

Basic Electrical Engineering

NOTES by
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Book: B.L. Thapar

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Syllabus:

1. Electric Circuits

Introduction to linear and non-linear circuits, circuit elements, various sources and source transformation, star delta transformation, solution of D.C. circuits using Kirchoff's laws, signal wave forms and passive elements specifications, basic theorems, generation of A.C. sinusoidal voltage and currents, average and r.m.s. values, form factor and peak factor, phasor representation, phasor in polar, rectangular and exponential forms, Terminal relationship for pure passive elements & their combination in series and parallel.

Analysis of single phase series, parallel and series-parallel circuits. Active and reactive power p.f. and volt-amperes, frequency response and Q-factor. Analysis of balanced three phase a.c. circuits - Introductory concepts, voltage, current and power in three phase balanced circuits.

Introduction to Domestic Electric Lighting & Storage Batteries

2. Electromagnetic and Transformer.

Magnetic circuit concept, B-H curves characteristic of magnetic materials, practical magnetic circuits, magnetic circuits with DC and AC excitation, hysteresis and eddy current losses.

3. Measuring Instruments

Introduction to galvanometer (Moving coil and Moving Iron), ammeter, voltmeter, wattmeter, energy meter, use of shunt and multiplier.

4. Electrical machines

Fundamentals of DC & AC machines.

Magnetic Force, self and mutual inductances, Faraday's laws, Lenz's laws, statically & dynamically induced emfs, energy stored in magnetic fields, Principle of Transformer operation, construction & equivalent circuit of transformer.

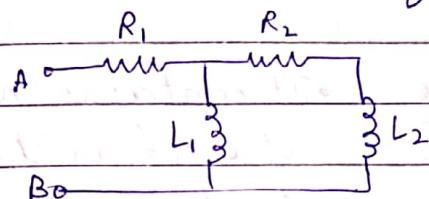
Ch-2 DC Network Theorems

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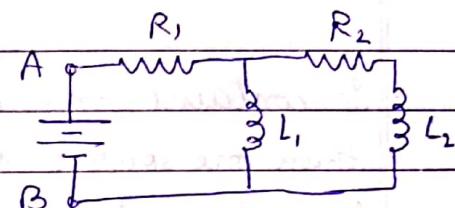
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2.1 Network - It is the interconnection of different circuit elements R, L, and C.

Circuit - It is a closed energized network where current flows or intended to flow.



Network

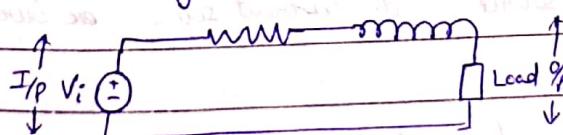


Circuit

Classification of Circuits

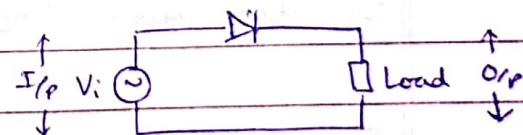
1. Linear circuit

- It is determined by R , L and C , waveforms and frequency
- Have linear V and I characteristics
- Linear input and output characteristics.
- It obeys Ohm's law.



Non-linear circuit

- It contains transistor, diode, transformer
- Non-linear V and I characteristics
- non-linear I/P and O/P characteristics
- Does not follow Ohm's law.



2. Bilateral element

Its properties or characteristics are the same in either direction.

e.g. transmission wires

Unilateral elements

Its parameters or characteristics change with the direction of its operation.

e.g. diode

Simple wire has only resistive and inductive elements but

Transmission wire has resistive, capacitive and inductive elements.

3. Active network

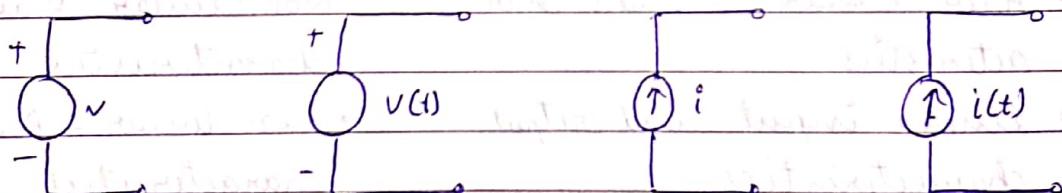
- It contains one or more than one source of e.m.f.
- It needs active elements like DC generator, AC generator, Batteries, photoelectric cell

Passive network

- It contains no source of e.m.f. in it.
- It does not contain active elements but contains passive like, R , L , C , diode, transistor

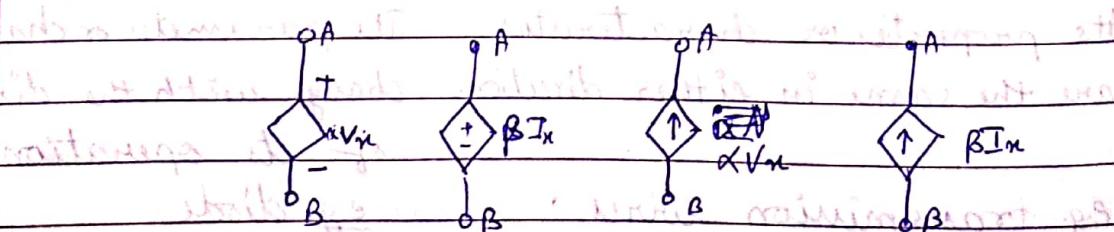
2.9

Independent sources - Those voltages or current sources, which do not depend on any other quantity in the circuit are called independent sources. e.g. generator, batteries, alternator.

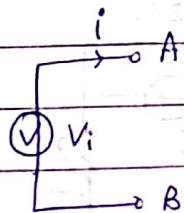


independent DC voltage source independent AC voltage source independent DC current source independent AC current source

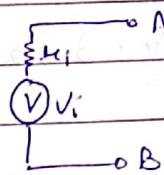
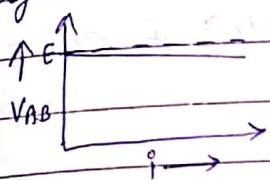
Dependent sources - These depend on some other quantity in the circuit which may be either a voltage or a current e.g. transistor, op-amp.



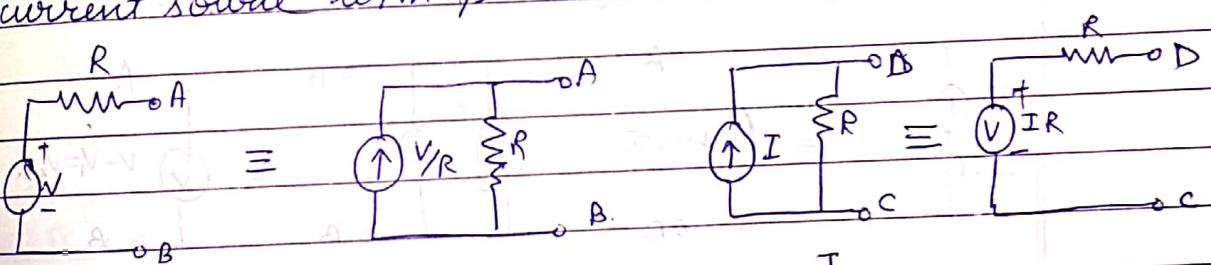
Ideal Source - There are no internal losses in source & it supplies the marked value.



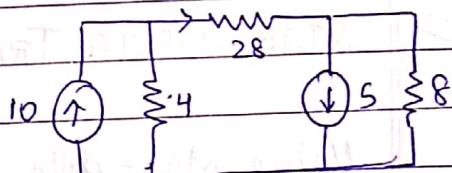
Practical Source - It has some internal resistance due to which voltage drop takes place, and it causes the terminal voltage to reduce.



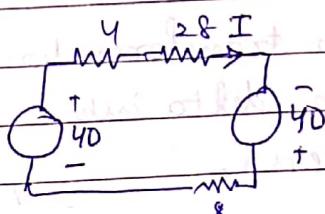
Source Conversion - A given voltage source with a series resistance can be converted into (replaced by) an equivalent current source with parallel resistance and vice versa.



Find the value of I



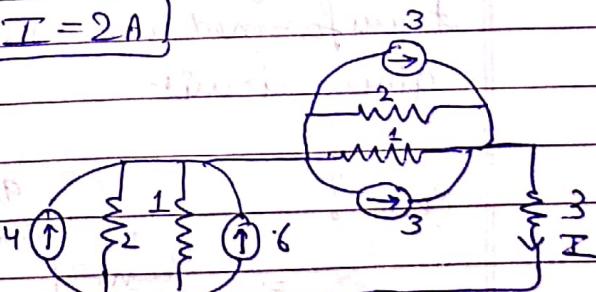
The given equivalent circuit is this by the method of source conversion



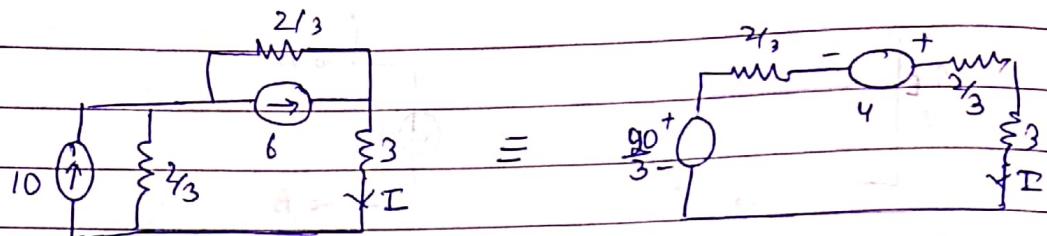
Now Applying KVL $40 - 4I - 28I + 40 - 8I = 0$

$$\Rightarrow 80 = 40I \Rightarrow I = 2A$$

Find the value of I



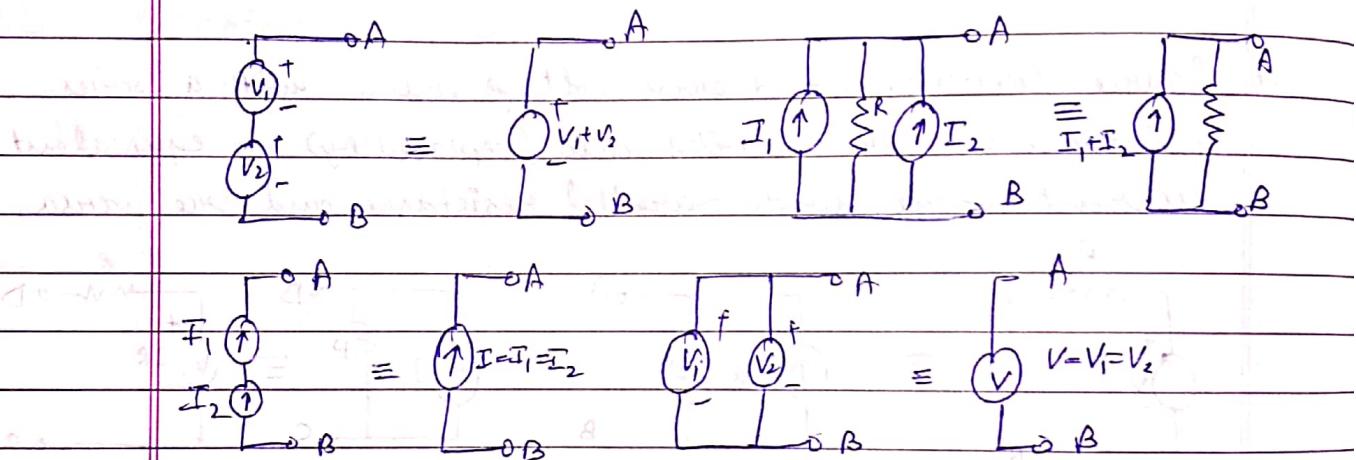
We apply source conversion on the given network after applying additive property



Applying KVL, $\frac{90}{3} - 2I + 4 - \frac{2}{3}I - 3 = 0$

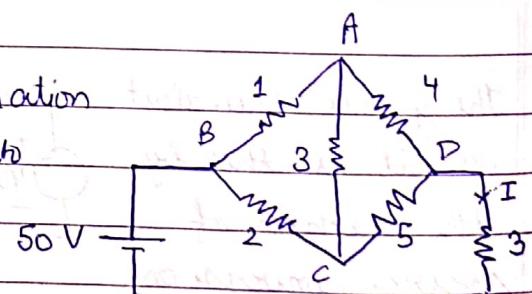
$$\Rightarrow \frac{20}{3} + 4 = \left(\frac{4+3}{3}\right)I \Rightarrow 20 + 12 = (4+3)I \Rightarrow I = \frac{32}{7} = 2.46 A$$

NOTE:

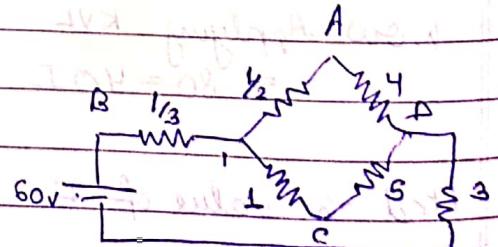


2.22 STAR-DELTA TRANSFORMATION

Using star-delta transformation we can convert a delta into a star & vice versa.



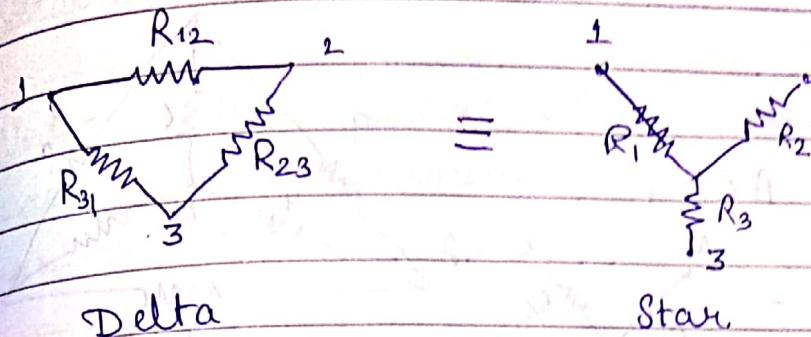
Thus the given circuit is transformed into the given circuit



$$R_{BD} = ?$$

$$R_{BD} = 2.9 \Omega$$

$$= \frac{18 + \frac{1}{3}}{\frac{27}{21}} = \frac{54 + 7}{21} = \frac{61}{21} = 2.9$$



Resistance between node 1 and 2 :

$$R_1 + R_2 = R_{12} \parallel (R_{31} + R_{23}) = \frac{R_{12}(R_{31} + R_{23})}{R_{12} + R_{23} + R_{31}} \quad \text{--- (1)}$$

Resistance between node 2 and 3 :

$$R_2 + R_3 = R_{23} \parallel (R_{12} + R_{31}) \quad \text{--- (2)}$$

Resistance between node 3 and 1 :

$$R_3 + R_1 = R_{31} \parallel (R_{12} + R_{23}) \quad \text{--- (3)}$$

Adding all the three equations, we get :

$$R_1 + R_2 + R_3 = \frac{R_{12}R_{31} + R_{12}R_{23} + R_{23}R_{31}}{R_{12} + R_{23} + R_{31}} \quad \text{--- (4)}$$

Subtracting (1) from (4) :

$$R_3 = \frac{R_{23}R_{31}}{R_{12} + R_{23} + R_{31}}$$

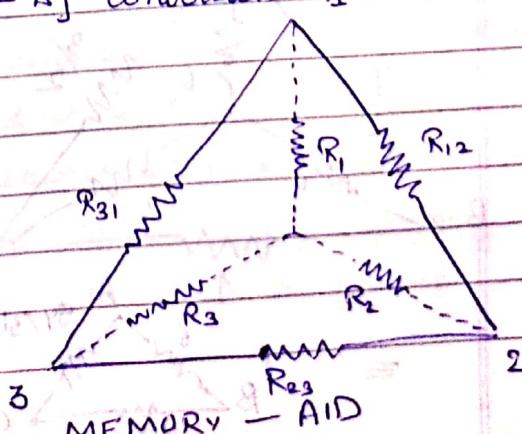
$$\text{Similarly : } R_1 = \frac{R_{12}R_{31}}{R_{12} + R_{23} + R_{31}}, \quad R_2 = \frac{R_{23}R_{12}}{R_{12} + R_{23} + R_{31}}$$

From Star to Delta [Y - A] conversion

$$R_{31} = R_3 + R_1 + \frac{R_3R_1}{R_2}$$

$$R_{12} = R_1 + R_2 + \frac{R_1R_2}{R_3}$$

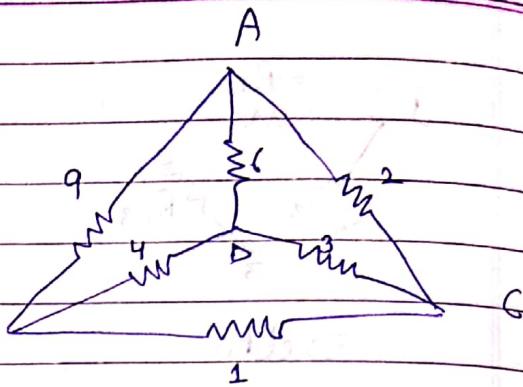
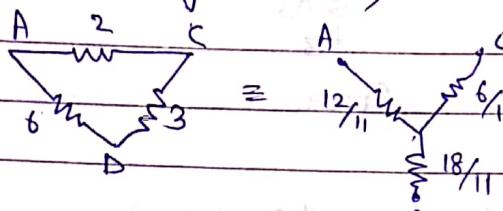
$$R_{23} = R_2 + R_3 + \frac{R_2R_3}{R_1}$$



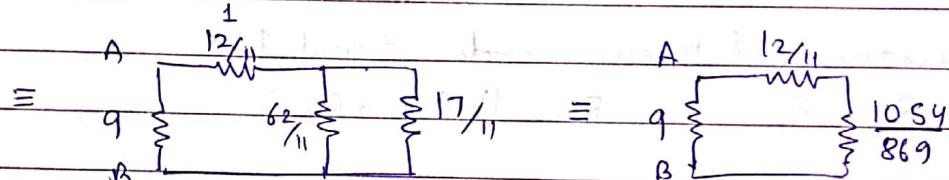
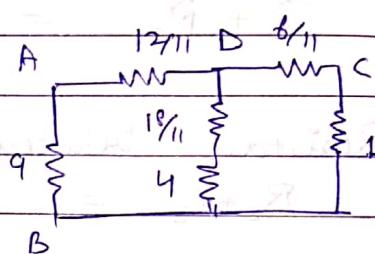
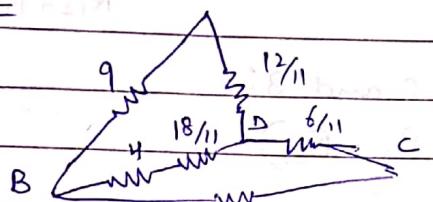
Example 2.9.2

Q. Find R_{AB}

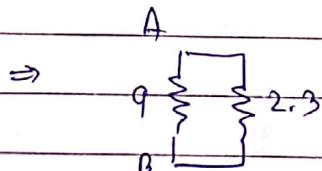
Considering ACD,



$$\textcircled{1} \quad (9+6) \parallel 2 = A(6+9) \parallel 18 = \frac{15}{2} = 7.5$$

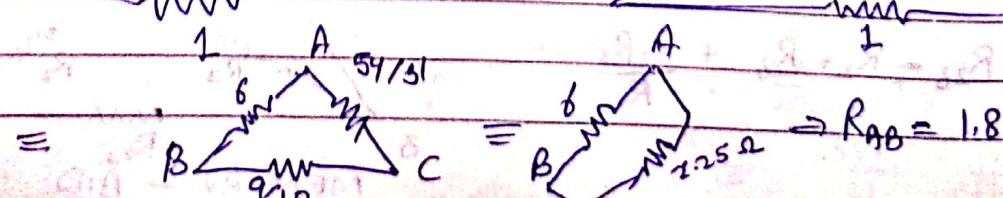
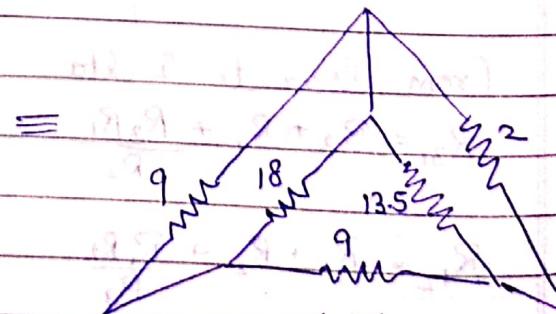
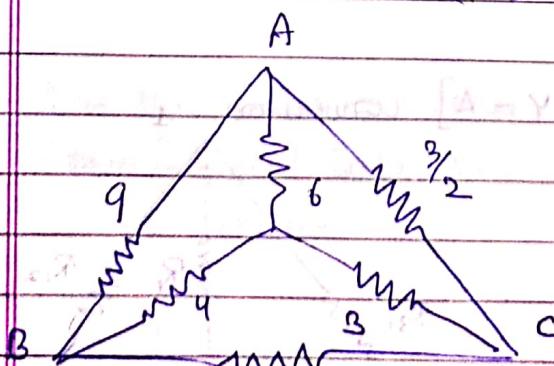


$$\textcircled{2} \quad \frac{\frac{17 \times 6^2}{4}}{\frac{62+17}{4}} = \frac{1054}{79 \times 11} = \frac{1054}{869} \quad \frac{1054 + 12}{869} = \frac{1066}{869} = 1.23$$



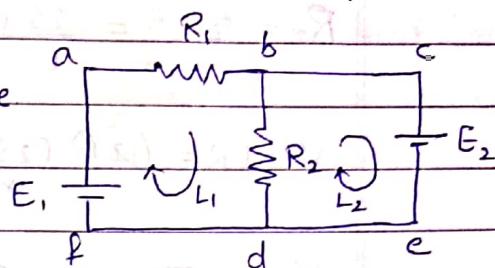
$$R_{AB} = 9 \times 2.3 = \frac{20.7}{11.3} = 1.83 \Omega$$

(OR)



Network Terminology:

1. Node - A node is a common point where at least two components are connected like connecting wire, resistor, battery etc. e.g. a, b, c, d, f, e



2. Junction - A common point

having at least three components connected to it. e.g. b, d.

3. Branch - It is that part of a network which lies between two junctions. e.g. baf, bcd, bd.

4. Loop - It is a close path in a circuit in which no element or node is encountered more than once. e.g. abdfa, bcedb, abcdefa.

5. Mesh - It is a subpart of a loop. It is the smallest loop which can not be subdivided. e.g. abdfa, bcedb

2.2 Kirchhoff's law

1. Kirchhoff's Current Law, KCL - Sum of all the incoming / entering current is equal to outgoing current.

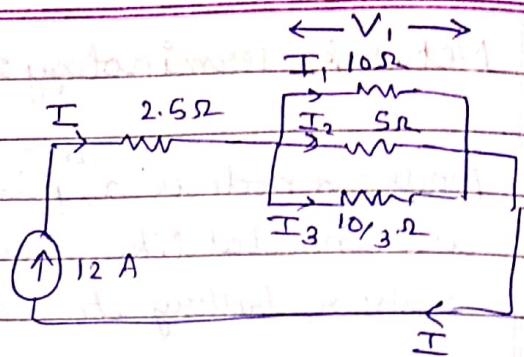
2. Kirchhoff's Voltage Law, KVL - The algebraic sum of products of currents and resistances in each of the conductors in any closed path (or mesh) in a network plus the algebraic sum of the emfs in that path is zero.

$$\sum IR + \sum \text{e.m.f.} = 0$$

Q. Calculate I_1, I_2, I_3 .

$$R_T = \frac{5}{2} + \frac{5}{3} = \frac{25}{6} \Omega.$$

$$V = IR = (12A) (2.5\Omega) = 30V.$$



$$V_{2.5} = IR_{2.5} = 12 \times 2.5 = 30V.$$

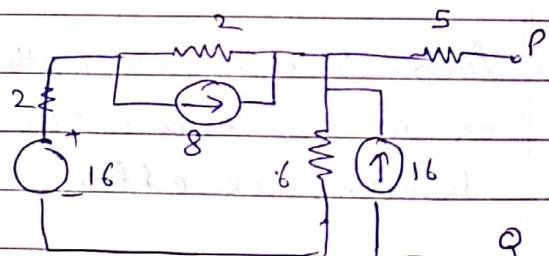
$$\Rightarrow V_1 = 50 - 30 = 20V.$$

$$I_1 = \frac{V_1}{R_1} = \frac{20}{10} = 2A. \quad I_3 = \frac{V_1}{R_3} = \frac{20}{10/3} = 6A$$

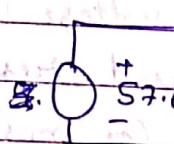
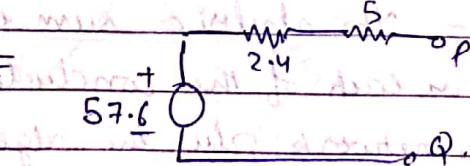
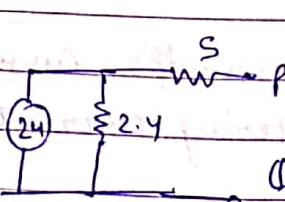
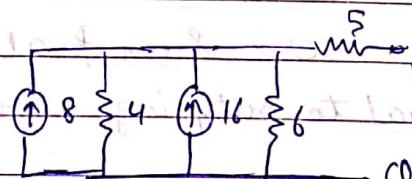
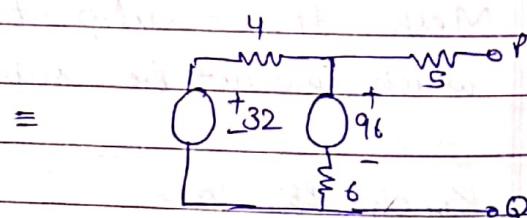
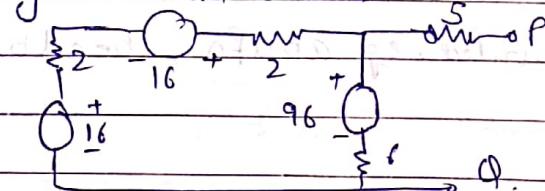
$$I_2 = \frac{V_1}{R_2} = \frac{20}{5} = 4A.$$

Q. Find single voltage

source equivalent of the full circuit.



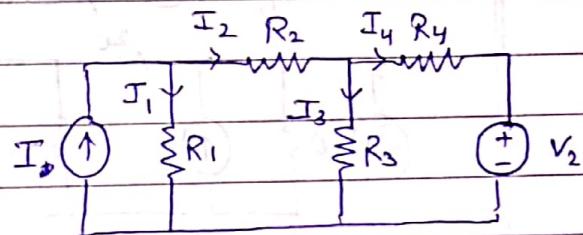
A. Given circuit becomes



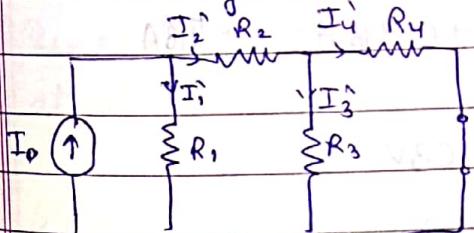
$$E_{eq} = \left[\frac{E_1 + E_2}{\tau_1 + \tau_2} \right] \tau_{eq} = \frac{-E_1 \tau_2 + E_2 \tau_1}{\tau_1 + \tau_2}$$

Superposition Theorem - It is applied to a multisource network. It states that the response in any branch of a linear, active bilateral network containing multisources is the algebraic sum of the responses obtained by considering each source acting separately during which all other sources are made zero.

