

DSP exercise

10.14

$$y(n) = x(n) + \left(\frac{2}{3}\right) \cdot y(n-1)$$

Taking Z.T.

$$Y(z) = X(z) + \left(\frac{2}{3}\right) \cdot z^{-1} \cdot Y(z)$$

$$Y(z) [1 - \left(\frac{2}{3}\right) z^{-1}] = X(z)$$

$$H(z) = \frac{1}{1 - \left(\frac{2}{3}\right) \cdot z^{-1}}$$

Back word filtering

$$H(z^{-1}) = \frac{1}{1 - \left(\frac{2}{3}\right) (z^{-1})^{-1}} = \frac{1}{1 - \left(\frac{2}{3}\right) (z)}$$

Forward- Backward Filter. ~~Reu~~ Transfer function

$$H_2(z) = H(z) \cdot H(z^{-1})$$

$$= \frac{1}{1 - \left(\frac{2}{3}\right) z^{-1}} \cdot \frac{1}{1 - \left(\frac{2}{3}\right) z}$$

$$= \frac{1}{1 - \left(\frac{2}{3}\right) z - \left(\frac{2}{3}\right) z^{-1} + \left(\frac{4}{9}\right)} = \frac{1}{\left(-\frac{2}{3}\right) z^2 + \left(\frac{13}{9}\right) z - \left(\frac{2}{3}\right)}$$

$$= \frac{(-3/2)}{z^2 - (13/6)z + 1} = \left(\frac{-3}{2}\right) \frac{1}{(z - 2/3)(z - 3/2)}$$

$$= \left(\frac{-3}{2}\right) \left[\frac{-6/5}{(z - 2/3)} + \frac{6/5}{z - 3/2} \right]$$

$$= \left(\frac{-3}{2}\right) \left(\frac{-6}{5}\right) \left[\frac{1}{(z - 2/3)} - \frac{1}{(z - 3/2)} \right]$$

$$H_2(z) = \left(\frac{9}{5}\right) \left[\frac{1}{(z - 2/3)} - \frac{1}{z - 3/2} \right]$$

Taking stable inverse Z transform.

$$h(n) = \left(\frac{9}{5}\right) \left[\left(\frac{2}{3}\right)^n \cdot u[n] + \left(\frac{2}{3}\right)^n \cdot u[-n-1] \right]$$

10.14

classmate

Date

Page

Frequency Response.

$$H_2(e^{j\omega}) = H(e^{j\omega}) \cdot H(e^{-j\omega})$$

$$\text{Mag}(H_2(e^{j\omega})) = |H(e^{j\omega})|^2$$

$$|H(e^{j\omega})| = \left| \frac{1}{1 - (2/3)e^{j\omega}} \right|^2$$

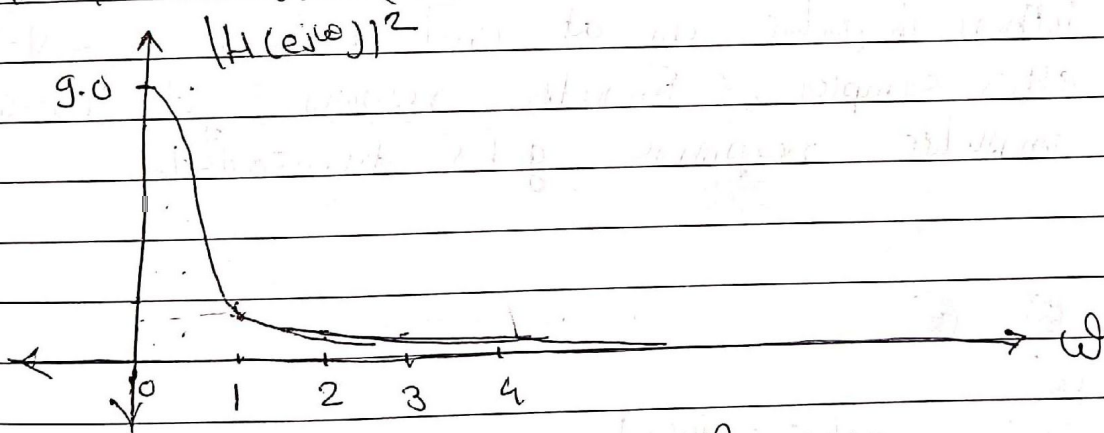
$$\omega = 0 \Rightarrow |H_2(e^{j0})|_{\omega=0}^2 = (3)^2 = 9$$

$$\omega = 1 \quad |H_2(e^{j1})| = (1.1752)^2 = 1.3810$$

$$\omega = 2 \quad |H_2(e^{j2})| = (0.7072)^2 = 0.5002$$

$$\omega = 3 \quad |H_2(e^{j3})| = (0.6014)^2 = 0.3617$$

$$\omega = 4 \quad |H_2(e^{j4})| = (0.6571)^2 = 0.4318$$



Frequency Response.

