 G_n((k-\lambda) \Delta f) \Delta f + 2 G_n((k+\lambda) \Delta f) \Delta f$$

$$\Rightarrow \boxed{G_n(\lambda \Delta f) = G_n((k-\lambda) \Delta f) + G_n((k+\lambda) \Delta f)} \quad \text{--- (4)}$$

We know that $f_0 = k\Delta f$, replace $\lambda \Delta f$ by a continuous freq variable f , \therefore eq (4) becomes

$$\boxed{G_n(f) = G_n(f_0 - f) + G_n(f + f_0)}$$

Similarly the same procedure can be adapted to find the PSD of $n_s(t)$ & a similar expression can be obtained for $n_s(t)$

$$\boxed{G_{ns}(f) = G_{nc}(f) = G_n(f_0 - f) + G_n(f_0 + f)}$$

$$\left\{ \begin{array}{l} \text{for } \underline{n_d(t)} : P_\lambda = \overline{b_{k-\lambda}^2} + \overline{b_{k+\lambda}^2} \\ \quad \quad \quad = \overline{a_{k-\lambda}^2} + \overline{a_{k+\lambda}^2} \end{array} \right. \quad \begin{array}{l} \text{we know that } \overline{a_k^2} = \overline{b_k^2} \\ \text{\& remaining is same} \end{array}$$

To get the PSD of either $n_c(t)$ or $n_s(t)$ graphically displace +ve & -ve portions of $G_n(f)$ towards left & right sides respectively by an amount f_0 , so that the portion of the plot originally located at f_0 now coincide with the ordinate and by adding these two plots, the plot for either $n_c(t)$ or $n_s(t)$ can be obtained

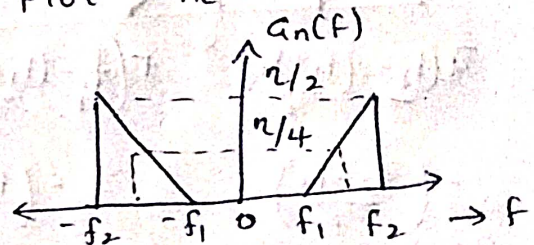
→ Problem :-

The noise PSD for noise $n(t)$ is shown in fig. Plot the PSD for $n_c(t)$ or $n_s(t)$ i.e., plot $G_{nc}(f)$ or $G_{ns}(f)$

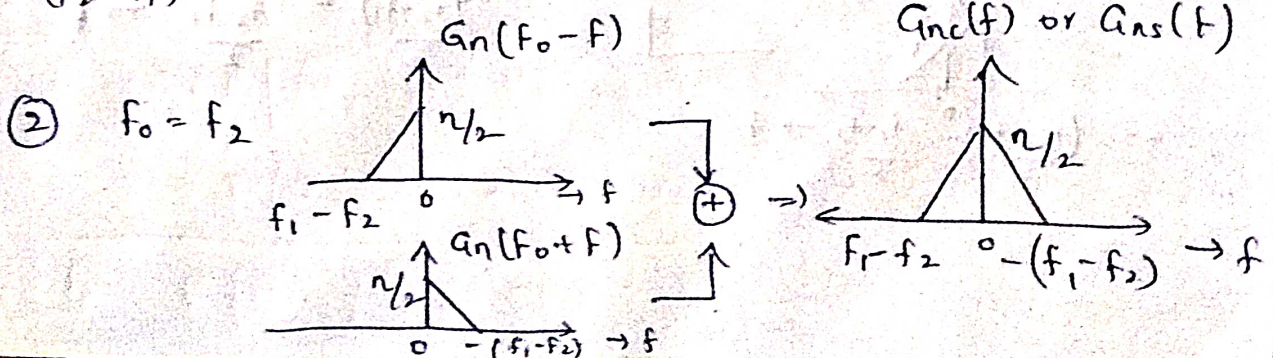
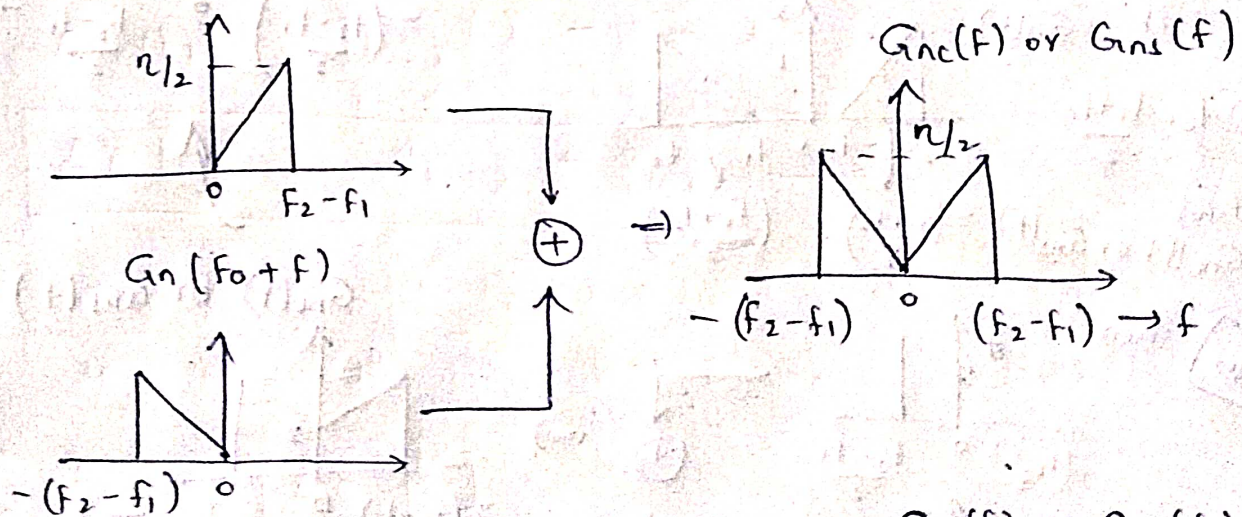
① $f_0 = f_1$