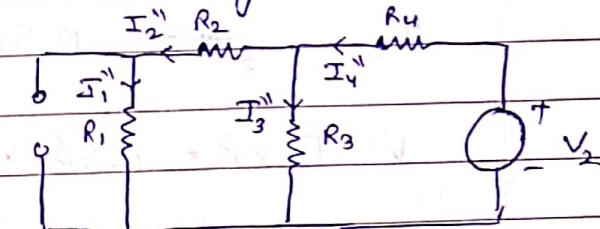
Since the given circuit is a multisource, linear, active and bilateral, we can apply superposition theorem.



Considering $V_2 = 0$.



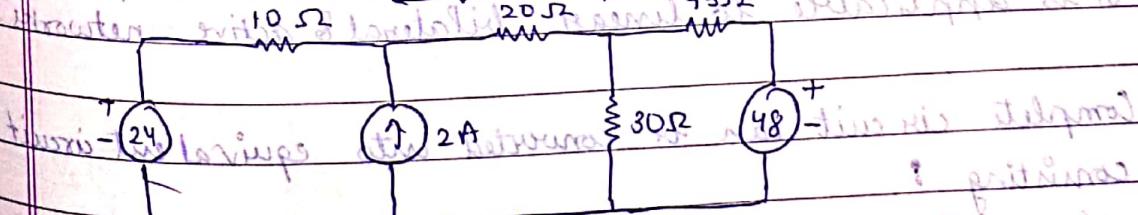
Considering I_1 to be open



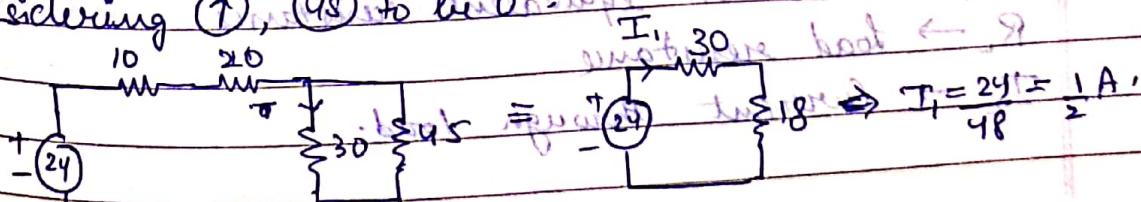
$$\text{Thus } I_2 = I_2' + I_2'' \text{ by } I_2 = I_2' - I_2'' \text{ in first}$$

Similarly express I_3 and I_4 by $I_3 = I_3' + I_3''$ and $I_4 = I_4' - I_4''$

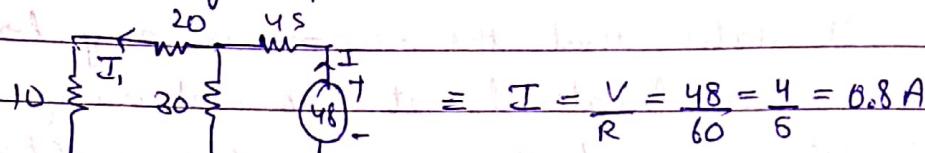
Q. Use superposition theorem to find voltage V_{ab}



Considering (1), (48) to be opening minuit $\leftarrow V$

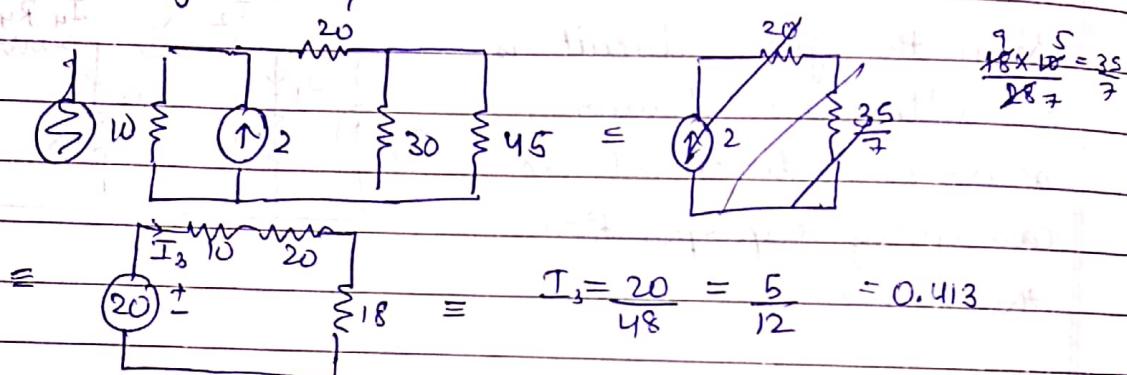


Considering (24), (48) to be 0.



$\Rightarrow I_1 = 0.4 \text{ A}$ & current divided by symmetry.

Considering (24), (48) to be zero



$$I_{20} = \frac{1}{2} I_3 = 0.5 \text{ A} - 0.8 \text{ A} + 0.413 \text{ A} = 0.516 \text{ A}$$

$$V = I_{20} \times R = 0.516 \text{ A} \times 20 = 10.3 \text{ V}$$

2.18 Thevenin Theorem - This theorem is applicable to determine current and voltage applied in element of network without going through complex procedures for going through solving a set of network equations. It is applicable to linear, bilateral & active networks.

Complete circuit can be converted into equivalent circuit consisting :

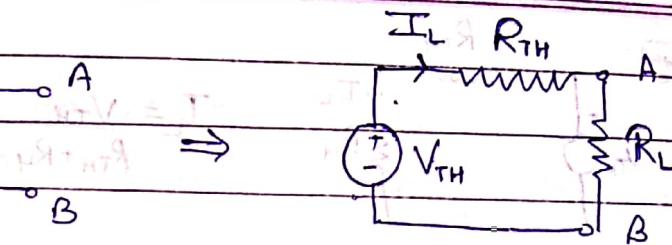
V_{th} \rightarrow Thevenin equivalent voltage source

R_{th} \rightarrow Thevenin equivalent resistance

R_L \rightarrow Load resistance

I_L \rightarrow Current through load.

Linear, active
bilateral
network

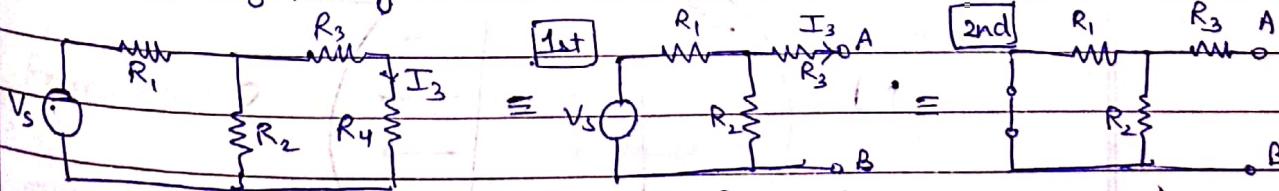


$$\text{We get } T_L = \frac{V_{TH}}{R_{TH} + R_L}$$

2.9 Steps to follow for converting a network to Thévenin's circuit

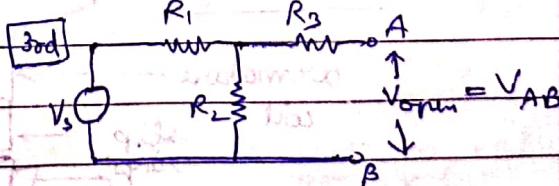
1. Remove load resistor R_L and calculate open circuit voltage across load terminal.
2. Determine Thévenin eq. Resistance R_{TH} by looking through open circuit terminal.
3. Replace voltage and current sources with no source at all.
4. Obtain Thévenin eq. circuit by connecting R_{TH}, R_L in series with open circuit voltage / V_{TH} .
5. Find / calculate the value of T_L .

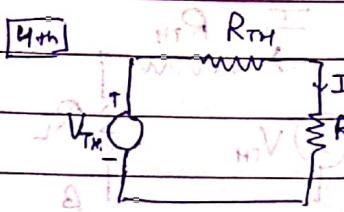
Q. Calculate I_3 using Thévenin's theorem.



$$\text{We get } R_{TH} = R_1 \parallel R_2 + R_3$$

$$V_{TH} = \frac{R_2 \times V_s}{R_2 + R_1}$$

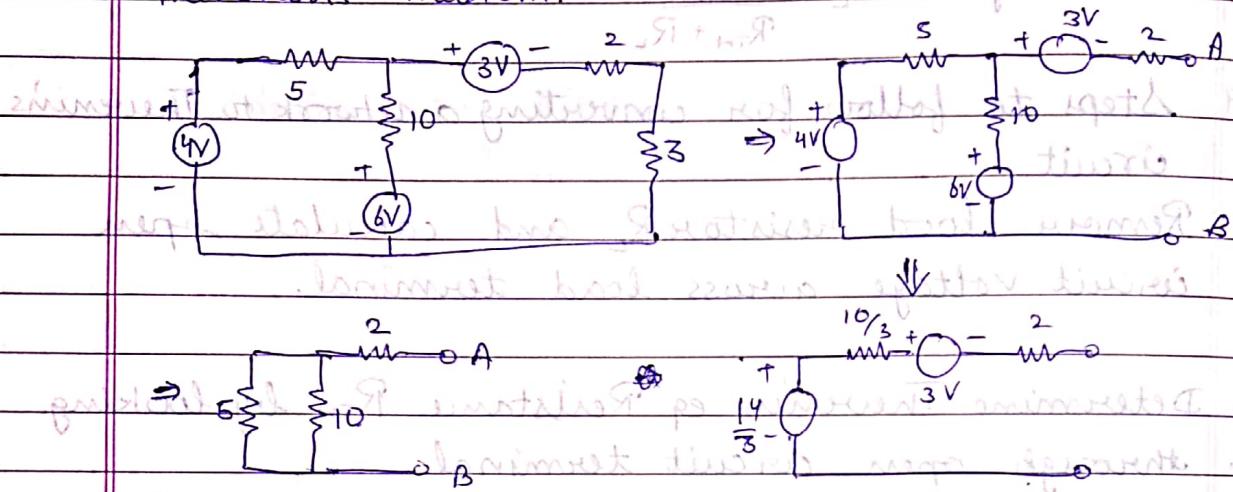




$$I_L = \frac{V_{TH}}{R_{TH} + R_L}$$

Induced
current

Q. Find the current in 3Ω resistor by using Thevenin's theorem.



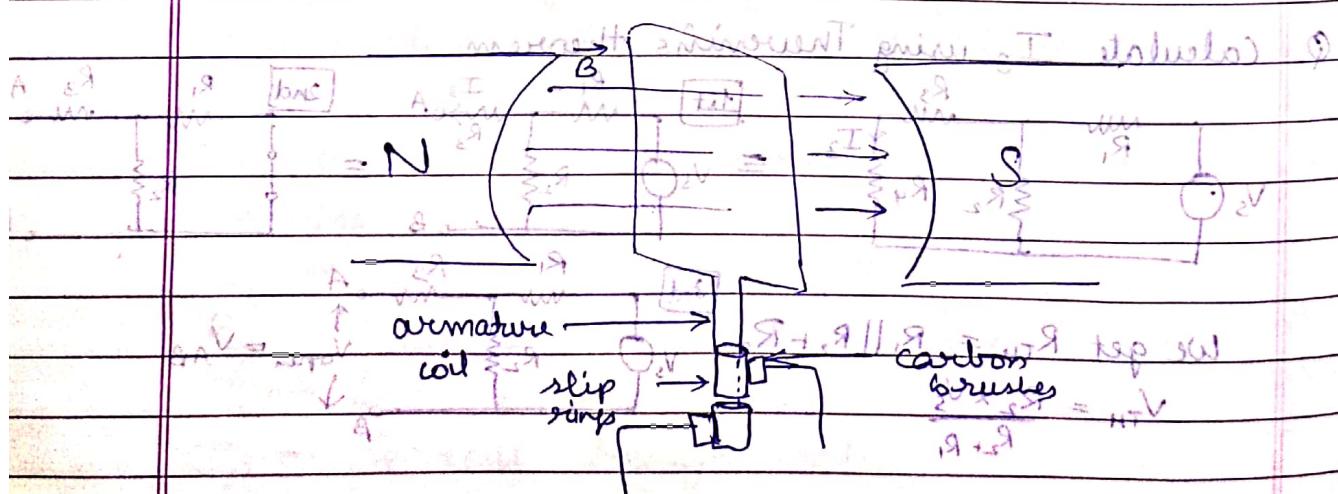
$$R_{TH} = \frac{5 \times 10}{5+10} + 2 = \frac{10+2}{3} = 16\Omega$$

$$V_{TH} = \frac{14}{3} - 3 = \frac{5}{3}V$$



$$\Rightarrow I = \frac{V_{TH}}{R_{TH} + R_L} = \frac{\frac{5}{3}}{16 + 3} = \frac{5}{25} = 0.2A$$

11.01 Generation of AC. to follow out induction built.



1. Mechanical rotation of the coil
2. Change in the flux passing through the coil.

$$\phi = \phi_m \cos \omega t$$

$$3. \text{ Induced emf } e = -N \frac{d\phi}{dt}$$

where N = no. of turns of coil

$$e = N \omega \sin \omega t \cdot \phi_m$$

$$\Rightarrow E_{\max} = \omega N \phi_m$$

$$e = E_m \sin \omega t$$

IF Z is the impedance.

$$I = \frac{e}{Z} = \frac{E_m \sin \omega t}{Z} = I_m \sin \omega t, I_m = \frac{E_m}{Z}$$

$$\Rightarrow i = i_m \sin \omega t$$

Cycle: A complete cycle is from 0 to 2π

Time period (T): It is the time required for a quantity to complete one cycle.

Amplitude: Maximum value of the quantity in a cycle is the amplitude.

Frequency: The no. of complete cycles per second.

RMS value (Root mean square value)

$$I_{\text{rms}} = \frac{I_m}{\sqrt{2}} \sin \omega t$$

$$V_{\text{rms}}^2 = \frac{1}{T} \int_0^T V^2 dt$$

$$\text{Mean or average value, } I_{\text{avg}} = \frac{2}{\pi} I_m = 0.636 I_m$$

$$V_{\text{avg}} = \frac{1}{T} \int_0^T V dt$$

$$\text{Form factor } K_f - K_f = \frac{\text{RMS value}}{\text{Average value}} = \frac{\pi}{2\sqrt{2}} = 1.11$$

It is different for different wave forms.

Peak or crest or amplitude factor $K_a = \frac{\text{maximum value}}{\text{rms. value}}$

$$K_a = 1.414 = \sqrt{2} \text{ for standard AC.}$$

Q. The equation of alternating current is $i = 42.42 \sin 628t$

Determine (i) its maximum value (ii) frequency (iii) its rms value (iv) its average

value. (v) form factor

A. (i) $i_m = 42.42$ (ii) $\omega = \frac{100}{\pi} = 25 \text{ rad/s}$ $f = 100 \text{ Hz}$

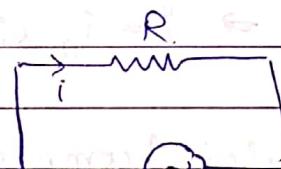
(iii) $i_m/\sqrt{2} = 30 = i_{\text{rms}}$

(iv) $i_{\text{avg}} = \frac{2}{\pi} i = 26.97 = 27$

(v) form factor $= \frac{i_{\text{rms}}}{i_{\text{avg}}} = \frac{30}{27} = 1.11$

11.28 AC series circuit

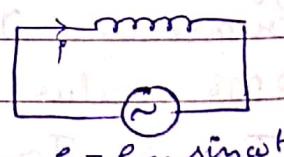
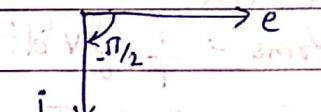
(i) Circuit with pure resistance



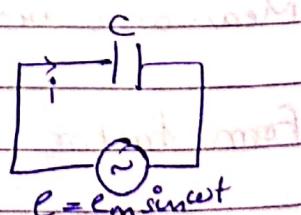
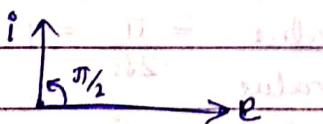
$$e = e_m \sin \omega t$$

$$I = \frac{e}{R} = \frac{e_m \sin \omega t}{R} = i_m \sin \omega t$$

(ii) Circuit with inductor



11.32 (iii) circuit with capacitor



(iv) R-L AC series circuit

A coil of resistance 12Ω and inductive reactance of 25Ω is connected in series with a capacitive reactance of 41Ω . The combination is connected with supply of $230V$, 50 Hz . Find (i) Z , (ii) I

$$Z = \sqrt{(X_L - X_C)^2 + R^2} = \sqrt{(41 - 25)^2 + 12^2} = \sqrt{16^2 + 12^2} = 20\Omega$$

$$I = \frac{V}{Z} = \frac{230}{20} = 11.5A$$

Waveforms → The shape of the curve obtained by plotting the instantaneous values of voltage or current as the ordinate against time as a abscissa is called its waveform or wave shape.

The rms value of alternating current is given by that steady (d.c.) current which when flowing through a given circuit for a given time produces the same heat as produced by the A.C. when flowing through the same circuit for the same time.

In electrical engineering work, unless indicated otherwise, the values of the given current & voltage are always the r.m.s. values.

The average value of an A.C. is expressed by that steady current (D.C.) which transfers across any circuit the same charge as is transferred by that A.C. during the same time.

* A.C. instruments measure r.m.s. value.

* $\text{rms value} > \text{avg. value}$ always except for rectangular wave where $I_{\text{rms}} = I_{\text{avg}}$

* The knowledge of form factor will enable the r.m.s.

Ch-19 Polyphase Circuits

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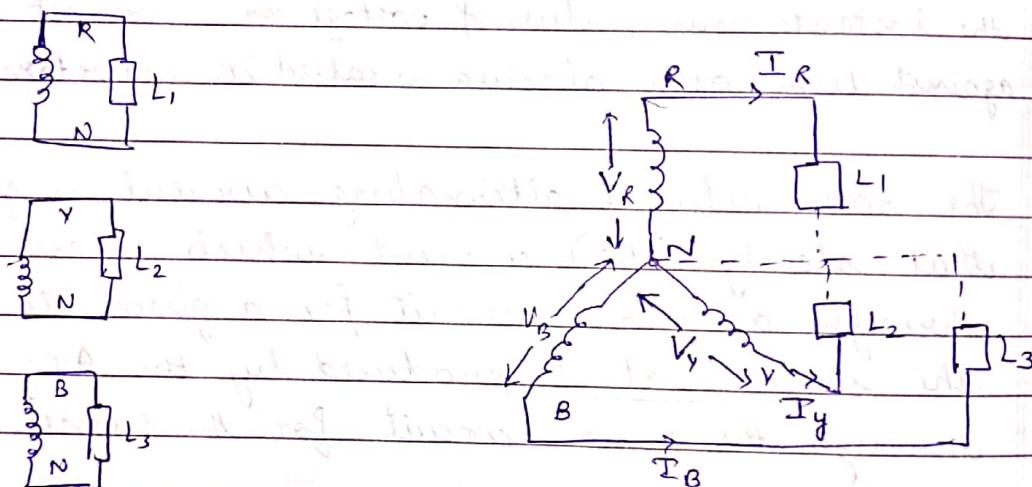
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value to be found from the arithmetic mean value and vice versa

- * Knowledge of crest factor is of importance in dielectric insulation testing, because the dielectric stress to which the insulation is subjected, is proportional to the max. or peak value of the applied voltage.

P-631

19.5 Three-phase a.c. circuit



Star or Wye (γ) connection

1- ϕ a.c.

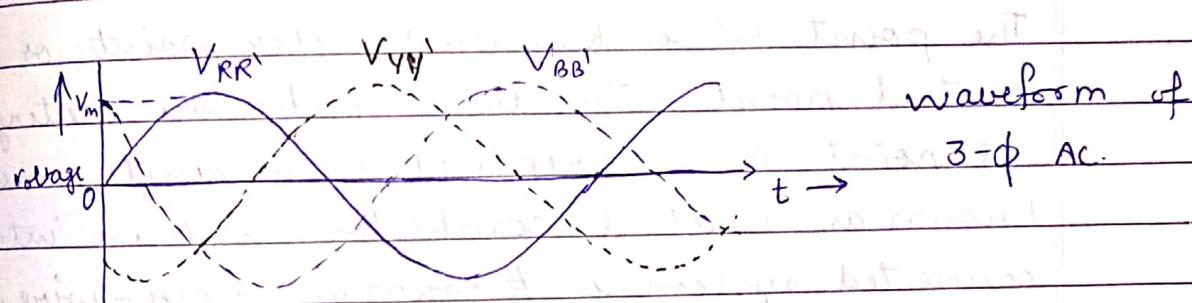
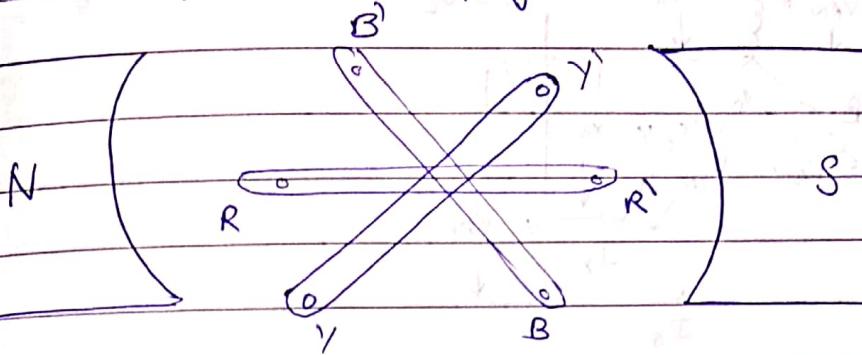
3- ϕ a.c.

Advantages of 3- ϕ a.c.

1. Less requirement of wire
2. Regular, stable, balanced a.c. supply
3. No disturbances
4. Current rating of 1- ϕ & 3- ϕ is same.
5. More power can be delivered. It is applicable only in case of power transmission & distribution. For higher power application 2- ϕ supply is not sufficient. e.g. industrial & commercial requirement of power.
- 6.

It contains 3 voltages, 3 currents, 3 conductors.

Generation of 3-φ supply.



If the phase difference is variable or amplitude is variable then the ~~the~~ 3-φ a.c. is unbalanced.
The 3-φ a.c. is balanced if 1φ & amplitude are constant.

$$V_{RR'} = V_m \sin \omega t$$

$$V_{BB'} = V_m \sin (\omega t - 2\pi/3)$$

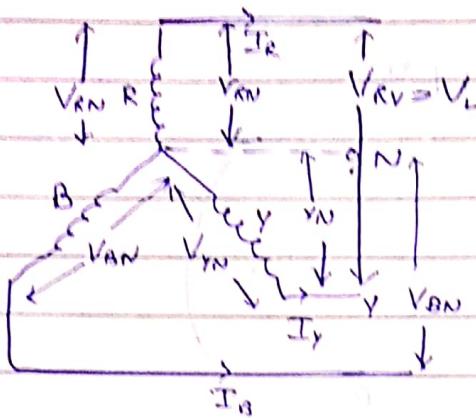
$$V_{YY'} = V_m \sin (\omega t - 4\pi/3) = V_m \sin (\omega t + 2\pi/3)$$

3-φ a.c. is almost constant always.

3-φ power is more constant compared to single phase
3-φ system has uniform torque & machines run more smoothly.

Power transmission by 3-φ is very economical.

3- ϕ Star connection (or) Wye connection (Y)



The point N is known as star point or neutral point. The three conductors meeting at point N are replaced by a single conductor known as neutral conductor. Such an interconnected system is known as four-wire, 3-phase system. If it is applied across a balanced symmetrical load, the neutral wire will be carrying three currents which are exactly equal in magnitude but are 120° out of phase with each other. Hence, their vector sum is zero $I_R + I_Y + I_B = 0$.

- (i) The arrows placed alongside I_R , I_Y & I_B flowing in the three phases indicate the directions of currents when they are assumed to be positive and not the directions at a particular instant.

At no instant will all the three currents flow in the same direction either outwards or inwards. The three arrows indicate that first the current flows outwards in phase R, then after a phase time of 120° , it will flow outwards from phase Y and after a further 120° , outwards from phase B.

the current flowing outwards in one or two conductors is always equal to that flowing inwards in the remaining conductor(s).

