

Procesamiento digital de señales

Sistemas discretos



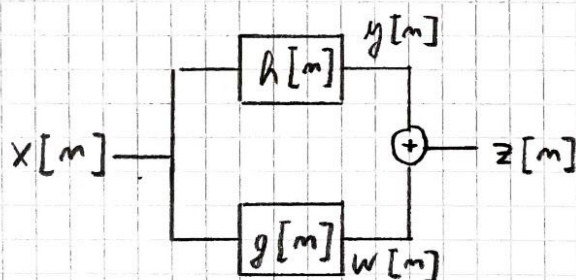
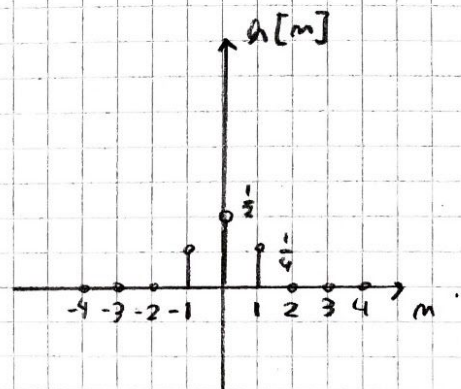
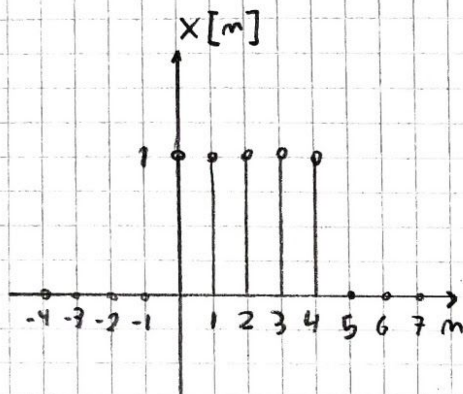
Integrantes:

- Barco Valentín
- Estrada Anselmo

1)

$$x[n] = u[n] - u[n-5]$$

$$h[n] = \frac{1}{4} \delta[n+1] + \frac{1}{2} \delta[n] + \frac{1}{4} \delta[n-1]$$



a) Como en este caso $h[n]$ solo presenta 3 terminos es mas sencilla efectuar la convolucion termino a termino

$$y[n] = x[n] * \left(\frac{1}{4} \delta[n+1] + \frac{1}{2} \delta[n] + \frac{1}{4} \delta[n-1] \right)$$

$$= \frac{1}{4} x[n+1] + \frac{1}{2} x[n] + \frac{1}{4} x[n-1]$$

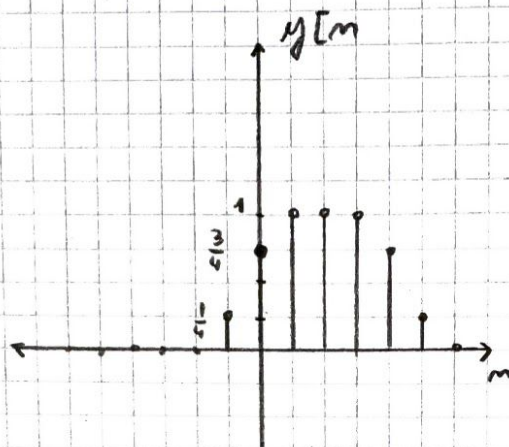
$$\bullet x[n] = \delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3] + \delta[n-4]$$

$$\bullet x[n+1] = \delta[n+1] + \delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3]$$

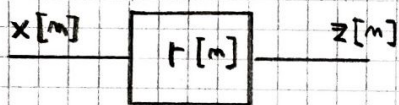
$$\bullet x[n-1] = \delta[n-1] + \delta[n-2] + \delta[n-3] + \delta[n-4] + \delta[n-5]$$

$$y[n] = \frac{1}{4} \delta[n+1] + \frac{1}{4} \delta[n] + \frac{1}{4} \delta[n-1] + \frac{1}{4} \delta[n-2] + \frac{1}{4} \delta[n-3] + \frac{1}{2} \delta[n] + \frac{1}{2} \delta[n-1] + \frac{1}{2} \delta[n-2] + \frac{1}{2} \delta[n-3] + \frac{1}{2} \delta[n-4] + \frac{1}{4} \delta[n-1] + \frac{1}{4} \delta[n-2] + \frac{1}{4} \delta[n-3] + \frac{1}{4} \delta[n-4] + \frac{1}{4} \delta[n-5]$$

$$y[m] = \frac{1}{4} \delta[m+1] + \frac{3}{4} \delta[m] + \delta[m-1] + \delta[m-2] + \delta[m-3] + \frac{3}{4} \delta[m-4] + \frac{1}{4} \delta[m-5]$$



b)