(a) Forward-Backward Filtering can be used to implement linear (Zero) phase filtering from nonlinear phase (IIR) filter.

(e) It is easier and possible to design FIR filter as linear.

(f) Phase
filter function in MATLAB implements forward-backward filtering

Least Squares in Signal Processing

classmate

Date _____

Page _____

(51)

$$\min_x \|y - Ax\|_2^2 + \lambda \|b - x\|_2^2$$

$$J(x) = \|y - Ax\|_2^2 + \lambda \|b - x\|_2^2$$

$$J(x) = (y - Ax)^T \cdot (y - Ax) + \lambda \cdot (b - x)^T \cdot (b - x)$$

$$= (y^T - x^T A^T) \cdot (y - Ax) + \lambda (b^T - x^T) \cdot (b - x)$$

$$= y^T \cdot y - y^T A x - x^T A^T y + x^T A^T A x + \lambda (b^T \cdot b - b^T x - x^T \cdot b + x^T x)$$

$$= y^T y - 2 \cdot y^T A x + x^T A^T A x + \lambda (b^T b - 2 b^T x + x^T x)$$

Minimizing x

$$\frac{\partial}{\partial x} J(x) = (-2 y^T A) + 2 A^T A x + \lambda (-2 b + 2 x)$$

Setting $\frac{\partial}{\partial x} J(x) = 0$

$$-2 A^T y + (2 A^T A) x + (-2 \lambda \cdot b) + (2 \lambda) x = 0$$
$$x (2 A^T A + 2 \lambda) = (2 A^T y + 2 \lambda \cdot b)$$

$$x = \frac{1}{(A^T A + \lambda)} (A^T y + \lambda \cdot b)$$

15.2

$$J(x) = \lambda_1 \|b_1 - A_1 x\|_2^2 + \lambda_2 \|b_2 - A_2 x\|_2^2 + \lambda_3 \|b_3 - A_3 x\|_2^2$$

$$= (\lambda_1)(b_1 - A_1 x)^T (b_1 - A_1 x) + (\lambda_2)(b_2 - A_2 x)^T (b_2 - A_2 x) + (\lambda_3)(b_3 - A_3 x)^T (b_3 - A_3 x)$$

$$= \lambda_1 (b_1^T b_1 - b_1^T A_1 x - x^T A_1^T b_1 + x^T A_1^T A_1 x) + \lambda_2 (b_2^T b_2 - b_2^T A_2 x - x^T A_2^T b_2 + x^T A_2^T A_2 x) + \lambda_3 (b_3^T b_3 - b_3^T A_3 x - x^T A_3^T b_3 + x^T A_3^T A_3 x)$$

$$= \lambda_1 (b_1^T b_1 - 2 b_1^T A_1 x + x^T A_1^T A_1 x) + \lambda_2 (b_2^T b_2 - 2 b_2^T A_2 x + x^T A_2^T A_2 x) + \lambda_3 (b_3^T b_3 - 2 b_3^T A_3 x + x^T A_3^T A_3 x)$$

$$\text{Setting } \frac{\partial}{\partial x} J(x) = 0$$

$$\lambda_1 (-2 A_1^T b_1 + 2 A_1^T A_1 x) + \lambda_2 (-2 A_2^T b_2 + 2 A_2^T A_2 x) + \lambda_3 (-2 A_3^T b_3 + 2 A_3^T A_3 x) = 0$$

$$2x (\lambda_1 A_1^T A_1 + \lambda_2 A_2^T A_2 + \lambda_3 A_3^T A_3) + (-2)(\lambda_1 A_1^T b_1 + \lambda_2 A_2^T b_2 + \lambda_3 A_3^T b_3) = 0$$

$$x (\lambda_1 A_1^T A_1 + \lambda_2 A_2^T A_2 + \lambda_3 A_3^T A_3)$$

$$= (\lambda_1 A_1^T b_1 + \lambda_2 A_2^T b_2 + \lambda_3 A_3^T b_3)$$

$$x = (\lambda_1 A_1^T A_1 + \lambda_2 A_2^T A_2 + \lambda_3 A_3^T A_3)^{-1} (\lambda_1 A_1^T b_1 + \lambda_2 A_2^T b_2 + \lambda_3 A_3^T b_3)$$