~~V_{YRN}~~

~~I = 100 A - outwards~~

~~A current flows from phase to neutral~~

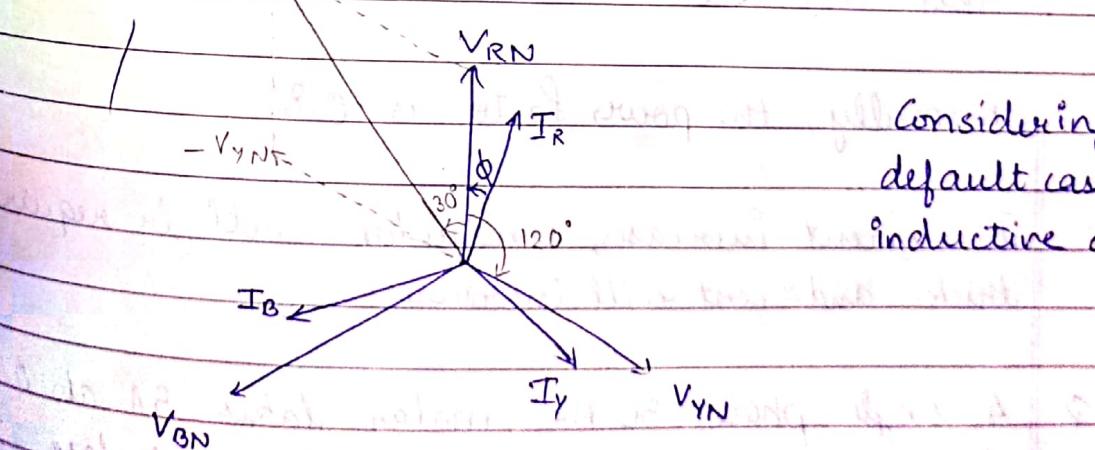
~~current flowing through neutral~~

~~Neutral current = zero~~

~~Phase current = zero~~

~~Line current = 0 A~~

In Star connection, line current and phase current is same. V_{YR}



Considering the default case i.e. inductive circuit

Applying KVL in RYL loop $V_{RY} = V_{RN} - V_{YN}$

$$V_{RY} = \sqrt{[V_{RN}]^2 + [V_{YN}]^2 + 2[V_{RN}] |V_{YN}| \cos 60^\circ}$$

$$V_L = \sqrt{3} V_{PH}$$

- * Line voltages are 120° apart
- * Line voltages are 30° ahead of phase voltages
- * Line voltage = $\sqrt{3}$ times phase voltage
- * Line current = phase current
- * In balanced condition, potential at neutral is 0.

$$\text{Power factor} \Rightarrow \cos\phi = \frac{R}{Z}$$

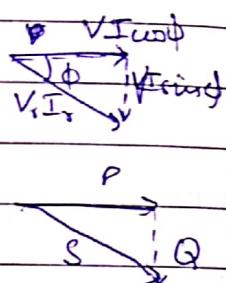


conversion of normal Δ to power Δ

$$\Rightarrow \text{actual power } P = \text{active power} = VI \cos\phi$$

$$Q_{ef} \text{ reactive power} = VI \sin\phi$$

$$S \text{ Apparent power} = V_r I_r$$



$$\cos\phi = \frac{P}{S} = \frac{P}{V_r I_r} \Rightarrow I_r = \frac{P}{V_r \cos\phi}$$

If power factor is low, machines will start taking more current which is dangerous
Heat loss will be more = $I_r^2 R$ losses

normally, the power factor is 0.8

If current increases, conductor will be required thick and cost will increase

- Q. A single phase 50 Hz motor takes 5A at 0.6 Pf lagging from 230 V power supply. Calculate
- active and reactive component of current
 - active power
 - The output of motor if the efficiency of motor at this load is 0.85.

$$I_r = 5 \text{ A}, f = 50 \text{ Hz}, \cos\phi = 0.6, V_r = 230 \text{ V}$$

A) active current component = $I_r \cos\phi = 5 \text{ A} \times 0.6 = 3 \text{ A}$

B) reactive component = $I_r \sin\phi = 5 \text{ A} \sqrt{1 - (0.6)^2} = 5 \times 0.8 = 4 \text{ A}$

C) active power = $I_r V_r \cos\phi = 5 \text{ A} \times 230 \text{ V} \times 0.6 = 690 \text{ W}$

D) Output = $n \times \text{input.} = n \times \text{active power}$
 $= 0.85 \times 690 \text{ W} = 586.5 \text{ W}$

3-φ power : Y-star connection

$$V_L = \sqrt{3} V_{ph}, I_L = I_{ph}$$

$$P = V_R I_R \cos\phi + V_Y I_Y \cos\phi + V_B I_B \cos\phi$$

For a balanced connection, phase difference is equal

$$\Rightarrow P = (V_R I_R + V_Y I_Y + V_B I_B) \cos\phi$$

$$\text{Also } V_R = V_Y = V_B = V_{ph} \text{ and } I_R = I_Y = I_B = I_{ph}$$

$$\Rightarrow \text{Total 3-φ power} = 3 V_{ph} I_{ph} \cos\phi \quad (\text{phase values})$$

$$= \sqrt{3} V_L I_L \cos\phi \quad (\text{line values})$$

Δ-Delta connection

$$\text{Total 3-φ power} = \sqrt{3} V_L I_L \cos\phi$$

- Q. A balanced star connected load of $(8+j6)\Omega$ per phase is connected to a balanced 3- ϕ , 400V supply. Find the line current, power factor and total volt-ampere.

A. $Z_L = (8+j6)\Omega$, $V_{ph} = 400V = V_L$

Line current $I = \frac{V_L}{Z_L} = \frac{\sqrt{3}V_{ph}}{R} = \frac{400\sqrt{3}}{8} = 10A$

Power factor $\cos\phi = R/Z$

$Z_{ph} = (R + jX) = (8 + j6) \Rightarrow R = 8, X = 6 = |X_L - X_C|$

cos when it is not given that is it line or phase voltage then voltage is given V_{line}

$Z = \sqrt{8^2 + 6^2} = 10$

$\cos\phi = \frac{R}{Z} = \frac{8}{10} = 0.8$ lagging (inductive is dominant)
as j is positive

P = power $= \sqrt{3}V_I \cos\phi = \sqrt{3}400 \times 23.1 \times 0.8 = 12803.32 \text{ Watt}$

$V_{ph} = \frac{V_L}{\sqrt{3}} = \frac{400}{\sqrt{3}} = 231V$

 $I_{ph} = \frac{V_{ph}}{Z_{ph}} = \frac{400}{\sqrt{3} \times 10} = \frac{40}{\sqrt{3}} = 23.094$
 $= 23.1A$

Total volt-ampere, apparent power $S = V_L I_L$

$S = \sqrt{3} \times 400 \times 23.1 = 16000 \text{ Watt} \quad 16 \text{ kW}$

$\Rightarrow P_{tan\phi} \quad Q = \sqrt{3} V_L I_L \sin\phi = 16000 \times 0.6 = 9.6 \text{ kVAR}$

$\Rightarrow P = 12.8 \text{ kW}$

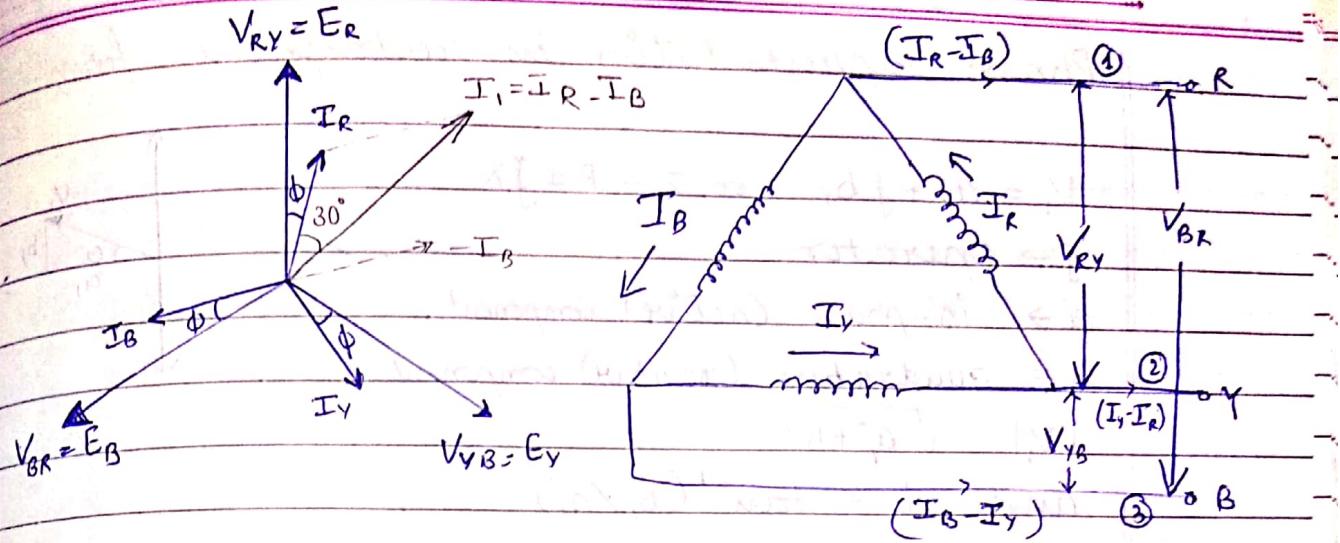
$Q = 9.6 \text{ kVAR}$

$\Rightarrow S = 16 \text{ kVA}$

Domestic Wiring Delta or Mesh Connection -

In this form of interconnection, the dissimilar ends of the three phase windings are joined together. In the system (balanced), the sum of three voltages round the closed mesh is zero.

In it Line voltage = Phase voltage $V_L = V_{ph}$



Current in line 1 is $I_1 = I_R - I_B$
 " line 2 is $I_2 = I_Y - I_R$
 " line 3 is $I_3 = I_B - I_Y$

If $I_B = I_R = I_Y$ = Phase current then

Current in line No. 1 is $I_1 = 2 \times I_{ph} \cos(60^\circ/2)$

$$I_1 = 2 \times I_{ph} \times \frac{\sqrt{3}}{2} = \sqrt{3} I_{ph}$$

$$\text{Similarly } I_2 = \sqrt{3} I_{ph}, I_3 = \sqrt{3} I_{ph}.$$

Since all line currents are equal in magnitude,

$$I_1 = I_2 = I_3 = I_L = \sqrt{3} I_{ph}$$

- * Line currents are 120° apart
- * line currents are 30° behind the respective phase currents
- * the angle between the line current and the corresponding line voltage is $(30 + \phi)$ with the current lagging.

$$\text{Power}_{ph} = V_{ph} I_{ph} \cos \phi$$

$$\begin{aligned} \text{Total power} &= 3 \times V_{ph} I_{ph} \cos \phi \\ &= \sqrt{3} \times V_L I_L \cos \phi \end{aligned}$$

Phasor representation :-
 (i) polar
 (ii) rectangular
 (iii) exponential

Ch-12 Complex Numbers

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Phasor representation in rectangular form.

$$V_1 = a_1 + j b_1 \quad \text{or} \quad Z = R \pm j X$$

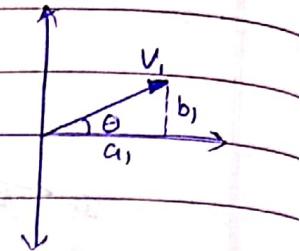
j → operator

a_1 → in-phase (active) component

b_1 → quadrature (reactive) component

$$|V_1| = \sqrt{a_1^2 + b_1^2}$$

$$\text{angle } \phi = \tan^{-1}(b_1/a_1)$$



j is used to indicate the counter-clockwise rotation of a vector through 90° : $j = \sqrt{-1}$

Phasor representation in Exponential form

$$\begin{aligned} \text{Euler's equation: } e^{\pm j\theta} &= \cos\theta \pm j\sin\theta \\ \Rightarrow V &= V e^{\pm j\theta} \end{aligned}$$

It represents a vector \vec{V} of a numerical value V and having a phase angle of $\pm\theta$ with the reference axis.

Phasor representation in polar form

$V \angle \pm\theta \quad (\equiv V e^{\pm j\theta})$ is polar form of Phasor.

Multiplication:

$$\vec{A} = a_1 + j b_1 = A \angle \alpha = A e^{j\alpha}, \quad a = \tan^{-1}(b_1/a_1)$$

$$\vec{B} = a_2 + j b_2 = A \angle \beta = A e^{j\beta}, \quad \beta = \tan^{-1}(b_2/a_2)$$

$$\vec{A} \vec{B} = A \angle \alpha \times B \angle \beta = AB(\alpha + \beta)$$

$$\vec{A} \vec{B} = A e^{j\alpha} \times B e^{j\beta} = A B e^{j(\alpha + \beta)}$$

$$\text{Division: } \frac{A}{B} = \frac{A \angle \alpha}{B \angle \beta} = \frac{A}{B} \angle (\alpha - \beta)$$

$$\frac{A}{B} = \frac{A e^{j\alpha}}{B e^{j\beta}} = \frac{A}{B} e^{j(\alpha - \beta)}$$

Designing aspect of domestic wiring

- I Type of supply
- II Expected load
- III Probable modification
- IV Electrical and mechanical Protection required
- V Cost

Sub Circuits:

- i) Light and fan sub-circuit (low power consumption appliances)
- ii) Power sub circuit [15A] (SA)

Important points about light-fan sub-circuit.

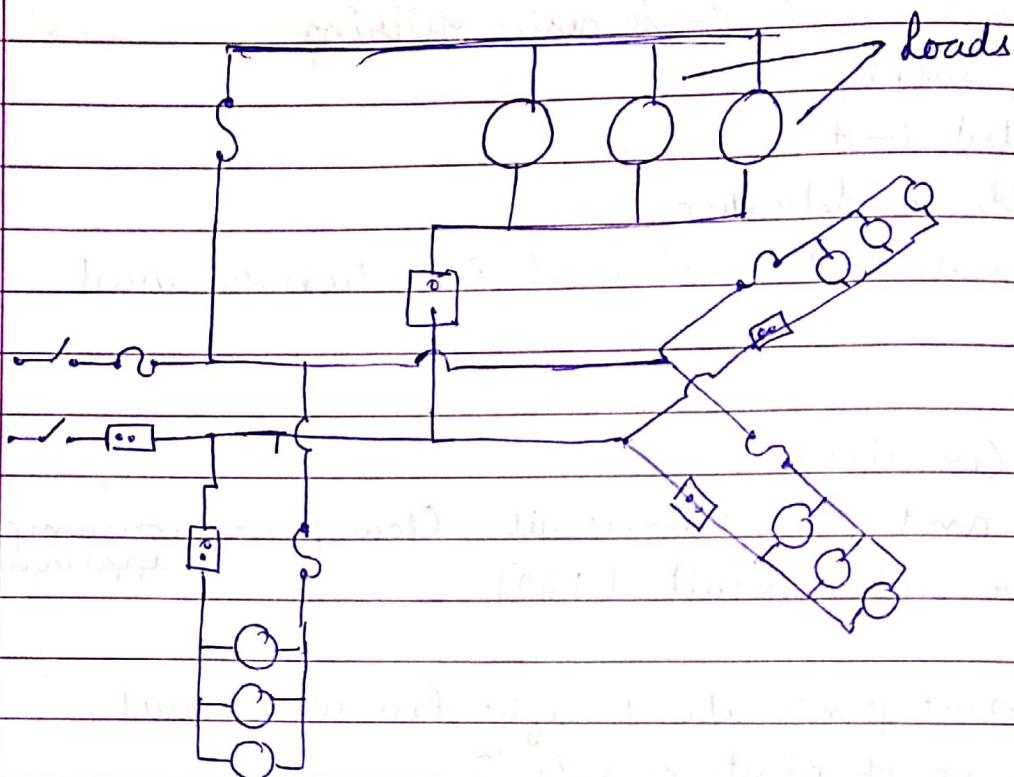
- (i) Max. no. of points can be 10
- (ii) The total load connected in sub-circuit should not be more than 800 W
- (iii) Control of light and fan points in any sub-circuit should be carried out by SA switches fitted on a switch board
- (iv) 1- ϕ supply should be fed to these circuits

Important points about power sub circuit

- (i) Hot plate, electric iron, refrigerator etc.
- (ii) No. of power point connected in one sub-circuit should not exceed than two.
- (iii) The load connected to a sub-circuit should not be greater than 3kW
- (iv) Control of these circuits should be done by 15A switches
- (v) Socket outlets provided in these circuits should be of 15A ratings.
- (vi) These sub circuits must be used on single phase supply.

Domestic Connections

1. Tree System

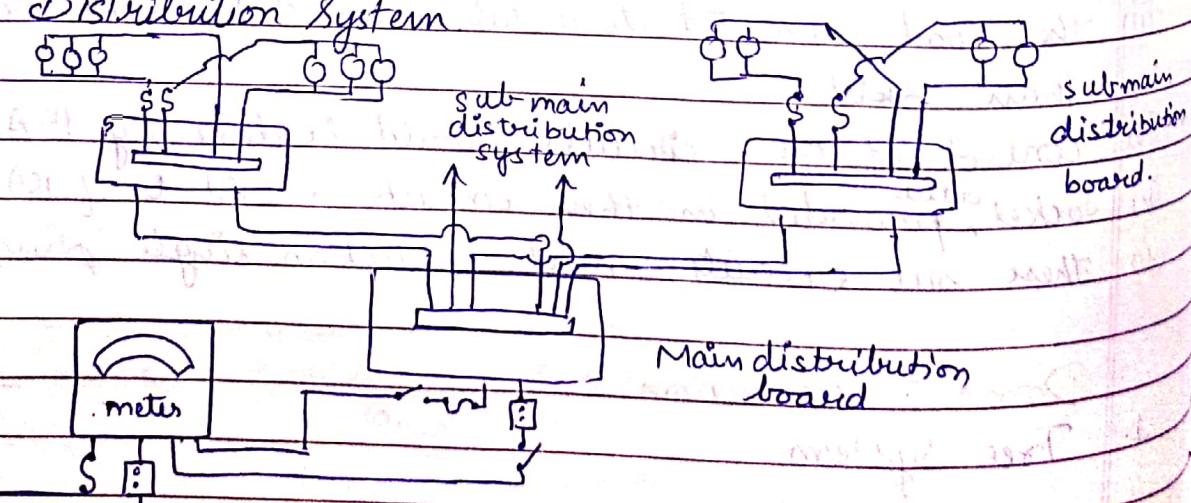


- 1) There are many joints for sub-circuits
- 2) Difficult to find fault
- 3) Voltage may vary
- 4) Appearance is not good.

- 1) Easy to build, modify
 - 2) Less requirement of wire. Hence cost effective.
- But it is not used in permanent connection requirement

II

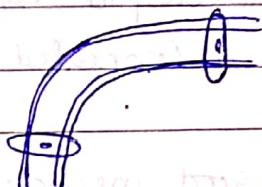
Distribution System



- 1) Less complexity, identification is easy
- 2) More wire required, fuse, board increased \Rightarrow costly
- 3) Voltage will be constant. Voltage division is equivalent in each circuit
- 4) No joints required.
- 5) Fault finding is easy.
- 6) Extension of circuit is easy.

Wiring Type

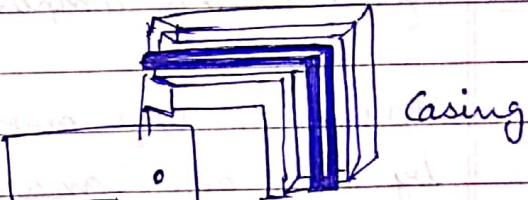
- I Cleat Wiring - e.g. inverter connection
simple PVC wire, easy, cheapest



not good in appearance, mechanical damage may increase, life = 5 yrs. Not applicable for some places where mechanical protection is required.

II Wooden casing and Capping Wiring -

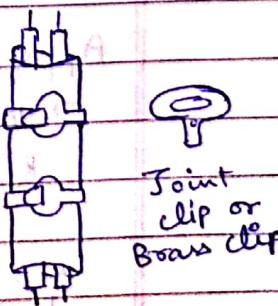
- * Suitable for low voltage.
- * ~~Suitable~~ mechanical protection.
- * ~~Not~~ suitable for damp places, it absorbs moisture
- * It is a bit complex to find screws \rightarrow capping
- * More skill needed. Its erection is not very simple.



III CTS [Cable Type Wiring] & IRS wiring System.

Tough Rubber Sheath

- * Easy installation and is durable. long life
- * Suitable for damp places, outer wiring
- * Cheaper. Give a good appearance
- * Lower risk of short circuit.
- * Suitable for medium voltage installation.



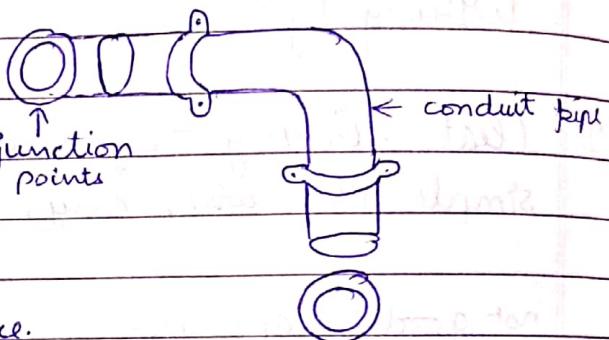
Joint clip or
Board clip

- * May be installed under exposed condition of the rain or dampness.
- * Provide good mechanical protection.
- * Skilled workmen are required.
- * Should not be exposed to direct sunlight.

IV Conduit Wiring

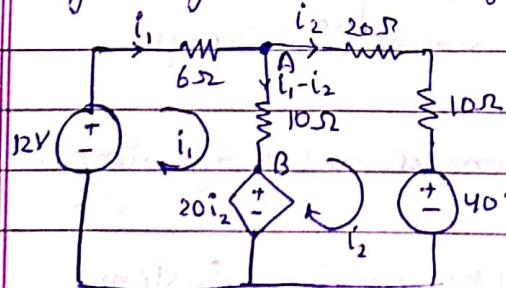
(i) Surface conduit

(ii) Concealed conduit



- * good mechanical protection
- * protection against moisture.
- * Surface conduit is easily repairable.
- * Fitting arrangement is costly. Initial cost is very high.
- * Very neat and attractive appearance in concealed conduit.
- * Requires skilled workmen.
- * Installation is complex and time consuming.

Q. Find the voltage across 10Ω resistor across terminal AB by using mesh analysis.



Applying KVL in mesh 1

$$12V = 6i_1 + 10(i_1 - i_2) + 20i_2$$

$$16i_1 + 10i_2 = 12$$

$$8i_1 + 5i_2 = 6$$

Applying KVL in mesh 2 $\Rightarrow 40 = -10i_2 + 10(i_1 - i_2) + 20i_2 - 20i_1$

$$40 = 10i_1 - 20i_2 - 20i_2 \Rightarrow i_1 - 2i_2 = 4 \Rightarrow i_1 = 4 + 2i_2$$

$$8(4 + 2i_2) + 5i_2 = 6 \Rightarrow 32 + 21i_2 = 6 \Rightarrow i_2 = -\frac{26}{21} = -1.23 A$$

$$i_1 = 4 + 2(-1.23) = 1.52 A$$

$$V_{AB} = 10(1.52 + 1.23) \\ = 10(2.75) = 27.5 V$$

Norton's Theorem

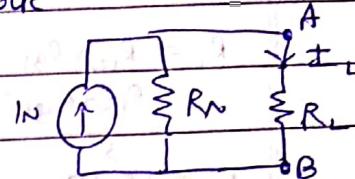
Step 1 - Short the circuit at the load resistance R_L

Step 2 - Determine the current I_{SC} in that shorted branch

Step 3 - Calculate $R_N = R_N$ in the closed network

Step 4 - Make the Norton circuit as

$$\text{Thus, } I_L = \frac{R_N \times I_N}{R_N + R_L}$$



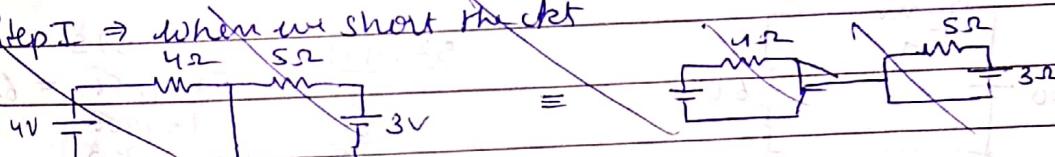
Q. Using Norton's theorem find the current in the branch CD containing 3Ω resistor of the network.

$$\text{Given: } 4V \text{ DC voltage source, } 4\Omega \text{ and } 5\Omega \text{ resistors, } 3V \text{ DC voltage source, } 3\Omega \text{ resistor.}$$

$$E_{eq} = \frac{E_1 r_2 + E_2 r_1}{r_1 + r_2} = \frac{20 + 12}{5+4} = \frac{32}{9} \text{ V}$$

$$R_{eq} = \frac{r_1 r_2}{r_1 + r_2} = \frac{5 \times 4}{5+4} = \frac{20}{9} \Omega$$

Step 1 \Rightarrow When we short the circuit



$$\text{Step 2: } \frac{20/9}{4\Omega} \parallel \frac{20/9}{5\Omega} \parallel \frac{32/9}{3\Omega} \Rightarrow \frac{32}{9} \Omega \text{ in parallel.}$$

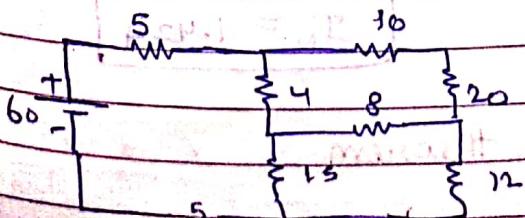
$$I_N = \frac{V}{R} = \frac{32/9}{20/9} = \frac{8}{5} = 1.6 \text{ A}$$

$$R_N = 20/9 \Omega$$

$$\text{Step 4: } I_L = \frac{R_N \times I_N}{R_N + R_L} = \frac{20/9 \times 1.6}{20/9 + 3} = \frac{32/9}{47} = 0.68 \text{ A}$$

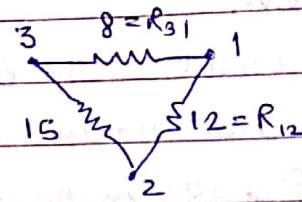
$$V_L = I_L R_L = \frac{32}{47} \times 3\Omega = 2.042 \text{ V}$$

Q. Find the current delivered by the battery.



Consider the delta

Given $R_{23} = 15$



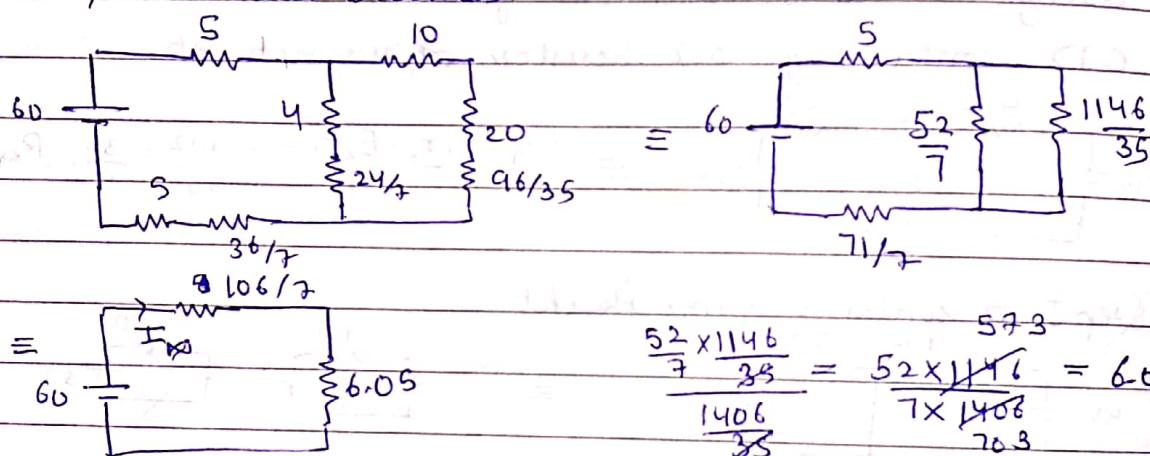
$$R_1 = \frac{R_{12} \times R_{31}}{R_{12} + R_{23} + R_{31}} = \frac{12 \times 8}{12 + 8 + 15} = \frac{96}{35} = 2.74 \Omega$$

$$R_2 = \frac{R_{12} \times R_{23}}{R_{12} + R_{23} + R_{31}} = \frac{12 \times 15}{12 + 8 + 15} = 5.14 \Omega$$

$$R_3 = \frac{R_{23} \times R_{31}}{R_{12} + R_{23} + R_{31}} = \frac{8 \times 15}{12 + 8 + 15} = 3.42 \Omega$$