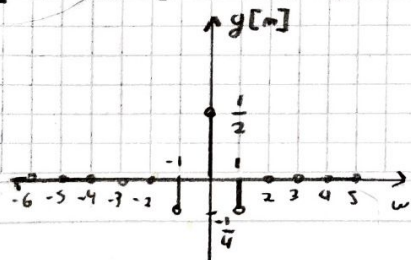


$$r[m] = h[m] + g[m]$$

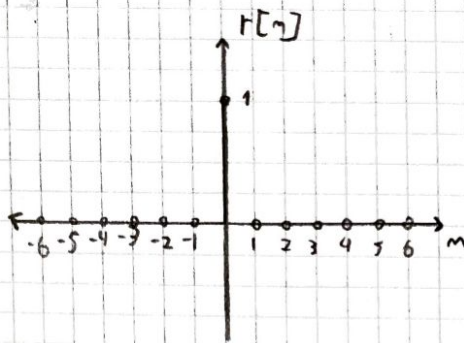
Para que $z[m] = x[m]$, $r[m]$ deve ser um impulso em caso
ya que convolucionado con $x[m]$ da o mesmo sinal
 $x[m]$.

$$r[m] = \delta[m]$$

$$g[m] = r[m] - h[m] = -\frac{1}{4} \delta[m+1] + \frac{1}{2} \delta[m] - \frac{1}{4} \delta[m-1]$$



c)



- d) Como $h[n]$, $g[n]$ y $r[n]$ son de respuesta impulsiva entonces son invariantes en el tiempo, por lo tanto para que sean causales no debe cumplir $b[n] = 0 \quad n < 0$
- $h[n]$: No causal $h[n] \neq 0 \quad n < 0$
- $g[n]$: No causal $g[n] \neq 0 \quad n < 0$
- $r[n]$: Causal $r[n] = 0 \quad n < 0$

Ejemplo 2

$$y[n] = a y[n-1] + b x[n-1], \quad y[-1] = 0$$

$$y[0] = a y[-1] + b x[-1] = b x[-1]$$

$$y[1] = a y[0] + b x[0] = a b x[-1] + b x[0]$$

$$y[2] = a y[1] + b x[1] = a^2 b x[-1] + a b x[0] + b x[1]$$

a)

$$h[0] = b \delta[-1] = 0$$

$$h[1] = a b \delta[-1] + b \delta[0] = b$$

$$h[2] = a^2 b \delta[-1] + a b \delta[0] + b x[1] = a b$$

$$h[3] = a^3 b \delta[-1] + a^2 b \delta[0] + a b x[1] + b x[0] = a^2 b$$

$$h[m] = b \cdot a^{m-1} \cdot u[m-1] = \frac{b}{a} \cdot a^m u[m] - \frac{b}{a} a^m \delta[m]$$

con $u[m] = \begin{cases} 1 & m \geq 0 \\ 0 & m < 0 \end{cases}$ y $\delta[m] = \begin{cases} 1 & m = 0 \\ 0 & m \neq 0 \end{cases}$

$$b) |h[m]| \leq \underbrace{\sum_{n=0}^{\infty} \left| \frac{b}{a} \right| |a|^n}_{\text{Como es una serie geometrica de razon } a, \text{ si } |a| < 1 \text{ la sumatoria converge a } \frac{|b/a|}{1-|a|}}$$

Como es una serie geometrica de razon a , si $|a| < 1$ la sumatoria converge a $\frac{|b/a|}{1-|a|}$

Por lo tanto se $h[m] = \sum \frac{b}{a} a^m u[m] - \frac{b}{a} a^m \delta[m]$

$$|h[m]| \leq \frac{|b/a|}{1-|a|} B_x + \frac{b}{a} |a|^m B_x < \infty$$

El sistema va a ser estable para $|a| < 1$ y $b < \infty$

c)

$$H(e^{j\omega}) = \sum_m h[m] e^{-j\omega m} = \sum_m \frac{b}{a} a^m \cdot e^{j\omega m} - \sum_m \frac{b}{a} a^m \delta[m] \cdot e^{j\omega m}$$

$$= \frac{b}{a} \sum_{m=0}^{\infty} (a e^{j\omega})^m - \frac{b}{a}$$

Esto es una serie geométrica de razón $a e^{j\omega}$, entonces la sumatoria converge si solo si $|a e^{j\omega}| < 1 \rightarrow |a| < 1$ y converge a $\frac{b}{a} \cdot \frac{1}{1 - a e^{j\omega}}$