② $f_0 = f_2$

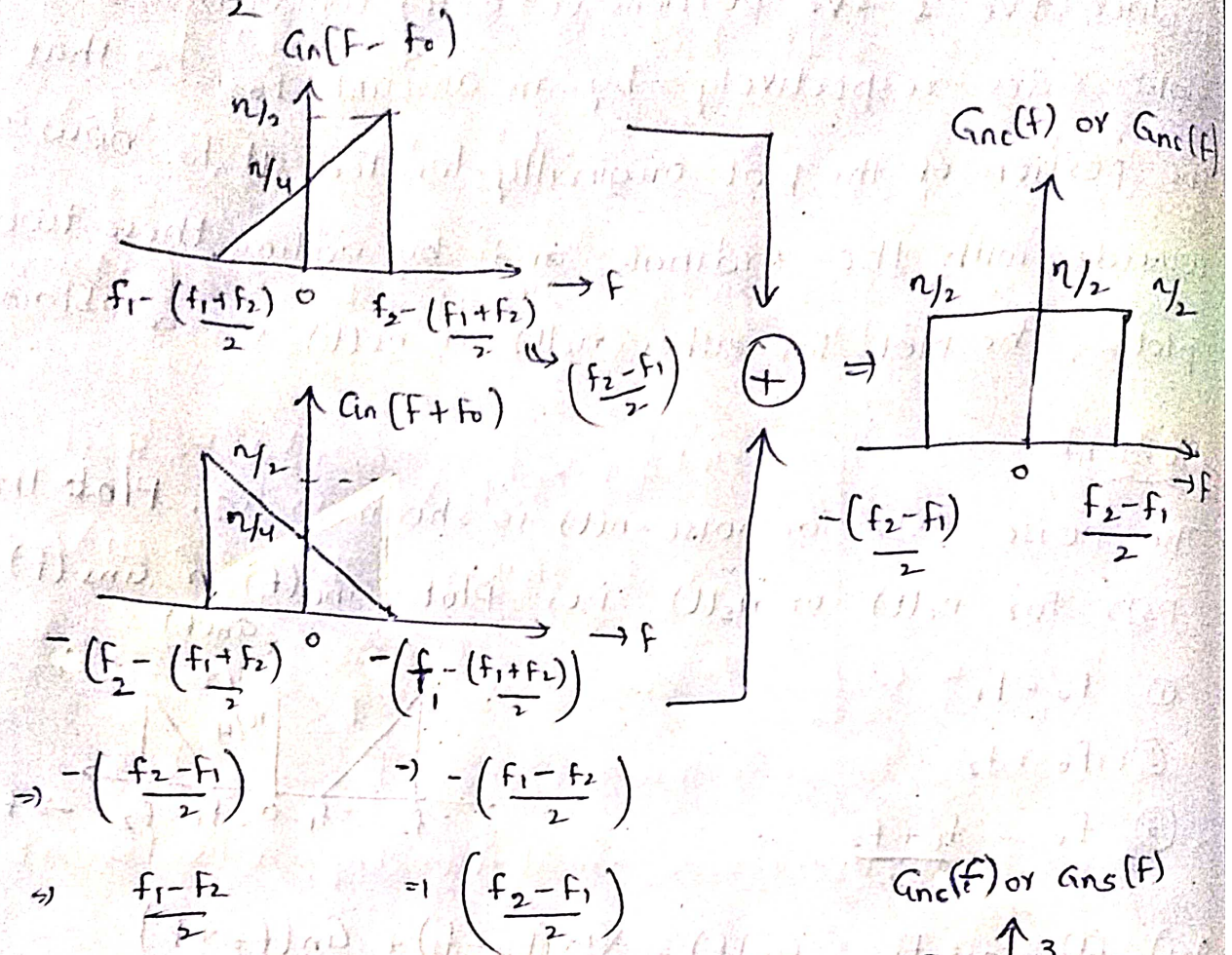
③ $f_0 = \frac{f_1 + f_2}{2}$



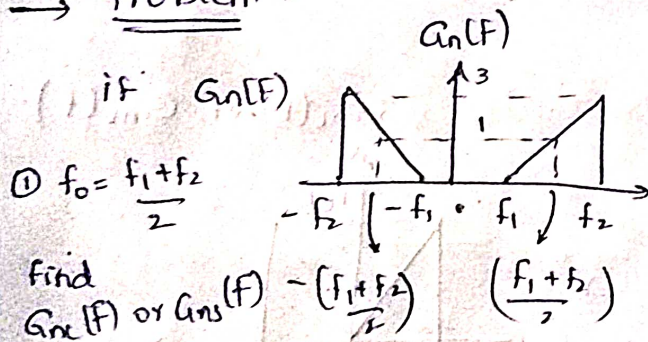
Ans] ① $f_0 = f_1$ $G_n(f) = G_n(f_0 - f) + G_n(f_0 + f)$
 $= G_{nc}(f)$



③ $f_0 = \frac{f_1 + f_2}{2}$



→ Problem ..



Ans) then