$I = 2.86 A$

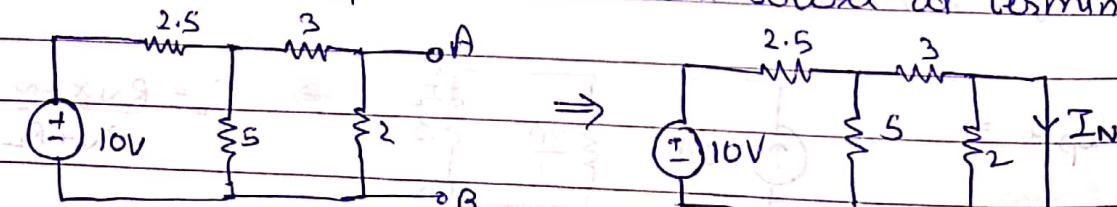
Thus, the circuit becomes



$$\frac{52}{7} \times \frac{1146}{35} = \frac{52 \times 1146}{7 \times 35} = \frac{573}{703} = 6.05 \Omega$$

$$I_1 = \frac{60}{15.14 + 6.05} = \frac{60}{21.19} = 2.83 A$$

Q. Find Norton's equivalent current source at terminal AB.



$$\Rightarrow R_N = 2.5 + \frac{3 \times 5}{3+5} = 2.5 + \frac{15}{8} = 4.37 \Omega$$

$$I_N = \frac{V}{R} = \frac{10}{35/8} = \frac{80}{35} = 2.285$$

$$I_N = \frac{5 \times 2.285}{8} = 1.42 A$$

Maximum Power Transfer theorem

DC source network

$$I_L = \frac{V_{TH}}{R_{TH} + R_L}$$

$$P_L = I_L^2 R_L = \left(\frac{V_{TH}}{R_{TH} + R_L} \right)^2 \times R_L$$

$$P_L = \frac{(V_{TH})^2}{\left(\frac{R_{TH}^2}{R_L} + 2R_{TH} + R_L \right)}$$

$$\text{Let } A = \frac{R_{TH}^2 + 2R_{TH} + R_L}{R_L}$$

For power to be maximum, A should be minimum

If $\frac{dA}{dR_L} = 0$ and $\frac{d^2A}{dR_L^2}$ is positive at a point then A is min.

$$A = \frac{(R_{TH})^2}{R_L} + 2R_{TH} + R_L \Rightarrow \frac{dA}{dR_L} = -\frac{(R_{TH})^2}{(R_L)^2} + 0 + 1 = 0$$

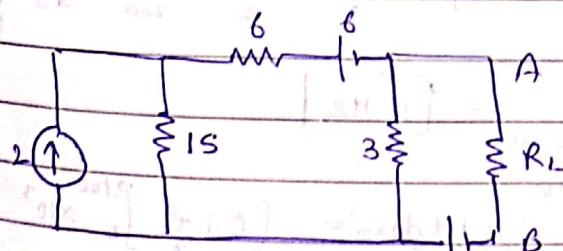
$$\Rightarrow \frac{(R_{TH})^2}{R_L^2} = 1 \Rightarrow (R_{TH})^2 = R_L^2 \Rightarrow R_{TH} = R_L$$

$$\frac{d^2A}{dR_L^2} = \frac{2(R_{TH})^2}{(R_L)^3} = \frac{2(R_{TH})^2}{(R_{TH})^3} = \frac{2}{R_{TH}} > 0$$

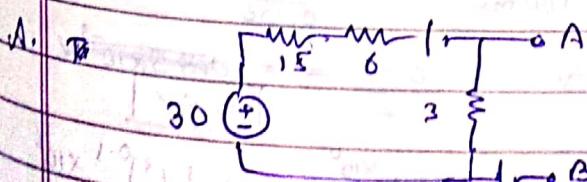
Thus A is minimum at

$$R_{TH} = R_L$$

$$P_{max} = \frac{(V_{TH})^2}{4R_{TH}}$$



Find R_L to have maximum power transfer in the circuit



$$30 - 6 = (15 + 6 + 3)I \Rightarrow I = \frac{24}{24} = 1A$$

$$V_{TH} = 1A \times 3 = 3V$$

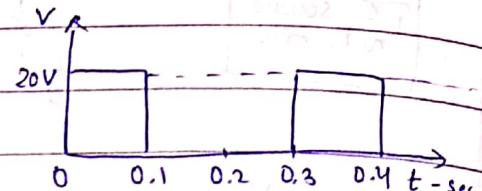
$$R_{TH} = 3 + \frac{21 \times 3}{21 + 3} = 3 + \frac{21 \times 3}{24} = 2.625$$

$$\text{For max. power } R_L = R_{TH} = 2.625 = \frac{21}{8}$$

Q. Compute the average and effective values of the square voltage wave.

$$A. V_{avg} = \frac{\int_0^{0.1} 20 dt}{\int_0^{0.3} dt}$$

$$= \frac{20[t]_0^{0.1}}{[t]_0^{0.3}} = \frac{20 \times 0.1}{0.3} = \frac{20}{3} V = 6.6 V$$



$$V_{rms} = \sqrt{\frac{\int_0^{0.1} V^2 dt}{\int_0^{0.3} dt}} = \sqrt{\frac{\int_0^{0.1} (20)^2 dt}{\int_0^{0.3} dt}} = \sqrt{\frac{400[t]_0^{0.1}}{[t]_0^{0.3}}}$$

$$= 20 \sqrt{\frac{0.1}{0.3}} = \frac{20}{\sqrt{3}} = 11.547 V$$

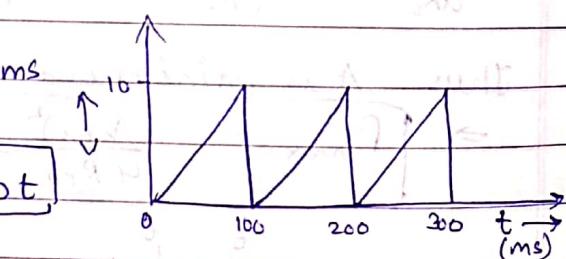
Q. Determine (a) freq. of waveform

(b) waveform eq. for $0 < t < 100$ ms

(c) average value (d) rms value of ~~for~~ & form factor

$$m = \text{slope } m = \frac{10}{100} = 0.1, T = 100 \text{ ms}$$

$$(b) \Rightarrow V = \frac{0.1}{m} t = V(t) \Rightarrow V(t) = 100t$$



$$(a) \text{frequency } f = \frac{1}{T} = \frac{1}{100 \text{ ms}} = 10 \text{ Hz}$$

$$(c) V_{avg} = \frac{1}{T} \int_0^{0.100} V dt = \frac{1}{100 \text{ ms}} \int_0^{0.100} 0.1t dt \times 10^3 = \left[\frac{0.1t^2}{2} \right]_0^{0.100} \times 10^3$$

$$= \frac{0.1}{2} \left(\frac{100 \times 100}{100} \right) = 5 \text{ V} = \frac{0.01}{2} = \frac{0.01}{2} = 5 \times 10^{-3} \text{ V}$$

$$(d) V_{rms} = \sqrt{\frac{1}{T} \int_0^{0.100} V^2 dt} = \sqrt{\frac{1}{100 \text{ ms}} \int_0^{0.100} 0.01t^2 dt} \times 10^6 = \frac{0.01}{0.1} \left[\frac{t^3}{3} \right]_0^{0.1} \times 10^6$$

$$V_{rms}^2 = 0.1 \frac{0.001}{3} = \frac{1}{3} \times 10^{-4} \text{ V} \times 10^6$$

$$V_{rms} = 0.577 \times 10^6 \text{ V}$$

$$V_{rms} = 6.77 \text{ V}$$

$$\text{form factor} = \frac{V_{avg}}{V_{rms}} = \frac{5}{6.77} = 0.74$$

$$\text{form factor} = \frac{V_{\text{rms}}}{V_{\text{avg}}} = \frac{5.77}{5} = 1.154$$

Q. calculate $(5 \angle 30^\circ + 8 \angle -30^\circ)$ V = V when they are in series

$$\begin{aligned} \Rightarrow V &= 5(\cos 30 + j \sin 30) + 8[\cos(-30) + j \sin(-30)] \\ &= 5 \cos 30 + 5j \sin 30 + 8 \cos 30 - 8j \sin 30 \\ &= 13 \cos 30 - 3j \sin 30 \\ &= 13 \times \frac{\sqrt{3}}{2} - 3j \frac{1}{2} = 11.25 - j 1.5 \end{aligned}$$

$$|\vec{V}| = \sqrt{(11.25)^2 + (-1.5)^2} = 11.35$$

$$\phi = \tan^{-1} \left(\frac{-1.5}{11.25} \right) = \tan^{-1}(-0.13) = -7.58$$

$$\Rightarrow V = 11.35 \angle -7.58$$

Ch-6 Magnetism and Electromagnetism

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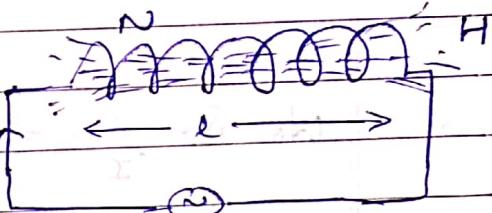
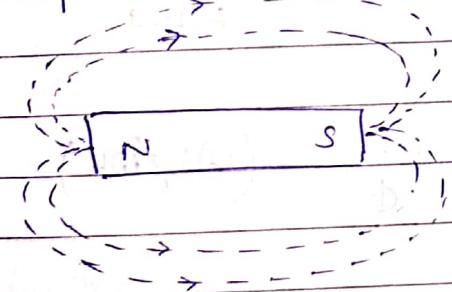
Magnetic circuit concept - It may be defined as the route or path followed by magnetic flux. The law of magnetic circuit is quite similar to electric circuit.

Magnetic flux (ϕ) - The total magnetic field passing perpendicularly through an area. Unit - Wb

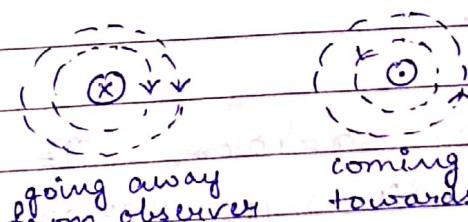
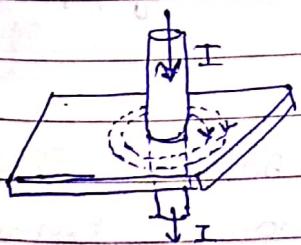
Magnetic flux density (\vec{B}) - It is given by the flux passing per unit area through a plane at right angles to the flux.
$$\vec{B} = \frac{\phi}{A} \text{ (Wbm}^{-2}\text{)} \div (\text{T})$$

Magnetic field strength or magnetic field intensity (H)

$$B = \mu H$$



$$H = \frac{nI}{l} \text{ (TA m}^{-1}\text{)}$$



Permeability (μ) -

$$\text{absolute } \mu_0 = \frac{B}{H}$$

absolute permeability of free space

$$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$$

$$\text{relative permeability, } \mu_r = \frac{\mu}{\mu_0}$$

magnetic motive force (mmf) - Flow of electron is basically

due to emf. Similarly the force behind the flow of flux or the production of flux in magnetic circuit is called magneto motive force or mmf.

$$\text{m.m.f} = N \cdot I = H \times l \quad \text{unit} \rightarrow \text{AT} \text{ (Amper-Turns)}$$

$N \rightarrow$ number of turns

$I \rightarrow$ current

$M^2 L^2 T^2 A^2$ Reluctance (S) or (R) - It is the opposition offered to the flow of flux.

$$S \propto \frac{l}{a} \Rightarrow S = \frac{l}{Ma} \Rightarrow S = \frac{l}{\mu_r H_0 A} \quad (R = \frac{l}{H})$$

$$\text{As } R = \frac{V}{I} \Rightarrow S = \frac{\text{mmf}}{\Phi} = \frac{NI}{\Phi} \quad (\text{AT/Wb})$$

Q. An iron ring of circular cross section area 3 cm^2 and mean diameter of 20 cm is wound with 500 turns of wire. The flux required is 0.5 mWb in the ring. Find the current required if $\mu_r = 833.33$ for iron.

$$A = 3 \text{ cm}^2, r = 10 \text{ cm}, N = 500 \text{ T}$$

$$\Phi = 0.5 \text{ mWb} \quad \mu_r = 833.33 = \frac{2500}{3}$$

$$\therefore S = \frac{l}{a} \Rightarrow \frac{l}{Ma} = \frac{NI}{\Phi}$$

$$\Rightarrow I = \frac{N \Phi}{N \mu_r \mu_0 A} = \frac{0.500 \times 2 \times \pi (10 \text{ cm})}{500 \times 2500 \times 3 \text{ cm}^2 \times 1.27 \times 10^{-6}} \text{ A}$$

$$\text{Required current} = \frac{10^{-3}}{5 \times 10^{-2} \cdot 2} = \frac{10}{50} = 0.2 \text{ A} = 2 \text{ A}$$

Comparison between Magnetic and Electric Circuits.

Similarities :-

Electric :

1. The path traced by the current is called electric circuit
2. Driving force is emf
3. Current flows in the circuit I
4. R resistance
5. $R = \frac{Fl}{a}$
6. $I = \frac{\text{emf}}{R}$

$$7. \text{Current density } J = \frac{I}{A} \text{ (Am}^{-2}\text{)}$$

$$8. \text{Conductance } G = \frac{1}{R} = \sigma \text{ (Siemen)}$$

9. KCL and KVL are applicable

Magnetic :

The path traced by flux is called magnetic circuit

Driving force is mmf

Flux flows in the circuit ϕ

R or S reluctance

$$R \text{ or } S = \frac{l}{\mu a}$$

$$\phi = \frac{\text{mmf}}{R}$$

$$\text{Flux density } B = \frac{\phi}{A} \text{ (Wbm}^{-2}\text{)}$$

$$\text{Permeance } l = \frac{1}{S} = \frac{l}{R} \text{ (Henry)}$$

Kirchoff's mmf and flux law is applicable

Differences :-

Electric

1. Current actually flows.
2. Many materials are electric insulators
3. Energy needs to be supplied to maintain current flow
4. At constant temperature, R and σ are free from current density.
5. Continuous consumption of power

Magnetic

Flux does not actually flow

No material is magnetic insulator.

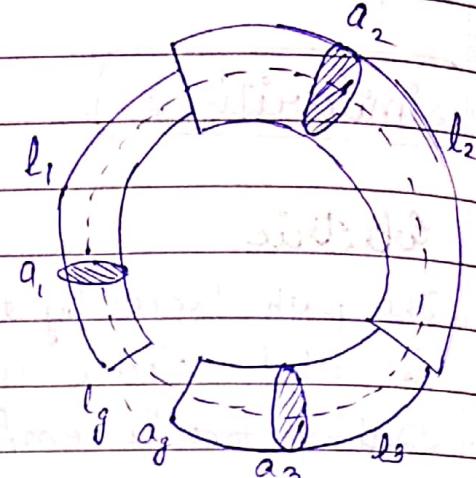
Energy required only to create not to maintain
 R and permeance are dependent on ϕ .

No continuous consumption of power

6.23 Series magnetic circuit

$$F_{\text{tot}} = \frac{l_1}{a_1 \mu_0 H_r} + \frac{l_2}{a_2 \mu_0 H_r} + \frac{l_3}{a_3 \mu_0 H_r} + \frac{l_g}{a_g \mu_0}$$

$$\text{Total mmf} = \phi \times F$$



$$= \phi \left[\frac{l_1}{a_1 \mu_0 H_r} + \frac{l_2}{a_2 \mu_0 H_r} + \frac{l_3}{a_3 \mu_0 H_r} + \frac{l_g}{a_g \mu_0} \right]$$

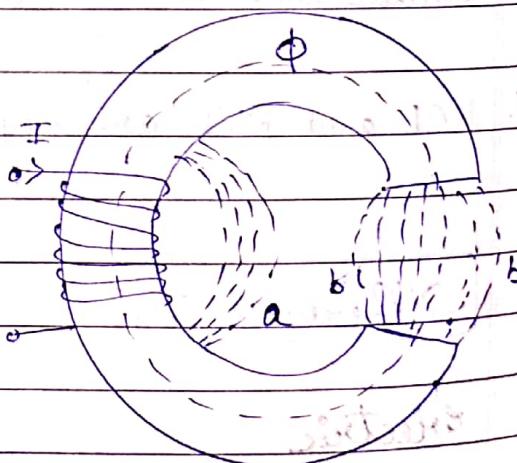
$$= \frac{B_1 l_1}{H_0 M_r} + \frac{B_2 l_2}{H_0 M_r} + \frac{B_3 l_3}{H_0 M_r} + \frac{B_g l_g}{H_0} \quad \left\{ B = \frac{\phi}{a} \right\}$$

$$= (H_1 l_1 + H_2 l_2 + H_3 l_3 + H_g l_g) \text{ AT} \quad \left\{ H = \frac{B}{\mu} \right\}$$

(Core) \rightarrow $B = \mu H$ and $B = \mu_0 H_r$

6.26 Parallel paramagnetic circuit -

Leakage flux - The flux that does not follow the desired path in a magnetic circuit is called leakage flux



$$\phi_{\text{leakage}} = \phi_i - \phi_g$$

$\phi_i \rightarrow$ Total flux produced

$\phi_g \rightarrow$ Useful flux across the air gap

a \rightarrow leakage (leaked flux)

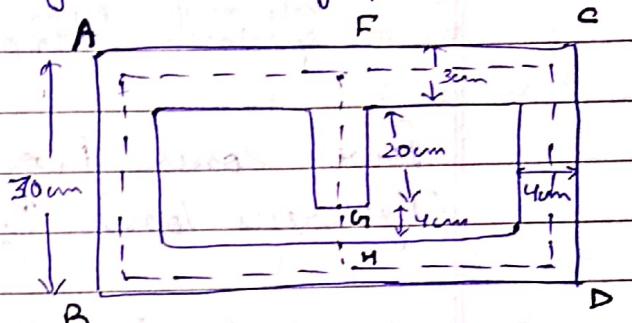
b \rightarrow bringing

A 350 turn coil is wound on FG branch of the cast steel frame as shown in fig. A total flux of 1.2×10^{-3} Wb is required in the gap. Find what current is required. Assume that the gap density is uniform & that all flux lines pass straight across gap.

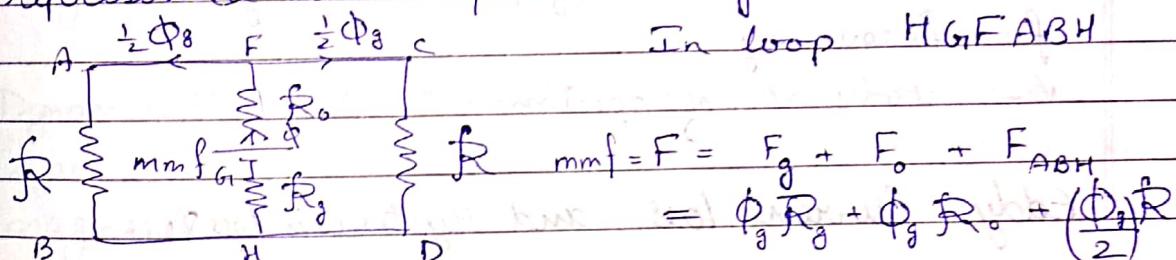
$$N = 350 \text{ turns}$$

$$\Phi = 1.2 \times 10^{-3} \text{ Wb}$$

$$I = ?$$



The given frame can be represented as the equivalent magnetic circuit.



$$= \Phi_g \left[\frac{l_g}{\mu_0 A_g} + \frac{l_o}{\mu_0 A_o} + \frac{l}{2\mu_0 A} \right]$$

for Iron

(99.8%)

$M_r = 5000$

$$\text{flux in air gap } \Phi_g = 1.2 \times 10^{-3} \text{ Wb}$$

$$\text{length of air gap } = 4 \times 10^{-2} \text{ m}$$

$$F = 24306$$

$$\text{Area of air gap } = 16 \times 10^{-4} \text{ m}^2$$

$$I = 69.45 \text{ A}$$

$$= \frac{\Phi_g}{4\pi \times 10^{-7}} \left(\frac{1}{4H_r} + \frac{1}{4M_r} + \frac{1}{8M_r} \right) \text{ cm}^{-1}$$

$$= \frac{\Phi_g}{4\pi \times 10^{-7}} \left[\frac{3}{4H_r} + \frac{1}{4} \right] \text{ cm}^{-1}$$

$$= \frac{1.2 \times 10^{-3}}{4\pi \times 4\pi \times 10^{-7}} \left[\frac{3}{4H_r} + \frac{1}{4} \right] \times 100 \text{ cm}^{-1} = \frac{3 \times 10^5}{4\pi} \left[\frac{3}{4H_r} + \frac{1}{4} \right] \text{ AT}$$

$$A_{AT} =$$

Ch-7 Electromagnetic Induction

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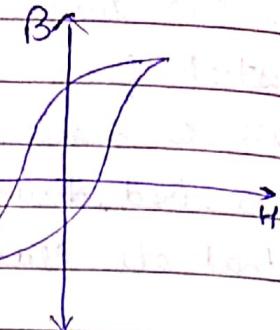
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B-H Curve.

Energy consumed,

$$W = \underbrace{a \cdot l}_{\text{volume of specimen}} \cdot \underbrace{H \cdot dB}_{\text{area of elementary strip in B-H curve}}$$

area of elementary strip in B-H curve



Power losses $P_h \propto B_{\max}^{1.6} f V$
(Hysteresis loss) $P_h = \eta B_{\max}^{1.6} f V$

$\eta \rightarrow$ hysteresis coefficient

$B_{\max} \rightarrow$ peak value of flux density

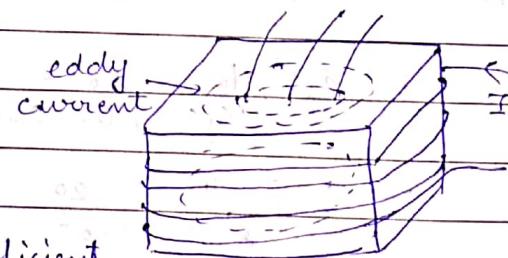
$f \rightarrow$ frequency

$V \rightarrow$ volume of specimen

Eddy Current loss and Hysteresis loss P_{Eddy} .

Eddy Current

$$\text{losses} = K_e (B_m)^2 f^2 t^2 V$$



$K_e \rightarrow$ Eddy current coefficient

$B_m \rightarrow$ Magnetic flux density peak

$f \rightarrow$ frequency

$t \rightarrow$ thickness of material

$V \rightarrow$ potential applied

To reduce it, we can use laminated material

Eddy current losses arise from the fact that the core itself is composed of conducting material, so that the voltage induced in it by the varying flux produces circulating currents in the material.

These are called eddy currents and are accompanied by ' i^2R ' loss in the core called eddy current loss

while Hysteresis loss is associated with the phenomena of hysteresis and is an expression of the fact when ferromagnetic materials involved, not all the energy of the magnetic field is returned to the circuit when mmf is reduced.

7.8 Static induced emf - The induced emf which is not dependent on the motion of the conductor or the magnetic flux is called static induced emf

7.7 Dynamically induced emf - The induced emf which is dependent on the motion of the conductor or the magnetic flux is called dynamic induced emf

e.g. Dynamo.

7.9 Self Inductance - $L = \frac{N_1 \phi_1}{I_1}$, $\phi = \text{mmf} = \frac{NI}{R}$

$$\Rightarrow L = \frac{N_1 N_1 I}{R} = \frac{N^2}{R} = (H) = \frac{\mu A N^2}{l} \Rightarrow L = \frac{N^2 \mu A}{l}$$

7.11 Mutual Inductance - When the magnetic field generated by a coil of wire induces voltage in an adjacent coil of wire, it is called mutual inductance

$$M = \frac{N_1 \phi_2}{I_2} = \frac{N_1 N_2 \mu A}{l}$$

$$1 \text{ Henry} = 1 \text{ Weber} \cdot \text{Turns} \quad L(H) = \frac{N \phi}{I} \quad 1 \text{ Weber} = 1 \text{ Ampere} \cdot \text{Turns}$$

Mutually induced emf

$$e_m = -N_1 \frac{d\phi_2}{dt} = -N_2 \frac{d\phi_1}{dt}$$

7.13 Coupling Coefficient

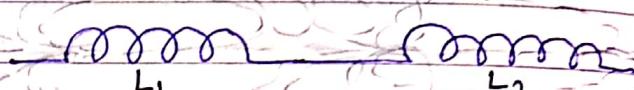
$$\Phi_{12} = k\phi_1$$

Flux

$$\Phi_{21} = k\phi_2$$

$$\phi_1$$

$$\phi_2$$



Flux linking

$$\phi_1 = \frac{N_1 i_1}{R} \quad \phi_2 = \frac{N_2 i_2}{R}$$

$$M_{12} = \frac{N_2 \phi_{12}}{i_1} = \frac{N_2 k \phi_1}{i_1} = \frac{k N_1 N_2}{R}$$

$$M_{21} = \frac{N_1 \phi_{21}}{i_2} = \frac{N_1 k \phi_2}{i_2} = \frac{N_1 N_2 k}{R} = M_{12}$$

$$\text{Self Inductance } L = \frac{N \phi}{i} = \frac{N^2 \phi}{N i} = \frac{N^2 B A}{H l} = \frac{N^2}{R}$$

$$L_1 = \frac{N_1^2}{R}, \quad L_2 = \frac{N_2^2}{R}$$

$$\text{then } L_1 L_2 = \frac{N_1^2 N_2^2}{R^2} \Rightarrow \frac{N_1 N_2}{R} = \sqrt{L_1 L_2}$$

$$\Rightarrow M = k \sqrt{L_1 L_2}$$

- Q. Two inductances of 15 mH and 33 mH are connected in series such that their flux oppose each other. The value of coefficient of coupling is 0.7. Calculate mutual inductance between two coils.

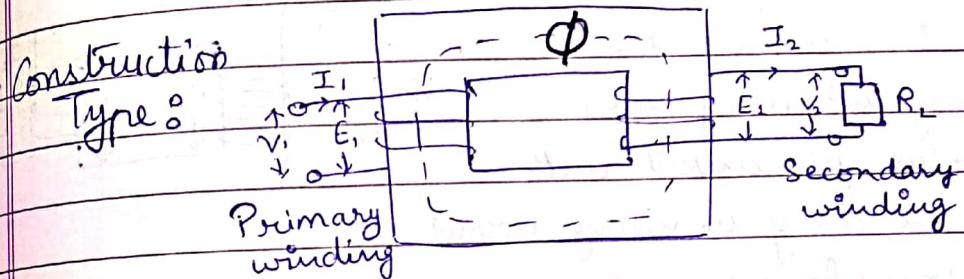
Chapter -32 Transformer

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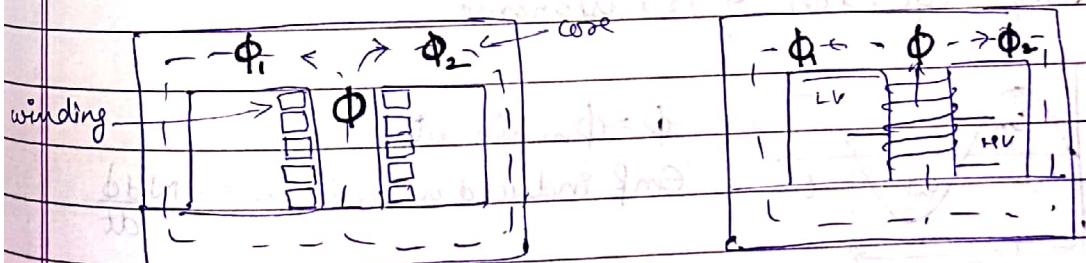
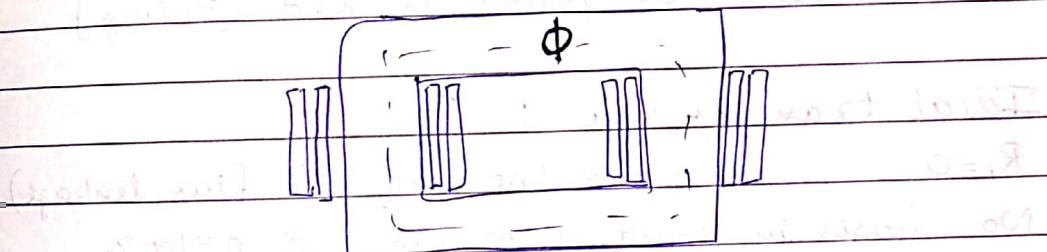
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$$M = k\sqrt{L_1 L_2} = 0.7 \sqrt{15 \times 35} = 0.7 \sqrt{3 \times 7 \times 3 \times 5} \\ = 16.039 \text{ mH} = 16.04 \text{ mH}$$

Transformer - It is a static device which converts one form of the energy to the other form.