Por lo tanto $H(e^{j\omega}) = \frac{b}{a} \left(\frac{1}{1 - a e^{j\omega}} - 1 \right)$

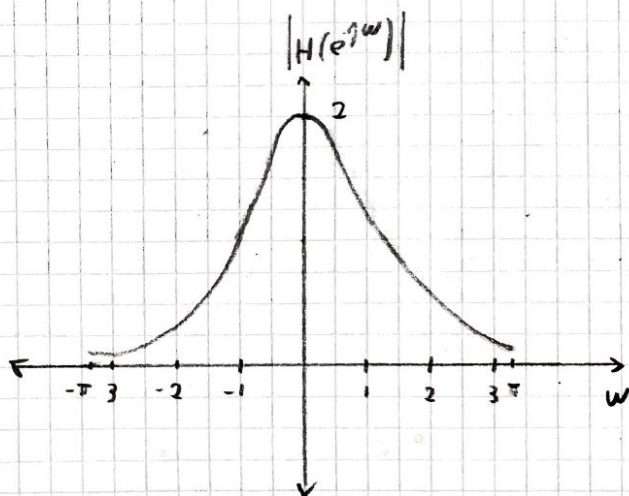
$$= \frac{b}{a} \cdot \frac{1 - 1 + a e^{j\omega}}{1 - a e^{j\omega}} = \frac{b e^{j\omega}}{1 - a e^{j\omega}}$$

$$|H(e^{j\omega})| = \left| \frac{b e^{j\omega}}{1 - a \cos(\omega) - j a \sin(\omega)} \right| = \frac{|b e^{j\omega}|}{\sqrt{(1 - a \cos(\omega))^2 + (a \sin(\omega))^2}} =$$

$$= \frac{b}{\sqrt{1 - 2a \cos(\omega) + a^2}}$$

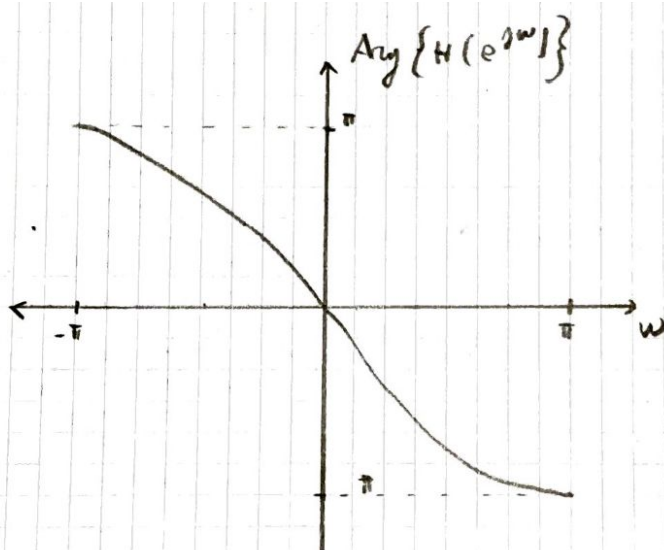
$$\text{Arg}\{H(e^{j\omega})\} = -\arctan\left(\frac{a \sin \omega}{1 - a \cos(\omega)}\right) - \omega$$

d)



$$\left| H(e^{j\omega}) \right|_{\omega=0} = \frac{b}{1-a} = 2$$

$$\left| H(e^{j\omega}) \right|_{\omega=\pi} = \frac{b}{1+a} = \frac{2}{3}$$



- e) Tomando $a=0,5$ y $b=1$ el sistema actúa como un
 paso bajo, filtrando los altos frecuencias y amplificando
 a dejando pasar los bajos, en cambio tomando valores
 negativos de " a " el sistema se comporta como un
 paso alto. El valor de b solo influye en la amplitud
 y el offset de la respuesta en frecuencia