(15.4)

$$J(x) = \|(Ax - b)\|_2^2$$

$$J(x) = (Ax - b)^T (Ax - b) = x^T A^T A x - x^T A^T b$$

$$- b^T A x + b^T b$$

$$\frac{\partial J(x)}{\partial x} = 2(A^T A x - A^T b)$$

$$= 2(A^T A x - A^T b)$$

$$= 2A^T(Ax - b)$$

(16.2)

$$x = (I + \lambda D^T D)^{-1} y, \quad \lambda \geq 0$$

$D \rightarrow$ first order difference

(15.3)

Matrix is invertible when its determinant is not equal to '0' (zero) by loading diagonals. of HTH its determinant becomes non-zero.

(15.5)

Large value λ can smooth out even large crest, and trough of signal along with noise in the data.

In an ECG signal large value of λ can destroy useful information in signal.

(15.6)

To check effect of higher order difference. ~~in~~ in smoothing demo amplitude of noise is increased.

Compare to second order third order output produces more curves. i.e. it tries to match the estimate with noise.

(16.1)

When impulse is at end $n_0=0$ or $n_0=N-1$ other samples of impulse response get cut-off. impulse response gets truncated.

(16.2)

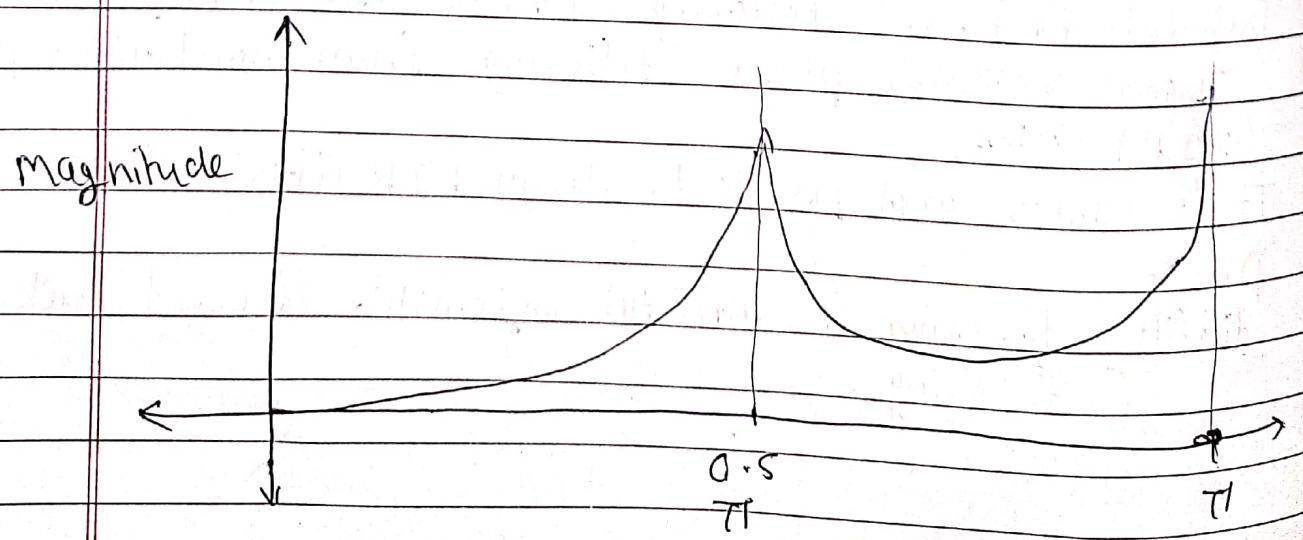
(a) (b)

(a)

(16.3)

(b)

$$|G(e^{j\omega})| =$$



(16.3)

(a)

$$G(z) = \frac{1}{H(z)}$$

$$G(e^{j\omega}) = \frac{1}{H(e^{j\omega})}$$