1- ϕ core type transformer



1- ϕ Shell type transformer

Shell type transformer is more protected than the core type transformer. But it has a large heating effect and thus must be cooled by external agent.

Supply type: is 1- ϕ transformer

(ii) 3- ϕ transformer

Rating type:

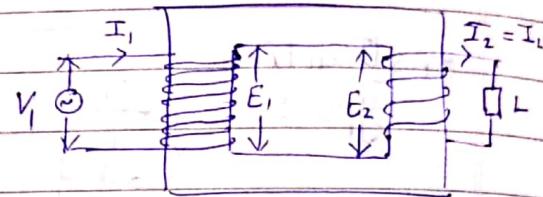
- (i) Power transformer - larger
- (ii) ^{Distribution} ~~Supply~~ transformer

Working of core type transformer - When supply is provided (V_1), current I_1 starts flowing.

The alternating current

generates flux in the primary winding core.

This flux gets linked with the secondary winding through the transformer core.

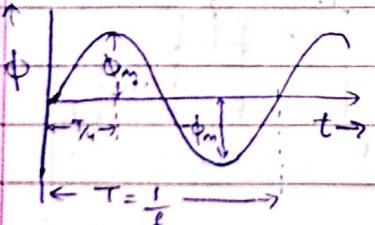


It represents its power as KVA [rating]

Ideal transformer :

- (i) $R_s = 0$
- (ii) $\mu = \infty$ (no magnetic flux leakage)
- (iii) No losses in winding or core. $\Rightarrow \eta = 100\%$.

32.6 Emf equation of transformer



$$\phi = \phi_m \sin \omega t$$

Emf induced in coil $e = -N \frac{d\phi}{dt}$

$$e = -N \frac{d}{dt} (\phi_m \sin \omega t) = -N \phi_m \omega \cos \omega t$$

$$e = N \omega \phi_m \sin (\omega t - \frac{\pi}{2})$$

$$e = E_m \sin(\omega t - \frac{\pi}{2}) \quad E_m = N \omega \phi_m = \text{max. emf}$$

$$E_{rms} = \frac{E_m}{\sqrt{2}} = \frac{N \omega \phi_m}{\sqrt{2}} = N \left(\frac{\pi f}{2} \right) \phi_m$$

$$E_{rms} = \sqrt{2} N \pi f \phi_m \Rightarrow E_{rms} = 4.44 f \phi_m N$$

This is called emf equation of transformer

$$\rightarrow E_1 = 4.44 f \phi N_1$$

$$E_2 = 4.44 f \phi N_2$$

Transformation ratio (K)

Voltage transformation ratio $K = \frac{E_2}{E_1} = \frac{N_2}{N_1} = \frac{I_1}{I_2}$

$$\frac{E_1}{E_2} = \frac{N_1}{N_2} = \frac{I_2}{I_1}$$

- i) A 50 kVA, single phase transformer has 600 turns in primary and 40 turns in secondary. The primary winding is connected to 22 kV, 50 Hz supply. Determine
- (i) E_2 , (ii) I_1 & I_2 at full load.

ii) $\frac{E_1}{E_2} = \frac{N_1}{N_2} \Rightarrow E_2 = \frac{E_1 N_2}{N_1} = \frac{2.2 \text{kV}}{15(600)} = \frac{2200 \text{ V}}{15} = 146.67 \text{ V}$

iii) $P = V_1 I_1 \Rightarrow I_1 = \frac{P}{V_1} = \frac{50 \text{kVA}}{2.2 \text{kV}} = \frac{500}{22} \text{ A} = 22.72 \text{ A}$

$$I_2 = \frac{P_{\text{full load}}}{V_2} = \frac{50 \text{kVA} \times 15}{2.2 \text{kV}} = \frac{500 \times 15}{22} \text{ A} = 340.9 \text{ A} = 341 \text{ A}$$

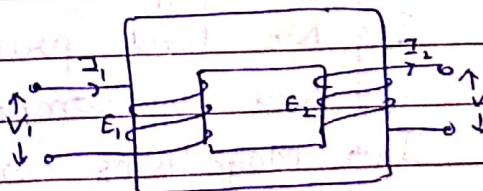
Phasor diagram of Ideal transformer

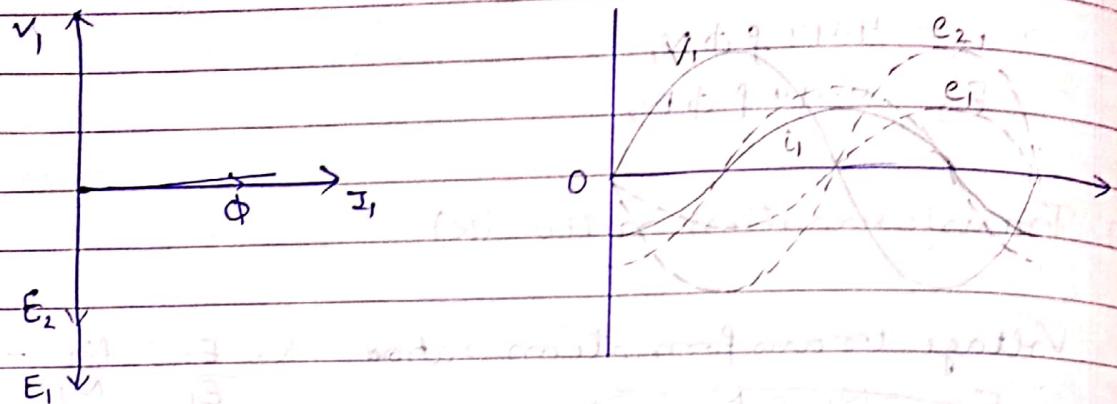
Ideal $\rightarrow R=0, x=0 \rightarrow$ pure inductor

$$V_1 \rightarrow L_1, R_1 = 0$$

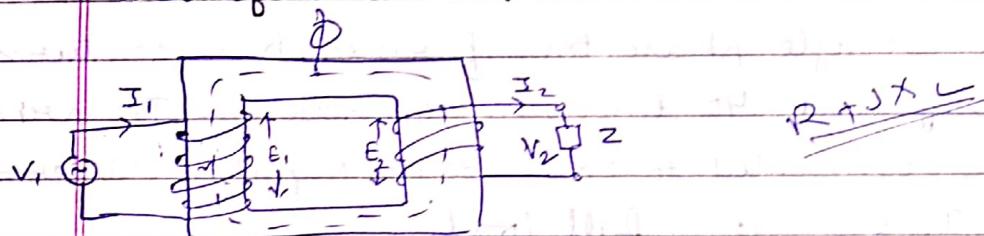
$$\phi = \frac{N_1 I_1}{f R}$$

$$E_1 = -N_1 \frac{d\phi}{dt}, E_2 = -N_2 \frac{d\phi}{dt}$$

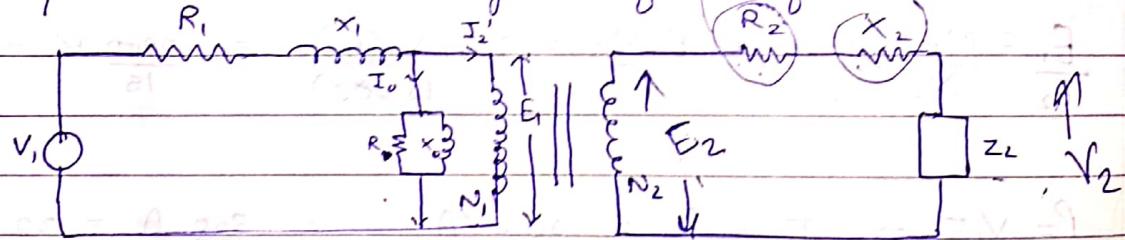




32.10 Transformer On Load



Equivalent circuit diagram of transformer



$$\text{At load, } I_1 = I_0 + I_2'$$

R → primary winding resistance

N → primary turns

V → primary / supplied voltage.

I → current in primary winding

X → primary winding leakage reactance

I_0 → No load primary current

I_0 → Core loss component of no load primary current

I_m → Magnetising component of I_1

R → secondary winding resistance

V_2 → secondary voltage

R_L → Resistance

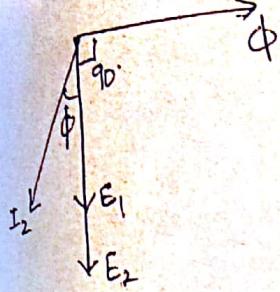
E_1 → primary induced emf

E_2 → secondary induced emf

X_2 → secondary leakage reactance / magnetizing reactance

I'_1 → component of I_1 to nullify effect of ϕ_2

Phasor diagram of 1- ϕ transformer on load

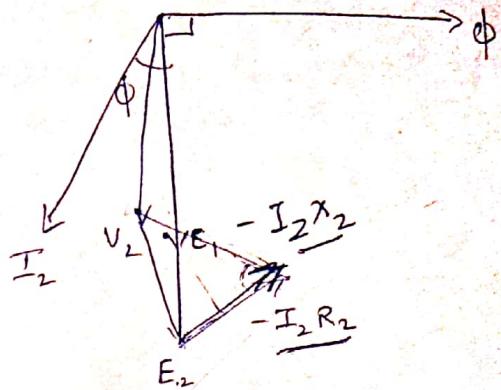


$$\vec{I}_1 = \vec{I}_o + \vec{I}'_1$$

$$\vec{I}'_1 = \vec{I}_c + \vec{I}'_\mu$$

$$V_2 = E_2 - I_2 R_2 - I_2 X_2$$

$$V_1 = -E_1 + I_1 R_1 + I_1 X_1$$



losses

losses in transformer

1. Cu loss → These losses occur in primary winding and secondary winding.
2. Core loss → flux leakage

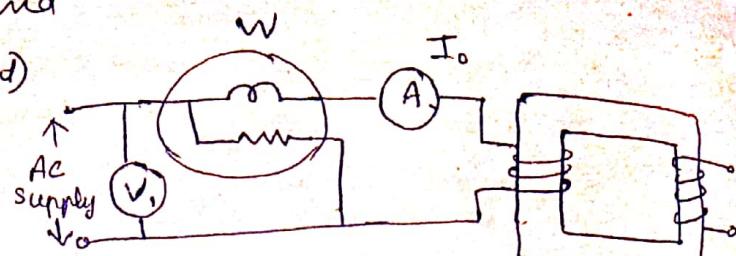
32.20

Open circuit test - If one winding is open, it becomes open circuit test and

$$I_r = I_o \quad (\text{5% of total current})$$

$$\text{Total loss} = \text{Iron} + \text{Cu loss}$$

Value of I_o is very small.
So Cu loss will be small.



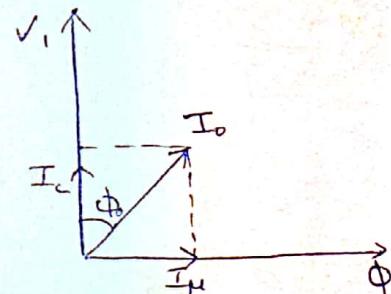
Primary winding Cu loss = $I_1^2 R_1 + I_2^2 R_2$ which will be very small. So the total losses in open circuit test will be equal to iron core losses.

- (a) To know iron losses
- (b) H.V. winding kept open.
- (c) L.V. is connected with supply voltmeter, ammeter and wattmeter. $W = V_i I_o \cos \phi$

$$\text{Iron losses} = W - I_o^2 R_o$$

$W \rightarrow$ wattmeter reading (active power)

$$\cos \phi_o = \frac{W}{V_i I_o} \quad R_o = \frac{V_i}{I_o} = \frac{V_i}{I_o \cos \phi_o}$$



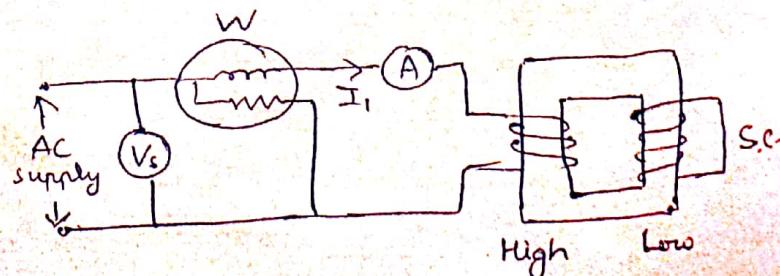
$$\text{Resistive component } I_c = I_o \cos \phi$$

$$\text{Magnetising component } I_\mu = I_o \sin \phi$$

32.22

Short circuit test - To know Cu losses [Impedance test]

- (a) Secondary winding is short circuited
- (b) H.V. is connected with supply, voltmeter, ammeter at wattmeter
- (c) L.V. is short circuited. It is never reversed.



$$Z_{o1} = \frac{V_{sc}}{I_1}$$

$$W = I_1^2 R_{o1}$$

$$R_{o1} = \frac{W}{I_1^2}$$

$$X_{o1} = \sqrt{Z_{o1}^2 - R_{o1}^2}$$

$Z_{o1} \rightarrow$ combined impedance of primary & secondary winding over short circuit.

Q. A 330/660 , 8 kVA , 50 Hz 1- ϕ transformer has full load open copper loss of 90 W. If it has an efficiency of 98% at full load unity power factor. Determine the core losses. What would be the efficiency of transformer at half load . 0.85 pf lagging?

~~$$A. K_F = \frac{330}{660} = 0.5$$~~

~~$$S = 8 \text{ kVA}$$~~

~~$$\text{Active power} = 8 \text{ kVA} \times \cos\phi = 8 \text{ kW}$$~~

~~$$\text{Cu loss } W = 90 \text{ W}$$~~

~~$$\eta = 98\%$$~~

~~$$\cos\phi = 1$$~~

~~$$\text{Core losses} = W - I^2 R_o$$~~

~~$$= 2\% \text{ of } 8 \text{ kW} - 90 \text{ W} =$$~~

~~$$\frac{160}{8000} \times 8000 - 90 = 70 \text{ W}$$~~

At half load, Power = ~~4 kVA~~

~~$$Q. P \quad \cos\phi_0 = \frac{W}{V \cdot I_0} = 0.85 = \frac{W}{4 \text{ kW}}$$~~

~~$$\Rightarrow W = 0.85 \times 4000 = 3400$$~~

~~$$\eta = 1 - \frac{W}{4 \text{ kW}} = 1 - \frac{3.4}{4} = 15\%$$~~

When the load value is changed
only current value gets affected

\Rightarrow At half load,
 $I_{\text{half}} = \frac{I}{2}$

~~$$A. (a) \eta = \frac{\text{O/P}}{\text{I/P}} \Rightarrow \text{I/P} = \frac{\text{O/P}}{\eta} = \frac{8 \text{ kVA} \times \cos\phi}{98\%}$$~~

~~$$\text{I/P} = \frac{8 \text{ kW} \times 1}{0.98} = 8.16 \text{ kW}$$~~

~~$$\text{Total losses} = \text{I/P} - \text{O/P} = 8.16 \text{ kW} - 8 \text{ kW} = 160 \text{ W}$$~~

~~$$\text{Iron losses} = \text{Total losses} - \text{Cu losses} = 160 - 90 - 70 \text{ W}$$~~

~~$$(b) \text{O/P} = 8 \text{ kVA} \times \cos\phi = 6.8 \text{ kW}$$~~

~~$$\text{at half load, O/P} = \frac{1}{2} (6.8 \text{ kW}) = 3.4 \text{ kW}$$~~

Iron losses are not affected by current change. So

~~$$\text{Iron losses} = 70 \text{ W at half load}$$~~

~~$$\text{Copper losses in half load} = \left(\frac{I}{2}\right)^2 \times \text{Cu loss at full load}$$~~

$$= \frac{90}{4} =$$

$$\eta = \frac{\text{output}}{\text{output + losses}} = \frac{3.4 \text{ kW}}{3.4 \text{ kW} + 92.5 \text{ W}} = \frac{3400}{3492.5} = 0.9735$$

$= 97.35\%$

Chapter-10 Electrical Instruments and Measurements

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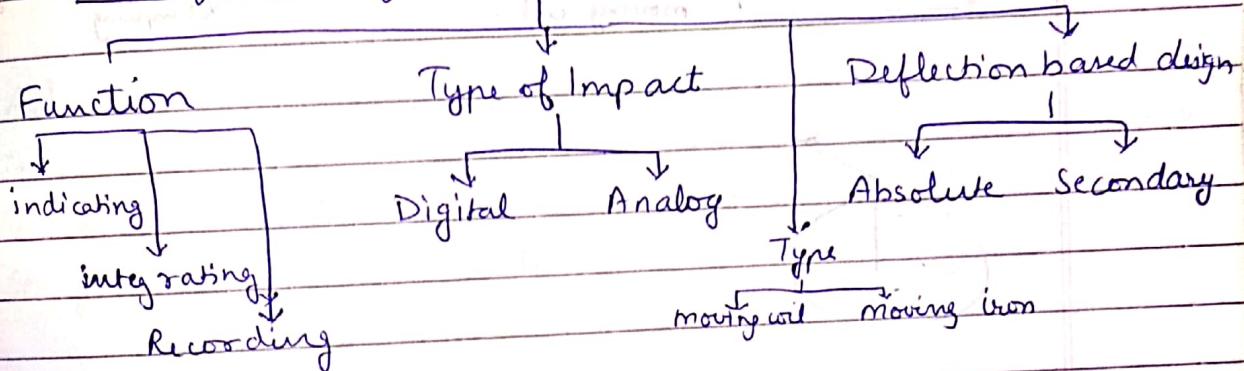
Date: 21/3/2018

Measurement - It is a process to know unknown quantity in terms of known standards.

Measuring Instrument - voltmeter, ammeter, wattmeter etc.

Instrumentation - The process which we undergo to take the measurement

Classification of measuring Instrument



indicating - ammeter, wattmeter, multimeter

integrating - summation, energy meter, odometer

Recording - CRO, ECG

Effect	Instrument utilize the effect
Magnetic	Voltmeter, Ammeter and Wattmeter
Heating	AC and DC ammeters & voltmeter
EMI	I, V, W
Electrostatic	Voltmeter, indirect application in ammeter and wattmeter.

10.3 Analog instruments

Different torque

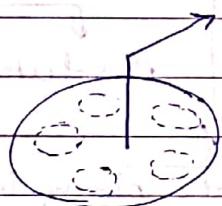
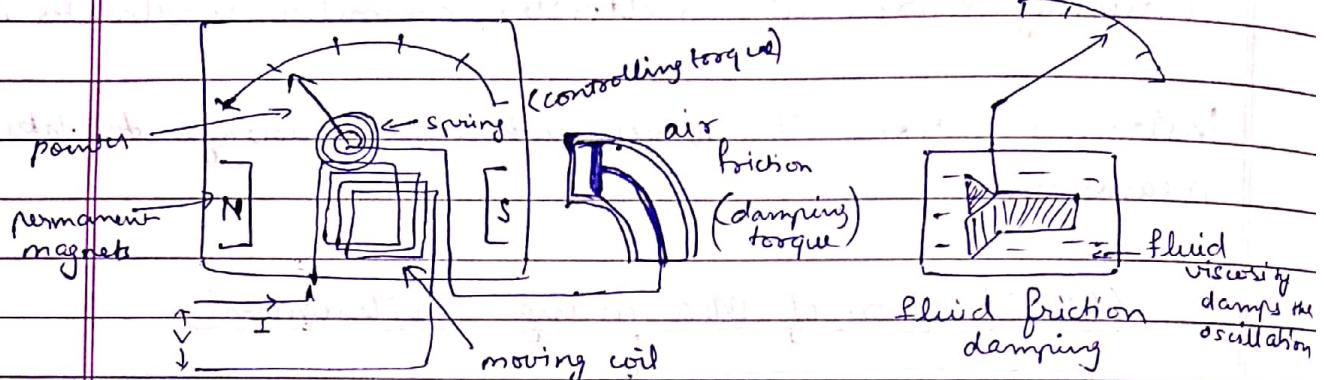
Deflection Torque \propto unknown quantity (V, I or other)

10.5

2) Controlling torque - spring, weight (gravity)

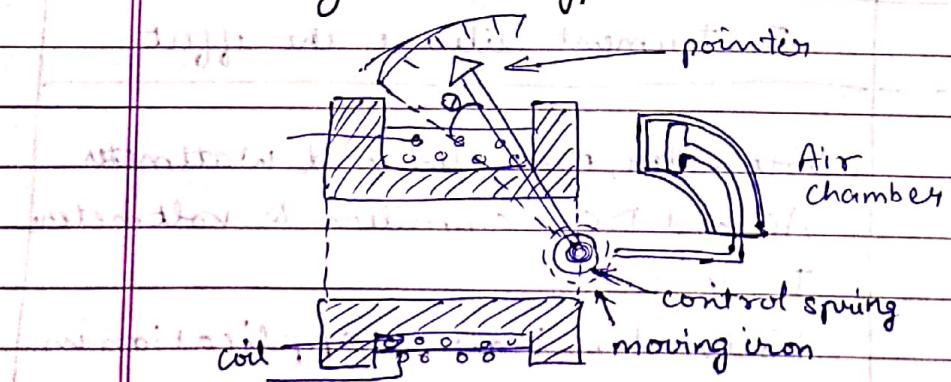
10.6 3) Damping torque - stability to pointer fluctuation / oscillation

At the right point $T_d = T_c$



eddy current damping
(mechanical damping)

10.7 Moving iron type



Attraction type moving iron.

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F \propto H

Now max

$m = ia$

$H \propto i$

 $m \rightarrow$ developed pole strength. $H \rightarrow$ field produced $F \rightarrow$ force pulling soft iron piece

$\Rightarrow F \propto i^2$

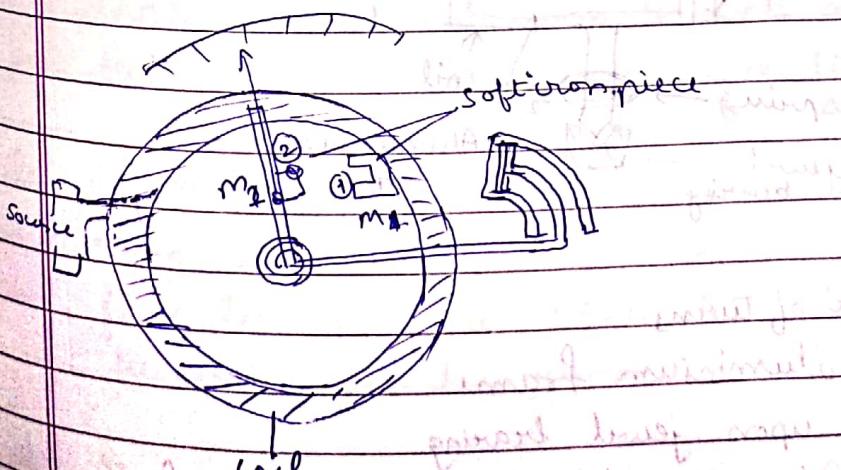
 \Rightarrow Deflection torque $T_d = i^2$ Controlling torque $T_c = \Theta$ [spring control.]At equilibrium, $T_d = T_c \Rightarrow \Theta = i^2$ Since $i = \sqrt{\Theta}$ \Rightarrow The scale is not linear.

Advantages -

1. It is less expensive, robust, and simple in construction
2. It can be used to measure AC and DC both
3. It gives torque/weight ratio

Disadvantage -

1. Non-uniform scale
2. Not sensitive as PMMC instrument.
3. Affected by stray magnetic field
4. Degree of precision of calibration is not high.



Repulsion type moving iron instrument

$$T_d \propto m_1 m_2 \propto i^2$$

$$T_d = i^2, \quad T_c = 0$$

$$\Rightarrow \text{At equilibrium} \quad \theta = i^2$$

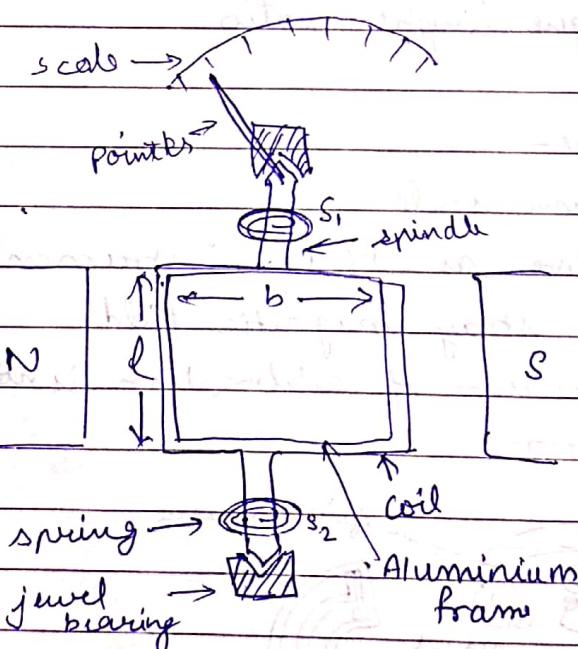
Advantages -

1. It is less expensive, simple construction
2. Can be used for AC and DC both
- 3.

Disadvantages

1. Non uniform scale
2. Less sensitive than PMMC type

10.15 Moving Coil type.



1. Coil has no. of turns
2. Rectangular aluminium frame
3. Coil is pivoted upon jewel bearing
4. Eddy current damping is used by aluminium frame

$B \rightarrow$ flux density in air gap (τ)

$N \rightarrow$ no. of turns in the coil $i \rightarrow$ current in the coil $l \rightarrow$ length of the coil $b \rightarrow$ breadth of the coil $A \rightarrow$ Area ($= l \times b$) $k_s \rightarrow$ spring constant $\theta \rightarrow$ deflection

$$F = N B i l$$

 $T_d \propto F \propto$ perpendicular distance (Nm)

$$T_d = N_i B l \times b = N_i B A = k_s i \quad k_s = N B A$$

$$T_c = k_s \theta$$

$$\text{At equilibrium} \quad T_d = T_c \Rightarrow k_s i = k_s \theta \Rightarrow \theta = k_i$$

Advantages

- (i) reduced size
- (ii) linear scale
- (iii) more sensitivity
- (iv) less power consumption
- (v) lesser losses
- (vi) very accurate & reliable.
- (vii) electrical clamping is advantageous

Disadvantage

- (i) can not measure AC

Application - DC Ammeter, voltmeter, galvanometer, ballistic galvanometer, multimeter

Q. A PMMC instrument has 40 turn coil with Al frame 18mm wide & 25mm effective length. It moves in 0.6T flux density. $k_s = 2.5 \times 10^{-6}$ Nm/degree. Calculate the current required to produce a deflection of 110° degrees

A. At equilibrium $T_d = T_c \Rightarrow k_s \theta = B i N A$

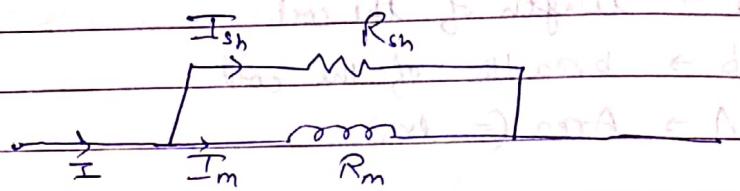
$$i = \frac{k_s \theta}{B N A} = \frac{2.5 \times 10^{-6} \text{ Nm/degree} \times 110^\circ \text{ degree}}{0.6 \text{ T} \times 40 \times 18 \times 25 \times 10^{-4} \text{ m}^2}$$

$$= 0.02 \text{ A}$$

10.17

Range extension:

Shunt - Range extension of ammeter

 $R_m \rightarrow$ resistance of meter $R_{sh} \rightarrow$ Resistance of shunt $I_m \rightarrow$ Current through meter $I_{sh} \rightarrow$ Current through shunt
 $I \rightarrow$ Current to be measured.