3)

$$y[n] = x[n] + x[n-1]$$

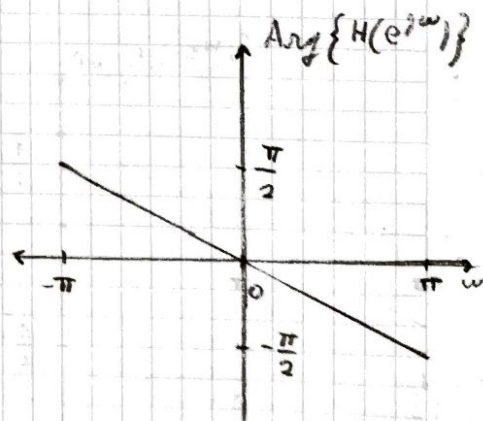
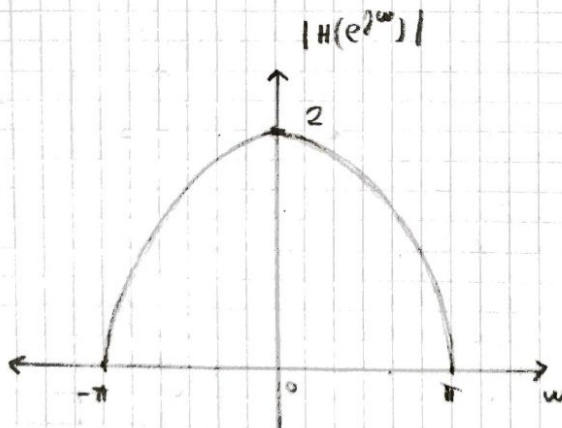
a) $h[n] = \delta[n] + \delta[n-1]$

b)
$$H(e^{j\omega}) = \sum_n h[n] \cdot e^{-j\omega n} = \sum_n (\delta[n] + \delta[n-1]) e^{-j\omega n}$$

$$= e^{-j\omega \cdot 0} + e^{-j\omega \cdot 1} = 1 + e^{-j\omega} = e^{-j\frac{\omega}{2}} (e^{j\frac{\omega}{2}} + e^{-j\frac{\omega}{2}})$$

$$= e^{-j\frac{\omega}{2}} 2 \cos\left(\frac{\omega}{2}\right)$$

c) $|H(e^{j\omega})| = \left| 2 \cos\left(\frac{\omega}{2}\right) \right|$ $\angle H(e^{j\omega}) = \frac{\omega}{2}$

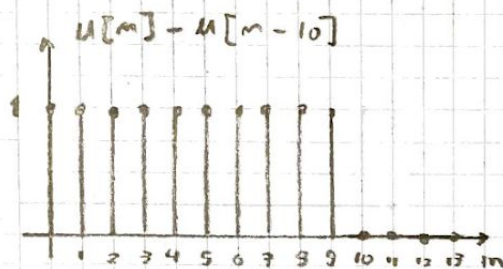


$$d-i) y[m] = \cos\left(\frac{\pi m}{4}\right) + \cos\left(\frac{\pi(m-1)}{4}\right)$$

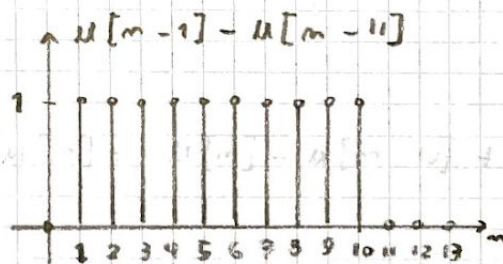
$$y[m] = \cos\left(\frac{\pi m}{4}\right) + \cos\left(\frac{\pi m}{4}\right) \cdot \overset{\sqrt{2}/2}{\cos\left(\frac{\pi}{4}\right)} + \overset{\sqrt{2}/2}{\sin\left(\frac{\pi m}{4}\right)} \cdot \overset{\sqrt{2}/2}{\sin\left(\frac{\pi}{4}\right)}$$

$$y[m] = \frac{1+\sqrt{2}}{2} \cos\left(\frac{\pi m}{4}\right) + \frac{\sqrt{2}}{2} \sin\left(\frac{\pi m}{4}\right)$$

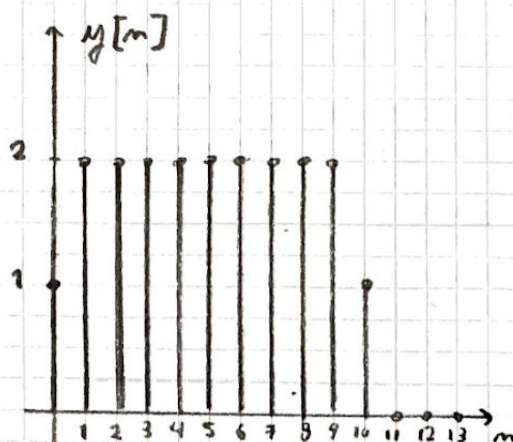
$$u) y[m] = u[m] - u[m-10] + u[m-1] - u[m-11]$$



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NOTA