A low shunt resistance connected in parallel with the ammeter to extend the range of current. Large current can be measured using a low current rated ammeter by using a shunt.

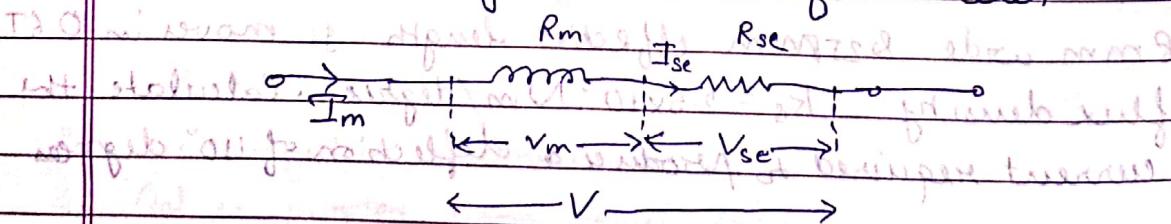
$$\text{Now } V_m = V_{sh} \Rightarrow I_m R_m = I_{sh} R_{sh}$$

$$\frac{I_m}{I_{sh}} = \frac{R_{sh}}{R_m}$$

$$I = I_m + I_{sh} \Rightarrow \frac{I}{I_m} = 1 + \frac{I_{sh}}{I_m}$$

$$\frac{I}{I_m} = 1 + \frac{R_m}{R_{sh}} \Rightarrow I = I_m (1 + \frac{R_m}{R_{sh}})$$

Multiplier - Range extension of voltmeter



$$I_m = A I_{se} \Rightarrow V_m = \frac{V_{se}}{R_m} A$$

$$V = V_m + V_{se} \Rightarrow \frac{V}{V_m} = 1 + \frac{V_{se}}{V_m} = 1 + \frac{R_{se}}{R_m}$$

$$\frac{V}{V_m} = 1 + \frac{R_{se}}{R_m}$$

$$\Rightarrow V = V_m \left(1 + \frac{R_{se}}{R_m} \right)$$

- Q. A moving coil instrument has a resistance of 75Ω and gives full scale deflection when carrying a current of 35 mA . Show how it can be adopted to measure
- current upto 80 A
 - voltage upto 550 V .

A. (i) To extend the range of current, we will use a shunt in parallel to the instrument. For, this

$$I = I_m \left(1 + \frac{R_m}{R_{sh}} \right)$$

$$\Rightarrow 80 = 35 \left(1 + \frac{75}{R_{sh}} \right)$$

$$\Rightarrow \frac{80000}{35} = 1 + \frac{75}{R_{sh}} \Rightarrow \frac{75}{R_{sh}} = \frac{80000 - 1}{35} = 2284.714$$

$$\Rightarrow R_{sh} = \frac{75}{2284.714} = 0.03287 \Omega = 32.87 \text{ m}\Omega$$

(ii) To extend the range of voltage, we will use a multiplier in series with the instrument.

$$V = V_m \left(1 + \frac{R_{se}}{R_m} \right)$$

$$550 = (35) \left(\frac{75}{1} \right) \left[1 + \frac{R_{se}}{75} \right]$$

$$\left(1 + \frac{R_{se}}{75} \right) = \frac{22000}{35 \times 75} = 0.2093 = 209.322$$

$$R_{se} = 208.523$$

$$R_{se} = 15639.22 \Omega$$

Measurement of power

DC power $P_{dc} = V_L I_L$ (in Amperes A)

AC power $P_{ac} = V_L I_L \cos\phi$ (in Ω^{-1})

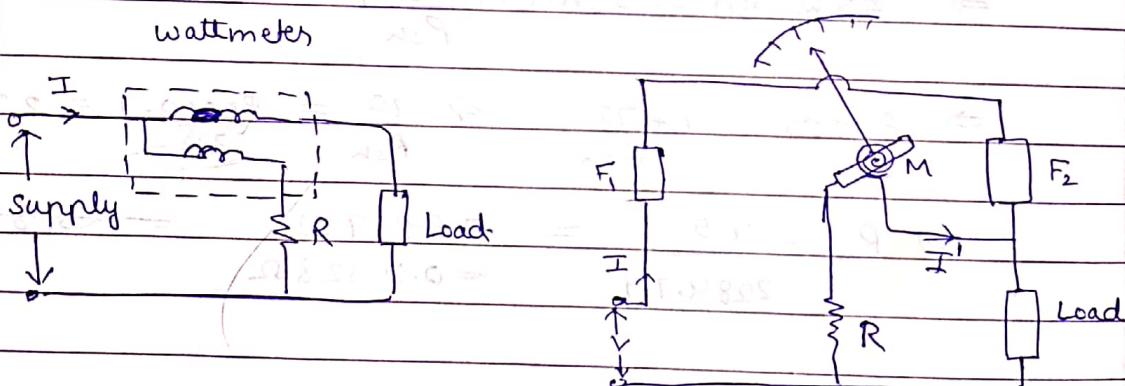
Two types of wattmeter -

- Dynamometer wattmeter \rightarrow a.c. & d.c. power
- Induction type wattmeter \rightarrow AC power.

10.37) Dynamometer wattmeter

Moving coil dynamometer and voltmeter

\Rightarrow mechanical force exists between two current carrying conductor or coil



EMMC \rightarrow electro magnetic moving coil

$F_1, F_2 \rightarrow$ Air core - fixed coils

$M \rightarrow$ moving coil

F_1, F_2 produce field & M moves & provides deflection Torque. Spring provides controlling Torque. Air friction provides damping torque

DC measurement -

$I \rightarrow$ Current coil or fixed coil $\rightarrow B$ (Wb/m^2)

$I' \rightarrow$ current through pressure coil

$$V \propto I' \quad \text{O.S.R. 1:1000}$$

$$T_d \propto BI \Rightarrow T_d \propto IV \Rightarrow T_d = kVI$$

AC measurement -

$$T_d = kVi, \quad V = V \sin \theta, \quad i = I_m \sin(\theta - \phi)$$

$$\Rightarrow T_d(\text{measured}) = \frac{kV_m I_m}{2\pi} \int_0^{2\pi} \sin \theta \sin(\theta - \phi) d\theta$$

$$= \frac{kV_m I_m}{4\pi} \int_0^{2\pi} [\cos \phi - \cos(2\theta - \phi)] d\theta$$

$$= \frac{kV_m I_m}{4\pi} [(2\pi - 0) \cancel{\cos \phi} + 0] = \frac{kV_m I_m}{2} \cos \phi$$

$$T_d = kV_{rms} I_{rms} \cos \phi$$

$$T_c = -k\theta \quad (\text{Spring control})$$

$$\text{At equilibrium, } T_d = T_c \Rightarrow \theta \propto V_{rms} I_{rms} \cos \phi = P$$

load power

Wattmeter will always measure active power $\propto P$
Thus, it has a uniform scale.

Advantages - 1. Uniform scale

2. higher accuracy

3. Can be used for both ac & dc measurement

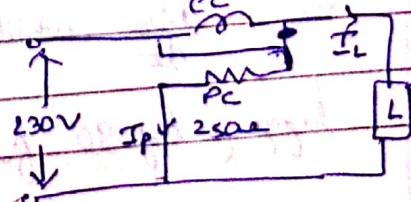
4. It can be used for calibration and standard.

Disadvantages - 1. Expensive

Q. The pressure coil of wattmeter has a resistance of 250Ω. The wattmeter is connected across a load.

Determine the actual load if wattmeter reading is 350W. Assume load voltage equal to 230V.

$$A. I_p = \frac{V}{R_p} = \frac{230}{250} = 0.92 A$$



$$W_p = V I = 230 \times 0.92 = 211.6 \text{ W}$$

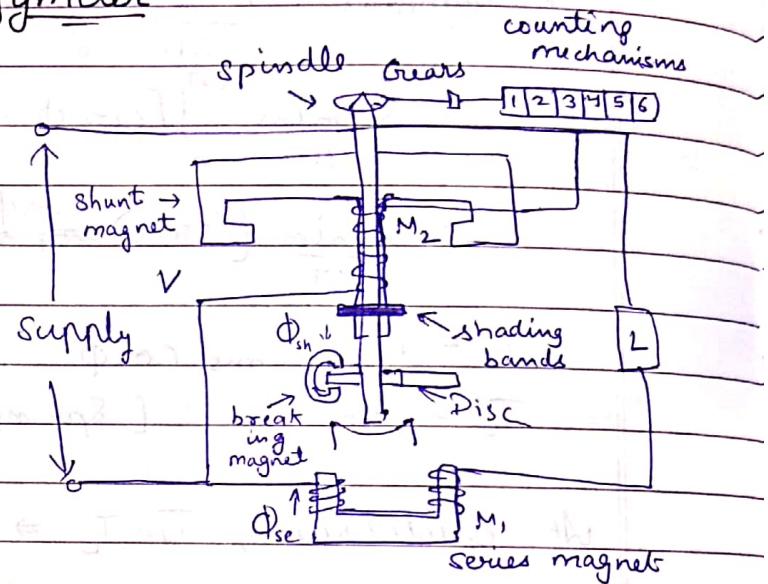
$$\text{Load power} = 350 \text{ W} - W_p = 138.4 \text{ W}$$

Induction type Energy meter

Deflection torque is produced by oddy currents and mutual induction.

Controlling torque is produced by the braking magnet.

But it does not bring the pointer to zero.



Shading band is a Cu band to produce phase difference in flux. It is used to adjust flux in M_2 magnet lag by 90° .

As M_1 and M_2 are silicon-steel laminated electromagnet. Disc is of Al. Torque is proportional to eddy current induced in disc.

V is connected with pressure coil

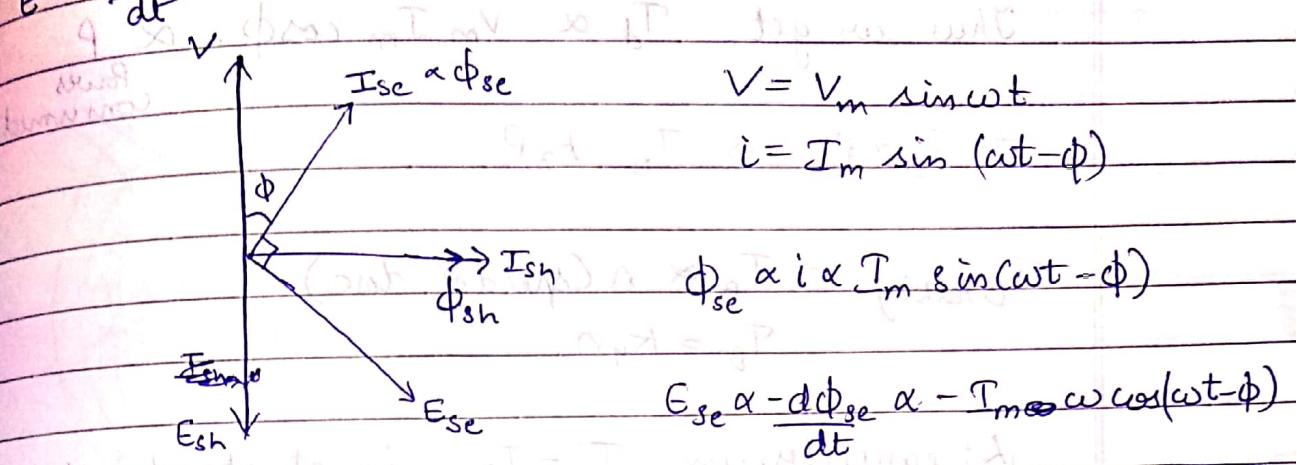
I is connected with current coil.

$V \rightarrow \Phi_{sh} - M_2 \rightarrow 90^\circ$ (due to shading band).

$E_{sh} = -\frac{d\Phi_{sh}}{dt}$, which will produce current. (eddy current).

Lagging by 90° from Φ_{sh}

In series magnet, current and flux will be in phase but
 $E = -\frac{d\phi_{se}}{dt} \Rightarrow E$ is 90° apart from ϕ_{se} .



Considering series eddy current

$$\text{Now, } \phi_{sh} \propto -\int V dt \propto -\int V_m \sin \omega t dt$$

$$\propto \frac{V_m \cos \omega t}{\omega}$$

$$E_{sh} \propto -\frac{d\phi_{sh}}{dt} \propto -V_m \sin \omega t$$

$$i_{sh} \propto -V_m \sin \omega t$$

If we assume eddy current path is purely resistive having a value R .

$$i_{se} \propto \frac{E_{se}}{R}, \quad I_{se} = k_1 \frac{E_{se}}{R}$$

$$i_{se} = -\frac{k_1}{R} I_m \cos \omega t \cos (\omega t - \phi)$$

$$i_{sh} = k_2 \frac{E_{sh}}{R} \Rightarrow i_{sh} = -k_2 \frac{V_m}{R} \sin \omega t$$

$$\tau \propto \phi_{sh} i_{se} - \phi_{se} i_{sh}$$

$$\Rightarrow \tau \propto \frac{k_2 V_m}{\omega} \times \frac{k_1 I_m}{R} \omega \cos \omega t \cos (\omega t - \phi)$$

$$+ \frac{k_1 I_m k_2 V_m}{R} \sin \omega t \sin (\omega t - \phi)$$

$$\Rightarrow T_d \propto \frac{k_1 k_2 V_m I_m \cos \phi}{R}$$

Thus, we get $T_d \propto V_m I_m \cos \phi$, $\propto P$
Powers consumed

$$\Rightarrow T_d \propto P \Rightarrow T_d = k_3 P$$

$$T_B = T_d \propto n (\text{Speed of disc})$$

$$T_B = k_4 n.$$

At equilibrium $T_d = T_B$ i.e. at steady state of rotation.

$$\Rightarrow k_4 n = k_3 P$$

$$k_4 n dt = k_3 P dt$$

$$\int_0^t n dt = \frac{k_3}{k_4} \int_0^t P dt$$

No. of revolution made = $k \times$ energy consumed by disc in time t in time t .

$$\frac{\omega}{\omega_0} t = \frac{P}{P_0} t \Rightarrow \frac{\omega}{\omega_0} = \frac{P}{P_0}$$

$$(Starting) \omega_0 = P_0 = 1$$

$$\omega = \frac{P}{P_0} \omega_0 = \frac{P}{P_0} \cdot 1 = \frac{P}{P_0}$$

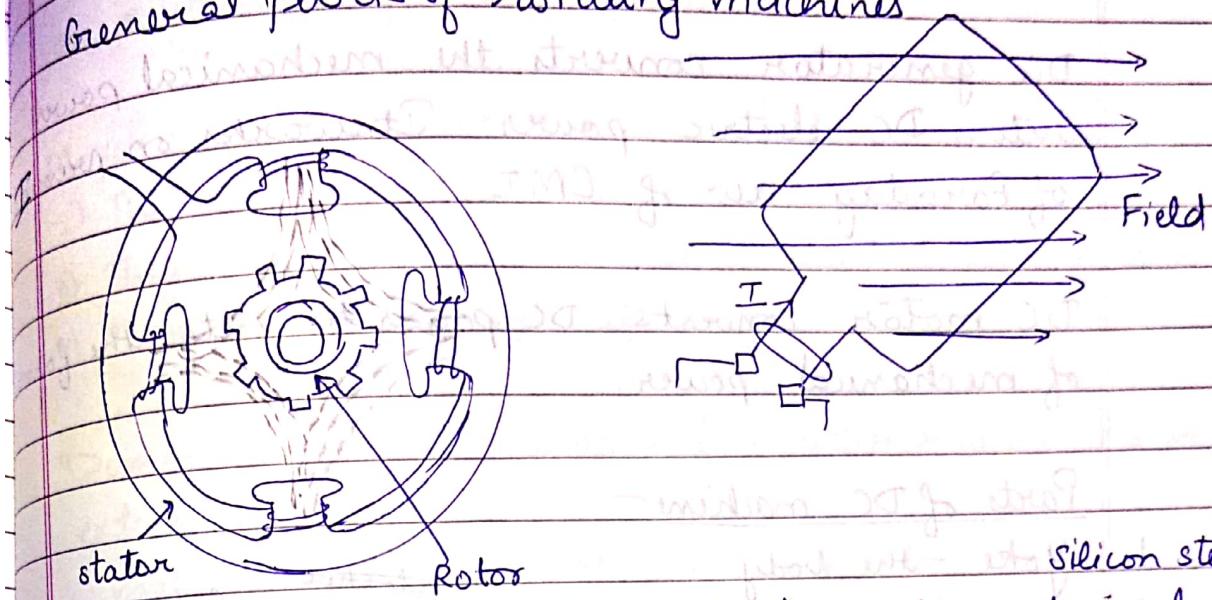
Electrical Machines

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Date: 11/4/2018

Rotating Machines -

General parts of rotating machines



Stator and rotor are made of same material
Rotor rotates in the field (flux pressure) of the
stator coil.

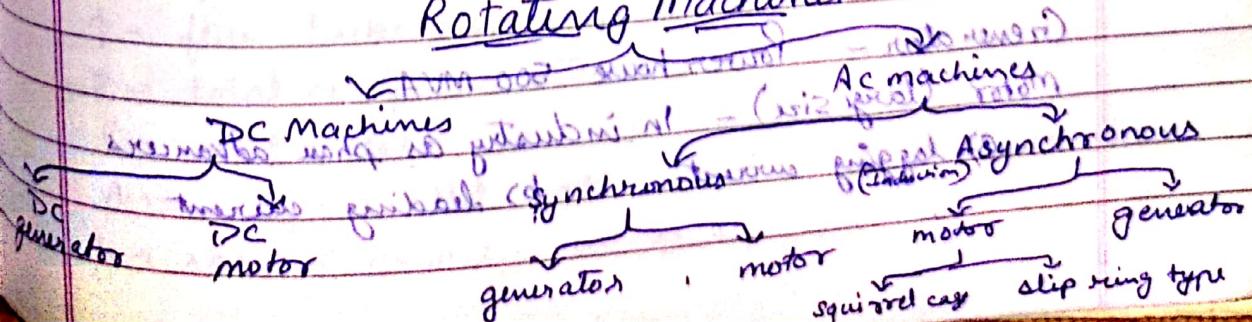
II Air gap - It separates stator and rotor (to avoid friction b/w the two). It needs to be small as it reduces power factor

III Two windings - Armature winding & Field winding.
Field winding produces flux / field (which can also be created by permanent magnet) b/w air gap
Armature winding rotates in the field

IV AC current in armature winding.

V Ventilating Ducts.

Rotating Machines



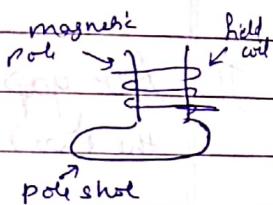
~~PP3 & M4~~ ~~Semidrum~~ ~~Induction~~ D.C. Machines - These work on DC

~~motor principle~~ ~~generator principle~~ ~~Faraday's law~~
DC generator converts the mechanical power into DC electric power. It works on principle of Faraday law of EMI.

DC motor converts DC power ~~to~~ into the form of mechanical power.

Parts of DC machine -

1. Yoke - the body
2. Magnetic pole or field coils ^{poleshot} _{Cu (Field coil)} - on which wire is wound
3. Armature core and armature winding
4. Commutator - terminal changer (converter)
5. Brushes (carbon)
6. Shaft bearing, pulley, fan
7. pole shoe → uniform flux distribution



Synchronous machine

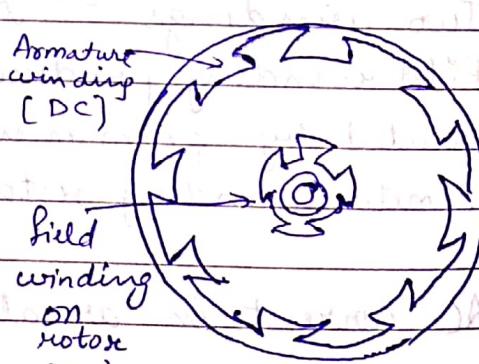
Doubly excited \rightarrow stator \rightarrow motor

synchronous speed N_s

$$N_s = \frac{120 f}{P}$$

$f \rightarrow$ supply freq. in Hz

$P \rightarrow$ no. of poles



Generator - Power house 500 MVA

Motor (large size) - In industry as ~~shunt~~ ^{excitation} ~~advancers~~ ^{advancers}

a) lagging current

b) leading current

It is normally not used in industrial sector. But it has an advantage of power factor.

- Merits - 1) It is used for power factor improvement
- 2) The speed is constant & independent of load.
- 3) They are better mechanically than induction motors.
- 4) They operate at high efficiency

Demerits - 1) DC excitation for motor requires a separate external source.

- 2) Variable speed control is not possible.
- 3) Cost /kW output is higher.
- 4) Motor is not self starting.

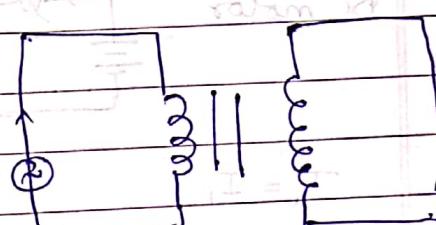
Induction machine motor

conductors winding

is shorted. Current is

Induced in it which is high.

Torque is applied on the
rotor winding.



DC Motor - When DC electric power is fetched to this machine, then it converts into the form of mechanical power.

Emf equation -

$\Phi \rightarrow$ flux linked with each pole.

$Z \rightarrow$ total no. of armature conductors.

$P \rightarrow$ no. of poles

$A \rightarrow$ no. of parallel paths in which conductors are distributed.

$N \rightarrow$ speed of armature is rpm .

and if $E_g \rightarrow$ generator emf as here there is no current in it.
Then, total flux = flux produced by each pole \times no. of poles
 $\Rightarrow d\phi = P\phi$

Time taken to complete one revolution by a conductor in air

$$dt = 1/60 \text{ sec}$$

~~then~~ $e = id\phi/dt = P\phi/N$ emf/conductor \times Total no. of armature

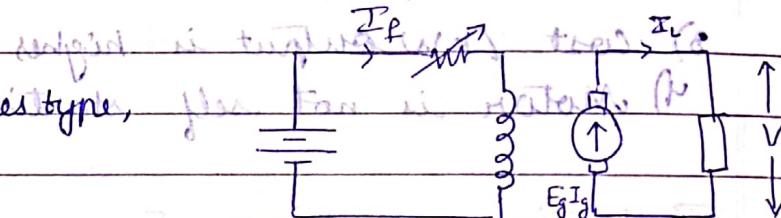
provided $d\phi/dt$ is due to N no. of pole path per

$$= \frac{P\phi}{60/N} \times N \Rightarrow e = \frac{P\phi N z}{60 \cdot A}$$

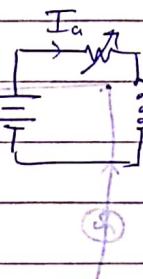
is proportional to number of turns \propto N \propto A

Types of DC Generators \rightarrow series excited CS

Self excited \rightarrow series type, shunt type.



DC motor



current direction

changes at field winding & armature winding are separated from each other

right in between time be what

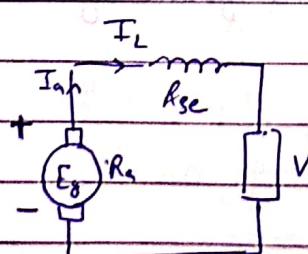
$$V = E_g - I_a R_a$$

electric power output = $V I_a$

Power generated = $E_g I_a$

Power at load = $E_g I_a - I_a^2 R_a$

lost power



It is self excited machine. Field winding is connected in series with armature winding

Power developed = $E_g I_a$

Series type generator Output power = $E_g I_a (R_a + R_{se})$

$$I_a = I_{se} = I_L = I$$

$$V = E_g - I_a R_a - I_a R_{se}$$

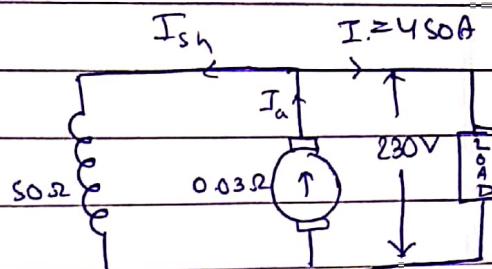
A shunt generator develops 450A at 230V and armature resistance is 5Ω. If the resistance of shunt field and armature are 50Ω and 0.03Ω respectively. Calculate the generated emf.

$$I_{sh} = \frac{230V}{50\Omega} = 4.6A$$

$$\text{load current } I = 450A$$

$$\Rightarrow I_a = I + I_{sh} = 454.6A$$

$$E_g = V + I_a R_a = 230V + (454.6)(0.03) = 230 + 13.6 \\ [E_g = 243.6V]$$

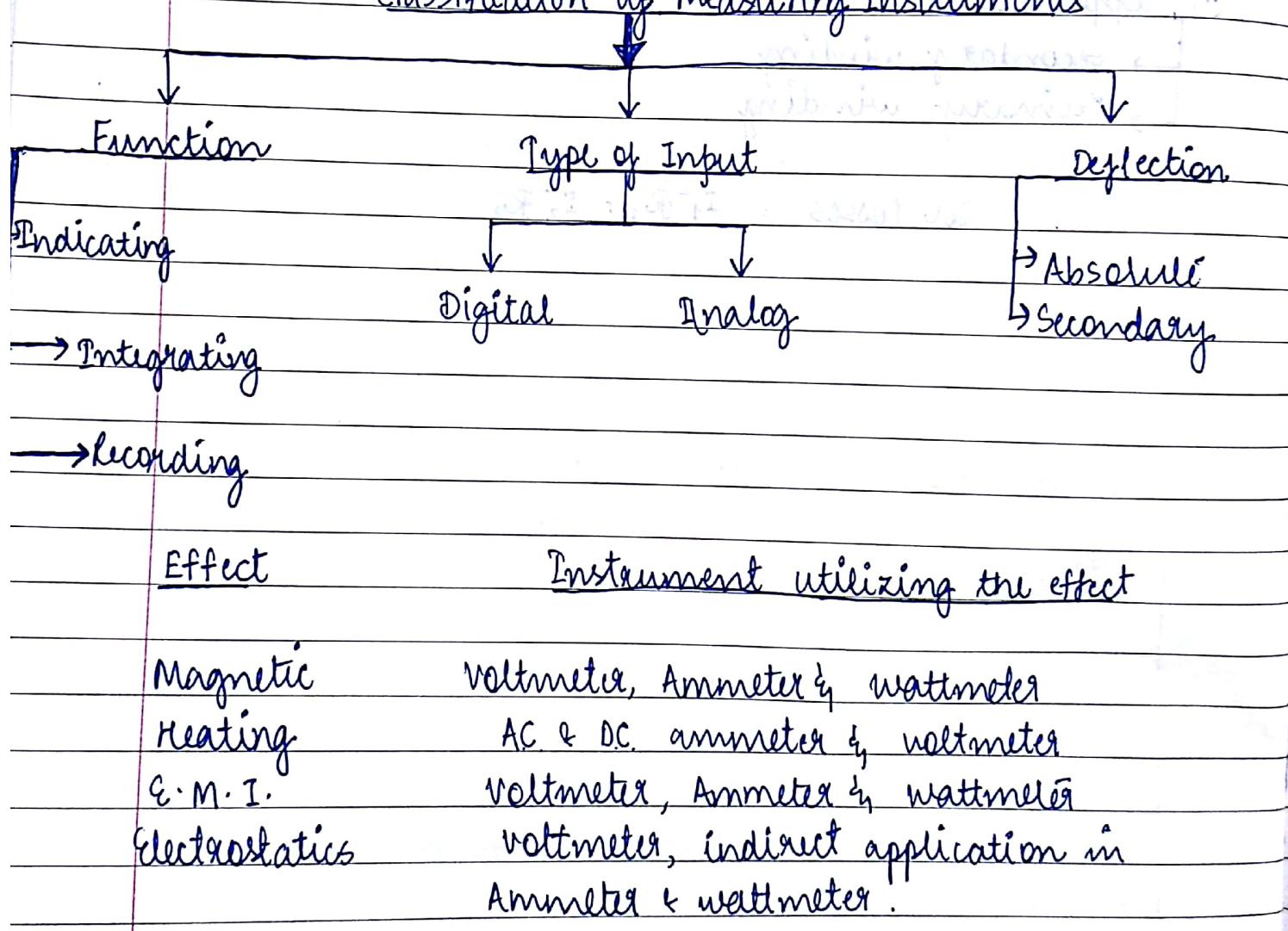


MEASURING INSTRUMENTS

TERMINOLOGY

1. Measurement: is a process to know unknown quantities in terms of known standards.
2. Measuring Instrument: Tools which are used for the purpose of measurement.
3. Instrumentation: Technology used for measurement

Classification of Measuring Instruments



I On the basis of function:

1. Indicating Instruments

The value of electrical quantity is indicated at the time when it is being measured.

Pointers moving over the scale give indication.

Eg: ammeters, voltmeters & wattmeters.

2. Recording Instruments

A continuous record of variations of the electrical qty. over a long period of time is given by these instrument.

It has a moving system which carries an inked pen which rests tightly on a graph chart.

Eg: Graphic recorders, galvanometers, etc.

3. Integrating Instruments

The total amount of electricity or electrical energy supplied over a period of time is measured.

Eg: Amper-hour meters, energy-meters.

II On the basis of Deflection:

1. Absolute Instruments

Gives the value of the quantity to be measured in terms of the constants of the instrument and their deflection only.

Eg: tangent galvanometer

gives value of I in terms of tangent of deflection produced, no. of turns & radius of wire used & horizontal component of earth's mag. field.

NOTE : No previous calibration is required in this case.

2. Secondary Instruments

The value of electrical qty. to be determined from the deflection of these instruments
eg: ammeter, voltmeter, etc.

Pre-calibrated with absolute instruments.

On the basis of Type of Input

1. Digital

2. Analog

Essentials of Indicating Instruments:

Indicating instruments consist of a pointer which moves over a calibrated scale and which is attached to a moving system pivoted in jewelled bearings.

The moving system is subjected to these 3 torques:

(a) Defecting / Operating Torque (T_d): Produced by utilizing any of the effects in table 3.1. It causes movement of the pointer.

(b) Controlling / Restoring Torque (T_c): Opposes the deflection torque & increases the deflection of the moving system of instrument. The pointer is brought to rest at a position where the two opposing torques are equal.

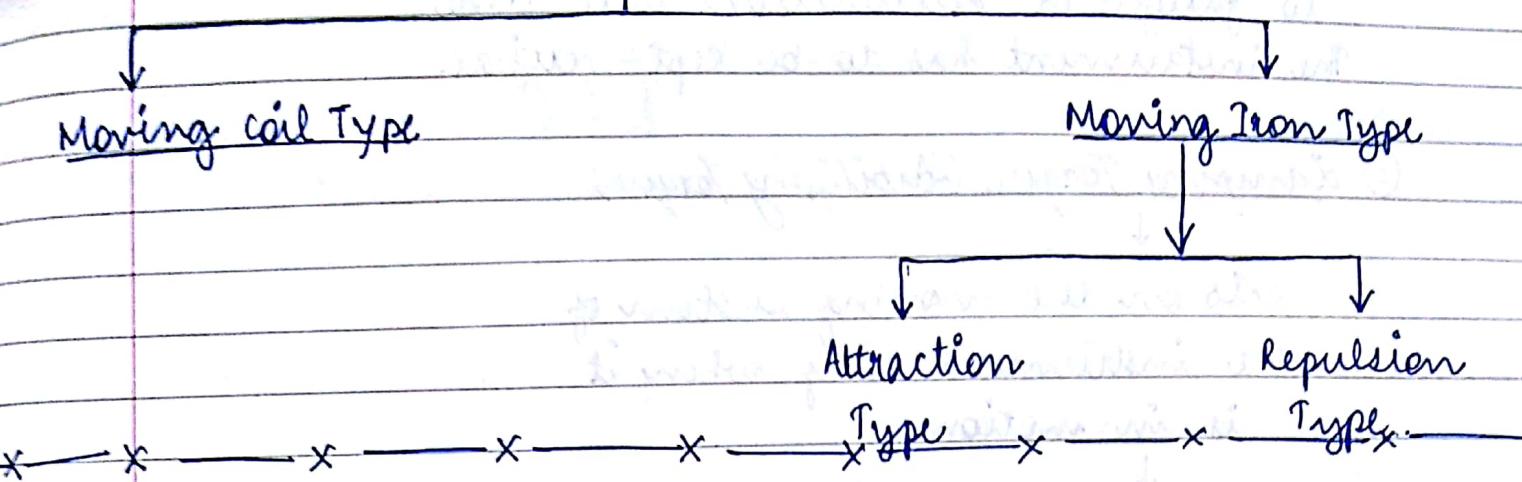
In the absence of restoring torque, the pointer once deflected will not return to its original position.

It is obtained either by

(i) Spring control

(ii) Gravity

Instrument Type



(i) Spring control:

A hair-spring, usually of phosphor-bronze, is attached to the moving system of the instrument.

With the deflection of the pointer, the spring is twisted in the opposite direction.

This twist in the spring produces restoring torque

$$T_c \propto \theta.$$

(ii) Gravity control

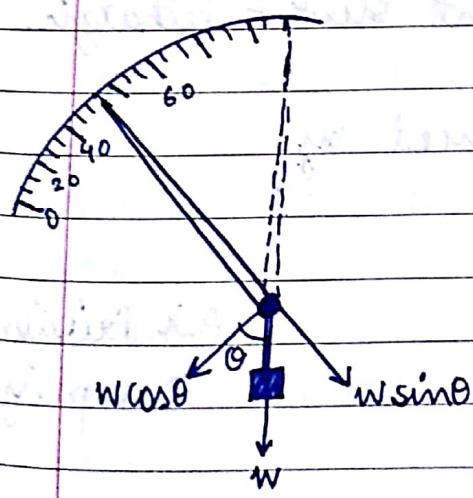
A small adjustable weight is attached to some part of the moving system such that the weights exert torques in the opposite direction.

$$T_d \propto I$$

$$T_c \propto \sin\theta$$

$$I \propto \sin\theta$$

The degree of control is adjusted by screwing the weight up or down the carrying system.



It gives a damped oscillation at the lower ends

$\because I \propto \sin\theta$ and at low value of θ value of $\sin\theta$ is also less]

It is cheap, unaffected by temp. and is not subjected to fatigue or deterioration with time.
The instrument has to be kept vertical.

(c) Damping Torque (Stabilizing Torque)

acts on the moving system of the instrument only when it is in motion

opposes its motion

Such stabilizing force or damping force is necessary to bring the pointer to rest quickly, otherwise due to inertia of the moving system, the pointer will oscillate about its final deflected position for a long time.

The degree of damping should be adjusted to a value which is sufficient to enable the pointer to rest quickly to its deflected position.

Overdamping makes the instrument slow & lethargic and is said to be dead beat.

The damping force can be produced by

Liquid Friction

Damping

OR

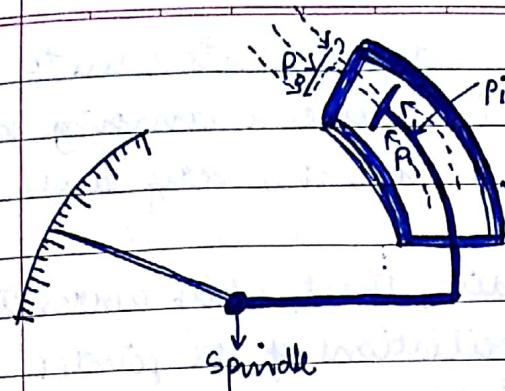
Fluid Damping

Eddy Current

Damping

Air Friction

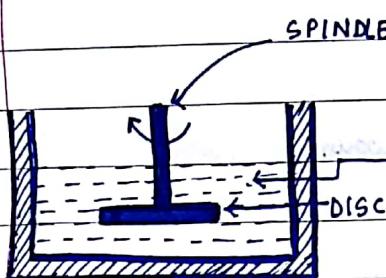
Damping



At piston

when there are oscillations
the piston moves in & out of
the chamber

Fluid friction damping



when piston moves inside the
chamber, $P > P_0$; so air opposes
the motion of piston & the moving
system

when piston moves out of the
chamber $P \downarrow$ and $P < P_0$, so
again there is an opposition to
the motion.

Advantages of fluid friction:

Due to more viscosity of fluid, more damping is provided.
Oil can also be used for insulation purposes.

Disadvantages:

Used only for instruments in vertical position.

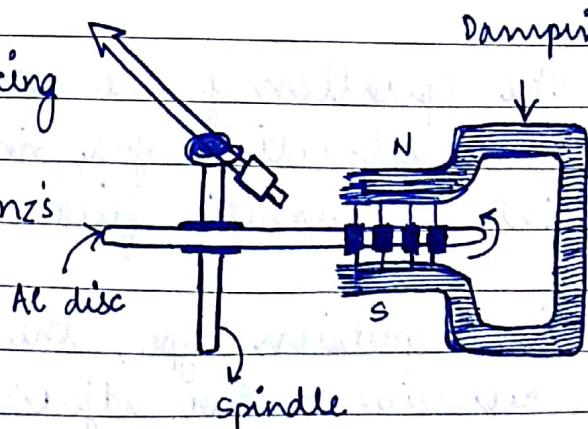
Objectionable creeping of oil.

Eddy Current Damping

Most efficient way of producing
damping.

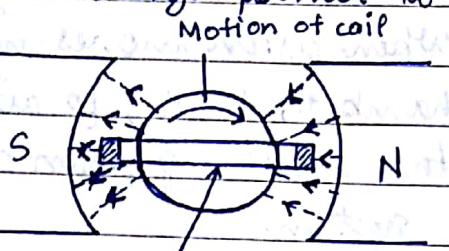
Based on Faraday's law & denis's
law.

When Al disc rotates, it
cuts the magnetic flux b/w
the poles of a permanent magnet.



This causes the production of eddy currents (due to induced emf) which flow and produce a damping force in such a direction so as to oppose the very cause of its production.

Since the cause is rotation of the disc, these eddy currents retard the motion of disc & oscillations of the pointer. This brings pointer to rest quickly.



Metal Former (on which coil is wound)

Moving Iron Type
→ Attraction
→ Repulsion

Hot-wire Type

Moving coil Type

→ Permanent Magn.
→ Electrodynami

Induction Type

Electrostatic
(Voltmeters only)

MOVING IRON INSTRUMENTS



moveable part is iron

The operation of the attraction type depends on the attraction of a single piece of soft iron into a magnetic field.

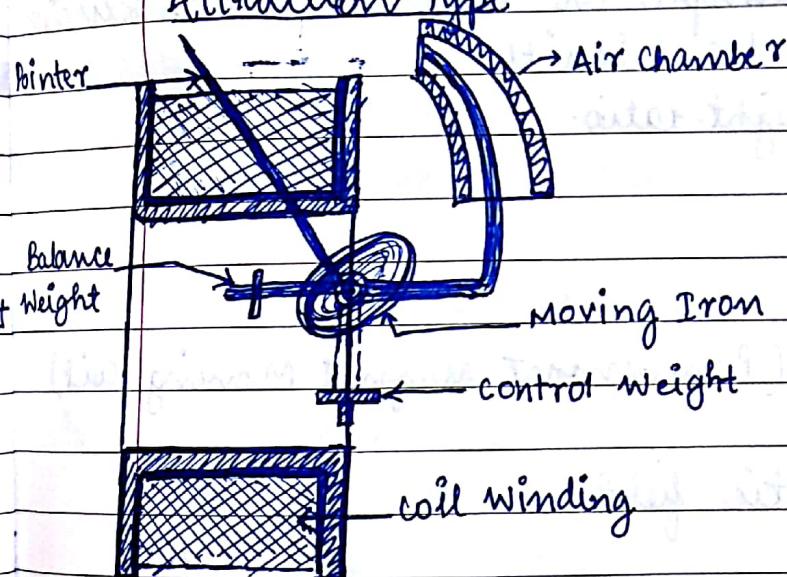
For repulsion type, the operation depends on the repulsion of two adjacent pieces of soft iron magnetised by the same magnetic field.

The necessary magnetic field is produced by the ampere-turns of a current carrying coil.

For ammeter \Rightarrow coil has fewer turns of wire so that ammeter has lower resistance since it is connected in series with a circuit.

for voltmeter \Rightarrow coil has high impedance (large no. of turns) so as to draw as small a current possible since it is connected in parallel with the circuit.

Attraction Type



Principle

If a piece of an unmagnetised soft iron is brought near a current carrying coil, it would be attracted into the coil.

* An oval-shaped disc of soft iron is pivoted on a spindle between bearings near the coil, the iron disc will swing into the coil when the latter has an electric current passing through it.

* whatever the direction of current through the coil be, the iron disc is always magnetised in such a way that it is pulled inwards.

* therefore, these type of instruments can be used for both d.c. & a.c.

Force $\rightarrow F \propto mH$
 pulling iron disc
 $H \Rightarrow$ field produced
 $m \Rightarrow$ developed pole strength

If current i flows through the loop.

$$m \propto i \quad H \propto i$$

$$F \propto i^2$$

$$T_d \propto i^2$$

$$T_c \propto \theta$$

$$T_d = T_c$$

$$\theta \propto i^2 \Rightarrow \text{Non-uniform scale}$$

Advantages of M.I. instruments

- less expensive, robust & simple in construction
- can be used for a.c. & d.c. circuits
- gives high torque to weight ratio.

Disadvantages of M.I. instruments

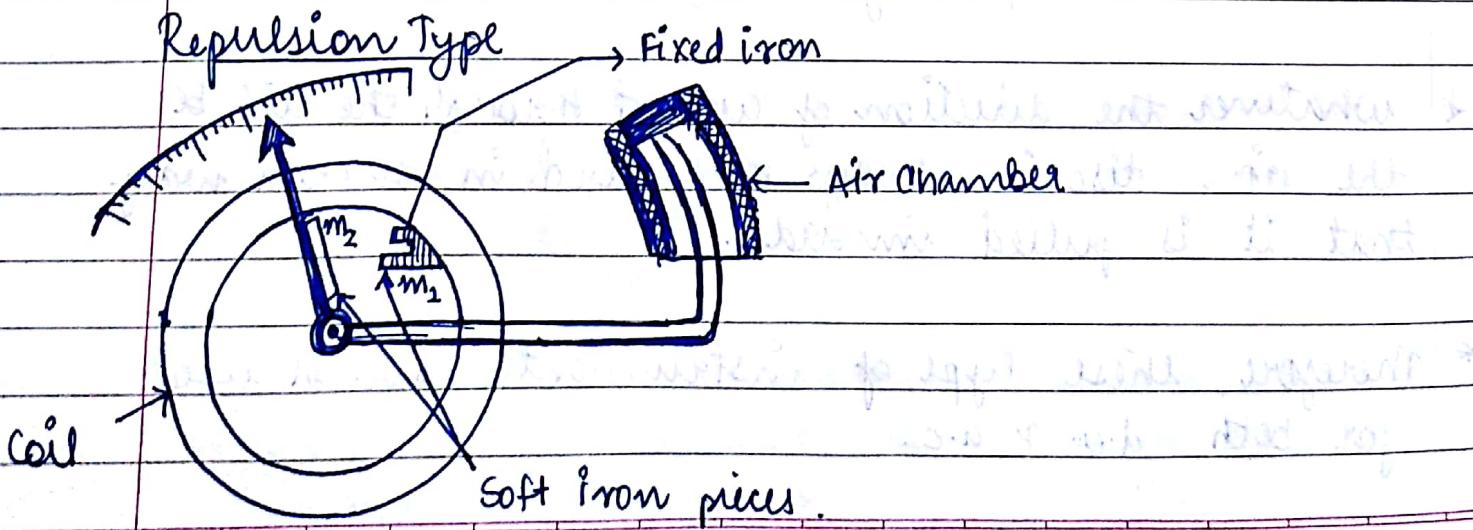
- Non-uniform scale
- Not sensitive as PMMC (Permanent Magnet Moving coil) instruments
- Affected by stray magnetic field.

Due to which instruments in lab

are enclosed in a wooden box.

Degree of precision of calibration is not high.

Repulsion Type



It consists of a fixed coil inside which 2 soft iron bars are placed parallel to one another & along the axis of the coil.

When current passes through the coil, it generates its own magnetic field which magnetizes the rods similarly and they start to repel.

$$T_d \propto m_1 m_2 \propto i^2$$

$$T_c = \theta$$

$$\theta \propto i^2$$

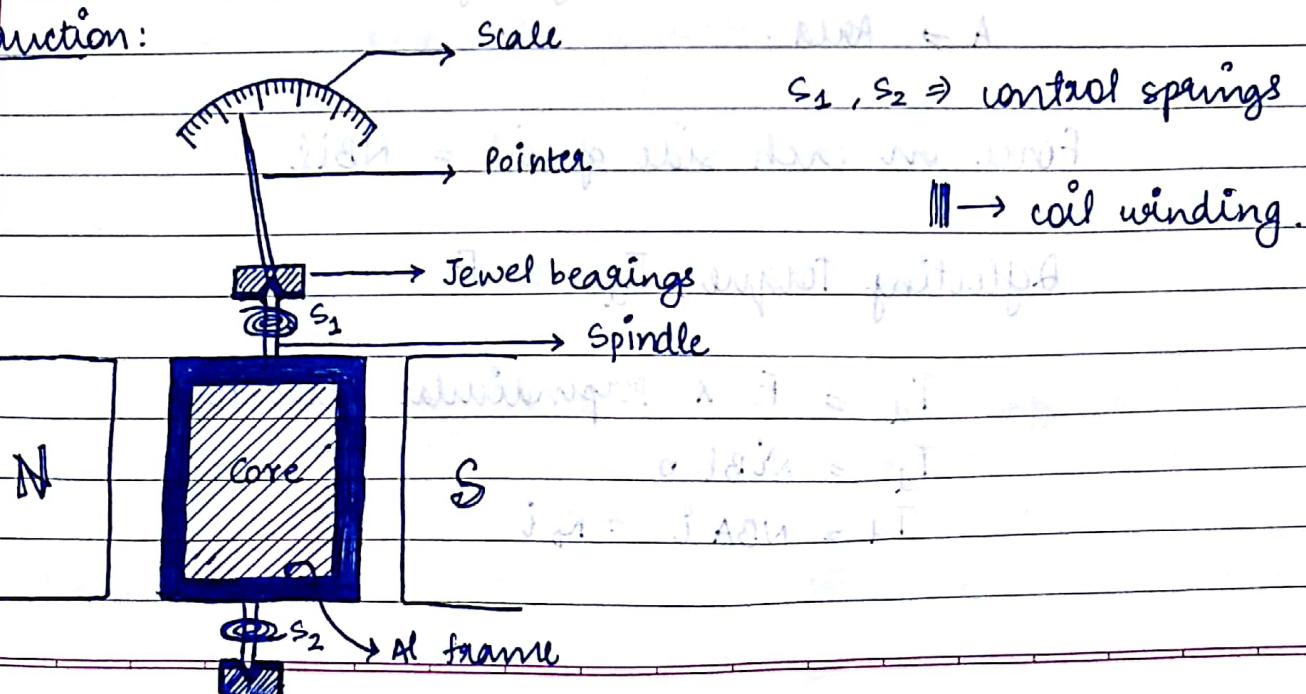
MOVING COIL INSTRUMENTS

- Permanent - Magnet Moving coil (PMMC)
- Dynamometer type

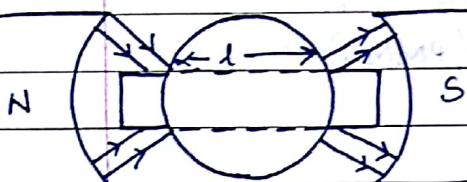
I. Permanent Magnet Type Instrument

Principle: When a current - carrying conductor is placed in a magnetic field, it is acted upon by a force which tends to move it to one side and out of the field.

Construction:



- * A rectangular coil of many turns wound on a light Al or copper frame inside which is an iron core.
- * Soft-iron cylinder is fixed b/w the poles
 → to make field radial & uniform
 → to decrease the reluctance of the air path b/w poles & hence increase the magnetic flux.
- * The Al frame provides support to the coil and also provides damping by eddy current induced in it.



when current is passed through the coil, force acts upon it both sides producing a deflecting torque.

Let, $B \rightarrow$ flux density in air gap

$N \rightarrow$ no. of turns in the coil

$i \rightarrow$ current in the coil

$b, l \rightarrow$ breadth & length of coil

$A \rightarrow$ Area.

Force on each side of coil = $NBil$

Deflecting Torque, $T_d = F \times d$

$$T_d = F \times \text{perpendicular distance}$$

$$T_d = Nibl b$$

$$T_d = NBA i = k_s i$$

Contact Torque.

$$T_c = K_s \theta$$

↳ spring constant.

At final deflected position

$$T_d = T_c$$

$$K_d i = K_s \theta$$

$$\Rightarrow [\theta \propto i] \Rightarrow \text{uniform scale}$$

- * Since the Al frame moves in an intense magnetic field, the induced eddy currents are large and damping is very effective.

Advantages:

- low power consumption
- uniform scale
- high (torque / weight) ratio
- no hysteresis loss
- effective & efficient eddy current damping.
- Not much affected by stray magnetic fields.

↓
due to strong operating fields

- accurate, reliable & sensitive.

Disadvantages:

- Cannot be used to measure a.c.

↓

direction of current will be reversed during -ve half cycle and hence direction of torque will also be reversed which gives average torque zero. The pointer will not deflect from its mean position showing zero reading.

→ quite expensive !

* PMMC instruments can be used as ammeters or as voltmeters.

Q:- The coil of PMMC instrument has 40 turns on an Al frame whose dimensions are 18mm wide & 25mm effective length. It moves in a uniform field of flux density 0.6T. The control spring constant is 2.5×10^6 Nm/degree. Calculate the current required to produce a deflection of 110° .

Sol:- $N = 40$ $l = 25 \text{ mm}$ $b = 18 \text{ mm}$

$B = 0.6 \text{ T}$ $K_s = 2.5 \times 10^6 \text{ Nm/degree}$.

$\theta = 110^\circ$

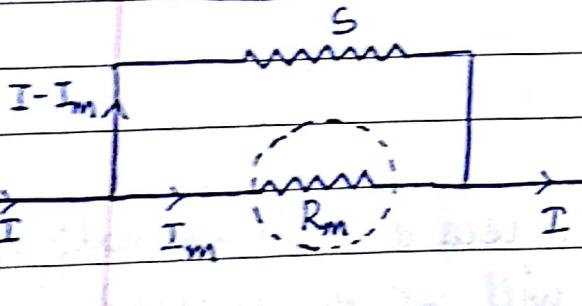
$$K_s \theta = NABi$$

$$K_s \theta = i \Rightarrow i = \frac{2.5 \times 10^6 \times 110}{40 \times 18 \times 25 \times 10^{-6} \times 0.6}$$

$$\Rightarrow i = \frac{110}{24 \times 18} = 0.02 \text{ A}$$

Extension of Range

(1) As ammeter



$R_m \rightarrow$ instrument resistance

$S \rightarrow$ shunt resistance

$I_m \rightarrow$ full-scale deflection

$I \rightarrow$ current of the instrument

$I \rightarrow$ line current to be measured

→ A low resistance shunt is connected in parallel with ammeter to extend the range of current.

→ Large current can be measured using low current rated ammeter using a shunt.

$$V_m = V_s$$

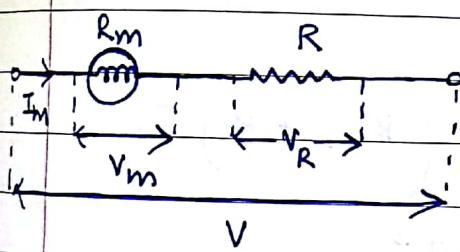
$$I_m R_m = (I - I_m) R_s$$

$$I_m (R_m + R_s) = I R_s$$

$$I = \left(1 + \frac{R_m}{R_s} \right) I_m$$

$\left(1 + \frac{R_m}{R_s} \right)$ ⇒ multiplying power

(2) As Voltmeter (multiplier)



I_m → full scale deflection current
 V → voltage to be measured

Since same current flows through R_m & R

$$V_m / R_m = V / R$$

$$V_m / R_m = (V - V_m) / R$$

$$V_m (R_m + R) = V / R \Rightarrow V_m \left(\frac{1}{R_m} + \frac{1}{R} \right) = \frac{V}{R}$$

$$V = \left(1 + \frac{R}{R_m} \right) V_m$$

Q: A moving coil instrument has a resistance of 75Ω and gives full scale deflection when carrying a current of 35mA . Show how it can be adapted to measure

- i) current upto 80A
- ii) voltage upto 550V .

$$(i) I = \left(1 + \frac{R_m}{R_s} \right) I_m$$

$$80 = \left(1 + \frac{75}{R_s} \right) 35 \times 10^{-3}$$

$$1 + \frac{75}{R_s} = \frac{80 \times 10^3}{35}$$

$$\frac{75}{R_s} = \frac{8 \times 10^4}{35} - 1$$

$$R_s = 0.032\Omega$$

$$(ii) V = \left(1 + \frac{R}{R_m} \right) V_m = (R_m + R) I_m$$

$$V = (75 + R) 35 \times 10^{-3}$$

$$75 + R = \frac{550 \times 10^3}{35}$$

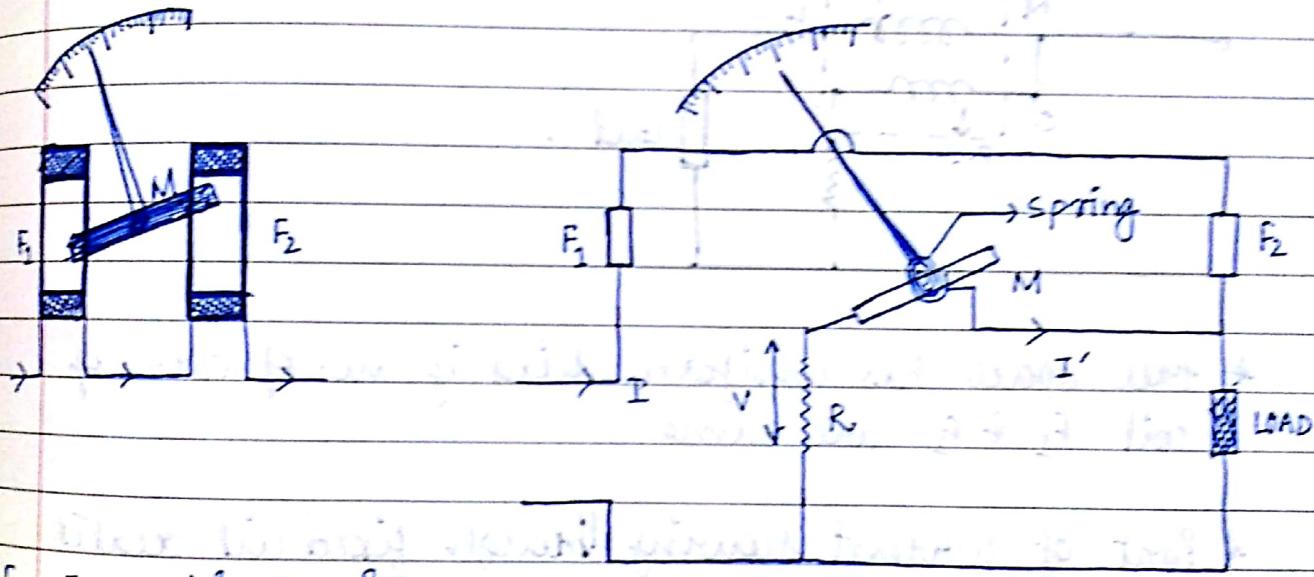
$$R = 15639\Omega$$

I. Dynamometer Type Instruments / Electrodynamics:
EMMC (Electromagnet moving coil)
operating field produced in moving coil by another
fixed coil.

Can be used either as an ammeter or voltmeter but
generally used as a wattmeter.

Electrodynamical Wattmeter

used to measure power taken
from ac or dc sources



F₁, F₂ \Rightarrow fixed coils
M \Rightarrow movable coil

* It consists of two coils, namely fixed coil and
movable coil.

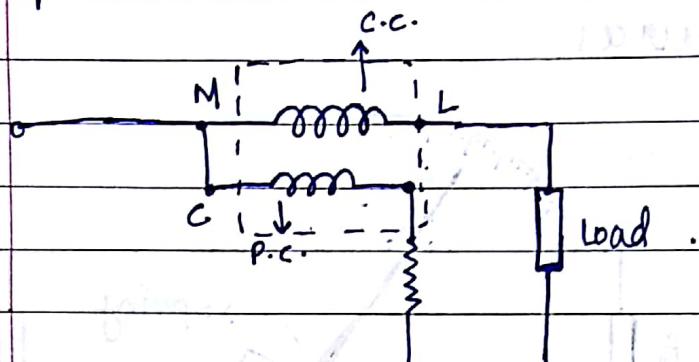
* The fixed coil is split into two equal parts which are
placed close together and parallel to each other.

* Moving coil is pivoted between the fixed coils & is
placed on the spindle.

* The fixed coils are connected in series with the load and carry circuit current.
It is therefore called current coil.

* The moving coil is connected across the load & carries current proportional to the voltage & is hence called potential or pressure coil.

* A high resistance is connected in series with pressure coil to limit the current through it.



* free space has uniform field if no. of turns of coil F_1 & F_2 are same

* Part of current flowing through fixed coil creates magnetic field & current flowing in moving coil creates deflection torque

* Spring contact used for controlling torque.

* Air friction damping is used.

(a) When using lamp test the lamp if not
with the meter it will not work

(b) In air gap we inserted bobbin in the primary &

DC measurement

Let $I_1 \Rightarrow$ current through fixed coil

$I_2 \Rightarrow$ current through moving coil

$$B \propto I_1 \Rightarrow B = K I_1.$$

Deflection torque,

$$T_d = N B A I_2$$

$$T_d = N A K I_1 I_2$$

$$T_d \propto I_1 I_2$$

Controlling torque,

$$T_c \propto \theta \Rightarrow T_c = K' \theta$$

When

$$T_d = T_c \Rightarrow N A K I_1 I_2 = K' \theta$$

$$\Rightarrow \theta \propto I_1 I_2$$

AC measurement

$$T_d \propto V i \Rightarrow T_d = K V i$$

$$V = V_m \sin \theta, \text{ for } i = i_m \sin(\theta - \phi)$$

$$(T_d)_{\text{mean}} = \frac{K V_m i_m}{2 \times 2\pi} \int_{-\pi}^{\pi} 2 \sin \theta \sin(\theta - \phi) d\theta$$

$$= \frac{K V_m i_m}{4\pi} \int_0^{2\pi} [\cos(\theta - \theta + \phi) + \cos(2\theta - \phi)] d\theta$$

$$= \frac{K V_m i_m}{4\pi} \left[\int_0^{2\pi} \cos \phi d\theta + \int_0^{2\pi} \cos(2\theta - \phi) d\theta \right]$$

$$= \frac{K V_m i_m}{4\pi} [2\pi \cos \phi - \frac{[\sin(2\theta - \phi)]_0}{2}]$$

$$(T_d)_{\text{mean}} = \frac{K V_{\text{min}} \times 2\pi \cos \phi}{2\pi K}$$

$$(T_d)_{\text{mean}} = \frac{K V_{\text{min}}^2 \cos \phi}{2}$$

$$(T_d)_{\text{mean}} = K V_{\text{rms}} I_{\text{rms}} \cos \phi$$

Controlling torque, $T_c \propto \theta$.

$$T_d = T_c$$

$$\theta \propto V_{\text{rms}} I_{\text{rms}} \cos \phi \propto P$$

↳ Active power.

Advantages:

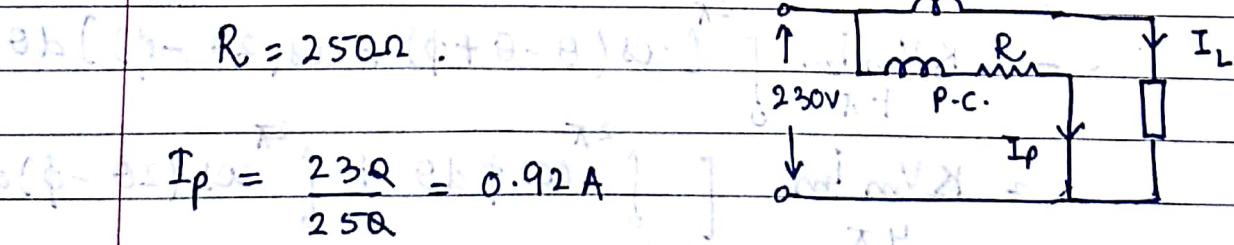
- uniform scale
- High degree of accuracy.
- Can be used for a.c. & d.c. measurement.
- Can be used for standardization & calibration.
- Free from hysteresis & eddy current losses.

Disadvantages:

- Quite expensive
- Small (torque/weight) ratio \Rightarrow low sensitivity.

Q: The pressure coil of wattmeter has a resistance of 250Ω . The wattmeter is connected across a load.

Determine the actual load if wattmeter is reading 350W. Assume the voltage equal to 230V.



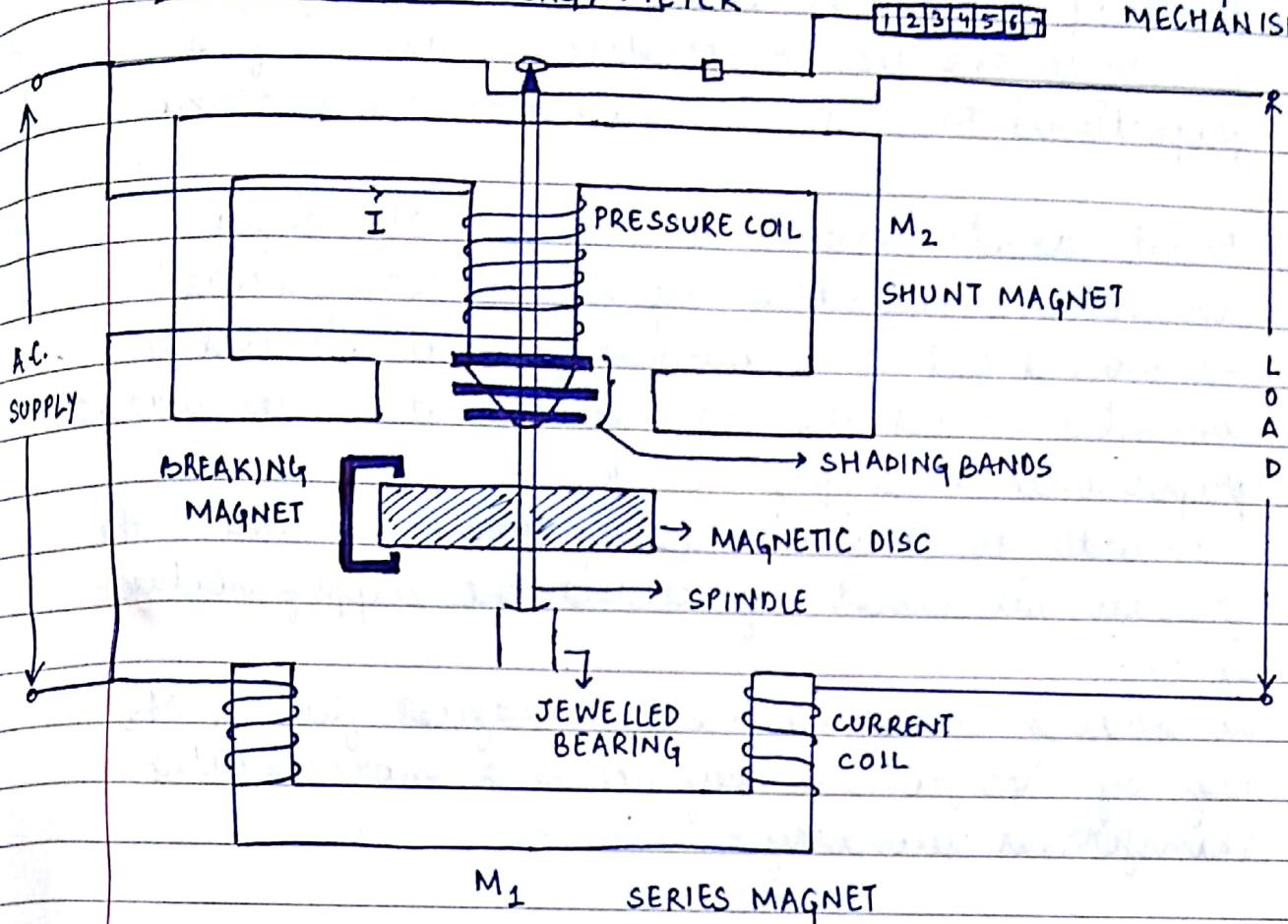
$$I_p = \frac{230}{250} = 0.92 \text{ A}$$

$$W_p = VI = 230 \times 0.92 = 211.6 \text{ W}$$

$$W_L = W - W_p = 350 - 211.6 = 138.4 \text{ W}$$

INDUCTION TYPE ENERGY METER

COUNTING MECHANISM



Construction

It generally has -

(1) Moving System

- Consists of a light Al disc mounted on a vertical spindle.
- No pointer & control spring \Rightarrow disc makes continuous rotations under the action of deflecting torque.

(2) Operating Mechanism

- Consists of series magnet, shunt magnet & breaking magnet.

(3) Series magnet : A thick wire of few turns is wound on both legs of the U-shaped laminated core. The wound coil is known as current coil and is connected in series with the load so that it carries

the load current. The series magnet is placed underside the disc & produces magnetic field proportional to and in phase with the current.

(b) Shunt magnet: consists of a no. of M-shaped laminations assembled together to form a core. The wound coil is known as pressure coil and is connected across the load so that it carries current proportional to supply voltage.

In order to obtain deflecting torque, current in the pressure coil must lag behind the supply voltage by 90° .

Cu shading bands are used to adjust flux in M_2 lag by 90° from V. They act as a short circuited transformer secondary.

(c) Breaking Magnet:



control the speed of the Al disc.

→ The magnet is mounted so that the disc revolves in the air gap between its poles.

→ As the disc rotates, currents are induced in the disc because it cuts the flux produced by the breaking magnet.

→ The direction of current in the disc is such that it opposes the rotation of the disc.

$$I_{\text{ind}} \propto N_{\text{disc}} \times \text{Breaking torque}$$

(ii) (3) Recording mechanism

The no. of revolutions of the disc is a measure of the electrical energy passing through the meter & is recorded on dials which are geared to the shaft.

Working

Energy meter connected to the circuit

\downarrow
C.C. \Rightarrow carries load current \Rightarrow B in phase with I_L

P.C. \Rightarrow carries current proportional to $V_s \Rightarrow$ B lags from I_L by 90°

\downarrow
C.C. field produces eddy currents
in the disc which reacts with
the field due to the pressure coil.

\downarrow
Thus a driving force is created which
causes the disc to rotate.

$I_L \Rightarrow$ line current
 $V_s \Rightarrow$ supply voltage

ROTATING MACHINES

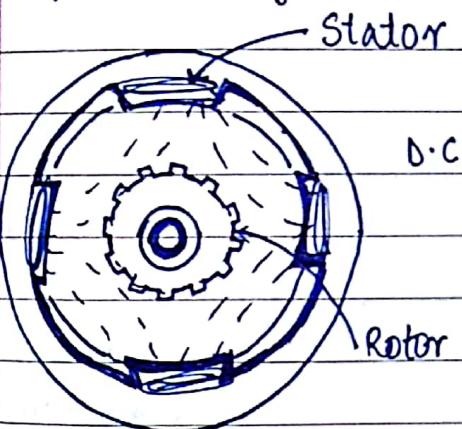
Voltages are generated in windings or group of coils by rotating these windings mechanically through a magnetic field.

Components of Rotating Machines

(1) Stator & Rotor

Stator is the stationary component of an electrical machine that is going around the rotor.

It contains windings & provides mechanical support & protection for the motor.



Rotor is general term for the main spinning part of machine. It is the part of the machine where electromechanical energy conversion occurs.

Laminated core type electrical steel used for construction of S&R.

(2) Air Gap

- Separates rotor and stator (to avoid friction b/w the two)
- Needs to be as small as possible (As it reduces power factor)

(3) Windings

(i) Armature winding : set of windings on a rotating machine which carry ac currents.

In ac. machines \Rightarrow armature winding is on the stator

In d.c. machines \Rightarrow armature winding is found on rotor

- (ii) Field winding: carry dc current & are used to produce the main operating flux in the machine.
 a.c. machines : found on rotor
 d.c. machines : found on stator

Only synchronous a.c. machines & d.c. machines have field winding

Permanent magnets also produce dc magnetic flux and are used in the place of field windings in some machines.

(4) Ventilating ducts provide cooling

ROTATING MACHINES

D.C. Machines

- D.C. Generator
- D.C. Motor

A.C. Machines

- Synchronous
- Motor
- Generator

- Asynchronous
(Induction)

Motor

Generator

Squirrel
Cage

Slipping
type

D.C. MACHINES

works on the principle of Faraday's law of E.M.F.

Armature winding is on rotor with current conducted from it by means of carbon brushes.

The field winding is on the stator & is excited by d.c.

D.C. MOTOR



converts electrical energy into mechanical energy.

Principle : When a current-carrying conductor is placed in a magnetic field, it faces & experiences a mechanical force given by

$$F = BIL$$

Direction of $F \Rightarrow$ Fleming's left-hand rule.

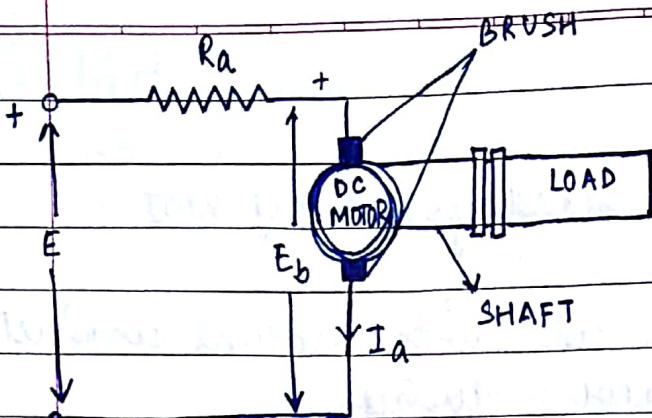
D.C. Motor



Shunt wound Series wound Compound wound

Parts of a D.C. motor

- Stator & Rotor
- Yoke (Magnetic frame \Rightarrow made of cast iron or steel)
- Poles & pole shoe or field coils
- Armature core & armature winding
- Commutator (to relay the supply current from mains to arm-winding)
- Brushes
- Shaft, bearing, pulley



$R_a \rightarrow$ armature resistance

$E \rightarrow$ input voltage

When voltage is applied across the motor, I_a produces torque in the presence of magnetic field

Due to this torque, armature rotates and it produces an induced emf E_b directed opposite to E . It is known as back emf.

$$E_b = \frac{P \cdot \phi \cdot Z \cdot N}{60 \cdot A}$$

where $P \rightarrow$ no. of poles

$\phi \rightarrow$ flux per pole

$Z \rightarrow$ no. of conductors

$A \rightarrow$ no. of parallel paths in which conductors are di-

$N \rightarrow$ speed of dc motor in rpm

$$E_b = \frac{P \cdot \phi \cdot Z \cdot N}{60} \times \frac{Z}{A}$$

$E_b = \text{emf per conductor} \times \frac{\text{Total no. of Armature}}{\text{conductors}}$

No. of parallel path

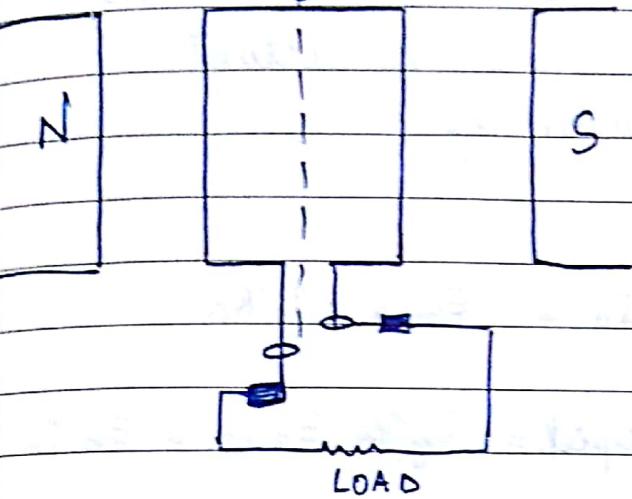
$$I_a = \frac{E - E_b}{R_a}$$

D.C. GENERATOR

Page No. _____

converts mechanical energy into electrical energy

This energy conversion is based on the principle of production of dynamically induced emf.

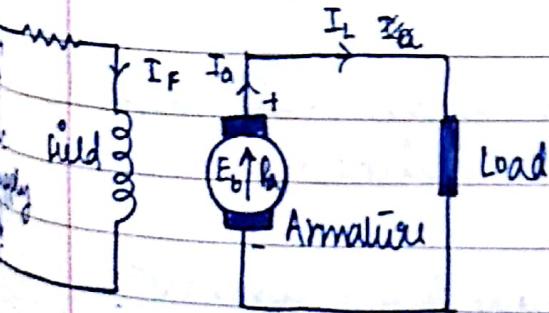


Need of a split ring commutator: to achieve single direction of current

D.C. Generator

Solely excited

Field coils are energized from an independent external DC source



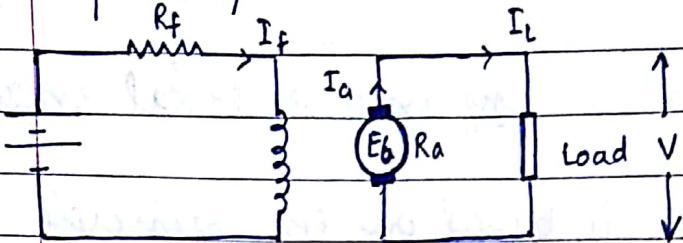
Self excited

Field coils are energized from the current produced by the generator itself.

It is further classified into

- series
- shunt
- compound.

(1) Separately Excited



$$I_a = I_L$$

$I_a \Rightarrow$ armature current

$I_L \Rightarrow$ line current

Terminal voltage V is given by

$$V = E_b - I_a R_a$$

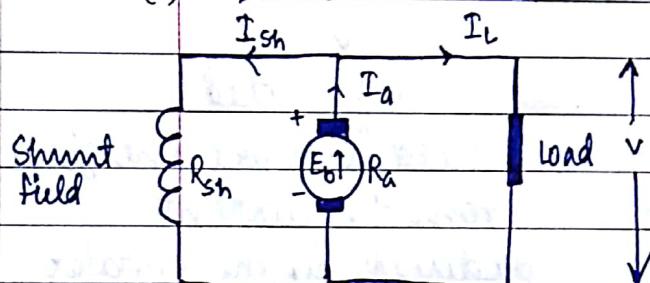
$$\text{Power at load, } P_L = V I_a = E_b I_a - I_a^2 R_a$$

$$\text{Power supplied / developed} = E_b I_a \quad E_b I_a = E_b I_a$$

$E_b \Rightarrow$ back emf.

(2) Self Excited

(i) Shunt wound



Field winding is connected across the armature winding forming a parallel / shunt circuit.

$$I_{sh} = \frac{V}{R_{sh}}$$

$R_{sh} \Rightarrow$ shunt field winding resistance

practically constant at all loads. $\Rightarrow \phi_{sh} = \text{constant}$

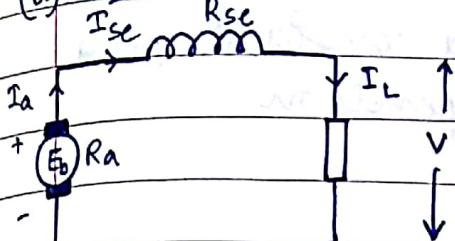
$$I_a = I_{sh} + I_L$$

$$V = E_b - I_a R_a$$

$$\text{Power developed} = E_b I_a$$

$$\text{Power output} = V I_L = (E_b - I_a R_a) \left(I_a - \frac{V}{R_{sh}} \right)$$

(ii) Series wound



$$I_{se} = I_a = I_L$$

$$V = E_b - I_a R_a - I_{se} R_{se} = E_b - I_a (R_a + R_{se})$$

$$\text{Power developed} = E_b I_a$$

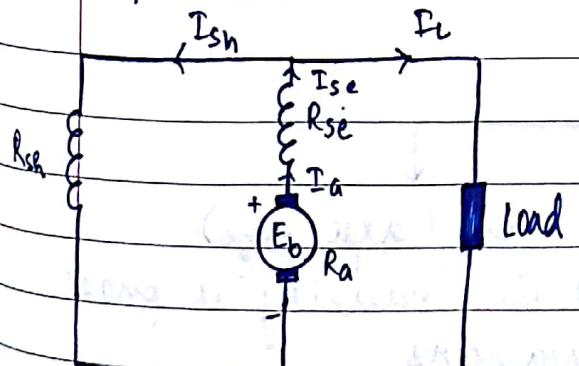
$$\text{Power output} = V I_L = V I_a = E_b I_a - I_a^2 (R_a + R_{se})$$

$$\phi_{se} \propto I_{se}$$

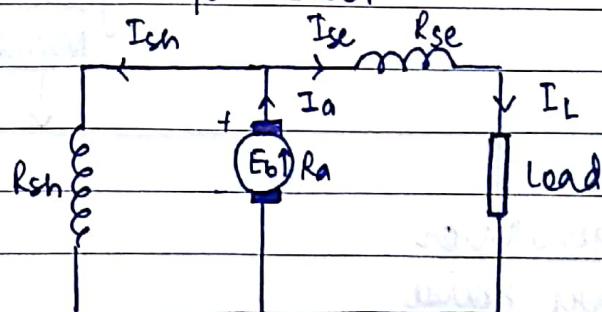
But flux becomes constant after magnetic saturation even if I_{se} is increased.

(iii) Compound wound

Long Shunt compound Generator



Short Shunt compound Generator



A compound wound generator, there are 2 sets of field winding on each pole. Has both shunt & series fields

A.C. MACHINES

1) SYNCHRONOUS MACHINE (S.M.)

Electromechanical transducer (Principle: Faraday's law of EMF)

* It runs either at synchronous speed or not at all, i.e. while running it maintains a constant speed.

Its satisfactory operation depends on the relation

$$N_s = \frac{120f}{P}$$

where,

N_s \Rightarrow synchronous speed in rpm

f \Rightarrow supply frequency

P \Rightarrow no. of poles of the machine.

* As a motor, if S.M. fails to maintain N_s the machine will not develop sufficient torque to maintain its rotation & will stop.

The motor is said to be pulled out of step.

* As a generator, it has to run at a fixed N_s to generate the power at a particular f .

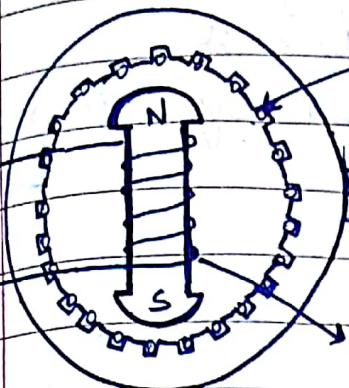
Synchronous
Machine

Generator
(Power House
500mVA)

Motor (large size)
used in industry as phase
advancers

- a) lagging current
- b) leading current.

Merits



- * used for power factor improvement
- * speed is const. & independent of load
- * mechanically better than induction motor
- *

2) ASYNCHRONOUS MOTOR / INDUCTION MOTOR

Merits

- Rotor rotates at a speed less than the synchronous speed
- Does not require any additional starting source
- less costly and reliable.
- requires min. maintenance & has high overloading property

Demerits

- Speed cannot be varied without sacrificing some of its efficiency.
- Speed \downarrow es with increase in load